



**TEXAS
INSTRUMENTS**

Linear Circuits

**Amplifiers, Comparators,
and Special Functions**

Data Book
Volume 1

Data Book
Volume 1

Linear Circuits
**Amplifiers, Comparators,
and Special Functions**

1989

1989

Linear Products

Linear Products Quick Reference Guide

Data Book	Contents	Document No.
<ul style="list-style-type: none"> ● Linear Circuits Vol 1 Amplifiers, Comparators, and Special Functions 	Operational Amplifiers Voltage Comparators Video Amplifiers Hall-Effect Devices Timers and Current Mirrors Magnetic-Memory Interface Frequency-to-Voltage Converters Sonar Ranging Circuits/Modules Sound Generators	SLYD003 1989
<ul style="list-style-type: none"> ● Linear Circuits Vol 2 Data Acquisition and Conversion 	A/D and D/A Converters DSP Analog Interface Analog Switches and Multiplexers Switched-Capacitor Filters	SLYD004 1989
<ul style="list-style-type: none"> ● Linear Circuits Vol 3 Voltage Regulators and Supervisors 	Supervisor Functions Series-Pass Voltage Regulators Shunt Regulators Voltage References DC-to-DC Converters PWM Controllers	SLYD005 1989
<ul style="list-style-type: none"> ● Telecommunications Circuits 	Equipment Line Interfaces Subscriber Line Interfaces Modems and Receiver/Transmitters Ringers, Detectors, Tone Encoders PCM Interface Transient Suppressors	SCTD001A 1988/89
<ul style="list-style-type: none"> ● Optoelectronics and Image Sensors 	Optocouplers CCD Image Sensors and Support Phototransistors IR-Emitting Diodes Hybrid Displays	SOYD002 1987
<ul style="list-style-type: none"> ● Interface Circuits 	High-Voltage (Display) Drivers High-Power (Peripheral/Motor) Drivers Line Drivers, Receivers, Transceivers EIA RS-232, RS-422, RS-423, RS-485 IBM 360/370, IEEE 802.3, CCITT Military Memory Interface	SLYD002 1987
<ul style="list-style-type: none"> ● Speech System Manuals 	TSP50C4X Family	SLPS025 1988

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***Linear Circuits
Data Book
1989***

***Volume 1
Amplifiers, Comparators, and Special Functions***



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Information contained in this data book supersedes all data for this technology published by TI in the United States of America before January 1989.

INTRODUCTION

Texas Instruments offers an extensive line of industry-standard integrated circuits designed for applications to sense, amplify, and condition analog or digital signals.

The technologies represented by the amplifiers, comparators, and many special functions in this book include traditional bipolar through BIFET, BIFET, IMPACT™, LinCMOS™, and Advanced LinCMOS™ processes. The IMPACT™ oxide-isolated technology provides the linear families with improved speed and power characteristics. LinCMOS™ and Advanced LinCMOS™ technologies feature a step-function improvement in impedance, speed, power requirements, and threshold stability.

This data book (Volume 1 of 3) provides information on the following types of products:

- Operational Amplifiers
- Video Amplifiers
- Voltage Comparators
- Timers
- Disk Drive Circuits for Control, Reading, or Writing
- Hall-Effect Circuits
- Current Mirrors
- Sonar Functions
- Sound Generators

These products cover the dynamic development of Linear circuits from the classical voltage converter to low-noise quad operational amplifiers. New surface-mount packages (8 to 28 leads) include both ceramic and plastic chip-carriers, and the small-outline (D) plastic packages that optimize board density with minimum impact on power-dissipation capability. Test equipment with handlers and automated assembly strengthen the production capabilities to provide an improved-cost-performance ratio. TI continues to enhance the quality and reliability of integrated circuits by improving materials, processes, test methods, and test equipment. In addition, specifications and programs are continuously updated. Quality and performance are monitored throughout all phases of manufacturing.

The alphanumeric listing in this data book includes all devices in Volumes 1, 2, and 3. Products in this data book are shown in **bold** type. The alphanumeric index provides a method of quickly locating the correct device type in this data book. The selection guide includes a functional description of each device that provides key parameter information and packaging types. Ordering information and mechanical data are in the last section of the data book.

While this volume offers design and specification data for linear circuit components only, complete technical data for any TI semiconductor product is available from your nearest TI Field Sales Office, local authorized TI distributor, or by writing directly to:

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P. O. Box 809066
DALLAS, TEXAS 75380-9066

We sincerely feel that the new 1989 Amplifiers and Comparators data book will be a significant addition to your library of technical literature from Texas Instruments.

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General Information

noncompensated, single

military temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_B (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
	MIN	MAX	MAX	MAX	MIN	TYP	TYP			
High Performance, Bipolar	± 5	± 22	2	75	50	1	0.5	LM101A	FK,JG,U,W	2-13
High Performance, Low Bias Current, Bipolar	± 2	± 20	2	2	50	1	0.3	LM108	L	2-21
High Performance, Low Bias Current, Bipolar	± 2	± 20	0.5	2	80	1	0.3	LM108A	L	2-21
BIFET, Low Power	± 3.5	± 18	6	0.2	4	1	3.5	TL060M	JG	2-357
BIFET, General Purpose	± 3.5	± 18	6	0.2	25	3	13	TL080M	JG	2-403
General Purpose, Precision Input, Bipolar	± 9	± 18	2	200	Typ 45	1	0.3	μ A709AM	J,JG,U,W	2-833
General Purpose, Bipolar	± 9	± 18	5	500	Typ 45	1	0.3	μ A709M	J,JG,U,W	2-833
General Purpose, Bipolar	± 2	± 22	5	500	50	1	0.5	μ A748M	JG,U	2-851

industrial temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_B (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
	MIN	MAX	MAX	MAX	MIN	TYP	TYP			
High Performance, Bipolar	± 5	± 22	2	75	50	1	0.5	LM201A	D,JG,P	2-13
BIFET, Low Power	± 3.5	± 18	6	0.2	4	1	3.5	TL060I	D,JG,P	2-357
BIFET, Low Noise	± 3.5	± 18	6	200	50	3	13	TL070I	D,JG,P	2-387
BIFET, Low Power	± 3.5	± 18	6	400	25	3	13	TL080I	D,JG,P	2-403

commercial temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_B (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
	MIN	MAX	MAX	MAX	MIN	TYP	TYP			
High Performance	± 5	± 18	7.5	250	15	1	7.5	LM301A	D,JG,P	2-13
High Performance	± 2	± 18	7.5	2	25	1	0.3	LM308	D,JG,P	2-21
High Performance	± 2	± 18	0.5	2	80	1	0.3	LM308A	D,JG,P	2-21
BIFET, Low Power	± 3.5	± 18	6	0.2	4	1	3.5	TL060AC	D,JG,P	2-357
BIFET, Low Power	± 3.5	± 18	3	0.2	4	1	3.5	TL060BC	D,JG,P	2-357
BIFET, Low Power	± 3.5	± 18	15	0.4	3	1	3.5	TL060C	D,JG,P	2-357
BIFET, Low Noise	± 3.5	± 18	6	0.2	50	3	13	TL070AC	D,JG,P	2-387
BIFET, Low Noise	± 3.5	± 18	10	0.2	25	3	13	TL070C	D,JG,P	2-387
BIFET, General Purpose	± 3.5	± 18	6	0.2	50	3	13	TL080AC	D,JG,P	2-403
BIFET, General Purpose	± 3.5	± 18	15	0.4	25	3	13	TL080C	D,JG,P	2-403
General Purpose, Bipolar	± 9	± 18	7.5	1500	15	1	0.3	μ A709C	D,JG,P	2-833
General Purpose, Bipolar	± 2	± 18	6	500	20	1	0.5	μ A748C	D,JG,P	2-851

OPERATIONAL AMPLIFIERS SELECTION GUIDE

internally compensated, single

military temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

General Information

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_{IB} (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
	MIN	MAX	MAX	MAX	MIN	TYP	TYP			
High Performance	± 5	± 20	2	75	50	1	0.5	LM107	J,JG,U,W	2-17
Precision	± 5	± 22	0.015	2	450	0.8	0.25	LT1001AM	JG,L	2-63
Precision	± 5	± 22	0.06	4	400	0.8	0.25	LT1001M	JG,L	2-63
Low Noise, High Speed, Precision Input	± 2.5	± 22	0.025	35	7000	8	2.5	LT1007AM	JG,L	2-83
Low Noise, High Speed, Precision Input	± 2.5	± 22	0.060	55	5000	8	2.5	LT1007M	JG,P	2-83
Ultra Precision	± 2.5	± 20	0.035	0.1	300	0.8	2	LT1012M	L	2-123
Low Noise, High Performance	± 4.5	± 16	0.12	150	300	25	15	LT1028AM	D,JG,L,P	2-127
Low Noise, High Performance	± 4.5	± 20	0.18	300	200	25	15	LT1028M	D,JG,L,P	2-127
Low Noise, High Speed, Noncompensated, $A_{VL} \geq 5$	± 2.5	± 22	0.025	35	7000	60	15	LT1037AM	JG,L	2-83
Low Noise, High Speed, Noncompensated, $A_{VL} \geq 5$	± 2.5	± 22	0.060	55	5000	60	15	LT1037M	JG,L	2-83
Chopper-Stabilized	± 1.9	± 8	0.005	0.03	1000	1.2	4	LTC1052M	J,JG,L	5-5
Low Noise, High Speed	± 3.5	± 22	0.025	40	1000	8	2.8	OP-27A	JG,L	2-151
Low Noise, High Speed	± 3.5	± 22	0.1	80	700	8	2.8	OP-27C	JG,L	2-151
Low Noise, High Speed Noncompensated, $A_{VL} \geq 5$	± 4	± 22	0.025	40	1000	40	17	OP-37A	JG,L	2-151
Low Noise, High Speed, Noncompensated, $A_{VL} \geq 5$	± 4	± 22	0.1	80	700	40	17	OP-37C	JG,L	2-151
Low Noise, High Performance	± 3	± 22	2	800	50	10	13	SE5534	FK,JG	2-181
Low Noise, High Performance	± 3	± 22	2	800	50	10	13	SE5534A	FK,JG	2-181
BIFET, Low Power, Precision	± 3.5	± 18	0.8	0.2	5	1.1	2.9	TL031AM	FK,JG,L	2-191
BIFET, Low-Power, Precision	± 3.5	± 18	1.5	0.2	5	1.1	2.9	TL031M	FK,JG,L	2-191
BIFET, Precision	± 3.5	± 18	0.8	0.2	50	3.1	20	TL051AM	FK,JG,L	2-273
BIFET, Precision	± 3.5	± 18	1.5	0.2	50	3.1	20	TL051M	FK,JG,L	2-273
BIFET, Low Power	± 1.5	± 18	6	0.2	4	1	3.5	TL061M	FK,JG,U	2-357
BIFET, Adjustable, Low-Power	± 1.2	± 18	6	0.2	4	1	3.5	TL066M	FK,JG	2-373
BIFET, Low Noise	± 3.5	± 18	6	0.2	35	3	13	TL071M	FK,JG	2-387
BIFET, General Purpose	± 3.5	± 18	6	0.2	25	3	13	TL081M	FK,JG	2-403
BIFET, Low V_{IO}	± 3.5	± 18	3	0.4	50	3	13	TL088M	JG,U	2-417
LinCMOS, Programmable, Low Bias	4	18	10	Typ 0.007	50	0.11	0.04	TLC271M	FK,JG	2-479
LinCMOS, Programmable, Medium Bias	4	18	10	Typ 0.007	25	0.64	0.56	TLC271M	FK,JG	2-479
LinCMOS, Programmable, High Bias	4	18	10	Typ 0.007	10	2.2	4.6	TLC271M	FK,JG	2-479
LinCMOS, Low Noise, Precision	4.6	16	0.2	Typ 0.001	400	1.9	2.7	TLC2201AM	D,FK,JG,L,P	2-763

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military temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	SUPPLY VOLTAGE (V)		V _{IO} (mV)	I _B (nA)	AVD (V/mV)	B ₁ (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
	MIN	MAX	MAX	MAX	MIN	TYP	TYP			
LinCMOS, Low Noise, Precision, 100% Noise Tested	4.6	16	0.2	Typ 0.001	400	1.9	2.7	TLC2201BM	D,FK,JG,L,P	2-763
LinCMOS, Low Noise, Precision	4.6	16	0.5	Typ 0.001	400	1.9	2.7	TLC2201M	D,FK,JG,L,P	2-763
LinCMOS, Precision, Chopper Stabilized	3.8	16	0.001	Typ 0.004	5600	1.9	2.8	TLC2652AM	D,FK,J,JG, L,N,P	2-789
LinCMOS, Precision Chopper Stabilized	3.8	16	0.003	Typ 0.004	1000	1.9	2.8	TLC2652M	D,FK,J,JG, L,N,P	2-789
LinCMOS, Low Noise, Precision, Chopper Stabilized	4.6	16	0.01	Typ 0.05	5600	1.9	2	TLC2654AM	D,FK,J,JG, L,N,P	2-811
LinCMOS, Low Noise, Precision, Chopper Stabilized	4.6	16	0.02	Typ 0.05	1000	2.2	2	TLC2654M	D,FK,J,JG, L,N,P	2-811
Excalibur, High-Speed, Precision	4	40	0.5	25	1000	2	0.9	TLE2021M	D,FK,JG,L,P	5-9
General Purpose	± 2	± 22	5	500	50	1	0.5	uA741M	FK,J,JG,U	2-837

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(Values specified for $T_A = 25^\circ\text{C}$)

General Information

DESCRIPTION	SUPPLY VOLTAGE (V)		V _{IO} (mV) MAX	I _B (nA) MAX	A _{VD} (V/mV) MIN	B ₁ (MHz) TYP	SR (V/μs) TYP	TYPE	PACKAGES	PAGE NO.
	MIN	MAX								
High Performance	±5	±22	2	75	50	1	0.5	LM207	D,J,JG,N,P,W	2-17
High Performance	±5	±20	4	250	50	15	70	LM218	D,JG,P	2-51
Chopper-Stabilized	±1.9	±8	0.005	0.03	1000	1.2	4	LTC1052C	J,JG,L,N,P	5-5
Chopper-Stabilized	±1.9	±8	0.005	0.03	1000	1.2	4	LTC7652C	L	5-5
Low Noise, High Speed	±3.5	±22	0.025	40	1000	8	2.8	OP-27E	JG,L,P	2-151
Low Noise, High Speed	±3.5	±22	0.1	80	700	8	2.8	OP-27G	JG,L,P	2-151
Low Noise, High Speed Noncompensated, A _{VL} ≥ 5	±4	±22	0.025	40	1000	40	17	OP-37E	JG,L,P	2-151
Low Noise, High Speed, Noncompensated, A _{VL} ≥ 5	±4	±22	0.1	80	700	40	17	OP-37G	JG,P	2-151
BIFET, Low Power, Precision	±3.5	±18	0.8	0.2	5	1.1	2.9	TL031AI	D,JG,L,P	2-191
BIFET, Low Power, Precision	±3.5	±18	1.5	0.2	5	1.1	2.9	TL031I	D,JG,L,P	2-191
BIFET, Precision	±3.5	±18	0.8	0.2	50	3.1	20	TL051AI	D,JG,L,P	2-273
BIFET, Precision	±3.5	±18	1.5	0.2	50	3.1	20	TL051I	D,JG,L,P	2-273
BIFET, Low Power	±3.5	±18	6	0.2	4	1	3.5	TL061I	D,JG,P	2-357
BIFET, Adjustable, Low Power	±1.2	±18	6	0.2	4	1	3.5	TL066I	D,JG,P	2-373
BIFET, Low Noise	±3.5	±18	6	0.2	50	3	13	TL071I	D,JG,P	2-387
BIFET, General Purpose	±3.5	±18	6	0.2	50	3	13	TL081I	D,JG,P	2-403
BIFET, Low Offset Voltage	±3.5	±18	0.5	0.2	50	3	13	TL087I	D,JG,P	2-417
BIFET, Low Offset Voltage	±3.5	±18	1	0.2	50	3	13	TL088I	D,JG,P	2-417
LinCMOS, Programmable, Low Bias	4	18	5	T _{YP} 0.007	50	0.11	0.04	TLC271AI	D,JG,P	2-479
LinCMOS, Programmable, Medium Bias	4	18	5	T _{YP} 0.007	25	0.64	0.56	TLC271AI	D,JG,P	2-479
LinCMOS, Programmable, High Bias	4	18	5	T _{YP} 0.007	10	2.2	4.6	TLC271AI	D,JG,P	2-479
LinCMOS, Programmable, Low Bias	4	18	2	T _{YP} 0.007	50	0.11	0.04	TLC271BI	D,JG,P	2-479
LinCMOS, Programmable, Medium Bias	4	18	2	T _{YP} 0.007	25	0.64	0.56	TLC271BI	D,JG,P	2-479
LinCMOS, Programmable, High Bias	4	18	2	T _{YP} 0.007	10	2.2	4.6	TLC271BI	D,JG,P	2-479
LinCMOS, Programmable, Low Bias	4	18	10	T _{YP} 0.007	50	0.11	0.04	TLC271I	D,JG,P	2-479
LinCMOS, Programmable, Medium Bias	4	18	10	T _{YP} 0.007	25	0.64	0.56	TLC271I	D,JG,P	2-479
LinCMOS, Programmable, High Bias	4	18	10	T _{YP} 0.007	10	2.2	4.6	TLC271I	D,JG,P	2-479
LinCMOS, Low Noise Precision	4.6	16	0.2	T _{YP} 0.001	400	1.9	2.7	TLC2201AI	D,JG,L,P	2-763

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(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_{IB} (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
	MIN	MAX	MAX	MAX	MIN	TYP	TYP			
LinCMOS, Low Noise Precision, 100% Noise Tested	4.6	16	0.2	Typ 0.001	400	1.9	2.7	TLC2201BI	D,JG,L,P	2-763
LinCMOS, Low Noise Precision	4.6	16	0.5	Typ 0.001	400	1.9	2.7	TLC2201I	D,JG,L,P	2-763
LinCMOS, Precision, Chopper Stabilized	3.8	16	0.001	Typ 0.004	5600	1.9	2.8	TLC2652AI	D,J,JG,L,N,P	2-789
LinCMOS, Precision, Chopper Stabilized	3.8	16	0.003	Typ 0.004	1000	1.9	2.8	TLC2652I	D,J,JG,L,N,P	2-789
LinCMOS, Low Noise, Precision, Chopper Stabilized	4.6	16	0.01	Typ 0.05	5600	1.9	2	TLC2654AI	D,J,JG,L,N,P	2-811
LinCMOS, Low Noise, Precision, Chopper Stabilized	4.6	16	0.02	Typ 0.05	1000	1.9	2	TLC2654I	D,J,JG,L,N,P	2-811
Excalibur, High-Speed, Precision	4	40	0.5	25	1000	2	0.9	TLE2021	D,FK,JG,L,P	5-9

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internally compensated, single
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(Values specified for $T_A = 25^\circ\text{C}$)

General Information

DESCRIPTION	SUPPLY VOLTAGE (V)		V _{IO} (mV)	I _B (nA)	AV _D (V/mV)	B ₁ (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.	
	MIN	MAX	MAX	MAX	MIN	TYP	TYP				
Chopper-Stabilized	± 1.9	± 8	0.005	0.03	1000	1.2	4	LTC1052C	J,JG,L,N,P	5-5	
Chopper-Stabilized	± 1.9	± 8	0.005	0.03	1000	1.2	4	LTC7652C	L	5-5	
Low Noise, High Speed	± 3.5	± 22	0.025	40	1000	8	2.8	OP-27E	JG,L,P	2-151	
Low Noise, High Speed	± 3.5	± 22	0.1	80	700	8	2.8	OP-27G	JG,L,P	2-151	
Low Noise, High Speed, Bipolar, Noncompensated, $A_{VL} \geq 5$	± 4	± 22	0.025	40	1000	40	17	OP-37E	JG,L,P	2-151	
Low Noise, High Speed, Bipolar, Noncompensated, $A_{VL} \geq 5$	± 4	± 22	0.1	80	700	40	17	OP-37G	JG,L,P	2-151	
BIFET, Low Power, Precision	± 3.5	± 18	0.8	0.2	5	1.1	2.9	TL031AI	D,JG,L,P	2-191	
BIFET, Low Power, Precision	± 3.5	± 18	1.5	0.2	5	1.1	2.9	TL031I	D,JG,L,P	2-191	
BIFET, Precision	± 3.5	± 18	0.8	0.2	50	3.1	20	TL051AI	D,JG,L,P	2-273	
BIFET, Precision	± 3.5	± 18	1.5	0.2	50	3.1	20	TL051I	D,JG,L,P	2-273	
BIFET, Low Power	± 3.5	± 18	6	0.2	4	1	3.5	TL061I	D,JG,P	2-357	
BIFET, Adjustable, Low-Power	± 1.2	± 18	6	0.2	4	1	3.5	TL066I	D,JG,P	2-373	
BIFET, Low Noise	± 3.5	± 18	6	0.2	50	3	13	TL071I	D,JG,P	2-387	
BIFET, General Purpose	± 3.5	± 18	6	0.2	50	3	13	TL081I	D,JG,P	2-403	
BIFET, Low Offset Voltage	± 3.5	± 18	0.5	0.2	50	3	13	TL087I	D,JG,P	2-417	
BIFET, Low Offset Voltage	± 3.5	± 18	1	0.2	50	3	13	TL088I	D,JG,P	2-417	
Single LM324, High Performance	S/S D/S	3 ± 15	30 ± 15	5	-150	50	0.6	0.3	TL321I	JG,P	2-441
LinCMOS, Programmable, Low Bias	4	18	5	Typ 0.007	50	0.11	0.04	TLC271AI	D,JG,P	2-479	
LinCMOS, Programmable, Medium Bias	4	18	5	Typ 0.007	25	0.64	0.56	TLC271AI	D,JG,P	2-479	
LinCMOS, Programmable, High Bias	4	18	5	Typ 0.007	10	2.2	4.6	TLC271AI	D,JG,P	2-479	
LinCMOS, Programmable, Low Bias	4	18	2	Typ 0.007	50	0.11	0.04	TLC271BI	D,JG,P	2-479	
LinCMOS, Programmable, Medium Bias	4	18	2	Typ 0.007	25	0.64	0.56	TLC271BI	D,JG,P	2-479	
LinCMOS, Programmable, High Bias	4	18	2	Typ 0.007	10	2.2	4.6	TLC271BI	D,JG,P	2-479	
LinCMOS, Programmable, Low Bias	4	18	10	Typ 0.007	50	0.11	0.04	TLC271I	D,JG,P	2-479	
LinCMOS, Programmable, Medium Bias	4	18	10	Typ 0.007	25	0.64	0.56	TLC271I	D,JG,P	2-479	
LinCMOS, Programmable, High Bias	4	18	10	Typ 0.007	10	2.2	4.6	TLC271I	D,JG,P	2-479	
LinCMOS, Precision, Low Noise	4.6	16	0.2	Typ 0.001	400	1.9	2.7	TLC2201AI	D,JG,L,P	2-763	
LinCMOS, Precision, Low Noise, 100% Noise Tested	4.6	16	0.2	Typ 0.001	400	1.9	2.7	TLC2201BI	D,JG,L,P	2-763	
LinCMOS, Precision, Low Noise	4.6	16	0.5	Typ 0.001	400	1.9	2.7	TLC2201I	D,JG,L,P	2-763	
LinCMOS, Precision, Chopper Stabilized	3.8	16	0.001	Typ 0.004	5600	1.9	2.8	TLC2652AI	D,J,JG,L,N,P	2-789	
LinCMOS, Precision, Chopper Stabilized	3.8	16	0.003	Typ 0.004	1000	1.9	2.8	TLC2652I	D,J,JG,L,N,P	2-789	
LinCMOS, Low-Noise, Precision, Chopper Stabilized	4.6	16	0.01	Typ 0.05	5600	1.9	2	TLC2654AI	D,J,JG,L,N,P	2-811	
LinCMOS, Low-Noise, Precision, Chopper Stabilized	4.6	16	0.02	Typ 0.05	1000	1.9	2	TLC2654I	D,J,JG,L,N,P	2-811	
Excalibur, High-Speed, Precision	4	40	0.5	25	1000	2	0.9	TLE2021I	D,FK,JG,L,P	5-9	

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(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_B (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
	MIN	MAX	MAX	MAX	MIN	TYP	TYP			
BIFET	± 3.5	± 18	10	0.2	25	3	13	LF351	D,JG,P	2-5
BIFET	± 3.5	± 18	2	0.2	25	3	13	LF411C	D,JG,P	2-9
High Performance	± 2	± 18	7.5	250	25	1	0.5	LM307	D,J,JG,N,P,W	2-17
High Performance	± 5	± 20	10	250	25	15	70	LM318	D,JG,P	2-51
Precision	± 5	± 22	0.025	2	450	0.8	0.25	LT1001AC	JG,L,P	2-63
Precision	± 5	± 22	0.06	4	400	0.8	0.25	LT1001C	JG,L,P	2-63
Low Noise, High Speed, Precision Input	± 2.5	± 22	0.025	35	7000	8	1.7	LT1007AC	JG,P	2-83
Low Noise, High Speed, Precision Input	± 2.5	± 22	0.060	55	5000	8	1.7	LT1007C	JG,P	2-83
Ultra Precision	± 2.5	± 20	0.05	0.15	200	—	0.2	LT1012C	L,P	2-123
Low Noise, High Performance	± 4.5	± 18	0.08	120	500	75	15	LT1028AC	D,JG,L,P	2-127
Low Noise, High Performance	± 4.5	± 18	0.13	240	300	75	15	LT1028C	D,JG,L,P	2-127
Low Noise, High Speed, Noncompensated, $A_{VL} \geq 5$	± 2.5	± 22	0.025	35	7000	60	15	LT1037AC	JG,P	2-83
Low Noise, High Speed, Noncompensated, $A_{VL} \geq 5$	± 2.5	± 22	0.060	55	5000	60	15	LT1037C	JG,P	2-83
Low Noise, High Performance	± 3	± 22	4	1500	25	10	13	NE5534	D,JG,P	2-181
Low Noise, High Performance	± 3	± 22	4	1500	25	10	13	NE5534A	D,JG,P	2-181
Ultra-Low Offset Voltage	± 3	± 22	0.15	7	120	0.6	0.3	OP-07C	D,JG,P	2-147
Ultra-Low Offset Voltage	± 3	± 22	0.15	12	120	0.6	0.3	OP-07D	D,JG,P	2-147
Ultra-Low Offset Voltage	± 3	± 22	0.075	4	200	0.6	0.3	OP-07E	D,JG,P	2-147

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(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_{IB} (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.	
	MIN	MAX	MAX	MAX	MIN	TYP	TYP				
BIFET, Low Power, Precision	± 3.5	± 18	0.8	0.2	5	1.1	2.9	TL031AC	D,FK,JG,L,P	2-191	
BIFET, Low Power, Precision	± 3.5	± 18	1.5	0.2	5	1.1	2.9	TL031C	D,FK,JG,L,P	2-191	
BIFET, Precision	± 3.5	± 18	0.8	0.2	50	3.1	20	TL051AC	D,FK,JG,L,P	2-273	
BIFET, Precision	± 3.5	± 18	1.5	0.2	50	3.1	20	TL051C	D,FK,JG,L,P	2-273	
BIFET, Low Power	± 3.5	± 18	6	0.2	4	1	3.5	TL061AC	D,JG,P	2-357	
BIFET, Low Power	± 3.5	± 18	3	0.2	4	1	3.5	TL061BC	D,JG,P	2-357	
BIFET, Low Power	± 3.5	± 18	15	0.2	3	1	3.5	TL061C	D,JG,P	2-357	
BIFET, Adjustable, Low-Power	± 1.2	± 18	6	0.2	4	1	3.5	TL066AC	D,JG,P	2-373	
BIFET, Adjustable, Low-Power	± 1.2	± 18	3	0.2	4	1	3.5	TL066BC	D,JG,P	2-373	
BIFET, Adjustable, Low-Power	± 1.2	± 18	15	0.4	3	1	3.5	TL066C	D,JG,P	2-373	
BIFET, Low Noise	± 3.5	± 18	6	0.2	50	3	13	TL071AC	D,JG,P	2-387	
BIFET, Low Noise	± 3.5	± 18	3	0.2	50	3	13	TL071BC	D,JG,P	2-387	
BIFET, Low Noise	± 3.5	± 18	10	0.2	25	3	13	TL071C	D,JG,P	2-387	
BIFET, General Purpose	± 3.5	± 18	6	0.2	50	3	13	TL081AC	D,JG,P	2-403	
BIFET, General Purpose	± 3.5	± 18	3	0.2	50	3	13	TL081BC	D,JG,P	2-403	
BIFET, General Purpose	± 3.5	± 18	15	0.4	25	3	13	TL081C	D,JG,P	2-403	
BIFET, Low V_{IO}	± 3.5	± 18	0.5	0.2	50	3	13	TL087C	D,JG,L,P	2-417	
BIFET, Low V_{IO}	± 3.5	± 18	1	0.2	50	3	13	TL088C	D,JG,L,P	2-417	
Single LM324, High Performance	S/S	3	30	7	-250	25	0.6	0.3	TL321C	JG,P	2-441
	D/S	1.5	15								
LinCMOS, Programmable, Low Bias	1.4	18	5	Typ 0.001	30	0.1	0.04	TLC251AC	D,JG,P	2-451	
LinCMOS, Programmable, Medium Bias	1.4	18	5	Typ 0.001	20	0.7	0.6	TLC251AC	D,JG,P	2-451	
LinCMOS, Programmable, High Bias	1.4	18	5	Typ 0.001	10	2.3	4.5	TLC251AC	D,JG,P	2-451	
LinCMOS, Programmable, Low Bias	1.4	18	2	Typ 0.001	30	0.1	0.04	TLC251BC	D,JG,P	2-451	
LinCMOS, Programmable, Medium Bias	1.4	18	2	Typ 0.001	20	0.7	0.6	TLC251BC	D,JG,P	2-451	
LinCMOS, Programmable, High Bias	1.4	18	2	Typ 0.001	10	2.3	4.5	TLC251BC	D,JG,P	2-451	
LinCMOS, Programmable, Low Bias	1.4	18	10	Typ 0.001	30	0.1	0.04	TLC251C	D,JG,P	2-451	
LinCMOS, Programmable, Medium Bias	1.4	18	10	Typ 0.001	20	0.7	0.6	TLC251C	D,JG,P	2-451	
LinCMOS, Programmable, High Bias	1.4	18	10	Typ 0.001	10	2.3	4.5	TLC251C	D,JG,P	2-451	
LinCMOS, Programmable, Low Bias	3	18	5	Typ 0.007	50	0.11	0.04	TLC271AC	D,JG,P	2-479	
LinCMOS, Programmable, Medium Bias	3	18	5	Typ 0.007	25	0.64	0.56	TLC271AC	D,JG,P	2-479	
LinCMOS, Programmable, High Bias	3	18	5	Typ 0.007	10	2.2	4.6	TLC271AC	D,JG,P	2-479	

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(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_B (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
	MIN	MAX	MAX	MAX	MIN	TYP	TYP			
LinCMOS, Programmable, Low Bias	3	18	2	Typ 0.007	50	0.11	0.04	TLC271BC	D,JG,P	2-479
LinCMOS, Programmable, Medium Bias	3	18	2	Typ 0.007	25	0.64	0.56	TLC271BC	D,JG,P	2-479
LinCMOS, Programmable, High Bias	3	18	2	Typ 0.007	10	2.2	4.6	TLC271BC	D,JG,P	2-479
LinCMOS, Programmable, Low Bias	3	18	10	Typ 0.007	50	0.11	0.04	TLC271C	D,JG,P	2-479
LinCMOS, Programmable, Medium Bias	3	18	10	Typ 0.007	25	0.64	0.56	TLC271C	D,JG,P	2-479
LinCMOS, Programmable, High Bias	3	18	10	Typ 0.007	10	2.2	4.6	TLC271C	D,JG,P	2-479
LinCMOS, Precision, Low Noise	4.6	16	0.2	Typ 0.001	400	1.9	2.7	TLC2201AC	D,JG,L,P	2-763
LinCMOS, Precision, Low Noise, 100% Noise Tested	4.6	16	0.2	Typ 0.001	400	1.9	2.7	TLC2201BC	D,JG,L,P	2-763
LinCMOS, Precision, Low Noise	4.6	16	0.5	Typ 0.001	400	1.9	2.7	TLC2201C	D,JG,L,P	2-763
LinCMOS, Precision, Chopper Stabilized	3.8	16	0.001	Typ 0.004	5600	1.9	2.8	TLC2652AC	D,J,JG,L,N,P	2-789
LinCMOS, Precision, Chopper Stabilized	3.8	16	0.003	Typ 0.004	1000	1.9	2.8	TLC2652C	D,J,JG,L,N,P	2-789
LinCMOS, Low-Noise, Precision, Chopper Stabilized	4.6	16	0.01	Typ 0.05	5600	1.9	2	TLC2654AC	D,J,JG,L,N,P	2-811
LinCMOS, Low-Noise, Precision, Chopper Stabilized	4.6	16	0.02	Typ 0.05	1000	1.9	2	TLC2654C	D,J,JG,L,N,P	2-811
Excalibur, High-Speed, Precision	4	40	0.5	25	1000	2	0.9	TLE2021C	D,FK,JG,L,P	5-9
General Purpose	± 2	± 18	6	500	20	1	0.5	uA741C	D,JG,P	2-837

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General Information

OPERATIONAL AMPLIFIERS SELECTION GUIDE

internally compensated, dual

military temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

General Information

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_B (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
	MIN	MAX	MAX	MAX	MIN	TYP	TYP			
High Gain, Low Power, Bipolar	S/S D/S	3 30	5	-150	50	0.6	0.2	LM158	FK,JG,U	2-43
General Purpose		± 2 ± 22	5	500	50	1	0.5	MC1558	FK,JG,U	2-133
Precision		± 5 ± 22	0.15	20	1500	0.7	0.4	LT1013AM	JG,L	5-3
Precision		± 5 ± 22	0.3	30	1200	0.7	0.4	LT1013M	JG,L	5-3
High Performance		± 4 ± 22	5	500	50	3.5	1.7	RM4558	JG	2-177
Low Power		± 2 ± 22	5	100	1	0.5	0.5	TL022M	U	2-187
BIFET, Low Power, Precision		± 3.5 ± 18	0.8	0.2	5	1.1	2.9	TL032AM	FK,JG,L	2-219
BIFET, Low Power, Precision		± 3.5 ± 18	1.5	0.2	5	1.1	2.9	TL032M	FK,JG,L	2-219
BIFET, Precision		± 3.5 ± 18	0.8	0.2	50	3	16	TL052AM	FK,JG,L	2-299
BIFET, Precision		± 3.5 ± 18	1.5	0.2	50	3	16	TL052M	FK,JG,L	2-299
BIFET, Low Power		± 3.5 ± 18	6	0.2	4	1	3.5	TL062M	FK,JG,U	2-357
BIFET, Low Noise		± 3.5 ± 18	6	0.2	35	3	13	TL072M	FK,JG	2-387
BIFET, General Purpose		± 3.5 ± 18	16	0.2	25	3	13	TL082M	FK,JG	2-403
BIFET, General Purpose		± 3.5 ± 18	6	0.2	25	3	13	TL083M	FK,J	2-403
BIFET, General Purpose		± 3.5 ± 18	3	0.4	50	3	13	TL287M	JG,U	2-417
BIFET, General Purpose		± 3.5 ± 18	3	0.4	50	3	13	TL288M	JG,U	2-417
LinCMOS, High Bias		4 18	10	Typ 0.005	10	2.2	5.3	TLC272M	FK,JG	2-543
LinCMOS, High Bias		4 18	0.5	Typ 0.005	10	2.2	5.3	TLC277M	FK,JG	2-543
LinCMOS, Low Bias		4 18	10	Typ 0.005	50	0.1	0.05	TLC27L2M	FK,JG	2-607
LinCMOS, Low Bias		4 18	0.5	Typ 0.005	50	0.1	0.05	TLC27L7M	FK,JG	2-607
LinCMOS, Medium Bias		4 18	10	Typ 0.005	25	0.6	0.6	TLC27M2M	FK,JG	2-671
LinCMOS, Medium Bias		4 18	0.5	Typ 0.005	25	0.6	0.6	TLC27M7M	FK,JG	2-671
LinCMOS, Micro Power, Precision		4 18	0.6	Typ 0.007	500	0.11	0.5	TLC1078M	FK,JG	2-735
Excalibur, High Speed, Precision		4 40	0.5	25	1000	2	0.9	TLE2022M	FK,JG,L	5-9

OPERATIONAL AMPLIFIERS SELECTION GUIDE

internally compensated, dual

automotive temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION		SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_B (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
		MIN	MAX	MAX	MAX	MIN	TYP	TYP			
High Gain, Low Power, Bipolar	S/S	3	30	5	-150	50	0.6	0.2	LM258	D,JG,P,U	2-43
	D/S	± 1.5	± 1.5								
High Gain, Low Power, Bipolar	S/S	3	30	3	-80	50	0.6	0.2	LM258A	D,JG,P,U	2-43
	D/S	± 1.5	± 1.5								
High Gain, Low Power, Bipolar	S/S	3	26	7	-250	Typ 100	0.6	0.2	LM2904	D,JG,P,U	2-43
	D/S	± 1.5	± 13								
High Performance		± 4	± 18	6	-500	20	3	1.7	RV4558	D,JG,P	2-177
BIFET, Low Power, Precision		± 3.5	± 18	0.8	0.2	5	1.1	2.9	TL032AI	D,JG,L,P	2-219
BIFET, Low Power, Precision		± 3.5	± 18	1.5	0.2	5	1.1	2.9	TL032I	D,JG,L,P	2-219
BIFET, Precision		± 3.5	± 18	0.8	0.2	50	3	16	TL052AI	D,JG,L,P	2-299
BIFET, Precision		± 3.5	± 18	1.5	0.2	50	3	16	TL052I	D,JG,L,P	2-299
BIFET, Low Power		± 3.5	± 18	6	0.2	4	1	3.5	TL062I	D,JG,P	2-357
BIFET, Low Noise		± 3.5	± 18	6	0.2	50	3	13	TL072I	D,JG,P	2-387
BIFET, General Purpose		± 3.5	± 18	6	0.2	50	3	13	TL082I	D,JG,P	2-403
BIFET, General Purpose		± 3.5	± 18	6	0.2	50	3	13	TL083I	D,JG,P	2-403
BIFET, General Purpose		± 3.5	± 18	0.5	0.2	50	3	13	TL287I	D,JG,P	2-417
BIFET, General Purpose		± 3.5	± 18	1	0.2	50	3	13	TL288I	D,JG,P	2-417
Low Power		± 1.5	± 18	8	-500	20	1	0.6	TL322I	D,JG,P	2-445
LinCMOS, High Bias		4	18	5	Typ 0.005	10	2.2	5.3	TLC272AI	D,JG,P	2-543
LinCMOS, High Bias		4	18	2	Typ 0.005	10	2.2	5.3	TLC272BI	D,JG,P	2-543
LinCMOS, High Bias		4	18	10	Typ 0.005	10	2.2	5.3	TLC272I	D,JG,P	2-543
LinCMOS, High Bias		4	18	0.5	Typ 0.005	10	2.2	5.3	TLC277I	D,JG,P	2-543
LinCMOS, Low Bias		4	18	5	Typ 0.005	50	0.1	0.05	TLC27L2AI	D,JG,P	2-607
LinCMOS, Low Bias		4	18	2	Typ 0.005	50	0.1	0.05	TLC27L2BI	D,JG,P	2-607
LinCMOS, Low Bias		4	18	10	Typ 0.005	50	0.1	0.05	TLC27L2I	D,JG,P	2-607
LinCMOS, Low Bias		4	18	0.5	Typ 0.005	50	0.1	0.05	TLC27L7I	D,JG,P	2-607
LinCMOS, Medium Bias		4	18	5	Typ 0.005	25	0.6	0.6	TLC27M2AI	D,JG,P	2-671
LinCMOS, Medium Bias		4	18	2	Typ 0.005	25	0.6	0.6	TLC27M2BI	D,JG,P	2-671
LinCMOS, Medium Bias		4	18	10	Typ 0.005	25	0.6	0.6	TLC27M2I	D,JG,P	2-671
LinCMOS, Medium Bias		4	18	0.5	Typ 0.005	25	0.6	0.6	TLC27M7I	D,JG,P	2-671
LinCMOS, Micro Power, Precision		4	18	0.6	Typ 0.007	500	0.11	0.05	TLC1078I	D,JG,P	2-735
Excalibur, High Speed, Precision		4	40	0.5	25	1000	2	0.9	TLE2022I	D,JG,L,P	5-9

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General Information

OPERATIONAL AMPLIFIERS SELECTION GUIDE

internally compensated, dual

industrial temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

General Information

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_B (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
	MIN	MAX	MAX	MAX	MIN	TYP	TYP			
BIFET, Low Power, Precision	± 3.5	± 18	0.8	0.2	5	1.1	2.9	TL032AI	D,JG,L,P	2-219
BIFET, Low Power, Precision	± 3.5	± 18	1.5	0.2	5	1.1	2.9	TL032I	D,JG,L,P	2-219
BIFET, Precision	± 3.5	± 18	0.8	0.2	50	3	16	TL052AI	D,JG,L,P	2-299
BIFET, Precision	± 3.5	± 18	1.5	0.2	50	3	16	TL052I	D,JG,L,P	2-299
BIFET, Low Power	± 3.5	± 18	6	0.2	4	1	3.5	TL062I	D,JG,P	2-357
BIFET, Low Noise	± 3.5	± 18	6	0.2	50	3	13	TL072I	D,JG,P	2-387
BIFET, General Purpose	± 3.5	± 18	6	0.2	50	3	13	TL082I	D,JG,P	2-403
BIFET, General Purpose	± 3.5	± 18	6	0.2	50	3	13	TL083I	D,JG,P	2-403
BIFET, General Purpose	± 3.5	± 18	0.5	0.2	50	3	13	TL287I	D,JG,P	2-417
BIFET, General Purpose	± 3.5	± 18	1	0.2	50	3	13	TL288I	D,JG,P	2-417
Low Power	± 1.5	± 18	8	-500	20	1	0.6	TL322I	D,JG,P	2-445
LinCMOS, High Bias	4	18	5	Typ 0.005	10	2.2	5.3	TLC272AI	D,JG,P	2-543
LinCMOS, High Bias	4	18	2	Typ 0.005	10	2.2	5.3	TLC272BI	D,JG,P	2-543
LinCMOS, High Bias	4	18	10	Typ 0.005	10	2.2	5.3	TLC272I	D,JG,P	2-543
LinCMOS, High Bias	4	18	0.5	Typ 0.005	10	2.2	5.3	TLC277I	D,JG,P	2-543
LinCMOS, Low Bias	4	18	5	Typ 0.005	50	0.1	0.05	TLC27L2AI	D,JG,P	2-607
LinCMOS, Low Bias	4	18	2	Typ 0.005	50	0.1	0.05	TLC27L2BI	D,JG,P	2-607
LinCMOS, Low Bias	4	18	10	Typ 0.005	50	0.1	0.05	TLC27L2I	D,JG,P	2-607
LinCMOS, Low Bias	4	18	0.5	Typ 0.005	50	0.1	0.05	TLC27L7I	D,JG,P	2-607
LinCMOS, Medium Bias	4	18	5	Typ 0.005	25	0.6	0.6	TLC27M2AI	D,JG,P	2-671
LinCMOS, Medium Bias	4	18	2	Typ 0.005	25	0.6	0.6	TLC27M2BI	D,JG,P	2-671
LinCMOS, Medium Bias	4	18	10	Typ 0.005	25	0.6	0.6	TLC27M2I	D,JG,P	2-671
LinCMOS, Medium Bias	4	18	0.5	Typ 0.005	25	0.6	0.6	TLC27M7I	D,JG,P	2-671
LinCMOS, Micro Power, Precision	4	18	0.6	Typ 0.007	500	0.11	0.05	TLC1078I	D,JG,P	2-735
Excalibur, High Speed, Precision	4	40	0.5	25	1000	2	0.9	TLE2022I	D,JG,L,P	5-9

OPERATIONAL AMPLIFIERS SELECTION GUIDE

internally compensated, dual

commercial temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_{IB} (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.	
	MIN	MAX									
BIFET, General Purpose	± 3.5	± 18	10	0.2	25	3	13	LF353	D,JG,P	2-7	
BIFET, Low Offset	± 3.5	± 18	3	0.2	25	3	13	LF412C	D,JG,P	2-11	
High Gain, Low Power, Bipolar	S/S	3	30	7	-250	25	0.6	0.2	LM358	D,JG,P,U	2-43
	D/S	± 1.5	± 15								
High Gain, Low Power, Bipolar	S/S	3	30	3	-100	25	0.6	0.2	LM358A	D,JG,P,U	2-43
	D/S	± 1.5	± 15								
Precision	± 5	± 22	0.15	20	1500	0.7	0.4	LT1013AC	JG,L,P	5-3	
Precision	± 5	± 22	0.3	30	1200	0.7	0.4	LT1013C	JG,L,P	5-3	
Precision	± 5	± 22	0.8	30	1200	0.7	0.4	LT1013D	D,JG,L,P	5-3	
General Purpose	± 1.5	± 18	6	500	20	1	0.5	MC1458	D,JG,P,U	2-133	
Low Noise	± 3	± 20	4	800	25	10	9	NE5532	JG,P	2-143	
Low Noise	± 3	± 20	4	800	25	10	9	NE5532A	JG,P	2-143	
High Performance	± 4	± 18	6	500	20	3	1.7	RC4558	D,JG,P	2-177	
High Performance	± 4	± 18	6	250	20	4	2	RC4559	D,P	2-171	
Low Power	± 2	± 18	5	250	1	0.5	0.5	TL022C	D,JG,P	2-187	
BIFET, Low Power, Precision	± 3.5	± 18	0.8	0.2	5	1.1	2.9	TL032AC	D,JG,L,P	2-219	
BIFET, Low Power, Precision	± 3.5	± 18	1.5	0.2	5	1.1	2.9	TL032C	D,JG,L,P	2-219	
BIFET, Precision	± 3.5	± 18	1.5	0.2	50	3	16	TL052AC	D,JG,L,P	2-299	
BIFET, Precision	± 3.5	± 18	4	0.2	50	3	16	TL052C	D,JG,L,P	2-299	
BIFET, Low Power	± 3.5	± 18	6	0.2	4	1	3.5	TL062AC	D,JG,P	2-357	
BIFET, Low Power	± 3.5	± 18	3	0.2	4	1	3.5	TL062BC	D,JG,P	2-357	
BIFET, Low Power	± 3.5	± 18	15	0.4	3	1	3.5	TL062C	D,JG,P	2-357	
BIFET, Low Noise	± 3.5	± 18	6	0.2	50	3	13	TL072AC	D,JG,P	2-387	
BIFET, Low Noise	± 3.5	± 18	3	0.2	50	3	13	TL072BC	D,JG,P	2-387	
BIFET, Low Noise	± 3.5	± 18	10	0.2	25	3	13	TL072C	D,JG,P	2-387	
BIFET, General Purpose	± 3.5	± 18	6	0.2	50	3	13	TL082AC	D,JG,P	2-403	
BIFET, General Purpose	± 3.5	± 18	3	0.2	50	3	13	TL082BC	D,JG,P	2-403	
BIFET, General Purpose	± 3.5	± 18	15	0.4	25	3	13	TL082C	D,JG,P	2-403	
BIFET, General Purpose	± 3.5	± 18	6	0.2	50	3	13	TL083AC	D,JG,N	2-403	

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General Information

OPERATIONAL AMPLIFIERS SELECTION GUIDE

internally compensated, dual
commercial temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_B (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
	MIN	MAX								
BIFET, General Purpose	± 3.5	± 18	15	0.4	25	3	13	TL083C	D,JG,N	2-403
BIFET, General Purpose	± 3.5	± 18	0.5	0.2	50	3	13	TL287C	D,JG,P	2-417
BIFET, General Purpose	± 3.5	± 18	1	0.2	50	3	13	TL288C	D,JG,P	2-417
Low Power	± 1.5	± 18	10	-500	20	1	0.6	TL322C	D,JG,P	2-445
LinCMOS, High Bias	1.4	18	5	Typ 0.005	10	2.2	5.3	TLC252AC	D,JG,P	2-461
LinCMOS, High Bias	1.4	18	2	Typ 0.005	10	2.2	5.3	TLC252BC	D,JG,P	2-461
LinCMOS, High Bias	1.4	18	10	Typ 0.005	10	2.2	5.3	TLC252C	D,JG,P	2-461
LinCMOS, Low Bias	1.4	18	5	Typ 0.005	30	0.1	0.05	TLC25L2AC	D,JG,P	2-461
LinCMOS, Low Bias	1.4	18	2	Typ 0.005	30	0.1	0.05	TLC25L2BC	D,JG,P	2-461
LinCMOS, Low Bias	1.4	18	10	Typ 0.005	30	0.1	0.05	TLC25L2C	D,JG,P	2-461
LinCMOS, Medium Bias	1.4	18	5	Typ 0.005	20	0.6	0.6	TLC25M2AC	D,JG,P	2-461
LinCMOS, Medium Bias	1.4	18	2	Typ 0.005	20	0.6	0.6	TLC25M2BC	D,JG,P	2-461
LinCMOS, Medium Bias	1.4	18	10	Typ 0.005	20	0.6	0.6	TLC25M2C	D,JG,P	2-461
LinCMOS, High Bias	3	18	5	Typ 0.005	10	2.2	5.3	TLC272AC	D,JG,P	2-543
LinCMOS, High Bias	3	18	2	Typ 0.005	10	2.2	5.3	TLC272BC	D,JG,P	2-543
LinCMOS, High Bias	3	18	10	Typ 0.005	10	2.2	5.3	TLC272C	D,JG,P	2-543
LinCMOS, High Bias	3	18	0.5	Typ 0.005	10	2.2	5.3	TLC277C	D,JG,P	2-543
LinCMOS, Low Bias	3	18	5	Typ 0.005	50	0.1	0.05	TLC27L2AC	D,JG,P	2-607
LinCMOS, Low Bias	3	18	2	Typ 0.005	50	0.1	0.05	TLC27L2BC	D,JG,P	2-607
LinCMOS, Low Bias	3	18	10	Typ 0.005	50	0.1	0.05	TLC27L2C	D,JG,P	2-607
LinCMOS, Low Bias	3	18	0.5	Typ 0.005	50	0.1	0.05	TLC27L7C	D,JG,P	2-607
LinCMOS, Medium Bias	3	18	5	Typ 0.005	25	0.6	0.6	TLC27M2AC	D,JG,P	2-671
LinCMOS, Medium Bias	3	18	2	Typ 0.005	25	0.6	0.6	TLC27M2BC	D,JG,P	2-671
LinCMOS, Medium Bias	3	18	10	Typ 0.005	25	0.6	0.6	TLC27M2C	D,JG,P	2-671
LinCMOS, Medium Bias	3	18	0.5	Typ 0.005	25	0.6	0.6	TLC27M7C	D,JG,P	2-671
LinCMOS, Micro Power, Precision	1.4	18	0.6	Typ 0.007	500	0.11	0.05	TLC1078C	D,JG,P	2-735
Excalibur, High Speed, Precision	4	40	0.5	25	1000	2	0.9	TLE2022C	D,JG,L,P	5-9
General Purpose	± 5	± 22	6	500	25	1	0.5	uA747C	D,J,N	2-845

OPERATIONAL AMPLIFIERS SELECTION GUIDE

internally compensated, quad

military temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_{IB} (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
	MIN	MAX	MAX	MAX	MIN	TYP	TYP			
General Purpose	3	30	5	-150	50	0.6	0.13	LM124	FK,J,W	2-33
General Purpose	± 4	± 22	5	100	50	1	0.5	LM148	FK,J	2-39
QUAD μ A741, High Performance	± 4	± 22	4	400	50	3.5	1.7	RM4136	FK,J,W	2-173
BIFET, Low Power, Precision	± 3.5	± 18	1.5	0.2	5	1.1	2.9	TL034AM	FK,J	2-245
BIFET, Low Power, Precision	± 3.5	± 18	4	0.2	5	1.1	2.9	TL034M	FK,J	2-245
Low Power	± 2	± 22	5	100	72	0.5	0.5	TL044M	FK,J,W	2-269
BIFET, Precision	± 3.5	± 18	1.5	0.2	50	2.7	16	TL054AM	FK,J	2-327
BIFET, Precision	± 3.5	± 18	4	0.2	50	2.7	16	TL054M	FK,J	2-327
BIFET, Low Power	± 3.5	± 18	9	0.2	4	1	3.5	TL064M	FK,J,W	2-357
BIFET, Low Noise	± 3.5	± 18	9	0.2	35	3	13	TL074M	FK,J,W	2-387
BIFET, General Purpose	± 3.5	± 18	9	0.2	25	3	13	TL084M	FK,J,W	2-403
LinCMOS, High Bias	4	18	10	T_{yp} 0.005	10	2.2	5.3	TLC274M	FK,J	2-575
LinCMOS, High Bias	4	18	1.2	T_{yp} 0.005	10	2.2	5.3	TLC279M	FK,J	2-575
LinCMOS, Low Bias	4	18	10	T_{yp} 0.005	50	0.1	0.05	TLC27L4M	FK,J	2-639
LinCMOS, Low Bias	4	18	0.9	T_{yp} 0.005	50	0.1	0.05	TLC27L9M	FK,J	2-639
LinCMOS, Medium Bias	4	18	10	T_{yp} 0.005	20	0.6	0.6	TLC27M4M	FK,J	2-703
LinCMOS, Medium Bias	4	18	0.9	T_{yp} 0.005	20	0.6	0.6	TLC27M9M	FK,J	2-703
LinCMOS, Micro Power, Precision	4	18	1.15	T_{yp} 0.007	500	0.11	0.05	TLC1079M	D,JG,P	2-749
Excalibur, High Speed, Precision	4	40	0.5	25	1000	2	0.9	TLE2024M	FK,J	5-9

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internally compensated, quad
automotive temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION		SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_{IB} (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
		MIN	MAX								
Norton Amplifier, Bipolar	S/S	4.5	32	-	200	1.2	2.5	0.5	LM2900	J,N	2-55
	D/S	± 2.2	± 16								
Extended Temperature Range LM324		3	26	7	-250	Typ 100	0.6	0.3	LM2902	D,J,N,W	2-33
Low Power, Bipolar	S/S	3	36	8	-500	20	1	0.6	MC3303	D,J,N	2-137
	D/S	± 1.5	± 18								
Quad μ A741		± 4.5	± 18	6	500	20	3	1.7	RV4136	D,J,N,W	2-173
BIFET, Low Power, Precision		± 3.5	± 18	1.5	0.2	5	1.1	2.9	TL034AI	D,J,N	2-245
BIFET, Low Power, Precision		± 3.5	± 18	4	0.2	5	1.1	2.9	TL034I	D,J,N	2-245
BIFET, Precision		± 3.5	± 18	1.5	0.2	50	2.7	16	TL054AI	D,J,N	2-327
BIFET, Precision		± 3.5	± 18	4	0.2	50	2.7	16	TL054I	D,J,N	2-327
BIFET, Low Power, Precision		± 3.5	± 18	6	0.2	4	1	3.5	TL064I	D,J,N	2-357
BIFET, Low Noise, Precision		± 3.5	± 18	6	0.2	50	3	13	TL074I	D,J,N	2-387
BIFET, General Purpose		± 3.5	± 18	6	0.2	50	3	13	TL084I	D,J,N	2-403
LinCMOS, High Bias		4	18	5	Typ 0.001	10	2.2	5.3	TLC274AI	D,J,N	2-575
LinCMOS, High Bias		4	18	2	Typ 0.001	10	2.2	5.3	TLC274BI	D,J,N	2-575
LinCMOS, High Bias		4	18	10	Typ 0.001	10	2.2	5.3	TLC274I	D,J,N	2-575
LinCMOS, High Bias		4	18	1.2	Typ 0.005	10	2.2	5.3	TLC279I	D,J,N	2-575
LinCMOS, Low Bias		4	18	5	Typ 0.005	50	0.1	0.05	TLC27L4AI	D,J,N	2-639
LinCMOS, Low Bias		4	18	2	Typ 0.005	50	0.1	0.05	TLC27L4BI	D,J,N	2-639
LinCMOS, Low Bias		4	18	10	Typ 0.005	50	0.1	0.05	TLC27L4I	D,J,N	2-639
LinCMOS, Low Bias		4	18	0.9	Typ 0.005	50	0.1	0.05	TLC27L9I	D,J,N	2-639
LinCMOS, Medium Bias		4	18	5	Typ 0.005	25	0.6	0.6	TLC27M4AI	D,J,N	2-703
LinCMOS, Medium Bias		4	18	2	Typ 0.005	25	0.6	0.6	TLC27M4BI	D,J,N	2-703
LinCMOS, Medium Bias		4	18	10	Typ 0.005	25	0.6	0.6	TLC27M4I	D,J,N	2-703
LinCMOS, Medium Bias		4	18	0.9	Typ 0.005	25	0.6	0.6	TLC27M9I	D,J,N	2-703
LinCMOS, Micro Power, Precision		4	18	1.15	Typ 0.007	500	0.11	0.05	TLC1079I	D,JG,P	2-749
Excalibur, High Speed, Precision		4	40	0.5	25	1000	2	0.9	TLE2024I	DW,FK,J,N	5-9

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internally compensated, quad

industrial temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_{IB} (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.	
	MIN	MAX	MAX	MAX	MIN	TYP	TYP				
General Purpose, Bipolar	3	30	5	-150	50	0.6	0.3	LM224	D,J,N,W	2-33	
General Purpose, Bipolar	3	30	3	-80	50	0.6	0.3	LM224A	D,J,N,W	2-33	
General Purpose, Bipolar	± 4	± 18	6	200	25	1	0.5	LM248	D,J,N	2-39	
High Gain, Low Power, Bipolar	S/S	3	32	5	-150	50	0.6	0.2	LM258	D,J,N	2-43
	D/S	± 1.5	± 22								
High Gain, Low Power, Bipolar	S/S	3	32	3	-80	50	0.6	0.2	LM258A	D,J,N	2-43
	D/S	± 1.5	± 22								
Single Supply, Norton Amplifier, Bipolar	S/S	4	32	-	200	1.2	2.5	0.5	LM2900	D,J,N	2-55
	D/S	± 2	± 16								
BIFET, Low Power	± 3.5	± 18	1.5	0.2	5	1.1	2.9	TL034AI	D,J,N	2-245	
BIFET, Low Power	± 3.5	± 18	4	0.2	5	1.1	2.9	TL034I	D,J,N	2-245	
BIFET, Precision	± 3.5	± 18	1.5	0.2	50	2.7	16	TL054AI	D,J,N	2-327	
BIFET, Precision	± 3.5	± 18	4	0.2	50	2.7	16	TL054I	D,J,N	2-327	
BIFET, Low Power, Precision	± 3.5	± 18	6	0.2	4	1	3.5	TL064I	D,J,N	2-357	
BIFET, Low Noise, Precision	± 3.5	± 18	6	0.2	50	3	13	TL074I	D,J,N	2-387	
BIFET, General Purpose	± 3.5	± 18	6	0.2	50	3	13	TL084I	D,J,N	2-403	
LinCMOS, High Bias	4	18	5	Typ 0.001	10	2.2	5.3	TLC274AI	D,J,N	2-575	
LinCMOS, High Bias	4	18	2	Typ 0.001	10	2.2	5.3	TLC274BI	D,J,N	2-575	
LinCMOS, High Bias	4	18	10	Typ 0.001	10	2.2	5.3	TLC274I	D,J,N	2-575	
LinCMOS, High Bias	4	18	0.9	Typ 0.005	10	2.2	5.3	TLC279I	D,J,N	2-575	
LinCMOS, Low Bias	4	18	5	Typ 0.005	50	0.1	0.05	TLC27L4AI	D,J,N	2-639	
LinCMOS, Low Bias	4	18	2	Typ 0.005	50	0.1	0.05	TLC27L4BI	D,J,N	2-639	
LinCMOS, Low Bias	4	18	10	Typ 0.005	50	0.1	0.05	TLC27L4I	D,J,N	2-639	
LinCMOS, Low Bias	4	18	0.9	Typ 0.005	50	0.1	0.05	TLC27L9I	D,J,N	2-639	
LinCMOS, Medium Bias	4	18	5	Typ 0.005	25	0.6	0.6	TLC27M4AI	D,J,N	2-703	
LinCMOS, Medium Bias	4	18	2	Typ 0.005	25	0.6	0.6	TLC27M4BI	D,J,N	2-703	
LinCMOS, Medium Bias	4	18	10	Typ 0.005	25	0.6	0.6	TLC27M4I	D,J,N	2-703	
LinCMOS, Medium Bias	4	18	0.9	Typ 0.005	25	0.6	0.6	TLC27M9I	D,J,N	2-703	
LinCMOS, Micro Power, Precision	4	18	1.15	Typ 0.007	500	0.11	0.05	TLC1079I	D,JG,P	2-749	
Excalibur, High Speed, Precision	4	40	0.5	25	1000	2	0.9	TLE2024I	DW,FK,J,N	5-9	

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commercial temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

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DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_B (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.	
	MIN	MAX									MAX
BIFET, General Purpose	± 3.5	± 18	10	0.2	25	3	13	LF347	D,J,N	2-3	
General Purpose	3	30	7	-250	25	0.6	0.3	LM324	D,J,N,W	2-33	
General Purpose	3	30	3	-100	25	0.6	0.3	LM324A	D,J,N,W	2-33	
General Purpose	± 4	± 18	6	200	25	1	0.5	LM348	D,J,N	2-39	
Single Supply, Norton Amplifier, Bipolar	S/S	4	32	-	200	1.2	2.5	0.5	LM3900	D,J,N	2-55
	D/S	± 2	± 16								
Low Power, Bipolar	S/S	3	36	10	-500	20	1	0.6	MC3403	D,J,N	2-137
	D/S	± 1.5	± 18								
Quad uA741, High Performance	± 4	± 18	6	500	20	3	1.7	RC4136	D,J,N,W	2-173	
BIFET, Low Power, Precision	± 3.5	± 18	1.5	0.2	5	1.1	2.9	TL034AC	D,J,N	2-245	
BIFET, Low Power, Precision	± 3.5	± 18	4	0.2	5	1.1	2.9	TL034C	D,J,N	2-245	
General Purpose	± 2	± 18	5	250	60	0.5	0.5	TL044C	J,N,W	2-269	
BIFET, Precision	± 3.5	± 18	1.5	0.2	50	2.7	16	TL054AC	D,J,N	2-327	
BIFET, Precision	± 3.5	± 18	4	0.2	50	2.7	16	TL054C	D,J,N	2-327	
BIFET, Low Power	± 3.5	± 18	6	0.2	4	1	3.5	TL064AC	D,J,N	2-357	
BIFET, Low Power	± 3.5	± 18	3	0.2	4	1	3.5	TL064BC	D,J,N	2-357	
BIFET, Low Power	± 3.5	± 18	15	0.4	3	1	3.5	TL064C	D,J,N	2-357	
BIFET, Low Noise	± 3.5	± 18	6	0.2	50	3	13	TL074AC	D,J,N	2-387	
BIFET, Low Noise	± 3.5	± 18	3	0.2	50	3	13	TL074BC	D,J,N	2-387	
BIFET, Low Noise	± 3.5	± 18	10	0.2	50	3	13	TL074C	D,J,N	2-387	
BIFET, Low Noise	± 3.5	± 18	10	0.2	25	3	13	TL075C	J,N	2-387	
BIFET, General Purpose	± 3.5	± 18	6	0.2	50	3	13	TL084AC	D,J,N	2-403	
BIFET, General Purpose	± 3.5	± 18	3	0.2	50	3	13	TL084BC	D,J,N	2-403	
BIFET, General Purpose	± 3.5	± 18	15	0.4	25	3	13	TL084C	D,J,N	2-403	
BIFET, General Purpose	± 3.5	± 18	15	0.4	25	3	13	TL085C	J,N	2-403	
High Performance, Bipolar	± 4	± 18	6	500	20	3	2	TL136C	N	2-437	
LinCMOS, High Bias	1.4	18	5	Typ 0.005	10	2.2	5.3	TLC254AC	D,J,N	2-469	
LinCMOS, High Bias	1.4	18	2	Typ 0.005	10	2.2	5.3	TLC254BC	D,J,N	2-469	
LinCMOS, High Bias	1.4	18	10	Typ 0.005	10	2.2	5.3	TLC254C	D,J,N	2-469	
LinCMOS, Low Bias	1.4	18	5	Typ 0.005	30	0.1	0.05	TLC25L4AC	D,J,N	2-469	
LinCMOS, Low Bias	1.4	18	2	Typ 0.005	30	0.1	0.05	TLC25L4BC	D,J,N	2-469	
LinCMOS, Low Bias	1.4	18	10	Typ 0.005	30	0.1	0.05	TLC25L4C	D,J,N	2-469	
LinCMOS, Medium Bias	1.4	18	5	Typ 0.005	20	0.6	0.6	TLC25M4AC	D,J,N	2-469	
LinCMOS, Medium Bias	1.4	18	2	Typ 0.005	20	0.6	0.6	TLC25M4BC	D,J,N	2-469	
LinCMOS, Medium Bias	1.4	18	10	Typ 0.005	20	0.6	0.6	TLC25M4C	D,J,N	2-469	

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(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	SUPPLY VOLTAGE (V)		V_{IO} (mV)	I_{IB} (nA)	A_{VD} (V/mV)	B_1 (MHz)	SR (V/ μ s)	TYPE	PACKAGES	PAGE NO.
	MIN	MAX	MAX	MAX	MIN	TYP	TYP			
LinCMOS, High Bias	3	18	5	Typ 0.005	10	2.2	5.3	TLC274AC	D,J,N	2-575
LinCMOS, High Bias	3	18	2	Typ 0.005	10	2.2	5.3	TLC274BC	D,J,N	2-575
LinCMOS, High Bias	3	18	10	Typ 0.005	10	2.2	5.3	TLC274C	D,J,N	2-575
LinCMOS, High Bias	3	18	0.9	Typ 0.005	10	2.2	5.3	TLC279C	D,J,N	2-575
LinCMOS, Low Bias	3	18	5	Typ 0.005	50	0.1	0.05	TLC27L4AC	D,J,N	2-639
LinCMOS, Low Bias	3	18	2	Typ 0.005	50	0.1	0.05	TLC27L4BC	D,J,N	2-639
LinCMOS, Low Bias	3	18	10	Typ 0.005	50	0.1	0.05	TLC27L4C	D,J,N	2-639
LinCMOS, Low Bias	3	18	0.9	Typ 0.005	50	0.1	0.05	TLC27L9C	D,J,N	2-639
LinCMOS, Medium Bias	3	18	5	Typ 0.005	25	0.6	0.6	TLC27M4AC	D,J,N	2-703
LinCMOS, Medium Bias	3	18	2	Typ 0.005	25	0.6	0.6	TLC27M4BC	D,J,N	2-703
LinCMOS, Medium Bias	3	18	10	Typ 0.005	25	0.6	0.6	TLC27M4C	D,J,N	2-703
LinCMOS, Medium Bias	3	18	0.9	Typ 0.005	25	0.7	0.6	TLC27M9C	D,J,N	2-703
LinCMOS, Micro Power, Precision	1.4	18	1.15	Typ 0.007	500	0.11	0.05	TLC1079C	D,J,N	2-749
Excalibur, High Speed, Precision	4	40	0.5	25	1000	2	0.9	TLE2024C	DW,FK,J,N	5-9

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Replacements are based on similarity of electrical and mechanical characteristics as shown in currently published data. Interchangeability in particular applications is not guaranteed. Before using a device as a substitute, the user should compare the specifications of the substitute device with the specifications of the original.

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Manufacturers are arranged in alphabetical order.

<p>ADVANCED LINEAR DEVICES ALD1701 or ALD1702 or ALD1703</p>		<p>SUGGESTED TI REPLACEMENT TLC271</p>	<p>PAGE NO. 2-479</p>
<p>ANALOG DEVICES AD510 or AD517</p>		<p>SUGGESTED TI REPLACEMENT OP-07</p>	<p>PAGE NO. 2-147</p>
<p>FAIRCHILD</p> <p>uA714 uA714E uA714L uA741 uA747 uA748 uA771 uA771A uA771B uA771L uA772 uA772A uA772B uA772L uA774 uA774B uA774L</p>	<p>DIRECT TI REPLACEMENT</p> <p>uA741 uA747 uA748</p>	<p>SUGGESTED TI REPLACEMENT</p> <p>OP-07C OP-07E OP-07D1</p> <p>TLO71 TLO71B or TLO81B TLO71A or TLO81A TLO81 TLO72 TLO72B TLO72A or TLO82A TLO82 TLO74 TLO74A or TLO74B TLO84</p>	<p>PAGE NO.</p> <p>2-147 2-147 2-147 2-837 2-845 2-851 2-387 2-387 2-387 2-403 2-387 2-387 2-387 2-403 2-387 2-387 2-403 2-387 2-387</p>

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TS27M4B		TLC27M4B	2-703



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Input Offset Voltage (V_{IO})

The d-c voltage that must be applied between the input terminals to force the quiescent d-c output voltage to zero or other level, if specified.

Average Temperature Coefficient of Input Offset Voltage (α_{VIO})

The ratio of the change in input offset voltage to the change in free-air temperature. This is an average value for the specified temperature range.

$$\alpha_{VIO} = \left[\frac{(V_{IO} @ T_{A(1)}) - (V_{IO} @ T_{A(2)})}{T_{A(1)} - T_{A(2)}} \right] \text{ where } T_{A(1)} \text{ and } T_{A(2)} \text{ are the specified temperature extremes.}$$

Input Offset Current (I_{IO})

The difference between the currents into the two input terminals with the output at zero volts.

Average Temperature Coefficient of Input Offset Current (α_{IIO})

The ratio of the change in input offset current to the change in free-air temperature. This is an average value for the specified temperature range.

$$\alpha_{IIO} = \left[\frac{(I_{IO} @ T_{A(1)}) - (I_{IO} @ T_{A(2)})}{T_{A(1)} - T_{A(2)}} \right] \text{ where } T_{A(1)} \text{ and } T_{A(2)} \text{ are the specified temperature extremes.}$$

Input Bias Current (I_{IB})

The average of the currents into the two input terminals with the output at zero volts.

Common-Mode Input Voltage (V_{IC})

The average of the two input voltages.

Common-Mode Input Voltage Range (V_{ICR})

The range of common-mode input voltage that if exceeded will cause the amplifier to cease functioning properly.

Differential Input Voltage (V_{ID})

The voltage at the noninverting input with respect to the inverting input.

Maximum Peak Output Voltage Swing (V_{OM})

The maximum positive or negative peak output voltage that can be obtained without waveform clipping when the quiescent d-c output voltage is zero.

Maximum Peak-to-Peak Output Voltage Swing (V_{OPP})

The maximum peak-to-peak output voltage that can be obtained without waveform clipping when the quiescent d-c output voltage is zero.

Large-Signal Voltage Amplification (A_V)

The ratio of the peak-to-peak output voltage swing to the change in input voltage required to drive the output.

Differential Voltage Amplification (A_{VD})

The ratio of the change in output voltage to the change in differential input voltage producing it.

OPERATIONAL AMPLIFIERS GLOSSARY

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General Information

Maximum-Output-Swing Bandwidth (B_{OM})

The range of frequencies within which the maximum output voltage swing is above a specified value.

Unity-Gain Bandwidth (B_1)

The range of frequencies within which the open-loop voltage amplification is greater than unity.

Phase Margin (ϕ_m)

The absolute value of the open-loop phase shift between the output and the inverting input at the frequency at which the modulus of the open-loop amplification is unity.

Gain Margin (A_m)

The reciprocal of the open-loop voltage amplification at the lowest frequency at which the open-loop phase shift is such that the output is in phase with the inverting input.

Input Resistance (r_i)

The resistance between the input terminals with either input grounded.

Differential Input Resistance (r_{id})

The small-signal resistance between the two ungrounded input terminals.

Output Resistance (r_o)

The resistance between the output terminal and ground.

Input Capacitance (C_i)

The capacitance between the input terminals with either input grounded.

Common-Mode Input Impedance (z_{ic})

The parallel sum of the small-signal impedance between each input terminal and ground.

Output Impedance (z_o)

The small-signal impedance between the output terminal and ground.

Common-Mode Rejection Ratio (k_{CMR} , $CMRR$)

The ratio of differential voltage amplification to common-mode voltage amplification.

NOTE: This is measured by determining the ratio of a change in input common-mode voltage to the resulting change in input offset voltage.

Supply Voltage Sensitivity (k_{SVS} , $\Delta V_{IO}/\Delta V_{CC}$)

The absolute value of the ratio of the change in input offset voltage to the change in supply voltages producing it.

NOTES: 1. Unless otherwise noted, both supply voltages are varied symmetrically.

2. This is the reciprocal of supply voltage rejection ratio.

Supply Voltage Rejection Ratio (k_{SVR} , $\Delta V_{CC}/\Delta V_{IO}$)

The absolute value of the ratio of the change in supply voltages to the change in input offset voltage.

NOTES: 1. Unless otherwise noted, both supply voltages are varied symmetrically.

2. This is the reciprocal of supply voltage sensitivity.

Equivalent Input Noise Voltage (V_n)

The voltage of an ideal voltage source (having an internal impedance equal to zero) in series with the input terminals of the device that represents the part of the internally generated noise that can properly be represented by a voltage source.

Equivalent Input Noise Current (I_n)

The current of an ideal current source (having an internal impedance equal to infinity) in parallel with the input terminals of the device that represents the part of the internally generated noise that can properly be represented by a current source.

Average Noise Figure (\bar{F})

The ratio of (1) the total output noise power within a designated output frequency band when the noise temperature of the input termination(s) is at the reference noise temperature, T_0 , at all frequencies to (2) that part of (1) caused by the noise temperature of the designated signal-input termination within a designated signal-input frequency band.

Short-Circuit Output Current (I_{OS})

The maximum output current available from the amplifier with the output shorted to ground, to either supply, or to a specified point.

Supply Current (I_{CC})

The current into the V_{CC} or V_{CC+} terminal of an integrated circuit.

Total Power Dissipation (P_D)

The total d-c power supplied to the device less any power delivered from the device to a load.

NOTE: At no load: $P_D = V_{CC+} \cdot I_{CC+} + V_{CC-} \cdot I_{CC-}$.

Crosstalk Attenuation (V_{O1}/V_{O2})

The ratio of the change in output voltage of a driven channel to the resulting change in output voltage of another channel.

Rise Time (t_r)

The time required for an output voltage step to change from 10% to 90% of its final value.

Total Response Time (Settling Time) (t_{tot})

The time between a step-function change of the input signal level and the instant at which the magnitude of the output signal reaches for the last time a specified level range ($\pm\epsilon$) containing the final output signal level.

Overshoot Factor

The ratio of (1) the largest deviation of the output signal value from its final steady-state value after a step-function change of the input signal, to (2) the absolute value of the difference between the steady-state output signal values before and after the step-function change of the input signal.

Slew Rate (SR)

The average time rate of change of the closed-loop amplifier output voltage for a step-signal input.





General Information

COMPARATORS SELECTION GUIDE

military temperature range

(Values specified at $T_A = 25^\circ\text{C}$)

DESCRIPTION	POWER SUPPLY		V_{IO} MAX (mV)	I_{IB} MAX (μA)	I_{OL} MIN (mA)	RESPONSE TIME TYP (ns)	TYPE	PACKAGE	PAGE NO.
	V_{CC+} NOM (V)	V_{CC-} NOM (V)							

single channel

Strobe	4-30	0	3	0.1	8	115	LM111	FK,J,JG,U	3-3
Ultra Low Power, Strobe	4-30	0	7.5	0.1	1.6	1200	LP111	FK,JG	3-29
Strobe	4-30	0	1.5	0.05	50	150	LT1011M	JG,L	3-37
Strobe	4-30	0	0.5	0.025	50	150	LT1011AM	JG,L	3-37
Ultra-Fast, Precision	5	-5	± 2	10	10	10	LT1016M	JG,L	3-61

dual channel

Low Power, Bipolar	4-30	0	5	0.1	6	300	LM193	FK,JG,L	3-21
Dual TL510M	12	-6	2	15	2	30	TL514M	FK,J,W	3-69
Ultra Low Supply, LinCMOS	1.4-18	0	10	†	6	200	TLC352M	FK,JG	3-103
High Speed, LinCMOS	4-18	0	10	†	6	200	TLC372M	FK,JG	3-119
Ultra Low Power, Open-Drain Output	4-18	0	5	†	6	1100	TLC393M	FK,JG	3-135
Ultra Low Power, Push-Pull Output	4-18	0	5	†	4	1300	TLC3702M	FK,JG	3-151

quad channel

Low Power, Bipolar	4-30	0	5	-0.1	6	300	LM139	FK,J	3-17
Precision Input	4-30	0	2	-0.1	6	300	LM139A	FK,J	3-17
Ultra Low Power, Open-Drain Output	4-18	0	5	†	6	1100	TLC339M	FK,J	3-89
Ultra Low Supply, LinCMOS	1.4-18	0	10	†	6	200	TLC354M	FK,J	3-111
High Speed, LinCMOS	4-18	0	10	†	6	200	TLC374M	FK,J	3-127
Ultra Low Power, Push-Pull Output	4-18	0	5	†	4	1300	TLC3704M	FK,J	3-157

† Typically 5 pA

COMPARATORS SELECTION GUIDE

automotive temperature range

(Values specified at $T_A = 25^\circ\text{C}$)

DESCRIPTION	POWER SUPPLY		V_{IO} MAX (mV)	I_{IB} MAX (μA)	I_{OL} MIN (mA)	RESPONSE TIME TYP (ns)	TYPE	PACKAGE	PAGE NO.
	V_{CC+} NOM (V)	V_{CC-} NOM (V)							

dual channel

Automotive LM393	4-30	0	7	0.25	6	300	LM2903	D,JG,P	3-21
Ultra-Low Supply, LinCMOS	1.4-18	0	10	†	6	200	TLC352I	D,JG,P	3-103
High Speed, LinCMOS	3-18	0	10	†	6	200	TLC372I	D,JG,P	3-119
Ultra Low Power, Open-Drain Output	3-18	0	5	†	6	1100	TLC393I	D,JG,P	3-135
Ultra Low Power, Push-Pull Output	3-18	0	5	†	4	1300	TLC3702I	D,JG,P	3-151

quad channel

Automotive Temp. LM339	4-30	0	7	-0.25	6	300	LM2901	D,J,N	3-17
Low-Cost LM2901	4-26	0	20	0.5	6	300	LM3302	D,J,N	3-25
Ultra Low Power, Automotive LP339, Bipolar	5	0	± 5	-0.025	20	8000	LP2901	D,J,N	3-33
Ultra Low Supply, LinCMOS	1.4-18	0	10	†	6	200	TLC354I	D,J,N	3-111
Open-Drain Output	3-18	0	5	†	6	1100	TLC339I	D,J,N	3-89
High Speed, LinCMOS	3-18	0	10	†	6	200	TLC374I	D,J,N	3-127
Push-Pull Output	3-18	0	5	†	4	1300	TLC3704I	D,J,N	3-157

† Typically 5 pA

COMPARATORS SELECTION GUIDE

industrial temperature range

(Values specified at $T_A = 25^\circ\text{C}$)

DESCRIPTION	POWER SUPPLY		V_{IO} MAX (mV)	I_{IB} MAX (μA)	I_{OL} MIN (mA)	RESPONSE TIME TYP (ns)	TYPE	PACKAGE	PAGE NO.
	V_{CC+} NOM (V)	V_{CC-} NOM (V)							

single channel

Strobe	4–30	0	3	0.1	8	115	LM211	D,JG,P	3-3
Ultra Low Power, Strobe	4–30	0	7.5	0.1	1.6	1200	LP211	D,JG,P	3-29
Single LM339	4–30	0	5	–0.1	6	300	TL3311	D,JG,P	3-65

dual channel

Industrial LM393	4–30	0	5	0.25	6	300	LM293	D,JG,P	3-21
Industrial LM393, Low Offset	4–30	0	2	0.25	6	300	LM293A	D,JG,P	3-21
Ultra Low Supply, LinCMOS	1.4–18	0	10	†	6	200	TLC352I	D,JG,P	3-103
High Speed, LinCMOS	3–18	0	10	†	6	200	TLC372I	D,JG,P	3-119
Ultra Low Power, Open-Drain Output	3–18	0	5	†	6	1100	TLC393I	D,JG,P	3-135
Ultra Low Power, Push-Pull Output	3–18	0	5	†	4	1300	TLC3702I	D,JG,P	3-151

quad channel

Industrial LM339	4–30	0	5	–0.25	6	300	LM239	D,J,N	3-17
Industrial LM339, Low Offset	4–30	0	2	–0.25	6	300	LM239A	D,J,N	3-17
Ultra Low Power, Industrial LP339, Bipolar	4–30	0	± 5	–0.025	20	8000	LP239	D,J,N	3-33
Ultra Low Power, Open-Drain Output	3–18	0	5	†	6	1100	TLC339I	D,J,N	3-89
Ultra Low Supply, LinCMOS	1.4–18	0	10	†	6	200	TLC354I	D,J,N	3-111
High Speed, LinCMOS	3–18	0	10	†	6	200	TLC374I	D,J,N	3-127
Ultra Low Power, Push-Pull Output	3–18	0	5	†	4	1300	TLC3704I	D,J,N	3-157

† Typically 5 pA

COMPARATORS SELECTION GUIDE

commercial temperature range

(Values specified at $T_A = 25^\circ\text{C}$)

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General Information

DESCRIPTION	POWER SUPPLY		V_{IO} MAX (mV)	I_{IB} MAX (μA)	I_{OL} MIN (mA)	RESPONSE TIME TYP (ns)	TYP	PACKAGE	PAGE NO.
	V_{CC+} NOM (V)	V_{CC-} NOM (V)							

single channel

Strobe	4-30	0	7.5	0.25	8	115	LM311	D,JG,P	3-3
Ultra Low Power, Strobe	4-30	0	7.5	0.1	1.6	1200	LP311	D,JG,P	3-29
Strobe	4-30	0	0.5	0.025	50	150	LT1011AC	JG,L,P	3-37
Strobe	4-30	0	1.5	0.05	50	150	LT1011C	JG,L,P	3-37
Ultra-Fast Precision	5	-5	± 3	10	10	10	LT1016C	D,JG,L,P	3-61
Single LM339	4-30	0	5	-0.25	6	300	TL331C	D,JG,P	3-65
Output Enable	5	0	± 1	-	MAX 16	25	TL712	D,JG,P	3-77
High Speed	5	0	-	-	MAX 16	7	TL714C	D,P	3-81
High Speed	0	-5.2	± 1	-	MAX 16	Max 12	TL721	D,JG,P	3-85

dual channel

Low Power, Bipolar	4-30	0	5	0.25	6	300	LM393	D,JG,P	3-21
Precision Input	4-30	0	2	0.25	6	300	LM393A	D,JG,P	3-21
Ultra Low Supply, LinCMOS	1.4-18	0	10	†	6	200	TLC352C	D,JG,P	3-103
High Speed, LinCMOS	3-18	0	10	†	6	200	TLC372C	D,JG,P	3-119
Ultra Low Power, Open Drain Output, CMOS	3-18	0	5	†	6	1100	TLC393C	D,JG,P	3-135
Ultra Low Power, Push-Pull Output, CMOS	3-18	0	5	†	4	1300	TLC3702C	D,JG,P	3-151

† Typically 5 pA

COMPARATORS SELECTION GUIDE

commercial temperature range

(Values specified at $T_A = 25^\circ\text{C}$)

DESCRIPTION	POWER SUPPLY		V_{IO} MAX (mV)	I_{IB} MAX (μA)	I_{OL} MIN (mA)	RESPONSE TIME TYP (ns)	TYP	PACKAGE	PAGE NO.
	V_{CC+} NOM (V)	V_{CC-} NOM (V)							

quad channel

Low Power, Bipolar	4-30	0	5	-0.25	6	300	LM339	D,J,N	3-17
Precision Input	4-30	0	2	-0.25	6	300	LM339A	D,J,N	3-17
Ultra Low Power, Bipolar	4-30	0	± 5	-0.025	6	8000	LP339	D,J,N	3-33
Ultra Low Power, Open-Drain Output, CMOS	3-18	0	5	†	6	1100	TLC339C	D,J,N	3-89
Ultra Low Supply, LinCMOS	1.4-18	0	10	†	6	200	TLC354C	D,J,N	3-111
High Speed, CMOS	3-18	0	10	†	6	200	TLC374C	D,J,N	3-127
Ultra Low Power, Push-Pull Output, CMOS	3-18	0	5	†	4	1300	TLC3704C	D,J,N	3-157

† Typically 5 pA

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General Information

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General Information

COMPARATORS CROSS-REFERENCE GUIDE

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General Information

Replacements are based on similarity of electrical and mechanical characteristics as shown in currently published data. Interchangeability in particular applications is not guaranteed. Before using a device as a substitute, the user should compare the specifications of the substitute device with the specifications of the original.

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Manufacturers are arranged in alphabetical order.

LINEAR TECHNOLOGY	DIRECT TI REPLACEMENT	SUGGESTED TI REPLACEMENT	PAGE NO.
LT1011A	LT1011A		3-37
LT1011	LT1011		3-37
LT1016	LT1016		3-61
LT1017		TLC352, TLC392, or TLC3702	3-103
LT1018		TLC352, TLC392, or TLC3702	3-151 3-103 3-151
NATIONAL	DIRECT TI REPLACEMENT	SUGGESTED TI REPLACEMENT	PAGE NO.
LM311	LM311		3-3
LM339	LM339	TLC339	3-17
LM393	LM393	TLC393	3-21
LM2901	LM2901	TLC339	3-17
LM3302	LM3302		3-25
LP339	LP339	TLC339	3-33
PMI		SUGGESTED TI REPLACEMENT	PAGE NO.
CMP04F		LM339 or LM3302 or LM2901 or TLC339	3-17 3-25 3-17 3-89



General Information

Input Offset Voltage (V_{IO})

The d-c voltage that must be applied between the input terminals to force the quiescent d-c output voltage to the specified level.

Average Temperature Coefficient of Input Offset Voltage (α_{VIO})

The ratio of the change in input offset voltage to the change in free-air temperature. This is an average value for the specified temperature range.

$$\alpha_{VIO} = \left[\frac{(V_{IO} @ T_{A(1)}) - (V_{IO} @ T_{A(2)})}{T_{A(1)} - T_{A(2)}} \right] \text{ where } T_{A(1)} \text{ and } T_{A(2)} \text{ are the specified temperature extremes.}$$

Input Offset Current (I_{IO})

The difference between the currents into the two input terminals with the output at the specified level.

Average Temperature Coefficient of Input Offset Current (α_{IIO})

The ratio of the change in input offset current to the change in free-air temperature. This is an average value for the specified temperature range.

$$\alpha_{IIO} = \left[\frac{(I_{IO} @ T_{A(1)}) - (I_{IO} @ T_{A(2)})}{T_{A(1)} - T_{A(2)}} \right] \text{ where } T_{A(1)} \text{ and } T_{A(2)} \text{ are the specified temperature extremes.}$$

Input Bias Current (I_{IB})

The average of the currents into the two input terminals with the output at the specified level.

High-Level Strobe Current (I_{IH(S)})

The current flowing into or out of* the strobe at a high-level voltage.

Low-Level Strobe Current (I_{IL(S)})

The current flowing out of* the strobe at a low-level voltage.

High-Level Strobe Voltage (V_{IH(S)})

For a device having an active-low strobe, a voltage within the range that is guaranteed not to interfere with the operation of the comparator.

Low-Level Strobe Voltage (V_{IL(S)})

For a device having an active-low strobe, a voltage within the range that is guaranteed to force the output high or low, as specified, independently of the differential inputs.

Input Voltage Range (V_I)

The range of voltage that if exceeded at either input terminal will cause the comparator to cease functioning properly.

Common-Mode Input Voltage (V_{IC})

The average of the two input voltages.

*Current out of a terminal is given as a negative value.

COMPARATORS GLOSSARY

Common-Mode Input Voltage Range (V_{ICR})

The range of common-mode input voltage that if exceeded will cause the comparator to cease functioning properly.

Differential Input Voltage (V_{ID})

The voltage at the noninverting input with respect to the inverting input.

Differential Input Voltage Range (V_{ID})

The range of voltage between the two input terminals that if exceeded will cause the comparator to cease functioning properly.

Differential Voltage Amplification (A_{VD})

The ratio of the change in output to the change in differential input voltage producing it with the common-mode input voltage held constant.

High-Level Output Voltage (V_{OH})

The voltage at an output with input conditions applied that according to the product specification will establish a high level at the output.

Low-Level Output Voltage (V_{OL})

The voltage at an output with input conditions applied that according to the product specification will establish a low level at the output.

High-Level Output Current, (I_{OH})

The current into* an output with input conditions applied that according to the product specification will establish a high level at the output.

Low-Level Output Current, (I_{OL})

The current into* an output with input conditions applied that according to the product specification will establish a low level at the output.

Output Resistance (r_o)

The resistance between an output terminal and ground.

Common-Mode Rejection Ratio (k_{CMR} , $CMRR$)

The ratio of differential voltage amplification to common-mode voltage amplification.

NOTE: This is measured by determining the ratio of a change in input common-mode voltage to the resulting change in input offset voltage.

Supply Current (I_{CC+} , I_{CC-})

The current into* the V_{CC+} or V_{CC-} terminal of an integrated circuit.

Total Power Dissipation (P_D)

The total d-c power supplied to the device less any power delivered from the device to a load.

NOTE: At no load: $P_D = V_{CC+} \cdot I_{CC+} + V_{CC-} \cdot I_{CC-}$.

*Current out of a terminal is given as a negative value.

Response Time

The interval between the application of an input step function and the instant the output crosses the logic threshold voltage.

NOTE: The input step drives the comparator from some initial condition sufficient to saturate the output (or in the case of high-to-low-level response time, to turn the output off) to an input level just barely in excess of that required to bring the output back to the logic threshold voltage. This excess is referred to as the voltage overdrive.

Strobe Release Time

The time required for the output to rise to the logic threshold voltage after the strobe terminal has been driven from its active logic level to its inactive logic level.

*Current out of a terminal is given as a negative value.



General Information

precision timers

military temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	OUTPUT CURRENT	TIMING		TYPE	PACKAGES	PAGE NO.
		FROM	TO			
Single Timer, Bipolar	± 200 mA	1 μs	Hours	SE555	FK,JG	4-37
Single Timer, Bipolar	± 200 mA	1 μs	Hours	SE555C	FK,JG	4-37
Dual Timer, Bipolar	± 200 mA	1 μs	Hours	SE556	FK,J	4-49
Dual Timer, Bipolar	± 200 mA	1 μs	Hours	SE556C	FK,J	4-49
LinCMOS, Single High-Speed Timer	100 mA – 10 mA	1 μs	Hours	TLC555M	FK,JG	4-195
LinCMOS, Dual High-Speed Timer	100 mA – 10 mA	1 μs	Hours	TLC556M	FK,J	4-203

automotive temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	OUTPUT CURRENT	TIMING		TYPE	PACKAGES	PAGE NO.
		FROM	TO			
Single Timer, Bipolar	± 200 mA	10 μs	Hours	SA555	D,JG,P	4-37
Dual Timer, Bipolar	± 200 mA	10 μs	Hours	SA556	D,J,N	4-49
LinCMOS, Single High-Speed Timer	100 mA – 10 mA	1 μs	Hours	TLC555I	D,JG,P	4-195
LinCMOS, Dual High-Speed Timer	100 mA – 10 mA	1 μs	Hours	TLC556I	D,J,N	4-203

commercial temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	OUTPUT CURRENT	TIMING		TYPE	PACKAGES	PAGE NO.
		FROM	TO			
Single Timer, Bipolar	± 200 mA	10 μs	Hours	NE555	D,JG,P	4-37
Dual Timer, Bipolar	± 200 mA	10 μs	Hours	NE556	D,J,N	4-49
LinCMOS, Single High-Speed Timer, 1-Volt Operation	100 mA – 10 mA	1 μs	Hours	TLC551C	D,P	4-179
LinCMOS, Dual High-Speed Timer, 1-Volt Operation	100 mA – 10 mA	1 μs	Hours	TLC552C	D,N	4-187
LinCMOS, Single High-Speed Timer	100 mA – 10 mA	1 μs	Hours	TLC555C	D,JG,P	4-195
LinCMOS, Dual High-Speed Timer	100 mA – 10 mA	1 μs	Hours	TLC556C	D,J,N	4-203
Programmable Timer/Counter	4 mA	10 μs	Days	uA2240C	N	4-221

SPECIAL FUNCTIONS SELECTION GUIDE

1 General Information

current mirrors

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	TEMPERATURE RANGE	CURRENT RATIO INPUT TO OUTPUT	INPUT CURRENT RANGE	TYPE	PACKAGES	PAGE NO.
Programmable	0°C to 70°C	3:1 to 1:15	Variable	TL010C	P	4-81
Programmable	-40°C to 85°C	3:1 to 1:15	Variable	TL010I	P	4-81
Fixed	0°C to 70°C	1:1	1 μA to 1 mA	TL011C	LP	4-85
Fixed	-40°C to 85°C	1:1	1 μA to 1 mA	TL011I	LP	4-85
Fixed	0°C to 70°C	1:2	1 μA to 1 mA	TL012C	LP	4-85
Fixed	-40°C to 85°C	1:2	1 μA to 1 mA	TL012I	LP	4-85
Fixed	0°C to 70°C	1:4	1 μA to 1 mA	TL014AC	LP	4-85
Fixed	0°C to 70°C	1:2	2 μA to 2 mA	TL021C	LP	4-85
Fixed	-40°C to 85°C	1:2	2 μA to 2 mA	TL021I	LP	4-85

Hall-Effect switches

(Values specified for $T_A = 25^\circ\text{C}$)

RELEASE POINT (GAUSS) MIN	OPERATING POINT (GAUSS) MAX	MINIMUM HYSTERESIS (GAUSS)	TYPE	PACKAGES	PAGE NO.
-250	250	50	TL170C	LP	4-121
100	600	230	TL172C	LP	4-123
25	450	30	TL3013C	LU	4-165
125	500	50	TL3019C	LU	4-167
50	350	20	TL3020C	LU	4-169
-250	250	50	TL3101	LU	4-171

Hall-Effect linear circuits

(Values specified for $T_A = 25^\circ\text{C}$)

LINEAR RANGE (GAUSS)	SENSITIVITY (mV/GAUSS)	TYPE	PACKAGES	PAGE NO.
± 500	1.4	TL173	LP	4-125
± 500	1.4	TL3103	LU	4-173

sonar ranging functions

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	TYPE	PACKAGES	PAGE NO.
Sonar Ranging Module	Sonar ranging module for measuring distances from a range of 6 inches to 35 feet using the TL851 and TL852	SN28827	4-61
Sonar Ranging Module	Sonar ranging module for measuring distances from a range of 6 inches to 35 feet using the TL852 and TL853	SN28828	4-67
Controller Circuit	Control integrated circuit for use in a sonar ranging module, capable of driving 50-kHz transducers with a simple interface	TL851	N
Receiver Circuit	Receiver integrated circuit for use in a sonar ranging module	TL852	N
Control Circuit	Control integrated circuit for use in a sonar ranging module, capable of driving 40-kHz transducers with a simple interface	TL853	N

SPECIAL FUNCTIONS SELECTION GUIDE

floppy-disk control circuits

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	TYPE	PACKAGES	PAGE NO.
Read-Amplifier System	MC3470 MC3470A	N	4-29
Tape-Read Signal Conditioner	TL041C	DW,NT	4-113
Disk-Memory Read-Chain Data Comparator	TL712	D,JG,P	3-77
Disk-Memory Read-Chain Data Comparator with MECL III and MECL 1000	TL721	D,JG,P	3-85

differential video amplifiers

military temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	BANDWIDTH (MHz)	GAIN	TYPICAL NOISE, V_n	TYPE	PACKAGES	PAGE NO.
Amplifier with internal frequency compensation and adjustable/selectable gain options	90	600 Max	12 μV	SE592	D,N	4-53
Amplifier with internal frequency compensation	200	10, 100, 400	12 μV	uA733M	J,U	4-213

commercial temperature range

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	BANDWIDTH (MHz)	GAIN	TYPICAL NOISE, V_n	TYPE	PACKAGES	PAGE NO.
Amplifier with 2 multiplexed inputs and wide AGC range	60	100 max	25 μV	MC1445	J,N	4-27
Amplifier with internal frequency compensation and adjustable/selectable gain options	90	600 max	12 μV	NE592	D,N	4-53
Similar to NE592 but with tighter gain distribution	90	600 max	12 μV	NE592A	D,N	4-53
Amplifier with a wide AGC range	50	100	12 μV	TLO26C	D,P	4-91
Amplifier with a wide AGC range	50	400 max	12 μV	TL027C	D,J,N	4-99
2-channel multiplexed Video Amp	20	600 max	< 5 μV	TLO40C	D,N	4-107
Similar to NE592 but in an 8-pin package	90	600 max	12 μV	TL592	D,P	4-143
Similar to NE592A but in an 8-pin package	90	600 max	12 μV	TL592A	D,P	4-143
Low-noise version of NE592 and TL592	90	600 max	3 μV	TL592B	D,N,P	4-147
Amplifier with internal frequency compensation	200	10, 100, 400	12 μV	uA733C	D,N	4-213

logarithmic amplifiers

military temperature range

(Values specified for operating temperature range)

DESCRIPTION	BANDWIDTH	GAIN	TYPE	PACKAGES	PAGE NO.
Logarithmic Amplifier	40 MHz	Logarithmic Curve	TL441AM	J,FK	4-129

1

General Information

SPECIAL FUNCTIONS SELECTION GUIDE

1

General Information

programmable tone/noise generators

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	TYPE	PACKAGES	PAGE NO.
<ul style="list-style-type: none"> Complex sound generators designed to provide low-cost digital tones or noise. Programmable white-noise and attenuation functions, and simultaneous sounds under microprocessor control. TTL compatible. 	SN76494/ SN76494A SN76496/ SN76496A	N	4-73

frequency-to-voltage-converters

(Values specified for $T_A = 25^\circ\text{C}$)

DESCRIPTION	TYPE	PACKAGES	PAGE NO.
<ul style="list-style-type: none"> Output swings to ground for zero-frequency input Only one RC network provides frequency doubling for low ripple. 8-pin version interfaces directly to variable reluctance magnetic pickups. 	LM2907 LM2917	D,N,P	4-21

sample-and-hold amplifiers

military temperature range

(Values specified for operating temperature range)

DESCRIPTION	OFFSET VOLTAGE	GAIN ERROR	TYPE	PACKAGES	PAGE NO.
Precision Sample-and-Hold Amplifier	1 mV	0.002%	LF198	L	4-3
	0.5 mV	0.001%	LF198A		

commercial temperature range

(Values specified for operating temperature range)

DESCRIPTION	OFFSET VOLTAGE	GAIN ERROR	TYPE	PACKAGES	PAGE NO.
Precision Sample-and-Hold Amplifier	2 mV	0.004%	LF398	JG,L	4-3
	1 mV	0.001%	LF398A	L,P	4-3

SPECIAL FUNCTIONS CROSS-REFERENCE GUIDE

Replacements are based on similarity of electrical and mechanical characteristics as shown in currently published data. Interchangeability in particular applications is not guaranteed. Before using a device as a substitute, the user should compare the specifications of the substitute device with the specifications of the original.

Texas Instruments makes no warranty as to the information furnished and buyer assumes all risk in the use thereof. No liability is assumed for damages resulting from the use of the information contained herein.

Manufacturers are arranged in alphabetical order.

INTERSIL		SUGGESTED TI REPLACEMENT	PAGE NO.
ICM7555		TLC555	4-195
MOTOROLA	DIRECT TI REPLACEMENT	SUGGESTED TI REPLACEMENT	PAGE NO.
MC1445	MC1445		4-27
MC1733		uA733	4-213
MC3470	MC3470		4-29
NE555	NE555		4-37
NE592	NE592		4-53
SIGNETICS	DIRECT TI REPLACEMENT	SUGGESTED TI REPLACEMENT	PAGE NO.
NE555	NE555	TLC555	4-195
NE556	NE556	TLC556	4-203
NE592	NE592		4-53
SA555	SA555	TLC555	4-195
SA556	SA556	TLC556	4-203
SE555	SE555	TLC555	4-195
SE555C	SE555C	TLC555	4-37
SE556	SE556	TLC556	4-203
SE556C	SE556C	TLC556	4-49
uA733	uA733	4-203	
SPRAGUE		SUGGESTED TI REPLACEMENT	PAGE NO.
UGN3019		TL3019C	4-167
UGS3019		TL3019I	4-167
UGN3020		TL3020C	4-169
UGS3020		TL3020I	4-169

General Information

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Operational Amplifiers

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Voltage Comparators

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Special Functions

4

Product Previews

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Mechanical Data

6

2

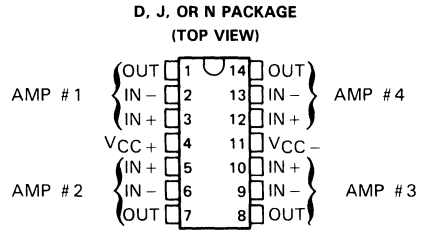
Operational Amplifiers

LF347, LF347B

WIDE-BANDWIDTH QUAD JFET-INPUT OPERATIONAL AMPLIFIERS

D2997, MARCH 1987

- Low Input Bias Current
Typically 50 pA
- Low Input Noise Current
Typically 0.01 pA/ $\sqrt{\text{Hz}}$
- Low Total Harmonic Distortion
- Low Supply Current . . . Typically 8 mA
- Wide Gain Bandwidth . . . Typically 3 MHz
- High Slew Rate . . . Typically 13 V/ μs
- Pin Compatible with the LM348



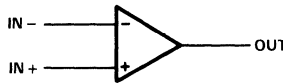
description

These devices are low-cost, high-speed, JFET-input operational amplifiers. They require low supply current yet maintain a large gain-bandwidth product and a fast slew rate. In addition, their matched high-voltage JFET inputs provide very low input bias and offset current.

The LF347 and LF347B can be used in applications such as high-speed integrators, digital-to-analog converters, sample-and-hold circuits, and many other circuits.

The LF347 and LF347B are characterized for operation from 0°C to 70°C.

symbol (each amplifier)



AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE		
		SMALL-OUTLINE (D)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C	10 mV	LF347D	LF347J	LF347N
to 70°C	5 mV	LF347BD	LF347BJ	LF347BN

D packages are available taped and reeled. Add "R" suffix to the device type. (e.g. LF347DR)

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V _{CC} +	18 V
Supply voltage, V _{CC} -	-18 V
Differential input voltage, V _{ID}	±30 V
Input voltage (see Note 1)	±15 V
Duration of output short circuit	Unlimited
Continuous total power dissipation	See Dissipation Rating Table
Operating temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds, J package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds, D or N package	260°C

NOTE 1: Unless otherwise specified, the absolute maximum negative input voltage is equal to the negative power supply voltage.

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Operational Amplifiers

LF347, LF347B

WIDE-BANDWIDTH QUAD JFET-INPUT OPERATIONAL AMPLIFIERS

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$ POWER RATING
D	680 mW	7.6 mW/ $^\circ\text{C}$	61 $^\circ\text{C}$	608 mW
J	680 mW	8.2 mW/ $^\circ\text{C}$	67 $^\circ\text{C}$	656 mW
N	680 mW	N/A	N/A	680 mW

electrical characteristics over operating free-air temperature range, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$ (unless otherwise specified)

2

Operational Amplifiers

PARAMETER	TEST CONDITIONS	LF347			LF347B			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = 0$, $R_S = 10\text{ k}\Omega$	$T_A = 25^\circ\text{C}$		5	10	$T_A = 70^\circ\text{C}$		mV
		Full range		13		7		
αV_{IO} Average temperature coefficient of input offset voltage	$V_{IC} = 0$, $R_S = 10\text{ k}\Omega$	18			18			$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current [†]	$V_{IC} = 0$	$T_J = 25^\circ\text{C}$		25	100	$T_J = 70^\circ\text{C}$		pA
		$T_J = 70^\circ\text{C}$		4		4		
I_{IB} Input bias current [†]	$V_{IC} = 0$	$T_J = 25^\circ\text{C}$		50	200	$T_J = 70^\circ\text{C}$		pA
		$T_J = 70^\circ\text{C}$		8		8		
V_{ICR} Common-mode input voltage range		± 11	-12 to 15	± 11	-12 to 15			V
V_{OM} Maximum peak output voltage swing	$R_L = 10\text{ k}\Omega$	± 12	± 13.5	± 12	± 13.5			V
A_{VD} Large-signal differential voltage	$V_O = \pm 10\text{ V}$, $R_L = 2\text{ k}\Omega$	$T_A = 25^\circ\text{C}$		25	100	$T_A = 70^\circ\text{C}$		V/mV
		Full range		15		25		
r_i Input resistance	$T_J = 25^\circ\text{C}$	10^{12}			10^{12}			Ω
CMRR Common-mode rejection ratio	$R_S \leq 10\text{ k}\Omega$	70	100	80	100			dB
kSVR Supply voltage rejection ratio	See Note 2	70	100	80	100			dB
I_{CC} Supply current		8		11				mA

operating characteristics, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{O1}/V_{O2} Crosstalk attenuation	$f = 1\text{ kHz}$	120			dB
SR Slew rate		8	13	$\text{V}/\mu\text{s}$	
B_1 Unity-gain bandwidth		3			MHz
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, $R_S = 100\ \Omega$	18			$\text{nV}/\sqrt{\text{Hz}}$
I_n Equivalent input noise current	$f = 1\text{ kHz}$	0.01			$\text{pA}/\sqrt{\text{Hz}}$

[†] Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive. Pulse techniques must be used that will maintain the junction temperatures as close to the ambient temperature as possible.

NOTE 2: Supply voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously.

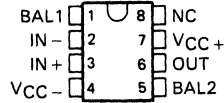
LF351

WIDE-BANDWIDTH JFET-INPUT OPERATIONAL AMPLIFIER

D2997, MARCH 1987

- **Low Input Bias Current**
Typically 50 pA
- **Low Input Noise Voltage**
Typically 18 nV/ $\sqrt{\text{Hz}}$
- **Low Input Noise Current**
Typically 0.01 pA/ $\sqrt{\text{Hz}}$
- **Low Supply Current . . . Typically 1.8 mA**
- **High Input Impedance**
Typically $10^{12} \Omega$
- **Low Total Harmonic Distortion**
- **Internally Trimmed Offset Voltage**
Typically 10 mV
- **High Slew Rate . . . Typically 13 V/ μs**
- **Wide Gain Bandwidth . . . Typically 3 MHz**
- **Pin Compatible with Standard 741**

P. D, OR JG PACKAGE
(TOP VIEW)



NC—No internal connection

2

Operational Amplifiers

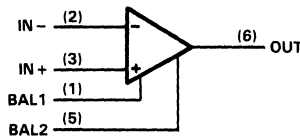
description

This device is a low-cost, high-speed, JFET-input operational amplifier with an internally trimmed input offset voltage. It requires low supply current yet maintains a large gain-bandwidth product and a fast slew rate. In addition, the matched high-voltage JFET input provides very low input bias and offset currents. It uses the same offset voltage adjustment circuits as the 741.

The LF351 can be used in applications such as high-speed integrators, digital-to-analog converters, sample-and-hold circuits, and many other circuits.

The LF351 is characterized for operation from 0°C to 70°C.

symbol (each amplifier)



AVAILABLE OPTIONS

SYMBOLIZATION		OPERATING TEMPERATURE RANGE	V _{IO} MAX at 25°C
DEVICE	PACKAGE SUFFIX		
LF351	D, JG, P	-0°C to 70°C	10 mV

The D packages are available taped and reeled. Add the suffix R to the device type when ordering. (ie., LF351DR)

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LF351 WIDE-BANDWIDTH JFET-INPUT OPERATIONAL AMPLIFIER

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+}	18 V
Supply voltage, V_{CC-}	-18 V
Differential input voltage, V_{ID}	± 30 V
Input voltage (see Note 1)	± 15 V
Duration of output short circuit	Unlimited
Continuous total power dissipation	500 mW
Operating temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds, JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds, D or P package	260°C

NOTE 1: Unless otherwise specified, the absolute maximum negative input voltage is equal to the negative power supply voltage.

electrical characteristics over operating free-air temperature range, $V_{CC+} = 15$ V, $V_{CC-} = -15$ V (unless otherwise specified)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	$V_{IC} = 0$,	$R_S = 10$ k Ω	$T_A = 25^\circ\text{C}$	5	10	mV
				Full range		13	
α_{VIO}	Average temperature coefficient of input offset voltage	$V_{IC} = 0$,	$R_S = 10$ k Ω		10		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current [†]	$V_{IC} = 0$		$T_J = 25^\circ\text{C}$	25	100	pA
				$T_J = 70^\circ\text{C}$		4	nA
I_{IB}	Input bias current [†]	$V_{IC} = 0$		$T_J = 25^\circ\text{C}$	50	200	pA
				$T_J = 70^\circ\text{C}$		8	nA
V_{ICR}	Common-mode input voltage range			± 11	-12 to 15		V
V_{OM}	Maximum peak output voltage swing	$R_L = 10$ k Ω		± 12	± 13.5		V
A_{VD}	Large-signal differential voltage	$V_O = \pm 10$ V,	$R_L = 2$ k Ω	$T_A = 25^\circ\text{C}$	25	200	V/mV
				Full range	15	200	
r_i	Input resistance	$T_J = 25^\circ\text{C}$			10^{12}		Ω
CMRR	Common-mode rejection ratio	$R_S \leq 10$ k Ω		70	100		dB
ksVR	Supply voltage rejection ratio	See Note 2		70	100		dB
I_{CC}	Supply current			1.8	3.4		mA

operating characteristics, $V_{CC+} = 15$ V, $V_{CC-} = -15$ V, $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
SR	Slew rate			8	13		V/ μs
B_1	Unity-gain bandwidth				3		MHz
V_n	Equivalent input noise voltage	$f = 1$ kHz,	$R_S = 100$ Ω		18		nV/ $\sqrt{\text{Hz}}$
I_n	Equivalent input noise current	$f = 1$ kHz			0.01		pA/ $\sqrt{\text{Hz}}$

[†] Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive. Pulse techniques must be used that will maintain the junction temperatures as close to the ambient temperature as possible.

NOTE 2: Supply voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously.

2

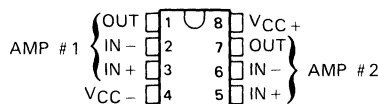
Operational Amplifiers

WIDE-BANDWIDTH DUAL JFET-INPUT OPERATIONAL AMPLIFIER

D2997, MARCH 1987—REVISED MAY 1988

- Low Input Bias Current
Typically 50 pA
- Low Input Noise Current
Typically 0.01 pA/√Hz
- Low Input Noise Voltage
Typically 18 nV/√Hz
- Low Supply Current . . . Typically 3.6 mA
- High Input Impedance
Typically 10¹² Ω
- Internally Trimmed Offset Voltage
- Wide Gain Bandwidth . . . Typically 3 MHz
- High Slew Rate . . . Typically 13 V/μs

D, JG, OR P PACKAGE
(TOP VIEW)



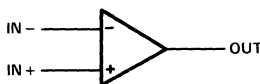
description

This device is a low-cost, high-speed, JFET-input operational amplifier with very low input offset voltage. It requires low supply current yet maintains a large gain-bandwidth product and a fast slew rate. In addition, the matched high-voltage JFET input provides very low input bias and offset currents.

The LF353 can be used in applications such as high-speed integrators, digital-to-analog converters, sample-and-hold circuits, and many other circuits.

The LF353 is characterized for operation from 0°C to 70°C.

symbol (each amplifier)



AVAILABLE OPTIONS

SYMBOLIZATION		OPERATING TEMPERATURE RANGE	V _{IO} MAX at 25°C
DEVICE	PACKAGE SUFFIX		
LF353	D,JG,P	0°C to 70°C	10 mV

The D packages are available taped and reeled. Add the suffix R to the device type when ordering. (i.e. LP353DR)

LF353 WIDE-BANDWIDTH DUAL JFET-INPUT OPERATIONAL AMPLIFIER

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+}	18 V
Supply voltage, V_{CC-}	-18 V
Differential input voltage, V_{ID}	± 30 V
Input voltage (see Note 1)	± 15 V
Duration of output short circuit	Unlimited
Continuous total power dissipation	500 mW
Operating temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds, JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds, D or P package	260°C

NOTE 1: Unless otherwise specified, the absolute maximum negative input voltage is equal to the negative power supply voltage.

electrical characteristics over operating free-air temperature range, $V_{CC+} = 15$ V, $V_{CC-} = -15$ V (unless otherwise specified)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{IO} Input offset voltage	$V_{IC} = 0$, $R_S = 10$ k Ω $T_A = 25^\circ\text{C}$ Full range		5	10 13	mV
α_{VIO} Average temperature coefficient of input offset voltage	$V_{IC} = 0$, $R_S = 10$ k Ω		10		$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current †	$V_{IC} = 0$ $T_J = 25^\circ\text{C}$ $T_J = 70^\circ\text{C}$		25	100 4	μA nA
I_{IB} Input bias current †	$V_{IC} = 0$ $T_J = 25^\circ\text{C}$ $T_J = 70^\circ\text{C}$		50	200 8	μA nA
V_{ICR} Common-mode input voltage range		± 11	-12 to 15		V
V_{OM} Maximum peak output voltage swing	$R_L = 10$ k Ω	± 12	± 13.5		V
A_{VD} Large-signal differential voltage	$V_O = \pm 10$ V, $R_L = 2$ k Ω $T_A = 25^\circ\text{C}$ Full range		25	100 15	V/mV
r_i Input resistance	$T_J = 25^\circ\text{C}$		10^{12}		Ω
CMRR Common-mode rejection ratio	$R_S \leq 10$ k Ω	70	100		dB
k_{SVR} Supply voltage rejection ratio	See Note 2	70	100		dB
I_{CC} Supply current			3.6	6.5	mA

operating characteristics, $V_{CC+} = 15$ V, $V_{CC-} = -15$ V, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{O1}/V_{O2} Crosstalk attenuation	$f = 1$ kHz		120		dB
SR Slew rate		8	13		V/ μs
B_1 Unity-gain bandwidth			3		MHz
V_n Equivalent input noise voltage	$f = 1$ kHz, $R_S = 100$ Ω		18		nV/ $\sqrt{\text{Hz}}$
I_n Equivalent input noise current	$f = 1$ kHz		0.01		pA/ $\sqrt{\text{Hz}}$

† Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive. Pulse techniques must be used that will maintain the junction temperatures as close to the ambient temperature as possible.

NOTE 2: Supply voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously.

2

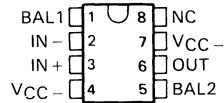
Operational Amplifiers

LF411C JFET-INPUT OPERATIONAL AMPLIFIER

D2997, MARCH 1987—REVISED MAY 1988

- **Low Input Bias Current**
Typically 50 pA
- **Low Input Noise Current**
Typically 0.01 pA/√Hz
- **Low Supply Current . . .** Typically 2.0 mA
- **High Input Impedance**
Typically 10¹² Ω
- **Low Total Harmonic Distortion**
- **Low 1/f Noise Corner . . .** Typically 50 Hz

D, JG, OR P PACKAGE
(TOP VIEW)



NC—No internal connection

2

Operational Amplifiers

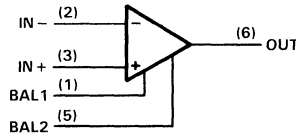
description

This device is a low-cost, high-speed, JFET-input operational amplifier with very low input offset voltage and a maximum input offset voltage drift. It requires low supply current yet maintains a large gain-bandwidth product and a fast slew rate. In addition, the matched high-voltage JFET input provides very low input bias and offset currents.

The LF411C can be used in applications such as high-speed integrators, digital-to-analog converters, sample-and-hold circuits, and many other circuits.

The LF411C is characterized for operation from 0°C to 70°C.

symbol



AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE		
		SMALL-OUTLINE (D)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	2 mV	LF411CD	LF411CJG	LF411CP

D package is available taped and reeled. Add "R" suffix to device type. (e.g. LF411CDR)

LF411C

JFET-INPUT OPERATIONAL AMPLIFIER

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+}	18 V
Supply voltage, V_{CC-}	-18 V
Differential input voltage, V_{ID}	± 30 V
Input voltage (see Note 1)	± 15 V
Duration of output short circuit	Unlimited
Continuous total power dissipation	500 mW
Operating temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds, JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds, D or P package	260°C

2

NOTE 1: Unless otherwise specified, the absolute maximum negative input voltage is equal to the negative power supply voltage.

Operational Amplifiers

electrical characteristics over operating free-air temperature range, $V_{CC+} = 15$ V, $V_{CC-} = -15$ V (unless otherwise specified)

PARAMETER		TEST CONDITIONS			MIN	TYP	MAX	UNIT	
V_{IO}	Input offset voltage	$V_{IC} = 0$, $R_S = 10$ k Ω , $T_A = 25^\circ\text{C}$			0.8		2	mV	
α_{VIO}	Average temperature coefficient of input offset voltage	$V_{IC} = 0$, $R_S = 10$ k Ω			10		20 [†]	$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current [‡]	$V_{IC} = 0$	$T_J = 25^\circ\text{C}$		25		100	pA	
			$T_J = 70^\circ\text{C}$				2	nA	
I_{IB}	Input bias current [‡]	$V_{IC} = 0$	$T_J = 25^\circ\text{C}$		50		200	pA	
			$T_J = 70^\circ\text{C}$				4	nA	
V_{ICR}	Common-mode input voltage range				± 11		-11.5 to 14.5	V	
V_{OM}	Maximum peak output voltage swing	$R_L = 10$ k Ω			± 12		± 13.5	V	
A_{VD}	Large-signal differential voltage	$V_O = \pm 10$ V, $R_L = 2$ k Ω	$T_A = 25^\circ\text{C}$		25		200	V/mV	
			Full range		15		200		
r_i	Input resistance	$T_J = 25^\circ\text{C}$					10^{12}	Ω	
CMRR	Common-mode rejection ratio	$R_S \leq 10$ k Ω			70		100	dB	
k_{SVR}	Supply voltage rejection ratio	See Note 2			70		100	dB	
I_{CC}	Supply current						2	3.4	mA

operating characteristics, $V_{CC+} = 15$ V, $V_{CC-} = -15$ V, $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS			MIN	TYP	MAX	UNIT
SR	Slew rate				8		13	V/ μs
B_1	Unity-gain bandwidth				2.7		3	MHz
V_n	Equivalent input noise voltage	$f = 1$ kHz, $R_S = 100$ Ω					18	nV/ $\sqrt{\text{Hz}}$
I_n	Equivalent input noise current	$f = 1$ kHz					0.01	pA/ $\sqrt{\text{Hz}}$

[†] At least 90% of the devices meet this limit for α_{VIO} .

[‡] Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive. Pulse techniques must be used that will maintain the junction temperatures as close to the ambient temperature as possible.

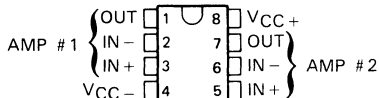
NOTE 2: Supply voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously.

LF412C DUAL JFET-INPUT OPERATIONAL AMPLIFIER

D2997, MARCH 1987—REVISED MAY 1988

- **Low Input Bias Current**
Typically 50 pA
- **Low Input Noise Current**
Typically 0.01 pA/√Hz
- **Low Supply Current . . .** Typically 4.5 mA
- **High Input Impedance**
Typically 10¹² Ω
- **Internally Trimmed Offset Voltage**
- **Wide Gain Bandwidth . . .** Typically 3 MHz
- **High Slew Rate . . .** Typically 13 V/μs

D, JG, OR P PACKAGE
(TOP VIEW)



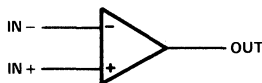
description

This device is a low-cost, high-speed, JFET-input operational amplifier with very low input offset voltage and a specified maximum input offset voltage drift. It requires low supply current yet maintains a large gain bandwidth product and a fast slew rate. In addition, the matched high-voltage JFET input provides very low input bias and offset currents.

The LF412C can be used in applications such as high-speed integrators, digital-to-analog converters, sample-and-hold circuits, and many other circuits.

The LF412C is characterized for operation from 0°C to 70°C.

symbol (each amplifier)



AVAILABLE OPTIONS

SYMBOLIZATION		OPERATING TEMPERATURE RANGE	V _{IO} MAX at 25°C
DEVICE	PACKAGE SUFFIX		
LF412C	D,JG,P	0°C to 70°C	3 mV

The D packages are available taped and reeled. Add the suffix R to the device type when ordering. (i.e. LF412CDR)

2

Operational Amplifiers

LF412C

DUAL JFET-INPUT OPERATIONAL AMPLIFIER

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+}	18 V
Supply voltage, V_{CC-}	-18 V
Differential input voltage, V_{ID}	± 30 V
Input voltage (see Note 1)	± 15 V
Duration of output short circuit	Unlimited
Continuous total power dissipation	500 mW
Operating temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds, JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds, D or P package	260°C

NOTE 1: Unless otherwise specified, the absolute maximum negative input voltage is equal to the negative power supply voltage.

electrical characteristics over operating free-air temperature range, $V_{CC+} = 15$ V, $V_{CC-} = -15$ V (unless otherwise specified)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{IO} Input offset voltage	$V_{IC} = 0$, $R_S = 10$ k Ω , $T_A = 25^\circ\text{C}$		1	3	mV
α_{VIO} Average temperature coefficient of input offset voltage	$V_{IC} = 0$, $R_S = 10$ k Ω		10	20 [†]	$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current [‡]	$V_{IC} = 0$, See Note 3	$T_J = 25^\circ\text{C}$	25	100	pA
		$T_J = 70^\circ\text{C}$		2	nA
I_{IB} Input bias current [‡]	$V_{IC} = 0$, See Note 3	$T_J = 25^\circ\text{C}$	50	200	pA
		$T_J = 70^\circ\text{C}$		4	nA
V_{ICR} Common-mode input voltage range		± 11	-11.5 to 14.5		V
V_{OM} Maximum peak output voltage swing	$R_L = 10$ k Ω	± 12	± 13.5		V
A_{VD} Large-signal differential voltage	$V_O = \pm 10$ V, $R_L = 2$ k Ω	$T_A = 25^\circ\text{C}$	25	200	V/mV
		Full range	15	200	
r_i Input resistance	$T_J = 25^\circ\text{C}$		10^{12}		Ω
CMRR Common-mode rejection ratio	$R_S \leq 10$ k Ω		70	100	dB
kSVR Supply voltage rejection ratio	See Note 2		70	100	dB
I_{CC} Supply current			4.5	6.8	mA

operating characteristics, $V_{CC+} = 15$ V, $V_{CC-} = -15$ V, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{O1}/V_{O2} Crosstalk attenuation	$f = 1$ kHz		120		dB
SR Slew rate		8	13		V/ μs
B_1 Unity-gain bandwidth		2.7	3		MHz
V_n Equivalent input noise voltage	$f = 1$ kHz, $R_S = 100$ Ω		18		nV/ $\sqrt{\text{Hz}}$
I_n Equivalent input noise current	$f = 1$ kHz		0.01		pA/ $\sqrt{\text{Hz}}$

[†] At least 90% of the devices meet this limit for α_{VIO} .

[‡] Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive. Pulse techniques must be used that will maintain the junction temperatures as close to the ambient temperature as possible.

NOTE 2: Supply voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously.

2

Operational Amplifiers

LM101A, LM201A, LM301A HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

D961, OCTOBER 1979—REVISED JUNE 1988

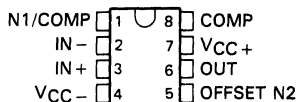
- Low Input Currents
- Low Input Offset Parameters
- Frequency and Transient Response Characteristics Adjustable
- Short-Circuit Protection
- Offset-Voltage Null Capability
- No Latch-Up
- Wide Common-Mode and Differential Voltage Ranges
- Same Pin Assignments as μ A709
- Designed to be Interchangeable with National Semiconductor LM101A and LM301A

description

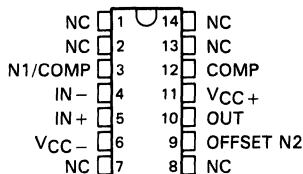
The LM101A, LM201A, and LM301A are high-performance operational amplifiers featuring very low input bias current and input offset voltage and current to improve the accuracy of high-impedance circuits using these devices. The high common-mode input voltage range and the absence of latch-up make these amplifiers ideal for voltage-follower applications. The devices are protected to withstand short circuits at the output. The external compensation of these amplifiers allows the changing of the frequency response (when the closed-loop gain is greater than unity) for applications requiring wider bandwidth or higher slew rate. A potentiometer may be connected between the offset-null inputs (N1 and N2), as shown in Figure 7, to null out the offset voltage.

The LM101A is characterized for operation over the full military temperature range of -55°C to 125°C , the LM201A is characterized for operation from -25°C to 85°C , and the LM301A is characterized for operation from 0°C to 70°C .

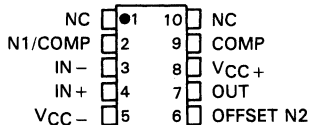
D, JG, OR P PACKAGE
(TOP VIEW)



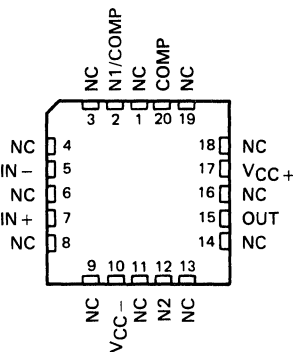
LM101A
W FLAT PACKAGE
(TOP VIEW)



LM101A
U FLAT PACKAGE
(TOP VIEW)

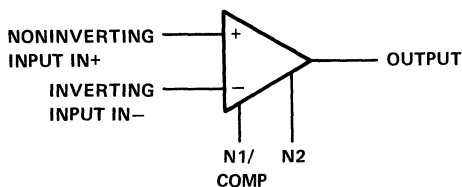


LM101A
FK CHIP-CARRIER PACKAGE
(TOP VIEW)



NC—No internal connection

symbol



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Operational Amplifiers

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LM101A, LM201A, LM301A HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

AVAILABLE OPTIONS

T _A	V _{IO} MAX at 25°C	PACKAGE					
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)	FLAT PACK (U)	FLAT PACK (W)
0°C to 70°C	7.5 mV	LM301AD	—	LM301AJG	LM301AP	—	—
–25°C to 85°C	2 mV	LM201AD	—	LM201AJG	LM201AP	—	—
–55°C to 125°C	2 mV	—	LM101AFK	LM101AJG	—	LM101AU	LM101AW

The D package is available taped and reeled. Add the suffix R to the device type. (i.e., LM301ADR)

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	LM101A	LM201A	LM301A	UNIT	
Supply voltage V _{CC+} (see Note 1)	22	22	18	V	
Supply voltage V _{CC–} (see Note 1)	–22	–22	–18	V	
Differential input voltage (see Note 2)	±30	±30	±30	V	
Input voltage (either input, see Notes 1 and 3)	±15	±15	±15	V	
Voltage between either offset null terminal (N1/N2) and V _{CC–}	–0.5 to 2	–0.5 to 2	–0.5 to 2	V	
Duration of output short-circuit (see Note 4)	unlimited	unlimited	unlimited		
Continuous total power dissipation	See Dissipation Rating Table				
Operating free-air temperature range	–55 to 125	–25 to 85	0 to 70	°C	
Storage temperature range	–65 to 150	–65 to 150	–65 to 150	°C	
Case temperature for 60 seconds: FK package	260			°C	
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	JG, U, or W package	300	300	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package		260	260	°C

NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC–}.

2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.

3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.

4. The output may be shorted to ground or either power supply. For the LM101A only, the unlimited duration of the short-circuit applies at (or below) 125°C case temperature or 75°C free-air temperature. For the LM201A only, the unlimited duration of the short-circuit applies at (or below) 85°C case temperature or 75°C free-air temperature.

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C	DERATING FACTOR	DERATE ABOVE T _A	T _A = 70°C	T _A = 85°C	T _A = 125°C
	POWER RATING			POWER RATING	POWER RATING	POWER RATING
D	500 mW	5.8 mW/°C	64°C	464 mW	377 mW	N/A
FK	500 mW	11.0 mW/°C	105°C	500 mW	500 mW	275 mW
JG (LM101A)	500 mW	8.4 mW/°C	90°C	500 mW	500 mW	210 mW
JG (LM201A, LM301A)	500 mW	6.6 mW/°C	74°C	500 mW	429 mW	N/A
P	500 mW	N/A	N/A	500 mW	500 mW	N/A
U	500 mW	5.4 mW/°C	57°C	432 mW	351 mW	135 mW
W	500 mW	8.0 mW/°C	87°C	500 mW	500 mW	200 mW

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Operational Amplifiers

LM101A, LM201A, LM301A HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $C_C = 30$ pF (see Note 5)

PARAMETER	TEST CONDITIONS†		LM101A, LM201A			LM301A			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0$ V	25°C	0.6	2		2	7.5	mV	
		Full range			3		10		
α_{VIO} Average temperature coefficient of input offset voltage	$V_O = 0$ V	Full range	3	15		6	30	$\mu\text{V}/^\circ\text{C}$	
I_{IO} Input offset current		25°C	1.5	10		3	50	nA	
		Full range			20		70		
α_{IIO} Average temperature coefficient of input offset current		$T_A = -55^\circ\text{C}$ to 25°C	0.02	0.2				nA/°C	
		$T_A = 25^\circ\text{C}$ to MAX	0.01	0.1					
		$T_A = 0^\circ\text{C}$ to 25°C				0.02	0.6		
		$T_A = 25^\circ\text{C}$ to 70°C				0.01	0.3		
I_{IB} Input bias current		25°C	30	75		70	250	nA	
		Full range			100		300		
V_{ICR} Common-mode input voltage range	See Note 6	Full range	± 15			± 12		V	
V_{OPP} Maximum peak-to-peak output voltage swing	$V_{CC\pm} = \pm 15$ V, $R_L = 10$ k Ω	25°C	24	28		24	28	V	
		Full range	24			24			
		25°C	20	26		20	26		
		Full range	20			20			
A_{VD} Large-signal differential voltage amplification	$V_{CC\pm} = \pm 15$ V, $V_O = \pm 10$ V, $R_L \geq 2$ k Ω	25°C	50	200		25	200	V/mV	
		Full range	25			15			
r_i Input resistance		25°C	1.5	4		0.5	2	M Ω	
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR}$ min	25°C	80	98		70	90	dB	
		Full range	80			70			
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IO}$)		25°C	80	98		70	96	dB	
		Full range	80			70			
I_{CC} Supply current	No load, $V_O = 0$, See Note 6	25°C	1.8	3		1.8	3	mA	
		MAX	1.2	2.5					

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for LM101A is -55°C to 125°C , for LM201A is -25°C to 85°C , and for LM301A is 0°C to 70°C .

NOTES: 5. Unless otherwise noted, $V_{CC\pm} = \pm 5$ V to ± 20 V for LM101A and LM201A, and $V_{CC\pm} = \pm 5$ V to ± 15 V for LM301A. All typical values are at $V_{CC\pm} = \pm 15$ V.

6. For LM101A and LM201A, $V_{CC\pm} = \pm 20$ V. For LM301A, $V_{CC\pm} = \pm 15$ V.

2

Operational Amplifiers

LM101A, LM201A, LM301A HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

2
Operational Amplifiers

TYPICAL CHARACTERISTICS

INPUT OFFSET CURRENT
VS
FREE-AIR TEMPERATURE

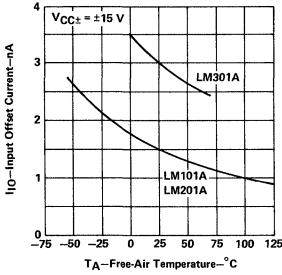


FIGURE 1

INPUT BIAS CURRENT
VS
FREE-AIR TEMPERATURE

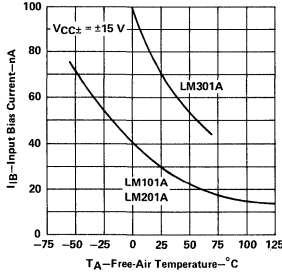


FIGURE 2

MAXIMUM PEAK-TO-PEAK
OUTPUT VOLTAGE (WITH
SINGLE-POLE COMPENSATION)
VS FREQUENCY

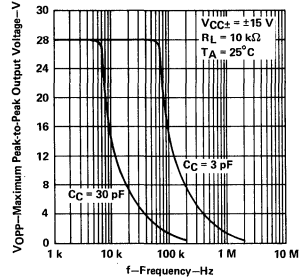


FIGURE 3

OPEN-LOOP LARGE-SIGNAL
DIFFERENTIAL
VOLTAGE AMPLIFICATION
VS
SUPPLY VOLTAGE

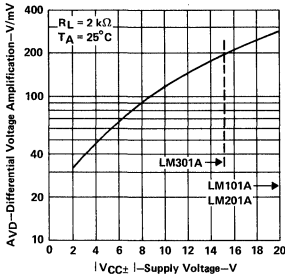


FIGURE 4

OPEN-LOOP LARGE-SIGNAL
DIFFERENTIAL
VOLTAGE AMPLIFICATION
VS
FREQUENCY

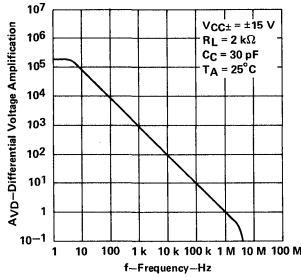


FIGURE 5

VOLTAGE-FOLLOWER
LARGE-SIGNAL PULSE RESPONSE

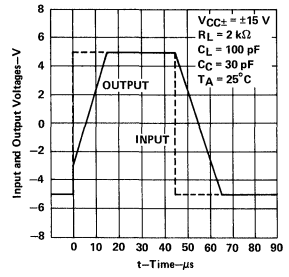


FIGURE 6

TYPICAL APPLICATION DATA

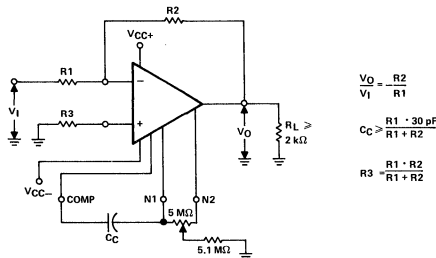


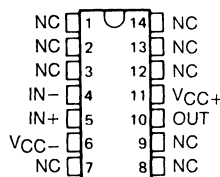
FIGURE 7. INVERTING CIRCUIT WITH ADJUSTABLE GAIN,
SINGLE-POLE COMPENSATION, AND OFFSET ADJUSTMENT

LM107, LM207, LM307 HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

D962, DECEMBER 1970—REVISED JUNE 1988

- Low Input Currents
- No Frequency Compensation Required
- Low Input Offset Parameters
- Short-Circuit Protection
- No Latch-Up
- Wide Common-Mode and Differential Voltage Ranges

LM107 . . . J OR W PACKAGE
(TOP VIEW)



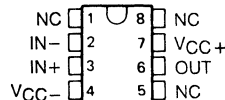
description

The LM107, LM207, and LM307 are high-performance operational amplifiers featuring very low input bias current and input offset voltage and current to improve the accuracy of high-impedance circuits using these devices.

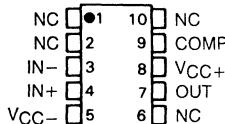
The high common-mode input voltage range and the absence of latch-up make these amplifiers ideal for voltage follower applications. The devices are short-circuit protected and the internal frequency compensation ensures stability without external components.

The LM107 is characterized for operation over the full military temperature range of -55°C to 125°C , the LM207 is characterized for operation from -25°C to 85°C , and the LM307 is characterized for operation from 0°C to 70°C .

LM107 . . . JG PACKAGE
LM207, LM307 . . . D, JG, OR P PACKAGE
(TOP VIEW)

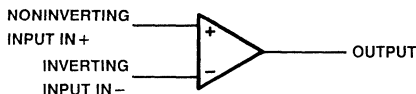


LM107 . . . U FLAT PACKAGE
(TOP VIEW)



NC—No internal connection

symbol



AVAILABLE OPTIONS

T _A	V _{IO} MAX at 25°C	PACKAGE					
		SMALL OUTLINE (D)	CERAMIC (J)	CERAMIC DIP (JG)	PLASTIC DIP (P)	FLAT PACK (U)	FLAT PACK (W)
0°C to 70°C	7.5 mV	LM307D	—	LM307JG	LM307P	—	—
-25°C to 85°C	2 mV	LM207D	—	LM207JG	LM207P	—	—
-55°C to 125°C	2 mV	—	LM107J	LM107JG	—	LM107U	LM107W

The D package is available taped and reeled. Add the suffix R to the device type when ordering. (e.g., LM307DR)

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LM107, LM207, LM307 HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

2

Operational Amplifiers

	LM107	LM207	LM307	UNIT
Supply voltage V_{CC+} (see Note 1)	22	22	18	V
Supply voltage V_{CC-} (see Note 1)	-22	-22	-18	V
Differential input voltage (see Note 2)	± 30	± 30	± 30	V
Input voltage (either input, see Notes 1 and 3)	± 15	± 15	± 15	V
Duration of output short-circuit (see Note 4)	unlimited	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table			
Operating free-air temperature range	-55 to 125	-25 to 85	0 to 70	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J, JG, U, or W package	300	300	300	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package		260	260	$^{\circ}\text{C}$

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
 4. The output may be shorted to ground or either power supply. For the LM107 only, the unlimited duration of the short-circuit applies at (or below) 125 $^{\circ}\text{C}$ case temperature or 75 $^{\circ}\text{C}$ free-air temperature. For the LM207 only, the unlimited duration of the short-circuit applies at (or below) 85 $^{\circ}\text{C}$ case temperature or 75 $^{\circ}\text{C}$ free-air temperature.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^{\circ}\text{C}$	$T_A = 85^{\circ}\text{C}$	$T_A = 125^{\circ}\text{C}$
	POWER RATING			POWER RATING	POWER RATING	POWER RATING
D	500 mW	5.8 mW/ $^{\circ}\text{C}$	64 $^{\circ}\text{C}$	464 mW	377 mW	—
J (LM107)	500 mW	11.0 mW/ $^{\circ}\text{C}$	105 $^{\circ}\text{C}$	500 mW	500 mW	275 mW
JG (LM107)	500 mW	8.4 mW/ $^{\circ}\text{C}$	90 $^{\circ}\text{C}$	500 mW	500 mW	210 mW
JG (LM207, LM307)	500 mW	6.6 mW/ $^{\circ}\text{C}$	74 $^{\circ}\text{C}$	500 mW	429 mW	—
P	500 mW	N/A	N/A	500 mW	500 mW	—
U	500 mW	5.4 mW/ $^{\circ}\text{C}$	57 $^{\circ}\text{C}$	432 mW	351 mW	135 mW
W	500 mW	8.0 mW/ $^{\circ}\text{C}$	87 $^{\circ}\text{C}$	500 mW	500 mW	200 mW

LM107, LM207, LM307 HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature (see Note 5)

PARAMETER		TEST CONDITIONS†		LM107, LM207			LM307			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	V _O = 0	25°C	0.6	2		2	7.5	mV	
			Full range			3		10		
α _{VIO}	Average temperature coefficient of input offset voltage	V _O = 0	Full range	3	15		6	30	μV/°C	
I _{IO}	Input offset current	V _O = 0	25°C	1.5	10		3	50	nA	
			Full range			20		70		
α _{IIO}	Average temperature coefficient of input offset current	T _A = -55°C to 25°C		0.02	0.2				nA/°C	
		T _A = 25°C to MAX		0.01	0.1					
		T _A = 0°C to 25°C					0.02	0.6		
		T _A = 25°C to 70°C					0.01	0.3		
I _{IB}	Input bias current		25°C	30	75		70	250	nA	
			Full range			100		300		
V _{ICR}	Common-mode input voltage range	See Note 6	Full range	±15			±12		V	
V _{OPP}	Maximum peak-to-peak output voltage swing	V _{CC±} = ±15 V, R _L = 10 kΩ		25°C	24	28	24	28	V	
				Full range	24		24			
		V _{CC±} = ±15 V, R _L = 2 kΩ		25°C	20	26	20	26		
				Full range	20		20			
A _{VD}	Large-signal differential voltage amplification	V _{CC±} = ±15 V, V _O = ±10 V, R _L ≥ 2 kΩ		25°C	50	200	25	200	V/mV	
				Full range	25		15			
r _i	Input resistance		25°C	1.5	4		0.5	2	MΩ	
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICR} min	25°C	80	98		70	90	dB	
			Full range	80			70			
k _{SVR}	Supply voltage rejection ratio (ΔV _{CC±} /ΔV _{IO})		25°C	80	98		70	96	dB	
			Full range	80			70			
I _{CC}	Supply current	No load, V _O = 0, See Note 6	25°C	1.8	3		1.8	3	mA	
			MAX	1.2	2.5					

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for LM107 is -55°C to 125°C, for LM207 is -25°C to 85°C, and for LM307 is 0°C to 70°C.

NOTES: 5. Unless otherwise noted V_{CC±} = ±5 V to ±20 V for LM107 and LM207, and V_{CC±} = ±5 V to ±15 V for LM307. All typical values are at V_{CC±} = ±15 V.

6. For LM107 and LM207, V_{CC±} = ±20 V. For LM307, V_{CC±} = ±15 V.

2

Operational Amplifiers

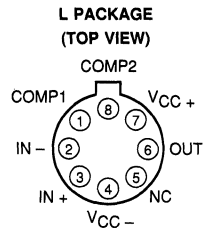
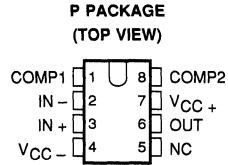
2

Operational Amplifiers

LM108, LM108A, LM308, LM308A OPERATIONAL AMPLIFIERS

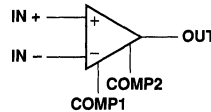
D2808, OCTOBER 1983 – REVISED MARCH 1989

- **Input Offset Current . . . 200 pA Max at 25°C for LM108, LM108A**
- **Input Bias Current . . . 2 nA Max at 25°C for LM108, LM108A**
- **Supply Current . . . 600 μA Max at 25°C for LM108, LM108A**
- **Input Offset Voltage . . . 500 μV Max at 25°C for LM108A, LM308A**
- **Offset Voltage Temperature Coefficient . . . 5 μV/°C Max for LM108A, LM308A**
- **Supply Voltage Range . . . ±2 V to ±18 V**
- **Applications:**
Integrators
Transducer Amplifiers
Analog Memories
Light Meters
- **Designed To Be Interchangeable with National LM108 Series and Linear Technology LM108 Series**



NC – No internal connection
Pin 4 (L package) is in electrical contact with the case.

symbol



description

The LM108 series of precision operational amplifiers is particularly well-suited for high-source-impedance applications requiring low input offset and bias currents as well as low power dissipation. Unlike FET input amplifiers, the input offset and bias currents of the LM108 series do not vary significantly with temperature. Advanced design, processing, and testing techniques make this series a superior choice over previous devices. For applications requiring higher performance, see the LT1008 and LT1012.

The LM108 and LM108A are characterized for operation over the full military temperature range of -55°C to 125°C. The LM308 and LM308A are characterized for operation from 0°C to 70°C.

AVAILABLE OPTIONS

T _A	V _{IO} max AT 25°C	PACKAGE	
		METAL CAN (L)	PLASTIC DIP (P)
0°C to 70°C	0.5 mV	LM308AL	LM308AP
	7.5 mV	LM308L	LM308P
-55°C to 125°C	0.5 mV	LM108AL	LM108AP
	2 mV	LM108L	LM108P

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS
INSTRUMENTS**

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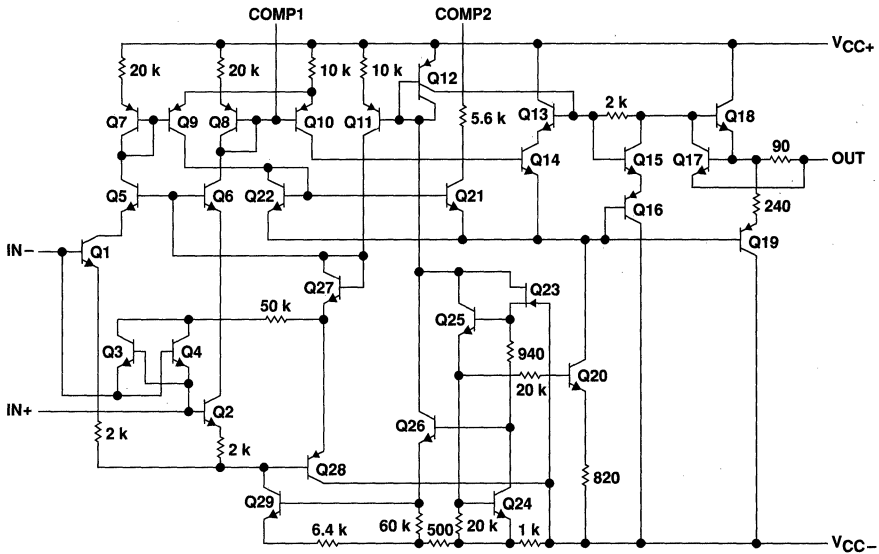
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Operational Amplifiers

LM108, LM108A, LM308, LM308A OPERATIONAL AMPLIFIERS

schematic



All resistor values shown are nominal and in ohms.

absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1): LM108, LM108A	20 V
LM308, LM308A	18 V
Supply voltage, V_{CC-} (see Note 1): LM108, LM108A	-20 V
LM308, LM308A	-18 V
Input voltage range, V_I (see Note 2)	± 15 V
Differential input current (see Notes 3 and 4)	± 10 mA
Duration of output short-circuit at (or below) 25°C (see Note 5)	unlimited
Operating free-air temperature, T_A : LM108, LM108A	-55°C to 125°C
LM308, LM308A	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: L package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: P package	260°C

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
2. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
3. The inputs are shunted with two opposite-facing base-emitter diodes for over-voltage protection. Therefore, excessive current will flow if a differential input voltage in excess of approximately 1 V is applied between the inputs unless some limiting resistance is used.
4. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
5. The output may be shorted to ground or either power supply.

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Operational Amplifiers

LM108, LM108A OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{CC\pm} = \pm 5\text{ V}$ to $\pm 20\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A^\dagger	LM108A			LM108			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$R_S = 50\ \Omega$	25°C		0.3	0.5		0.7	2	mV
		Full range			1			3	
α_{VIO} Temperature coefficient of input offset voltage		Full range		1	5		3	15	$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current		25°C		0.05	0.2		0.05	0.2	nA
		Full range			0.4			0.4	
α_{IIO} Temperature coefficient of input offset current		Full range		0.5	2.5		0.5	2.5	$\text{pA}/^\circ\text{C}$
I_{IB} Input bias current		25°C		0.5	2		0.5	2	nA
		Full range			3			3	
V_{ICR} Common-mode input voltage range	$V_{CC\pm} = \pm 15\text{ V}$	Full range	± 13.5			± 13.5			V
V_{OM} Maximum peak output voltage swing	$V_{CC\pm} = \pm 15\text{ V}$, $R_L = 10\text{ k}\Omega$	Full range	± 13			± 13			V
A_{VD} Large-signal differential voltage amplification	$V_{CC\pm} = \pm 15\text{ V}$, $V_O = \pm 10\text{ V}$, $R_L \geq 10\text{ k}\Omega$	25°C	80	300		50	300		V/mV
		Full range	40			25			
r_i Input resistance		25°C	30	70		30	70		$\text{M}\Omega$
CMRR Common-mode rejection ratio		Full range	96			85			dB
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{CC\pm} / \Delta V_{IO}$)		Full range	96			80			dB
I_{CC} Supply current		25°C		0.3	0.6		0.3	0.6	mA
		125°C		0.15	0.4		0.15	0.4	

† Full range is -55°C to 125°C .

LM308, LM308A OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{CC\pm} = \pm 5 \text{ V}$ to $\pm 15 \text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A †	LM308A			LM308			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$R_S = 50 \Omega$	25°C		0.3	0.5		2	7.5	mV
		Full range			0.73			10	
α_{VIO} Temperature coefficient of input offset voltage		Full range		2	5		6	30	$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current		25°C		0.2	1		0.2	1	nA
		Full range			1.5			1.5	
α_{IIO} Temperature coefficient of input offset current		Full range		2	10		2	10	$\text{pA}/^\circ\text{C}$
I_{IB} Input bias current		25°C		1.5	7		1.5	7	nA
		Full range			10			10	
V_{ICR} Common-mode input voltage range	$V_{CC\pm} = \pm 15 \text{ V}$	Full range	± 14			± 14			V
V_{OM} Maximum peak output voltage swing	$V_{CC\pm} = \pm 15 \text{ V}$, $R_L = 10 \text{ k}\Omega$	Full range	± 13			± 13			V
A_{VD} Large-signal differential voltage amplification	$V_{CC\pm} = \pm 15 \text{ V}$, $V_O = \pm 10 \text{ V}$, $R_L \geq 10 \text{ k}\Omega$	25°C	80	300		25	300		V/mV
		Full range	60			15			
r_i Input resistance		25°C	10	40		10	40		M Ω
CMRR Common-mode rejection ratio		Full range	96			80			dB
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{CC\pm} / \Delta V_{IO}$)		Full range	96			80			dB
I_{CC} Supply current		25°C		0.3	0.8		0.3	0.8	mA

†Full range is 0°C to 70°C.

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Operational Amplifiers

TYPICAL CHARACTERISTICS†

EQUIVALENT INPUT OFFSET VOLTAGE
VS
MATCHED SOURCE RESISTANCE

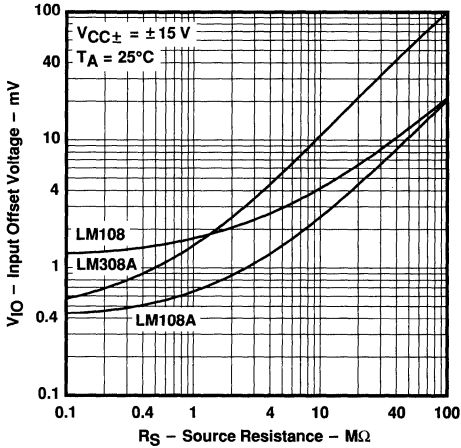


FIGURE 1

TEMPERATURE COEFFICIENT
OF EQUIVALENT INPUT OFFSET VOLTAGE
VS
MATCHED SOURCE RESISTANCE

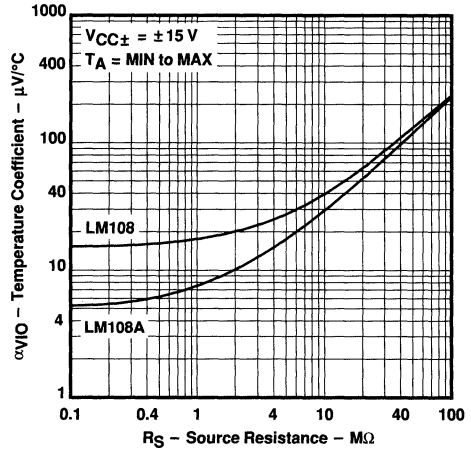


FIGURE 2

LM108, LM108A
INPUT BIAS and OFFSET CURRENTS
VS
FREE-AIR TEMPERATURE

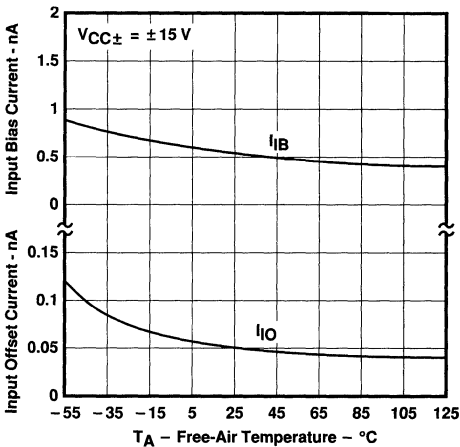


FIGURE 3

SUPPLY CURRENT
VS
SUPPLY VOLTAGE

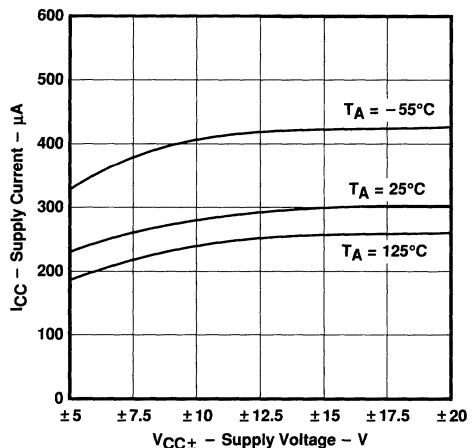


FIGURE 4

†Data above 70°C , below 0°C , or from $V_{CC\pm} = \pm 18\text{ V}$ to $\pm 20\text{ V}$ are applicable to LM108 and LM108A devices only.

TYPICAL CHARACTERISTICS†

**MAXIMUM PEAK OUTPUT VOLTAGE SWING
 VS
 FREQUENCY**

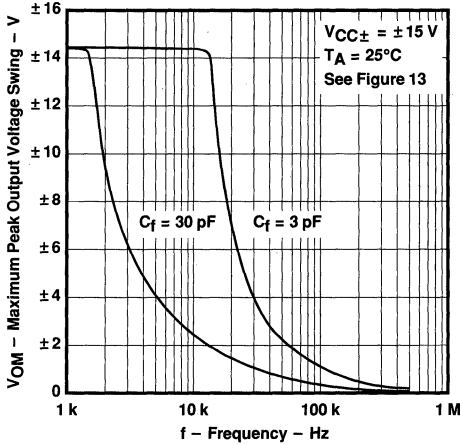


FIGURE 5

**MAXIMUM PEAK OUTPUT VOLTAGE SWING
 VS
 OUTPUT CURRENT**

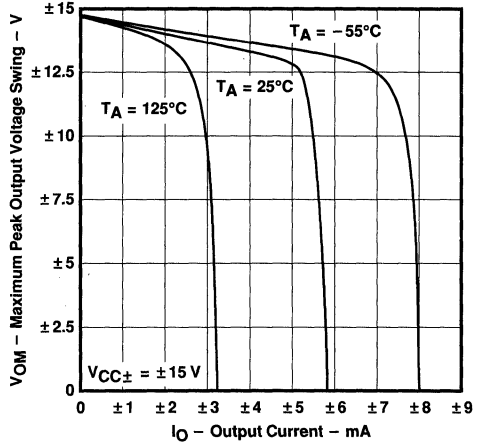


FIGURE 6

**DIFFERENTIAL VOLTAGE AMPLIFICATION
 VS
 SUPPLY VOLTAGE**

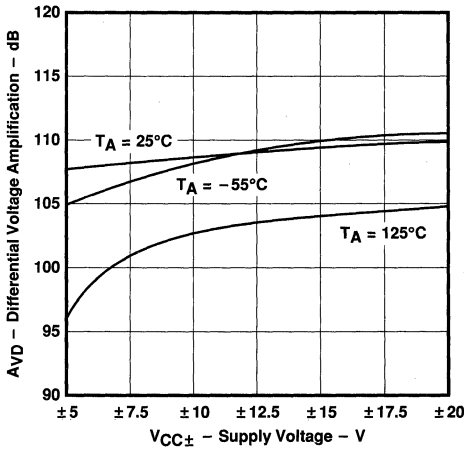


FIGURE 7

**DIFFERENTIAL VOLTAGE AMPLIFICATION
 and PHASE DELAY
 VS
 FREQUENCY**

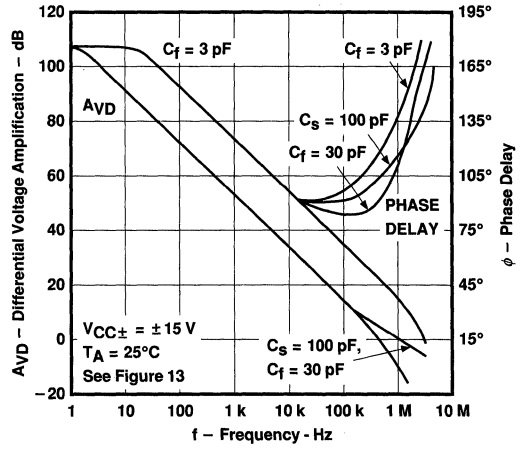


FIGURE 8

†Data above 70°C, below 0°C, or from $V_{CC\pm} = \pm 18$ V to ± 20 V are applicable to LM108 and LM108A devices only.

TYPICAL CHARACTERISTICS

SUPPLY VOLTAGE REJECTION RATIO
VS
FREQUENCY

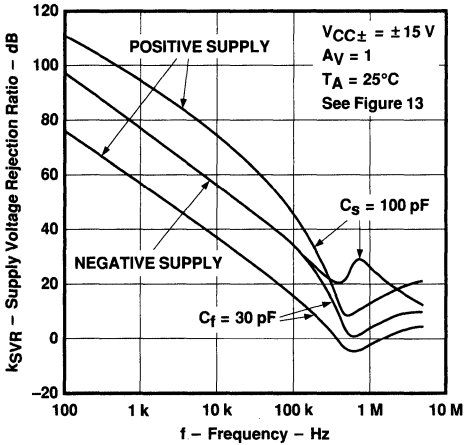


FIGURE 9

CLOSED-LOOP OUTPUT IMPEDANCE
VS
FREQUENCY

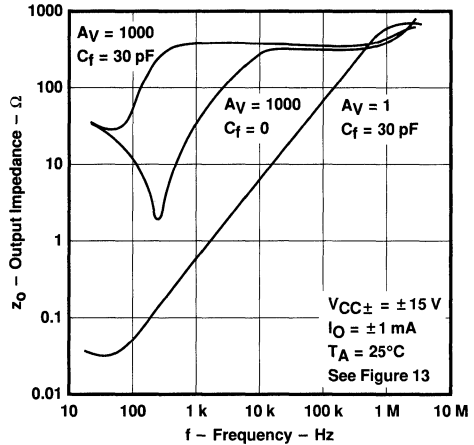


FIGURE 10

EQUIVALENT INPUT NOISE VOLTAGE
VS
FREQUENCY

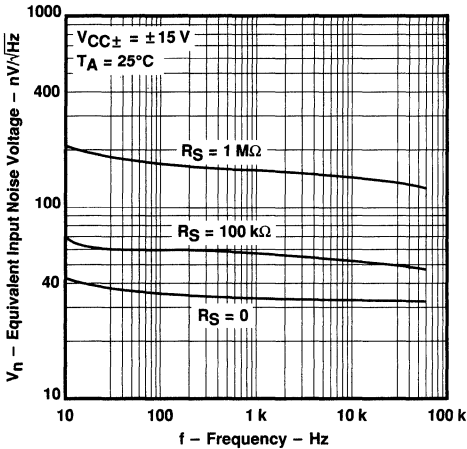


FIGURE 11

VOLTAGE FOLLOWER
PULSE RESPONSE

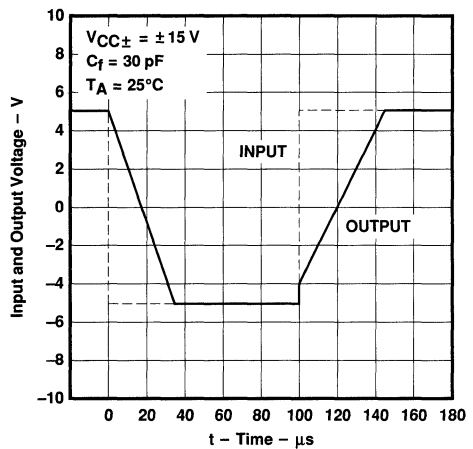


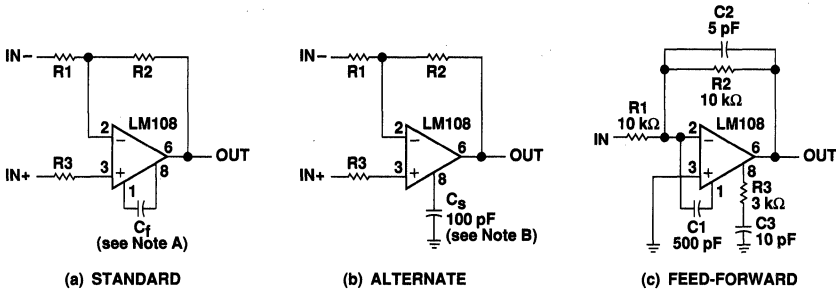
FIGURE 12

LM108, LM108A, LM308, LM308A OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

frequency compensation

Figure 13 shows the frequency compensation circuits for standard compensation, alternate compensation, and feed-forward compensation. The alternate compensation circuit improves supply voltage rejection by a factor of ten.



NOTES: A. $C_f \geq R_1 C_O / (R_1 + R_2)$, $C_O = 30 \text{ pF}$, bandwidth and slew rate are proportional to $1/C_f$.
B. Bandwidth and slew rate are proportional to $1/C_s$.

FIGURE 13. FREQUENCY COMPENSATION CIRCUITS

input guarding

Input guarding is used to reduce surface leakage (see Figure 14). Both sides of the board must be guarded. Bulk leakage reduction is less than surface leakage reduction and depends on the guard-ring width. The guard ring is connected to a low-impedance point at the same potential as the sensitive input leads. Connections for various op-amp configurations are shown in Figure 15.

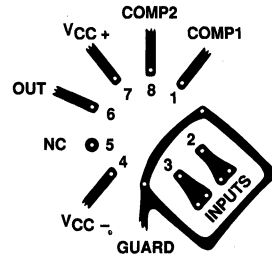


FIGURE 14. INPUT GUARDING

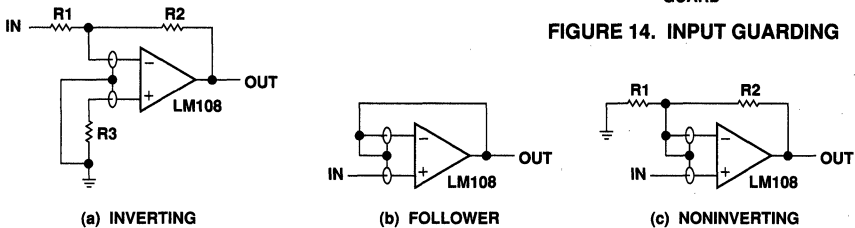


FIGURE 15. GUARD RING CONNECTIONS FOR VARIOUS OP AMP CONFIGURATIONS

TYPICAL APPLICATION DATA

input protection

Current is limited by R2 even when the input is connected to a voltage source outside the common-mode range [see Figure 16(a)]. If one supply reverses, current is controlled by R1. These resistors do not affect normal operation. The input resistor controls the current when the input exceeds the supply voltages, when the power for the op amp is turned off, or when the output is shorted [see Figure 16(b)].

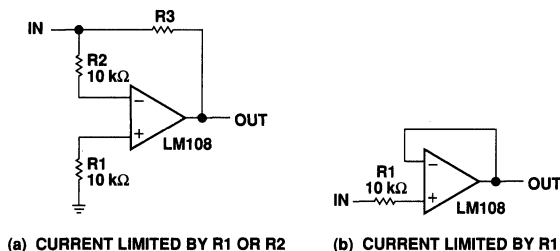
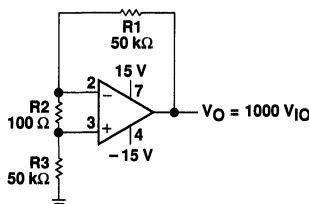


FIGURE 16. INPUT PROTECTION

input offset voltage testing

The test circuit for input offset voltage is shown in Figure 17. This circuit is also used as the burn-in configuration with supply voltages equal to ± 20 V, $R_1 = R_3 = 10$ k Ω , $R_2 = 200$ Ω , $AV = 100$.



NOTE A: Resistors must have low thermoelectric potential.

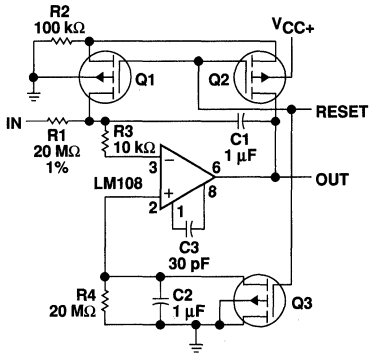
FIGURE 17. TEST CIRCUIT FOR INPUT OFFSET VOLTAGE

LM108, LM108A, LM308, LM308A OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

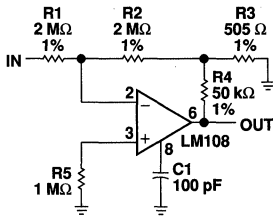
2

Operational Amplifiers



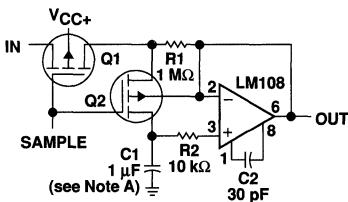
NOTE A: Q1 and Q3 should not have internal gate-protection diodes.

FIGURE 18. LOW-DRIFT INTEGRATOR WITH RESET



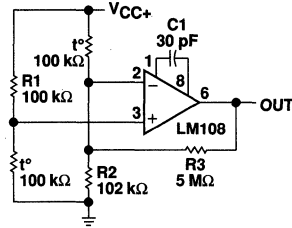
NOTE A: $R2 > R1$, $R2 \gg R3$,
 $A_v = R2(R3 + R4)/R1R3$.

FIGURE 20. INVERTING AMPLIFIER WITH HIGH INPUT RESISTANCE



NOTES: A. Teflon, polyethylene, or poly-carbonate dielectric capacitor.
B. Worst case drift is less than 2.5 mV/s.

FIGURE 22. SAMPLE-AND-HOLD AMPLIFIER



NOTE A: $R1 = R2R3/(R2 + R3)$.

FIGURE 19. AMPLIFIER FOR BRIDGE TRANSDUCERS

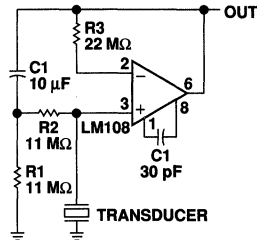
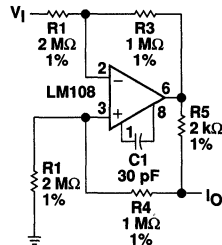


FIGURE 21. AMPLIFIER FOR PIEZOELECTRIC TRANSDUCERS



NOTE A: $I_O = (R3)V_1/R1R5$
 $R3 = R4 + R5$
 $R1 = R2$

FIGURE 23. BILATERAL CURRENT SOURCE

LM108, LM108A, LM308, LM308A OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

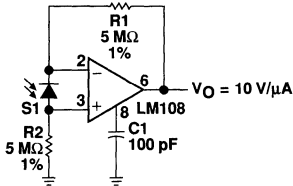
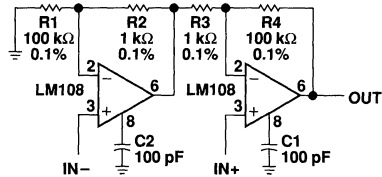
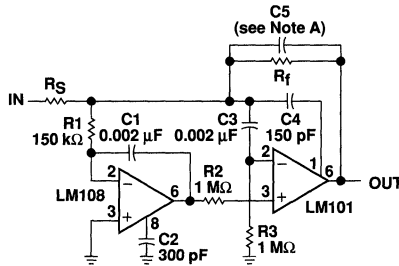


FIGURE 24. AMPLIFIER FOR
PHOTODIODE SENSOR



NOTE A: $R_1 = R_4, R_2 = R_3, A_V = 1 + R_1/R_2$

FIGURE 25. DIFFERENTIAL-INPUT
INSTRUMENTATION AMPLIFIER



- NOTES: A. $C_5 = 6 \times 10^{-8}/R_f$
 B. Power bandwidth = 250 kHz
 C. Small-signal bandwidth = 3.5 MHz
 D. Slew Rate = 10 V/μs
 E. The LM101 increases speed, raises high- and low-frequency gain, increases output drive capability, and eliminates thermal feedback.

FIGURE 26. FAST SUMMING AMPLIFIER

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Operational Amplifiers

LM124, LM224, LM224A, LM324, LM324A, LM2902 QUADRUPE OPERATIONAL AMPLIFIERS

D1990, SEPTEMBER 1975—REVISED JANUARY 1989

- **Wide Range of Supply Voltages:**
Single Supply . . . 3 V to 30 V
(LM2902 . . . 3 V to 26 V),
or Dual Supplies
- **Low Supply Current Drain Independent of Supply Voltage . . . 0.8 mA Typ**
- **Common-Mode Input Voltage Range Includes Ground Allowing Direct Sensing near Ground**
- **Low Input Bias and Offset Parameters:**
Input Offset Voltage . . . 3 mV Typ
A Versions . . . 2 mV Typ
Input Offset Current . . . 2 nA Typ
Input Bias Current . . . 20 nA Typ
A Versions . . . 15 nA Typ
- **Differential Input Voltage Range Equal to Maximum-Rated Supply Voltage . . . 32 V (26 V for LM2902)**
- **Open-Loop Differential Voltage Amplification . . . 100 V/mV Typ**
- **Internal Frequency Compensation**

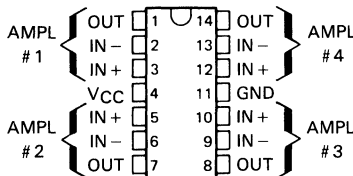
description

These devices consist of four independent, high-gain frequency-compensated operational amplifiers that were designed specifically to operate from a single supply over a wide range of voltages. Operation from split supplies is also possible so long as the difference between the two supplies is 3 V to 30 V (for the LM2902, 3 V to 26 V), and Pin 4 is at least 1.5 V more positive than the input common-mode voltage. The low supply current drain is independent of the magnitude of the supply voltage.

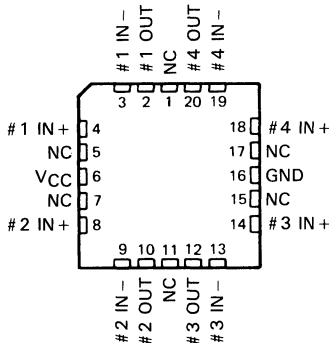
Applications include transducer amplifiers, d-c amplification blocks, and all the conventional operational amplifier circuits that now can be more easily implemented in single-supply-voltage systems. For example, the LM124 can be operated directly off of the standard 5-V supply that is used in digital systems and will easily provide the required interface electronics without requiring additional ± 15 -V supplies.

The LM124 is characterized for operation over the full military temperature range of -55°C to 125°C . The LM2902 is characterized for operation from -40°C to 105°C , the LM224 and LM224A from -25°C to 85°C , and the LM324 and LM324A from 0°C to 70°C .

LM124 . . . J OR W PACKAGE
ALL OTHERS . . . D, J, OR N PACKAGES
(TOP VIEW)

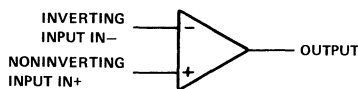


LM124
FK CHIP CARRIER PACKAGE
(TOP VIEW)



NC—No internal connection

symbol (each amplifier)



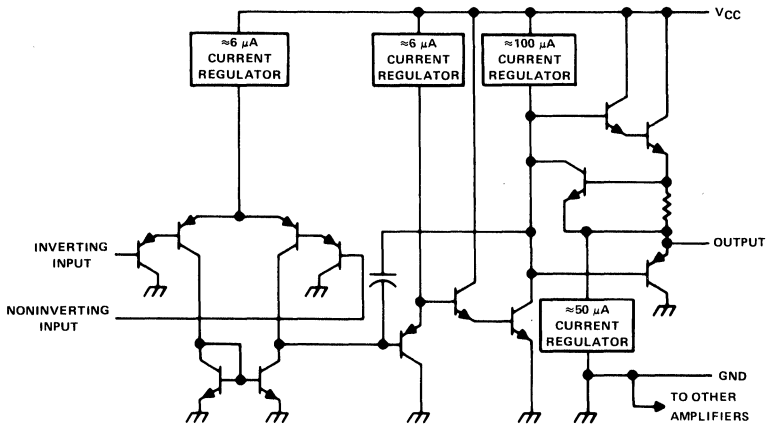
**LM124, LM224, LM224A,
LM324, LM324A, LM2902
QUADRUPLE OPERATIONAL AMPLIFIERS**

AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE				
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	FLAT PACK (W)
0°C to 70°C	7 mV 3 mV	LM324D	—	LM324J LM324AJ	LM324N LM324AN	—
-25°C to 85°C	5 mV 3 mV	LM224D	—	LM224J LM224AJ	LM224N LM224AN	—
-40°C to 105°C	7 mV	LM2902D	—	LM2902J	LM2902N	—
-55°C to 125°C	5 mV	—	LM124FK	LM124J	—	LM124W

The D package is available taped and reeled. Add the suffix R to the device type when ordering. (e.g., LM324DR)

schematic (each amplifier)



**LM124, LM224, LM224A,
LM324, LM324A, LM2902
QUADRUPLE OPERATIONAL AMPLIFIERS**

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

		LM124 LM224, LM224A, LM324, LM324A	LM2902	UNIT
Supply voltage, V_{CC} (see Note 1)		32	26	V
Differential voltage (see Note 2)		± 32	± 26	V
Input voltage range (either input)		-0.3 to 32	-0.3 to 26	V
Duration of output short-circuit (one amplifier) to ground at (or below) 25°C free-air temperature ($V_{CC} \leq 15$ V) (see Note 3)		unlimited	unlimited	
Continuous total dissipation		See Dissipation Rating Table		
Operating free-air temperature range	LM124	-55 to 125		°C
	LM224, LM224A	-25 to 85		
	LM324, LM324A	0 to 70		
	LM2902		-40 to 105	
Storage temperature range		-65 to 150	-65 to 150	°C
Case temperature for 60 seconds		FK package	260	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds		J or W package	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds		D or N package	260	°C

- NOTES: 1. All voltage values, except differential voltages and V_{CC} specified for the measurement of I_{OS} , are with respect to the network ground terminal.
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. Short circuits from outputs to V_{CC} can cause excessive heating and eventual destruction.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	900 mW	7.6 mW/°C	32°C	608 mW	494 mW	N/A
FK	900 mW	11.0 mW/°C	68°C	880 mW	715 mW	275 mW
J (LM124)	900 mW	11.0 mW/°C	68°C	880 mW	715 mW	275 mW
J (all others)	900 mW	8.2 mW/°C	40°C	656 mW	533 mW	N/A
N	900 mW	9.2 mW/°C	52°C	736 mW	598 mW	N/A
W	900 mW	8.0 mW/°C	37°C	640 mW	520 mW	200 mW

2

Operational Amplifiers

electrical characteristics at specified free-air temperature, $V_{CC} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS [†]	LM124, LM224			LM324			LM2902			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{CC} = 5\text{ V to MAX}$, $V_{IC} = V_{ICR\text{ min}}$, $V_O = 1.4\text{ V}$	25°C	3	5	3	7	3	7	mV		
		Full range		7		9		10			
I_{IO} Input offset current	$V_O = 1.4\text{ V}$	25°C	2	30	2	50	2	50	nA		
		Full range		100		150		200			
I_{IB} Input bias current	$V_O = 1.4\text{ V}$	25°C	-20	-150	-20	-250	-20	-250	nA		
		Full range		-300		-500		-500			
V_{ICR} Common-mode input voltage range	$V_{CC} = 5\text{ V to MAX}$	25°C	0 to $V_{CC}-1.5$		0 to $V_{CC}-1.5$		0 to $V_{CC}-1.5$		V		
		Full range	0 to $V_{CC}-2$		0 to $V_{CC}-2$		0 to $V_{CC}-2$				
V_{OH} High-level output voltage	$R_L = 2\text{ k}\Omega$	25°C							V		
	$R_L = 10\text{ k}\Omega$	25°C					$V_{CC}-1.5$				
	$V_{CC} = \text{MAX}$, $R_L = 2\text{ k}\Omega$	Full range	26		26		22				
	$V_{CC} = \text{MAX}$, $R_L = 10\text{ k}\Omega$	Full range	27	28	27	28	23	24			
V_{OL} Low-level output voltage	$R_L \leq 10\text{ k}\Omega$	Full range		5	20		5	100	mV		
A_{VD} Large-signal differential voltage amplification	$V_{CC} = 15\text{ V}$, $V_O = 1\text{ V to }11\text{ V}$, $R_L \geq 2\text{ k}\Omega$	25°C	50	100	25	100	100	V/mV			
		Full range	25		15		15				
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$	25°C	70	80	65	80	50	80	dB		
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IO}$)		25°C	65	100	65	100	50	100	dB		
V_{O1}/V_{O2} Crosstalk attenuation	$f = 1\text{ kHz to }20\text{ kHz}$	25°C		120		120		120	dB		
I_O Output current	$V_{CC} = 15\text{ V}$, $V_{ID} = 1\text{ V}$, $V_O = 0$	25°C	-20	-30	-60	-20	-30	-60	mA		
		Full range	-10		-10		-10				
	$V_{CC} = 15\text{ V}$, $V_{ID} = -1\text{ V}$, $V_O = 15\text{ V}$	25°C	10	20	10	20	10	20	μA		
		Full range	5		5		5				
I_{OS} Short-circuit output current	V_{CC} at 5 V, GND at -5 V, $V_O = 0$	25°C	± 40	± 60	± 40	± 60	± 40	± 60	mA		
		Full range	0.7	1.2	0.7	1.2	0.7	1.2			
I_{CC} Supply current (four amplifiers)	$V_O = 2.5\text{ V}$, No load	Full range	0.7	1.2	0.7	1.2	0.7	1.2	mA		
		$V_{CC} = \text{MAX}$, $V_O = 0.5 V_{CC}$, No load	Full range	1.1	3	1.1	3	1.1		3	

[†] All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. "MAX" V_{CC} for testing purposes is 26 V for LM2902, 30 V for the others. Full range is -55°C to 125°C for LM124, -25°C to 85°C for LM224, 0°C to 70°C for LM324, and -40°C to 105°C for LM2902.

electrical characteristics at specified free-air temperature, $V_{CC} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	LM224A			LM324A			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{CC} = 5\text{ V to }30\text{ V}$, $V_{IC} = V_{ICR\text{ min}}$, $V_O = 1.4\text{ V}$	25 °C	2	3	2	3	mV	
		Full range		4		5		
I_O Input offset current	$V_O = 1.4\text{ V}$	25 °C	2	15	2	30	nA	
		Full range		30		75		
I_{IB} Input bias current	$V_O = 1.4\text{ V}$	25 °C	-15	-80	-15	-100	nA	
		Full range		-100		-200		
V_{ICR} Common-mode input voltage range	$V_{CC} = 30\text{ V}$	25 °C	0 to $V_{CC}-1.5$		to $V_{CC}-1.5$		V	
		Full range	0 to $V_{CC}-2$		0 to $V_{CC}-2$			
V_{OH} High-level output voltage	$R_L = 2\text{ k}\Omega$	25 °C		$V_{CC}-1$		$V_{CC}-1$	V	
	$V_{CC} = 30\text{ V}$, $R_L = 2\text{ k}\Omega$	Full range	26		26			
	$V_{CC} = 30\text{ V}$, $R_L = 10\text{ k}\Omega$	Full range	27	28	27	28		
V_{OL} Low-level output voltage	$R_L \leq 10\text{ k}\Omega$	Full range	5	20	5	20	mV	
A_{VD} Large-signal differential voltage amplification	$V_{CC} = 15\text{ V}$, $V_O = 1\text{ V to }11\text{ V}$, $R_L \geq 2\text{ k}\Omega$	25 °C	50	100	25	100	V/mV	
		Full range	25		15			
CMRR Common-mode rejection ratio	$V_{IC} \geq V_{ICR\text{ min}}$	25 °C	70	80	65	80	dB	
kSVR Supply voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IO}$)		25 °C	65	100	65	100	dB	
V_{O1}/V_{O2} Crosstalk attenuation	$f = 1\text{ kHz to }20\text{ kHz}$	25 °C		120		120	dB	
I_O Output current	$V_{CC} = 15\text{ V}$, $V_{ID} = 1\text{ V}$, $V_O = 0$	25 °C	-20	-30	-60	-20	-30	mA
		Full range	-10			-10		
	$V_{CC} = 15\text{ V}$, $V_{ID} = -1\text{ V}$, $V_O = 15\text{ V}$	25 °C	10	20		10	20	
		Full range	5			5		
I_{OS} Short-circuit output current	V_{CC} at 5 V, GND at -5 V, $V_O = 0$	25 °C	±40	±60	±40	±60	mA	
		Full range	1.5	2.4	1.5	2.4		
I_{CC} Supply current (four amplifiers)	$V_O = 2.5\text{ V}$, No load	Full range	1.1	3	1.1	3	mA	
		$V_{CC} = 30\text{ V}$, $V_O = 15\text{ V}$, No load	Full range	1.1	3	1.1		3

†All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range is -25 °C to 85 °C for LM224A and 0 °C to 70 °C for LM324A.

2

Operational Amplifiers

LM148, LM248, LM348 QUADRUPLE OPERATIONAL AMPLIFIERS

D2551, OCTOBER 1979—REVISED MAY 1988

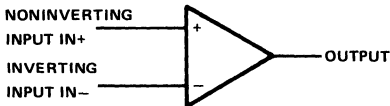
- μ A741 Operating Characteristics
- Low Supply Current Drain . . . 0.6 mA Typ (per amplifier)
- Low Input Offset Voltage
- Low Input Offset Current
- Class AB Output Stage
- Input/Output Overload Protection
- Designed to be Interchangeable with National LM148, LM248, and LM348.

description

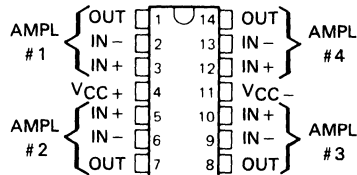
The LM148, LM248, and LM348 are quadruple, independent, high-gain, internally compensated operational amplifiers designed to have operating characteristics similar to the μ A741. These amplifiers exhibit low supply current drain, and input bias and offset currents that are much less than those of the μ A741.

The LM148 is characterized for operation over the full military temperature range of -55°C to 125°C , the LM248 is characterized for operation from -25°C to 85°C , and the LM348 is characterized for operation from 0°C to 70°C .

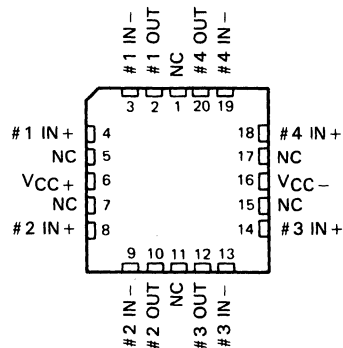
symbol (each amplifier)



LM148 . . . J PACKAGE
LM248, LM348 . . . D, J, OR N PACKAGE
(TOP VIEW)



LM148 . . . FK PACKAGE
(TOP VIEW)



NC—No internal connection

AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	6 mV	LM348D	—	LM348J	LM348N
-25°C to 85°C	6 mV	LM248D	—	LM248J	LM248N
-55°C to 125°C	5 mV	—	LM148FK	LM148J	—

The D package is available taped and reeled. Add the suffix R to the device type when ordering. (e.g., LM348DR)

2

Operational Amplifiers

LM148, LM248, LM348 QUADRUPLE OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

		LM148	LM248	LM348	UNIT
Supply voltage V_{CC+} (see Note 1)		22	18	18	V
Supply voltage V_{CC-} (see Note 1)		-22	-18	-18	V
Differential input voltage (see Note 2)		44	36	36	V
Input voltage (either input, see Notes 1 and 3)		± 22	± 18	± 18	V
Duration of output short-circuit (see Note 4)		unlimited	unlimited	unlimited	
Continuous total power dissipation		See Dissipation Rating Table			
Operating free-air temperature range		-55 to 125	-25 to 85	0 to 70	$^{\circ}\text{C}$
Storage temperature range		-65 to 150	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Case temperature for 60 seconds	FK package	260			$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J package	300	300	300	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or N package		260	260	$^{\circ}\text{C}$

NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .

2. Differential voltages are at the noninverting input terminal with respect to the inverting terminal.

3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or the value specified in the table, whichever is less.

4. The output may be shorted to ground or either power supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^{\circ}\text{C}$	$T_A = 85^{\circ}\text{C}$	$T_A = 125^{\circ}\text{C}$
	POWER RATING			POWER RATING	POWER RATING	POWER RATING
D	900 mW	7.6 mW/ $^{\circ}\text{C}$	32 $^{\circ}\text{C}$	608 mW	494 mW	N/A
FK	900 mW	11.0 mW/ $^{\circ}\text{C}$	68 $^{\circ}\text{C}$	880 mW	715 mW	275 mW
J (LM148)	900 mW	11.0 mW/ $^{\circ}\text{C}$	68 $^{\circ}\text{C}$	880 mW	715 mW	275 mW
J (LM248, LM348)	900 mW	8.2 mW/ $^{\circ}\text{C}$	40 $^{\circ}\text{C}$	656 mW	533 mW	N/A
N	900 mW	9.2 mW/ $^{\circ}\text{C}$	52 $^{\circ}\text{C}$	736 mW	598 mW	N/A

2

Operational Amplifiers

electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$

PARAMETER	TEST CONDITIONS†		LM148			LM248			LM348			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0$	25°C	1	5		1	6		1	6	mV	
		Full range		6		7.5		7.5				
I_{IO} Input offset current	$V_O = 0$	25°C	4	25		4	50		4	50	nA	
		Full range		75		125		100				
I_{IB} Input bias current	$V_O = 0$	25°C	30	100		30	200		30	200	nA	
		Full range		325		500		400				
V_{ICR} Common-mode input voltage range		Full range	±12			±12			±12		V	
V_{OM} Maximum peak output voltage swing	$R_L = 10\text{ k}\Omega$	25°C	±12	±13		±12	±13		±12	±13	V	
		Full range	±12			±12			±12			
		25°C	±10	±12		±10	±12		±10	±12		
		Full range	±10			±10			±10			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 10\text{ V}$, $R_L \geq 2\text{ k}\Omega$	25°C	50	160		25	160		25	160	V/mV	
		Full range	25			15			15			
r_i Input resistance		25°C	0.8	2.5		0.8	2.5		0.8	2.5	MΩ	
B_1 Unity-gain bandwidth	$A_{VD} = 1$	25°C	1			1			1			MHz
ϕ_M Phase margin	$A_{VD} = 1$	25°C	60°			60°			60°			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min}$, $V_O = 0$	25°C	70	90		70	90		70	90	dB	
		Full range	70			70			70			
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC\pm}/\Delta V_{IO}$)	$V_{CC\pm} = \pm 9\text{ V to } \pm 15\text{ V}$, $V_O = 0$	25°C	77	96		77	96		77	96	dB	
		Full range	77			77			77			
I_{OS} Short-circuit output current		25°C	±25			±25			±25			mA
I_{CC} Supply current (four amplifiers)	No load	$V_O = 0$	25°C			2.4 4.5			2.4 4.5			mA
		$V_O = V_{OM}^{\dagger}$										
V_{O1}/V_{O2} Crosstalk attenuation	$f = 1\text{ Hz to } 20\text{ kHz}$,	25°C	120			120			120			dB

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for T_A is -55°C to 125°C for LM148, -25°C to 85°C for LM248, and 0°C to 70°C for LM348.

**LM148, LM248, LM348
QUADRUPLER OPERATIONAL AMPLIFIERS**

operating characteristics, $V_{CC} \pm = \pm 15 \text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	$R_L = 2 \text{ k}\Omega$, $C_L = 100 \text{ pF}$, See Figure 1		0.5		$\text{V}/\mu\text{s}$

PARAMETER MEASUREMENT INFORMATION

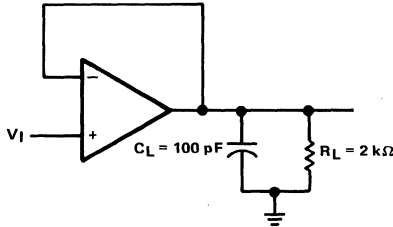


FIGURE 1. UNITY-GAIN AMPLIFIER

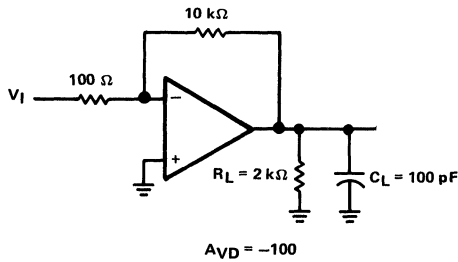


FIGURE 2. INVERTING AMPLIFIER

LM158, LM258, LM358 LM258A, LM358A, LM2904 DUAL OPERATIONAL AMPLIFIERS

D2231, JUNE 1976—REVISED AUGUST 1988

- **Wide Range of Supply Voltages:**
Single Supply . . . 3 V to 30 V
(LM2904 . . . 3 V to 26 V),
or Dual Supplies
- **Low Supply Current Drain Independent of Supply Voltage . . . 0.7 mA Typ**
- **Common-Mode Input Voltage Range**
Includes Ground Allowing Direct Sensing
near Ground
- **Low Input Bias and Offset Parameters:**
Input Offset Voltage . . . 3 mV Typ
A Versions . . . 2 mV Typ
Input Offset Current . . . 2 nA Typ
Input Bias Current . . . 20 nA Typ
A Versions . . . 15 nA Typ
- **Differential Input Voltage Range Equal to Maximum-Rated Supply Voltage . . . ± 32 V**
(± 26 V for LM2904)
- **Open-Loop Differential Voltage Amplification . . . 100 V/mV Typ**
- **Internal Frequency Compensation**

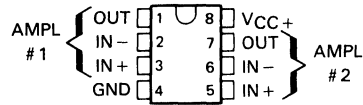
description

These devices consist of two independent, high-gain, frequency-compensated operational amplifiers that were designed specifically to operate from a single supply over a wide range of voltages. Operation from split supplies is also possible so long as the difference between the two supplies is 3 V to 30 V (3 V to 26 V for the LM2904), and the V_{CC} pin is at least 1.5 V more positive than the input common-mode voltage. The low supply current drain is independent of the magnitude of the supply voltage.

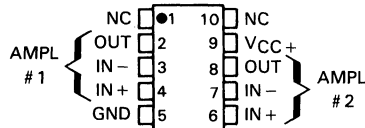
Applications include transducer amplifiers, d-c amplification blocks, and all the conventional operational amplifier circuits that now can be more easily implemented in single-supply-voltage systems. For example, these devices can be operated directly off of the standard 5-V supply that is used in digital systems and will easily provide the required interface electronics without requiring additional ± 15 -V supplies.

The LM158 is characterized for operation over the full military temperature range of -55°C to 125°C . The LM258 and LM258A are characterized for operation from -25°C to 85°C , the LM358 and LM358A from 0°C to 70°C , and the LM2904 from -40°C to 105°C .

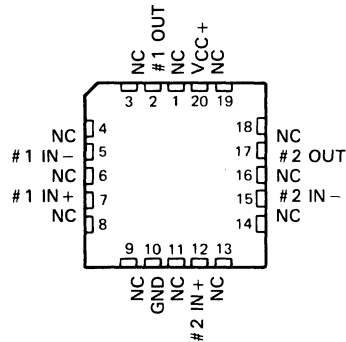
D, JG, OR P PACKAGE
(TOP VIEW)



U FLAT PACKAGE
(TOP VIEW)

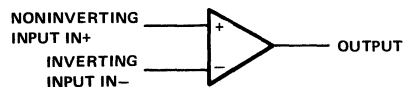


LM 158
FK CHIP CARRIER PACKAGE
(TOP VIEW)



NC—No internal connection

schematic (each amplifier)



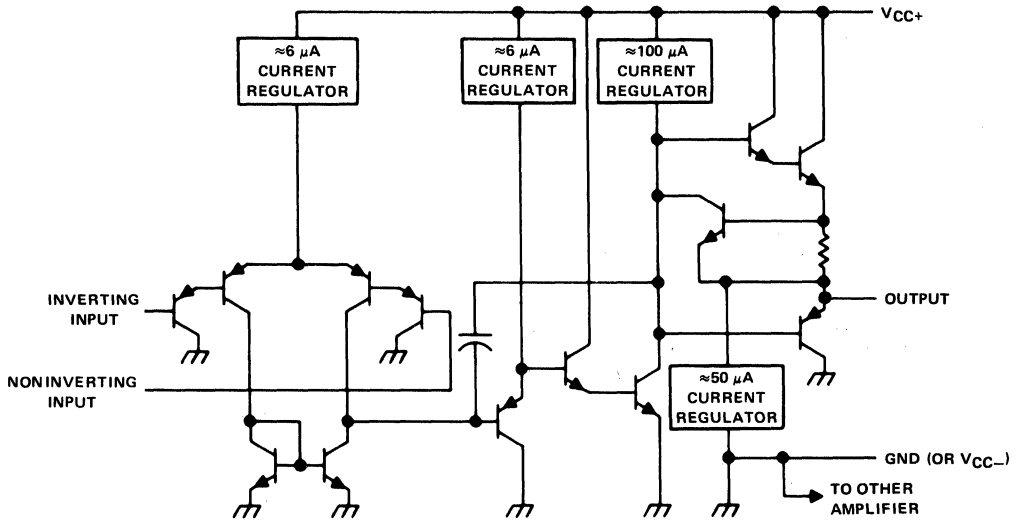
LM158, LM258, LM358, LM258A, LM358A, LM2904
DUAL OPERATIONAL AMPLIFIERS

AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE				
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)	FLAT PACK (U)
0°C to 70°C	7 mV 3 mV	LM358D LM358AD	—	LM358JG LM358AJG	LM358P LM358AP	—
-25°C to 85°C	5 mV 3 mV	LM258D LM258AD	—	LM258JG LM258AJG	LM258P LM258AP	—
-40°C to 105°C	7 mV	LM2904D	—	LM2904JG	LM2904P	—
-55°C to 125°C	5 mV	—	LM158FK	LM158JG	—	LM158U

The D package is available taped and reeled. Add the suffix R to the device type. (e.g., LM358DR)

schematic (each amplifier)



2

Operational Amplifiers

LM158, LM258, LM358, LM258A, LM358A, LM2904 DUAL OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

		LM158, LM258, LM258A LM358, LM358A	LM2904	UNIT	
Supply voltage, V_{CC} (see Note 1)		32	26	V	
Differential voltage (see Note 2)		± 32	± 26	V	
Input voltage range (either input)		0.3 to 32	0.3 to 26	V	
Duration of output short-circuit (one amplifier) to ground at (or below) 25°C free-air temperature ($V_{CC} \leq 15$ V) (see Note 3)		unlimited	unlimited		
Continuous total dissipation		See Dissipation Rating Table			
Operating free-air temperature range	LM158	-55 to 125		°C	
	LM258, LM258A	-25 to 85			
	LM358, LM358A	0 to 70			
	LM2904		-40 to 105		
Storage temperature range		-65 to 150	-65 to 150	°C	
Case temperature for 60 seconds		FK package		260	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds		JG, or U package		300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds		D or P package		260	°C

- NOTES: 1. All voltage values, except differential voltages, and V_{CC} specified for measurement of I_{OS} , are with respect to the network ground terminal.
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. Short circuits from outputs to V_{CC} can cause excessive heating and eventual destruction.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING		POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	—
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG (LM158)	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
JG (all others)	825 mW	6.6 mW/°C	528 mW	429 mW	—
P	1000 mW	8.0 mW/°C	640 mW	520 mW	—
U	675 mW	5.4 mW/°C	432 mW	351 mW	135 mW

2

Operational Amplifiers

electrical characteristics at specified free-air temperature, $V_{CC} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	LM158, LM258			LM358			LM2904			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{CC} = 5\text{ V to MAX,}$ $V_{IC} = V_{ICR\text{ min,}}$ $V_O = 1.4\text{ V}$	25°C	3	5	3	7	3	7	mV		
		Full range	7		9		10				
α_{VIO} Average temperature coefficient of input offset voltage		Full range	7			7			$\mu\text{V}/^\circ\text{C}$		
I_{IO} Input offset current	$V_O = 1.4\text{ V}$	25°C	2	30	2	50	2	50	nA		
		Full range	100		150		200				
α_{IIO} Average temperature coefficient of input offset current		Full range	10			10			$\text{pA}/^\circ\text{C}$		
I_{IB} Input bias current	$V_O = 1.4\text{ V}$	25°C	-20	-150	-20	-250	-20	-250	nA		
		Full range	-300		-500		-500				
V_{ICR} Common-mode input voltage range	$V_{CC} = 5\text{ V to MAX}$	25°C	0 to $V_{CC} - 1.5$	0 to $V_{CC} - 1.5$	0 to $V_{CC} - 1.5$	0 to $V_{CC} - 1.5$	0 to $V_{CC} - 1.5$	V			
		Full range	0 to $V_{CC} - 2$	0 to $V_{CC} - 2$	0 to $V_{CC} - 2$	0 to $V_{CC} - 2$	0 to $V_{CC} - 2$				
V_{OH} High-level output voltage	$R_L \geq 2\text{ k}\Omega$	25°C	$V_{CC} - 1.5$			$V_{CC} - 1.5$			V		
		25°C				$V_{CC} - 1.5$					
		Full range	26	26		22		22			
V_{OH} High-level output voltage	$R_L \geq 2\text{ k}\Omega$	Full range	27	28	27	28	23	24	V		
		Full range	27	28	27	28	23	24			
V_{OL} Low-level output voltage	$R_L \leq 10\text{ k}\Omega$	Full range	5	20	5	20	5	100	mV		

AVD	Large-signal differential voltage amplification	$V_{CC} = 15\text{ V},$ $V_O = 1\text{ V to }11\text{ V},$ $R_L = \geq 2\text{ k}\Omega$	25 °C	50	100	25	100	100	V/mV		
			Full range	25		15		15			
CMRR	Common-mode rejection ratio	$V_{CC} = 5\text{ V to MAX},$ $V_{IC} = V_{ICR\text{ min}}$	25 °C	70	80	65	80	50	80	dB	
kSVR	Supply voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IO}$)	$V_{CC} = 5\text{ V to MAX}$	25 °C	65	100	65	100	50	100	dB	
V_{O1}/V_{O2}	Crosstalk attenuation	$f = 1\text{ kHz to }20\text{ kHz}$	25 °C		120		120		120	dB	
I _O	Output current	$V_{CC} = 15\text{ V},$ $V_{ID} = 1\text{ V},$ $V_O = 0$	25 °C	-20	-30	-20	-30	-20	-30	mA	
			Full range	-10		-10		-10			
		$V_{CC} = 15\text{ V},$ $V_{ID} = -1\text{ V},$ $V_O = 5\text{ V}$	25 °C	10	20	10	20	10	20	μA	
			Full range	5		5		5			
I _{OS}	Short-circuit output current	V_{CC} at 5 V, GND at -5 V, $V_O = 0$	25 °C		± 40	± 60	± 40	± 60	± 40	± 60	mA
			Full range		0.7	1.2	0.7	1.2	0.7	1.2	
I _{CC}	Supply current (two amplifiers)	$V_{CC} = 2.5\text{ V},$ No load	Full range		0.7	1.2	0.7	1.2	0.7	1.2	mA
			Full range		1	2	1	2	1	2	
		$V_{CC} = \text{MAX},$ $V_O = 0.5\text{ }V_{CC},$ No load	Full range		1	2	1	2	1	2	

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. "MAX" V_{CC} for testing purposes is 26 V for LM2904, 30 V for the others. Full range is -55 °C to 125 °C for LM158, -25 °C to 85 °C for LM258, 0 °C to 70 °C for LM358, and -40 °C to 105 °C for LM2904.

electrical characteristics at specified free-air temperature, $V_{CC} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS [†]	LM258A			LM358A			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{CC} = 5\text{ V to }30\text{ V}$ $V_{IC} = V_{ICR}\text{ min,}$ $V_O = 1.4\text{ V}$	25°C	2	3	2	3	mV	
		Full range		4		5		
α_{VIO} Average temperature coefficient of input offset voltage		Full range	7	15	7	20	$\mu\text{V}/^\circ\text{C}$	
I_{IO} Input offset current	$V_O = 1.4\text{ V}$	25°C	2	15	2	30	nA	
		Full range		30		75		
α_{IIO} Average temperature coefficient of input offset current		Full range	10	200	10	300	$\text{pA}/^\circ\text{C}$	
I_{IB} Input bias current	$V_O = 1.4\text{ V}$	25°C	-15	-80	-15	-100	nA	
		Full range		-100		-200		
V_{ICR} Common-mode input voltage range	$V_{CC} 30\text{ V}$	25°C	0 to $V_{CC}-1.5$		0 to $V_{CC}-1.5$		V	
		Full range	0 to $V_{CC}-2$		0 to $V_{CC}-2$			
V_{OH} High-level output voltage	$R_L \geq 2\text{ k}\Omega$	25°C	$V_{CC}-1.5$		$V_{CC}-1.5$		V	
	$V_{CC} = 30\text{ V,}$ $R_L = 2\text{ k}\Omega$	Full range	26		26			
V_{OL} Low-level output voltage	$V_{CC} = 30\text{ V,}$ $R_L \geq 10\text{ k}\Omega$	Full range	27	28	27	28	mV	
	$R_L \leq 10\text{ k}\Omega$	Full range		5 20		5 20		

A _{VD}	Large-signal differential voltage amplification	V _{CC} = 15 V, V _O = 1 V to 11 V, R _L = ≅ 2 kΩ	25°C	50	100	25	100	V/mV		
			Full range	25		15				
CMRR	Common-mode rejection ratio		25°C	70	80	65	80	dB		
k _{SVR}	Supply voltage rejection ratio (ΔV _{CC} /ΔV _{IO})		25°C	65	100	65	100	dB		
V _{O1} /V _{O2}	Crosstalk attenuation	f = 1 kHz to 20 kHz	25°C	120		120		dB		
I _O	Output current	V _{CC} = 15 V, V _{ID} = 1 V, V _O = 0	25°C	20	30	60	20	30	60	mA
			Full range	10			10			
		V _{CC} = 15 V, V _{ID} = -1 V, V _O = 15 V	25°C	10	20		10	20		μA
			Full range	5			5			
I _{OS}	Short-circuit output current	V _{CC} at 5 V, GND at -5 V, V _O = 0	25°C		±40	±60		±40	±60	mA
I _{CC}	Supply current (two amplifiers)	V _O = 2.5 V, No load	Full range		0.7	1.2		0.7	1.2	mA
		V _{CC} = 30 V, V _O = 15 V, No load	Full range		1	2			2	

†All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range is -25°C to 85°C for LM258A and 0°C to 70°C to LM358A.

2

Operational Amplifiers

LM218, LM318 HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

D2219, JUNE 1976—REVISED OCTOBER 1988

- Small-Signal Bandwidth . . . 15 MHz Typ
- Slew Rate . . . 50 V/ μ s Min
- Bias Current . . . 250 nA Max (LM218)
- Supply Voltage Range . . . ± 5 V to ± 20 V
- Internal Frequency Compensation
- Input and Output Overload Protection
- Same Pin Assignments as General-Purpose Operational Amplifiers

description

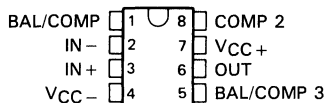
The LM218 and LM318 are precision, high-speed operational amplifiers designed for applications requiring wide bandwidth and high slew rate. They feature a factor-of-ten increase in speed over general purpose devices without sacrificing dc performance.

These operational amplifiers have internal unity-gain frequency compensation. This considerably simplifies their application since no external components are necessary for operation. However, unlike most internally compensated amplifiers, external frequency compensation may be added for optimum performance. For inverting applications, feed-forward compensation will boost the slew rate to over 150 V/ μ s and almost double the bandwidth. Over compensation may be used with the amplifier for greater stability when maximum bandwidth is not needed. Further, a single capacitor may be added to reduce the settling time for 0.1% error band to under 1 μ s.

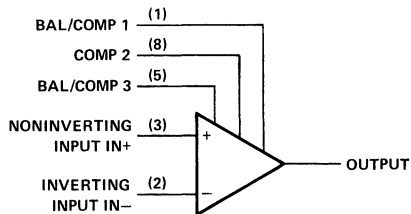
The high speed and fast settling time of these operational amplifiers make them useful in A/D converters, oscillators, active filters, sample and hold circuits, and general purpose amplifiers.

The LM218 is characterized for operation from -25°C to 85°C , and the LM318 is characterized for operation from 0°C to 70°C .

D, JG, OR P PACKAGE (TOP VIEW)



symbol



AVAILABLE OPTIONS

T _A	V _{IO} MAX at 25°C	PACKAGE		
		SMALL- OUTLINE (D)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	10 mV	LM318D	LM318JG	LM318P
-25°C to 85°C	4 mV	LM218D	LM218JG	LM218P

The D packages are available taped and reeled. Add the suffix R to the device type (e.g., LM318DR).

2

Operational Amplifiers

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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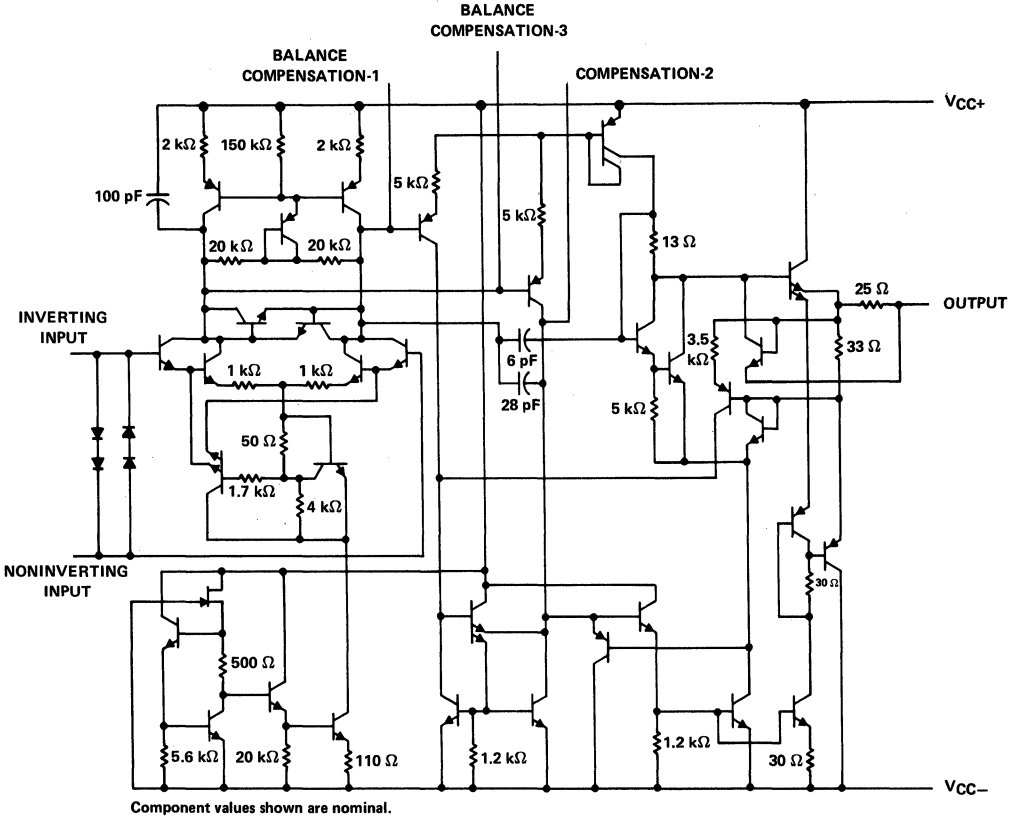
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LM218, LM318 HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

schematic

2

Operational Amplifiers



LM218, LM318 HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	LM218	LM318	UNIT
Supply voltage, V_{CC+} (see Note 1)	20	20	V
Supply voltage, V_{CC-} (see Note 1)	-20	-20	V
Input voltage (either input, see Notes 1 and 2)	± 15	± 15	V
Differential input current (see Note 3)	± 10	± 10	mA
Duration of output short-circuit (see Note 4)	unlimited	unlimited	
Continuous total power dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-25 to 85	0 to 70	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	JG package	300	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package	260	$^{\circ}\text{C}$

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
2. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
3. The inputs are shunted with two opposite-facing base-emitter diodes for over voltage protection. Therefore, excessive current will flow if a differential input voltage in excess of approximately 1 V is applied between the inputs unless some limiting resistance is used.
4. The output may be shorted to ground or either power supply. For the LM218 only, the unlimited duration of the short-circuit applies at (or below) 85 $^{\circ}\text{C}$ case temperature or 75 $^{\circ}\text{C}$ free-air temperature.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^{\circ}\text{C}$	$T_A = 85^{\circ}\text{C}$
	POWER RATING			POWER RATING	POWER RATING
D	500 mW	5.8 mW/ $^{\circ}\text{C}$	64 $^{\circ}\text{C}$	464 mW	377 mW
JG	500 mW	6.6 mW/ $^{\circ}\text{C}$	74 $^{\circ}\text{C}$	500 mW	429 mW
P	500 mW	N/A	N/A	500 mW	500 mW

2

Operational Amplifiers

LM218, LM318 HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature (see Note 5)

PARAMETER	TEST CONDITIONS†	LM218			LM318			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0$	25°C	2	4	4	10	mV	
		Full range	6			15		
I_{IO} Input offset current	$V_O = 0$	25°C	6	50	30	200	nA	
		Full range	100			300		
I_{IB} Input bias current	$V_O = 0$	25°C	120	250	150	500	nA	
		Full range	500			750		
V_{ICR} Common-mode input voltage range	$V_{CC\pm} = \pm 15\text{ V}$	Full range	±11.5		±11.5		V	
V_{OM} Maximum peak output voltage swing	$V_{CC\pm} = \pm 15\text{ V}$, $R_L = 2\text{ k}\Omega$	Full range	±12	±13	±12	±13	V	
A_{VD} Large-signal differential voltage amplification	$V_{CC\pm} = \pm 15\text{ V}$, $V_O = \pm 10\text{ V}$, $R_L \geq 2\text{ k}\Omega$	25°C	50	200	25	200	V/mV	
		Full range	25		20			
B_1 Unity-gain bandwidth	$V_{CC\pm} = \pm 15\text{ V}$	25°C	15		15		MHz	
r_i Input resistance		25°C	1	3	0.5	3	MΩ	
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min}$	Full range	80	100	70	100	dB	
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IO}$)		Full range	70	80	65	80	dB	
I_{CC} Supply current	No load, $V_O = 0$	25°C	5	8	5	10	mA	

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for LM218 is -25°C to 85°C and for LM318 is 0°C to 70°C.

Note 5: Unless otherwise noted, $V_{CC} = \pm 5\text{ V}$ to $\pm 20\text{ V}$. All typical values are at $V_{CC\pm} = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$.

operating characteristics, $V_{CC+} = 15\text{ V}$, $V_{CC-} = 15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	$\Delta V_I = 10\text{ V}$, $C_L = 10\text{ pF}$, See Figure 1	50	70		V/μs

PARAMETER MEASUREMENT INFORMATION

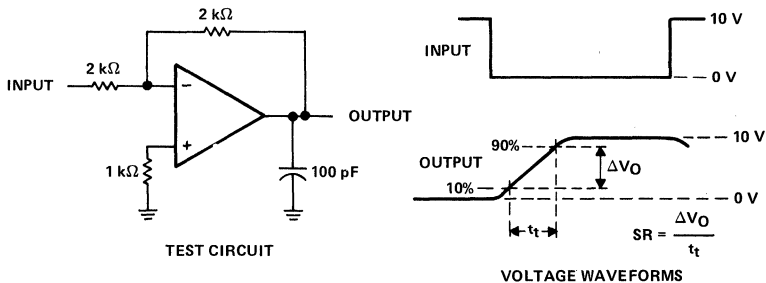


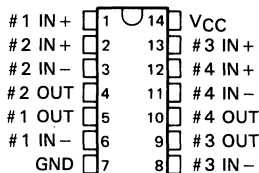
FIGURE 1. SLEW RATE

LM2900, LM3900 QUADRUPLE OPERATIONAL AMPLIFIERS

D2531, JULY 1979—REVISED AUGUST 1988

- Wide Range of Supply Voltages, Single or Dual Supplies
- Wide Bandwidth
- Large Output Voltage Swing
- Output Short-Circuit Protection
- Internal Frequency Compensation
- Low Input Bias Current
- Designed to be Interchangeable with National Semiconductor LM2900 and LM3900, Respectively

J OR N DUAL-IN-LINE PACKAGE
(TOP VIEW)



AVAILABLE OPTIONS

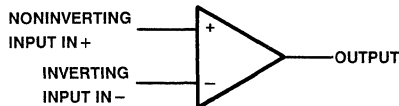
T _A	PACKAGE	
	PLASTIC DIP (N)	CERAMIC DIP (J)
0°C to 70°C	LM3900N	LM3900J
-40°C to 85°C	LM2900N	LM2900J

description

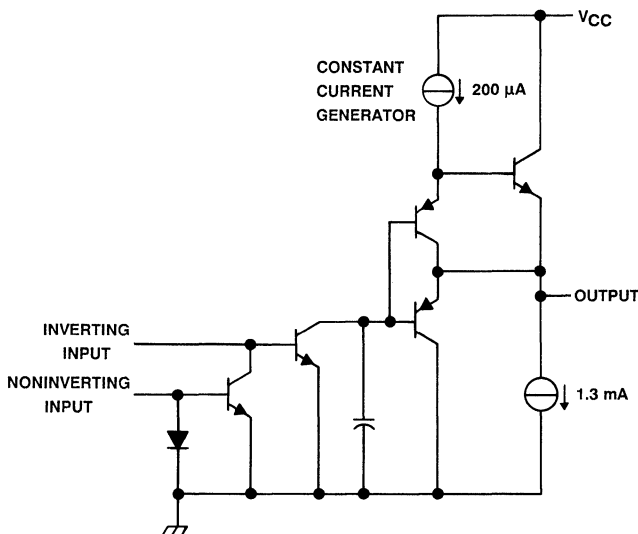
These devices consist of four independent, high-gain frequency-compensated Norton operational amplifiers that were designed specifically to operate from a single supply over a wide range of voltages. Operation from split supplies is also possible. The low supply current drain is essentially independent of the magnitude of the supply voltage. These devices provide wide bandwidth and large output voltage swing.

The LM2900 is characterized for operation from -40°C to 85°C, and the LM3900 is characterized for operation from 0°C to 70°C.

symbol (each amplifier)



schematic (each amplifier)



PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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LM2900, LM3900 QUADRUPLE OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	LM2900	LM3900	UNIT
Supply voltage, V_{CC} (see Note 1)	36	36	V
Input current	20	20	mA
Duration of output short circuit (one amplifier) to ground at (or below) 25°C free-air temperature (see Note 2)	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-40 to 85	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J Package	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	N Package	260	°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the network ground terminal.
2. Short circuits from outputs to V_{CC} can cause excessive heating and eventual destruction.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
J	1025 mW	8.2 mW/°C	656 mW	533 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW

recommended operating conditions

	LM2900		LM3900		UNIT
	MIN	MAX	MIN	MAX	
Input current (see Note 3)		-1		-1	mA
Operating free-air temperature, T_A	-40	85	0	70	°C

- NOTE 3: Clamp transistors are included that prevent the input voltages from swinging below ground more than approximately -0.3 V. The negative input currents that may result from large signal overdrive with capacitive input coupling must be limited externally to values of approximately -1 mA. Negative input currents in excess of -4 mA will cause the output voltage to drop to a low voltage. These values apply for any one of the input terminals. If more than one of the input terminals are simultaneously driven negative, maximum currents are reduced. Common-mode current biasing can be used to prevent negative input voltages.

LM2900, LM3900 QUADRUPLE OPERATIONAL AMPLIFIERS

electrical characteristics, $V_{CC} = 15\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		LM2900			LM3900			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
I_B	Input bias current (inverting input)	$I_{I+} = 0$	$T_A = 25^\circ\text{C}$	30	200		30	200	nA	
			$T_A = \text{Full range}$	300		300				
$\frac{I_{I-}}{I_{I+}}$	Mirror gain	$I_{I+} = 20\ \mu\text{A}$ to $200\ \mu\text{A}$, $T_A = \text{Full range}$, See Note 4		0.9	1.1	0.9	1.1	$\mu\text{A}/\mu\text{A}$		
	Change in mirror gain			2	5	2	15	%		
	Mirror current	$V_{I+} = V_{I-}$, See Note 4	$T_A = \text{Full range}$,	10	500	10	500	μA		
A_{VD}	Large-signal differential voltage amplification	$V_O = 10\text{ V}$, $R_L = 10\text{ k}\Omega$, $f = 100\text{ Hz}$		1.2	2.8	1.2	2.8	V/mV		
r_i	Input resistance (inverting input)			1		1		$\text{M}\Omega$		
r_o	Output resistance			8		8		$\text{k}\Omega$		
B_1	Unity-gain bandwidth (inverting input)			2.5		2.5		MHz		
k_{SVR}	Supply voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IO}$)			70		70		dB		
V_{OH}	High-level output voltage	$I_{I+} = 0$, $I_{I-} = 0$	$R_L = 2\text{ k}\Omega$	13.5		13.5		V		
			$V_{CC} = 30\text{ V}$, No load	29.5		29.5				
V_{OL}	Low-level output voltage	$I_{I+} = 0$, $R_L = 2\text{ k}\Omega$	$I_{I-} = 10\ \mu\text{A}$,	0.09	0.2	0.09	0.2	V		
I_{OHS}	Short-circuit output current (output internally high)	$I_{I+} = 0$, $V_O = 0$	$I_{I-} = 0$,	-6	-18	-6	-10	mA		
	Pull-down current			0.5	1.3	0.5	1.3	mA		
I_{OL}	Low-level output current‡	$I_{I-} = 5\ \mu\text{A}$, $V_{OL} = 1\text{ V}$		5		5		mA		
I_{CC}	Supply current (four amplifiers)	No load		6.2	10	6.2	10	mA		

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified. Full range for T_A is -40°C to 85°C for LM2900, and 0°C to 70°C for LM3900.

‡ The output current-sink capability can be increased for large-signal conditions by overdriving the inverting input.

NOTE 4: These parameters are measured with the output balanced midway between V_{CC} and ground.

operating characteristics, $V_{CC\pm} = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS			MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	Low-to-high output	$V_O = 10\text{ V}$, $R_L = 2\text{ k}\Omega$	$C_L = 100\text{ pF}$,		0.5		$\text{V}/\mu\text{s}$
		High-to-low output				20		

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Operational Amplifiers

LM2900, LM3900 QUADRUPLE OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

INPUT BIAS CURRENT (INVERTING INPUT)
vs
FREE-AIR TEMPERATURE

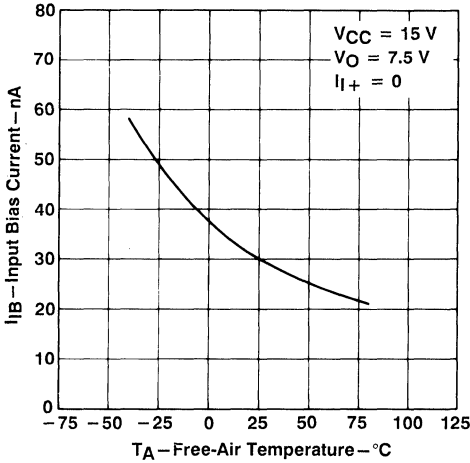


FIGURE 1

MIRROR GAIN
vs
FREE-AIR TEMPERATURE

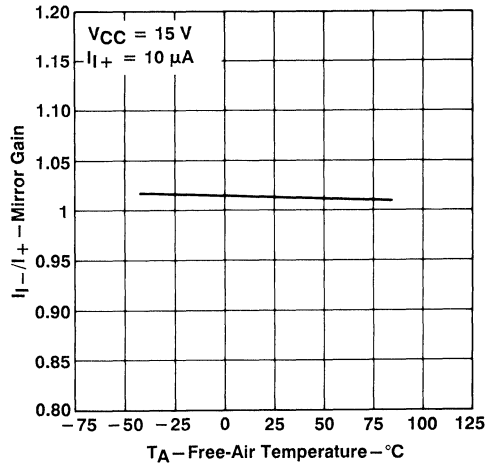


FIGURE 2

LARGE SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREQUENCY

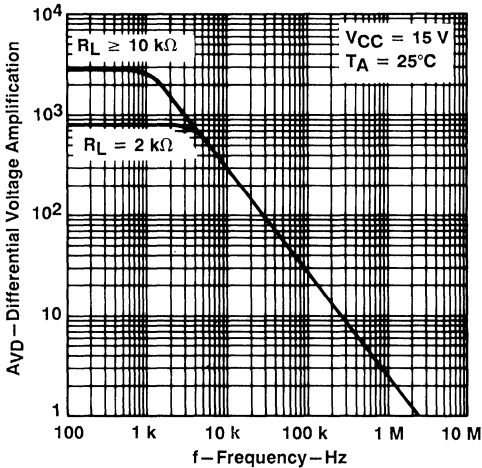


FIGURE 3

LARGE SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
SUPPLY VOLTAGE

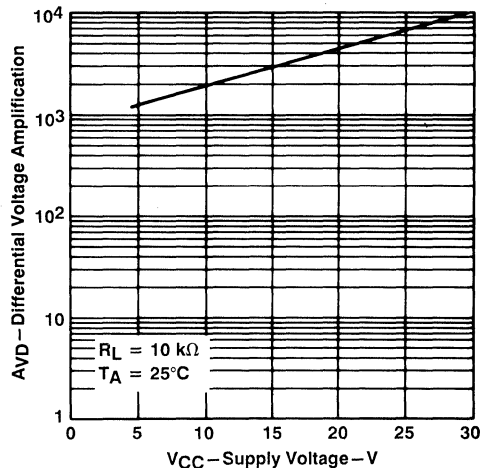


FIGURE 4

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

LM2900, LM3900 QUADRUPLE OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

LARGE SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREE-AIR TEMPERATURE

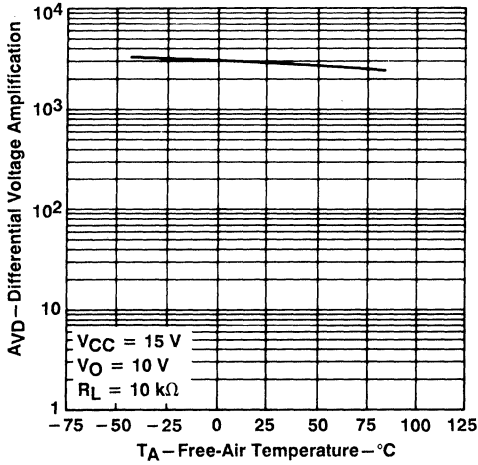


FIGURE 5

SUPPLY VOLTAGE REJECTION RATIO
vs
FREQUENCY

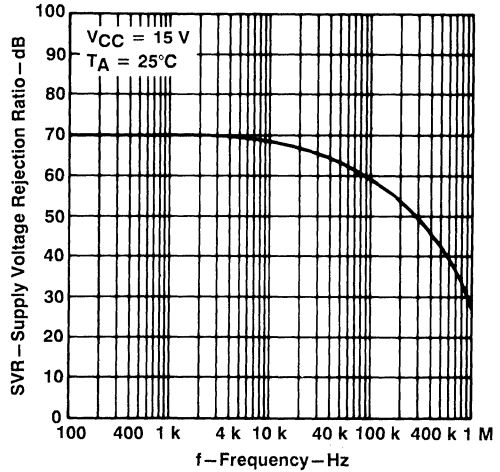


FIGURE 6

PEAK-TO-PEAK OUTPUT VOLTAGE
vs
FREQUENCY

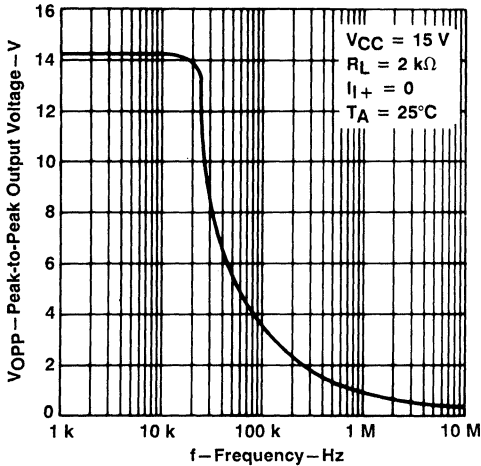


FIGURE 7

LM2900
SHORT-CIRCUIT OUTPUT CURRENT
(OUTPUT INTERNALLY HIGH)
vs
SUPPLY VOLTAGE

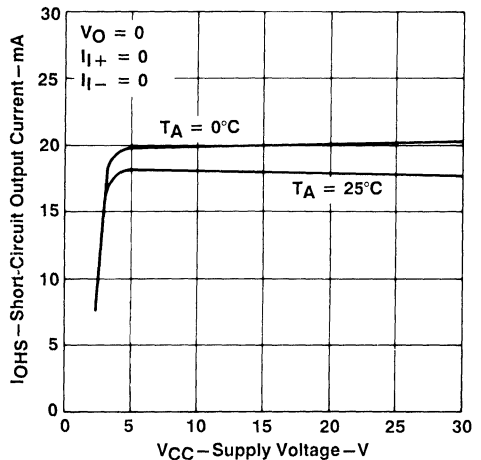


FIGURE 8

2

Operational Amplifiers

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

LM2900, LM3900 QUADRUPLE OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT CURRENT
vs
SUPPLY VOLTAGE

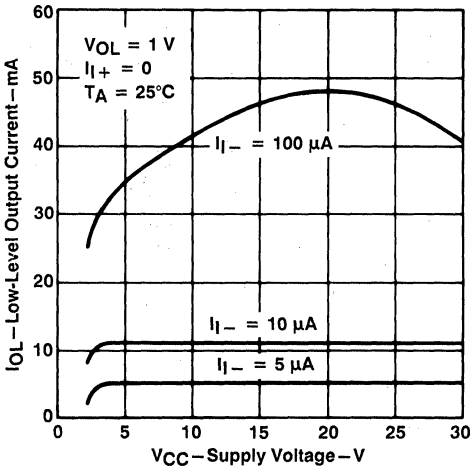


FIGURE 9

PULL-DOWN CURRENT
vs
SUPPLY VOLTAGE

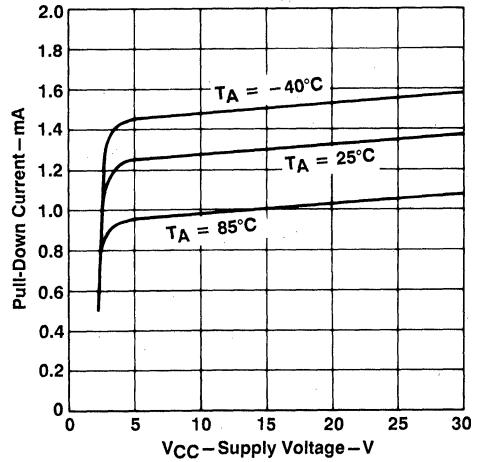


FIGURE 10

PULL-DOWN CURRENT
vs
FREE-AIR TEMPERATURE

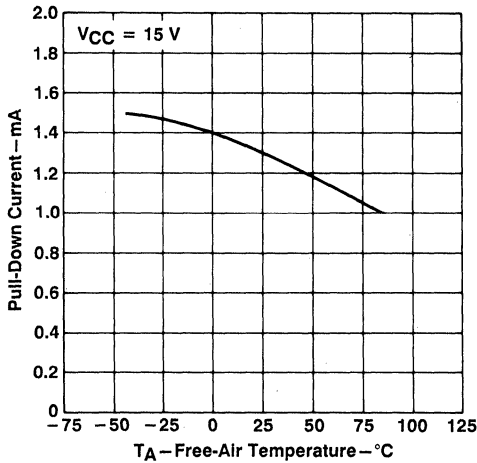


FIGURE 11

TOTAL SUPPLY CURRENT
vs
SUPPLY VOLTAGE

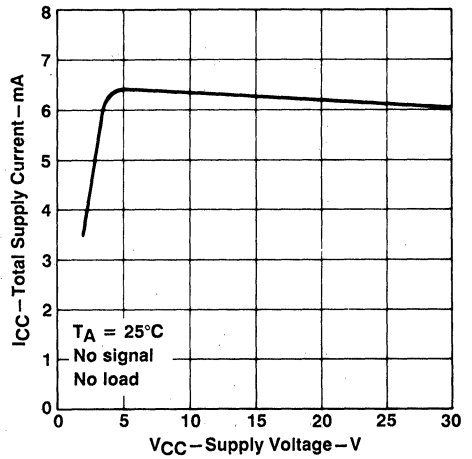


FIGURE 12

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL APPLICATION DATA

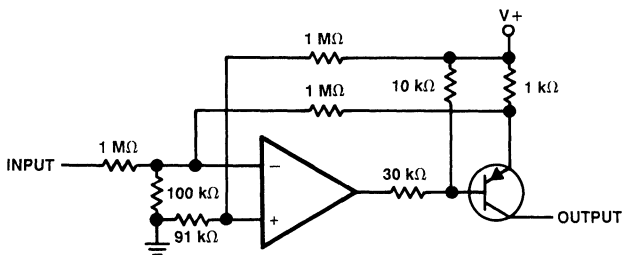
Norton (or current-differencing) amplifiers can be used in most standard general-purpose op-amp applications. Performance as a dc amplifier in a single-power-supply mode is not as precise as a standard integrated-circuit operational amplifier operating from dual supplies. Operation of the amplifier can be best be understood by noting that input currents are differenced at the inverting input terminal and this current then flows through the external feedback resistor to produce the output voltage. Common-mode current biasing is generally useful to allow operating with signal levels near (or even below) ground.

Internal transistors clamp negative input voltages at approximately -0.3 V but the magnitude of current flow has to be limited by the external input network. For operation at high temperature, this limit should be approximately $-100\ \mu\text{A}$.

Noise immunity of a Norton amplifier is less than that of standard bipolar amplifiers. Circuit layout is more critical since coupling from the output to the noninverting input can cause oscillations. Care must also be exercised when driving either input from a low-impedance source. A limiting resistor should be placed in series with the input lead to limit the peak input current. Current up to $20\ \text{mA}$ will not damage the device but the current mirror on the noninverting input will saturate and cause a loss of mirror gain at higher current levels, especially at high operating temperatures.

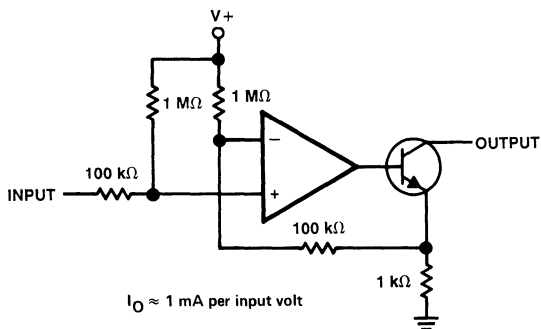
2

Operational Amplifiers



$I_O \approx 1\ \text{mA per input volt}$

FIGURE 13. VOLTAGE-CONTROLLED CURRENT SOURCE



$I_O \approx 1\ \text{mA per input volt}$

FIGURE 14. VOLTAGE-CONTROLLED CURRENT SINK

2

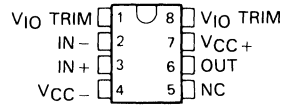
Operational Amplifiers

LT1001 PRECISION OPERATIONAL AMPLIFIER

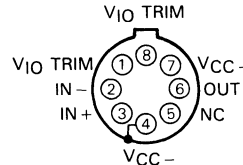
D3192, JANUARY 1989

- **Low Input Offset Voltage:**
 LT1001AM . . . 15 μV Max
 LT1001AC . . . 25 μV Max
 LT1001M, LT1001C . . . 60 μV Max
- **Low Offset Voltage Temperature Coefficient:**
 LT1001AM, LT1001AC . . . 0.6 $\mu\text{V}/^\circ\text{C}$ Max
 LT1001M, LT1001C . . . 1 $\mu\text{V}/^\circ\text{C}$ Max
- **Low Input Bias Current:**
 LT1001AM, LT1001AC . . . ± 2 nA Max
 LT1001M, LT1001C . . . ± 4 nA Max
- **Low Common-Mode Rejection Ratio:**
 LT1001AM, LT1001AC . . . 114 dB Min
 LT1001M, LT1001C . . . 110 dB Min
- **Low Supply Voltage Rejection Ratio:**
 LT1001AM, LT1001AC . . . 110 dB Min
 LT1001M, LT1001C . . . 106 dB Min
- **Low Power Dissipation:**
 LT1001AM, LT1001AC . . . 75 mW Max
 LT1001M, LT1001C . . . 80 mW Max
- **Low Peak-to-Peak Equivalent Input Noise Voltage . . . 0.3 μV Typ**

D, JG, OR P PACKAGE
(TOP VIEW)



L PACKAGE
(TOP VIEW)



NC—No internal connection
 Pin 4 (L package) is in electrical contact with the case.

2
Operational Amplifiers

description

The LT1001 is a precision operational amplifier suited for applications such as thermocouple amplifiers, strain gauge amplifiers, low-level signal processing, and high-accuracy data acquisition. In the design, processing, and testing of the device, particular attention has been paid to optimizing the entire distribution of several key parameters. The input offset voltage of all units is less than 60 μV , and the LT1001AM is specified at 15 μV maximum. Power dissipation is nearly halved compared to the most popular precision operational amplifiers without adversely affecting noise or speed performance. The output drive capability of the LT1001 is enhanced with voltage gain at a load current of 10 mA.

The specifications of the low-cost commercial-temperature device, the LT1001C, have been significantly improved when compared to equivalent grades of similar precision amplifiers. The input bias current, input offset current, and common-mode and supply voltage rejection ratios of the LT1001C offer performance previously attainable only with high-cost, selected grades of other devices.

The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The C-suffix devices are characterized for operation from 0°C to 70°C .

AVAILABLE OPTIONS

T _A	V _{IO} MAX	PACKAGE			
		SMALL OUTLINE (D)	CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)
0°C	60 μV	LT1001CD	LT1001CJG	LT1001CL	LT1001CP
to 70°C	25 μV		LT1001ACJG	LT1001ACL	LT1001ACP
-55°C	60 μV		LT1001MJG	LT1001ML	
to 125°C	15 μV		LT1001AMJG	LT1001AML	

The D package is available in tape and reel. Add the suffix R to the device type (e.g., LT1001CDR).

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

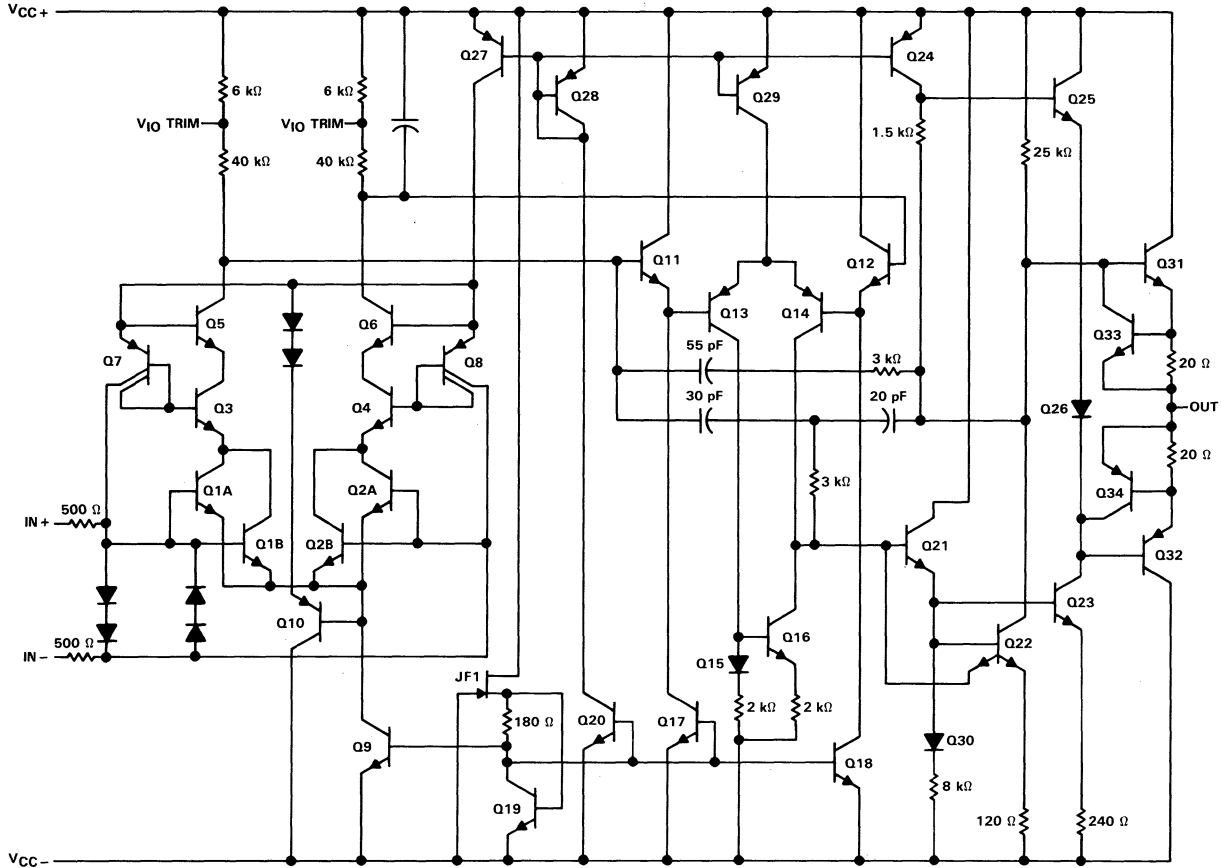


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LT1001
PRECISION OPERATIONAL AMPLIFIER

Schematic



Component values shown are nominal.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	22 V
Supply voltage, V_{CC-}	-22 V
Differential input voltage (see Note 2)	± 30 V
Input voltage range, V_I	± 22 V
Duration of short-circuit current at (or below) 25°C	unlimited
Continuous power dissipation	See Dissipation Rating Table
Operating free-air temperature range: M-suffix	-55°C to 125°C
C-suffix	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG or L package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C

NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
2. Differential voltages are at the noninverting input with respect to the inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	145 mW
JG (M-suffix)	1050 mW	8.4 mW/°C	672 mW	210 mW
JG (C-suffix)	825 mW	6.6 mW/°C	528 mW	N/A
L (M-suffix)	825 mW	6.6 mW/°C	528 mW	165 mW
L (C-suffix)	650 mW	5.2 mW/°C	416 mW	N/A
P	1000 mW	8.0 mW/°C	640 mW	200 mW

recommended operating conditions

	M-SUFFIX			C-SUFFIX			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, V_{CC+}	4	15	22	4	15	22	V
Supply voltage, V_{CC-}	-4	-15	-22	-4	-15	-22	V
Common-mode input voltage, V_{IC}	$V_{CC\pm} = \pm 15$ V			± 13			V
Operating free-air temperature, T_A	-55			125			70 °C

LT1001M, LT1001AM PRECISION OPERATIONAL AMPLIFIERS

electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T _A	LT1001M			LT1001AM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	See Note 3	25°C	18	60	7	15	μV	
			-55°C to 125°C	160		60			
α _{VIO}	Average temperature coefficient of input offset voltage	-55°C to 125°C	0.3	1	0.2	0.6	μV/°C		
	Long-term drift of input offset voltage	See Note 4	0.3	1.5	0.2	1	μV/mo		
I _{IO}	Input offset current	V _O = 0, V _{IC} = 0	25°C	0.4	3.8	0.3	2	nA	
			-55°C to 125°C	7.6		4			
I _{IB}	Input bias current	V _O = 0, V _{IC} = 0	25°C	±0.7	±4	±0.5	±2	nA	
			-55°C to 125°C	±8		±4			
V _{OH}	Maximum peak output voltage swing	R _L ≥ 2 kΩ	25°C	±13	±14	±13	±14	V	
		R _L ≥ 1 kΩ		±12	±13.5	±12	±13.5		
		R _L ≥ 2 kΩ	-55°C to 125°C	±12		±12.5			
A _{VD}	Large-signal differential voltage amplification	R _L ≥ 2 kΩ, V _O = ±12 V	25°C	400	800	450	800	V/mV	
		R _L ≥ 1 kΩ, V _O = ±10 V		250	500	300	500		
		R _L ≥ 2 kΩ, V _O = ±10 V	-55°C to 125°C	200		300			
r _{id}	Differential input resistance	25°C	15	80	30	100	MΩ		
CMRR	Common-mode rejection ratio	V _{IC} = ±13 V	25°C	110	126	114	126	dB	
			-55°C to 125°C	106		110			
K _{SVR}	Supply voltage rejection ratio	V _{CC±} = ±3 V to ±18 V	25°C	106	123	110	123	dB	
			-55°C to 125°C	100		104			
P _D	Total power dissipation	No load	25°C	48	80	46	75	mW	
		No load, V _{CC±} = ±3 V		4	8	4	6		
		No load	-55°C to 125°C	100		90			

NOTES: 3. The input offset voltage for all devices is measured with high-speed test equipment approximately 1 second after power is applied. The LT1001AM receives a 168-hour burn-in at 125°C or equivalent.

4. Long-term drift of input offset voltage refers to the average trend line of offset voltage versus time over extended periods after the first 30 days of operation. Excluding the initial hour of operation, the change in V_{IO} during the first 30 days is typically 2.5 μV.

operating characteristics, $V_{CC\pm} = \pm 15\text{ V}$, T_A = 25°C (unless otherwise noted)

PARAMETER	TEST CONDITIONS	LT1001M			LT1001AM			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain	R _L ≥ 2 kΩ	0.1	0.25	0.1	0.25	V/μs	
φ _m	Phase margin at unity gain	A _V = 40 dB, T _A = 25°C	60°		63°			
		R _S = 100 Ω, T _A = MIN	63°		63°			
		C _L = 10 pF, T _A = MAX	57°		57°			
V _{NPP}	Peak-to-peak equivalent input noise voltage	f = 0.1 Hz to 10 Hz	0.3	0.6	0.3	0.6	μV	
V _n	Equivalent input noise voltage	f = 10 Hz	10.5	18	10.3	18	nV/√Hz	
		f = 1 MHz	9.8	11	9.6	11		
GBW	Gain bandwidth product		0.4	0.8	0.4	0.8	MHz	

LT1001C, LT1001AC PRECISION OPERATIONAL AMPLIFIERS

2

Operational Amplifiers

electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A	LT1001C			LT1001AC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	See Note 5	25°C	18		60	10		25	μV
		0°C to 70°C			110			60	
α_{VIO} Average temperature coefficient of input offset voltage		0°C to 70°C	0.3	1		0.2	0.6		$\mu\text{V}/^\circ\text{C}$
Long-term drift of input offset voltage	See Note 4		0.3	1.5		0.2	1		$\mu\text{V}/\text{mo}$
I_{IO} Input offset current	$V_O = 0, V_{IC} = 0$	25°C	0.4	3.8		0.3	2		nA
		0°C to 70°C			5.3			3.5	
I_{IB} Input bias current	$V_O = 0, V_{IC} = 0$	25°C	± 0.7	± 4		± 0.5	± 2		nA
		0°C to 70°C			± 5.5			± 3.5	
V_{OH} Maximum peak output voltage swing	$R_L \geq 2\text{ k}\Omega$	25°C	± 13	± 14		± 13	± 14		V
	$R_L \geq 1\text{ k}\Omega$		± 12	± 13.5		± 12	± 13.5		
	$R_L \geq 2\text{ k}\Omega$	0°C to 70°C	± 12.5		± 12.5				
A_{VD} Large-signal differential voltage amplification	$R_L \geq 2\text{ k}\Omega, V_O = \pm 12\text{ V}$	25°C	400	800		450	800		V/mV
	$R_L \geq 1\text{ k}\Omega, V_O = \pm 10\text{ V}$		250	500		300	500		
	$R_L \geq 2\text{ k}\Omega, V_O = \pm 10\text{ V}$	0°C to 70°C	250		350				
r_{id} Differential input resistance		25°C	15	80		30	100		M Ω
CMRR Common-mode rejection ratio	$V_{IC} = \pm 13\text{ V}$	25°C	110	126		114	126		dB
		0°C to 70°C	106		110				
k_{SVR} Supply voltage rejection ratio	$V_{CC\pm} = \pm 3\text{ V to } \pm 18\text{ V}$	25°C	106	123		110	123		dB
		0°C to 70°C	103		106				
P_D Total power dissipation	No load	25°C	48		80	46		75	mW
	No load, $V_{CC\pm} = \pm 3\text{ V}$		4		8	4		6	
	No load	0°C to 70°C			90			85	

- Notes: 4. Long-term drift of input offset voltage refers to the average trend line of offset voltage versus time over extended periods after the first 30 days of operation. Excluding the initial hour of operation, the change in V_{IO} during the first 30 days is typically 2.5 μV .
 5. The input offset voltage for all devices is measured with high-speed test equipment approximately 1 second after power is applied.

operating characteristics, $V_{CC\pm} = \pm 15\text{ V}, T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	LT1001C			LT1001AC			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L \geq 2\text{ k}\Omega$	0.1	0.25		0.1	0.25		V/ μs
ϕ_m Phase margin at unity gain	$A_V = 40\text{ dB}, R_S = 100\ \Omega, C_L = 10\text{ pF}$	$T_A = 25^\circ\text{C}$	60°		63°			
		$T_A = \text{MIN}$	63°		63°			
		$T_A = \text{MAX}$	57°		57°			
V_{NPP} Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ Hz to } 10\text{ Hz}$		0.3	0.6	0.3	0.6		μV
V_n Equivalent input noise voltage	$f = 10\text{ Hz}$		10.5	18	10.3	18		nV/ $\sqrt{\text{Hz}}$
	$f = 1\text{ MHz}$		9.8	11	9.6	11		
GBW Gain bandwidth product		0.4	0.8		0.4	0.8		MHz

**LT1001M, LT1001AM, LT1001C, LT1001AC
PRECISION OPERATIONAL AMPLIFIER**

TYPICAL CHARACTERISTICS†

**INPUT OFFSET VOLTAGE
OF REPRESENTATIVE UNITS
VS
FREE-AIR TEMPERATURE**

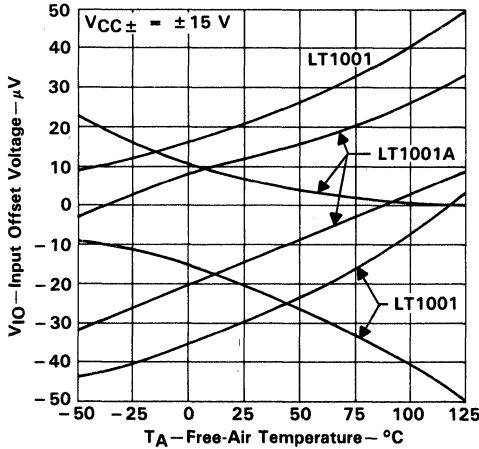


FIGURE 1

**WARM-UP CHANGE
IN INPUT OFFSET VOLTAGE
VS
ELAPSED TIME**

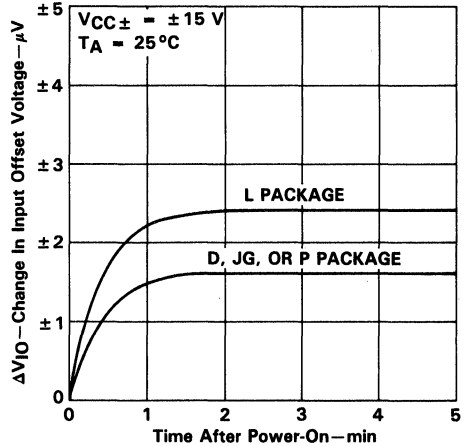


FIGURE 2

**LONG-TERM DRIFT OF
INPUT OFFSET VOLTAGE
OF REPRESENTATIVE UNITS**

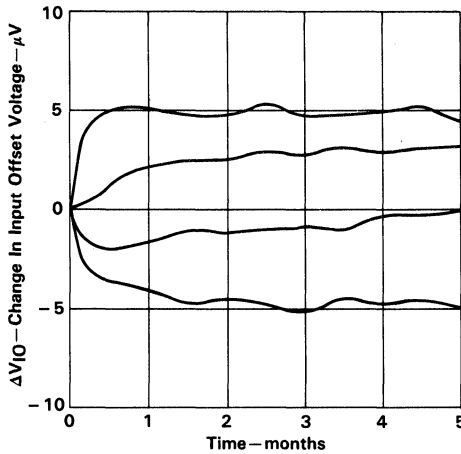


FIGURE 3

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

2
Operational Amplifiers

TYPICAL CHARACTERISTICS†

INPUT OFFSET CURRENT
and INPUT BIAS CURRENT
vs
FREE-AIR TEMPERATURE

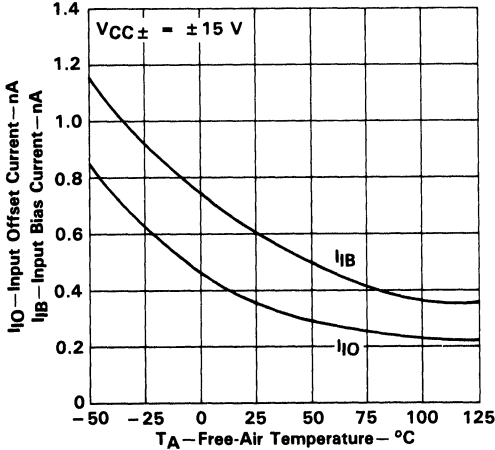


FIGURE 4

INPUT BIAS CURRENT
vs
DIFFERENTIAL INPUT VOLTAGE

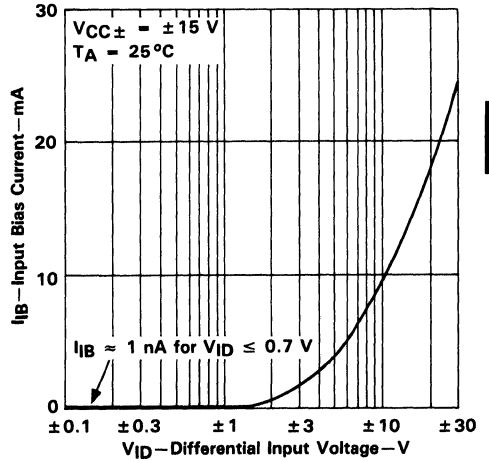


FIGURE 5

INPUT BIAS CURRENT
vs
COMMON-MODE INPUT VOLTAGE

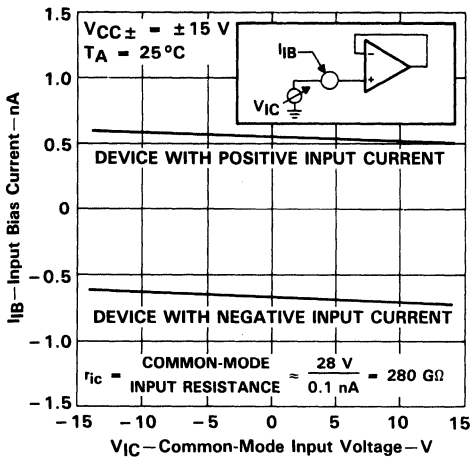


FIGURE 6

COMMON-MODE INPUT VOLTAGE RANGE LIMITS
vs
FREE-AIR TEMPERATURE

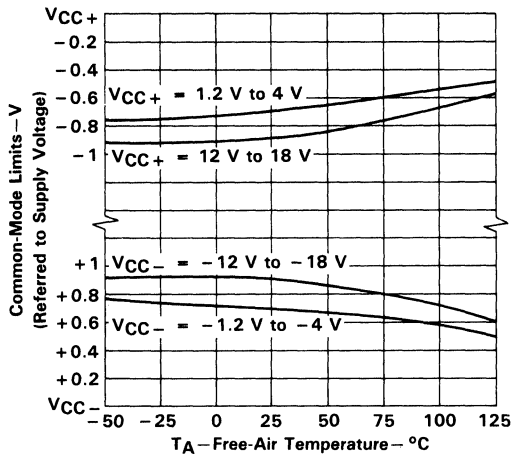


FIGURE 7

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

MAXIMUM PEAK
OUTPUT VOLTAGE SWING
vs
LOAD RESISTANCE

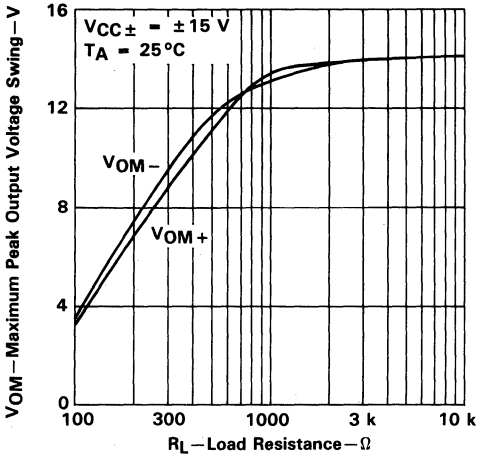


FIGURE 8

MAXIMUM PEAK-TO-PEAK
OUTPUT VOLTAGE SWING
vs
FREQUENCY

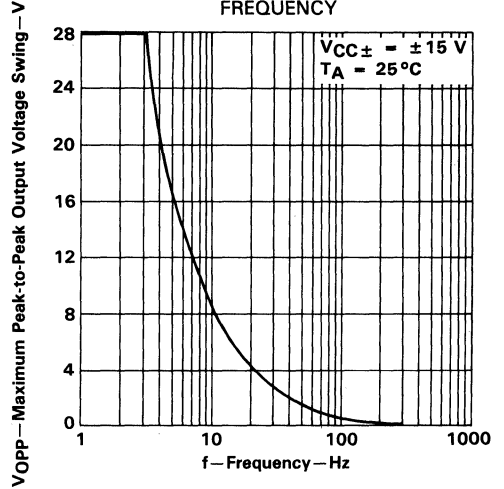


FIGURE 9

DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREQUENCY

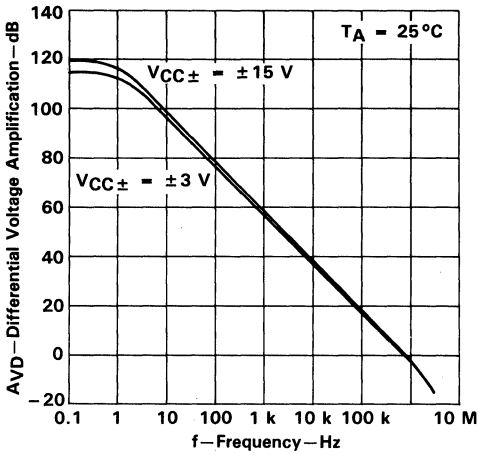


FIGURE 10

DIFFERENTIAL VOLTAGE AMPLIFICATION
and PHASE SHIFT
vs
FREQUENCY

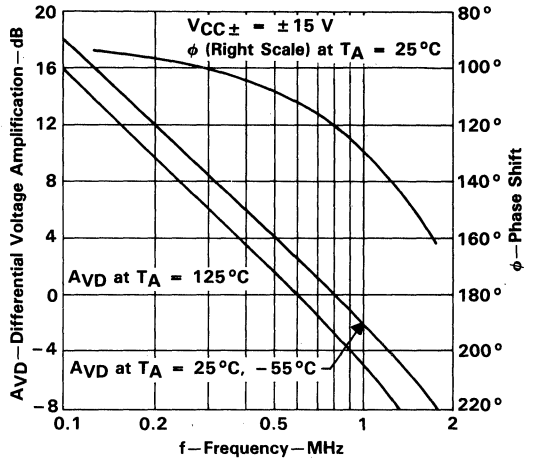


FIGURE 11

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREE-AIR TEMPERATURE

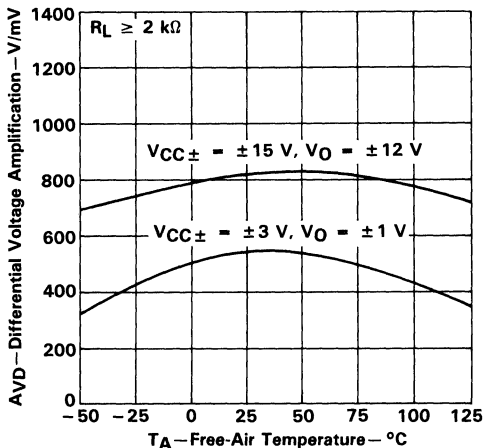


FIGURE 12

OUTPUT IMPEDANCE
vs
FREQUENCY

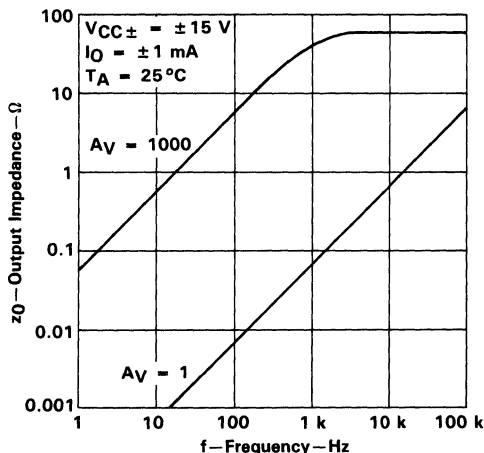


FIGURE 13

COMMON-MODE REJECTION RATIO
vs
FREQUENCY

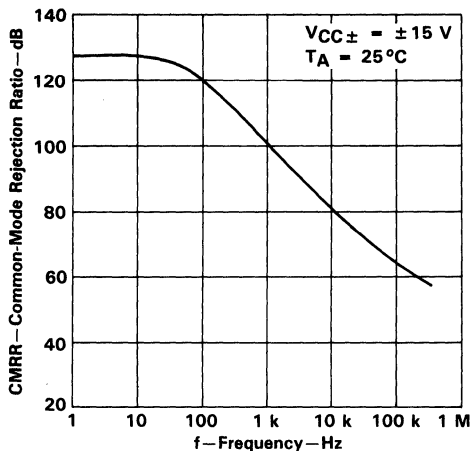


FIGURE 14

SUPPLY VOLTAGE REJECTION RATIO
vs
FREQUENCY

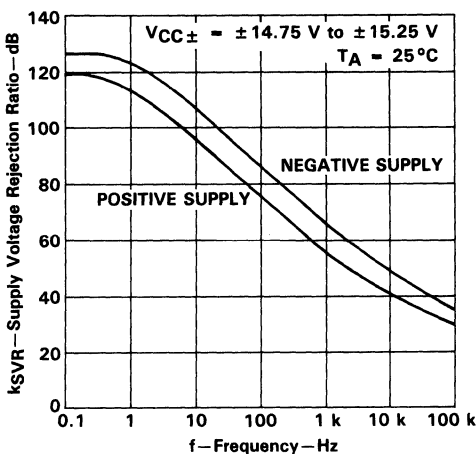


FIGURE 15

2
Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

SUPPLY CURRENT
vs
SUPPLY VOLTAGE

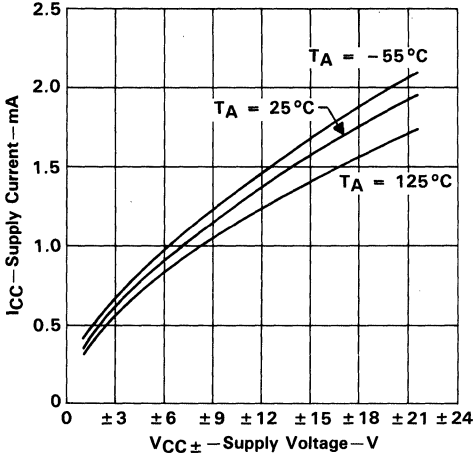


FIGURE 16

SHORT-CIRCUIT OUTPUT CURRENT
vs
ELAPSED TIME

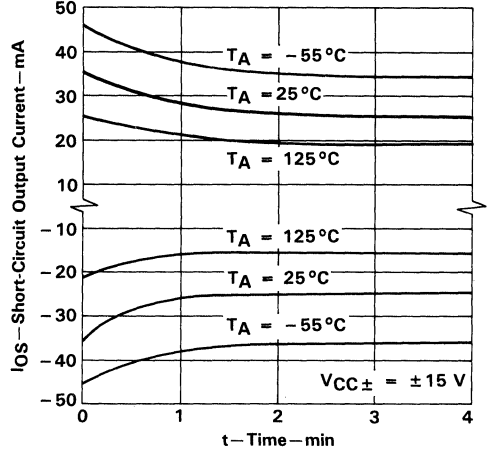


FIGURE 17

EQUIVALENT INPUT NOISE VOLTAGE
and EQUIVALENT INPUT NOISE CURRENT
vs
FREQUENCY

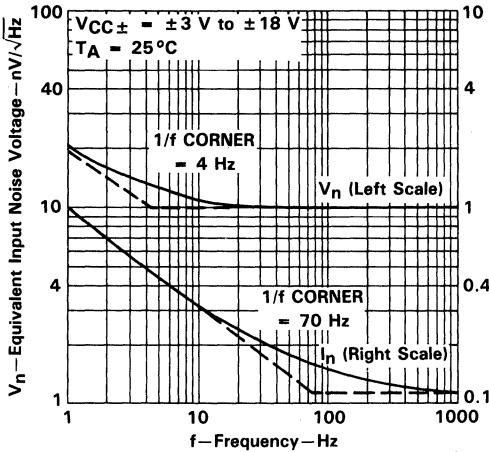


FIGURE 18

OUTPUT NOISE VOLTAGE
OVER A
10-SECOND PERIOD

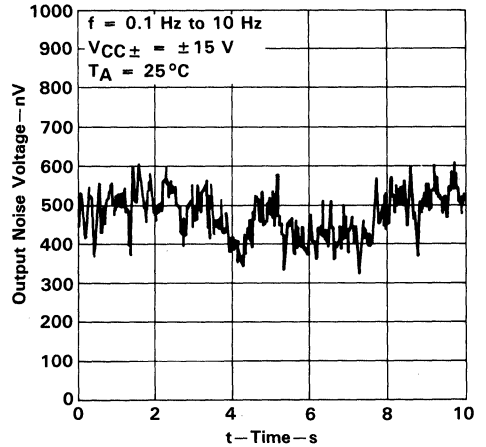


FIGURE 19

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

VOLTAGE-FOLLOWER
SMALL-SIGNAL
PULSE RESPONSE

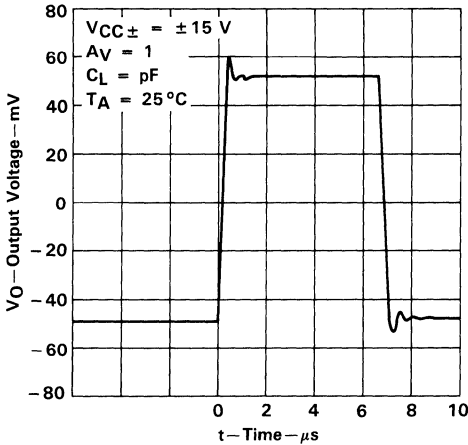


FIGURE 20

VOLTAGE-FOLLOWER
SMALL-SIGNAL
PULSE RESPONSE

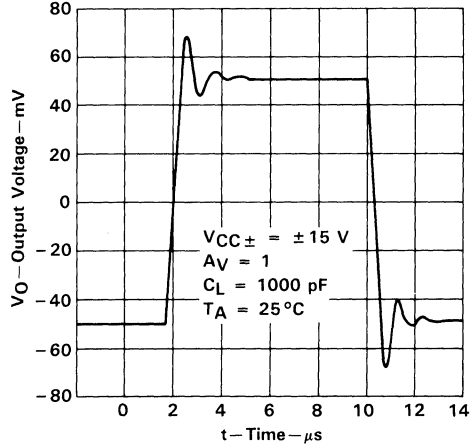


FIGURE 21

VOLTAGE-FOLLOWER
LARGE-SIGNAL
PULSE-RESPONSE

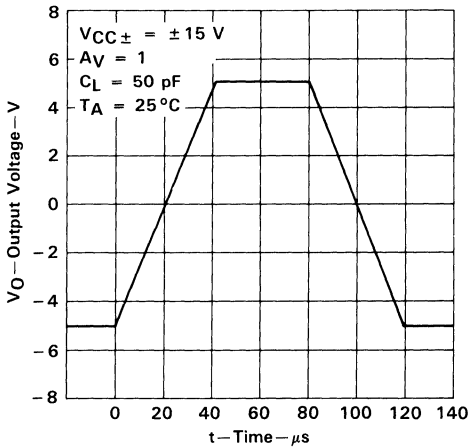


FIGURE 22

VOLTAGE-FOLLOWER OVERSHOOT
vs
LOAD CAPACITANCE

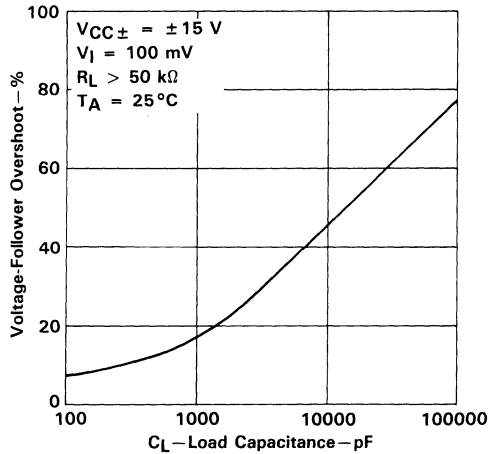
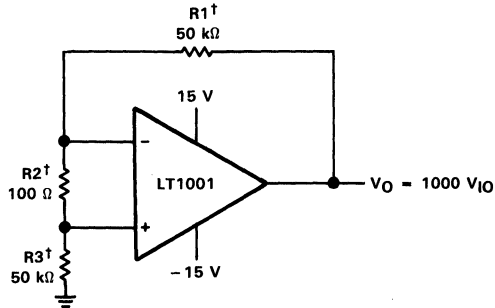


FIGURE 23

2
Operational Amplifiers

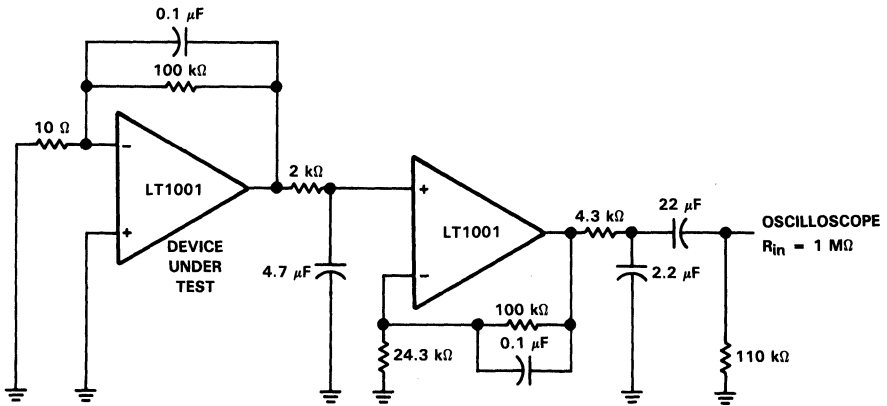
PARAMETER MEASUREMENT INFORMATION



†Resistors must have low thermoelectric potential.

NOTE: This circuit is also used as the burn-in configuration for the LT1001 with supply voltages increased to ± 20 V, $R1 = R3 = 10$ k Ω , $R2 = 200$ Ω , and $A_V = 100$.

FIGURE 24. TEST CIRCUIT FOR INPUT OFFSET VOLTAGE AND ITS TEMPERATURE COEFFICIENT



NOTES: A. $A_V = 50,000$.

B. The device under test should be warmed up for three minutes and shielded from air currents.

FIGURE 25. TEST CIRCUIT FOR 0.1-Hz TO 10-Hz PEAK-TO-PEAK NOISE VOLTAGE (MEASURED OVER A 10-SECOND INTERVAL)

TYPICAL APPLICATION DATA

application notes

The LT1001 series units may be inserted directly into OP-07 or LM108A sockets with or without removing external frequency compensation or nulling components.

The LT1001 is specified over a wide range of supply voltages from ± 3 V to ± 18 V. Operation with lower supply voltages (e.g., two Ni-Cad batteries) is possible down to ± 1.2 V. However, with ± 1.2 -V supplies, the device is stable only in closed-loop gains of 2 or higher (or inverting gains of one or higher).

Unless proper care is exercised, thermocouple effects caused by temperature gradients across dissimilar metals at the contacts to the input terminals can exceed the inherent temperature coefficient of the amplifier. Air currents over device leads should be minimized, package leads should be short, and the two input leads should be as close together as possible and maintained at the same temperature.

input offset voltage adjustment

The input offset voltage and temperature coefficient of the LT1001 are permanently trimmed to a low level at wafer test. However, if further adjustment of V_{IO} is necessary, nulling with a 10-k Ω or 20-k Ω potentiometer will not degrade the temperature coefficient. Trimming to a value other than zero creates a temperature coefficient change of $(V_{IO}/300)$ $\mu\text{V}/^\circ\text{C}$. For example, if V_{IO} is adjusted to 300 μV , the change in the temperature coefficient will be 1 $\mu\text{V}/^\circ\text{C}$. The adjustment range with a 10-k Ω or 20-k Ω potentiometer is approximately ± 2.5 mV. If less adjustment range is needed, the sensitivity and resolution of the nulling can be improved by using a smaller potentiometer in conjunction with fixed resistors. The example in Figure 26 has an approximate null range of ± 100 μV .

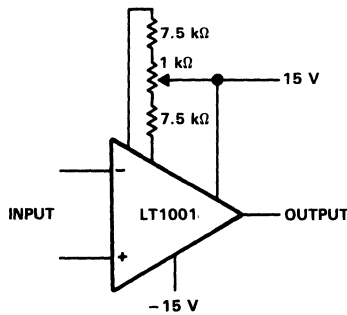
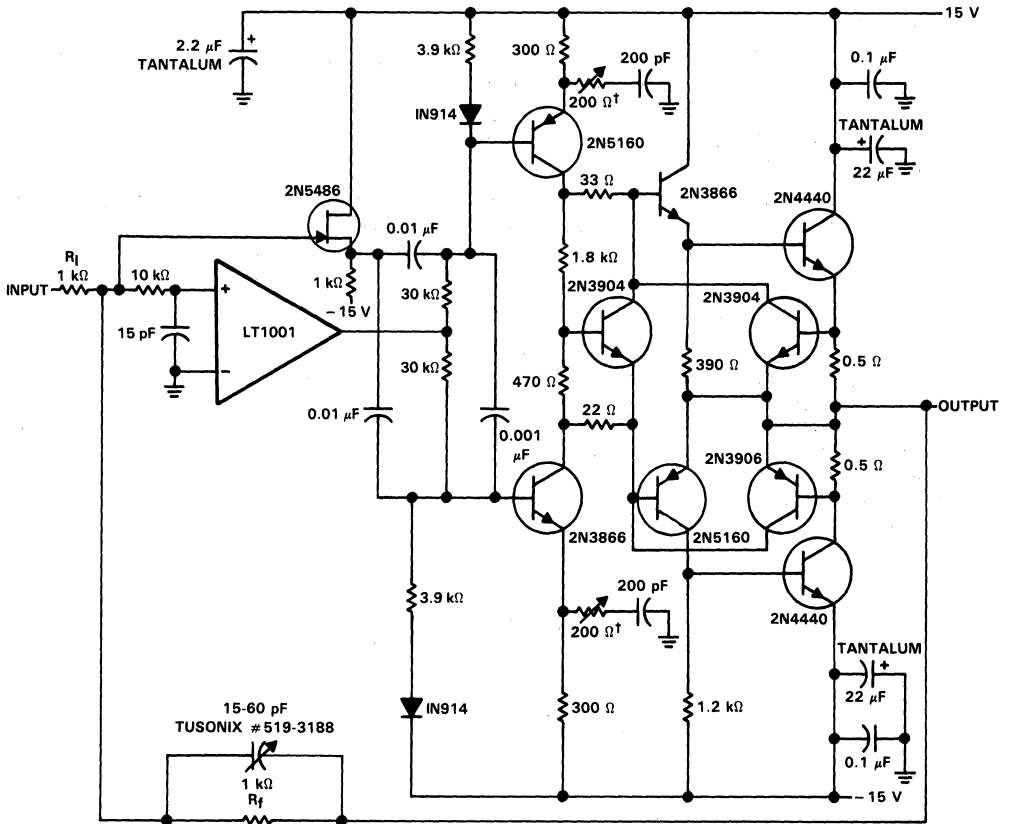


FIGURE 26. IMPROVED SENSITIVITY ADJUSTMENT

TYPICAL APPLICATION DATA

2
Operational Amplifiers



†Adjust for best square wave at output.
NOTE: Full-power bandwidth is 8 MHz.

FIGURE 27. DC-STABILIZED 1000-V/μs OPERATIONAL AMPLIFIER

TYPICAL APPLICATION DATA

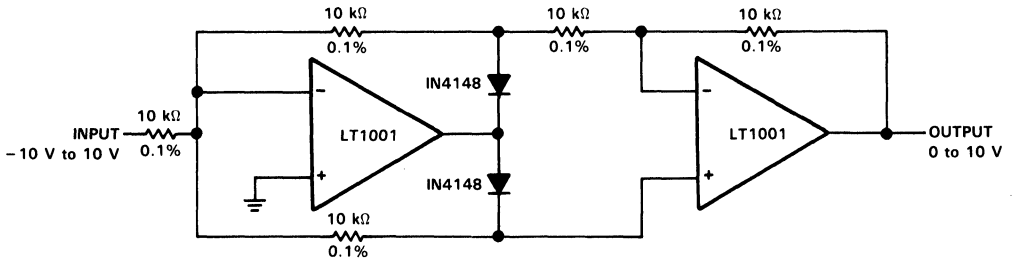


FIGURE 28. PRECISION ABSOLUTE VALUE CIRCUIT

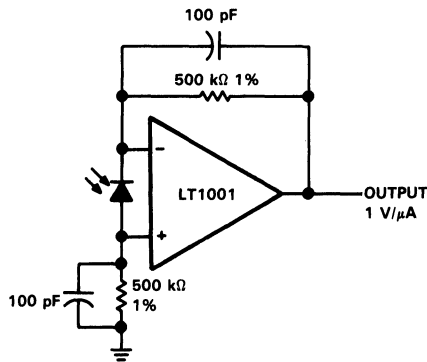


FIGURE 29. PHOTODIODE AMPLIFIER.

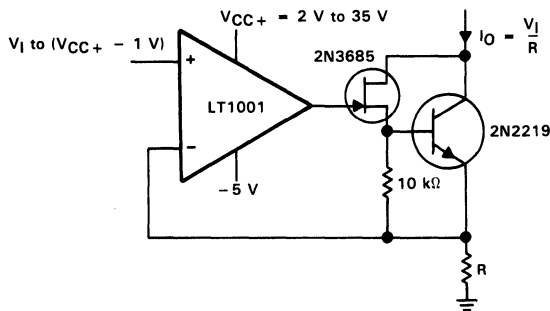
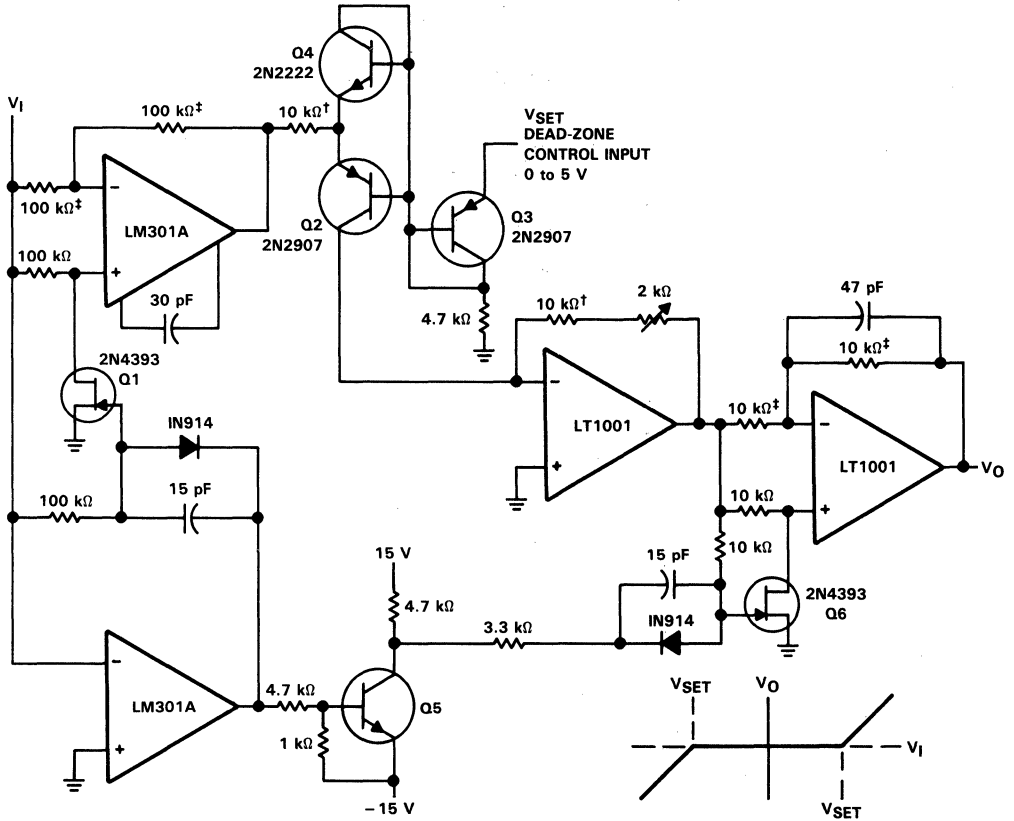


FIGURE 30. PRECISION CURRENT SINK

LT1001
PRECISION OPERATIONAL AMPLIFIER

TYPICAL APPLICATION DATA

2
Operational Amplifiers



† 1% film
‡ Ratio match 0.05%

NOTES: A. The bipolar symmetry for this application is excellent because one device, Q2, sets both limits.
B. Q2-Q5 are a CA 3096 transistor array.

FIGURE 31. DEAD-ZONE GENERATOR

TYPICAL APPLICATION DATA

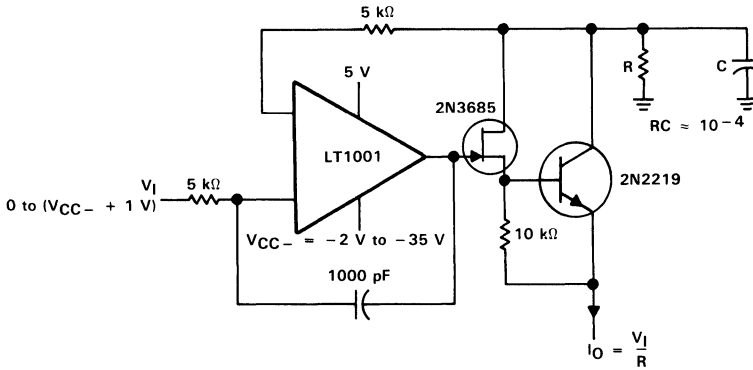
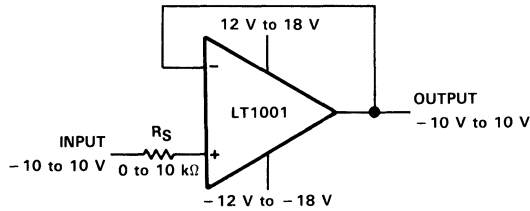


FIGURE 32. PRECISION CURRENT SOURCE



OUTPUT ACCURACY

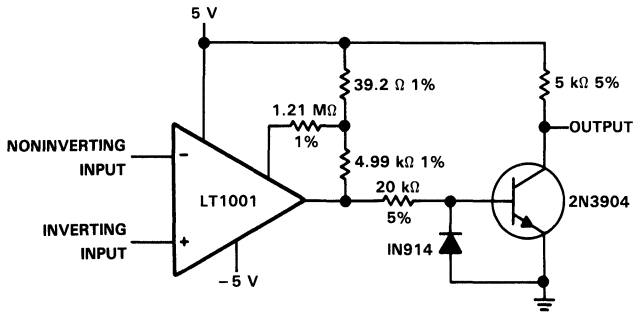
ERROR	LT1001AM $T_A = 25^\circ\text{C}$ MAX	LT1001C $T_A = 25^\circ\text{C}$ MAX	LT1001AM $T_A = -55^\circ\text{C to } 125^\circ\text{C}$ MAX	LT1001C $T_A = 0^\circ\text{C to } 70^\circ\text{C}$ MAX
Input Offset Voltage	15 μV	60 μV	60 μV	110 μV
Input Bias Current	20 μV	40 μV	40 μV	55 μV
Common-Mode Rejection Ratio	20 μV	30 μV	30 μV	50 μV
Supply Voltage Rejection Ratio	18 μV	30 μV	36 μV	42 μV
Differential Voltage Amplification	22 μV	25 μV	33 μV	40 μV
Worst-case Sum	95 μV	185 μV	199 μV	297 μV
Percent of Full Scale (= 20 V)	0.0005%	0.0009%	0.0010%	0.0015%

NOTE: The contributing error terms are due to input offset voltage, input bias current, voltage gain, common-mode rejection ratio, and supply voltage rejection ratio. The worst-case specifications are given in the above table.

FIGURE 33. LARGE-SIGNAL VOLTAGE FOLLOWER WITH 0.001% WORST-CASE ACCURACY

LT1001 PRECISION OPERATIONAL AMPLIFIER

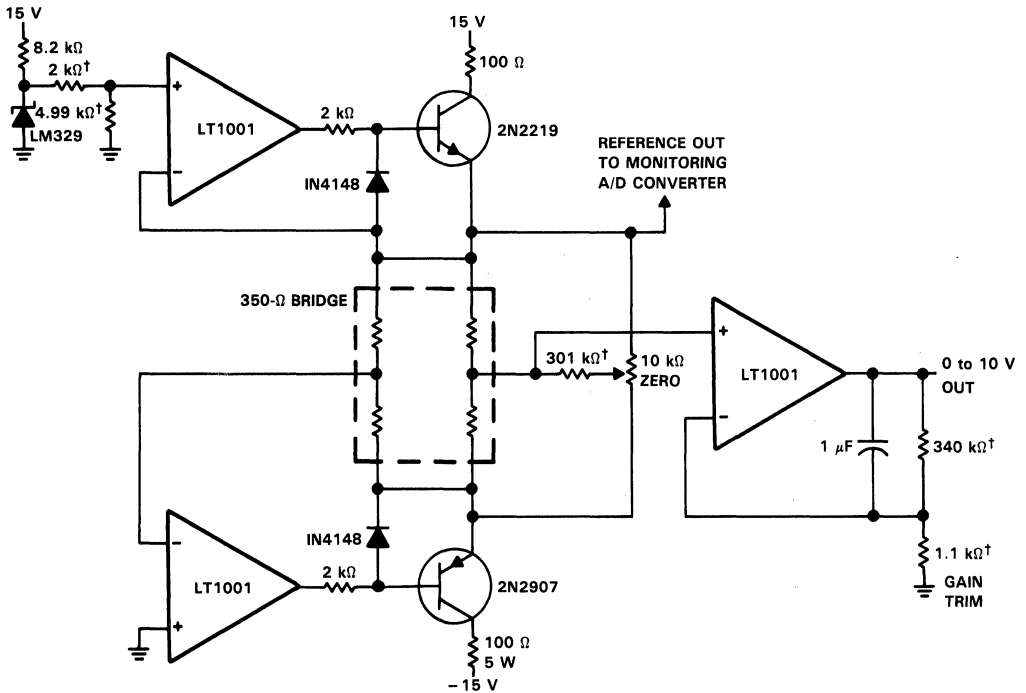
TYPICAL APPLICATION DATA



NOTE: Positive feedback to one of the nulling terminals creates 5 μV to 20 μV of hysteresis. The input offset voltage is typically changed by less than 5 μV due to the feedback.

FIGURE 34. MICROVOLT COMPARATOR WITH TTL OUTPUT

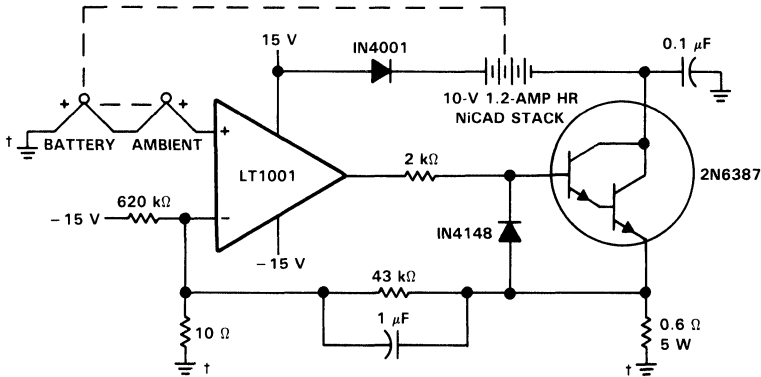
2
Operational Amplifiers



†RN60C film resistors

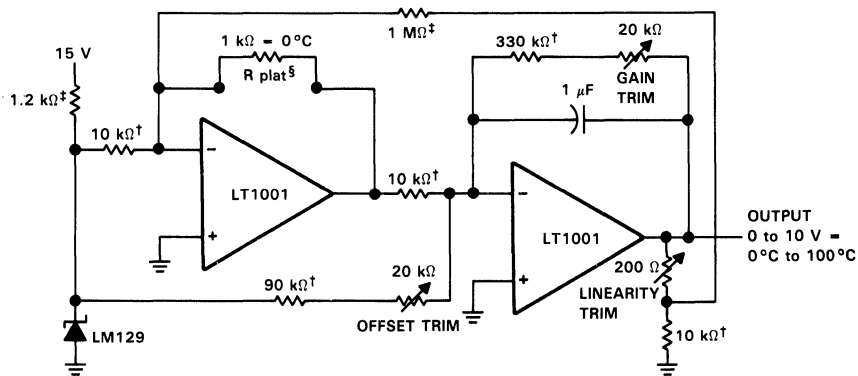
FIGURE 35. STRAIN-GAUGE SIGNAL CONDITIONER WITH BRIDGE EXCITATION

TYPICAL APPLICATION DATA



†Single point ground thermocouples are $40 \mu\text{V}/^\circ\text{C}$ chromel-alumel (type K).
NOTE: This circuit uses the temperature difference between the battery pack mounted thermocouple and the ambient thermocouple to set the battery charge current. The peak charging current is 1 A.

FIGURE 36. THERMALLY CONTROLLED NICAD CHARGER



†ULTRONIX 105A wirewound

‡1% film

§Platinum RTD 118MF (Rosemount, Inc.)

NOTE: Trim sequence: trim offset ($0^\circ\text{C} = 1000 \Omega$), trim linearity ($35^\circ\text{C} = 1138.7 \Omega$), trim gain ($100^\circ\text{C} = 1392.6 \Omega$). Repeat until all three points are fixed with $\pm 0.025^\circ\text{C}$.

FIGURE 37. LINEARIZED PLATINUM RESISTANCE THERMOMETER WITH $\pm 0.025^\circ\text{C}$ ACCURACY FOR $T_A = 0^\circ\text{C}$ TO 100°C

2

Operational Amplifiers

LT1007, LT1007A, LT1037, LT1037A LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

D3195, FEBRUARY 1989

- **Maximum Equivalent Input Noise Voltage:**
3.8 nV/ $\sqrt{\text{Hz}}$ at 1 kHz
4.5 nV/ $\sqrt{\text{Hz}}$ at 10 Hz
- **Low Peak-to-Peak Equivalent Input Noise Voltage:** 60 nV Typ from 0.1 Hz to 10 Hz
- **Slew Rate (LT1037 and LT1037A):**
11 V/ μs Min

LT1007A and LT1037A Specifications:

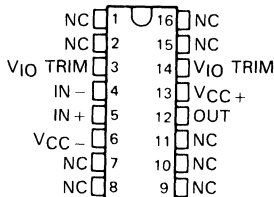
- **High Voltage Amplification:**
7 V/ μV Min, $R_L = 2 \text{ k}\Omega$
3 V/ μV Min, $R_L = 600 \Omega$
- **Low Input Offset Voltage 25 μV Max**
- **Low Input Offset Voltage Temperature Coefficient: 0.6 $\mu\text{V}/^\circ\text{C}$ Max**
- **Common-Mode Rejection Ratio: 117 dB Min**

description

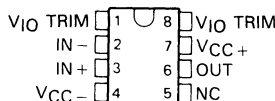
These monolithic operational amplifiers feature extremely low noise performance and outstanding precision and speed specifications. The typical differential voltage amplification (at $T_A = 25^\circ\text{C}$) of these devices is an extremely high 20 V/ μV driving a 2-k Ω load to $\pm 12 \text{ V}$ and 12 V/ μV driving a 600- Ω load to $\pm 10 \text{ V}$.

In the design, processing, and testing of the device, particular attention has been paid to the optimization of the entire distribution of several key parameters. Consequently, the specifications of even the lowest-cost grades (the LT1007C and the LT1037C) have been greatly improved compared to equivalent grades of competing amplifiers.

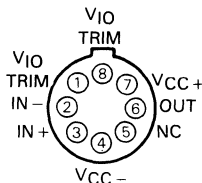
**DW SMALL-OUTLINE PACKAGE
(TOP VIEW)**



**JG AND P
DUAL-IN-LINE PACKAGES
(TOP VIEW)**



**L PLUG-IN PACKAGE
(TOP VIEW)**



Pin 4 (L Package) is in electrical contact with the case
NC—No internal connection

AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE			
		SMALL-OUTLINE (DW)	CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)
0°C to 70°C	60 μV	LT1007CDW	LT1007CJG	LT1007CL	LT1007CP
	25 μV	—	LT1007ACJG	LT1007ACL	LT1007ACP
—55°C to 125°C	60 μV	LT1037CDW	LT1037CJG	LT1037CL	LT1037CP
	25 μV	—	LT1037ACJG	LT1037ACL	LT1037ACP
—55°C to 125°C	60 μV	—	LT1007MJG	LT1007ML	LT1007MP
	25 μV	—	LT1007AMJG	LT1007AML	LT1007AMP
—55°C to 125°C	60 μV	—	LT1037MJG	LT1037ML	LT1037MP
	25 μV	—	LT1037AMJG	LT1037AML	LT1037AMP

The DW packages are available taped and reeled. Add the suffix "R" to the device type. (e.g., LT1007CDWR).

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



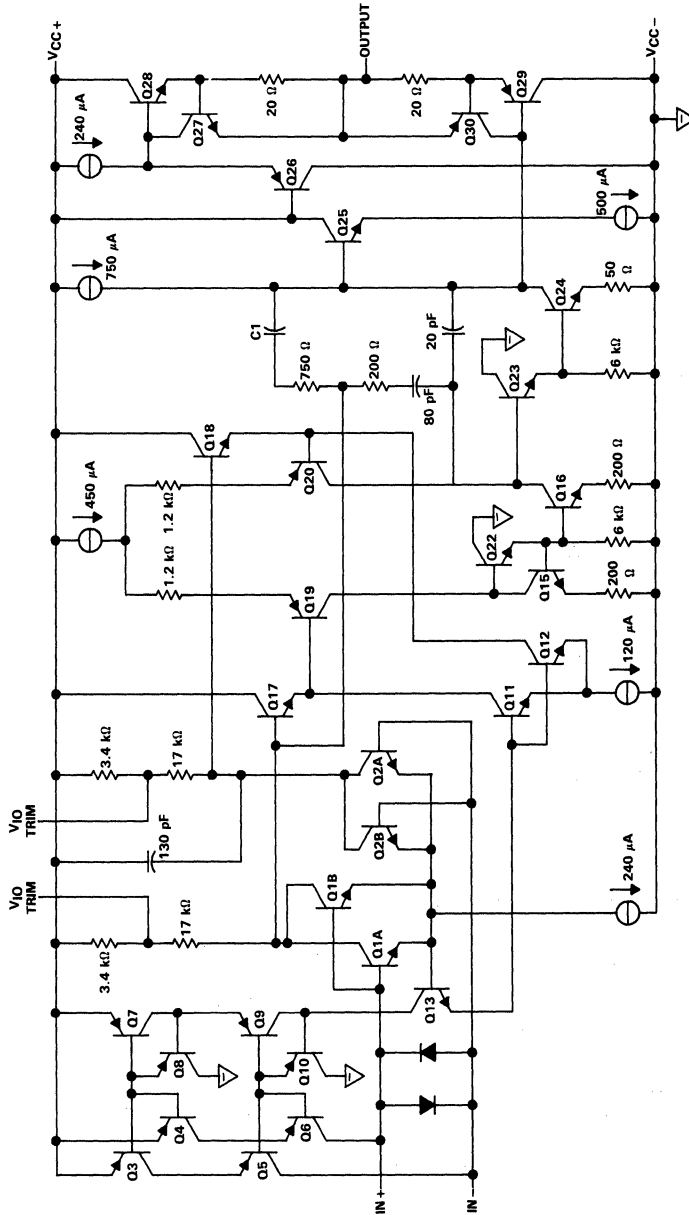
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LT1007, LT1007A, LT1037, LT1037A
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

schematic

2
Operational Amplifiers



C1 = 110 pF for LT1007
 C1 = 12 pF for LT1037
 All component values shown are nominal.

LT1007, LT1007A, LT1037, LT1037A LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	22 V
Supply voltage, V_{CC-}	-22 V
Input voltage	$V_{CC\pm}$
Duration of output short-circuit	Unlimited
Differential input current (see Note 2)	± 25 mA
Power dissipation	see Dissipation Rating Table
Operating free-air temperature range:	
LT1007M, LT1007AM, LT1037M, LT1037AM	-55°C to 125°C
LT1007C, LT1007AC, LT1037C, LT1037AC	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: DW and P packages	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG and L packages	300°C

2
Operational Amplifiers

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. The inputs are protected by back-to-back diodes. Current limiting resistors are not used in order to achieve low noise. Excessive input current will flow if a differential input voltage in excess of approximately ± 0.7 V is applied between the inputs, unless some limiting resistance is used.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
DW	1025 mW	8.2 mW/°C	656 mW	N/A	N/A
JG (M-suffix)	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
JG (C-suffix)	825 mW	6.6 mW/°C	528 mW	429 mW	N/A
L (M-suffix)	825 mW	6.6 mW/°C	528 mW	429 mW	165 mW
L (C-suffix)	650 mW	5.2 mW/°C	416 mW	338 mW	N/A
P	1000 mW	8 mW/°C	640 mW	520 mW	200 mW

recommended operating conditions

	LT1007M, LT1037M			LT1007C, LT1037C			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, V_{CC+}	4	15	22	4	15	22	V
Supply voltage, V_{CC-}	-4	-15	-22	-4	-15	-22	V
Input voltage, V_I	$T_A = 25^\circ\text{C}$			± 11			V
	$T_A = \text{full range}$			± 10.5			V
Operating free-air temperature, T_A	-55		125	0		70	°C

LT1007M, LT1037M, LT1007AM, LT1037AM
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIER

electrical characteristics, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$

PARAMETER	TEST CONDITIONS	T _A	LT1007, LT1037			LT1007A, LT1037A			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	See Note 3	25°C	20 60			10 25			μV
		-55°C to 125°C	160			60			
αV _{IO} Average temperature coefficient of input offset voltage		-55°C to 125°C	1			0.6			μV/°C
I _{IO} Input offset current		25°C	12 50			7 30			nA
		-55°C to 125°C	85			50			
I _{IB} Input bias current		25°C	±15 ±55			±10 ±35			nA
		-55°C to 125°C	±95			±60			
V _{OM} Peak output voltage swing	R _L = 2 kΩ	25°C	±12.5 ±13.5			±13 ±13.8			V
	R _L = 600 Ω	25°C	±10.5 ±12.5			±11 ±12.5			
	R _L = 2 kΩ	-55°C to 125°C	±12			±12.5			
A _{VD} Large-signal differential voltage amplification	R _L ≥ 2 kΩ, V _O = ±12 V	25°C	5 20			7 20			V/μV
	R _L ≥ 1 kΩ, V _O = ±10 V	25°C	3.5 16			5 16			
	R _L ≥ 600 Ω, V _O = ±10 V	25°C	2 12			3 12			
	R _L ≥ 2 kΩ, V _O = ±10 V	-55°C to 125°C	2			3			
	R _L ≥ 1 kΩ, V _O = ±10 V	-55°C to 125°C	1.5			2			
r _{i(CM)} Common-mode input resistance		25°C	5			7			GΩ
r _o Open-loop output resistance		25°C	70			70			Ω
CMRR Common-mode rejection ratio	V _{IC} = ±11 V	25°C	110 126			117 130			dB
	V _{IC} = ±10.3 V	-55°C to 125°C	104			112			
k _{SVR} Supply voltage rejection ratio	V _{CC±} = ±4 V to ±18 V	25°C	106 126			110 130			dB
	V _{CC±} = ±4.5 V to ±18 V	-55°C to 125°C	100			104			
P _D Power dissipation	LT1007M, LT1007AM	25°C	80 140			80 120			mW
	LT1037M, LT1037AM	25°C	85 140			80 130			
		-55°C to 125°C	170			150			

NOTE 3: V_{IO} measurements are performed by automatic test equipment approximately 0.5 seconds after application of power.

LT1007C, LT1037C, LT1007AC, LT1037AC
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIER

electrical characteristics, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	LT1007, LT1037			LT1007A, LT1037A			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	See Note 3	25°C	20 60			10 25			μV
		0°C to 70°C	110			50			
αV_{IO} Average temperature coefficient of input offset voltage		0°C to 70°C	1			0.6			$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current		25°C	12 50			7 30			nA
		0°C to 70°C	70			40			
I_{IB} Input bias current		25°C	$\pm 15 \pm 55$			$\pm 10 \pm 35$			nA
		0°C to 70°C	± 75			± 45			
V_{OM} Peak output voltage swing	$R_L = 2\text{ k}\Omega$	25°C	$\pm 12.5 \pm 13.5$			$\pm 13 \pm 13.8$			V
	$R_L = 600\ \Omega$	25°C	$\pm 10.5 \pm 12.5$			$\pm 11 \pm 12.5$			
	$R_L = 2\text{ k}\Omega$	0°C to 70°C	± 12			± 12.5			
A_{VD} Large-signal differential voltage amplification	$R_L \geq 2\text{ k}\Omega$, $V_O = \pm 12\text{ V}$	25°C	5 20			7 20			$\text{V}/\mu\text{V}$
	$R_L \geq 1\text{ k}\Omega$, $V_O = \pm 10\text{ V}$	25°C	3.5 16			5 16			
	$R_L \geq 600\ \Omega$, $V_O = \pm 10\text{ V}$	25°C	2 12			3 12			
	$R_L \geq 2\text{ k}\Omega$, $V_O = \pm 10\text{ V}$	0°C to 70°C	2.5			4			
$R_L \geq 1\text{ k}\Omega$, $V_O = \pm 10\text{ V}$	0°C to 70°C	2			2.5				
$r_{i(CM)}$ Common-mode input resistance		25°C	5			7			$\text{G}\Omega$
r_o Open-loop output resistance		25°C	70			70			Ω
CMRR Common-mode rejection ratio	$V_{IC} = \pm 11\text{ V}$	25°C	110 126			117 130			dB
	$V_{IC} = \pm 10.5\text{ V}$	0°C to 70°C	106			114			
k_{SVR} Supply voltage rejection ratio	$V_{CC\pm} = \pm 4\text{ V to } \pm 18\text{ V}$	25°C	106 126			110 130			dB
	$V_{CC\pm} = \pm 4.5\text{ V to } \pm 18\text{ V}$	0°C to 70°C	102			106			
P_D Power dissipation	LT1007M, LT1007AM	25°C	80 140			80 120			mW
		25°C	85 140			80 130			
	LT1037M, LT1037AM	0°C to 70°C	160			144			

NOTE 3: V_{IO} measurements are performed by automatic test equipment approximately 0.5 seconds after application of power.

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Operational Amplifiers

LT1007, LT1007A, LT1037, LT1037A

LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

operating characteristics $V_{CC\pm} = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	LT1007/LT1007A			LT1037/LT1037A			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate	$R_L \geq 2\text{ k}\Omega$, $A_{VD} \geq 1$ (LT1007, LT1007A)	1.7	2.5		11	15		$\text{V}/\mu\text{s}$
	$R_L \geq 2\text{ k}\Omega$, $A_{VD} \geq 5$ (LT1037, LT1037A)							
V_{NPP} Peak-to-peak equivalent input noise voltage	0.1 Hz to 10 Hz, See Note 4	0.06	0.13		0.06	0.13		μV
V_n Equivalent input noise voltage	10 Hz	2.8	4.5		2.8	4.5		$\text{nV}/\sqrt{\text{Hz}}$
	1 kHz	2.5	3.8		2.5	3.8		
I_n Equivalent input noise current	10 Hz, See Note 5	1.5	4		1.5	4		$\text{pA}/\sqrt{\text{Hz}}$
	1 kHz, See Note 5	0.4	0.6		0.4	0.6		
GBW Gain bandwidth product	100 kHz	5	8					MHz
	10 kHz, $A_V \geq 15$				45	60		

NOTES: 4. See the test circuit and frequency response curve for 0.1 Hz to 10 Hz noise (Figure 39) in the Applications Information section.
5. See the test circuit for current noise measurement (Figure 40) in the Applications Information section.

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Operational Amplifiers

LT1007, LT1007A, LT1037, LT1037A
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

table of graphs

			FIGURE
V_{IO}	Input offset voltage	vs Temperature	1
ΔV_{IO}	Change in input offset voltage	vs Time after power on	2
		vs Time (long-term stability)	3
I_{IO}	Input offset current	vs Temperature	4
I_{IB}	Input bias current	vs Temperature	5
		over common-mode range	6
	Common-mode limit voltage	vs Free-air temperature	7
V_{OM}	Maximum peak output voltage swing	vs Load resistance	8
		vs Frequency	9
A_{VD}	Differential voltage amplification	vs Frequency	10
		vs Frequency (LT1007)	11
		vs Frequency (LT1037)	12
		vs Temperature	13
		vs Load resistance	14
		vs Supply voltage	15
	at 2 k Ω and 600 Ω	16	
V_{ID}	Differential input voltage	vs Output voltage	16
CMRR	Common-mode rejection ratio	vs Frequency	17
k_{SVR}	Supply voltage rejection ratio	vs Frequency	18
SR	Slew rate	vs Free-air temperature (LT1007)	19
		vs Free-air temperature (LT1037)	20
ϕ	Phase shift	vs Frequency (LT1007)	11
		vs Frequency (LT1037)	12
ϕ_m	Phase margin	vs Free-air temperature (LT1007)	19
		vs Free-air temperature (LT1037)	20
V_n	Equivalent input noise voltage	vs Free-air temperature	21
		vs Time (0.01-Hz to 1-Hz noise)	22
		vs Frequency	23
		vs Bandwidth	24
		vs Supply voltage	25
I_n	Equivalent input noise current Total noise	vs Frequency	26
		vs Source resistance	27
GBW	Gain bandwidth product	vs Free-air temperature (LT1007)	19
		vs Free-air temperature (LT1037)	20
I_{OS}	Short-circuit output current	vs Time (from short to GND)	28
I_{CC}	Supply current	vs Supply voltage	29
z_o	Closed-loop output impedance	vs Frequency	30
	Pulse response (LT1007)	Small-signal ($C_{load} = 15$ pF)	31
		Large-signal	32
	Pulse response (LT1037)	Small-signal ($C_{load} = 15$ pF)	33
		Large-signal	34

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Operational Amplifiers

LT1007, LT1007A, LT1037, LT1037A
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

INPUT OFFSET VOLTAGE
 OF REPRESENTATIVE UNITS
 VS
 FREE-AIR TEMPERATURE

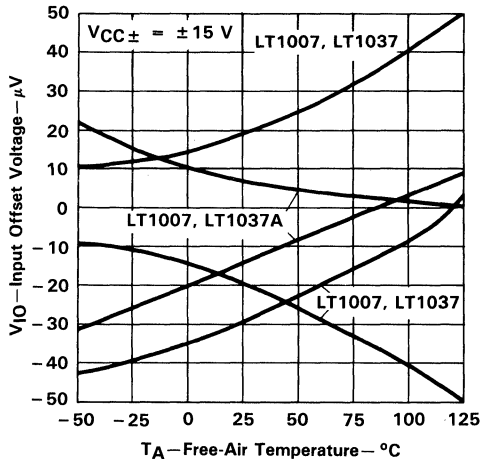


FIGURE 1

INPUT OFFSET VOLTAGE CHANGE
 VS
 TIME AFTER POWER ON

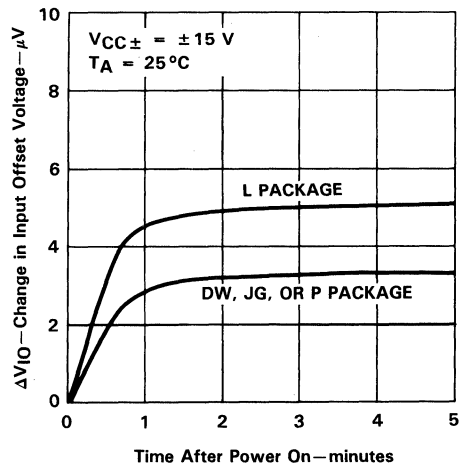


FIGURE 2

LONG TERM STABILITY OF
 INPUT OFFSET VOLTAGE
 FOR FOUR REPRESENTATIVE UNITS

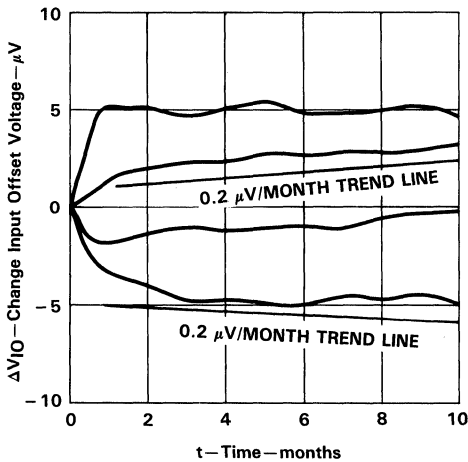


FIGURE 3

INPUT OFFSET CURRENT
 VS
 TEMPERATURE

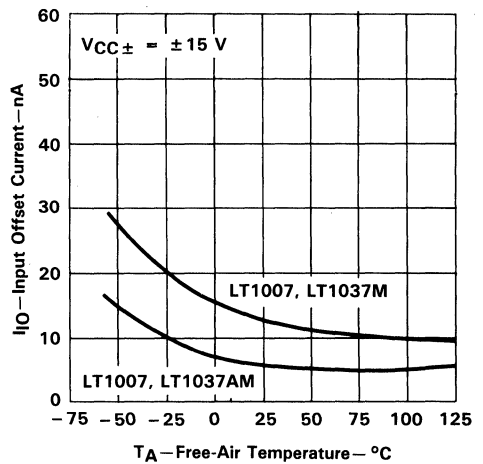


FIGURE 4

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

LT1007, LT1007A, LT1037, LT1037A
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

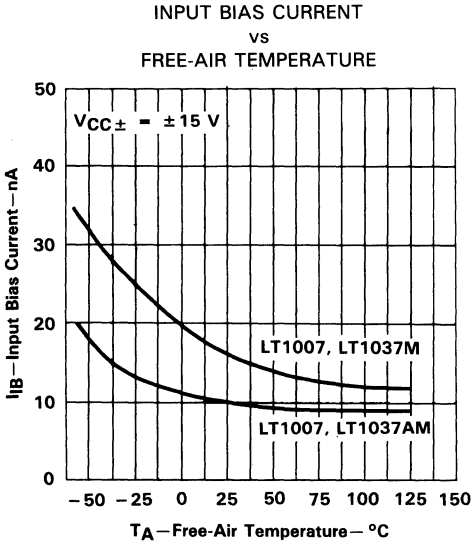


FIGURE 5

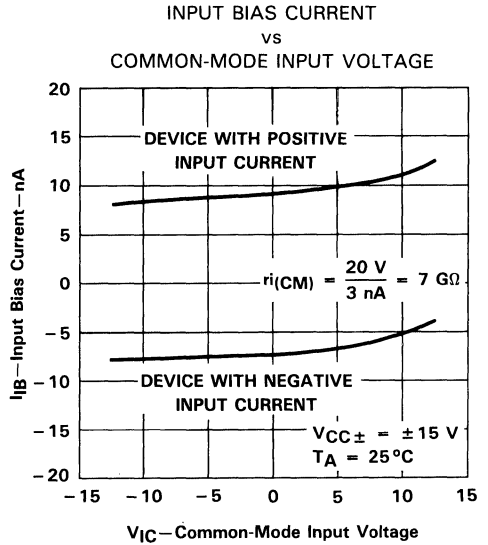


FIGURE 6

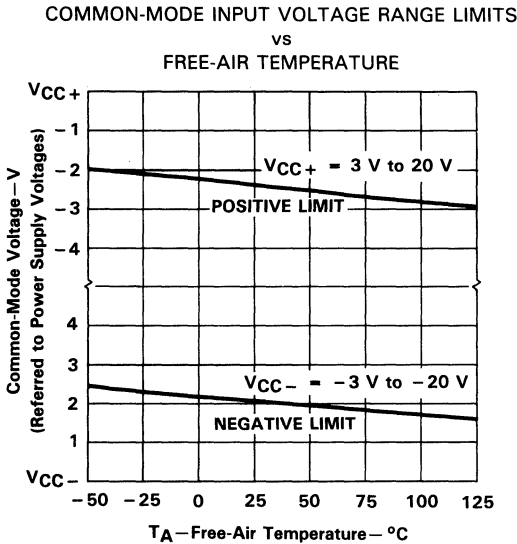


FIGURE 7

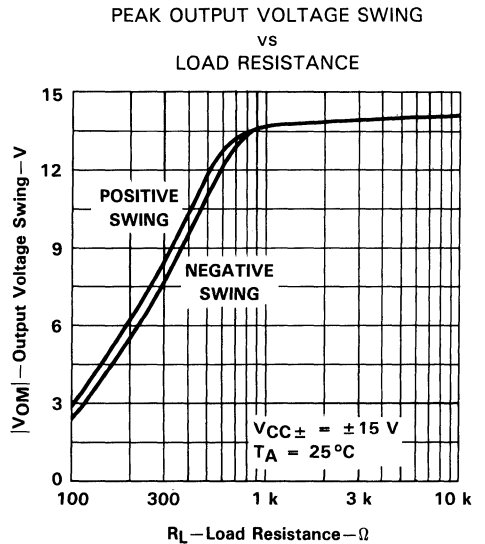


FIGURE 8

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

LT1007, LT1007A, LT1037, LT1037A
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

**PEAK-TO-PEAK OUTPUT VOLTAGE SWING
 vs
 FREQUENCY**

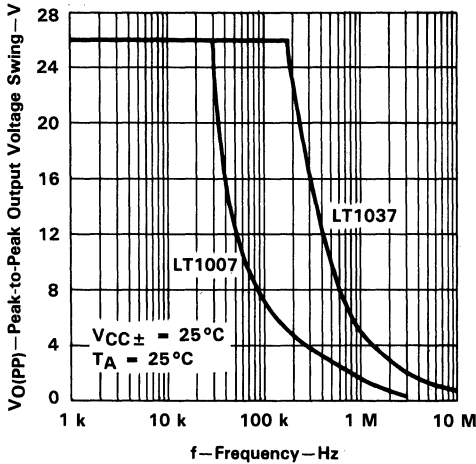


FIGURE 9

**DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 FREQUENCY**

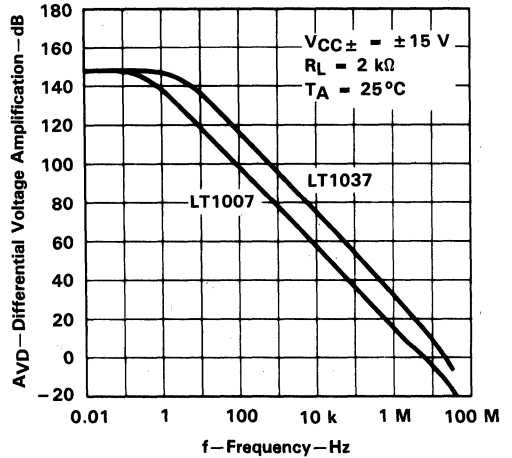


FIGURE 10

**LT1007
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 AND PHASE SHIFT
 vs
 FREQUENCY**

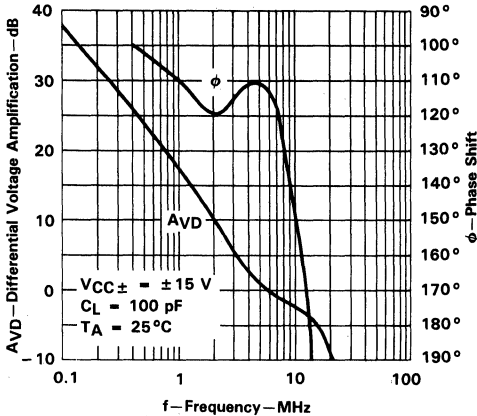


FIGURE 11

**LT1037
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 AND PHASE SHIFT
 vs
 FREQUENCY**

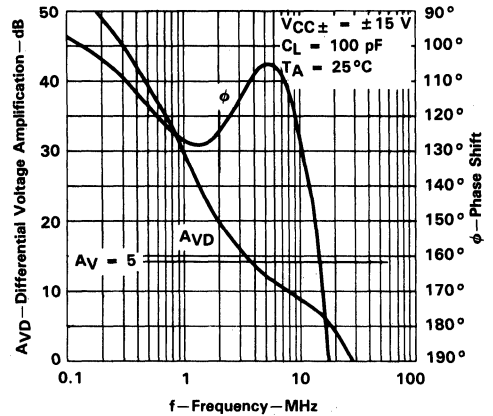


FIGURE 12

TYPICAL CHARACTERISTICS†

DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 FREE-AIR TEMPERATURE

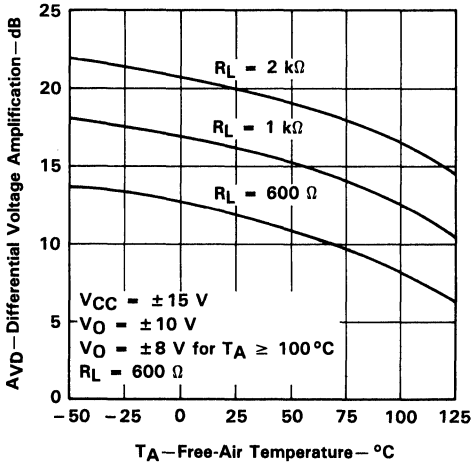


FIGURE 13

DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 LOAD RESISTANCE

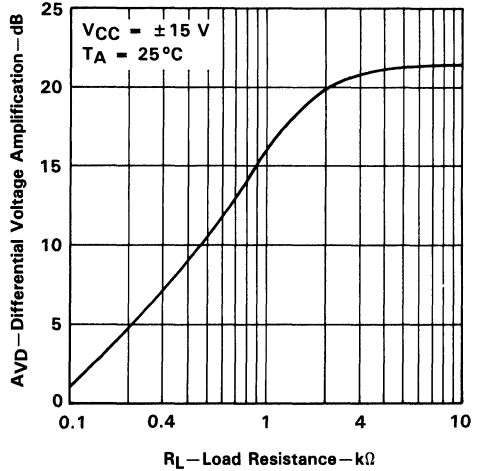


FIGURE 14

DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 SUPPLY VOLTAGE

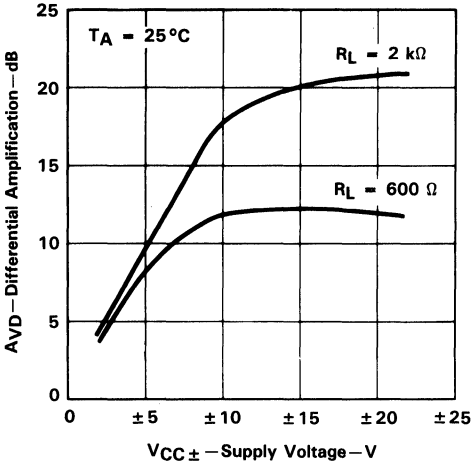


FIGURE 15

DIFFERENTIAL INPUT VOLTAGE
 vs
 OUTPUT VOLTAGE

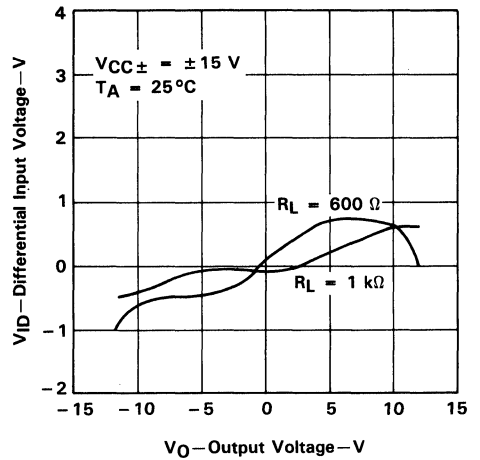


FIGURE 16

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

LT1007, LT1007A, LT1037, LT1037A
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

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Operational Amplifiers

COMMON-MODE REJECTION RATIO
vs
FREQUENCY

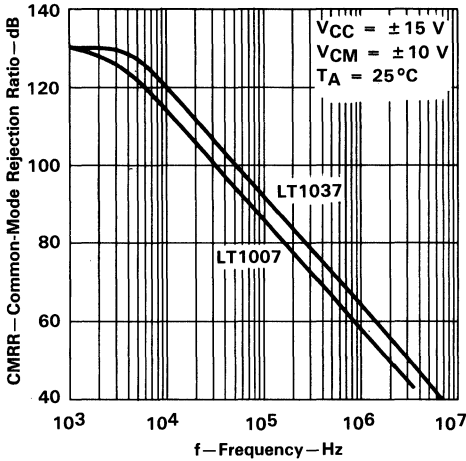


FIGURE 17

SUPPLY VOLTAGE REJECTION RATIO
vs
FREQUENCY

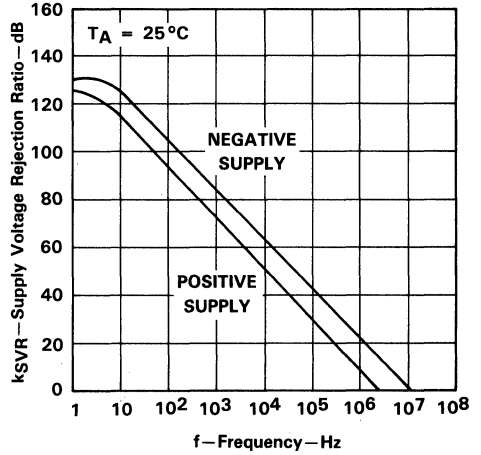


FIGURE 18

LT1007
SLEW RATE, PHASE MARGIN AND
GAIN BANDWIDTH PRODUCT
vs
FREE-AIR TEMPERATURE

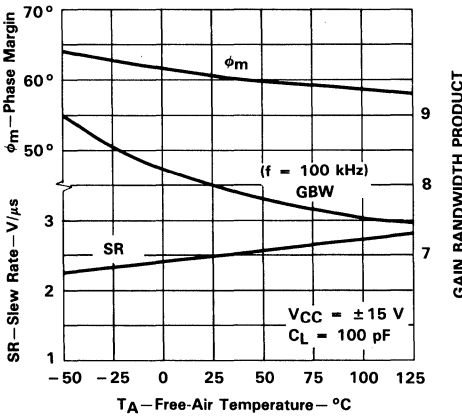


FIGURE 19

LT1037
SLEW RATE, PHASE MARGIN AND
GAIN BANDWIDTH PRODUCT
vs
FREE-AIR TEMPERATURE

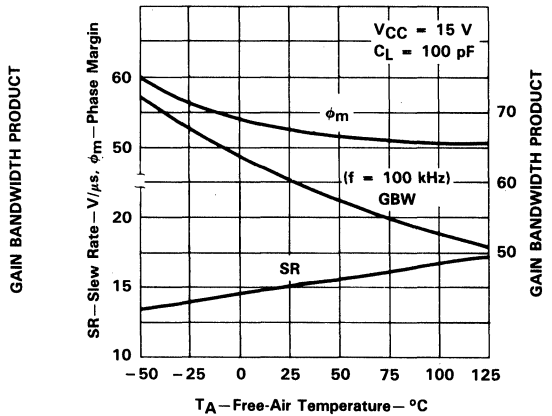


FIGURE 20

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

LT1007, LT1007A, LT1037, LT1037A
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

EQUIVALENT INPUT NOISE VOLTAGE
 vs
 FREE-AIR TEMPERATURE

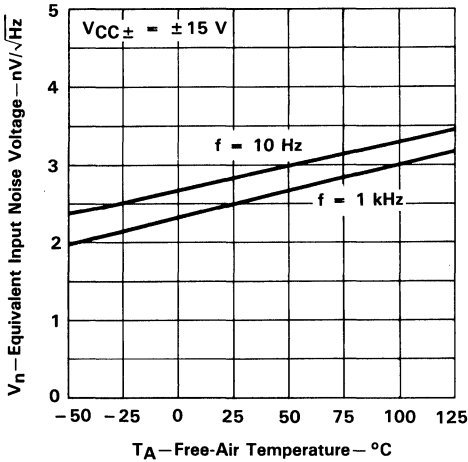


FIGURE 21

EQUIVALENT INPUT NOISE VOLTAGE
 OVER A 100-SECOND TIME PERIOD

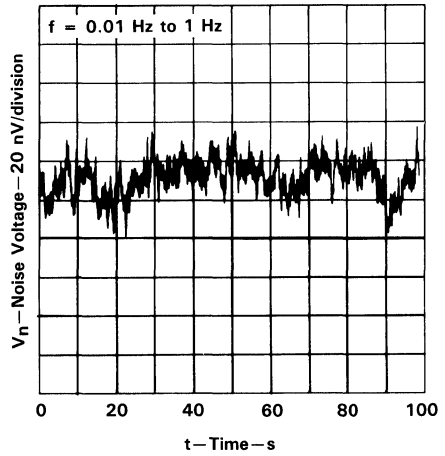


FIGURE 22

EQUIVALENT INPUT NOISE VOLTAGE
 vs
 FREQUENCY

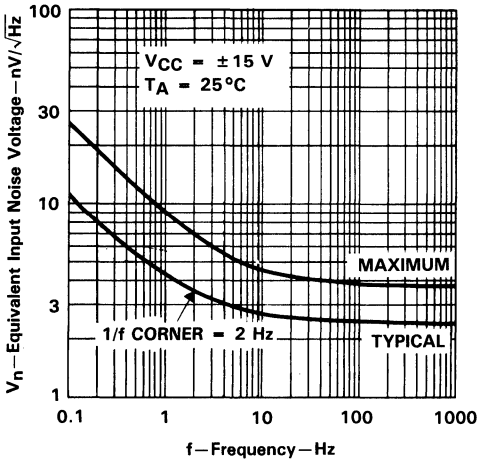


FIGURE 23

BROADBAND NOISE VOLTAGE
 0.1 Hz TO INDICATED FREQUENCY

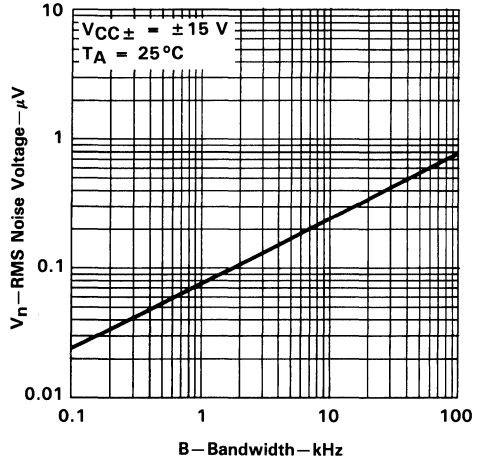


FIGURE 24

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Operational Amplifiers

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

LT1007, LT1007A, LT1037, LT1037A
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

EQUIVALENT INPUT NOISE VOLTAGE
vs
SUPPLY VOLTAGE

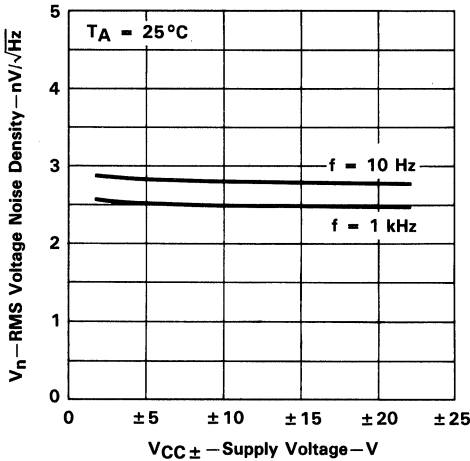


FIGURE 25

EQUIVALENT INPUT NOISE CURRENT
vs
FREQUENCY

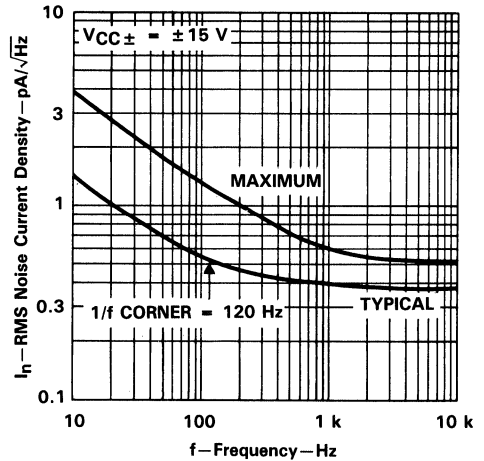


FIGURE 26

TOTAL NOISE VOLTAGE
vs
SOURCE RESISTANCE

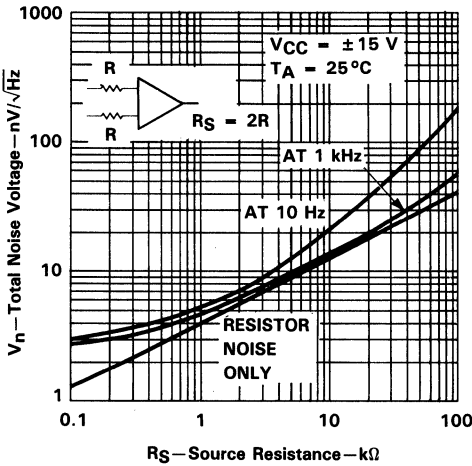


FIGURE 27

SHORT-CIRCUIT OUTPUT CURRENT
vs
ELAPSED TIME

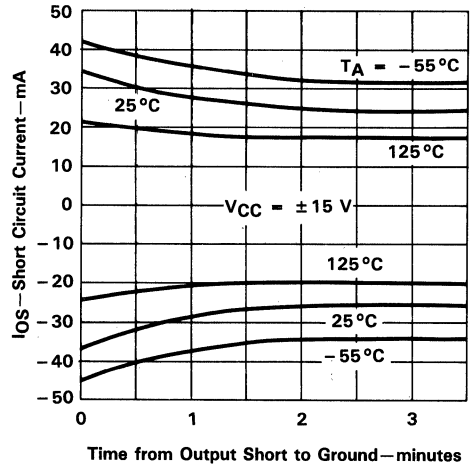


FIGURE 28

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

2 Operational Amplifiers

LT1007, LT1007A, LT1037, LT1037A LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

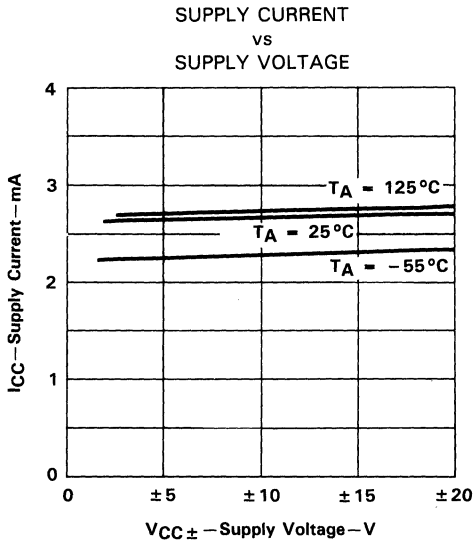


FIGURE 29

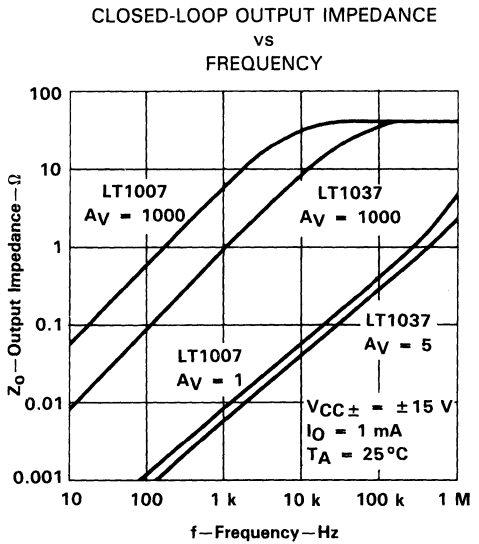


FIGURE 30

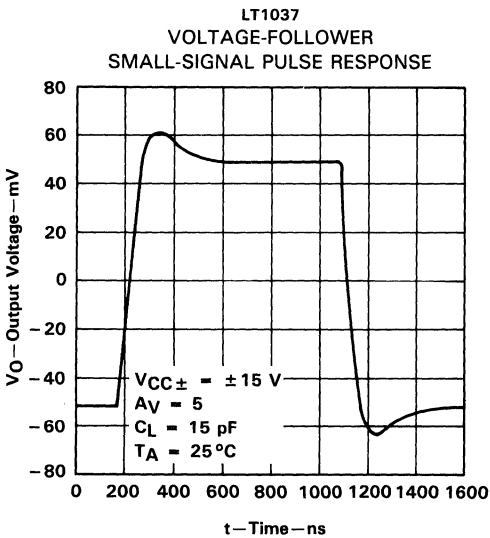


FIGURE 31

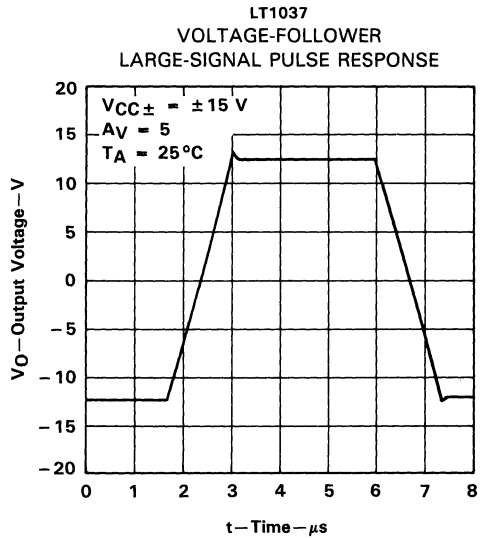


FIGURE 32

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Operational Amplifiers

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

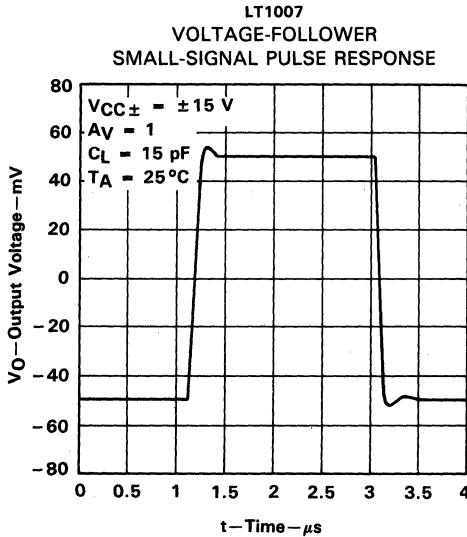


FIGURE 33

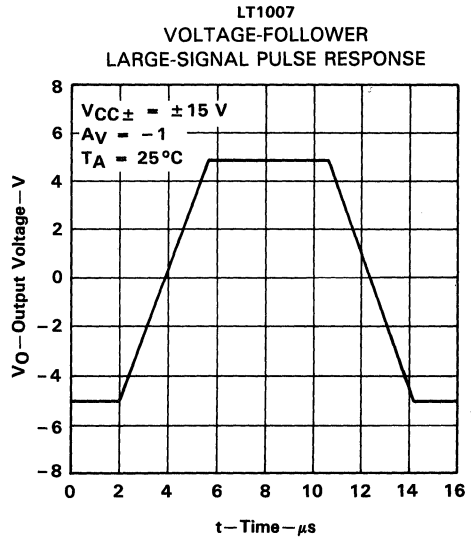


FIGURE 34

TYPICAL APPLICATION DATA

general

The LT1007- and LT1037-series devices may be inserted directly into OP-07, OP-27, OP-37, and 5534 sockets with or without removal of external-compensation or nulling components. In addition, the LT1007 and LT1037 may be fitted to μ A741 sockets by removing or modifying external nulling components.

offset voltage adjustment

The input offset voltage and its change with temperature of the LT1007 and LT1037 are permanently trimmed to a low level at wafer testing. However, if further adjustment of V_{IO} is necessary, the use of a 10-k Ω nulling potentiometer, as shown in Figure 35, will not degrade drift with temperature. Trimming to a value other than zero creates a drift of $V_{IO}/300 \mu\text{V}/^\circ\text{C}$ (e.g., if V_{IO} is adjusted to 300 μV , the change in temperature coefficient will be 1 $\mu\text{V}/^\circ\text{C}$).

The adjustment range with a 10-k Ω potentiometer is approximately $\pm 2.5 \text{ mV}$. If a smaller adjustment range is needed, the sensitivity and resolution of the nulling can be improved by using a smaller potentiometer in conjunction with fixed resistors. The example in Figure 36 has an approximate null range of $\pm 200 \mu\text{V}$.

offset voltage and drift

Unless proper care is exercised, thermocouple effects at the contacts to the input terminals, caused by temperature gradients across dissimilar metals, can exceed the inherent temperature coefficient of the amplifier. Air currents should be minimized, package leads should be short, input leads should be close together, and input leads should be at the same temperature.

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Operational Amplifiers

TYPICAL APPLICATION DATA

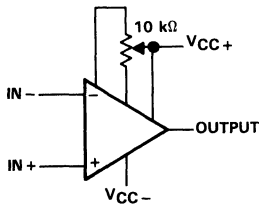


FIGURE 35. STANDARD ADJUSTMENT

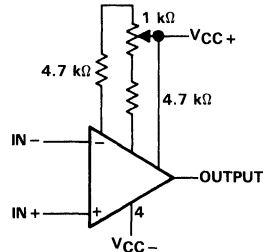


FIGURE 36. IMPROVED SENSITIVITY ADJUSTMENT

2

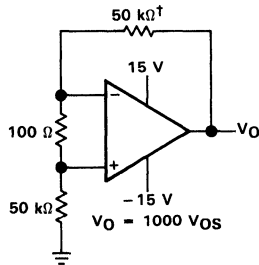
Operational Amplifiers

The circuit shown in Figure 37 can be used to measure offset voltage. In addition, with the supply voltages increased to ± 20 V, it can be used as the burn-in configuration for the LT1007 and LT1037.

When $R_F \leq 100 \Omega$ and the input is driven with a fast large-signal pulse (> 1 V), the output waveform will be as shown in Figure 38.

During the fast-feedthrough-like portion of the output, the input protection diodes effectively short the output to the input and a current, limited only by the output short-circuit protection, is drawn by the signal generator. When $R_F \geq 500 \Omega$, the output is capable of handling the current requirements ($I_L \leq 20$ mA at 10 V), the amplifier stays in its active mode, and a smooth transition occurs.

When R_F is > 2 k Ω , a pole will be created with R_F and the amplifier's input capacitance, creating additional phase shift and reducing the phase margin. A small capacitor (20 pF to 50 pF) in parallel with R_F will eliminate this problem.



†Resistors must have low thermoelectric potential

FIGURE 37. TEST CIRCUIT FOR OFFSET VOLTAGE AND OFFSET VOLTAGE DRIFT WITH TEMPERATURE

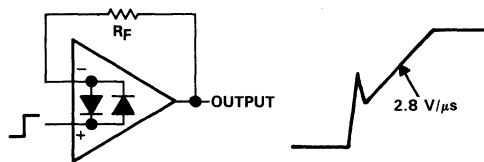


FIGURE 38. PULSE OPERATION

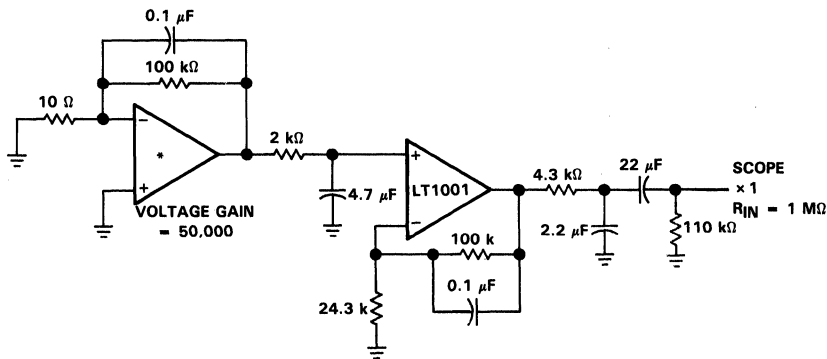
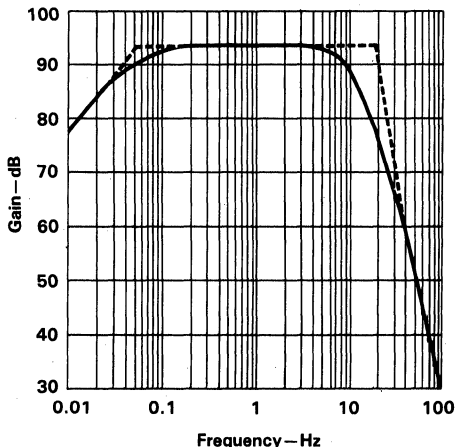
TYPICAL APPLICATION DATA

noise testing

Figure 39 shows a test circuit for 0.1-Hz to 10-Hz peak-to-peak noise measurement of the LT1007 and LT1037. The frequency response of this noise tester indicates that the 0.1 Hz corner is defined by only one zero. Because the time limit acts as an additional zero to eliminate noise contributions from the frequency band below 0.1 Hz, the test time to measure 0.1-Hz to 10-Hz noise should not exceed 10 seconds.

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Operational Amplifiers

0.1 Hz to 10 Hz p-p NOISE TESTER
 FREQUENCY RESPONSE



*Device under test
 NOTE: All capacitor values are for non-polarized capacitors only.

FIGURE 39. 0.1-Hz TO 10-Hz NOISE TEST CIRCUIT

LT1007, LT1007A, LT1037, LT1037A LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

Special test precautions are required to measure the typical 60-nV peak-to-peak noise performance of the LT1007 and LT1037:

1. The device should be warmed up for at least five minutes. As the operational amplifier warms up, the offset voltage typically changes $3\ \mu\text{V}$, due to the chip temperature increasing 10°C to 20°C from the moment the power supplies are turned on. In the 10-second measurement interval, these temperature-induced effects can easily exceed tens of nanovolts.
2. The device must be well shielded from air currents to eliminate thermoelectric effects. In excess of a few nanovolts, thermoelectric effects would invalidate the measurements.
3. Sudden motion in the vicinity of the device can produce a feedthrough effect that increases observed noise.

When measuring noise on a large number of units, a noise-voltage density test is recommended. A 10-Hz noise-voltage density measurement will correlate well with a 0.1-Hz to 10-Hz peak-to-peak noise reading since both results are determined by the white noise and the location of the $1/f$ corner frequency.

Figure 40 shows a circuit that measures noise current and presents the formula for calculating noise current.

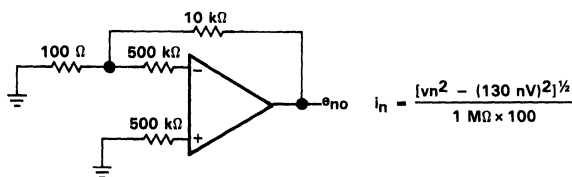


FIGURE 40. NOISE TEST CIRCUIT

The LT1007 and LT1037 achieve low noise, in part, by operating the input stage at $120\ \mu\text{A}$ versus the typical $10\ \mu\text{A}$ for most other operational amplifiers. Voltage noise is directly proportional to the square root of the stage current; therefore, the LT1007 and LT1037 noise current is relatively high. At low frequencies, the low $1/f$ current-noise corner frequency ($\approx 120\ \text{Hz}$) minimizes noise current to some extent.

In most practical applications, however, noise current will not limit system performance; this is illustrated in Figure 27, where:

$$\text{total noise} = [(\text{noise voltage})^2 (\text{noise current} \times R_S)^2 + (\text{resistor noise})^2]^{1/2}$$

Three regions can be identified as a function of source resistance:

- | | |
|--|---|
| (i) $R_S \leq 400\ \Omega$ | Voltage noise dominates in region (i) |
| (ii) $R_S = 400\ \Omega$ to $50\ \text{k}\Omega$ at $1\ \text{kHz}$
$R_S = 400\ \Omega$ to $8\ \text{k}\Omega$ at $10\ \text{Hz}$ | Resistor noise dominates in region (ii) |
| (iii) $R_S > 50\ \text{k}\Omega$ at $1\ \text{kHz}$
$R_S > 8\ \text{k}\Omega$ at $10\ \text{Hz}$ | Current noise dominates in region (iii) |

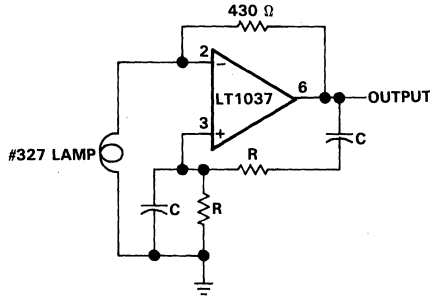
The LT1007 and LT1037 should not be used in region (iii) where total system noise is at least six times higher than the noise voltage of the operational amplifier (i.e., the low-voltage noise specification is completely wasted).

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Operational Amplifiers

TYPICAL APPLICATIONS

The sine wave generator application shown below, utilizes the low-noise and low-distortion characteristics of the LT1037.



$f = \frac{1}{2\pi RC}$
 $R = 1591.5\Omega \pm 0.1\%$
 $C = 0.1 \mu F \pm 0.1\%$

TOTAL HARMONIC DISTORTION $\leq 0.0025\%$
 NOISE $\leq 0.001\%$
 AMPLITUDE = $\pm 8 V$
 OUTPUT FREQUENCY = 1.000 kHz FOR VALUES GIVEN $\pm 0.4\%$

FIGURE 41. ULTRA-PURE 1-kHz SINE-WAVE GENERATOR

EQUIVALENT INPUT NOISE VOLTAGE
 OVER A 10-SECOND PERIOD

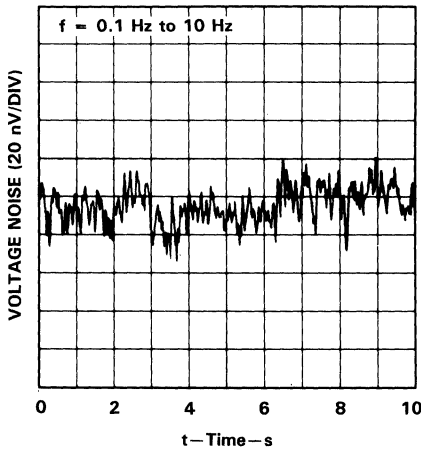
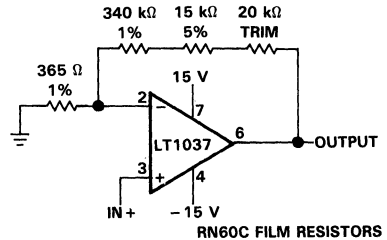


FIGURE 42



The high gain and wide bandwidth of the LT1037 and (LT1007) is useful in low-frequency high-closed-loop-gain amplifier applications. A typical precision Op Amp may have an open loop gain of one million with 500 kHz bandwidth. As the gain error plot shows, this device is capable of 0.1% amplifying accuracy up to 0.3 Hz only. Even instrumentation range signals can vary at a faster rate. The LT1037's "gain precision—bandwidth product" is 200 times higher, as shown.

FIGURE 43. GAIN 1000 AMPLIFIER WITH
 0.01% ACCURACY, DC to 5 Hz

LT1007, LT1007A, LT1037, LT1037A
 LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL APPLICATIONS

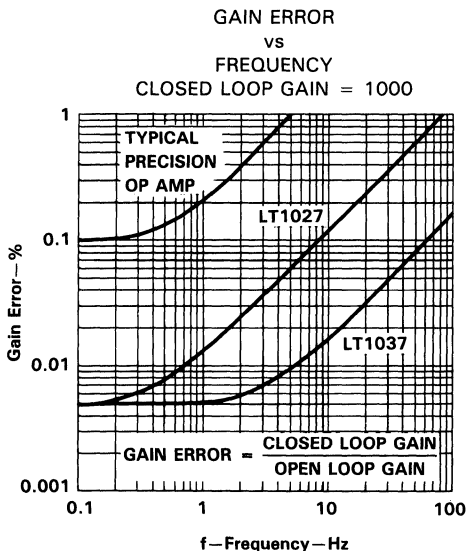
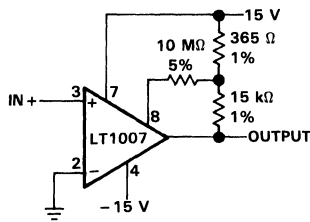


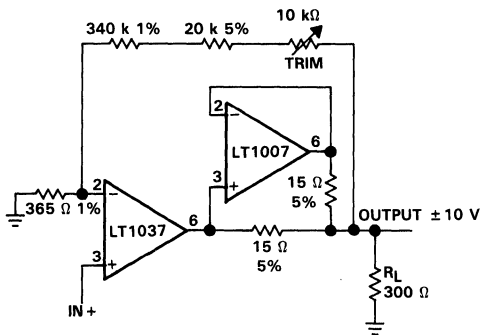
FIGURE 44



Positive feedback to one of the nulling terminals creates approximately 5 μV of hysteresis. Output can sink 16 mA.

Input offset voltage is typically changed less than 5 μV due to the feedback.

FIGURE 45. MICROVOLT COMPARATOR WITH HYSTERESIS



The addition of the LT1007 doubles the amplifier's output drive to ± 33 mA. Gain accuracy is 0.02%, slightly degraded compared to above because of self heating of the LT1037 under load.

FIGURE 46. PRECISION AMPLIFIER DRIVES 300-Ω LOAD TO ± 10 V

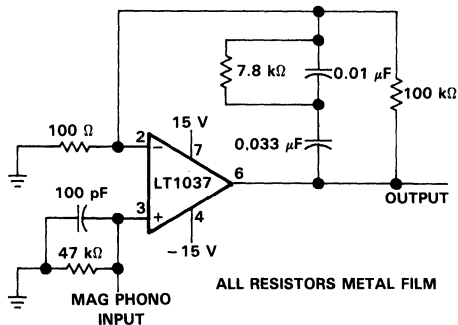


FIGURE 47. PHONO PREAMPLIFIER

2
 Operational Amplifiers

TYPICAL APPLICATIONS

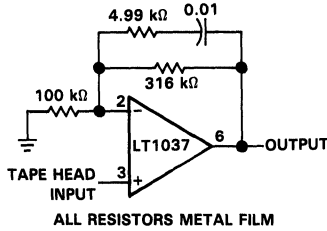


FIGURE 48. TAPE HEAD AMPLIFIER

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Operational Amplifiers

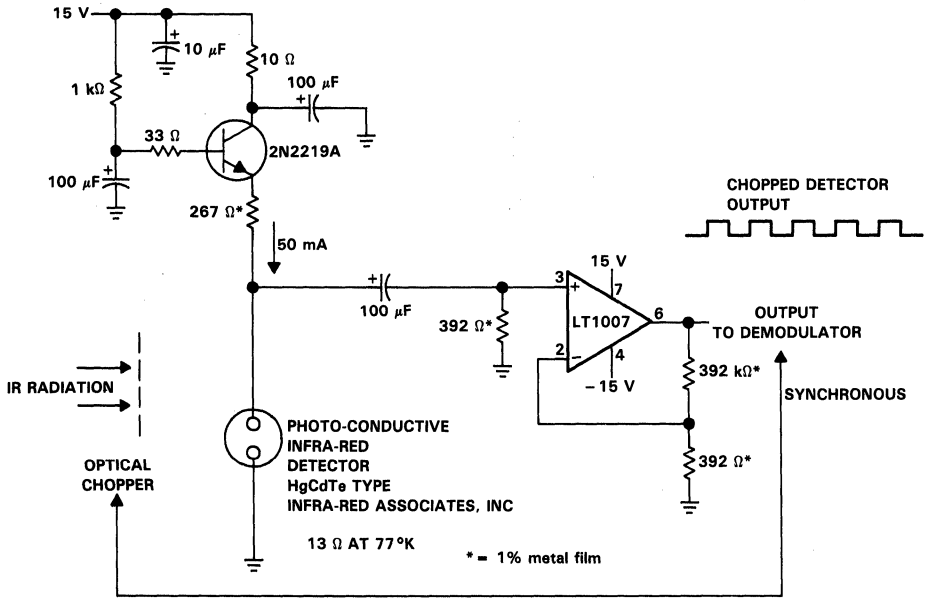


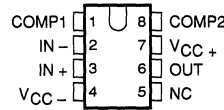
FIGURE 49. INFRA-RED DETECTOR PREAMPLIFIER

LT1008M, LT1008C PICOAMP INPUT CURRENT, MICROVOLT OFFSET LOW-NOISE OPERATIONAL AMPLIFIERS

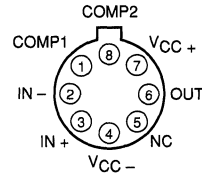
D3233, MAY 1988 – REVISED FEBRUARY 1989

- **Input Bias Current** . . . ± 30 pA Typ,
 ± 100 pA Max at 25°C
- **Input Offset Voltage** . . . 30 μ V Typ,
120 μ V Max at 25°C
- **Offset Voltage Temperature Coefficient** . . .
1.5 μ V/°C Max
- **Low Peak-to-Peak Noise Voltage** at
0.1 Hz to 10 Hz . . . 0.5 μ V
- **Low Supply Current** . . . 380 μ A Typ,
600 μ A Max at 25°C
- **Supply Voltage Rejection Ratio** . . . 114 dB
Min at 25°C
- **Common-Mode Rejection Ratio** . . . 114 dB
Min at 25°C
- **High Voltage Amplification with 5-mA Load
Current**
- **Applications:**
Precision Instrumentation
Charge Integrators
Wide-Dynamic-Range Logarithmic
Amplifiers
Light Meters
Low-Frequency Active Filters
Standard Cell Buffers
Thermocouple Amplifiers

**JG OR P PACKAGE
(TOP VIEW)**

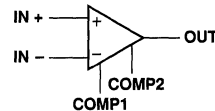


**L PACKAGE
(TOP VIEW)**



NC – No internal connection
Pin 4 (L Package) is in electrical contact
with the case.

symbol



description

The LT1008 is a precision operational amplifier that can be used in practically all precision applications. The LT1008 offers picoampere bias currents (maintained over the full temperature range), microvolt offset voltage, low offset voltage temperature coefficient and long-term drift, low voltage and current noise, and low power dissipation. Additionally, the LT1008's precision specifications include high common-mode and supply voltage rejection ratios. The LT1008 can deliver a 5-mA load current with high voltage amplification.

The LT1008 is externally compensated with a single capacitor to add flexibility in shaping the frequency response of the amplifier. The LT1008 is a pin-for-pin replacement for the LM108 series.

The LT1008M is characterized for operation over the full military temperature range of -55°C to 125°C . The LT1008C is characterized for operation from 0°C to 70°C .

AVAILABLE OPTIONS

T _A	PACKAGE		
	CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)
0°C to 70°C	LT1008CJG	LT1008CL	LT1008CP
-55°C to 125°C	LT1008MJG	LT1008ML	LT1008MP

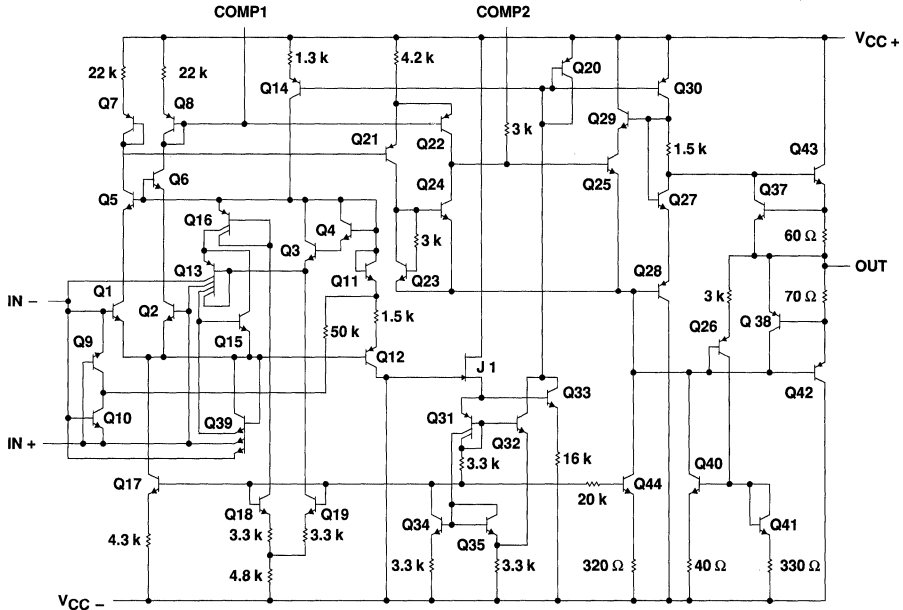
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Operational Amplifiers

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

LT1008M, LT1008C PICOAMP INPUT CURRENT, MICROVOLT OFFSET LOW-NOISE OPERATIONAL AMPLIFIERS

schematic

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Operational Amplifiers



All resistor values shown are nominal and in ohms.

absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	20 V
Supply voltage, V_{CC-}	- 20 V
Input voltage range, V_I	± 20 V
Differential input current (see Note 2)	± 10 mA
Duration of output short-circuit at (or below) 25°C (see Note 3)	unlimited
Operating free-air temperature, T_A : LT1008M	- 55°C to 125°C
LT1008C	0°C to 70°C
Storage temperature range	- 65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: JG or L package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: P package	260°C

- NOTES: 1. All voltage values are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential input voltages greater than 1 V will cause excessive current to flow through the input protection diodes unless current-limiting resistors are used.
 3. The output may be shorted to either supply.

recommended operating conditions

	LT1008M		LT1008C		UNIT
	MIN	NOM MAX	MIN	NOM MAX	
Supply voltage, V_{CC}	± 20		± 20		V
Common-mode input voltage, V_{IC}	$V_{CC\pm} \pm 15$ V		± 13.5	± 13.5	V
Operating free-air temperature, T_A	- 55	125	0	70	°C

LT1008M, LT1008C PICOAMP INPUT CURRENT, MICROVOLT OFFSET LOW-NOISE OPERATIONAL AMPLIFIERS

electrical characteristics, $V_{CC\pm} = \pm 15V$, $V_{IC} = 0$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A^\dagger	LT1008M			LT1008C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{CC\pm} = \pm 15V$, $V_{IC} = 0$	25°C		30	120		30	120	μV
		Full range			250			180	
	$V_{CC\pm} = \pm 15V$, $V_{IC} = \pm 13.5V$	25°C		40	180		40	180	
		Full range			320			250	
	$V_{CC\pm} = \pm 2V$ to $\pm 20V$	25°C		40	180		40	180	
$V_{CC\pm} = \pm 2.5V$ to $\pm 20V$	Full range			320			250		
α_{VIO} Average temperature coefficient of input offset voltage		Full range		0.2	1.5		0.2	1.5	$\mu V/^\circ C$
Long-term drift of input offset voltage		25°C		0.3			0.3		$\mu V/mo$
I_{IO} Input offset current	$V_{CC\pm} = \pm 15V$, $V_{IC} = 0$	25°C		30	100		30	100	pA
		Full range			250			180	
	$V_{CC\pm} = \pm 15V$, $V_{IC} = \pm 13.5V$	25°C		40	150		40	150	
		Full range			350			250	
	$V_{CC\pm} = \pm 2V$ to $\pm 20V$	25°C		40	150		40	150	
$V_{CC\pm} = \pm 2.5V$ to $\pm 20V$	Full range			350			250		
α_{IIO} Average temperature coefficient of input offset current		Full range		0.4	2.5		0.4	2.5	$pA/^\circ C$
I_{IB} Input bias current	$V_{CC\pm} = \pm 15V$, $V_{IC} = 0$	25°C		± 30	± 100		± 30	± 100	pA
		Full range			± 600			± 180	
	$V_{CC\pm} = \pm 15V$, $V_{IC} = \pm 13.5V$	25°C		± 40	± 150		± 40	± 150	
		Full range			± 800			± 250	
	$V_{CC\pm} = \pm 2V$ to $\pm 20V$	25°C		± 40	± 150		± 40	± 150	
$V_{CC\pm} = \pm 2.5V$ to $\pm 20V$	Full range			± 800			± 250		
α_{IIB} Average temperature coefficient of input bias current		Full range		0.6	6		0.4	2.5	$pA/^\circ C$
V_{ICR} Common-mode input voltage range		25°C		± 13.5	± 14		± 13.5	± 14	V
		Full range		± 13.5			± 13.5		
V_{OM} Maximum peak output voltage swing	$R_L = 10k\Omega$	25°C		± 13	± 14		± 13	± 14	V
		Full range		± 13			± 13		
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 12V$, $R_L \geq 10k\Omega$	25°C		200	2000		200	2000	V/mV
		Full range		100			150		
		$V_O = \pm 10V$, $R_L \geq 2k\Omega$	25°C		120	600		120	
CMRR Common-mode rejection ratio	$V_{IC} = \pm 13.5V$	25°C		114	132		114	132	dB
		Full range		108			110		
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{CC\pm} / \Delta V_{IO}$)	$V_{CC\pm} = \pm 2V$ to $\pm 20V$	25°C		114	132		114	132	dB
	$V_{CC\pm} = \pm 2.5V$ to $\pm 20V$	Full range		108			110		
I_{CC} Supply current	$V_{CC\pm} = \pm 15V$, $V_{IC} = \pm 13.5V$	25°C		380	600		380	600	μA
		Full range			800			800	
	$V_{CC\pm} = \pm 2V$ to $\pm 20V$	25°C		380	600		380	600	
	$V_{CC\pm} = \pm 15V$, $V_{IC} = 0$	Full range			800			800	

† Full range is $-55^\circ C$ to $125^\circ C$ for the LT1008M and $0^\circ C$ to $70^\circ C$ for the LT1008C.

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Operational Amplifiers

LT1008M, LT1008C
PICOAMP INPUT CURRENT, MICROVOLT OFFSET
LOW-NOISE OPERATIONAL AMPLIFIERS

operating characteristics, $V_{CC} \pm = \pm 15 \text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$C_f = 30 \text{ pF}$, See Figure 29(a)	0.1	0.2		$\text{V}/\mu\text{s}$
V_{NPP}	Peak-to-peak equivalent input noise voltage	$f = 0.1 \text{ Hz to } 10 \text{ Hz}$		0.5		μV
V_n	Equivalent input noise voltage	$f = 10 \text{ Hz}$		17	30	$\text{nV}/\sqrt{\text{Hz}}$
		$f = 1 \text{ kHz}$		14	22	
I_n	Equivalent input noise current	$f = 10 \text{ Hz}$		20		$\text{fA}/\sqrt{\text{Hz}}$

TYPICAL CHARACTERISTICS

table of graphs

			FIGURE
V_{IO}	Input offset voltage	vs Temperature	1
		vs Source resistance	5
ΔV_{IO}	Change in input offset voltage	vs Time – minutes	2
		vs Time – months	3
α_{VIO}	Temperature coefficient of input offset voltage	vs Source resistance	6
I_{IB}	Input bias current	vs Common-mode input voltage	7
		vs Temperature	8
A_{VD}	Differential voltage amplification	vs Load resistance	9
		vs Frequency	10, 11, 12
CMRR	Common-mode rejection ratio	vs Frequency	13
k_{SVR}	Supply-voltage rejection ratio	vs Frequency	14
I_{OS}	Short-circuit output current	vs Time	15
I_{CC}	Supply current	vs Supply voltage	4
SR	Slew rate	vs Compensation capacitance	16
V_{NPP}	Peak-to-peak equivalent input noise voltage	vs Time	17
V_n, I_n	Equivalent input noise voltage and equivalent input noise current	vs Frequency	18
		Total equivalent input noise voltage	19
	Phase shift	vs Frequency	11, 12
Pulse response		Small-signal	20, 21, 22
		Large-signal	23, 24

2

Operational Amplifiers

LT1008M, LT1008C

PICOAMP INPUT CURRENT, MICROVOLT OFFSET LOW-NOISE OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

INPUT OFFSET VOLTAGE
OF REPRESENTATIVE UNITS
VS
FREE-AIR TEMPERATURE

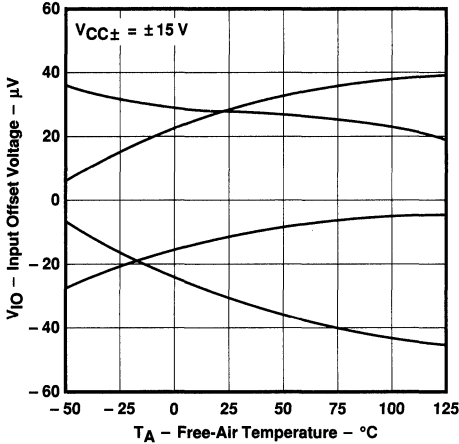


FIGURE 1

WARM-UP CHANGE IN
INPUT OFFSET VOLTAGE
VS
ELAPSED TIME

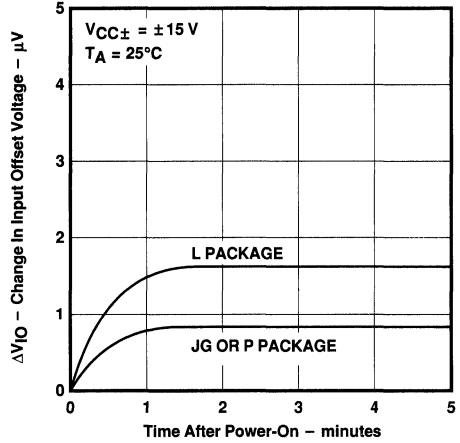


FIGURE 2

LONG-TERM DRIFT OF
INPUT OFFSET VOLTAGE
OF REPRESENTATIVE UNITS

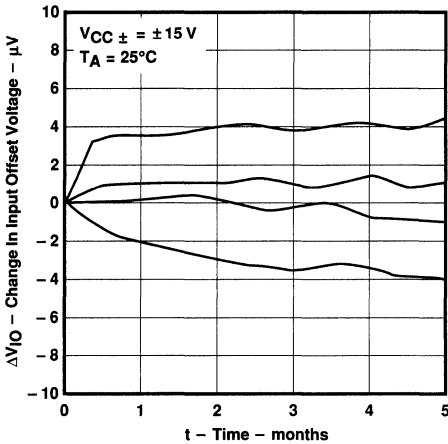


FIGURE 3

SUPPLY CURRENT
VS
SUPPLY VOLTAGE

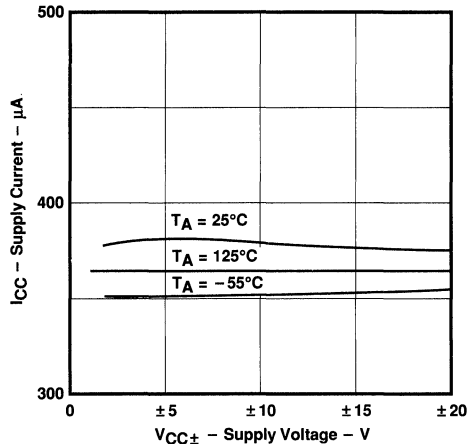


FIGURE 4

2
Operational Amplifiers

†Data for temperatures below 0°C and above 70°C are applicable to the LT1008M only.

LT1008M, LT1008C
PICOAMP INPUT CURRENT, MICROVOLT OFFSET
LOW-NOISE OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

INPUT OFFSET VOLTAGE
VS
SOURCE RESISTANCE

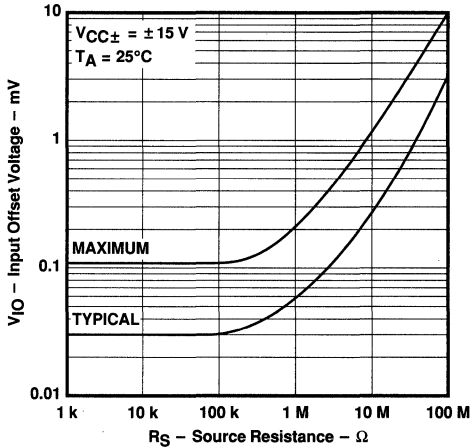


FIGURE 5

AVERAGE TEMPERATURE COEFFICIENT
OF INPUT OFFSET VOLTAGE
VS
SOURCE RESISTANCE

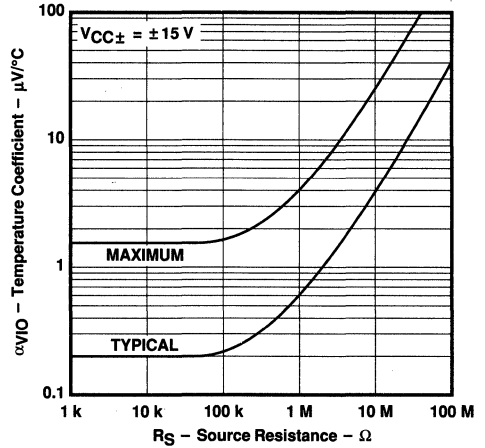


FIGURE 6

INPUT BIAS CURRENT
VS
COMMON-MODE INPUT VOLTAGE

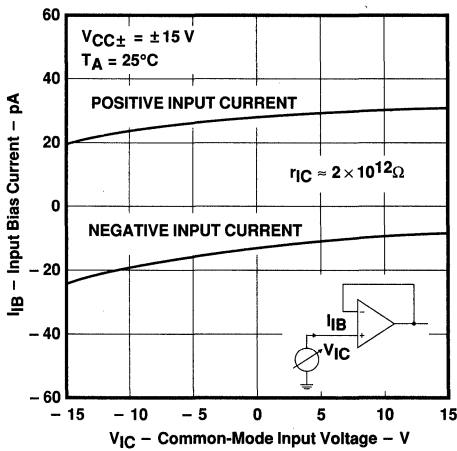


FIGURE 7

INPUT BIAS CURRENT
VS
FREE-AIR TEMPERATURE

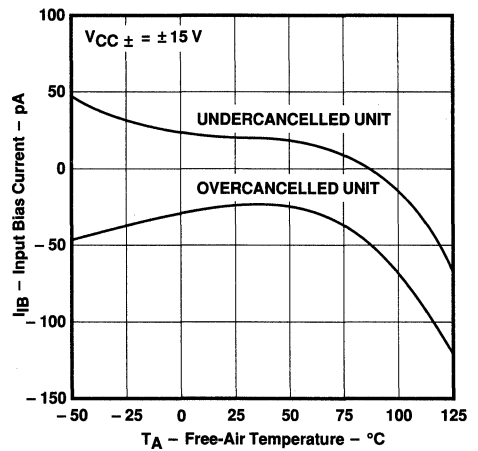


FIGURE 8

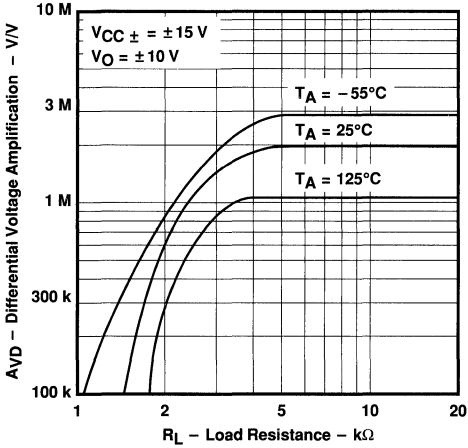
†Data for temperatures below $0^\circ C$ and above $70^\circ C$ are applicable to the LT1008M only.

LT1008M, LT1008C

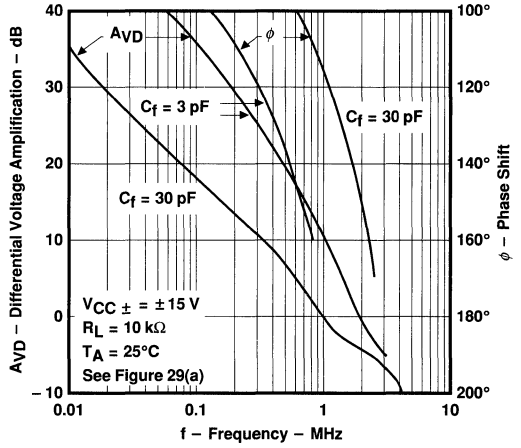
PICOAMP INPUT CURRENT, MICROVOLT OFFSET LOW-NOISE OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

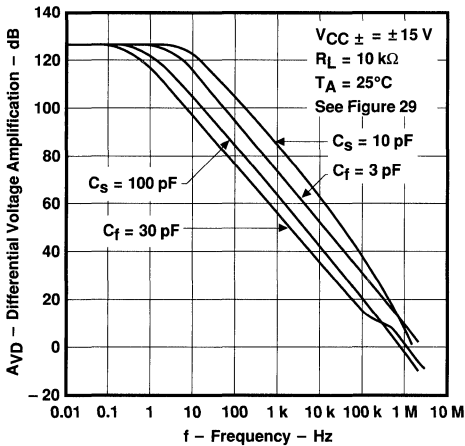
LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
VS
LOAD RESISTANCE



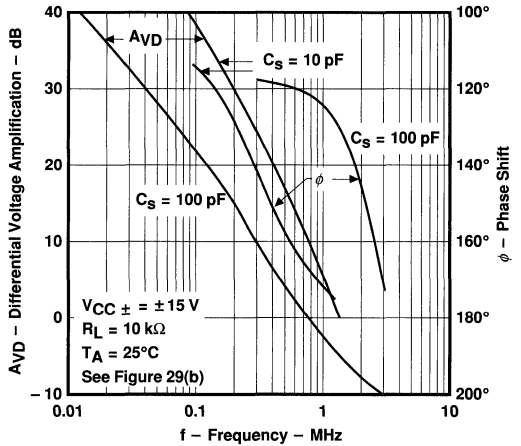
LARGE-SIGNAL DIFFERENTIAL
VOLTAGE AMPLIFICATION and PHASE SHIFT
VS
FREQUENCY



LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
VS
FREQUENCY



LARGE-SIGNAL DIFFERENTIAL
VOLTAGE AMPLIFICATION and PHASE SHIFT
VS
FREQUENCY



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Operational Amplifiers

†Data for temperatures below 0°C and above 70°C are applicable to the LT1008M only.

LT1008M, LT1008C
PICOAMP INPUT CURRENT, MICROVOLT OFFSET
LOW-NOISE OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

COMMON-MODE REJECTION RATIO
VS
FREQUENCY

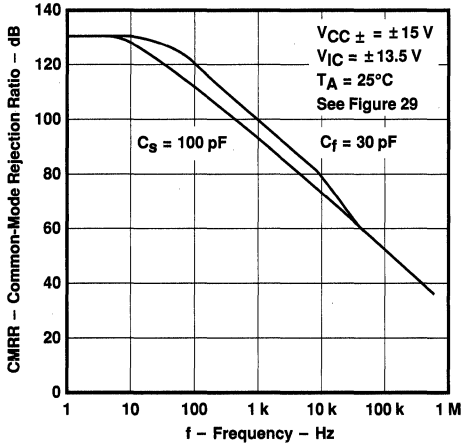


FIGURE 13

SUPPLY-VOLTAGE REJECTION RATIO
VS
FREQUENCY

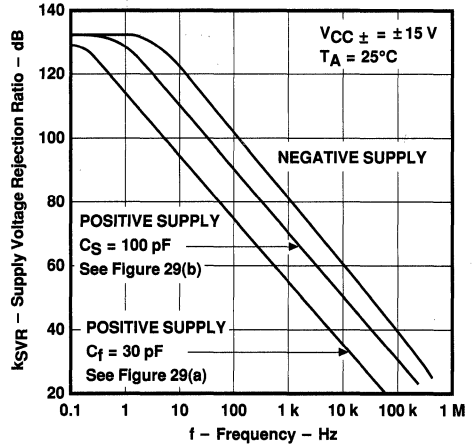


FIGURE 14

SHORT-CIRCUIT OUTPUT CURRENT
VS
ELAPSED TIME

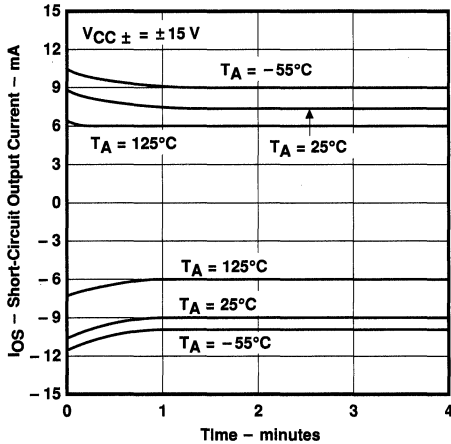


FIGURE 15

SLEW RATE
VS
COMPENSATION CAPACITANCE

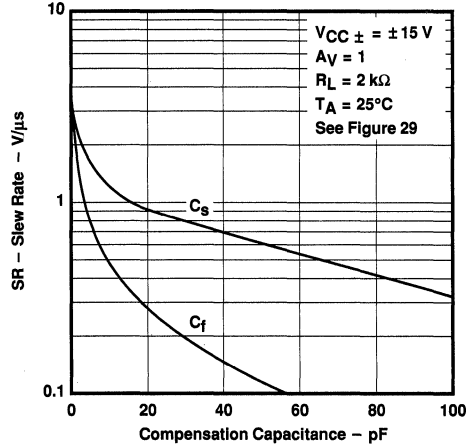


FIGURE 16

†Data for temperatures below 0°C and above 70°C are applicable to the LT1008M only.

LT1008M, LT1008C

PICOAMP INPUT CURRENT, MICROVOLT OFFSET LOW-NOISE OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

OUTPUT NOISE VOLTAGE
OVER A 10-SECOND PERIOD

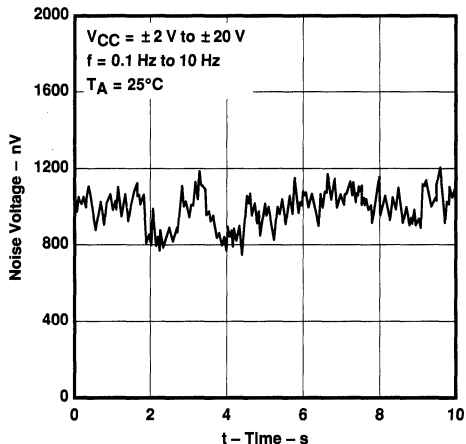


FIGURE 17

EQUIVALENT INPUT NOISE VOLTAGE
and EQUIVALENT INPUT NOISE CURRENT
vs
FREQUENCY

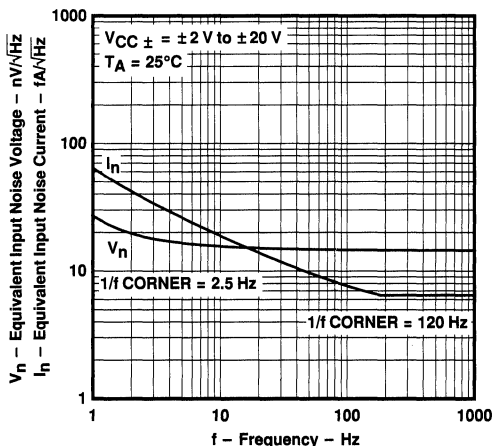


FIGURE 18

TOTAL EQUIVALENT INPUT NOISE VOLTAGE
vs
SOURCE RESISTANCE

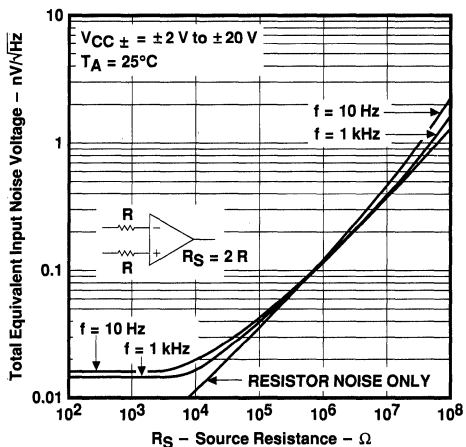


FIGURE 19

VOLTAGE-FOLLOWER
SMALL-SIGNAL
PULSE RESPONSE

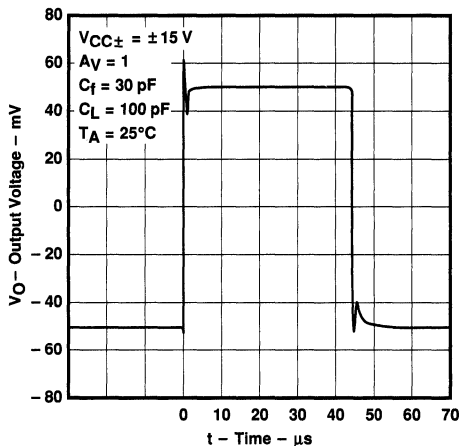


FIGURE 20

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Operational Amplifiers

LT1008M, LT1008C
PICOAMP INPUT CURRENT, MICROVOLT OFFSET
LOW-NOISE OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

**VOLTAGE-FOLLOWER
 SMALL-SIGNAL
 PULSE RESPONSE**

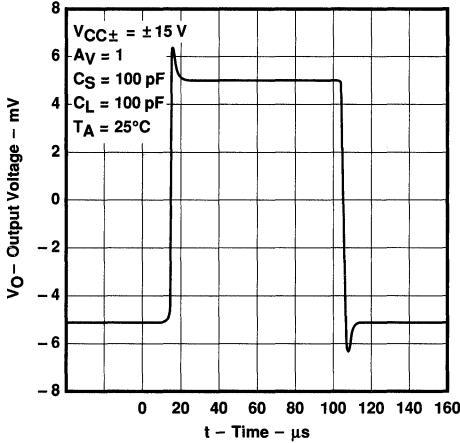


FIGURE 21

**VOLTAGE-FOLLOWER
 SMALL-SIGNAL
 PULSE RESPONSE**

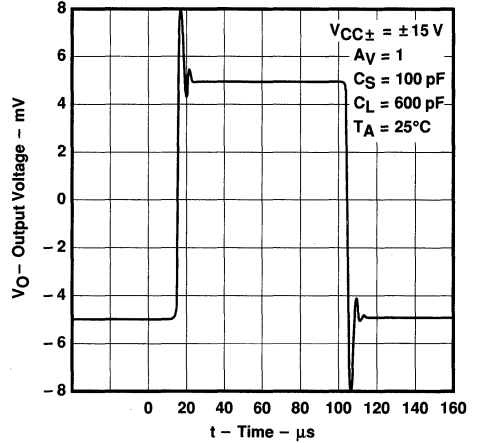


FIGURE 22

**VOLTAGE-FOLLOWER
 LARGE-SIGNAL
 PULSE RESPONSE**

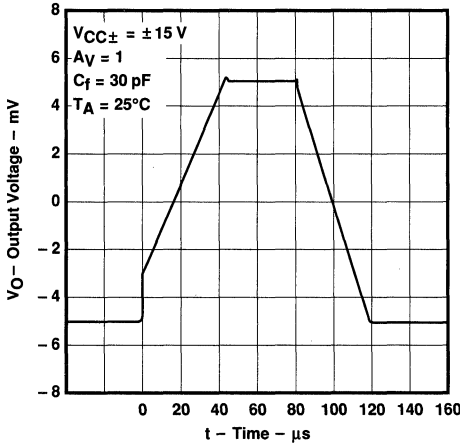


FIGURE 23

**VOLTAGE-FOLLOWER
 LARGE-SIGNAL
 PULSE RESPONSE**

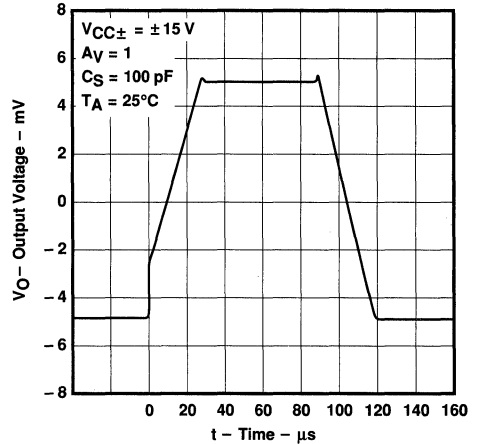


FIGURE 24

2

Operational Amplifiers

TYPICAL APPLICATION DATA

achieving picoampere, microvolt performance

Proper care should be exercised to realize the picoampere, microvolt accuracy of the LT1008. Because leakage currents in external circuitry can significantly degrade performance, high-quality insulation should be used (e. g., Teflon, Kel-F). All insulating surfaces should be cleaned to remove fluxes and other residues. Surface coating may be necessary to provide a moisture barrier in high-humidity environments.

Board leakage can be minimized by encircling the input circuitry with a guard ring operated at a potential close to that of the inputs (see Figure 25). In inverting configurations, the guard ring should be tied to ground; in noninverting configurations, the guard ring should be tied to the inverting input (pin 2). Both sides of the printed circuit board should be guarded. Bulk leakage reduction depends on the guard ring width. Nanoampere-level leakage into the compensation terminals can affect input offset voltage and its temperature coefficient (see Figure 26).

Microvolt-level error voltages can also be generated in the external circuitry. Thermocouple effects, caused by temperature gradients across dissimilar metals at the contacts to the input terminals, can exceed the inherent temperature coefficient of the amplifier. Air currents over device leads should be minimized, package leads should be short, and the two input leads should be as close together as possible and maintained at the same temperature. The LT1008 is specified over a wide range of supply voltages from ± 2 V to ± 18 V. Operation with lower supplies (down to ± 1 V) is possible with two Ni-Cad batteries.

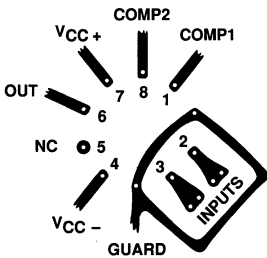
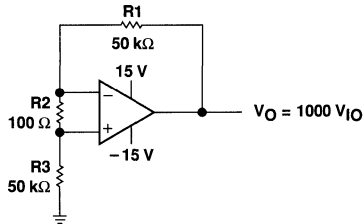


FIGURE 25. GUARD RING



- NOTES: A. Resistors must have low thermoelectric potential.
 B. This circuit is also used as the burn-in configuration for the LT1008, with supply voltages increased to ± 20 V, $R1 = R3 = 20$ k Ω , $R2 = 200$ Ω , and $A_V = 100$.

FIGURE 26. TEST CIRCUIT FOR V_{IO} AND α_{VIO}

noise testing

The peak-to-peak equivalent input noise voltage of the LT1008 is measured in the test circuit shown in Figure 27. The frequency response of this noise tester indicates that the 0.1-Hz to 10-Hz noise should not exceed 10 seconds, as this time limit acts as an additional zero to eliminate noise contributions from the frequency band below 0.1 Hz.

An input noise voltage test is recommended when measuring noise in a large number of units. A 10-Hz input noise voltage measurement correlates well with a 0.1-Hz peak-to-peak noise reading because both results are determined by the white noise and the location of the $1/f$ corner frequency.

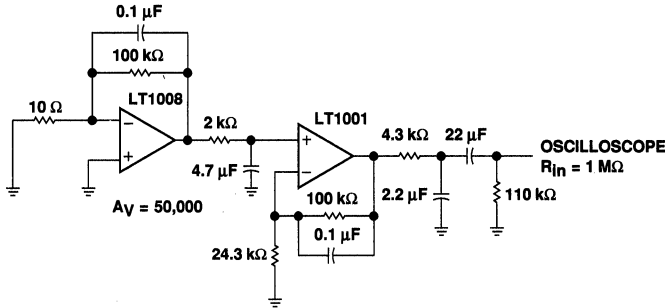
Current noise is measured by the current shown in Figure 28 and calculated by the following formula in which the noise of the source resistors is subtracted:

$$I_n = \frac{[V_{no}^2 - (820 \text{ nV})^2]^{1/2}}{40 \text{ M}\Omega \times 100}$$

LT1008M, LT1008C
PICOAMP INPUT CURRENT, MICROVOLT OFFSET
LOW-NOISE OPERATIONAL AMPLIFIERS

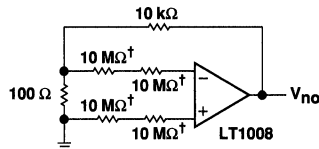
TYPICAL APPLICATION DATA

noise testing (continued)



NOTE A : All capacitor values are for nonpolarized capacitors only.

FIGURE 27. 0.1-Hz TO 10-Hz PEAK-TO-PEAK NOISE VOLTAGE TEST CIRCUIT

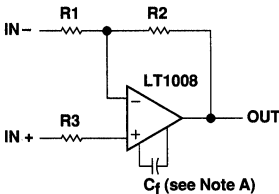


† Metal film.

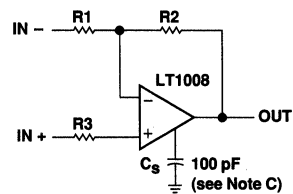
FIGURE 28. NOISE CURRENT TEST CIRCUIT

frequency compensation

The LT1008 is externally frequency compensated with a single capacitor. The two compensation circuits shown in Figure 29 are identical to the frequency compensation circuits for the LM108A series. Therefore, the LT1008 operational amplifiers can be inserted directly into LM108A or LM308A sockets, with similar ac and upgraded dc performance.



(a) STANDARD COMPENSATION



(b) ALTERNATE COMPENSATION (see Note B)

- NOTES: A. $C_f \geq (R1 \times C_0) / (R1 + R2)$, $C_0 = 30$ pF. Bandwidth and slew rate are proportional to $1/C_f$.
 B. This circuit improves the supply voltage rejection ratio by a factor of 5.
 C. Bandwidth and slew rate are proportional to $1/C_s$.
 D. For $(R2/R1) > 200$, no external frequency compensation is necessary.

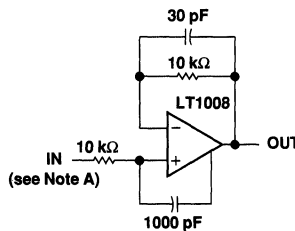
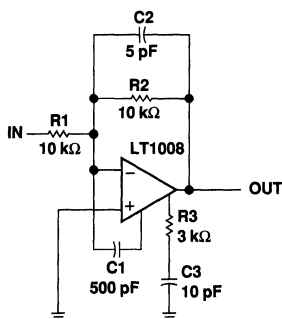
FIGURE 29. FREQUENCY COMPENSATION CIRCUITS (see Note D)

TYPICAL APPLICATION DATA

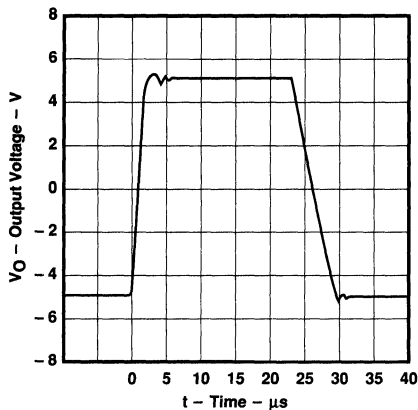
frequency compensation (continued)

External frequency compensation provides additional flexibility in shaping the frequency response of the amplifier. For example, for a voltage gain of 10 and $C_f = 3 \text{ pF}$, a gain-bandwidth product of 5 MHz and slew rate of $1.2 \text{ V}/\mu\text{s}$ can be realized. For closed-loop gains greater than 200, no external compensation is necessary, and the slew rate increases to $4 \text{ V}/\mu\text{s}$. The LT1008 can also be overcompensated (e.g., $C_f > 30 \text{ pF}$ or $C_S > 100 \text{ pF}$) to improve capacitive-load-handling capability or to narrow noise bandwidth. In applications in which the feedback loop around the amplifier has gain, overcompensation can stabilize the circuit with a single capacitor.

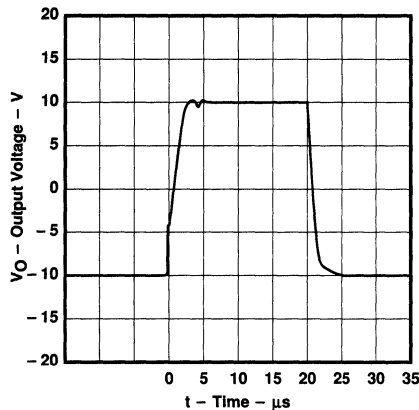
The availability of the compensation terminals permits the use of feed-forward frequency compensation to enhance slew rate in low closed-loop-gain configurations (see Figure 30). The inverter slew rate is increased to $1.4 \text{ V}/\mu\text{s}$. The voltage-follower feed-forward scheme bypasses the amplifier's gain stages and slews at nearly $10 \text{ V}/\mu\text{s}$.



NOTE A: $R_S \leq 15 \text{ k}\Omega$ for stability.



(a) INVERTER FEED-FORWARD



(b) VOLTAGE-FOLLOWER FEED-FORWARD

FIGURE 30. FREQUENCY COMPENSATION CIRCUITS
and VOLTAGE-FOLLOWER PULSE RESPONSES

LT1008M, LT1008C
PICOAMP INPUT CURRENT, MICROVOLT OFFSET
LOW-NOISE OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

other considerations

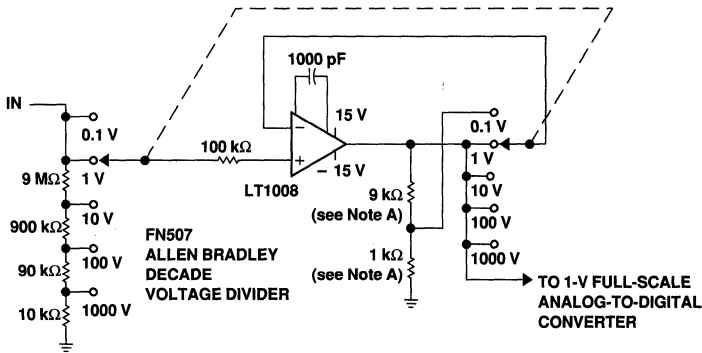
The inputs of the LT1008 are protected by back-to-back diodes. Current-limiting resistors are not used because the leakage of these resistors would prevent the realization of picoampere-level bias currents at elevated temperatures. In the voltage-follower configuration, when the input is driven by a fast, large-signal pulse ($> 1\text{ V}$), the input protection diodes effectively short the output to the input during slewing, and a current, limited only by the output short-circuit protection, flows through the diodes.

The use of a feedback resistor, as shown in the voltage-follower feed-forward diagram, is recommended because this resistor keeps the current below the short-circuit limit, resulting in faster recovery and settling of the output.

2

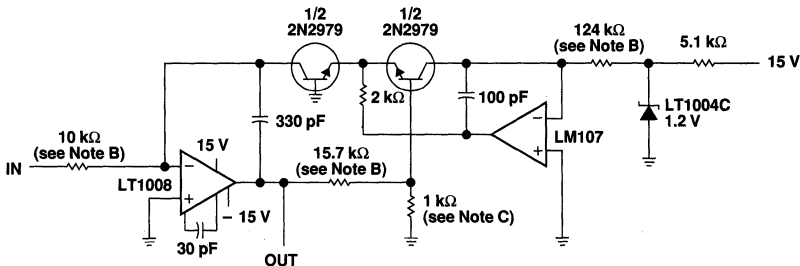
Operational Amplifiers

typical applications



NOTES: A. Ratio match $\pm 0.01\%$.
 B. This application requires low bias current, low offset voltage and offset voltage temperature coefficient, low noise, and low long-term offset voltage drift.

FIGURE 31. INPUT AMPLIFIER FOR 4 1/2-DIGIT VOLTMETER

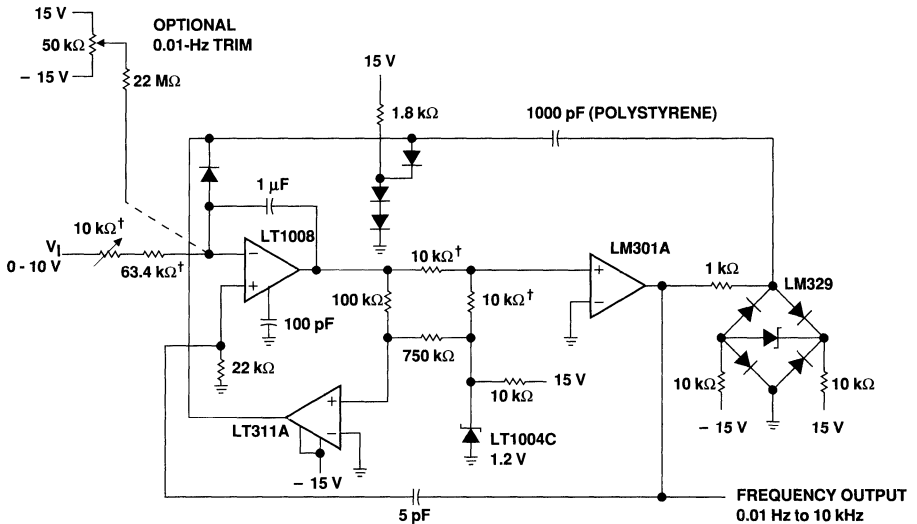


NOTES: A. The low bias current and offset voltage of the LT1008 allow 4 1/2 decades of voltage input logging.
 B. 1% film resistor.
 C. Tel. Labs, Type Q81.

FIGURE 32. LOGARITHMIC AMPLIFIER

LT1008M, LT1008C PICOAMP INPUT CURRENT, MICROVOLT OFFSET LOW-NOISE OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA



†1% metal film resistor

NOTES: A. The LT1008 integrator extends the low frequency range. The total dynamic range is 0.01 Hz to 10 kHz (or 120 dB) with 0.01% linearity.

B. All diodes 1N4148.

FIGURE 33. EXTENDED RANGE CHARGE PUMP VOLTAGE-TO-FREQUENCY CONVERTER

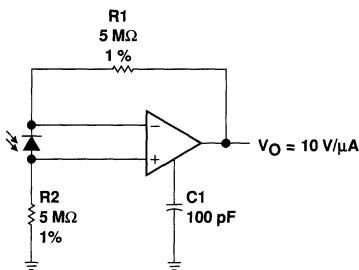
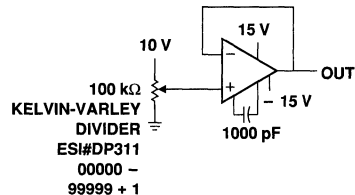


FIGURE 34. AMPLIFIER FOR PHOTODIODE SENSOR



NOTE A: Approximate error due to noise, bias current, common-mode rejection, and voltage gain of the amplifier is 1/5 of a least significant bit.

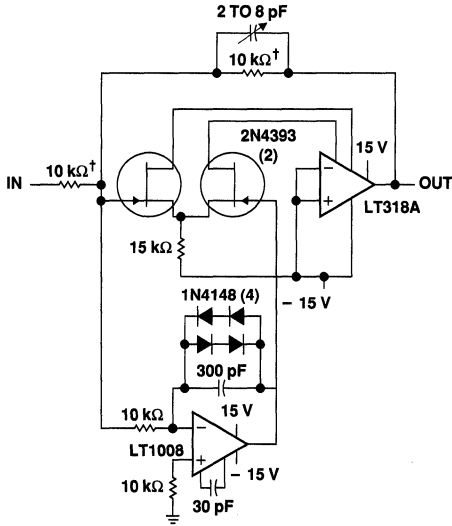
FIGURE 35. FIVE-DECADE KELVIN-VARLEY DIVIDER BUFFERED BY THE LT1008

LT1008M, LT1008C
PICOAMP INPUT CURRENT, MICROVOLT OFFSET
LOW-NOISE OPERATIONAL AMPLIFIERS

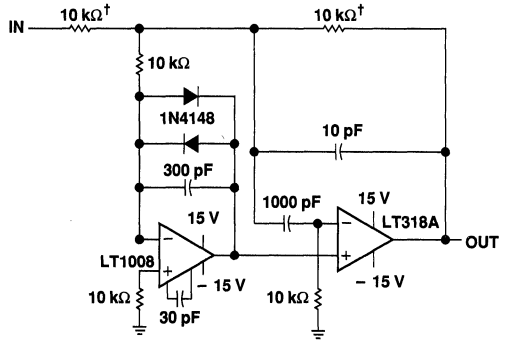
TYPICAL APPLICATION DATA

2

Operational Amplifiers



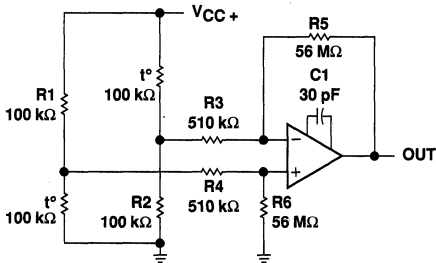
- (a) SR = 100 V/μs
 $I_{IB} = 30 \text{ pA}$
 $V_{IO} = 30 \text{ μV}$
 Settling = 5 μs to 0.01%/10-V step



- (b) SR = 50 V/μs
 $I_{IB} = 30 \text{ pA}$
 $V_{IO} = 30 \text{ μV}$
 $\alpha V_{IO} = 0.3 \text{ μV/°C}$
 BW = 2 MHz
 Settling = 12 μs to 0.01%/10-V step

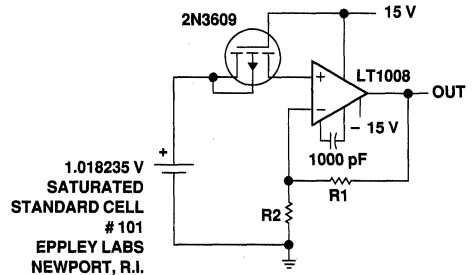
†1% metal film resistor.

FIGURE 36. FAST PRECISION INVERTERS



NOTE A : $A_{VD} = 100$.

FIGURE 37. AMPLIFIER FOR BRIDGE TRANSDUCERS

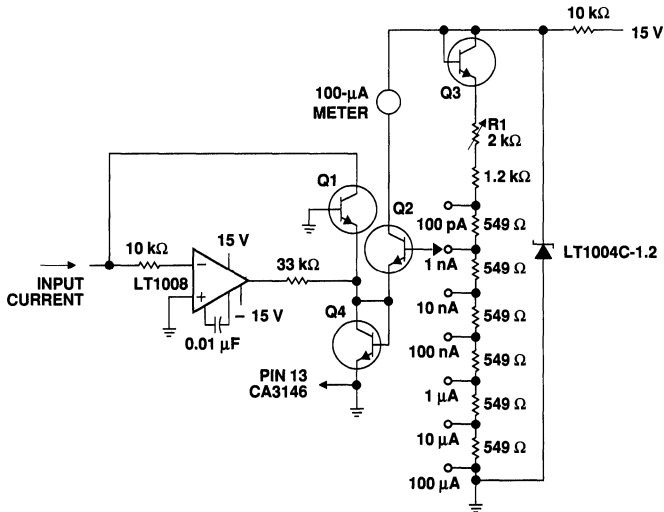


NOTE A : The typical 30-pA input bias current of the LT1008 will degrade the standard cell by only 1 ppm/year. Noise is a fraction of a ppm. Unprotected gate MOSFET isolates standard cell on power down.

FIGURE 38. SATURATED STANDARD-CELL AMPLIFIER

LT1008M, LT1008C
PICOAMP INPUT CURRENT, MICROVOLT OFFSET
LOW-NOISE OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA



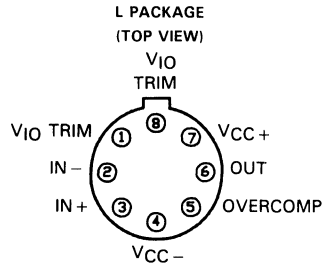
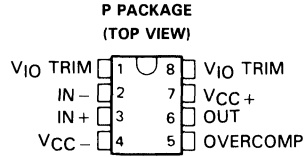
- NOTES: A. This ammeter measures currents from 100 pA to 100 μA without the use of expensive high-value resistors. Accuracy at 100 μA is limited by the offset voltage between Q1 and Q2 and, at 100 pA, by the inverting bias current of the LT1008.
 B. Q1-Q4 RCA CA3146 transistor array.
 C. Adjust R1 for full-scale deflection with 1-μA input current.

FIGURE 39. AMMETER WITH 6-DECADE RANGE

LT1012M, LT1012C HIGH-PERFORMANCE, LOW-NOISE OPERATIONAL AMPLIFIERS

D3186, MARCH 1989

- Internally Compensated
- Input Offset Voltage:
LT1012M . . . 35 μV Max
LT1012C . . . 50 μV Max
- Input Bias Current (LT1012M):
100 pA Max at 25°C
600 pA Max from -55°C to 125°C
- αV_{IO} . . . 1.5 $\mu\text{V}/^\circ\text{C}$ Max
- Typical Peak-To-Peak Noise Voltage . . .
0.5 μV at $f = 0.1$ Hz to 10 Hz
- Low Supply Current . . . 600 μA Max
- CMRR . . . 114 dB Min (LT1012M)
- k_{SVR} . . . 114 dB Min (LT1012M)
- 5-mA Load Current with Voltage Gain of
200,000 Min (LT1012M)



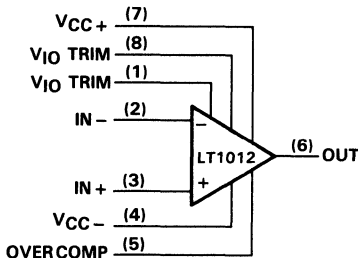
Pin 4 (L package) is in electrical contact with the case.

description

The LT1012 is an internally compensated operational amplifier that can be used in practically all precision applications. The LT1012 combines picoampere bias currents (maintained over the full temperature range), microvolt offset voltage, low offset voltage temperature coefficient and long-term drift, low voltage and current noise, and low power dissipation. High common-mode and supply voltage rejection ratios, low warm-up drift, and the capability to deliver 5-mA load current with a voltage gain of 200,000 complete the LT1012's precision specifications.

The LT1012M is characterized for operation over the full military temperature range of -55°C to 125°C. The LT1012C is characterized for operation from 0°C to 70°C.

symbol



AVAILABLE OPTIONS

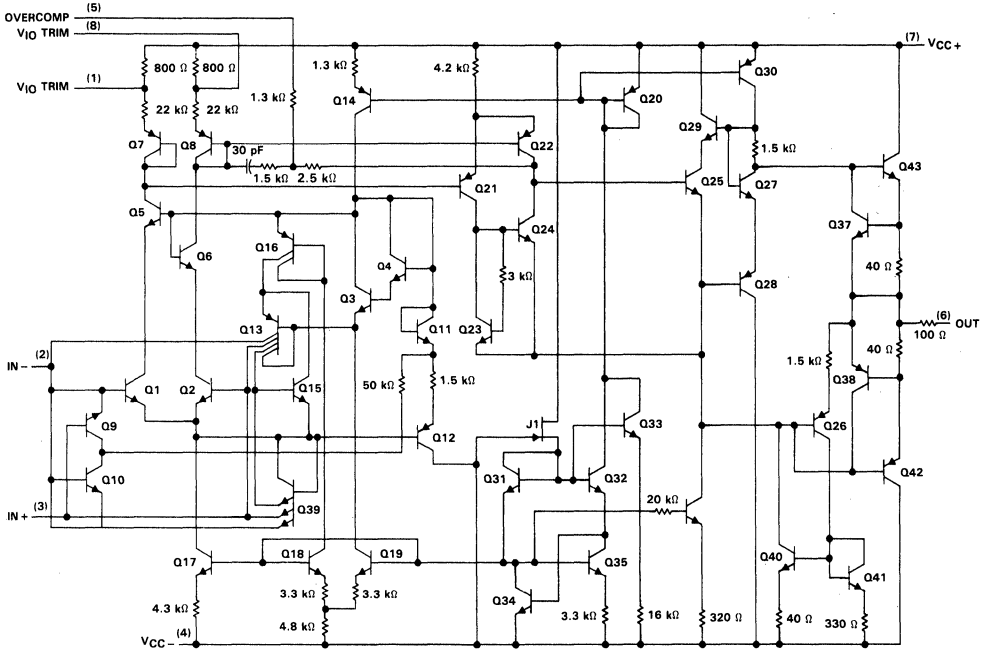
T _A	V _{IO} MAX at 25°C	PACKAGE	
		METAL CAN (L)	PLASTIC DIP (P)
0°C to 70°C	50 μV	LT1012CL	LT1012CP
-55°C to 125°C	35 μV	LT1012ML	—

LT1012M, LT1012C HIGH-PERFORMANCE, LOW-NOISE OPERATIONAL AMPLIFIERS

2

Operational Amplifiers

schematic



All resistor values shown are nominal

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, VCC+ (see Note 1)	+20 V
Supply voltage, VCC- (see Note 1)	-20 V
Input voltage	VCC±
Differential input current (see Note 2)	±10 mA
Duration of output short-circuit at or below 25°C	unlimited
Operating free-air temperature range: LT1012M	-55°C to 125°C
LT1012C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: L package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: P package	260°C

NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between VCC+ and VCC-.
2. Differential input voltages greater than 1 V cause excessive current to flow through the input protection diodes unless limiting resistance is used.

LT1012M, LT1012C HIGH-PERFORMANCE, LOW-NOISE OPERATIONAL AMPLIFIERS

2

Operational Amplifiers

electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$, $V_{IC} = 0$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A^\dagger	LT1012M			LT1012C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{CC\pm} = \pm 15\text{ V}$, $V_{IC} = 0$	25°C	8 35		10 50		μV		
		Full range	180		120				
	$V_{CC\pm} = \pm 15\text{ V}$, $V_{IC} = \pm 13.5\text{ V}$	25°C	20 90		25 120				
		Full range	250		200				
	$V_{CC\pm} = \pm 2\text{ V to } \pm 20\text{ V}$	25°C	20 90		25 120				
Full range		250		200					
α_{VIO} Average temperature coefficient of input offset voltage		Full range	0.2	1.5	0.2	1.5	$\mu\text{V}/^\circ\text{C}$		
Long-term drift of input offset voltage		25°C	0.3		0.3		$\mu\text{V}/\text{mo}$		
I_{IO} Input offset current	$V_{CC\pm} = \pm 15\text{ V}$, $V_{IC} = 0$	25°C	15	100	20	150	pA		
		Full range	250		230				
	$V_{CC\pm} = \pm 15\text{ V}$, $V_{IC} = \pm 13.5\text{ V}$	25°C	25	150	30	200			
		Full range	350		300				
	$V_{CC\pm} = \pm 2\text{ V to } \pm 20\text{ V}$	25°C	25	150	30	200			
Full range		350		300					
α_{IIO} Average temperature coefficient of input offset current		Full range	0.3	2.5	0.3	2.5	$\text{pA}/^\circ\text{C}$		
I_{IB} Input bias current	$V_{CC\pm} = \pm 15\text{ V}$, $V_{IC} = 0$	25°C	± 25	± 100	± 30	± 150	pA		
		Full range	± 600		± 230				
	$V_{CC\pm} = \pm 15\text{ V}$, $V_{IC} = \pm 13.5\text{ V}$	25°C	± 35	± 150	± 40	± 200			
		Full range	± 800		± 300				
	$V_{CC\pm} = \pm 2\text{ V to } \pm 20\text{ V}$	25°C	± 35	± 150	± 40	± 200			
Full range		± 800		± 300					
α_{IIB} Average temperature coefficient of input bias current		Full range	0.6	6	0.3	2.5	$\text{pA}/^\circ\text{C}$		
V_{ICR} Common-mode input voltage range		25°C	± 13.5	± 14	± 13.5	± 14	V		
		Full range	± 13.5		± 13.5				
V_{OM} Maximum peak output voltage swing	$R_L = 10\text{ k}\Omega$	25°C	± 13	± 14	± 13	± 14	V		
		Full range	± 13		± 13				
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 12\text{ V}$, $R_L \geq 10\text{ k}\Omega$	25°C	300	2000	200	2000	V/mV		
		Full range	100		100				
	$V_O = \pm 10\text{ V}$, $R_L \geq 2\text{ k}\Omega$	25°C	200	1000	120	1000			
		Full range	100		100				
CMRR Common-mode rejection ratio	$V_{IC} = \pm 13.5\text{ V}$	25°C	114	132	110	132	dB		
		Full range	108		108				
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{CC\pm}/\Delta V_{IO}$)	$V_{CC\pm} = \pm 2\text{ V to } \pm 20\text{ V}$	25°C	114	132	110	132	dB		
	$V_{CC\pm} = \pm 2.5\text{ V to } \pm 20\text{ V}$	Full range	108		108				
I_{CC} Supply current	$V_{CC\pm} = \pm 15\text{ V}$, $V_{IC} = \pm 13.5\text{ V}$	25°C	380	600	380	600	μA		
		$V_{CC\pm} = \pm 2\text{ V to } \pm 20\text{ V}$	25°C	380	600	380		600	
	Full range		800		800				

† Full range is -55°C to 125°C for the LT1012M and 0°C to 70°C for the LT1012C.

LT1012M, LT1012C HIGH-PERFORMANCE, LOW-NOISE OPERATIONAL AMPLIFIERS

operating characteristics at $T_A = 25^\circ\text{C}$, $V_{CC\pm} = \pm 15\text{ V}$

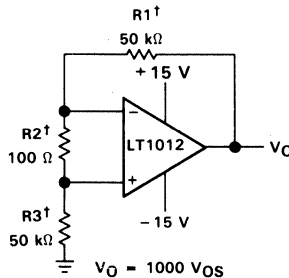
PARAMETER	TEST CONDITIONS	LT1012M			LT1012C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain	0.1	0.2		0.1	0.2		V/ μs
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage	f = 0.1 Hz to 10 Hz			0.5	0.65		μV
V_n	Equivalent input noise voltage	f = 10 Hz, See Note 3			17	30		nV/ $\sqrt{\text{Hz}}$
		f = 1 kHz			14	22		
I_n	Equivalent input noise current	f = 10 Hz			20	20		fA/ $\sqrt{\text{Hz}}$

NOTE 3: This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

2

Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION



† Resistors must have low thermoelectric potential.

This circuit is also used as the burn-in configuration for the LT1012, with supply voltages increased to $\pm 20\text{ V}$, $R_1 = R_3 = 20\text{ k}\Omega$, $R_2 = 200\ \Omega$, $A_V = 100$.

FIGURE 1. TEST CIRCUIT FOR OFFSET VOLTAGE AND ITS TEMPERATURE COEFFICIENT

LT1028, LT1028A ULTRALOW-NOISE, HIGH-SPEED PRECISION OPERATIONAL AMPLIFIERS

D3239, MAY 1988 – REVISED MARCH 1989

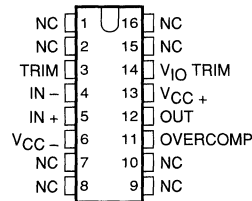
- **Very Low Input Noise Voltage:**
1.1 nV/ $\sqrt{\text{Hz}}$ Max , 0.85 nV/ $\sqrt{\text{Hz}}$ Typ at 1 kHz
for LT1028AM, LT1028AC
- **Low Peak-To-Peak Input Noise Voltage . . .**
35 nV Typ at $f = 0.1$ Hz to 10 Hz
- **Noise Voltage and Current 100% Tested**
- **Gain-Bandwidth Product . . . 50 MHz Min**
- **Slew Rate . . . 11 V/ μs Min**
- **Input Offset Voltage . . . 40 μV Max at 25°C**
for LT1028AM, LT1028AC
- **Offset Voltage Temperature Coefficient . . .**
0.8 $\mu\text{V}/^\circ\text{C}$ Max for LT1028AM, LT1028AC
- **Applications:**
 - Low-Noise Frequency Synthesizers
 - High-Quality Audio
 - Infrared Detectors
 - Accelerometer and Gyro Amplifiers
 - 350- Ω Bridge Signal Conditioning
 - Magnetic Search Coil Amplifiers
 - Hydrophone Amplifiers

description

The LT1028 features excellent noise performance combined with high-speed specifications, distortion-free output, and true precision parameters. Although the LT1028 input stage operates at collector currents of nearly 1 mA to achieve low voltage noise, the input bias current is only 25 or 30 nA at 25°C. The noise voltage of the LT1028 is less than the noise of a 50- Ω resistor. Therefore, even in very-low-source-impedance transducer or audio amplifier applications, the device's contribution to total system noise will be negligible.

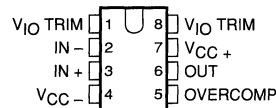
The LT1028AM and LT1028M are characterized for operation over the full military temperature range of -55°C to 125°C. The LT1028AC and LT1028C are characterized for operation from 0°C to 70°C.

**DW PACKAGE
(TOP VIEW)**

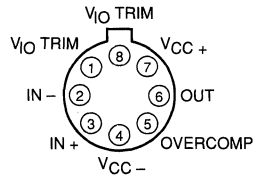


NC – No internal connection

**JG OR P PACKAGE
(TOP VIEW)**



**L PACKAGE
(TOP VIEW)**



Pin 4 (L package) is in electrical contact with the case.

2
Operational Amplifiers

AVAILABLE OPTIONS

T _A	V _{IO} max AT 25°C	PACKAGE			
		SMALL OUTLINE (DW)	CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)
0°C to 70°C	40 μV	—	LT1028ACJG	LT1028ACL	LT1028ACP
	80 μV	LT1028ACDW	LT1028CJG	LT1028CL	LT1028CP
-55°C to 125°C	40 μV	—	LT1028AMJG	LT1028AML	—
	80 μV	—	LT1028MJG	LT1028ML	—

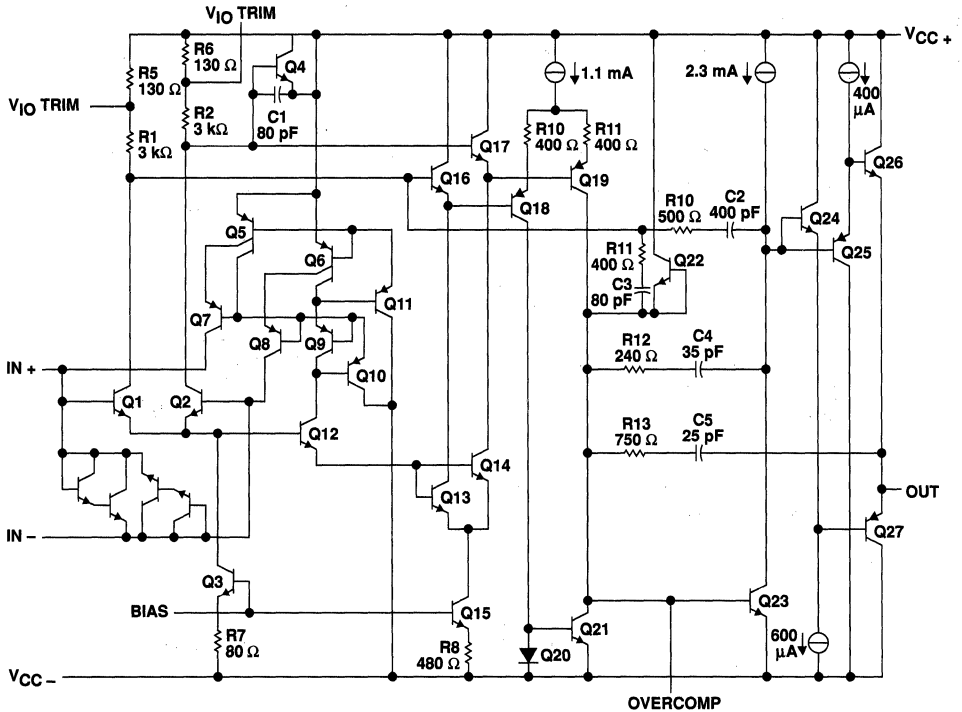
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LT1028, LT1028A ULTRALOW-NOISE, HIGH-SPEED PRECISION OPERATIONAL AMPLIFIERS

schematic



All component and current values shown are nominal.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1): LT1028AM, LT1028M	22 V
LT1028AC, LT1028C	16 V
Supply voltage, V_{CC-} (see Note 1) LT1028AM, LT1028M	-22 V
LT1028AC, LT1028C	-16 V
Differential input current (see Note 2)	± 25 mA
Input voltage range, V_I (any input, see Note 1)	$V_{CC\pm}$
Duration of output short-circuit at (or below) 25°C (see Note 2)	unlimited
Operating free-air temperature, T_A : LT1028AM, LT1028M	-55°C to 125°C
LT1028AC, LT1028C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG or L package	260°C

NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .

2. The specified values for this parameter takes into account junction temperature increase due to supply and output currents.

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Operational Amplifiers

LT1028M, LT1028AM
ULTRALOW-NOISE, HIGH SPEED
PRECISION OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{CC} \pm = \pm 15$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A	LT1028M			LT1028AM			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
V_{IO} Input offset voltage	See Note 3	25°C		20	80		10	40	μ V	
		-55°C to 125°C			180			120		
α_{VIO} Temperature coefficient of input offset voltage		-55°C to 125°C		0.25	1		0.2	0.8	μ V/°C	
Input offset voltage long-term drift	See Note 4	25°C		0.3			0.3	μ V/mo		
I_{IO} Input offset current	$V_{IC} = 0$	25°C		18	100		12	50	nA	
		-55°C to 125°C			180			90		
I_{IB} Input bias current	$V_{IC} = 0$	25°C		30	180		25	90	nA	
		-55°C to 125°C			300			150		
V_{ICR} Common-mode input voltage range		25°C		± 11	± 12.2		± 11	± 12.2	V	
		-55°C to 125°C		± 10.3			± 10.3			
V_{OM} Maximum peak output voltage swing	$R_L \geq 2$ k Ω	25°C		± 12	± 13		± 12.3	± 13	V	
		-55°C to 125°C		± 10.3			± 10.3			
		$R_L \geq 600$ Ω	25°C		± 10.5	± 12.2		± 11		± 12.2
A_{VD} Large-signal differential voltage amplification		$V_O = \pm 12$ V, $R_L \geq 2$ k Ω	25°C	5	30		7	30	V/ μ V	
		$V_O = \pm 10$ V, $R_L \geq 2$ k Ω	-55°C to 125°C		2		3			
		$V_O = \pm 10$ V, $R_L \geq 1$ k Ω	25°C		3.5	20		5		20
		$V_O = \pm 10$ V, $R_L \geq 1$ k Ω	-55°C to 125°C		1.5			2		
		$V_O = \pm 10$ V, $R_L \geq 600$ Ω	25°C		2	15		3		15
r_{ic} Common-mode input resistance		25°C		300		300		M Ω		
r_{id} Differential-mode input resistance		25°C		20		20		k Ω		
c_i Input capacitance		25°C		5		5		pF		
z_o Output impedance	$V_O = 0$, $I_O = 0$, Open loop	25°C		80		80		Ω		
CMRR Common-mode rejection ratio	$V_{IC} = \pm 11$ V	25°C		110	126		114	126	dB	
	$V_{IC} = \pm 10.3$ V	-55°C to 125°C		100			106			
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{CC} \pm / \Delta V_{IO}$)	$V_{CC} \pm = \pm 4$ V to ± 18 V	25°C		110	132		117	133	dB	
	$V_{CC} \pm = \pm 4.5$ V to ± 16 V	-55°C to 125°C		104			110			
I_{CC} Supply current		25°C		7.6	10.5		7.4	9.5	mA	
		-55°C to 125°C			13			11.5		

- NOTES: 3. Input offset voltage measurements are performed by automatic test equipment approximately 0.5 seconds after applying power.
4. Input offset voltage long-term drift refers to the average trend line of offset voltage versus time over extended periods after the first 30 days of operation. Excluding the initial hour of operation, changes in V_{IO} during the first 30 days are typically 2.5 μ V.

2
Operational Amplifiers

LT1028C, LT1028AC

ULTRALOW-NOISE, HIGH SPEED

PRECISION OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{CC\pm} = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A	LT1028C			LT1028AC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO}	Input offset voltage	See Note 3	25°C	20	80	10	40	μV	
			0°C to 70°C	125		80			
α_{VIO}	Temperature coefficient of input offset voltage	0°C to 70°C	0.2	1	0.1	0.8	$\mu\text{V}/^\circ\text{C}$		
	Input offset voltage long-term drift	See Note 4	25°C	0.3		0.3		$\mu\text{V}/\text{mo}$	
I_{IO}	Input offset current	$V_{IC} = 0$	25°C	18	100	12	50	nA	
			0°C to 70°C	130		65			
I_{IB}	Input bias current	$V_{IC} = 0$	25°C	30	180	25	90	nA	
			0°C to 70°C	240		120			
V_{ICR}	Common-mode input voltage range		25°C	$\pm 11 \pm 12.2$		$\pm 11 \pm 12.2$		V	
			0°C to 70°C	± 10.5		± 10.5			
V_{OM}	Maximum peak output voltage swing	$R_L \geq 2\text{ k}\Omega$	25°C	± 12	± 13	± 12.3	± 13	V	
			0°C to 70°C	± 11.5		± 11.5			
			$R_L \geq 600\ \Omega$	25°C	± 10.5	± 12.2	± 11		± 12.2
			$R_L \geq 600\ \Omega$, See Note 2	70°C	± 9	± 10.5	± 9.5		± 11
A_{VD}	Large-signal differential voltage amplification	$V_O = \pm 12\text{ V}$, $R_L \geq 2\text{ k}\Omega$	25°C	5	30	7	30	$\text{V}/\mu\text{V}$	
			$V_O = \pm 10\text{ V}$, $R_L \geq 2\text{ k}\Omega$	0°C to 70°C	3		5		
			$V_O = \pm 10\text{ V}$, $R_L \geq 1\text{ k}\Omega$	25°C	3.5	20	5		20
			$V_O = \pm 10\text{ V}$, $R_L \geq 1\text{ k}\Omega$	0°C to 70°C	2.5		4		
			$V_O = \pm 10\text{ V}$, $R_L \geq 600\ \Omega$	25°C	2	15	3		15
r_{ic}	Common-mode input resistance		25°C	300		300		M Ω	
r_{id}	Differential-mode input resistance		25°C	20		20		k Ω	
c_i	Input capacitance		25°C	5		5		pF	
z_o	Output impedance	$V_O = 0$, $I_O = 0$, Open loop	25°C	80		80		Ω	
CMRR	Common-mode rejection ratio	$V_{IC} = \pm 11\text{ V}$	25°C	110	126	114	126	dB	
		$V_{IC} = \pm 10.3\text{ V}$	0°C to 70°C	106		110			
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{CC\pm} / \Delta V_{IO}$)	$V_{CC\pm} = \pm 4\text{ V to } \pm 18\text{ V}$	25°C	110	132	117	133	dB	
		$V_{CC\pm} = \pm 4.5\text{ V to } \pm 16\text{ V}$	0°C to 70°C	107		114			
I_{CC}	Supply current		25°C	7.6	10.5	7.4	9.5	mA	
			0°C to 70°C	11.5		10.5			

- NOTES: 2. The specified values for this parameter takes into account junction temperature increase due to supply and output currents.
3. Input offset voltage measurements are performed by automatic test equipment approximately 0.5 seconds after applying power.
4. Input offset voltage long-term drift refers to the average trend line of offset voltage versus time over extended periods after the first 30 days of operation. Excluding the initial hour of operation, changes in V_{IO} during the first 30 days are typically 2.5 μV .

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Operational Amplifiers

LT1028, LT1028A
ULTRALOW-NOISE, HIGH SPEED
PRECISION OPERATIONAL AMPLIFIERS

operating characteristics, $V_{CC \pm} = \pm 15 \text{ V}$, $V_{IC} = 0$, $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS	LT1028M, LT1028C			LT1028AM, LT1028AC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate	$A_v = -1$	11	15		11	15		$\text{V}/\mu\text{s}$
V_{NPP}	Peak-to-peak equivalent input noise voltage	$f = 0.1 \text{ Hz to } 10 \text{ Hz}$		35	90		35	75	nV
V_n	Equivalent input noise voltage	$f = 10 \text{ Hz}$, See Note 5		1	1.9		1	1.7	$\text{nV}/\sqrt{\text{Hz}}$
		$f = 1 \text{ kHz}$		0.9	1.2		0.85	1.1	
I_n	Equivalent input noise current	$f = 10 \text{ Hz}$, See Note 6		4.7	12		4.7	10	$\text{pA}/\sqrt{\text{Hz}}$
		$f = 1 \text{ kHz}$, See Note 6		1	1.8		1	1.6	
GBW	Gain bandwidth product	$f = 20 \text{ kHz}$	50	75		50	75		MHz

- NOTES: 5. 10-Hz equivalent input noise voltage is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.
6. Noise current is defined and measured with balanced source resistors. The resulting voltage (after subtracting the resistor noise on an RMS basis) is divided by the sum of the two source resistors to obtain noise current. Maximum 10-Hz noise current can be inferred from testing at 1 kHz.

2

Operational Amplifiers

2

Operational Amplifiers

MC1558, MC1458 DUAL GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

D972, FEBRUARY 1971—REVISED MAY 1988

- Short-Circuit Protection
- Wide Common-Mode and Differential Voltage Ranges
- No Frequency Compensation Required
- Low Power Consumption
- No Latch-up
- Designed to be Interchangeable with Motorola MC1558/MC1458 and Signetics S5558/N5558

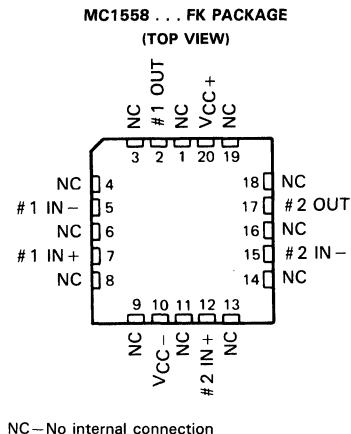
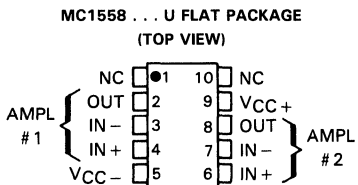
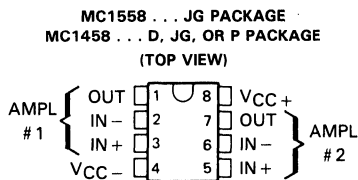
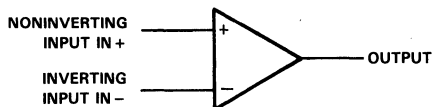
description

The MC1558 and MC1458 are dual general-purpose operational amplifiers with each half electrically similar to the uA741 except that offset null capability is not provided.

The high-common-mode input voltage range and the absence of latch-up make these amplifiers ideal for voltage-follower applications. The devices are short-circuit protected and the internal frequency compensation ensures stability without external components.

The MC1558 is characterized for operation over the full military temperature range of -55°C to 125°C ; the MC1458 is characterized for operation from 0°C to 70°C .

symbol (each amplifier)



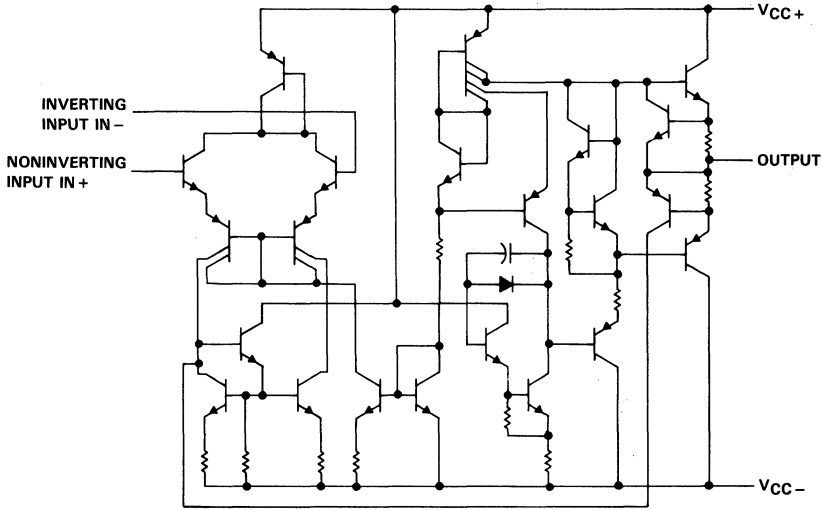
AVAILABLE OPTIONS

DEVICE	SYMBOLIZATION PACKAGE SUFFIX	OPERATING TEMPERATURE RANGE	V_{IO} MAX at 25°C
MC1458	D, JG, P	0°C to 70°C	6 mV

The D packages are available taped and reeled. Add the suffix R to the device type when ordering. (i.e., MC1458DR)

MC1558, MC1458 DUAL GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

schematic (each amplifier)



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	MC1558	MC1458	UNIT
Supply voltage V_{CC+} (see Note 1)	22	18	V
Supply voltage V_{CC-} (see Note 1)	-22	-18	V
Differential input voltage (see Note 2)	± 30	± 30	V
Input voltage at either input (see Notes 1 and 3)	± 15	± 15	V
Duration of output short-circuit (see Note 4)	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	65 to 150	$^{\circ}\text{C}$
Case temperature for 60 seconds: FK package	260		$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	JG or U package	300	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package	260	$^{\circ}\text{C}$

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 volts, whichever is less.
 4. The output may be shorted to ground or either power supply. For the MC1558 only, the unlimited duration of the short-circuit applies at (or below) 125 $^{\circ}\text{C}$ case temperature or 70 $^{\circ}\text{C}$ free-air temperature.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
D	680 mW	5.8 mW/ $^{\circ}\text{C}$	33 $^{\circ}\text{C}$	464 mW	—
FK	680 mW	11.0 mW/ $^{\circ}\text{C}$	88 $^{\circ}\text{C}$	880 mW	275 mW
JG (MC1558)	680 mW	8.4 mW/ $^{\circ}\text{C}$	69 $^{\circ}\text{C}$	672 mW	210 mW
JG (MC1458)	680 mW	6.6 mW/ $^{\circ}\text{C}$	47 $^{\circ}\text{C}$	528 mW	—
P	680 mW	8.0 mW/ $^{\circ}\text{C}$	65 $^{\circ}\text{C}$	640 mW	—
U	675 mW	5.4 mW/ $^{\circ}\text{C}$	25 $^{\circ}\text{C}$	432 mW	135 mW

MC1558, MC1458

DUAL GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$

PARAMETER	TEST CONDITIONS†	MC1558			MC1458			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0$	25°C	1 5		1 6		mV	
		Full range	6		7.5			
I_{IO} Input offset current	$V_O = 0$	25°C	20	200	20	200	nA	
		Full range	500		300			
I_{IB} Input bias current	$V_O = 0$	25°C	80	500	80	500	nA	
		Full range	1500		800			
V_{ICR} Common-mode input voltage range		25°C	±12	±13	±12	±13	V	
		Full range	±12		±12			
V_{OM} Maximum peak output voltage swing	$R_L = 10\text{ k}\Omega$	25°C	±12	±14	±12	±14	V	
	$R_L \geq 10\text{ k}\Omega$	Full range	±12		±12			
	$R_L = 2\text{ k}\Omega$	25°C	±10	±13	±10	±13		
	$R_L \geq 2\text{ k}\Omega$	Full range	±10		±10			
A_{VD} Large-signal differential voltage amplification	$R_L \geq 2\text{ k}\Omega$, $V_O \geq \pm 10\text{ V}$	25°C	50	200	20	200	V/mV	
	$V_O = \pm 10\text{ V}$	Full range	25		15			
B_{OM} Maximum-output-swing bandwidth (closed-loop)	$R_L = 2\text{ k}\Omega$, $V_O \geq \pm 10\text{ V}$, $A_{VD} = 1$, $THD \leq 5\%$	25°C	14		14		kHz	
B_1 Unity-gain bandwidth		25°C	1		1		MHz	
ϕ_m Phase margin	$A_{VD} = 1$	25°C	65°		65°			
A_m Gain margin		25°C	11		11		dB	
r_i Input resistance		25°C	0.3	2	0.3	2	MΩ	
r_o Output resistance	$V_O = 0$, See Note 5	25°C	75		75		Ω	
C_i Input capacitance		25°C	1.4		1.4		pF	
z_{ic} Common-mode input impedance	$f = 20\text{ Hz}$	25°C	200		200		MΩ	
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min}$, $V_O = 0$	25°C	70	90	70	90	dB	
		Full Range	70		70			
k_{SVS} Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	$V_{CC} = \pm 9\text{ V to } \pm 15\text{ V}$, $V_O = 0$	25°C	30	150	30	150	μV/V	
		Full range	150		150			
V_n Equivalent input noise voltage (closed-loop)	$A_{VD} = 100$, $R_S = 0$, $f = 1\text{ kHz}$, $BW = 1\text{ Hz}$	25°C	45		45		nV/√Hz	
I_{OS} Short-circuit output current		25°C	±25	±40	±25	±40	mA	
I_{CC} Supply current (both amplifiers)	No load, $V_O = 0$	25°C	3.4	5	3.4	5.6	mA	
		Full range	6.6		6.6			
P_D Total power dissipation (both amplifiers)	No load, $V_O = 0$	25°C	100	150	100	170	mW	
		Full range	200		200			
V_{O1}/V_{O2} Crosstalk attenuation		25°C	120		120		dB	

† All characteristics are specified under open-loop operating conditions with zero common-mode input voltage unless otherwise specified. Full range for MC1558 is -55°C to 125°C and for MC1458 is 0°C to 70°C .

NOTE 5: This typical value applies only at frequencies above a few hundred hertz because of the effects of drift and thermal feedback.

2 Operational Amplifiers

**MC1558, MC1458
DUAL GENERAL-PURPOSE OPERATIONAL AMPLIFIERS**

operating characteristics, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MC1558			MC1458			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
t_r Rise time	$V_I = 20\text{ mV}$, $R_L = 2\text{ k}\Omega$,		0.3			0.3		μs
Overshoot factor	$C_L = 100\text{ pF}$, See Figure 1		5%			5%		
SR Slew rate at unity gain	$V_I = 10\text{ V}$, $C_L = 100\text{ pF}$, See Figure 1		0.5			0.5		$\text{V}/\mu\text{s}$

PARAMETER MEASUREMENT INFORMATION

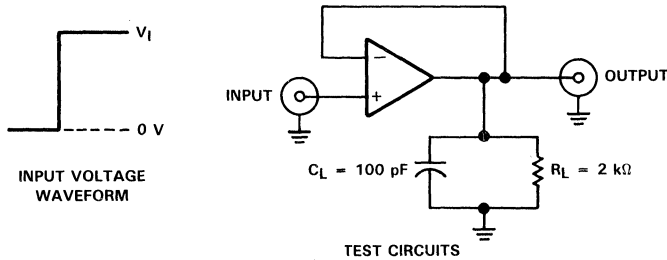


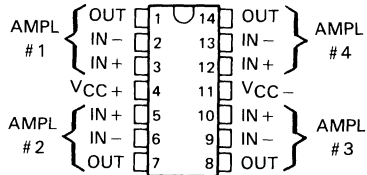
FIGURE 1. RISE TIME, OVERSHOOT, AND SLEW RATE

MC3303, MC3403 QUADRUPLE LOW-POWER OPERATIONAL AMPLIFIERS

D2517, FEBRUARY 1979—REVISED MAY 1988

- Wide Range of Supply Voltages
Single Supply . . . 3 V to 36 V
or Dual Supplies
- Class AB Output Stage
- True Differential Input Stage
- Low Input Bias Current
- Internal Frequency Compensation
- Short-Circuit Protection
- Designed to be Interchangeable with Motorola
MC3303, MC3403

MC3303, MC3403 . . . D, J, OR N PACKAGE
(TOP VIEW)



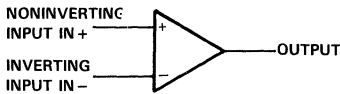
2
Operational Amplifiers

description

The MC3303 and the MC3403 are quadruple operational amplifiers similar in performance to the uA741 but with several distinct advantages. They are designed to operate from a single supply over a range of voltages from 3 V to 36 V. Operation from split supplies is also possible provided the difference between the two supplies is 3 V to 36 V. The common-mode input range includes the negative supply. Output range is from the negative supply to $V_{CC} - 1.5$ V. Quiescent supply currents are less than one-half those of the uA741.

The MC3303 is characterized for operation from -40°C to 85°C and the MC3403 is characterized for operation from 0°C to 70°C .

symbol (each amplifier)



AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE		
		SMALL-OUTLINE (D)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	10 mV	MC3403D	MC3403J	MC3403N
-40°C to 85°C	8 mV	MC3303D	MC3303J	MC3303N

D packages are available taped and reeled. Add "R" suffix to the device type when ordering. (e.g., MC3403DR)

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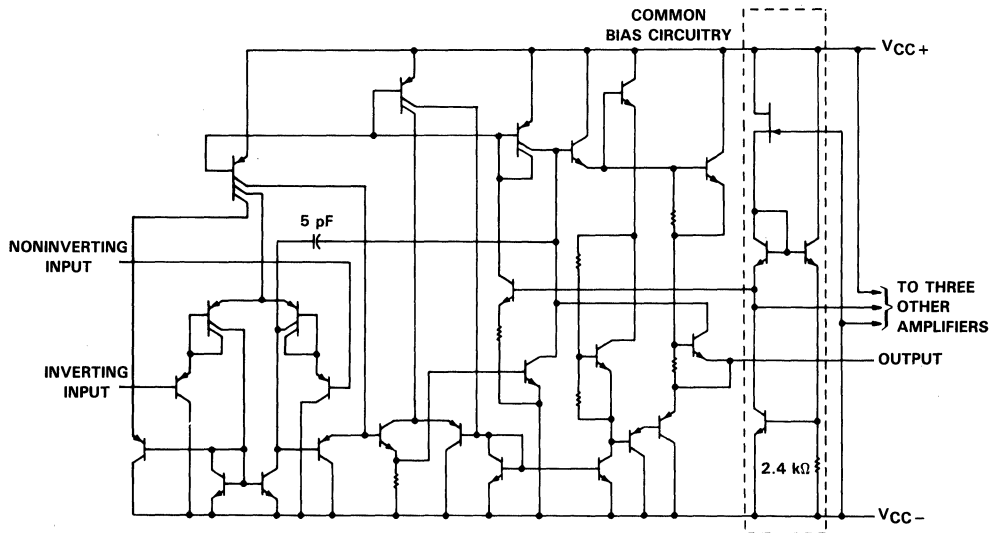
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MC3303, MC3403 QUADRUPLE LOW-POWER OPERATIONAL AMPLIFIERS

schematic (each amplifier)

2

Operational Amplifiers



Component values shown are nominal.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	MC3303	MC3403	UNIT
Supply voltage V_{CC+} (see Note 1)	18	18	V
Supply voltage V_{CC-} (see Note 1)	-18	-18	V
Supply voltage V_{CC+} with respect to V_{CC-}	36	36	V
Differential input voltage (see Note 2)	± 36	± 36	V
Input voltage (see Notes 1 and 3)	± 18	± 18	V
Continuous total power dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-40 to 85	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J package	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or N package	260	°C

- NOTES: 1. These voltage values are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting terminal.
 3. Neither input must ever be more positive than V_{CC+} or more negative than V_{CC-} .

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW
J	1025 mW	8.2 mW/°C	656 mW	533 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW

MC3303, MC3403
QUADRUPLE LOW-POWER OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature; $V_{CC+} = 14\text{ V}$, $V_{CC-} = 0\text{ V}$ for MC3303; $V_{CC\pm} = \pm 15\text{ V}$ for MC3403

PARAMETER	TEST CONDITIONS†	MC3303			MC3403			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX		
V_{IO} Input offset voltage	See Note 4	25 °C	2	8	2	10	mV		
		Full range	10			12			
αV_{IO} Temperature coefficient of input offset voltage	See Note 4	Full range	10			10	$\mu\text{V}/^\circ\text{C}$		
I_{IO} Input offset current	See Note 4	25 °C	30	75	30	50	nA		
		Full range	250			200			
αI_{IO} Temperature coefficient of input offset current	See Note 4	Full range	50			50	$\text{pA}/^\circ\text{C}$		
I_{IB} Input bias current	See Note 4	25 °C	-0.2	-0.5	-0.2	-0.5	μA		
		Full range	-1			-0.8			
V_{ICR} Common-mode input voltage range‡		25 °C	V_{CC-} to 12	V_{CC-} to 12.5	V_{CC-} to 13	V_{CC-} to 13.5	V		
V_{OM} Peak output voltage swing	$R_L = 10\text{ k}\Omega$	25 °C	12	12.5	± 12	± 13.5	V		
	$R_L = 2\text{ k}\Omega$	25 °C	10	12	± 10	± 13			
	$R_L = 2\text{ k}\Omega$	Full range	10			± 10			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 10\text{ V}$, $R_L = 2\text{ k}\Omega$	25 °C	20	200	20	200	V/mV		
		Full range	15			15			
B_{OM} Maximum-output-swing bandwidth	$V_{OPP} = 20\text{ V}$, $A_{VD} = 1$, $\text{THD} \leq 5\%$, $R_L = 2\text{ k}\Omega$	25 °C	9			9	kHz		
B_1 Unity-gain bandwidth	$V_O = 50\text{ mV}$, $R_L = 10\text{ k}\Omega$	25 °C	1			1	MHz		
ϕ_m Phase margin	$C_L = 200\text{ pF}$, $R_L = 2\text{ k}\Omega$	25 °C	60°			60°			
r_i Input resistance	$f = 20\text{ Hz}$	25 °C	0.3	1	0.3	1	M Ω		
r_o Output resistance	$f = 20\text{ Hz}$	25 °C	75			75	Ω		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min}$	25 °C	70	90	70	90	dB		
k_{SVS} Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	$V_{CC} = \pm 2.5$ to $\pm 15\text{ V}$	25 °C	30	150	30	150	$\mu\text{V}/\text{V}$		
I_{OS} Short-circuit output current‡		25 °C	± 10	± 30	± 45	± 10	± 30	± 45	mA
I_{CC} Total supply current	No load, See Note 4	25 °C	2.8			7	2.8	7	mA

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified. Full range for T_A is -40°C to 85°C for MC3303, and 0°C to 70°C for MC3403.

‡ The V_{ICR} limits are directly linked volt-for-volt to supply voltage; the positive limit is 2 V less than V_{CC+} .

§ Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

NOTE 4: V_{IO} , I_{IO} , I_{IB} , and I_{CC} are defined at $V_O = 0$ for MC3403, and $V_O = 7\text{ V}$ for MC3303.

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Operational Amplifiers

MC3303, MC3403 QUADRUPLE LOW-POWER OPERATIONAL AMPLIFIERS

electrical characteristics, $V_{CC+} = 5\text{ V}$, $V_{CC-} = 0\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	MC3303			MC3403			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 2.5\text{ V}$			10	2	10	mV	
I_{IO} Input offset current	$V_O = 2.5\text{ V}$			75	30	50	nA	
I_{IB} Input bias current	$V_O = 2.5\text{ V}$			-0.5	-0.2	-0.5	pA	
V_{OM} Peak output voltage swing‡	$R_L = 10\text{ k}\Omega$ $R_L = 10\text{ k}\Omega$, $V_{CC+} = 5\text{ V to } 30\text{ V}$	3.3	3.5		3.3	3.5	V	
A_{VD} Large-signal differential voltage amplification	$V_O = 1.7\text{ V to } 3.3\text{ V}$, $R_L = 2\text{ k}\Omega$	20	200		20	200	V/mV	
k_{SVS} Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC\pm}$)	$V_{CC} = \pm 15\text{ V to } \pm 2.5\text{ V}$			150		150	$\mu\text{V/V}$	
I_{CC} Supply current	No load, $V_O = 2.5\text{ V}$		2.5	7	2.5	7	mA	
V_{O1}/V_{O2} Crosstalk attenuation	$f = 1\text{ kHz to } 20\text{ kHz}$		120		120		dB	

†All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified.

‡Output will swing essentially to ground.

operating characteristics, $V_{CC+} = 14\text{ V}$, $V_{CC-} = 0\text{ V}$ for MC3303; $V_{CC\pm} = \pm 15\text{ V}$ for MC3403; $T_A = 25^\circ\text{C}$, $A_{VD} = 1$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	$V_I = \pm 10\text{ V}$, See Figure 1 $C_L = 100\text{ pF}$ $R_L = 2\text{ k}\Omega$		0.6		$\text{V}/\mu\text{s}$
t_r Rise time	$\Delta V_O = 50\text{ mV}$, See Figure 1 $C_L = 100\text{ pF}$, $R_L = 10\text{ k}\Omega$		0.35		μs
t_f Fall time			0.35		μs
Overshoot factor			20%		
Crossover distortion	$V_{PP} = 30\text{ mV}$, $V_{Opp} = 2\text{ V}$, $f = 10\text{ kHz}$		1%		

PARAMETER MEASUREMENT INFORMATION

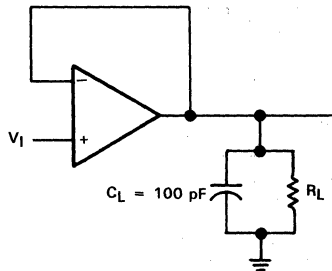


FIGURE 1. UNITY-GAIN AMPLIFIER

MC3303, MC3403
QUADRUPLE LOW-POWER OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

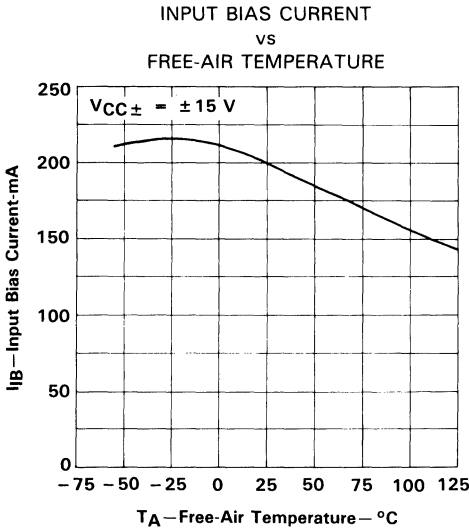


FIGURE 2

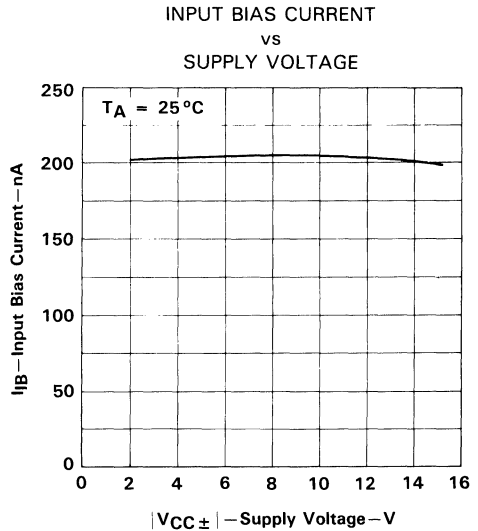


FIGURE 3

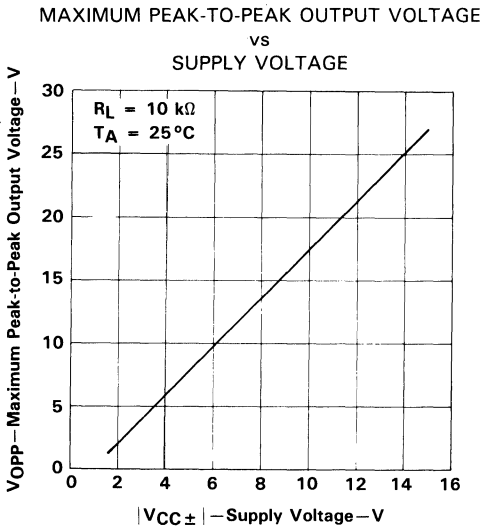


FIGURE 4

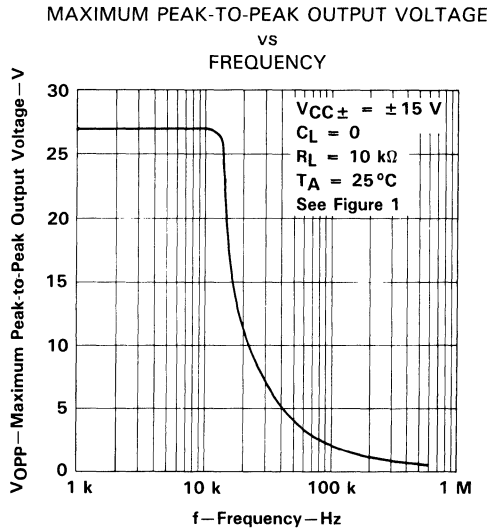


FIGURE 5

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

2
Operational Amplifiers

MC3303, MC3403
QUADRUPLE LOW-POWER OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

**LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREQUENCY**

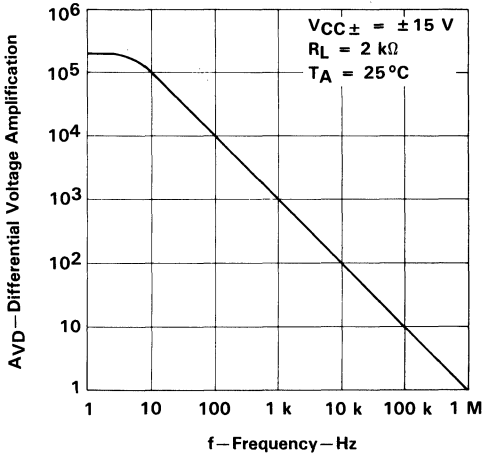


FIGURE 6

**VOLTAGE-FOLLOWER
LARGE-SIGNAL PULSE RESPONSE**

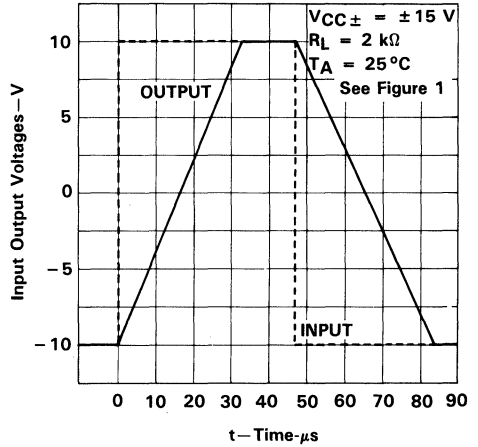


FIGURE 7

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

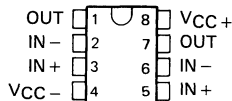
2 Operational Amplifiers

NE5532, NE5532A DUAL LOW-NOISE OPERATIONAL AMPLIFIERS

D2563, NOVEMBER 1979 – REVISED MAY 1988

- Equivalent Input Noise Voltage . . . 5 nV/√Hz Typ at 1 kHz
- Unity-Gain Bandwidth . . . 10 MHz Typ
- Common-Mode Rejection Ratio . . . 100 dB Typ
- High DC Voltage Gain . . . 100 V/m Typ
- Peak-to-Peak Output Voltage Swing . . . 32 V Typ with $V_{CC\pm} = \pm 18$ V and $R_L = 600 \Omega$
- High Slew Rate . . . 9 V/μs Typ
- Wide Supply Voltage Range . . . ± 3 V to ± 20 V
- Designed to be Interchangeable with Signetics NE5532 and NE5532A

NE5532, NE5532A . . . JG OR P
DUAL-IN-LINE PACKAGE
(TOP VIEW)



description

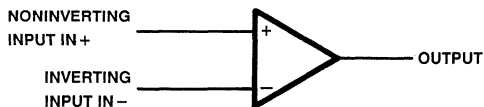
The NE5532 and NE5532A are monolithic high-performance operational amplifiers combining excellent dc and ac characteristics. They feature very low noise, high output drive capability, high unity-gain and maximum-output-swing bandwidths, low distortion, high slew rate, input-protection diodes, and output short-circuit protection. These operational amplifiers are internally compensated for unity gain operation. The NE5532A has specified maximum limits for equivalent input noise voltage.

The NE5532 and NE5532A are characterized for operation from 0°C to 70°C.

AVAILABLE OPTIONS

T _A	V _{IO} MAX at 25°C	PACKAGE	
		CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	4 mV	NE5532JG NE5532AJG	NE5532P NE5532AP

symbol (each amplifier)



PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS
INSTRUMENTS

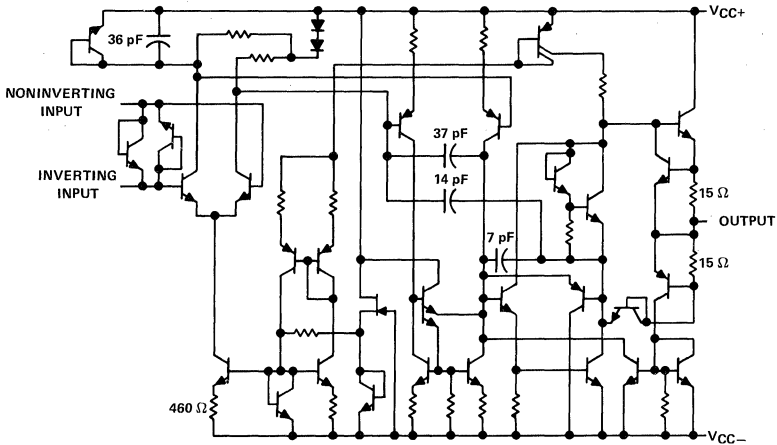
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2-143

NE5532, NE5532A DUAL LOW-NOISE OPERATIONAL AMPLIFIERS

schematic (each amplifier)



Component values shown are nominal.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	22 V
Supply voltage, V_{CC-} (see Note 1)	-22 V
Input voltage, either input (see Notes 1 and 2)	$V_{CC\pm}$
Input current (see Note 3)	± 10 mA
Duration of output short-circuit (see Note 4)	unlimited
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: P package	260°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. The magnitude of the input voltage must never exceed the magnitude of the supply voltage.
 3. Excessive input current will flow if a differential input voltage in excess of approximately 0.6 V is applied between the inputs unless some limiting resistance is used.
 4. The output may be shorted to ground or either power supply. Temperature and/or supply voltages must be limited to ensure the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING
JG	825 mW	6.6 mW/°C	528 mW
P	1000 mW	8.0 mW/°C	640 mW

NE5532, NE5532A DUAL LOW-NOISE OPERATIONAL AMPLIFIERS

electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	$V_O = 0$	$T_A = 25^\circ\text{C}$		0.5	4	mV
			$T_A = 0^\circ\text{C to } 70^\circ\text{C}$			5	
I_{IO}	Input offset current	$T_A = 25^\circ\text{C}$			10	150	nA
		$T_A = 0^\circ\text{C to } 70^\circ\text{C}$				200	
I_{IB}	Input bias current	$T_A = 25^\circ\text{C}$			200	800	nA
		$T_A = 0^\circ\text{C to } 70^\circ\text{C}$				1000	
V_{ICR}	Common-mode input voltage range			± 12	± 13		V
V_{OPP}	Maximum peak-to-peak output voltage swing	$R_L \geq 600\ \Omega$	$V_{CC\pm} = \pm 15\text{ V}$	24	26		V
			$V_{CC\pm} = \pm 18\text{ V}$	30	32		
A_{VD}	Large-signal differential voltage amplification	$R_L \geq 600\ \Omega$, $V_O = \pm 10\text{ V}$	$T_A = 25^\circ\text{C}$	15	50		V/mV
			$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	10			
		$R_L \geq 2\text{ k}\Omega$, $V_O = \pm 10\text{ V}$	$T_A = 25^\circ\text{C}$	25	100		
			$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	15			
A_{vd}	Small-signal differential voltage amplification	$f = 10\text{ kHz}$			2.2		V/mV
B_{OM}	Maximum-output-swing bandwidth	$R_L = 600\ \Omega$, $V_O = \pm 10\text{ V}$			140		kHz
		$R_L = 600\ \Omega$, $V_{CC\pm} = \pm 18\text{ V}$, $V_O = \pm 14\text{ V}$			100		
B_1	Unity-gain bandwidth	$R_L = 600\ \Omega$, $C_L = 100\text{ pF}$			10		MHz
r_i	Input resistance			30	300		k Ω
z_o	Output impedance	$A_{VD} = 30\text{ dB}$, $R_L = 600\ \Omega$, $f = 10\text{ kHz}$			0.3		Ω
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		70	100		dB
k_{SVR}	Supply voltage rejection ratio ($\Delta V_{CC\pm}/\Delta V_{IO}$)	$V_{CC\pm} = \pm 9\text{ V to } \pm 15\text{ V}$, $V_O = 0$		80	100		dB
I_{OS}	Output short-circuit current				38		mA
I_{CC}	Total supply current	No load, $V_O = 0$			8	16	mA
V_{O1}/V_{O2}	Crosstalk attenuation	$V_{O1} = 10\text{ V peak}$, $f = 1\text{ kHz}$			110		dB

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified.

operating characteristics, $V_{CC\pm} = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	NE5532			NE5532A			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain	9			9			V/ms
	Overshoot factor	10%			10%			
V_n	Equivalent input noise voltage	$f = 30\text{ Hz}$	8		8	10		nV/ $\sqrt{\text{Hz}}$
		$f = 1\text{ kHz}$	5		5	6		
I_n	Equivalent input noise current	$f = 30\text{ Hz}$	2.7		2.7		pA/ $\sqrt{\text{Hz}}$	
		$f = 1\text{ kHz}$	0.7		0.7			

2

Operational Amplifiers

2

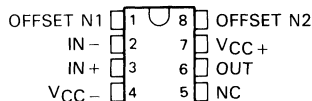
Operational Amplifiers

OP-07C, OP-07D, OP-07E ULTRA-LOW-OFFSET-VOLTAGE OPERATIONAL AMPLIFIERS

D2757, OCTOBER 1983—REVISED JUNE 1988

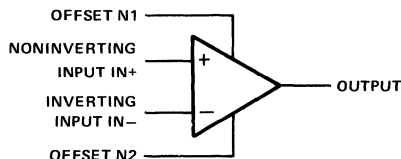
- Ultra-Low Offset Voltage . . .30 μV Typ (OP-07E)
- Ultra-Low Offset Voltage Temperature Coefficient . . . 0.3 $\mu\text{V}/^\circ\text{C}$ Typ (OP-07E)
- Ultra-Low Noise
- No External Components Required
- Replaces Chopper Amplifiers at a Lower Cost
- Single-Chip Monolithic Fabrication
- Wide Input Voltage Range
0 to $\pm 14\text{ V}$ Typ
- Wide Supply Voltage Range
 $\pm 3\text{ V}$ to $\pm 18\text{ V}$
- Essentially Equivalent to Fairchild $\mu\text{A}714$ Operational Amplifiers
- Direct Replacement for PMI OP-07C, OP-07D, OP-07E

D, JG, OR P PACKAGE
(TOP VIEW)



NC—No internal connection

symbol



description

These devices represent a breakthrough in operational amplifier performance. Low offset and long-term stability are achieved by means of a low-noise, chopperless, bipolar-input-transistor amplifier circuit. For most applications, external components are required for offset nulling and frequency compensation. The true differential input, with a wide input voltage range and outstanding common-mode rejection, provides maximum flexibility and performance in high-noise environments and in noninverting applications. Low bias currents and extremely high input impedances are maintained over the entire temperature range. The OP-07 is unsurpassed for low-noise, high-accuracy amplification of very-low-level signals.

These devices are characterized for operation from 0°C to 70°C .

AVAILABLE OPTIONS

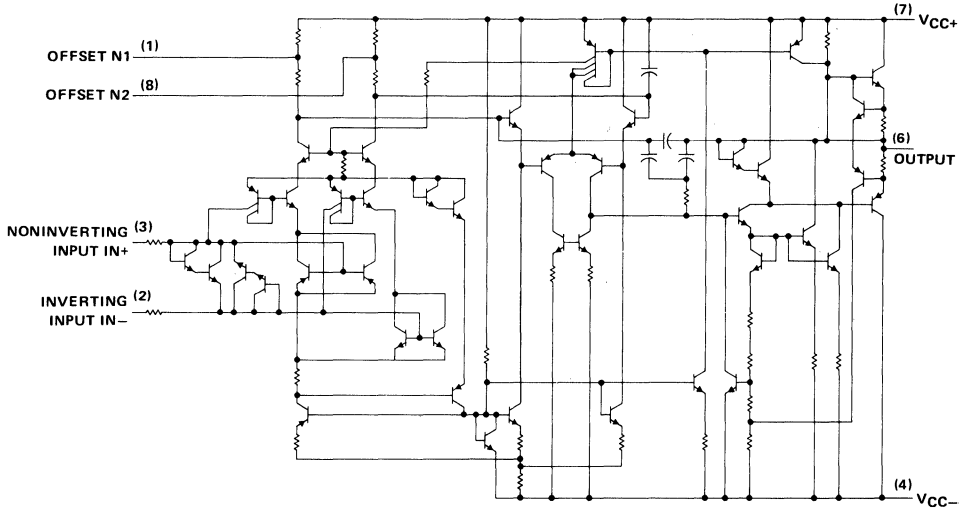
T _A	V _{IO} MAX at 25°C	PACKAGE		
		SMALL OUTLINE (D)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C	150 μV	OP-07CD	OP-07CJG	OP-07CP
to		OP-07DD	OP-07DJG	OP-07DP
70°C	75 μV	OP-07ED	OP-07EJG	OP-07EP

The D package is available taped and reeled. Add the suffix R to the device type when ordering. (e.g., OP-07CDR)

2
Operational Amplifiers

OP-07C, OP-07D, OP-07E
ULTRA-LOW-OFFSET VOLTAGE OPERATIONAL AMPLIFIERS

schematic



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage V_{CC+} (see Note 1)	22 V
Supply voltage V_{CC-}	-22 V
Differential input voltage (see Note 2)	± 30 V
Input voltage (either input, see Note 3)	± 22 V
Duration of output short circuit (see Note 4)	unlimited
Continuous total dissipation at (or below) 25°C free-air temperature (see Note 5)	500 mW
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 volts, whichever is less.
 4. The output may be shorted to ground or either power supply.
 5. For operation above 64°C free-air temperature, derate the D package to 464 mW at 70°C at the rate of 5.8 mW/°C.

2

Operational Amplifiers

electrical characteristics at specified free-air temperature, $V_{CC} \pm = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†		OP-7C			OP-7D			OP-7E			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0$, $R_S = 50\ \Omega$	25°C	60	150	60	150	30	75	μV			
		0°C to 70°C	85	250	85	250	45	130				
α_{VIO} Temperature coefficient of input offset voltage	$V_O = 0$, $R_S = 50\ \Omega$	0°C to 70°C	0.5	1.8	0.7	2.5	0.3	1.3	$\mu\text{V}/^\circ\text{C}$			
Long-term drift of input offset voltage	See Note 6		0.4		0.5		0.3		$\mu\text{V}/\text{mo}$			
Offset adjustment range	$R_S = 20\ \text{k}\Omega$, See Figure 1	25°C	± 4			± 4			mV			
I_{IO} Input offset current		25°C	0.8	6	0.8	6	0.5	3.8	nA			
		0°C to 70°C	1.6	8	1.6	8	0.9	5.3				
α_{IIO} Temperature coefficient of input offset current		0°C to 70°C	12	50	12	50	8	35	$\text{pA}/^\circ\text{C}$			
I_{IB} Input bias current		25°C	± 1.8	± 7	± 2	± 12	± 1.2	± 4	nA			
		0°C to 70°C	± 2.2	± 9	± 3	± 14	± 1.5	± 5.5				
α_{IIB} Temperature coefficient of input bias current		0°C to 70°C	18	50	18	50	13	35	$\text{pA}/^\circ\text{C}$			
V_{ICR} Common-mode input voltage range		25°C	± 13	± 14	± 13	± 14	± 13	± 14	V			
		0°C to 70°C	± 13	± 13.5	± 13	± 13.5	± 13	± 13.5				
V_{OM} Peak output voltage	$R_L \geq 10\ \text{k}\Omega$	25°C	± 12	± 13	± 12	± 13	± 12.5	± 13	V			
	$R_L \geq 2\ \text{k}\Omega$		± 11.5	± 12.8	± 11.5	± 12.8	± 12	± 12.8				
	$R_L \geq 1\ \text{k}\Omega$			± 12		± 12	± 10.5	± 12				
	$R_L \geq 2\ \text{k}\Omega$		0°C to 70°C	± 11	± 12.6	± 11	± 12.6	± 12		± 12.6		
A_{VD} Large-signal differential voltage amplification	$V_{CC} \pm = \pm 3\text{ V}$, $V_O = \pm 0.5\text{ V}$, $R_L \geq 500\ \text{k}\Omega$	25°C	100	400	400		150	400	V/mV			
	$V_O = \pm 10\text{ V}$, $R_L = 2\ \text{k}\Omega$	25°C	120	400	120	400	200	500				
		0°C to 70°C	100	400	100	400	180	450				
B_1 Unity gain bandwidth		25°C	0.4	0.6	0.4	0.6	0.4	0.6	MHz			
r_i Input resistance		25°C	8	33	7	31	15	50	$\text{M}\Omega$			
CMRR Common-mode rejection ratio	$V_{IC} = \pm 13\text{ V}$, $R_S = 50\ \Omega$	25°C	100	120	94	110	106	123	dB			
		0°C to 70°C	97	120	94	106	103	123				
k_{SVS} Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	$V_{CC} \pm = \pm 3\text{ V}$ to $\pm 18\text{ V}$, $R_S = 50\ \Omega$	25°C	7	32	7	32	5	20	$\mu\text{V}/\text{V}$			
		0°C to 70°C	10	51	10	51	7	32				
P_D Power dissipation	$V_O = 0$, No load	25°C	80	150	80	150	75	120	mW			
	$V_{CC} \pm = \pm 3\text{ V}$, $V_O = 0$, No load		4	8	4	8	4	6				

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise noted.

NOTE 6: Since long-term drift cannot be measured on the individual devices prior to shipment, this specification is not intended to be a guarantee or warranty. It is an engineering estimate of the averaged trend line of drift versus time over extended periods after the first thirty days of operation.

OP-07C, OP-07D, OP-07E
ULTRA-LOW-OFFSET VOLTAGE OPERATIONAL AMPLIFIERS

operating characteristics at specified free-air temperature, $V_{CC} \pm = \pm 15$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONS†		OP-7C			OP-7D			OP-7E			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_n Equivalent input noise voltage	$T_A = 25^\circ\text{C}$	$f = 10$ Hz	10.5			10.5			10.3			$\text{nV}/\sqrt{\text{Hz}}$
		$f = 100$ Hz	10.2			10.3			10.0			
		$f = 1$ kHz	9.8			9.8			9.6			
V_{NPP} Peak-to-peak equivalent input noise voltage	$f = 0.1$ Hz to 10 Hz, $T_A = 25^\circ\text{C}$		0.38			0.38			0.35			μV
I_n Equivalent input noise current	$T_A = 25^\circ\text{C}$	$f = 10$ Hz	0.35			0.35			0.32			$\text{pA}/\sqrt{\text{Hz}}$
		$f = 100$ Hz	0.15			0.15			0.14			
		$f = 1$ kHz	0.13			0.13			0.12			
I_{NPP} Peak-to-peak equivalent input noise current	$f = 0.1$ Hz to 10 Hz, $T_A = 25^\circ\text{C}$		15			15			14			pA
SR Slew rate	$R_L \geq 2$ k Ω , $T_A = 25^\circ\text{C}$		0.3			0.3			0.3			$\text{V}/\mu\text{s}$

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified.

TYPICAL APPLICATION DATA

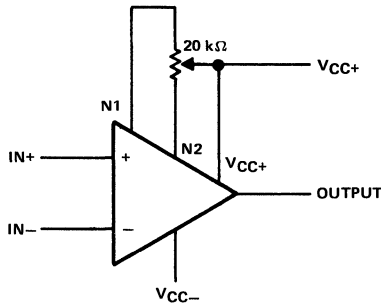


FIGURE 1. INPUT OFFSET VOLTAGE NULL CIRCUIT

OP-27A, OP-27C, OP-27E, OP-27G OP-37A, OP-37C, OP-37E, OP-37G

LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

D3176, FEBRUARY 1989

- Direct Replacements for PMI and LTC OP-27 and OP-37 Series

Features of OP-27A, OP-27C, OP-37A, and OP-37C:

- Maximum Equivalent Input Noise Voltage:
3.8 nV/ $\sqrt{\text{Hz}}$ at 1 kHz
5.5 nV/ $\sqrt{\text{Hz}}$ at 10 Hz
- Very Low Peak-to-Peak Noise Voltage at 0.1 Hz to 10 Hz . . . 80 nV Typ
- Low Input Offset Voltage . . . 25 μV Max
- High Voltage Amplification . . . 1 V/ μV Min

Feature of OP-37 Series:

- Minimum Slew Rate . . . 11 V/ μs

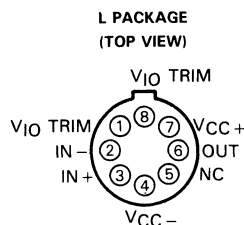
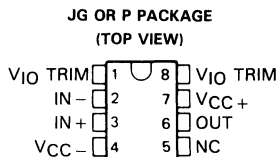
description

The OP-27 and OP-37 operational amplifiers combine outstanding noise performance with excellent precision and high-speed specifications. The wideband noise is only 3 nV/ $\sqrt{\text{Hz}}$, and with the 1/f noise corner at 2.7 Hz, low noise is maintained for all low-frequency applications.

The outstanding characteristics of the OP-27 and OP-37 make these devices excellent choices for low-noise amplifier applications requiring precision performance and reliability. Additionally, the OP-37 is free of latch-up in high-gain, large-capacitive-feedback configurations.

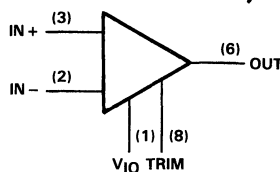
The OP-27 series is compensated for unity gain. The OP-37 series is decompensated for increased bandwidth and slew rate and is stable down to a gain of 5.

The OP-27A, OP-27C, OP-37A, and OP-37C are characterized for operation over the full military temperature range of -55°C to 125°C . The OP-27E, OP-27G, OP-37E, and OP-37G are characterized for operation from -25°C to 85°C .



NC—No internal connection

symbol



AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	STABLE GAIN	PACKAGE		
			CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)
-25°C to 85°C	25 μV	1	OP27EJG	OP27EL	OP27EP
		5	OP37EJG	OP37EL	OP37EP
	100 μV	1	OP27GJG	OP27GL	OP27GP
		5	OP37GJG	OP37GL	OP37GP
-55°C to 125°C	25 μV	1	OP27AJG	OP27AL	OP27AP
		5	OP37AJG	OP37AL	OP37AP
	100 μV	1	OP27CJG	OP27CL	OP27CP
		5	OP37CJG	OP37CL	OP37CP

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



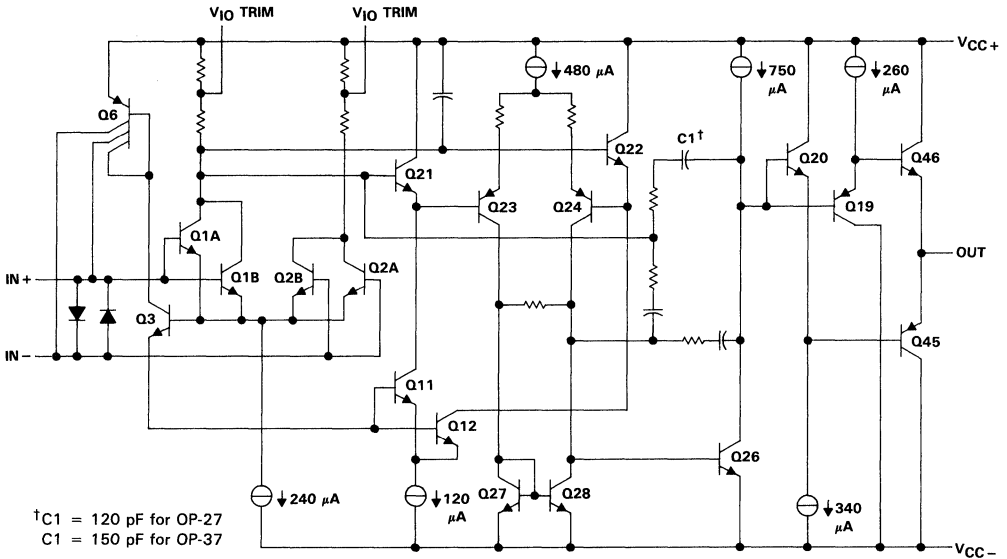
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**OP-27A, OP-27C, OP-27E, OP-27G
OP-37A, OP-37C, OP-37E, OP-37G
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS**

schematic

2
Operational Amplifiers



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V _{CC+} (see Note 1)	22 V
Supply voltage, V _{CC-} (see Note 1)	-22 V
Input voltage	V _{CC±}
Duration of output short circuit	unlimited
Differential input current (see Note 2)	±25 mA
Continuous power dissipation	see Dissipation Rating Table
Operating free-air temperature range: OP-27A, OP-27C, OP-37A, OP-37C	-55°C to 125°C
OP-27E, OP-27G, OP-37E, OP-37G	-25°C to 85°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG or L package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: P package	260°C

- NOTES: 1. All voltage values are with respect to the midpoint between V_{CC+} and V_{CC-} unless otherwise noted.
2. The inputs are protected by back-to-back diodes. Current-limiting resistors are not used in order to achieve low noise. Excessive input current will flow if a differential input voltage in excess of approximately ±0.7 V is applied between the inputs unless some limiting resistance is used.

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C POWER RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 85°C POWER RATING	T _A = 125°C POWER RATING
JG (OP-27A, OP-27C, OP-37A, OP-37C)	1050 mW	8.4 mW/°C	546 mW	210 mW
JG (OP-27E, OP-27G, OP-37E, OP-37G)	825 mW	6.6 mW/°C	429 mW	N/A
L (OP-27A, OP-27C, OP-37A, OP-37C)	825 mW	6.6 mW/°C	429 mW	165 mW
L (OP-27E, OP-27G, OP-37E, OP-37G)	650 mW	5.2 mW/°C	338 mW	N/A
P	1000 mW	8.0 mW/°C	520 mW	N/A

OP-27A, OP-27C, OP-37A, OP-37C

LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

recommended operating conditions

	OP-27A, OP-37A			OP-27C, OP-37C			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, V_{CC+}	4	15	22	4	15	22	V
Supply voltage, V_{CC-}	-4	-15	-22	-4	-15	-22	V
Common-mode input voltage, V_{ICR}	$V_{CC\pm} = \pm 15\text{ V}, T_A = 25^\circ\text{C}$			± 11			V
	$V_{CC\pm} = \pm 15\text{ V}, T_A = -55^\circ\text{C to } 125^\circ\text{C}$			± 10.3			
Operating free-air temperature, T_A	-55	125		-55	125		$^\circ\text{C}$

electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A	OP-27A, OP-37A			OP-27C, OP-37C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0, V_{IC} = 0$ $R_S = 50\ \Omega$, See Note 3	25°C	10	25		30	100	μV	
		$-55^\circ\text{C to } 125^\circ\text{C}$		60			300		
$\propto V_{IO}$ Average temperature coefficient of input offset voltage		$-55^\circ\text{C to } 125^\circ\text{C}$	0.2	0.6		0.4	1.8	$\mu\text{V}/^\circ\text{C}$	
Long-term drift of input offset voltage	See Note 4		0.2	1		0.4	2	$\mu\text{V}/\text{mo}$	
I_{IO} Input offset current	$V_O = 0, V_{IC} = 0$	25°C	7	35		12	75	nA	
		$-55^\circ\text{C to } 125^\circ\text{C}$		50			135		
I_{IB} Input bias current	$V_O = 0, V_{IC} = 0$	25°C	± 10	± 40		± 15	± 80	nA	
		$-55^\circ\text{C to } 125^\circ\text{C}$		± 60			± 150		
V_{ICR} Common-mode input voltage range		25°C	± 11			± 11		V	
		$-55^\circ\text{C to } 125^\circ\text{C}$	± 10.3			± 10.2			
V_{OM} Peak output voltage swing	$R_L \geq 2\ \text{k}\Omega$	25°C	± 12	± 13.8		± 11.5	± 13.5	V	
	$R_L \geq 0.6\ \text{k}\Omega$		± 10	± 11.5		± 10	± 11.5		
	$R_L \geq 2\ \text{k}\Omega$		$-55^\circ\text{C to } 125^\circ\text{C}$	± 11.5			± 10.5		
A_{VD} Large-signal differential voltage amplification	$R_L \geq 2\ \text{k}\Omega, V_O = \pm 10\ \text{V}$	25°C	1000	1800		700	1500	V/mV	
	$R_L \geq 1\ \text{k}\Omega, V_O = \pm 10\ \text{V}$		800	1500		1500			
	$R_L \geq 0.6\ \text{k}\Omega, V_O = \pm 1\ \text{V}, V_{CC} = \pm 4\ \text{V}$		250	700		200	500		
	$R_L \geq 2\ \text{k}\Omega, V_O = \pm 10\ \text{V}$		$-55^\circ\text{C to } 125^\circ\text{C}$	600			300		
$r_{i(CM)}$ Common-mode input resistance			3			2	G Ω		
r_o Output resistance	$V_O = 0, I_O = 0$	25°C	70			70	Ω		
CMRR Common-mode rejection ratio	$V_{IC} = \pm 11\ \text{V}$	25°C	114	126		100	120	dB	
	$V_{IC} = \pm 10\ \text{V}$	$-55^\circ\text{C to } 125^\circ\text{C}$	108			94			
k_{SVR} Supply voltage rejection ratio	$V_{CC\pm} = \pm 4\ \text{V to } \pm 18\ \text{V}$	25°C	100	120		94	118	dB	
	$V_{CC\pm} = \pm 4.5\ \text{V to } \pm 18\ \text{V}$	$-55^\circ\text{C to } 125^\circ\text{C}$	96			86			

- NOTES: 3. Input offset voltage measurements are performed by automatic test equipment approximately 0.5 seconds after applying power.
4. Long-term drift of input offset voltage refers to the average trend line of offset voltage versus time over extended periods after the first 30 days of operation. Excluding the initial hour of operation, changes in V_{IO} during the first 30 days are typically 2.5 μV . See Figure 3.

2

Operational Amplifiers

OP-27E, OP-37E, OP-27G, OP-37G
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, V_{CC+}		4	15	22	V
Supply voltage, V_{CC-}		-4	-15	-22	V
Common-mode input voltage, range	$V_{CC\pm} = \pm 15\text{ V}, T_A = 25^\circ\text{C}$				± 11
	$V_{CC\pm} = \pm 15\text{ V}, T_A = -55^\circ\text{C to } 125^\circ\text{C}$				± 10.5
Operating free-air temperature, T_A		-25	85		$^\circ\text{C}$

electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A	OP-27E, OP-37E			OP-27G, OP-37G			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0, V_{IC} = 0$ $R_S = 50\ \Omega$, See Note 3	25 $^\circ\text{C}$	10		25	30		100	μV
		-25 $^\circ\text{C to } 85^\circ\text{C}$			50			220	
$\propto V_{IO}$ Average temperature coefficient of input offset voltage		-25 $^\circ\text{C to } 85^\circ\text{C}$	0.2	0.6		0.4	1.8		$\mu\text{V}/^\circ\text{C}$
Long-term drift of input offset voltage	See Note 4		0.2	1		0.4	2		$\mu\text{V}/\text{mo}$
I_{IO} Input offset current	$V_O = 0, V_{IC} = 0$	25 $^\circ\text{C}$	7		35	12		75	nA
		-25 $^\circ\text{C to } 85^\circ\text{C}$			50			135	
I_{IB} Input bias current	$V_O = 0, V_{IC} = 0$	25 $^\circ\text{C}$	± 10	± 40		± 15	± 80		nA
		-25 $^\circ\text{C to } 85^\circ\text{C}$		± 60			± 150		
V_{ICR} Common-mode input voltage range		25 $^\circ\text{C}$	± 11			± 11			V
		-25 $^\circ\text{C to } 85^\circ\text{C}$	± 10.5			± 10.5			
V_{OM} Peak output voltage swing	$R_L \geq 2\ \text{k}\Omega$	25 $^\circ\text{C}$	$\pm 12 \pm 13.8$			$\pm 11.5 \pm 13.5$			V
	$R_L \geq 0.6\ \text{k}\Omega$		$\pm 10 \pm 11.5$			$\pm 10 \pm 11.5$			
	$R_L \geq 2\ \text{k}\Omega$	-25 $^\circ\text{C to } 85^\circ\text{C}$	± 11.7			± 11			
A_{VD} Large-signal differential voltage amplification	$R_L \geq 2\ \text{k}\Omega, V_O = \pm 10\ \text{V}$	25 $^\circ\text{C}$	1000	1800		700	1500		V/mV
	$R_L \geq 1\ \text{k}\Omega, V_O = \pm 10\ \text{V}$		800	1500		1500			
	$R_L \geq 0.6\ \text{k}\Omega, V_O = \pm 1\ \text{V}, V_{CC} = \pm 4\ \text{V}$		250	700		200	500		
	$R_L \geq 2\ \text{k}\Omega, V_O = \pm 10\ \text{V}$		-25 $^\circ\text{C to } 85^\circ\text{C}$	750		450			
$r_{i(CM)}$ Common-mode input resistance			3			2			$\text{G}\Omega$
r_o Output resistance	$V_O = 0, I_O = 0$	25 $^\circ\text{C}$	70			70			Ω
CMRR Common-mode rejection ratio	$V_{IC} = \pm 11\ \text{V}$	25 $^\circ\text{C}$	114	126		100	120		dB
	$V_{IC} = \pm 10\ \text{V}$	-25 $^\circ\text{C to } 85^\circ\text{C}$	110		96				
kSVR Supply voltage rejection ratio	$V_{CC\pm} = \pm 4\ \text{V to } \pm 18\ \text{V}$	25 $^\circ\text{C}$	100	120		94	118		dB
	$V_{CC\pm} = \pm 4.5\ \text{V to } \pm 18\ \text{V}$	-25 $^\circ\text{C to } 85^\circ\text{C}$	97		90				

- NOTES: 3. Input offset voltage measurements are performed by automatic test equipment approximately 0.5 seconds after applying power.
 4. Long-term drift of input offset voltage refers to the average trend line of offset voltage versus time over extended periods after the first 30 days of operation. Excluding the initial hour of operation, changes in V_{IO} during the first 30 days are typically 2.5 μV . See Figure 3.

2

Operational Amplifiers

OP-27A, OP-27C, OP-27E, OP-27G
OP-37A, OP-37C, OP-37E, OP-37G
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

OP-27 operating characteristics over operating free-air temperature range, $V_{CC} \pm = \pm 15$ V

PARAMETER	TEST CONDITIONS	OP-27A, OP-27E			OP-27C, OP-27G			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain	$A_{VD} \geq 1, R_L \geq 2 \text{ k}\Omega$						$V/\mu\text{s}$
V_{NPP}	Peak-to-peak equivalent input noise voltage	$f = 0.1 \text{ Hz to } 10 \text{ Hz}, R_S = 100 \Omega$, See Figure 34						μV
V_n	Equivalent input noise voltage	$f = 10 \text{ Hz}, R_S = 100 \Omega$,						$\text{nV}/\sqrt{\text{Hz}}$
		$f = 30 \text{ Hz}, R_S = 100 \Omega$						
		$f = 1 \text{ kHz}, R_S = 100 \Omega$						
I_n	Equivalent input noise current	$f = 10 \text{ Hz}$, See Figure 35						$\text{pA}/\sqrt{\text{Hz}}$
		$f = 30 \text{ Hz}$, See Figure 35						
		$f = 1 \text{ kHz}$, See Figure 35						
GBW	Gain bandwidth product	$f = 100 \text{ kHz}$						MHz

OP-37 operating characteristics over operating free-air temperature range, $V_{CC} \pm = \pm 15$ V

PARAMETER	TEST CONDITIONS	OP-37A, OP-37E			OP-37C, OP-37G			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain	$A_{VD} \geq 5, R_L \geq 2 \text{ k}\Omega$						$V/\mu\text{s}$
V_{NPP}	Peak-to-peak equivalent input noise voltage	$f = 0.1 \text{ Hz to } 10 \text{ Hz}, R_S = 100 \Omega$, See Figure 34						μV
V_n	Equivalent input noise voltage	$f = 10 \text{ Hz}, R_S = 100 \Omega$						$\text{nV}/\sqrt{\text{Hz}}$
		$f = 30 \text{ Hz}, R_S = 100 \Omega$						
		$f = 1 \text{ kHz}, R_S = 100 \Omega$						
I_n	Equivalent input noise current	$f = 10 \text{ Hz}$, See Figure 35						$\text{pA}/\sqrt{\text{Hz}}$
		$f = 30 \text{ Hz}$, See Figure 35						
		$f = 1 \text{ kHz}$, See Figure 35						
GBW	Gain bandwidth product	$f = 10 \text{ kHz}$						MHz
		$A_V \geq 5, f = 1 \text{ MHz}$						

2

Operational Amplifiers

**OP-27A, OP-27C, OP-27E, OP-27G
 OP-37A, OP-37C, OP-37E, OP-37G
 LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS

table of graphs

			FIGURE
V_{IO}	Input offset voltage	vs Temperature	1
ΔV_{IO}	Change in input offset voltage	vs Time after power-on	2
		vs Time (long-term drift)	3
I_{IO}	Input offset current	vs Temperature	4
I_{IB}	Input bias current	vs Temperature	5
V_{ICR}	Common-mode input voltage range	vs Supply voltage	6
V_{OM}	Maximum peak output voltage	vs Load resistance	7
V_{OPP}	Maximum peak-to-peak output voltage	vs Frequency	8, 9
A_{VD}	Differential voltage amplification	vs Supply voltage	10
		vs Load resistance	11
		vs Frequency	12, 13, 14
CMRR	Common-mode rejection ratio	vs Frequency	15
k_{SVR}	Supply voltage rejection ratio	vs Frequency	16
SR	Slew rate	vs Temperature	17
		vs Supply voltage	18
		vs Load resistance	19
ϕ_m	Phase margin	vs Temperature	20, 21
ϕ	Phase shift	vs Frequency	12, 13
V_n	Equivalent input noise voltage	vs Bandwidth	22
		vs Source resistance	23
		vs Supply voltage	24
		vs Temperature	25
		vs Frequency	26
I_n	Equivalent input noise current	vs Frequency	27
GBW	Gain bandwidth product	vs Temperature	20, 21
I_{OS}	Short-circuit output current	vs Time	28
I_{CC}	Supply current	vs Supply voltage	29
	Pulse response	Small-signal	30, 32
		Large-signal	31, 33

2

Operational Amplifiers

OP-27A, OP-27C, OP-27E, OP-27G
OP-37A, OP-37C, OP-37E, OP-37G
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

INPUT OFFSET VOLTAGE
OF REPRESENTATIVE UNIT
VS
FREE-AIR TEMPERATURE

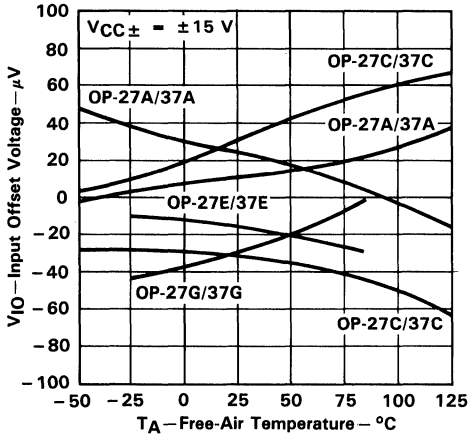


FIGURE 1

WARM-UP CHANGE IN
INPUT OFFSET VOLTAGE
VS
ELAPSED TIME

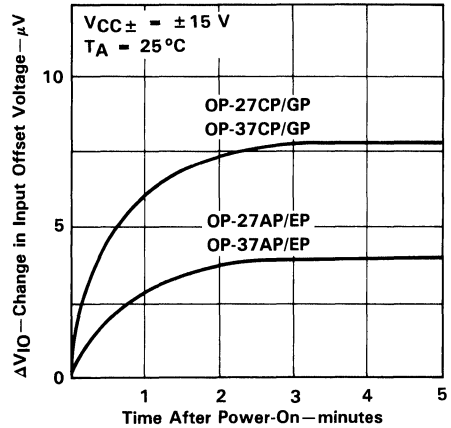


FIGURE 2

LONG-TERM DRIFT OF
INPUT OFFSET VOLTAGE
OF REPRESENTATIVE UNITS

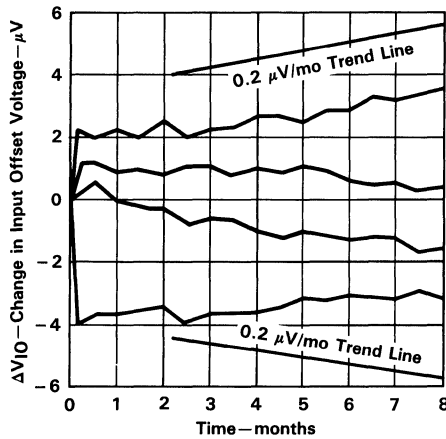


FIGURE 3

†Data for temperatures below -25°C and above 85°C are applicable to the OP-27A, OP-27C, OP-37A, and OP-37C only.

**OP-27A, OP-27C, OP-27E, OP-27G
 OP-37A, OP-37C, OP-37E, OP-37G
 LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS†

**INPUT OFFSET CURRENT
 vs
 FREE-AIR TEMPERATURE**

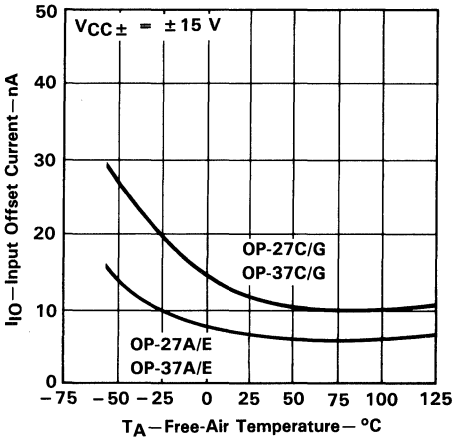


FIGURE 4

**INPUT BIAS CURRENT
 vs
 FREE-AIR TEMPERATURE**

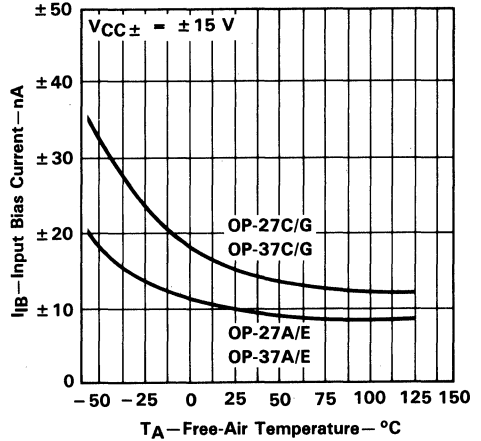


FIGURE 5

**2
Operational Amplifiers**

**COMMON-MODE INPUT VOLTAGE RANGE LIMITS
 vs
 SUPPLY VOLTAGE**

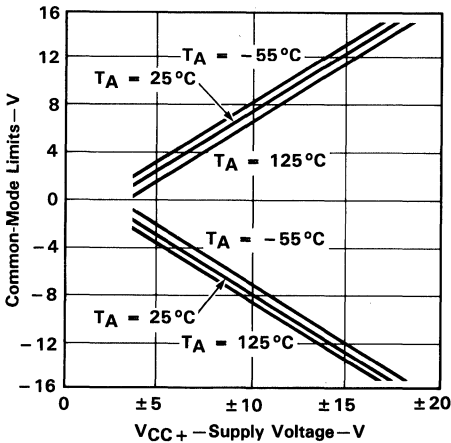


FIGURE 6

**MAXIMUM PEAK OUTPUT VOLTAGE
 vs
 LOAD RESISTANCE**

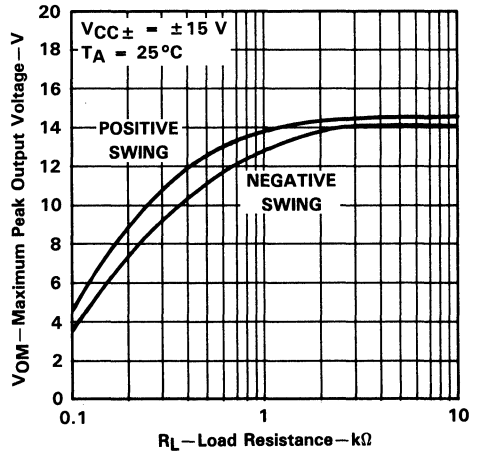


FIGURE 7

†Data for temperatures below -25°C and above 85°C are applicable to the OP-27A, OP-27C, OP-37A, and OP-37C only.

**OP-27A, OP-27C, OP-27E, OP-27G
OP-37A, OP-37C, OP-37E, OP-37G
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS

**OP-27
MAXIMUM PEAK-TO-PEAK
OUTPUT VOLTAGE
VS
FREQUENCY**

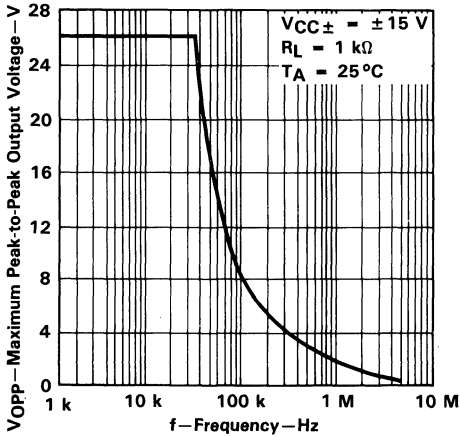


FIGURE 8

**OP-37
MAXIMUM PEAK-TO-PEAK
OUTPUT VOLTAGE
VS
FREQUENCY**

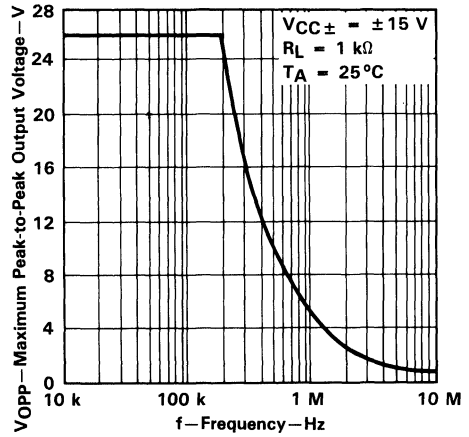


FIGURE 9

**OP-27A, OP-27E, OP-37A, OP-37E
LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
VS
TOTAL SUPPLY VOLTAGE**

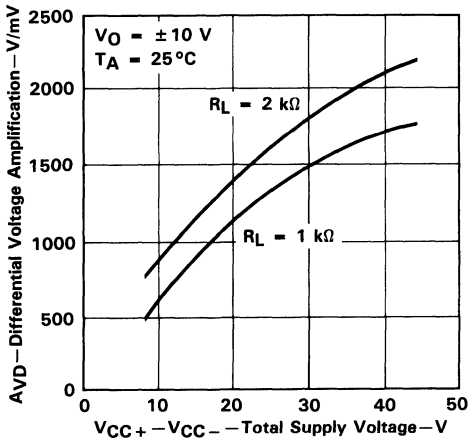


FIGURE 10

**OP-27A, OP-27E, OP-37A, OP-37E
LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
VS
LOAD RESISTANCE**

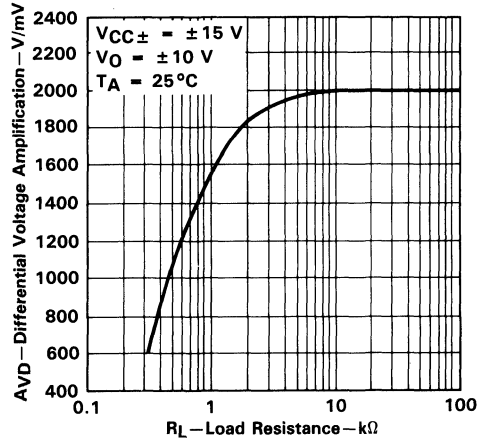


FIGURE 11

2
Operational Amplifiers

**OP-27A, OP-27C, OP-27E, OP-27G
 OP-37A, OP-37C, OP-37E, OP-37G
 LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS

**2
Operational Amplifiers**

**OP-27
 LARGE-SIGNAL DIFFERENTIAL
 VOLTAGE AMPLIFICATION AND PHASE SHIFT
 VS
 FREQUENCY**

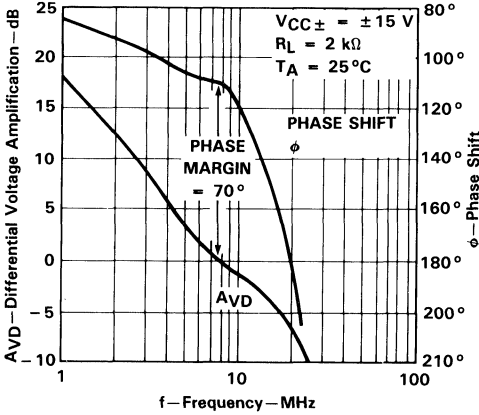


FIGURE 12

**OP-37
 LARGE-SIGNAL DIFFERENTIAL
 VOLTAGE AMPLIFICATION AND PHASE SHIFT
 VS
 FREQUENCY**

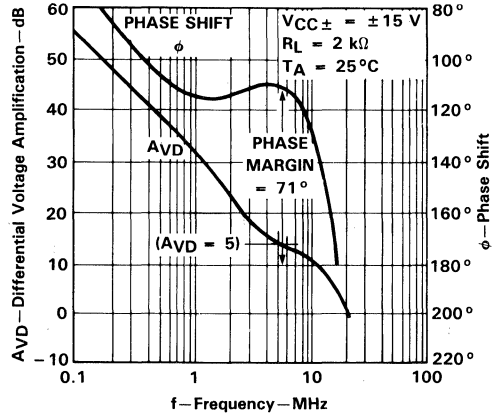


FIGURE 13

**OP-27A, OP-27E, OP-37A, OP-37E
 LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 VS
 FREQUENCY**

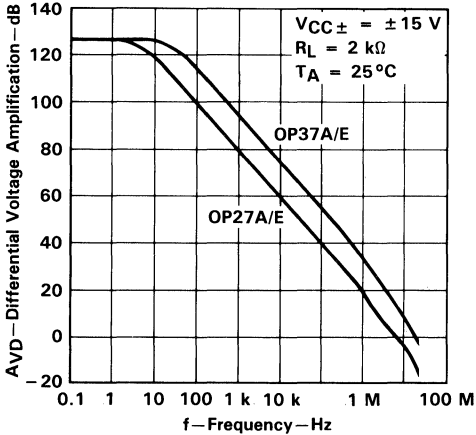


FIGURE 14

**OP-27A, OP-27E, OP-37A, OP-37E
 COMMON-MODE REJECTION RATIO
 VS
 FREQUENCY**

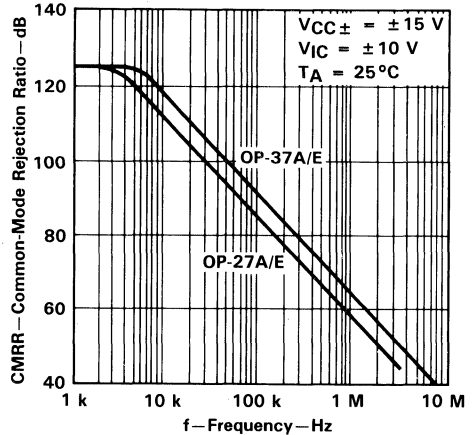


FIGURE 15

**OP-27A, OP-27C, OP-27E, OP-27G
OP-37A, OP-37C, OP-37E, OP-37G**
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

SUPPLY VOLTAGE REJECTION RATIO
vs
FREQUENCY

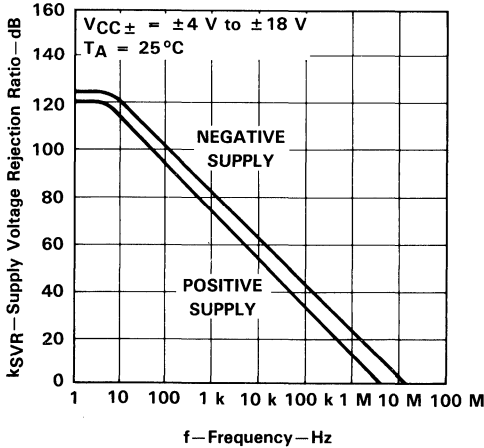


FIGURE 16

SLEW RATE
vs
FREE-AIR TEMPERATURE

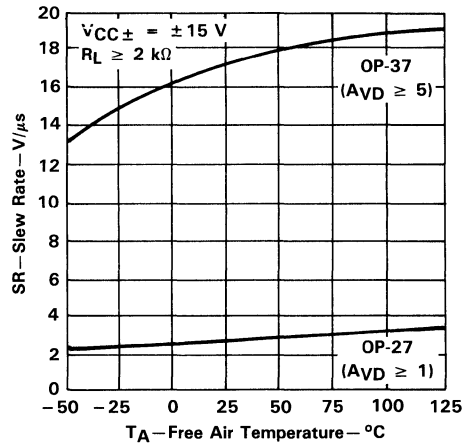


FIGURE 17

OP-37
SLEW RATE
vs
SUPPLY VOLTAGE

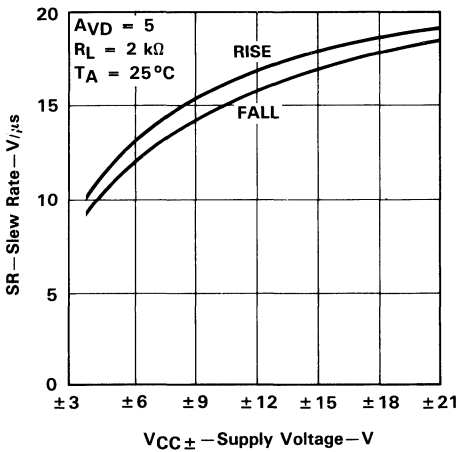


FIGURE 18

OP-37
SLEW RATE
vs
LOAD RESISTANCE

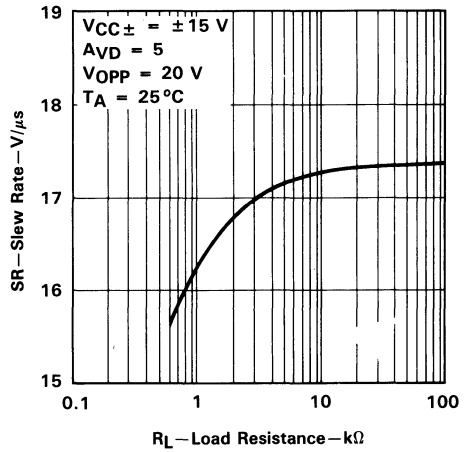


FIGURE 19

2

Operational Amplifiers

†Data for temperatures below -25°C and above 85°C are applicable to the OP-27A, OP-27C, OP-37A, and OP-37C only.

**OP-27A, OP-27C, OP-27E, OP-27G
 OP-37A, OP-37C, OP-37E, OP-37G
 LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS†

**OP-27
 PHASE MARGIN AND
 GAIN-BANDWIDTH PRODUCT
 VS
 FREE-AIR TEMPERATURE**

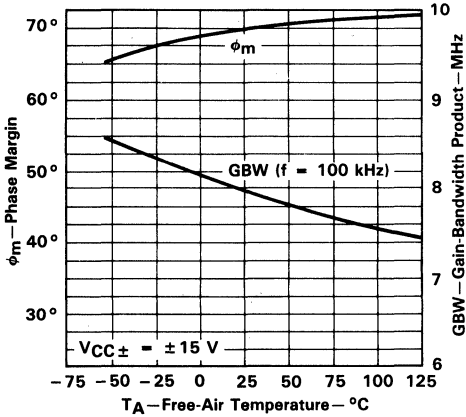


FIGURE 20

**OP-37
 PHASE MARGIN AND
 GAIN-BANDWIDTH PRODUCT
 VS
 FREE-AIR TEMPERATURE**

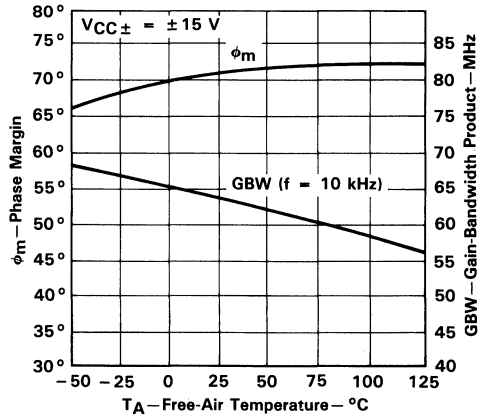


FIGURE 21

**EQUIVALENT INPUT NOISE VOLTAGE
 VS
 BANDWIDTH**

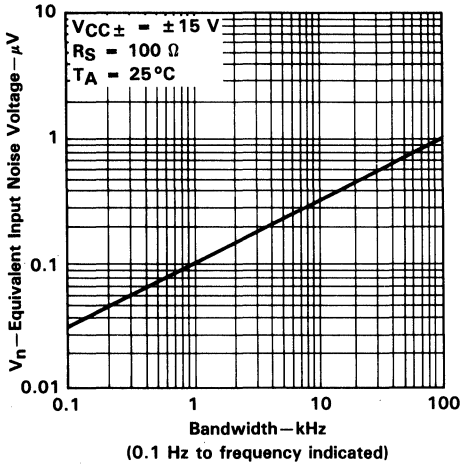


FIGURE 22

**TOTAL EQUIVALENT INPUT NOISE VOLTAGE
 VS
 SOURCE RESISTANCE**

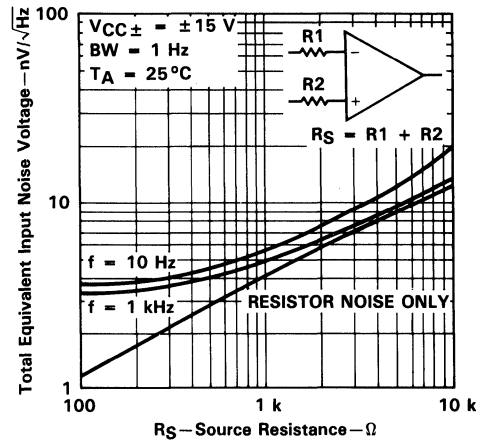


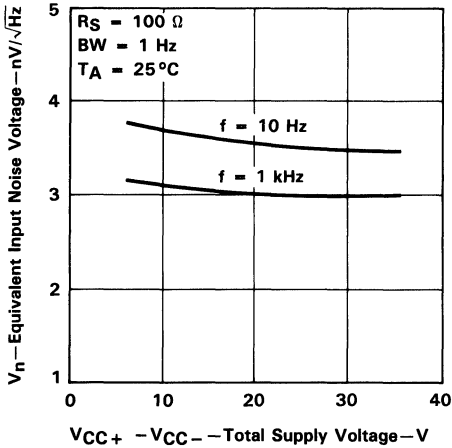
FIGURE 23

†Data for temperatures below -25°C and above 85°C are applicable to the OP-27A, OP-27C, OP-37A, and OP-37C only.

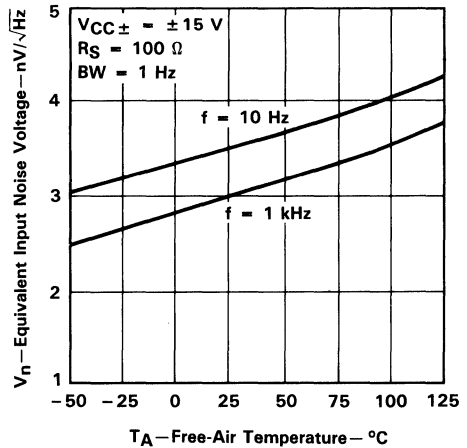
**OP-27A, OP-27C, OP-27E, OP-27G
OP-37A, OP-37C, OP-37E, OP-37G**
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

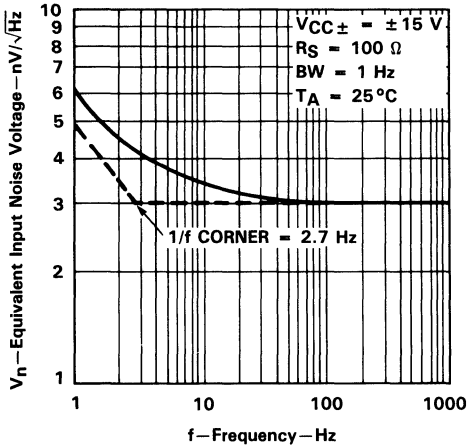
OP-27A, OP-27E, OP-37A, OP-37E
EQUIVALENT INPUT NOISE VOLTAGE
vs
TOTAL SUPPLY VOLTAGE



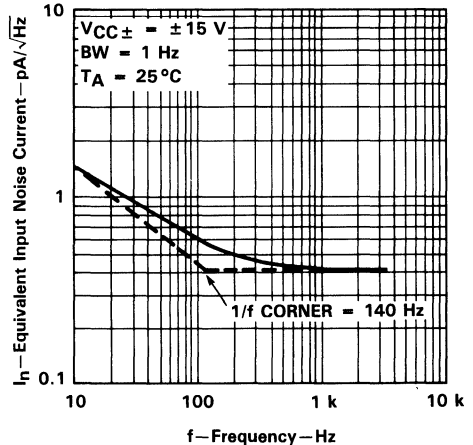
OP-27A, OP-27E, OP-37A, OP-37E
EQUIVALENT INPUT NOISE VOLTAGE
vs
FREE-AIR TEMPERATURE



OP-27A, OP-27E, OP-37A, OP-37E
EQUIVALENT INPUT NOISE VOLTAGE
vs
FREQUENCY



EQUIVALENT INPUT NOISE CURRENT
vs
FREQUENCY



†Data for temperatures below -25°C and above 85°C are applicable to the OP-27A, OP-27C, OP-37A, and OP-37C only.

**OP-27A, OP-27C, OP-27E, OP-27G
 OP-37A, OP-37C, OP-37E, OP-37G
 LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS†

**SHORT-CIRCUIT OUTPUT CURRENT
 VS
 ELAPSED TIME**

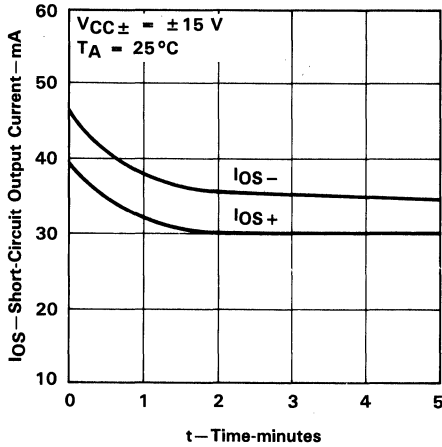


FIGURE 28

**SUPPLY CURRENT
 VS
 TOTAL SUPPLY VOLTAGE**

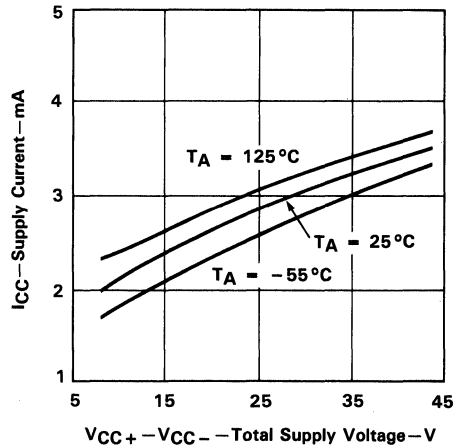


FIGURE 29

**OP-27
 VOLTAGE FOLLOWER
 SMALL-SIGNAL
 PULSE RESPONSE**

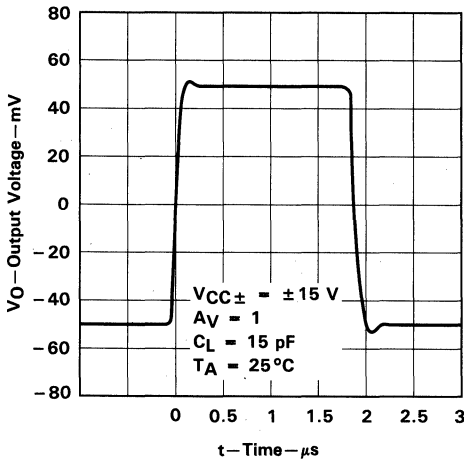


FIGURE 30

**OP-27
 VOLTAGE FOLLOWER
 LARGE-SIGNAL
 PULSE RESPONSE**

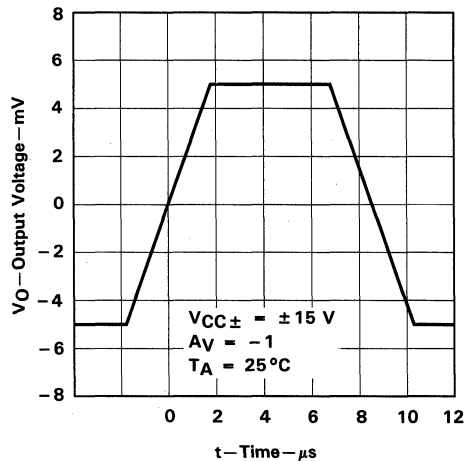
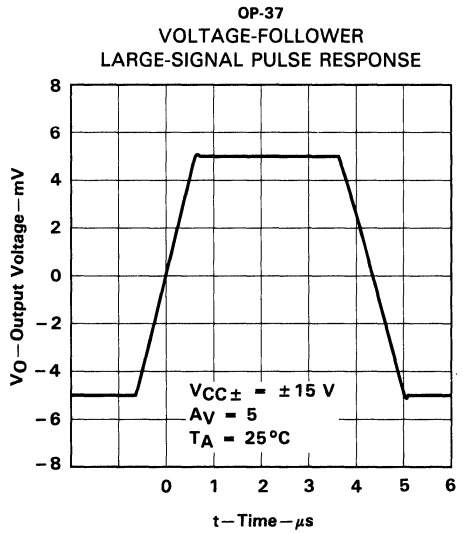
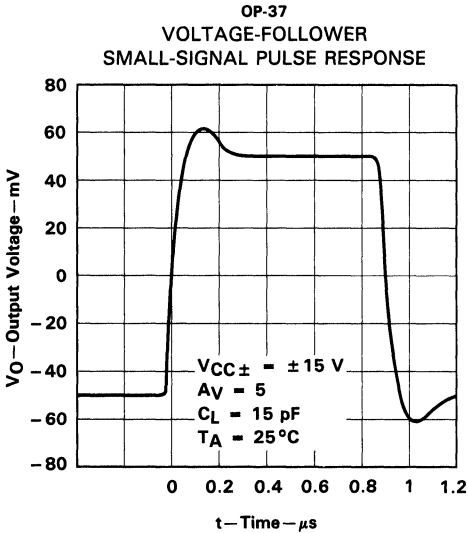


FIGURE 31

†Data for temperatures below -25°C and above 85°C are applicable to the OP-27A, OP-27C, OP-37A, and OP-37C only.

OP-27A, OP-27C, OP-27E, OP-27G
OP-37A, OP-37C, OP-37E, OP-37G
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS



TYPICAL APPLICATION DATA

general

The OP-27 and OP-37 series devices may be inserted directly into OP-07, OP-05, μ A725, and SE5534 sockets with or without removing external compensation or nulling components. In addition, the OP-27 and OP-37 may be fitted to μ A741 sockets by removing or modifying external nulling components.

noise testing

Figure 34 shows a test circuit for 0.1-Hz to 10-Hz peak-to-peak noise measurement of the OP-27 and OP-37. The frequency response of this noise tester indicates that the 0.1-Hz corner is defined by only one zero. Because the time limit acts as an additional zero to eliminate noise contributions from the frequency band below 0.1 Hz, the test time to measure 0.1-Hz to 10-Hz noise should not exceed 10 seconds.

TYPICAL APPLICATION DATA

When measuring noise on a large number of units, a noise-voltage density test is recommended. A 10-Hz noise-voltage density measurement correlates well with a 0.1-Hz to 10-Hz peak-to-peak noise reading since both results are determined by the white noise and the location of the 1/f corner frequency.

Figure 35 shows a circuit measuring current noise and the formula for calculating current noise.

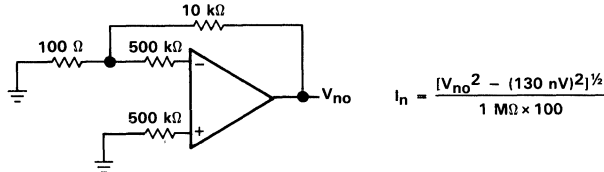


FIGURE 35. CURRENT NOISE TEST CIRCUIT AND FORMULA

offset voltage adjustment

The input offset voltage and temperature coefficient of the OP-27 and OP-37 are permanently trimmed to a low level at wafer testing. However, if further adjustment of V_{IO} is necessary, using a 10-kΩ nulling potentiometer, as shown in Figure 36, does not degrade the temperature coefficient $\propto V_{IO}$. Trimming to a value other than zero creates an $\propto V_{IO}$ of $V_{IO}/300 \mu\text{V}/^\circ\text{C}$. For example, if V_{IO} is adjusted to 300 μV , the change in $\propto V_{IO}$ is 1 $\mu\text{V}/^\circ\text{C}$.

The adjustment range with a 10-kΩ potentiometer is approximately $\pm 2.5 \text{ mV}$. If a smaller adjustment range is needed, the sensitivity and resolution of the nulling can be improved by using a smaller potentiometer in conjunction with fixed resistors. The example in Figure 37 has an approximate null range of $\pm 200 \mu\text{V}$.

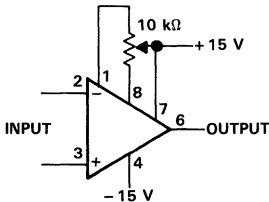


FIGURE 36. STANDARD INPUT OFFSET VOLTAGE ADJUSTMENT

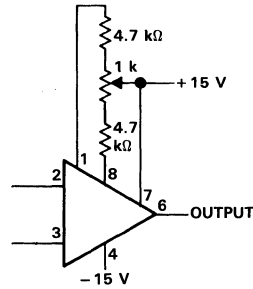


FIGURE 37. INPUT OFFSET VOLTAGE ADJUSTMENT WITH IMPROVED SENSITIVITY

offset voltage and drift

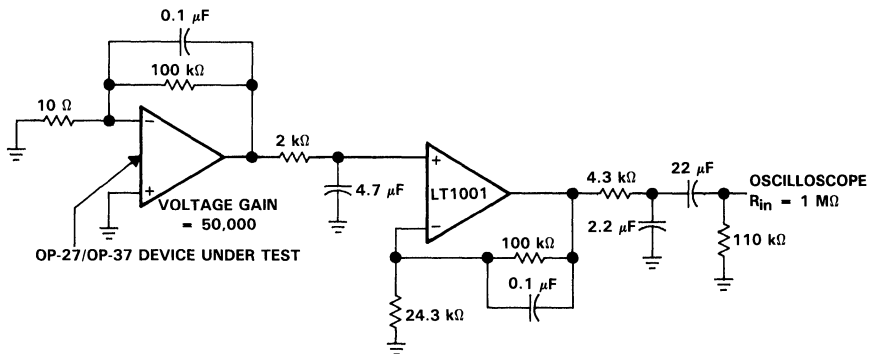
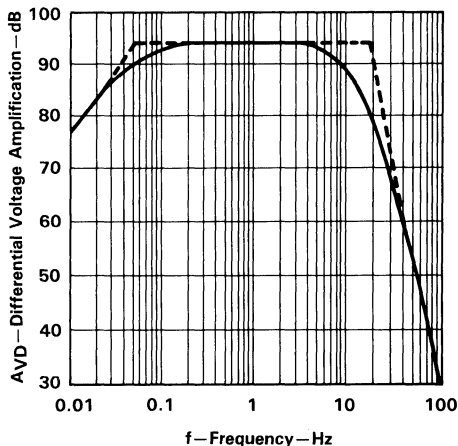
Unless proper care is exercised, thermocouple effects caused by temperature gradients across dissimilar metals at the contacts to the input terminals can exceed the inherent temperature coefficient $\propto V_{IO}$ of the amplifier. Air currents should be minimized, package leads should be short, and the two input leads should be close together and at the same temperature.

The circuit shown in Figure 38 measures offset voltage. This circuit can also be used as the burn-in configuration for the OP-27 and OP-37, with the supply voltage increased to $\pm 20 \text{ V}$, $R_1 = R_3 = 10 \text{ k}\Omega$, $R_2 = 200 \Omega$, and $A_{VD} = 100$.

**OP-27A, OP-27C, OP-27E, OP-27G
 OP-37A, OP-37C, OP-37E, OP-37G
 LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS**

TYPICAL APPLICATION DATA

noise testing (continued)



NOTE: All capacitor values are for non-polarized capacitors only.

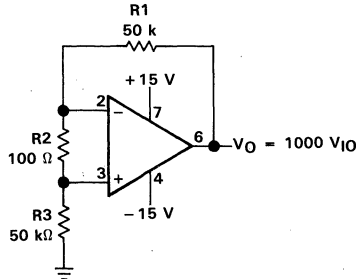
FIGURE 34. 0.1-Hz TO 10-Hz PEAK-TO-PEAK NOISE TEST CIRCUIT AND FREQUENCY RESPONSE

Measuring the typical 80-nV peak-to-peak noise performance of the OP-27 and OP-37 requires the following special test precautions:

1. The device should be warmed up for at least five minutes. As the operational amplifier warms up, the offset voltage typically changes $4 \mu\text{V}$ due to the chip temperature increasing from 10°C to 20°C starting from the moment the power supplies are turned on. In the 10-s measurement interval, these temperature-induced effects can easily exceed tens of nanovolts.
2. For similar reasons, the device should be well shielded from air currents to eliminate the possibility of thermoelectric effects in excess of a few nanovolts, which would invalidate the measurements.
3. Sudden motion in the vicinity of the device should be avoided, as it produces a feedthrough effect that increases observed noise.

**OP-27A, OP-27C, OP-27E, OP-27G
 OP-37A, OP-37C, OP-37E, OP-37G
 LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS**

TYPICAL APPLICATION DATA



NOTE: Resistors must have low thermoelectric potential.

FIGURE 38. TEST CIRCUIT FOR OFFSET VOLTAGE AND OFFSET VOLTAGE TEMPERATURE COEFFICIENT

unity gain buffer applications

The resulting output waveform when $R_f \leq 100 \Omega$ and the input is driven with a fast large-signal pulse ($> 1 \text{ V}$) is shown in the pulsed-operation diagram in Figure 39.

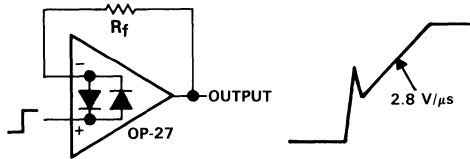
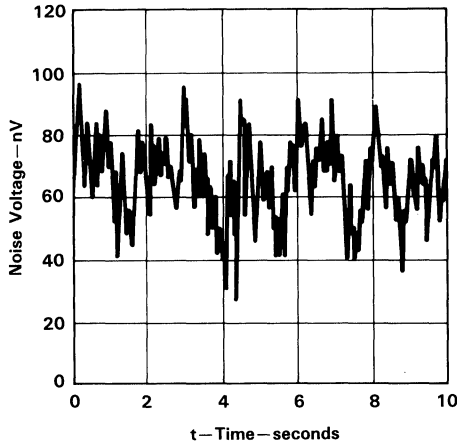


FIGURE 39. PULSED OPERATION

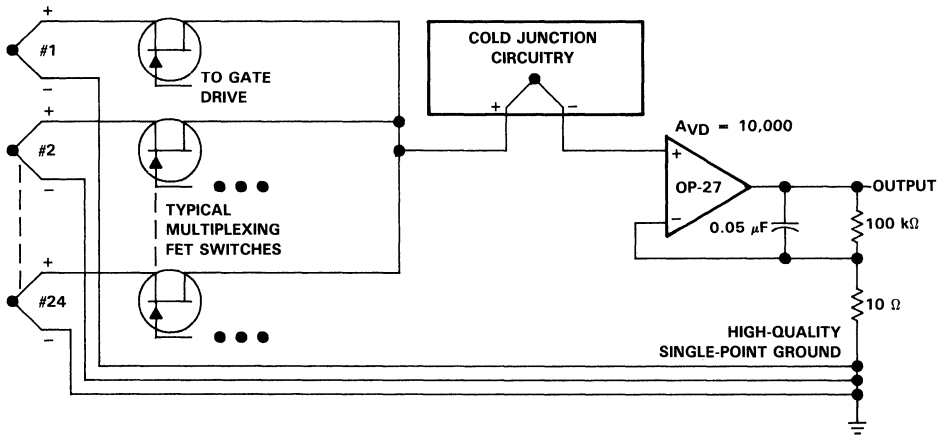
During the initial (fast-feedthrough-like) portion of the output waveform, the input protection diodes effectively short the output to the input, and a current, limited only by the output short-circuit protection, is drawn by the signal generator. When $R_f \geq 500 \Omega$, the output is capable of handling the current requirements (load current $\leq 20 \text{ mA}$ at 10 V), the amplifier stays in its active mode, and a smooth transition occurs. When $R_f > 2 \text{ k}\Omega$, a pole is created with R_f and the amplifier's input capacitance, creating additional phase shift and reducing the phase margin. A small capacitor (20 pF to 50 pF) in parallel with R_f eliminates this problem.

OP-27A, OP-27C, OP-27E, OP-27G
OP-37A, OP-37C, OP-37E, OP-37G
LOW-NOISE, HIGH-SPEED, PRECISION OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION



TYPE S THERMOCOUPLES
 $5.4 \mu\text{V}/^\circ\text{C}$ AT 0°C



NOTE A: If 24 channels are multiplexed per second, and the output is required to settle to 0.1% accuracy, the amplifier's bandwidth cannot be limited to less than 30 Hz. The peak-to-peak noise contribution of the OP-27 will still be only $0.11 \mu\text{V}$, which is equivalent to an error of only 0.02°C .

FIGURE 40. LOW-NOISE, MULTIPLEXED THERMOCOUPLE AMPLIFIER AND 0.1-Hz TO 10-Hz PEAK-TO-PEAK NOISE VOLTAGE

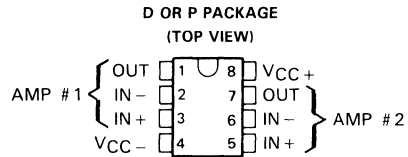
2

Operational Amplifiers

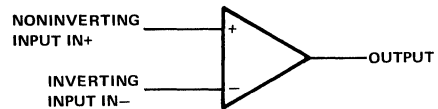
RC4559 DUAL HIGH-PERFORMANCE OPERATIONAL AMPLIFIER

D2785, OCTOBER 1983—REVISED JUNE 1988

- Matched Gain and Offset Between Amplifiers
- Unity-Gain Bandwidth . . . 3 MHz Min
- Slew Rate . . . 1.5 V/ns Min
- Low Equivalent Input Noise Voltage . . . $2 \mu\text{V}/\sqrt{\text{Hz}}$ Max (20 Hz to 20 kHz)
- No Frequency Compensation Required
- No Latch Up
- Wide Common-Mode Voltage Range
- Low Power Consumption
- Designed to be Interchangeable with Raytheon RC4559



symbol (each amplifier)



AVAILABLE OPTIONS

SYMBOLIZATION		OPERATING TEMPERATURE RANGE	V_{IO} MAX at 25°C
DEVICE	PACKAGE SUFFIX		
RC4559	D,P	-0°C to 70°C	6 mV

The D packages are available taped and reeled. Add the suffix R to the device type when ordering. (i.e., RC4559DR)

description

The RC4559 is a dual high-performance operational amplifier. The high common-mode input voltage and the absence of latch-up make this amplifier ideal for low-noise signal applications such as audio preamplifiers and signal conditioners. This amplifier features a guaranteed dynamic performance and output drive capability that far exceeds that of the general-purpose type amplifiers.

The RC4559 is characterized for operation from 0°C to 70°C.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage V_{CC+} (see Note 1)	18 V
Supply voltage V_{CC-} (see Note 1)	-18 V
Differential input voltage (see Note 2)	± 30 V
Input voltage (any input, see Notes 1 and 3)	± 15 V
Duration of output short-circuit to ground, one amplifier at a time (see Note 4)	unlimited
Continuous total dissipation	500 mW
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 125°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the zero reference level (ground) of the supply voltages where the zero reference level is the midpoint between V_{CC+} and V_{CC-} .
2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 volts, whichever is less.
4. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS
INSTRUMENTS**

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Operational Amplifiers

RC4559
DUAL HIGH-PERFORMANCE OPERATIONAL AMPLIFIER

electrical characteristics at specified free-air temperature, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$

PARAMETER		TEST CONDITIONS†		MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	$V_O = 0$	25°C	2	6	mV	
			0°C to 70°C	7.5			
I_{IO}	Input offset current	$V_O = 0$	25°C	5	100	nA	
			0°C to 70°C	200			
I_{IB}	Input bias current	$V_O = 0$	25°C	40	250	nA	
			0°C to 70°C	500			
V_I	Input voltage range		25°C	±12	±13	V	
V_{OM}	Maximum peak output voltage swing		$R_L \geq 3\text{ k}\Omega$	25°C	±12	±13	V
			$R_L = 600\ \Omega$	25°C	±9.5	±10	
			$R_L \geq 2\text{ k}\Omega$	0°C to 70°C	±10		
A_{VD}	Large-signal differential voltage amplification	$V_O = \pm 10\text{ V}$, $R_L = 2\text{ k}\Omega$	25°C	20	300	V/mV	
			0°C to 70°C	15			
B_{OM}	Maximum output-swing bandwidth	$V_{OPP} = 20\text{ V}$, $R_L = 2\text{ k}\Omega$	25°C	24	32	kHz	
B_1	Unity-gain bandwidth		25°C	3	4	MHz	
r_i	Input resistance		25°C	0.3	1	M Ω	
CMRR	Common-mode rejection ratio	$V_O = 0$	25°C	80	100	dB	
k_{SVS}	Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	$V_O = 0$	25°C	10	75	$\mu\text{V}/\text{V}$	
V_n	Equivalent input noise voltage (closed-loop)	$A_{VD} = 100$, $R_S = 1\text{ k}\Omega$, $f = 20\text{ Hz to } 20\text{ kHz}$	25°C	1.4	2	μV	
I_n	Equivalent input noise current	$f = 20\text{ Hz to } 20\text{ kHz}$	25°C	25		pA	
I_{CC}	Supply current (both amplifiers)	No load, No signal	25°C	3.3	5.6	mA	
			0°C	4	6.6		
			70°C	3	5		
V_{O1}/V_{O2}	Crosstalk attenuation	$A_{VD} = 100$, $R_S = 1\text{ k}\Omega$, $f = 10\text{ kHz}$	25°C	90		dB	
			25°C	90			

†All characteristics are specified under open-loop operation, unless otherwise noted.

matching characteristics at $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{IO}	$V_O = 0$		±0.2		mV
I_{IO}	$V_O = 0$		±7.5		nA
I_{IB}	$V_O = 0$		±15		nA
A_{VD}	$V_O = \pm 10\text{ V}$, $R_L = 2\text{ k}\Omega$		±1		dB

operating characteristics, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_r	$V_I = 20\text{ mV}$, $R_L = 2\text{ k}\Omega$		80		μs
			Overshoot	18%	
SR	$V_I = 10\text{ V}$, $C_L = 100\text{ pF}$	1.5	2		V/ μs

2 Operational Amplifiers

RM4136, RV4136, RC4136 QUAD HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

D2142, MARCH 1978—REVISED NOVEMBER 1988

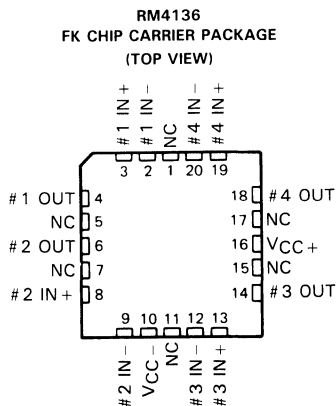
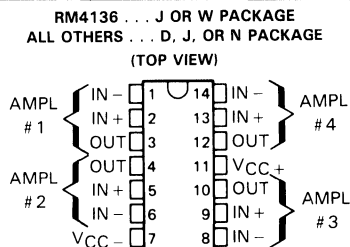
- Continuous-Short-Circuit Protection
- Wide Common-Mode and Differential Voltage Ranges
- No Frequency Compensation Required
- Low Power Consumption
- No Latch-Up
- Unity Gain Bandwidth 3 MHz Typical
- Gain and Phase Match Between Amplifiers
- Designed to be Interchangeable with Raytheon RM4136, RV4136, and RC4136
- Low Noise . . . 8 nV/ $\sqrt{\text{Hz}}$ Typ at 1 kHz

description

The RM4136, RV4136, and RC4136 are quad high-performance operational amplifiers with each amplifier electrically similar to the uA741 except that offset null capability is not provided.

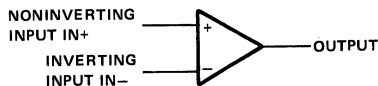
The high common-mode input voltage range and the absence of latch-up make these amplifiers ideal for voltage-follower applications. The devices are short-circuit protected and the internal frequency compensation ensures stability without external components.

The RM4136 is characterized for operation over the full military temperature range of -55°C to 125°C , the RV4136 is characterized for operation from -40°C to 85°C , and the RC4136 is characterized for operation from 0°C to 70°C .



NC—No internal connection

symbol (each amplifier)



AVAILABLE OPTIONS

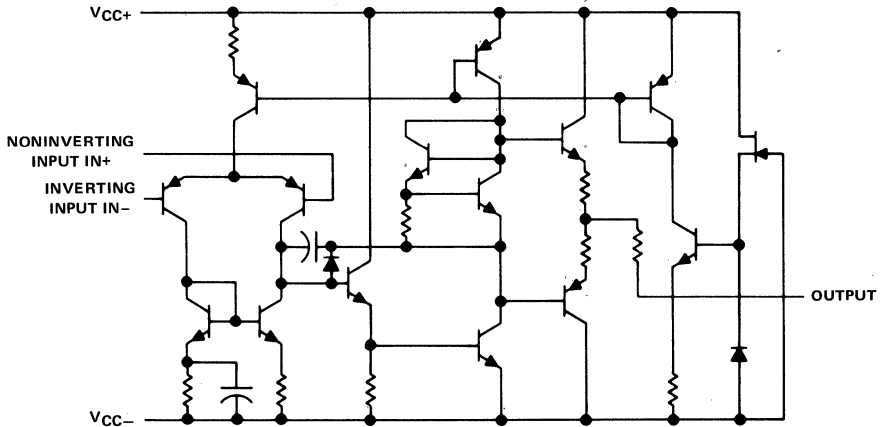
T _A	V _{IO} MAX at 25°C	PACKAGE				
		SMALL-OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	FLAT (W)
0°C to 70°C	6 mV	RC4136D	—	RC4136J	RC4136N	—
-40°C to 85°C	6 mV	RV4136D	—	RV4136J	RV4136N	—
-55°C to 125°C	4 mV	—	RM4136FK	RM4136J	—	RM4136W

The D packages are available taped and reeled. Add the suffix R to the device type. (e.g., RC4136DR)

2
Operational Amplifiers

RM4136, RV4136, RC4136 QUAD HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

schematic (each amplifier)



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	RM4136	RV4136	RC4136	UNIT
Supply voltage V_{CC+} (see Note 1)	22	18	18	V
Supply voltage V_{CC-} (see Note 1)	-22	-18	-18	V
Differential input voltage (see Note 2)	± 30	± 30	± 30	V
Input voltage (any input, see Notes 1 and 3)	± 15	± 15	± 15	V
Duration of output short-circuit to ground, one amplifier at a time (see Note 4)	unlimited	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table			
Operating free-air temperature range	-55 to 125	-40 to 85	0 to 70	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Case temperature for 60 seconds	FK package 260 $^{\circ}\text{C}$			
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J or W package 300 $^{\circ}\text{C}$			
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or N package 260 $^{\circ}\text{C}$			

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 volts, whichever is less.
 4. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^{\circ}\text{C}$ POWER RATING	$T_A = 85^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
D	800 mW	7.6 mW/ $^{\circ}\text{C}$	45 $^{\circ}\text{C}$	608 mW	494 mW	—
FK	800 mW	11.0 mW/ $^{\circ}\text{C}$	77 $^{\circ}\text{C}$	800 mW	715 mW	275 mW
J (RM4136)	800 mW	11.0 mW/ $^{\circ}\text{C}$	77 $^{\circ}\text{C}$	800 mW	715 mW	275 mW
J (others)	800 mW	8.2 mW/ $^{\circ}\text{C}$	52 $^{\circ}\text{C}$	656 mW	533 mW	—
N	800 mW	9.2 mW/ $^{\circ}\text{C}$	63 $^{\circ}\text{C}$	736 mW	598 mW	—
W	800 mW	8.0 mW/ $^{\circ}\text{C}$	50 $^{\circ}\text{C}$	640 mW	520 mW	200 mW

RM4136, RV4136, RC4136 QUAD HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

2
Operational Amplifiers

electrical characteristics at specified free-air temperature, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$

PARAMETER	TEST CONDITIONS†	RM4136			RV4136			RC4136			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0$	25°C	0.5	4		0.5	6		0.5	6	mV
		Full range		6		7.5		7.5			
I_{IO} Input offset current	$V_O = 0$	25°C	5	150		5	200		5	200	nA
		Full range		500		500		300			
I_{IB} Input bias current	$V_O = 0$	25°C	140	400		140	500		140	500	nA
		Full range		1500		1500		800			
V_I Input voltage range		25°C	±12	±14		±12	±14		±12	±14	V
V_{OM} Maximum peak output voltage swing	$R_L = 10\text{ k}\Omega$	25°C	±12	±14		±12	±14		±12	±14	V
	$R_L = 2\text{ k}\Omega$	25°C	±10	±13		±10	±13		±10	±13	
	$R_L \geq 2\text{ k}\Omega$	Full range	±10			±10			±10		
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 10\text{ V}$, $R_L \geq 2\text{ k}\Omega$	25°C	50	350		20	300		20	300	V/mV
		Full range	25			15			15		
B_1 Unity-gain bandwidth		25°C	3.5			3			3		MHz
r_i Input resistance		25°C	0.3	5		0.3	5		0.3	5	MΩ
CMRR Common-mode rejection ratio	$V_O = 0$, $R_S = 50\text{ }\Omega$	25°C	70	90		70	90		70	90	dB
k_{SVS} Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	$V_{CC} = \pm 9\text{ V}$ to $\pm 15\text{ V}$, $V_O = 0$	25°C	30	150		30	150		30	150	$\mu\text{V/V}$
V_n Equivalent input noise voltage (closed-loop)	$A_{VD} = 100$, BW = 1 Hz, f = 1 kHz, $R_S = 100\text{ }\Omega$	25°C	8			8			8		nV/√Hz
I_{CC} Supply current (All four amplifiers)	$V_O = 0$, No load	25°C	5	11.3		5	11.3		5	11.3	mA
		MIN T_A	6	13.3		6	13.7		6	13.7	
		MAX T_A	4.5	10		4.5	10		4.5	10	
P_D Total power dissipation (All four amplifiers)	$V_O = 0$, No load	25°C	150	340		150	340		150	340	mW
		MIN T_A	180	400		180	400		180	400	
		MAX T_A	135	300		135	300		135	300	
V_{O1}/V_{O2} Crosstalk attenuation	$A_{VD} = 100$, f = 10 kHz, $R_S = 1\text{ k}\Omega$	25°C	105			105			105		dB

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range is -55°C to 125°C for RM4136, -40°C to 85°C for RV4136, and 0°C to 70°C for RC4136.

operating characteristics, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	RM4136			RV4136, RC4136			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
t_r Rise time	$V_I = 20\text{ mV}$, $R_L = 2\text{ k}\Omega$,		0.13			0.13		μs
Overshoot factor	$C_L = 100\text{ pF}$		5%			5%		
SR Slew rate at unity gain	$V_I = 10\text{ V}$, $C_L = 100\text{ pF}$		1.7			1.7		V/ μs

2

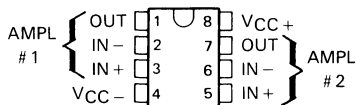
Operational Amplifiers

RM4558, RV4558, RC4558 DUAL HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

D2141, MARCH 1976—REVISED DECEMBER 1988

- Continuous-Short-Circuit Protection
- Wide Common-Mode and Differential Voltage Ranges
- No Frequency Compensation Required
- Low Power Consumption
- No Latch-up
- Unity Gain Bandwidth 3 MHz Typical
- Gain and Phase Match Between Amplifiers
- Low Noise . . . 8 nV/ $\sqrt{\text{Hz}}$ Typ at 1 kHz
- Designed to be Interchangeable with Raytheon RM4558, RV4558, and RC4558

D, JG, OR P PACKAGE
(TOP VIEW)



description

The RM4558, RV4558, and RC4558 are dual general-purpose operational amplifiers with each half electrically similar to $\mu\text{A}741$ except that offset null capability is not provided.

The high common-mode input voltage range and the absence of latch-up make these amplifiers ideal for voltage-follower applications. The devices are short-circuit protected and the internal frequency compensation ensures stability without external components.

The RM4558 is characterized for operation over the full military temperature range of -55°C to 125°C , the RV4558 is characterized for operation from -40°C to 85°C , and the RC4558 is characterized for operation from 0°C to 70°C .

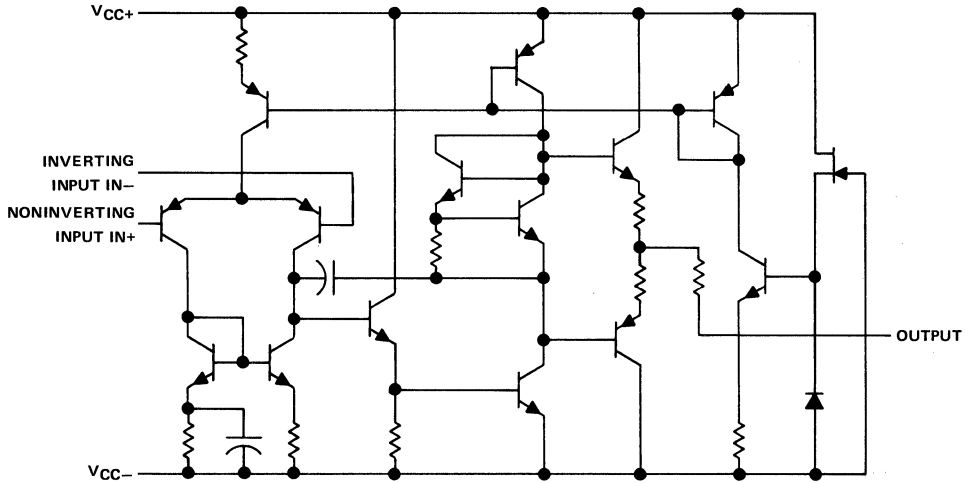
AVAILABLE OPTIONS

T_A	$V_{IO \text{ MAX}}$ at 25°C	PACKAGES		
		SMALL OUTLINE (D)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	6 mV	RC4558D	RC4558JG	RC4558P
-40°C to 85°C	6 mV	RV4558D	RV4558JG	RV4558P
-55°C to 125°C	5 mV	—	RM4558JG	—

The D packages are available taped and reeled. Add the suffix "R" to the device type (e.g., RC4558DR).

RM4558, RV4558, RC4558 DUAL HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

schematic (each amplifier)



absolute maximum ratings over operating free-air temperature (unless otherwise noted)

	RM4558	RV4558	RC4558	UNIT
Supply voltage V_{CC+} (see Note 1)	22	18	18	V
Supply voltage V_{CC-} (see Note 1)	-22	-18	-18	V
Differential input voltage (see Note 2)	± 30	± 30	± 30	V
Input voltage (any input, see Notes 1 and 3)	± 15	± 15	± 15	V
Duration of output short-circuit to ground, one amplifier at a time (see Note 4)	unlimited	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table			
Operating free-air temperature range	-55 to 125	-40 to 85	0 to 70	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300	300	300	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package		260	260	$^{\circ}\text{C}$

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 volts, whichever is less.
 4. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$	DERATING FACTOR		$T_A = 70^{\circ}\text{C}$	$T_A = 85^{\circ}\text{C}$	$T_A = 85^{\circ}\text{C}$
	POWER RATING	ABOVE $T_A = 25^{\circ}\text{C}$	DERATE ABOVE T_A	POWER RATING	POWER RATING	POWER RATING
D	680 mW	5.8 mW/ $^{\circ}\text{C}$	33 $^{\circ}\text{C}$	464 mW	377 mW	N/A
JG (RM4558)	680 mW	8.4 mW/ $^{\circ}\text{C}$	69 $^{\circ}\text{C}$	672 mW	546 mW	210 mW
JG (RV4558)	680 mW	6.6 mW/ $^{\circ}\text{C}$	47 $^{\circ}\text{C}$	528 mW	429 mW	N/A
(RC4558)	680 mW	8.0 mW/ $^{\circ}\text{C}$	65 $^{\circ}\text{C}$	640 mW	520 mW	N/A
P	680 mW					

RM4558, RV4558, RC4558 DUAL HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$

PARAMETER	TEST CONDITIONS†	RM4558			RV4558			RC4558			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0$	25°C	0.5	5	0.5	6	0.5	6	mV		
		Full range	6			7.5				7.5	
I_{IO} Input offset current	$V_O = 0$	25°C	5			5			5	nA	
		Full range	500			500			300		
I_{IB} Input bias current	$V_O = 0$	25°C	140	500	140	500	150	500	nA		
		Full range	1500			1500				800	
V_{ICR} Common-mode input voltage range		25°C	±12	±14	±12	±14	±12	±14	V		
V_{OM} Maximum output voltage swing	$R_L = 10\text{ k}\Omega$	25°C	±12	±14	±12	±14	±12	±14	V		
	$R_L = 2\text{ k}\Omega$	25°C	±10	±13	±10	±13	±10	±13			
	$R_L \geq 2\text{ k}\Omega$	Full range	±10			±10				±10	
A_{VD} Large-signal differential voltage amplification	$R_L \geq 2\text{ k}\Omega$, $V_O = \pm 10\text{ V}$	25°C	50	350	20	300	20	300	V/mV		
		Full range	25			15				15	
B_1 Unity-gain bandwidth		25°C	2	3.5	3			3	MHz		
r_i Input resistance		25°C	0.3	5	0.3	5	0.3	5	M Ω		
CMRR Common-mode rejection ratio		25°C	70	90	70	90	70	90	dB		
k_{SVS} Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	$V_{CC} = \pm 15\text{ V}$ to $\pm 9\text{ V}$	25°C	30			30			150	$\mu\text{V/V}$	
V_n Equivalent input noise voltage (closed-loop)	$A_{VD} = 100$, $R_S = 100\ \Omega$, $f = 1\text{ kHz}$, $BW = 1\text{ Hz}$	25°C	8			8			8	nV/ $\sqrt{\text{Hz}}$	
I_{CC} Supply current (Both amplifiers)	No load, $V_O = 0$	25°C	2.5	5.6	2.5	5.6	2.5	5.6	mA		
		MIN T_A	3			3				6.6	
		MAX T_A	2			2.3				5	
P_D Total power dissipation (Both amplifiers)	No load, $V_O = 0$	25°C	75	170	75	170	75	170	mW		
		MIN T_A	90			90				200	
		MAX T_A	60			70				150	
V_{O1}/V_{O2} Crosstalk attenuation	Open loop $A_{VD} = 100$	$R_S = 1\text{ k}\Omega$, $f = 10\text{ kHz}$	25°C	85			85			dB	
		25°C	105			105			105		

†All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range is -55°C to 125°C for RM4558, -40°C to 85°C for RV4558, and 0°C to 70°C for RC4558.

operating characteristics, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	RM4558			RV4558			RC4558			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
t_r Rise time	$V_I = 20\text{ mV}$, $R_L = 2\text{ k}\Omega$,	0.13			0.13			0.13			ns
	$C_L = 100\text{ pF}$	5%			5%			5%			
SR Slew rate at unity gain	$V_I = 10\text{ V}$, $R_L = 2\text{ k}\Omega$, $C_L = 100\text{ pF}$	1.1	1.7		1.1	1.7		1.1	1.7	V/ μs	

2

Operational Amplifiers

2

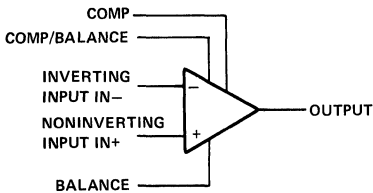
Operational Amplifiers

SE5534, SE5534A, NE5534, NE5534A LOW-NOISE OPERATIONAL AMPLIFIERS

D2532, JULY 1979—REVISED MAY 1988

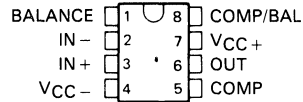
- Equivalent Input Noise Voltage
3.5 nV/ $\sqrt{\text{Hz}}$ Typ
- Unity-Gain Bandwidth 10 MHz Typ
- Common-Mode Rejection Ratio
100 dB Typ
- High DC Voltage Gain 100 V/mV Typ
- Peak-to-Peak Output Voltage Swing
32 V Typ with $V_{CC\pm} = \pm 18$ V and
 $R_L = 600 \Omega$
- High Slew Rate 13 V/ μs Typ
- Wide Supply Voltage Range
 ± 3 V to ± 20 V
- Low Harmonic Distortion
- Designed to be Interchangeable with Signetics
SE5534, SE5534A, NE5534, and NE5534A

symbol



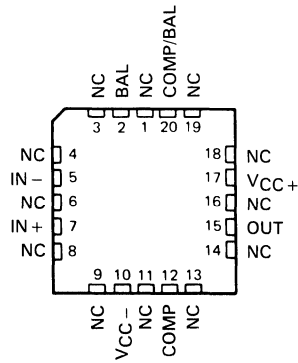
SE5534, SE5534A . . . JG PACKAGE
NE5534, NE5534A . . . D, JG OR P PACKAGE

(TOP VIEW)



SE5534, SE5534A
FK CHIP CARRIER PACKAGE

(TOP VIEW)



AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE			
		SMALL OUTLINE (D)	CERAMIC (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	4 mV	NE5534D	—	NE5534JG	NE5534P
		NE5534AD	—	NE5534AJG	NE5534AP
-55°C to 125°C	2 mV	—	SE5534FK	SE5534JG	—
		—	SE5534AFK	SE5534AJG	—

The D package is available taped and reeled. Add the suffix R to the device type when ordering (e.g., NE5534DR).

SE5534A FROM TI NOT RECOMMENDED FOR NEW DESIGNS

description

The SE5534, SE5534A, NE5534, and NE5534A are monolithic high-performance operational amplifiers combining excellent dc and ac characteristics. Some of the features include very low noise, high output drive capability, high unity-gain and maximum-output-swing bandwidths, low distortion, and high slew rate.

These operational amplifiers are internally compensated for a gain equal to or greater than three. Optimization of the frequency response for various applications can be obtained by use of an external compensation capacitor between pins 5 and 8. The devices feature input-protection diodes, output short-circuit protection, and offset-voltage nulling capability.

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS
INSTRUMENTS

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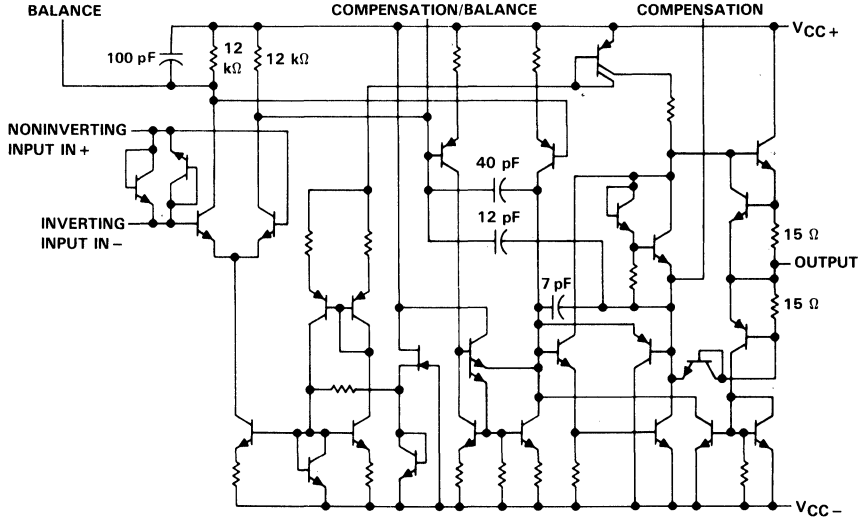
SE5534, SE5534A, NE5534, NE5534A LOW-NOISE OPERATIONAL AMPLIFIERS

description (continued)

For the NE5534A, a maximum limit is specified for equivalent input noise voltage.

The SE5534 and SE5534A are characterized for operation over the full military temperature range of -55°C to 125°C ; the NE5534 and NE5534A are characterized for operation from 0°C to 70°C .

schematic



All component values shown are nominal.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	22 V
Supply voltage, V_{CC-} (see Note 1)	-22 V
Input voltage either input (see Notes 1 and 2)	V_{CC+}
Input current (see Note 3)	± 10 mA
Duration of output short-circuit (see Note 4)	unlimited
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range: SE5534, SE5534A	-55°C to 125°C
NE5534, NE5534A	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C

- NOTES:
1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. The magnitude of the input voltage must never exceed the magnitude of the supply voltage.
 3. Excessive current will flow if a differential input voltage in excess of approximately 0.6 V is applied between the inputs unless some limiting resistance is used.
 4. The output may be shorted to ground or either power supply. Temperature and/or supply voltages must be limited to ensure the maximum dissipation rating is not exceeded.

SE5534, SE5534A, NE5534, NE5534A LOW-NOISE OPERATIONAL AMPLIFIERS

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	N/A
FK (see Note 5)	1375 mW	11.0 mW/°C	880 mW	275 mW
JG (SE5534_)	1050 mW	8.4 mW/°C	672 mW	210 mW
JG (NE5534_)	825 mW	6.6 mW/°C	528 mW	N/A
P	1000 mW	8.0 mW/°C	640 mW	N/A

NOTE 5: For the FK package, power rating and derating factor will vary with actual mounting technique used. The values stated here are believed to be conservative.

electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†		SE5534, SE5534A			NE5534, NE5534A			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0$, $R_S = 50\ \Omega$	$T_A = 25^\circ\text{C}$	0.5	2	0.5	4		mV	
		$T_A = \text{full range}$		3		5			
I_{IO} Input offset current	$V_O = 0$	$T_A = 25^\circ\text{C}$	10	200	20	300		nA	
		$T_A = \text{full range}$		500		400			
I_{IE} Input bias current	$V_O = 0$	$T_A = 25^\circ\text{C}$	400	800	500	1500		nA	
		$T_A = \text{full range}$		1500		2000			
V_{ICR} Common-mode input voltage range			± 12	± 13	± 12	± 13		V	
V_{OPP} Maximum peak-to-peak output voltage swing	$R_L \geq 600\ \Omega$	$V_{CC\pm} = \pm 15\text{ V}$	24	26	24	26		V	
		$V_{CC\pm} = \pm 18\text{ V}$	30	32	30	32			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 10\text{ V}$, $R_L \geq 600\ \Omega$	$T_A = 25^\circ\text{C}$	50	100	25	100		V/mV	
		$T_A = \text{full range}$		25		15			
A_{vd} Small-signal differential voltage amplification	$f = 10\text{ kHz}$	$C_C = 0$		6		6		V/mV	
		$C_C = 22\text{ pF}$		2.2		2.2			
B_{OM} Maximum-output-swing bandwidth	$V_O = \pm 10\text{ V}$, $C_C = 0$			200		200		kHz	
	$V_O = \pm 10\text{ V}$, $C_C = 22\text{ pF}$			95		95			
	$V_{CC\pm} = \pm 18\text{ V}$, $V_O = \pm 14\text{ V}$, $R_L = 600\ \Omega$, $C_C = 22\text{ pF}$			70		70			
B_1 Unity-gain bandwidth	$C_C = 22\text{ pF}$, $C_L = 100\text{ pF}$			10		10		MHz	
r_i Input resistance			50	100	30	100		k Ω	
z_c Output impedance	$A_{VD} = 30\text{ dB}$, $C_C = 22\text{ pF}$,	$R_L = 600\ \Omega$, $f = 10\text{ kHz}$		0.3		0.3		Ω	
CMRR Common-mode rejection ratio	$V_O = 0$, $R_S = 50\ \Omega$	$V_{IC} = V_{ICR\text{ min}}$	80	100	70	100		dB	
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IO}$)	$V_{CC+} = \pm 9\text{ V to } \pm 15\text{ V}$, $V_O = 0$, $R_S = 50\ \Omega$		86	100	80	100		dB	
I_{OS} Output short-circuit current				38		38		mA	
I_{CC} Supply current	No load, $V_O = 0$	$T_A = 25^\circ\text{C}$	4	6.5	4	8		mA	
		$T_A = \text{full range}$		9					

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range is $T_A = -55^\circ\text{C to } 125^\circ\text{C}$ for SE5534 and SE5534A and $0^\circ\text{C to } 70^\circ\text{C}$ for NE5534 and NE5534A.

2
Operational Amplifiers

SE5534, SE5534A, NE5534, NE5534A LOW-NOISE OPERATIONAL AMPLIFIERS

operating characteristics, $V_{CC\pm} = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	SE5534, NE5534			SE5534A, NE5534A			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$C_C = 0$	13			13			$\text{V}/\mu\text{s}$
	$C_C = 22\text{ pF}$	6			6			
t_r Rise time	$V_I = 50\text{ mV}$, $A_{VD} = 1$, $R_L = 600\ \Omega$, $C_C = 22\text{ pF}$,	20			20			ns
Overshoot factor	$C_L = 100\text{ pF}$	20%			20%			
t_r Rise time	$V_I = 50\text{ mV}$, $A_{VD} = 1$, $R_L = 600\ \Omega$, $C_C = 47\text{ pF}$,	50			50			ns
	$C_L = 500\text{ pF}$	35%			35%			
V_n Equivalent input noise voltage	$f = 30\text{ Hz}$	7			5.5	7		$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\text{ kHz}$	4			3.5	4.5		
I_n Equivalent input noise current	$f = 30\text{ Hz}$	2.5			1.5			$\text{pA}/\sqrt{\text{Hz}}$
	$f = 1\text{ kHz}$	0.6			0.4			
F Average noise figure	$R_S = 5\text{ k}\Omega$, $f = 10\text{ Hz to }20\text{ kHz}$				0.9			dB

2

Operational Amplifiers

TYPICAL CHARACTERISTICS†

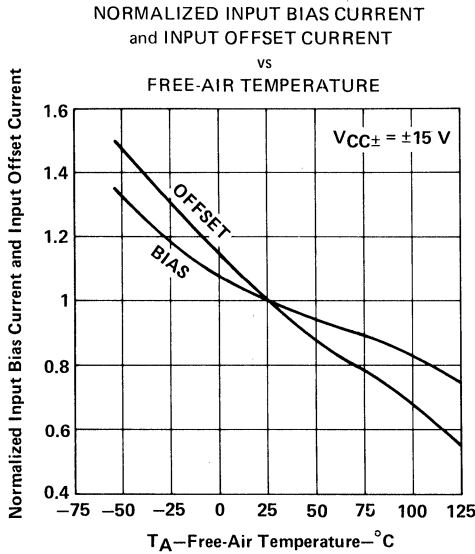


FIGURE 1

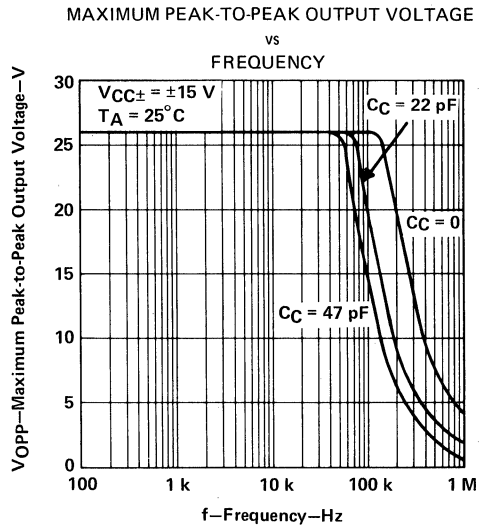


FIGURE 2

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

SE5534, SE5534A, NE5534, NE5534A LOW-NOISE OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREQUENCY

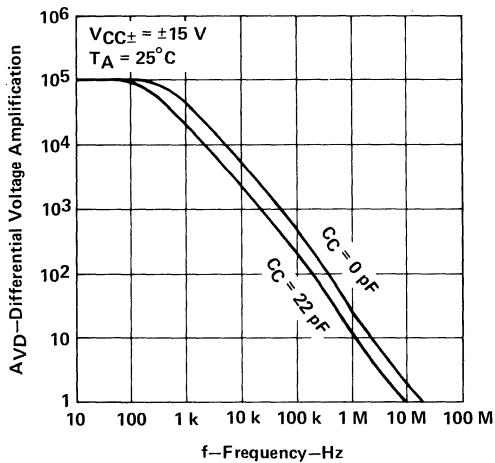


FIGURE 3

NORMALIZED SLEW RATE and
UNITY-GAIN BANDWIDTH
vs
SUPPLY VOLTAGE

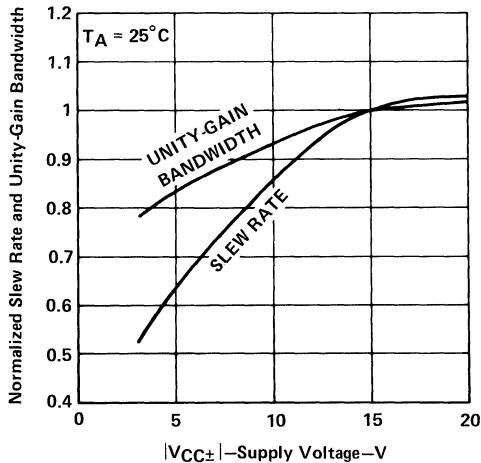


FIGURE 4

NORMALIZED SLEW RATE and
UNITY-GAIN BANDWIDTH
vs
FREE-AIR TEMPERATURE

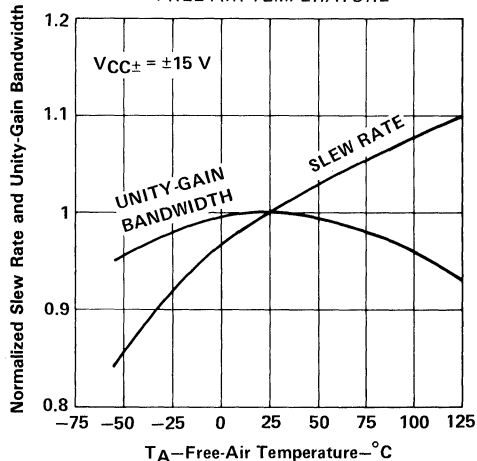


FIGURE 5

TOTAL HARMONIC DISTORTION
vs
FREQUENCY

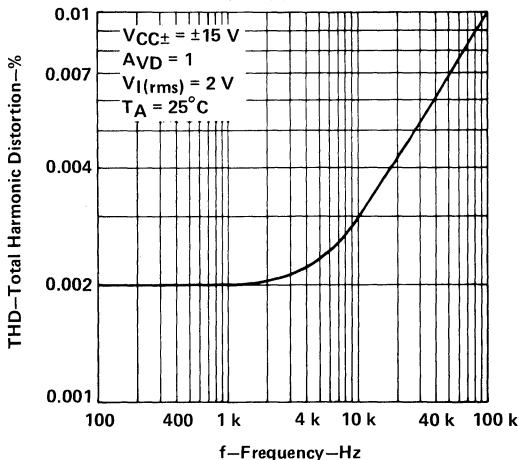


FIGURE 6

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

SE5534, SE5534A, NE5534, NE5534A
LOW-NOISE OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

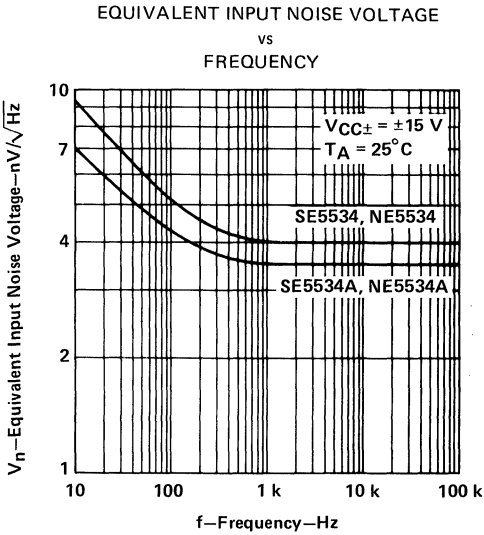


FIGURE 7

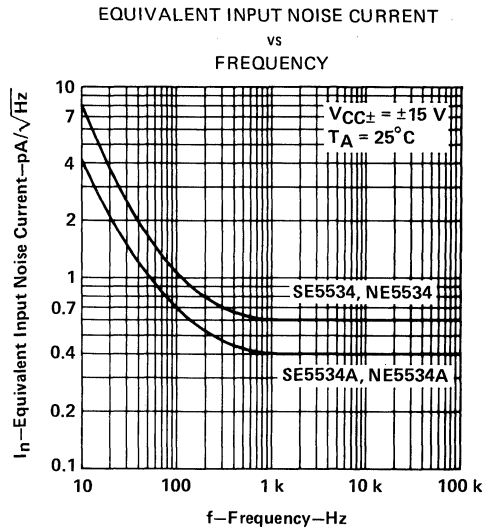


FIGURE 8

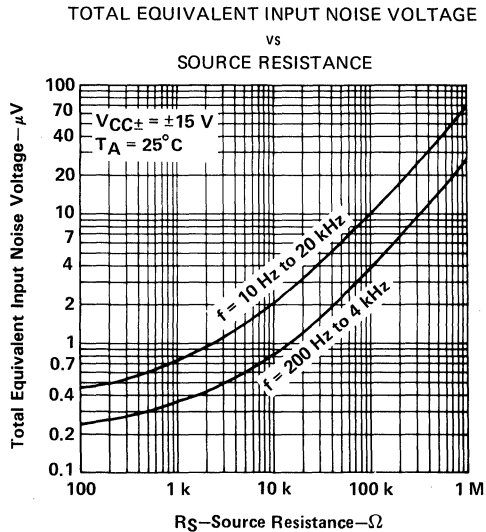


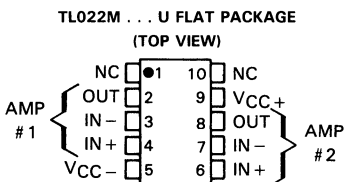
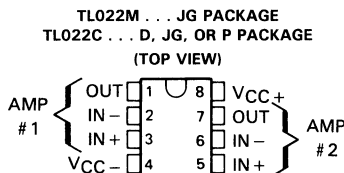
FIGURE 9

TL022M, TL022C DUAL LOW-POWER OPERATIONAL AMPLIFIERS

D1661, SEPTEMBER 1973—REVISED JULY 1988

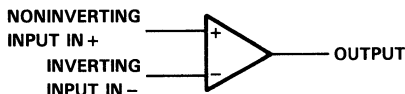
- Very Low Power Consumption
- Power Dissipation with $\pm 2\text{-V}$ Supplies . . . $170 \mu\text{W}$ Typ
- Low Input Bias and Offset Currents
- Output Short-Circuit Protection
- Low Input Offset Voltage
- Internal Frequency Compensation
- Latch-Up-Free Operation
- Popular Dual Op Amp Pin-Out

**TL022M IS NOT RECOMMENDED FOR
NEW DESIGNS**



NC—No internal connection

symbol (each amplifier)



description

The TL022 is a dual low-power operational amplifier designed to replace higher power devices in many applications without sacrificing system performance. High input impedance, low supply currents, and low equivalent input noise voltage over a wide range of operating supply voltages result in an extremely versatile operational amplifier for use in a variety of analog applications including battery-operated circuits. Internal frequency compensation, absence of latch-up, high slew rate, and output short-circuit protection assure ease of use.

The TL022M is characterized for operation over the full military temperature range of -55°C to 125°C ; the TL022C is characterized for operation from 0°C to 70°C .

AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE			
		SMALL OUTLINE (D)	CERAMIC DIP (JG)	PLASTIC DIP (P)	CERAMIC FLAT PACK (U)
0°C to 70°C	5 mV	TL022CD	TL022CJG	TL022CP	—
-55°C to 125°C	5 mV	—	TL022MJG	—	TL022MU

The D package is available taped and reeled. Add the suffix R to the device type (i.e. TL022CDR)

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Operational Amplifiers

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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TL022M, TL022C DUAL LOW-POWER OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	TL022M	TL022C	UNIT
Supply voltage V_{CC+} (see Note 1)	22	18	V
Supply voltage V_{CC-} (see Note 1)	-22	-18	V
Differential input voltage (see Note 2)	± 30	± 30	V
Input voltage (any input, see Notes 1 and 3)	± 15	± 15	V
Duration of output short-circuit (see Note 4)	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	°C
Storage temperature range	-65 to 150		°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	JG or U package	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package	260	°C

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
 4. The output may be shorted to ground or either power supply. For the TL022M only, the unlimited duration of the short-circuit applies at (or below) 125°C case temperature or 75°C free-air temperature.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	680 mW	5.8 mW/°C	33°C	464 mW	—
JG (TL022M)	680 mW	8.4 mW/°C	69°C	672 mW	210 mW
JG (TL022C)	680 mW	6.6 mW/°C	47°C	528 mW	—
P	680 mW	8.0 mW/°C	65°C	640 mW	—
U	675 mW	5.4 mW/°C	25°C	432 mW	135 mW

TL022M, TL022C DUAL LOW-POWER OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$

PARAMETER	TEST CONDITIONS†		TL022M			TL022C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0$, $R_S = 50\ \Omega$	25°C	1	5		1	5	mV	
		Full range		6		7.5			
I_{IO} Input offset current	$V_O = 0$	25°C	5	40		15	80	nA	
		Full range		100		200			
I_{IB} Input bias current	$V_O = 0$	25°C	50	100		100	250	nA	
		Full range		250		400			
V_{ICR} Common-mode input voltage range		25°C	±12	±13		±12	±13	V	
		Full range	±12			±12			
V_{OPP} Maximum peak-to-peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	20	26		20	26	V	
	$R_L \geq 10\ \text{k}\Omega$	Full range	20			20			
A_{VD} Large-signal differential voltage amplification	$R_L \geq 10\ \text{k}\Omega$, $V_O = \pm 10\ \text{V}$	25°C	72	86		60	80	dB	
		Full range	66			60			
B_1 Unity-gain bandwidth		25°C		0.5		0.5	MHz		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR}\ \text{min}$, $R_S = 50\ \Omega$	25°C	60	72		60	72	dB	
		Full range	60			60			
kSVS Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	$V_{CC} = \pm 9\ \text{V}$ to $\pm 15\ \text{V}$, $R_S = 50\ \Omega$	25°C		30	150		30	200	$\mu\text{V}/\text{V}$
		Full range			150			200	
V_n Equivalent input noise voltage	$A_{VD} = 20\ \text{dB}$, $B = 1\ \text{Hz}$, $f = 1\ \text{kHz}$	25°C		50		50		$\text{nV}/\sqrt{\text{Hz}}$	
I_{OS} Short-circuit output current		25°C		±6		±6		mA	
I_{CC} Supply current (both amplifiers)	No load, $V_O = 0$	25°C		130	250		130	250	μA
		Full range			250			250	
P_D Total dissipation (both amplifiers)	No load, $V_O = 0$	25°C		3.9	6		3.9	7.5	mW
		Full range			6			7.5	

†All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for TL022M is -55°C to 125°C and for TL022C is 0°C to 70°C .

operating characteristics, $V_{CC+} = 15\ \text{V}$, $V_{CC-} = -15\ \text{V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS		TL022M			TL022C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
t_r Rise time	$V_I = 20\ \text{mV}$, $C_L = 100\ \text{pF}$, See Figure 1	$R_L = 10\ \text{k}\Omega$, See Figure 1	0.3			0.3			μs
Overshoot factor			5%			5%			
SR Slew rate at unity gain	$V_I = 10\ \text{V}$, $C_L = 100\ \text{pF}$, See Figure 1	$R_L = 10\ \text{k}\Omega$, See Figure 1	0.5			0.5			$\text{V}/\mu\text{s}$

2
Operational Amplifiers

**TL022M, TL022C
DUAL LOW-POWER OPERATIONAL AMPLIFIERS**

2
Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

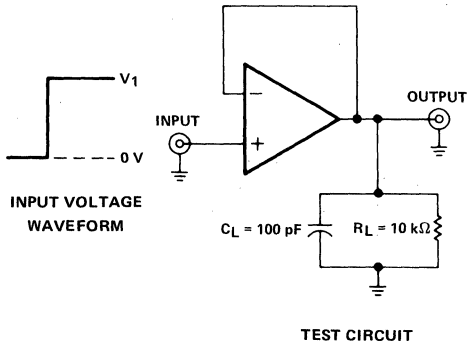


FIGURE 1. RISE TIME, OVERSHOOT FACTOR, AND SLEW RATE

TYPICAL CHARACTERISTICS

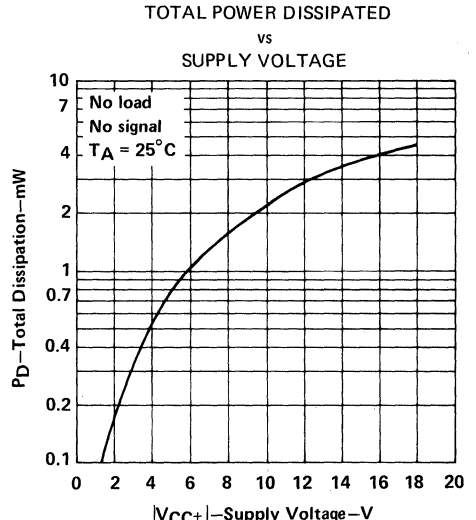
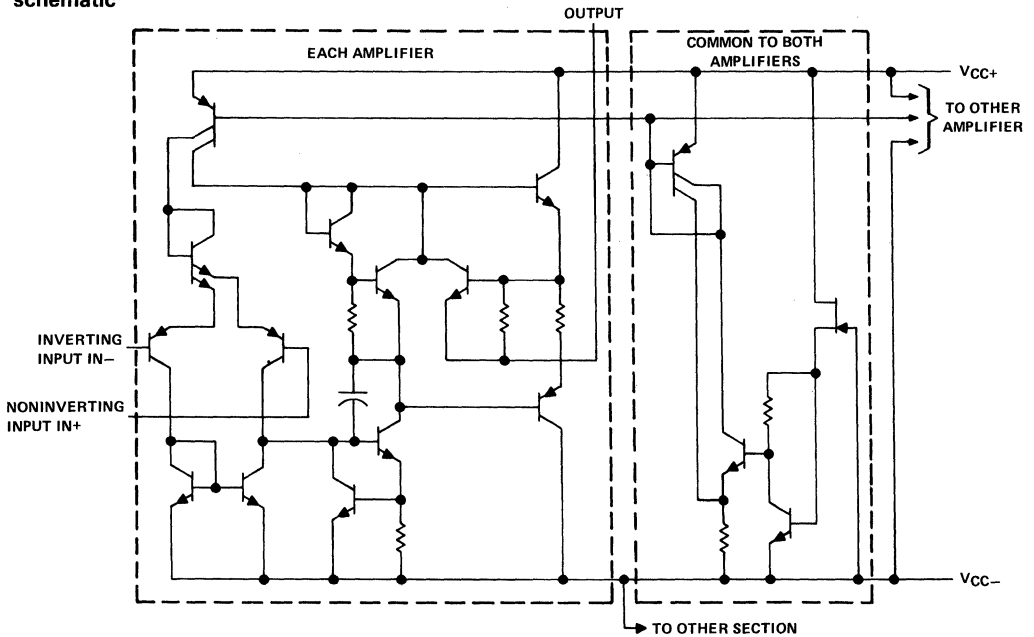


FIGURE 2

schematic

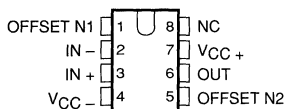


TL031, TL031A ENHANCED JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

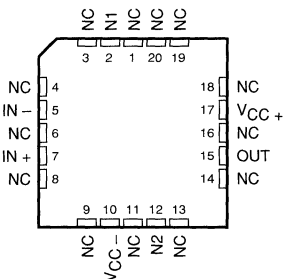
D3151, JULY 1988 – REVISED JANUARY 1989

- Maximum Offset Voltage ... 800 μV
- High Slew Rate ... 2.9 $\text{V}/\mu\text{s}$ Typ
- Low Input Bias Current ... 2 pA Typ
- Very Low Power Consumption ... 6.5 mW Typ
- Output Short-Circuit Protection

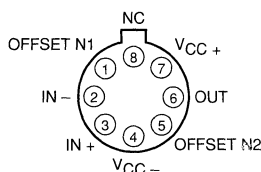
D, JG, or P PACKAGE
(TOP VIEW)



FK PACKAGE
(TOP VIEW)



L PACKAGE
(TOP VIEW)



Pin 4 (L Package) is in electrical contact with the case

NC – No internal connection

description

The TL031 and TL031A operational amplifiers incorporate well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. These devices offer the significant advantages of Texas Instruments new enhanced JFET process. This process affords not only low initial offset voltage due to the on-chip zener trim capability but also stable offset voltage over time and temperature. In comparison, traditional JFET processes are plagued by significant offset voltage drift.

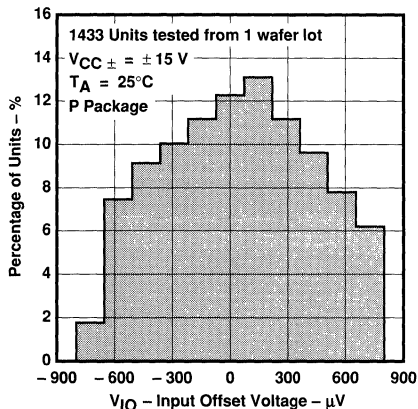
This new enhanced process still maintains the traditional JFET advantages of fast slew rates and low input bias and offset currents. These advantages, coupled with low power consumption, make the TL031 well-suited for new state-of-the-art designs as well as existing design upgrades. The TL031 has been designed to be functionally compatible and pin compatible with the TL061.

AVAILABLE OPTIONS

TA	V _{IO} max AT 25°C	PACKAGE				
		SMALL- OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)
0°C to 70°C	0.8 mV	TL031ACD	—	TL031ACJG	TL031ACL	TL031ACP
	1.5 mV	TL031CD	—	TL031CJG	TL031CL	TL031CP
-40°C to 85°C	0.8 mV	TL031AID	—	TL031AJG	TL031AIL	TL031AIP
	1.5 mV	TL031ID	—	TL031JG	TL031IL	TL031IP
-55°C to 125°C	0.8 mV	TL031AMD	TL031AMFK	TL031AMJG	TL031AML	TL031AMP
	1.5 mV	TL031MD	TL031MFK	TL031MJG	TL031ML	TL031MP

D packages are available taped-and-reeled. Add "R" suffix to device type (e.g., TL031CDR).

DISTRIBUTION OF TL031A
INPUT OFFSET VOLTAGE



PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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TL031, TL031A ENHANCED JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

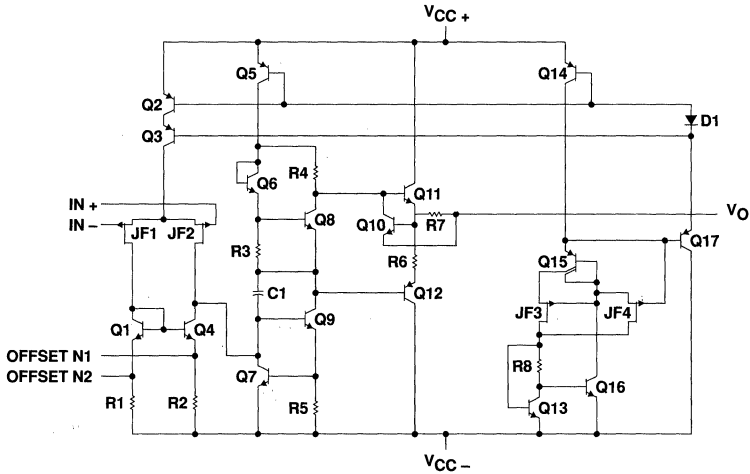
description (continued)

Two offset voltage grades are available: TL031 (1.5 mV max) and TL031A (800 μ V max).

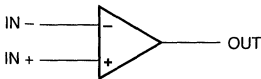
A variety of available packaging options includes small-outline and chip carrier versions for high density system applications.

The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I-suffix devices are characterized for operation from -40°C to 85°C . The C-suffix devices are characterized for operation from 0°C to 70°C .

equivalent schematic



symbol



2

Operational Amplifiers

TL031, TL031A ENHANCED JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

2

Operational Amplifiers

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	18 V
Supply voltage, V_{CC-} (see Note 1)	- 18 V
Differential input voltage (see Note 2)	± 30 V
Input voltage range, V_I (any input, see Notes 1 and 3)	± 15 V
Input current, I_I (each input)	± 1 mA
Output current, I_O	± 40 mA
Total current into V_{CC+} terminal	160 mA
Total current out of V_{CC-} terminal	160 mA
Duration of short-circuit current at (or below) 25°C (see Note 4)	Unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, T_A : M-suffix	- 55°C to 125°C
I-suffix	- 40°C to 85°C
C-suffix	0°C to 70°C
Storage temperature range	- 65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG or L package	300°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
2. Differential voltages are at the noninverting input with respect to the inverting input.
3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
4. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR		$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$		POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C		464 mW	377 mW	145 mW
FK	1375 mW	11.0 mW/°C		880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C		672 mW	546 mW	210 mW
L	825 mW	6.6 mW/°C		528 mW	429 mW	165 mW
P	1000 mW	8.0 mW/°C		640 mW	520 mW	200 mW

recommended operating conditions

		M-SUFFIX		I-SUFFIX		C-SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, V_{CC}		± 5	± 15	± 5	± 15	± 5	± 15	V
Common-mode input voltage, V_{IC}	$V_{CC} \pm \pm 5$ V	- 1.5	4	- 1.5	4	- 1.5	4	V
	$V_{CC} \pm \pm 15$ V	- 11.5	14	- 11.5	14	- 11.5	14	V
Operating free-air temperature, T_A		- 55	125	- 40	85	0	70	°C

TL031M, TL031AM

ENHANCED JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± ± 5 V			V _{CC} ± ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL031M	25°C	0.54	3.5	0.5	1.5	mV	
			Full range	6.5		4.5			
		TL031AM	25°C	0.41	2.8	0.34	0.8		
			Full range	5.8		3.8			
α _{VIO} Temperature coefficient of input offset voltage	TL031M	25°C to 125°C	5.1			4.3			μV/°C
		TL031AM	25°C to 125°C	5.1			4.3		
Input offset voltage long-term drift (see Note 5)		25°C	0.04			0.04			μV/mo
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	1	100	1	100	pA		
		125°C	0.2	10	0.2	10	nA		
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	2	200	2	200	pA		
		125°C	7	20	8	20	nA		
V _{ICR} Common-mode input voltage range		25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4	V		
		Full range	-1.5 to 4		-11.5 to 14				
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.3	13	14	V		
		-55°C	3	4.1	13	14			
		125°C	3	4.4	13	14			
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-3	-4.2	-12.5	-13.9	V		
		-55°C	-3	-4	-12.5	-13.8			
		125°C	-3	-4.3	-12.5	-14			
A _{VD} Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	4	12	5	14.3	V/mV		
		-55°C	3	7.1	4	10.4			
		125°C	3	12.9	4	15			
r _i Input resistance		25°C	10 ¹²			10 ¹²	Ω		
C _i Input capacitance		25°C	5			4	pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	70	87	75	94	dB		
		-55°C	70	87	70	94			
		125°C	70	87	70	94			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75	96	75	96	dB		
		-55°C	75	95	75	95			
		125°C	75	96	75	96			
P _D Total power dissipation	No load, V _O = 0	25°C	1.9	2.5	6.5	8.4	mW		
		-55°C	1.1	2.5	4.7	8.4			
		125°C	1.8	2.5	5.8	8.4			
I _{CC} Supply current	No load, V _O = 0	25°C	192	250	217	280	μA		
		-55°C	114	250	156	280			
		125°C	178	250	197	280			

[†] Full range is -55°C to 125°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. At V_{CC} ± = ± 5 V, V_O = ± 2.3 V; at V_{CC} ± = ± 15 V, V_O = ± 10 V.

TL031M, TL031AM ENHANCED JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER		TEST CONDITIONS		T _A	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT	
					MIN	TYP	MAX	MIN	TYP	MAX		
SR +	Positive slew rate at unity gain	R _L = 10 kΩ, C _L = 100 pF, See Figure 1, See Note 7		25°C	2.0			2			V/μs	
				-55°C	1.4			1.2				
				125°C	2.4			1.2				
SR -	Negative slew rate at unity gain	See Note 7		25°C	3.9			3			V/μs	
				-55°C	3.2			2.5				
				125°C	4.1			2.5				
t _r	Rise time	V _I PP = ± 10 mV, R _L = 10 kΩ, C _L = 100 pF, See Figures 1 and 2		25°C	138			132			ns	
				-55°C	142			123				
				125°C	166			158				
t _f	Fall time	See Figures 1 and 2		25°C	138			132			ns	
				-55°C	142			123				
				125°C	166			158				
Overshoot factor				25°C	11%			5%				
				-55°C	16%			6%				
				125°C	14%			8%				
V _n	Equivalent input noise voltage	TL031M	R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C	61			61			nV/√Hz
				f = 1 kHz	41			41				
		TL031AM		f = 10 Hz	25°C	61			61			
				f = 1 kHz	41			41				
I _n	Equivalent input noise current	f = 1 kHz		25°C	0.003			0.003			pA/√Hz	
B1	Unity-gain bandwidth	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	1			1.1			MHz	
				-55°C	1			1.1				
				125°C	0.9			0.9				
φ _m	Phase margin at unity gain	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	61°			65°				
				-55°C	57°			64°				
				125°C	59°			62°				

NOTE 7: For V_{CC} ± = ± 5 V, V_IPP = ± 1 V; for V_{CC} ± = ± 15 V, V_IPP = ± 5 V.

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Operational Amplifiers

TL031I, TL031AI ENHANCED JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC±} = ± 5 V			V _{CC±} = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL031I	25°C	0.54	3.5	0.5	1.5	mV	
			Full range	5.3			3.3		
		TL031AI	25°C	0.41	2.8	0.34	0.8		
			Full range	4.6			2.6		
α _{VIO} Temperature coefficient of input offset voltage (see Note 8)	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL031I	25°C to 85°C	6.5			6.2	μV/°C	
		TL031AI	25°C to 85°C	6.5			6.2	25	
Input offset voltage long-term drift (see Note 5)		25°C	0.04			0.04	μV/mo		
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	1			100	1	100	pA
		85°C	0.02			0.45	0.02	0.45	nA
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	2			200	2	200	pA
		85°C	0.2			0.9	0.2	0.9	nA
V _{ICR} Common-mode input voltage range		25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4	V		
		Full range	-1.5 to 4		-11.5 to 14				
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.3	13	14	V		
		-40°C	3	4.1	13	14			
		85°C	3	4.4	13	14			
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-3	-4.2	-12.5	-13.9	V		
		-40°C	-3	-4.1	-12.5	-13.8			
		85°C	-3	-4.2	-12.5	-14			
A _{VD} Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	4	12	5	14.3	V/mV		
		-40°C	3	8.4	4	11.6			
		85°C	4	13.5	5	15.3			
r _i Input resistance		25°C	10 ¹²			10 ¹²	Ω		
C _i Input capacitance		25°C	5			4	pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	70	87	75	94	dB		
		-40°C	70	87	75	94			
		85°C	70	87	75	94			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC±} / ΔV _{IO})	V _{CC±} = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75	96	75	96	dB		
		-40°C	75	96	75	96			
		85°C	75	96	75	96			
P _D Total power dissipation	No load, V _O = 0	25°C	1.9	2.5	6.5	8.4	mW		
		-40°C	1.4	2.5	5.4	8.4			
		85°C	1.9	2.5	6.2	8.4			
I _{CC} Supply current	No load, V _O = 0	25°C	192	250	217	280	μA		
		-40°C	144	250	181	280			
		85°C	189	250	207	280			

[†] Full range is -40°C to 85°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. At V_{CC±} = ± 5 V, V_O = ± 2.3 V; at V_{CC±} = ± 15 V, V_O = ± 10 V.

8. This parameter is tested on a sample basis for the TL031A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

TL031I, TL031AI ENHANCED JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER		TEST CONDITIONS		T _A	V _{CC±} = ±5 V			V _{CC±} = ±15 V			UNIT
					MIN	TYP	MAX	MIN	TYP	MAX	
SR +	Positive slew rate at unity gain	R _L = 10 kΩ, C _L = 100 pF, See Figure 1, See Note 7		25°C	2.0			2 2.9			V/μs
				-40°C	1.6			1.5 2.1			
				85°C	2.3			2 3.3			
SR -	Negative slew rate at unity gain	See Figure 1, See Note 7		25°C	3.9			3.5 5.1			V/μs
				-40°C	3.3			3.2 4.8			
				85°C	4.1			3.2 4.9			
t _r	Rise time	V _{Ipp} = ±10 mV, R _L = 10 kΩ, C _L = 100 pF, See Figures 1 and 2		25°C	138			132			ns
t _f	Fall time			-40°C	132			123			
				85°C	154			146			
Overshoot factor				25°C	138			132			ns
				-40°C	132			123			
				85°C	154			146			
V _n	Equivalent input noise voltage (see Note 9)	TL031I	R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C			61			nV/√Hz
				f = 1 kHz	41			41			
		TL031AI		f = 10 Hz	25°C			61			
				f = 1 kHz	41			41 60			
I _n	Equivalent input noise current	f = 1 kHz		25°C	0.003			0.003			pA/√Hz
B1	Unity-gain bandwidth	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	1			1.1			MHz
				-40°C	1			1.1			
				85°C	0.9			1			
φ _m	Phase margin at unity gain	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	61°			65°			
				-40°C	60°			65°			
				85°C	60°			64°			

NOTES: 7. For V_{CC±} = ±5 V, V_{Ipp} = ±1 V; for V_{CC±} = ±15 V, V_{Ipp} = ±5 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

TL031C, TL031AC ENHANCED JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL031C	25°C	0.54	3.5	0.5	1.5	mV	
			Full range	4.5		2.5			
		TL031AC	25°C	0.41	2.8	0.34	0.8		
			Full range	3.8		1.8			
α _{VIO} Temperature coefficient of input offset voltage (see Note 8)	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL031C	25°C to 70°C	7.1		5.9		μV/°C	
		TL031AC	25°C to 70°C	7.1		5.9		25	
Input offset voltage long-term drift (see Note 5)		25°C	0.04		0.04		μV/mo		
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	1	100	1	100	pA		
		70°C	9	200	12	200			
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	2	200	2	200	pA		
		70°C	50	400	80	400			
V _{ICR} Common-mode input voltage range		25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4	V		
		Full range	-1.5 to 4		-11.5 to 14				
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.3	13	14	V		
		0°C	3	4.2	13	14			
		70°C	3	4.3	13	14			
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-3	-4.2	-12.5	-13.9	V		
		0°C	-3	-4.1	-12.5	-13.9			
		70°C	-3	-4.2	-12.5	-14			
A _{VD} Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	4	12	5	14.3	V/mV		
		0°C	3	11.1	4	13.5			
		70°C	4	13.3	5	15.2			
r _i Input resistance		25°C	10 ¹²			10 ¹²	Ω		
C _i Input capacitance		25°C	5			4	pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	70	87	75	94	dB		
		0°C	70	87	75	94			
		70°C	70	87	75	94			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75	96	75	96	dB		
		0°C	75	96	75	96			
		70°C	75	96	75	96			
P _D Total power dissipation	No load, V _O = 0	25°C	1.9	2.5	6.5	8.4	mW		
		0°C	1.8	2.5	6.3	8.4			
		70°C	1.9	2.5	6.3	8.4			
I _{CC} Supply current	No load, V _O = 0	25°C	192	250	217	280	μA		
		0°C	184	250	211	280			
		70°C	189	250	210	280			

[†] Full range is 0°C to 70°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. At V_{CC} ± = ± 5 V, V_O = ± 2.3 V; at V_{CC} ± = ± 15 V, V_O = ± 10 V.

8. This parameter is tested on a sample basis for the TL031A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

TL031C, TL031AC ENHANCED JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

operating characteristics

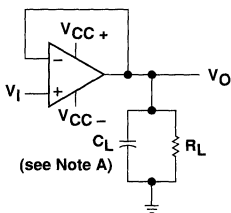
PARAMETER	TEST CONDITIONS	T _A	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR + Positive slew rate at unity gain	R _L = 10 kΩ, C _L = 100 pF, See Figure 1, See Note 7	25°C	2.0			2 2.9			V/μs
		0°C	1.8			1.5 2.6			
		70°C	2.2			2 3.2			
SR - Negative slew rate at unity gain		25°C	3.9			3.5 5.1			V/μs
		0°C	3.7			3.2 5.0			
		70°C	4.0			3.2 5.0			
t _r Rise time	V _I PP = ± 10 mV, R _L = 10 kΩ, C _L = 100 pF, See Figures 1 and 2	25°C	138			132			ns
		0°C	134			127			
		70°C	150			142			
t _f Fall time		25°C	138			132			ns
		0°C	134			127			
		70°C	150			142			
Overshoot factor	25°C	11%			5%				
	0°C	10%			4%				
	70°C	12%			6%				
V _n Equivalent input noise voltage (see Note 9)	TL031C See Figure 3 TL031AC	f = 10 Hz	61			61			nV/√Hz
			f = 1 kHz	41			41		
		f = 10 Hz	61			61			
			f = 1 kHz	41			41 60		
I _n Equivalent input noise current	f = 1 kHz	25°C	0.003			0.003			pA/√Hz
B1 Unity-gain bandwidth	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4	25°C	1			1.1			MHz
		0°C	1			1.1			
		70°C	1			1			
φ _m Phase margin at unity gain	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4	25°C	61°			65°			
		0°C	61°			65°			
		70°C	60°			64°			

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Operational Amplifiers

NOTES: 7. For V_{CC} ± = ± 5 V, V_IPP = ± 1 V; for V_{CC} ± = ± 15 V, V_IPP = ± 5 V.
9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

PARAMETER MEASUREMENT INFORMATION



NOTE A: C_L includes fixture capacitance.

FIGURE 1. SLEW RATE, RISE/FALL TIME, AND OVERSHOOT TEST CIRCUIT

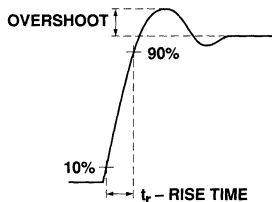


FIGURE 2. RISE TIME AND OVERSHOOT WAVEFORM

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ENHANCED JFET LOW-POWER LOW-OFFSET
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PARAMETER MEASUREMENT INFORMATION

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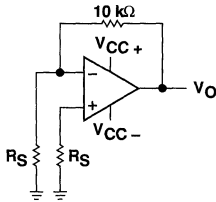
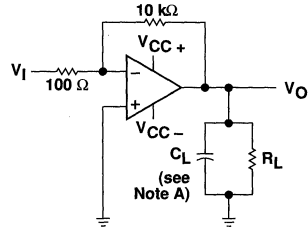


FIGURE 3. NOISE VOLTAGE TEST CIRCUIT



NOTE A: C_L includes fixture capacitance.

FIGURE 4. UNITY-GAIN BANDWIDTH AND PHASE MARGIN TEST CIRCUIT

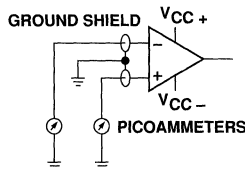


FIGURE 5. INPUT BIAS AND OFFSET CURRENT TEST CIRCUIT

typical values

Typical values as presented in this data sheet represent the median (50% point) of device parametric performance.

input bias and offset current

At the picoamp bias current level typical of the TL031 and TL031A, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To accurately measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied but with no device in the socket. The device is then inserted into the socket and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

noise

Because of the increasing emphasis on low noise levels in many of today's applications, the input noise voltage density is sample-tested at $f = 1$ kHz. Texas Instruments also has additional noise testing capability to meet specific application requirements. Please contact the factory for details.

TL031, TL031A ENHANCED JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

table of graphs

			FIGURE
V_{IO}	Input offset voltage	Distribution	6
α_{VIO}	Temperature coefficient of input offset voltage	Distribution	7
I_{IO}	Input offset current	vs Temperature	8
I_{IB}	Input bias current	vs V_{IC}	9
		vs Temperature	8
V_I	Input voltage range	vs V_{CC}	10
		vs Temperature	11
V_{ID}	Differential input voltage	vs Output voltage	12, 13
		vs V_{CC}	14
V_{OM}	Maximum peak output voltage swing	vs Output current	16, 17
		vs Frequency	15
		vs Temperature	18, 19
		vs R_L	20
A_{VD}	Differential voltage amplification	vs Frequency	21
		vs Temperature	22
		vs Frequency	23
z_o	Output impedance	vs Frequency	23
CMRR	Common-mode rejection ratio	vs Frequency	24, 25
		vs Temperature	26
k_{SVR}	Supply-voltage rejection ratio	vs Temperature	27
I_{OS}	Short-circuit output current	vs V_{CC}	28
		vs Time	29
		vs Temperature	30
I_{CC}	Supply current	vs V_{CC}	32
		vs Temperature	33
SR	Slew rate	vs R_L	34, 35
		vs Temperature	36, 37
	Overshoot factor	vs C_L	38
V_n	Equivalent input noise voltage	vs Frequency	31
THD	Total harmonic distortion	vs Frequency	39
B_1	Unity-gain bandwidth	vs V_{CC}	40
		vs Temperature	41
ϕ_m	Phase margin	vs V_{CC}	42
		vs C_L	43
		vs Temperature	44
	Phase shift	vs Frequency	21
	Pulse response	Small-signal	45
		Large-signal	46, 47

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TL031, TL031A
ENHANCED JFET LOW-POWER LOW-OFFSET
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

**DISTRIBUTION OF TL031
 INPUT OFFSET VOLTAGE**

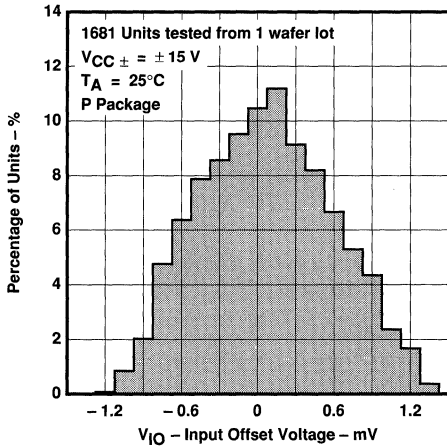


FIGURE 6

**DISTRIBUTION OF TL031
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT**

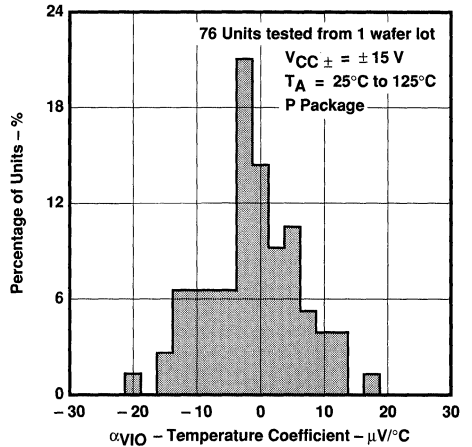


FIGURE 7

**INPUT BIAS CURRENT AND
 INPUT OFFSET CURRENT
 VS
 FREE-AIR TEMPERATURE**

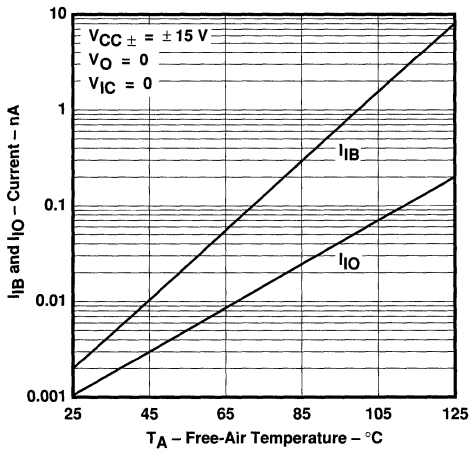


FIGURE 8

**INPUT BIAS CURRENT
 VS
 COMMON-MODE INPUT VOLTAGE**

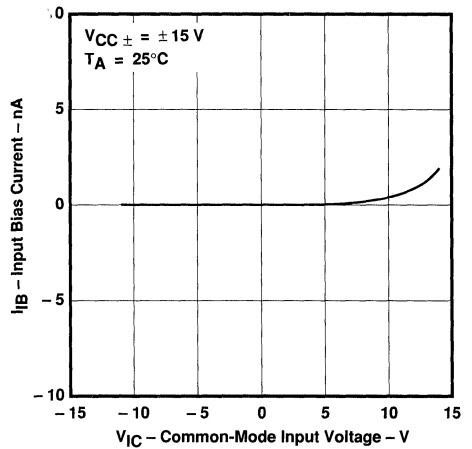


FIGURE 9

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Operational Amplifiers

TYPICAL CHARACTERISTICS

COMMON-MODE
 INPUT VOLTAGE RANGE LIMITS
 vs
 SUPPLY VOLTAGE

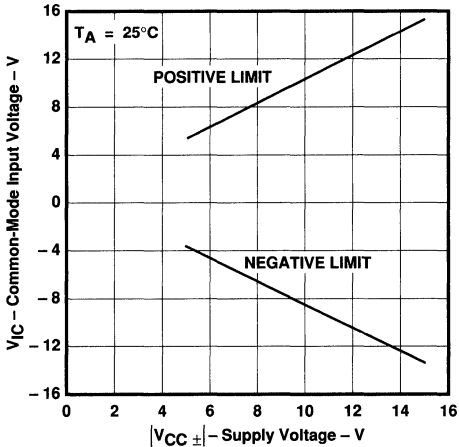


FIGURE 10

COMMON-MODE
 INPUT VOLTAGE RANGE LIMITS
 vs
 FREE-AIR TEMPERATURE

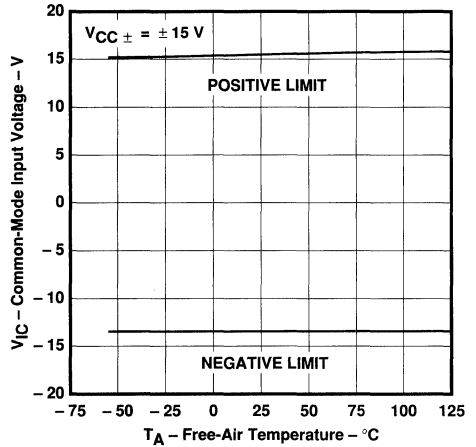


FIGURE 11

DIFFERENTIAL INPUT VOLTAGE
 vs
 OUTPUT VOLTAGE

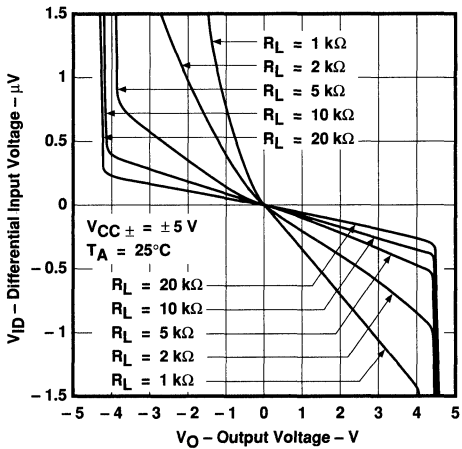


FIGURE 12

DIFFERENTIAL INPUT VOLTAGE
 vs
 OUTPUT VOLTAGE

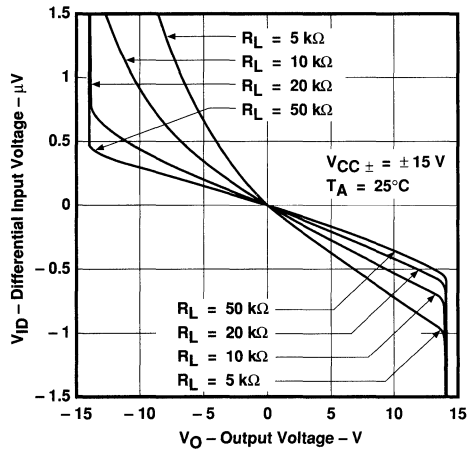


FIGURE 13

TLO31, TLO31A
ENHANCED JFET LOW-POWER LOW-OFFSET
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

MAXIMUM PEAK OUTPUT VOLTAGE
 vs
 SUPPLY VOLTAGE

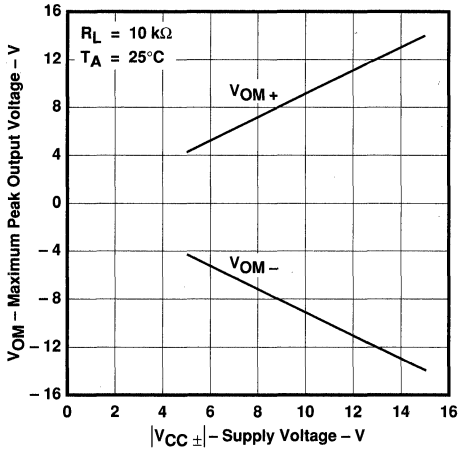


FIGURE 14

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 vs
 FREQUENCY

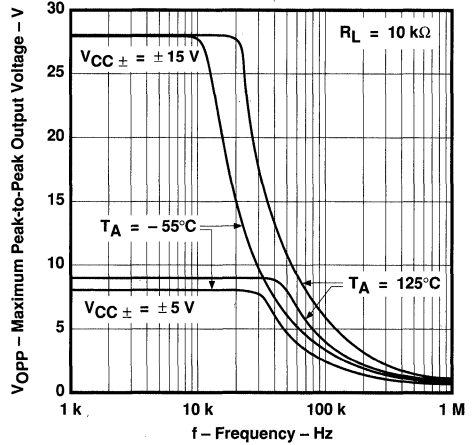


FIGURE 15

MAXIMUM PEAK OUTPUT VOLTAGE
 vs
 OUTPUT CURRENT

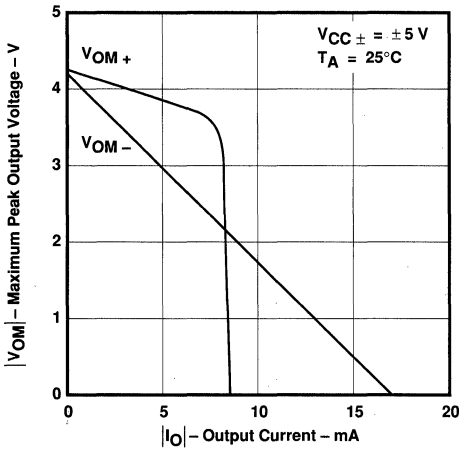


FIGURE 16

MAXIMUM PEAK OUTPUT VOLTAGE
 vs
 OUTPUT CURRENT

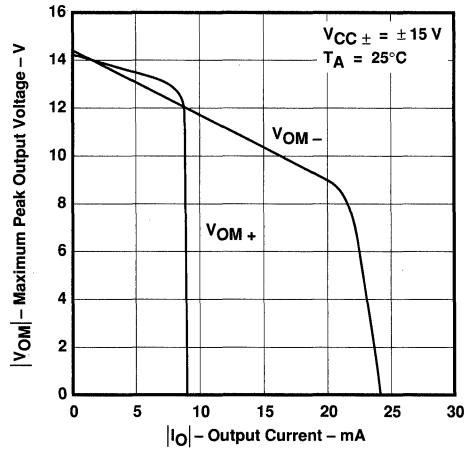


FIGURE 17

TYPICAL CHARACTERISTICS

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

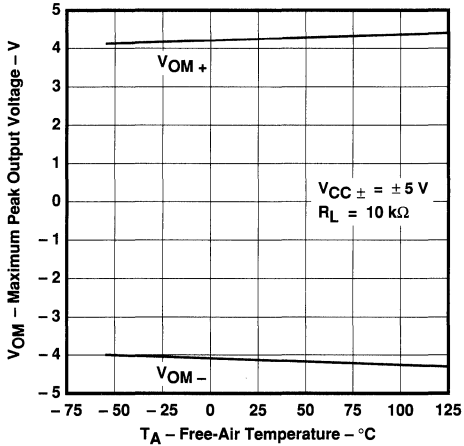


FIGURE 18

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

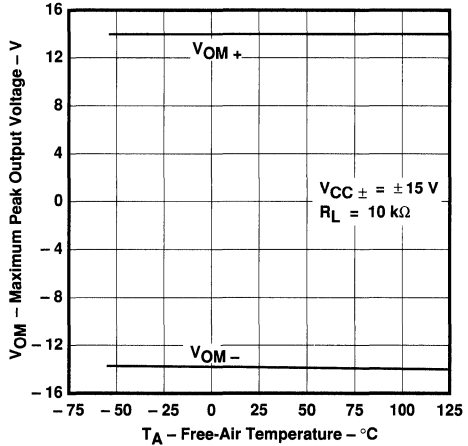


FIGURE 19

LARGE-SIGNAL VOLTAGE AMPLIFICATION
 VS
 LOAD RESISTANCE

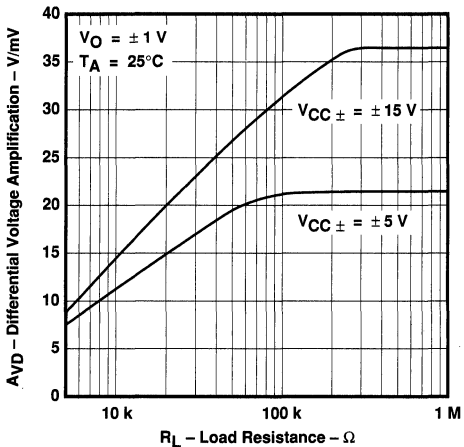


FIGURE 20

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 VS
 FREQUENCY

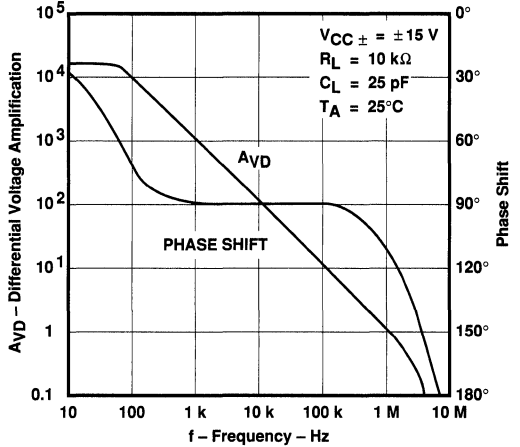


FIGURE 21

TL031, TL031A
ENHANCED JFET LOW-POWER LOW-OFFSET
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

LARGE-SIGNAL VOLTAGE AMPLIFICATION
VS
FREE-AIR TEMPERATURE

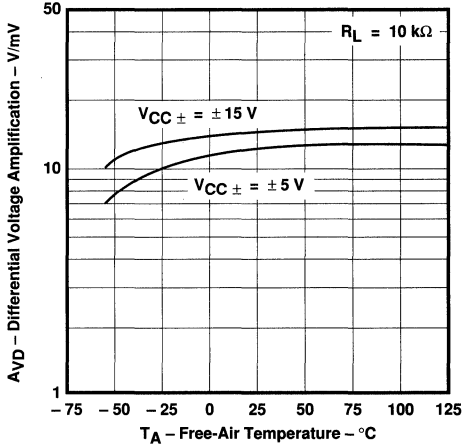


FIGURE 22

OUTPUT IMPEDANCE
VS
FREQUENCY

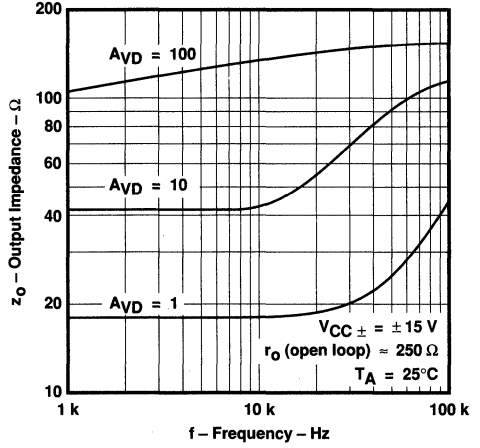


FIGURE 23

COMMON-MODE REJECTION RATIO
VS
FREQUENCY

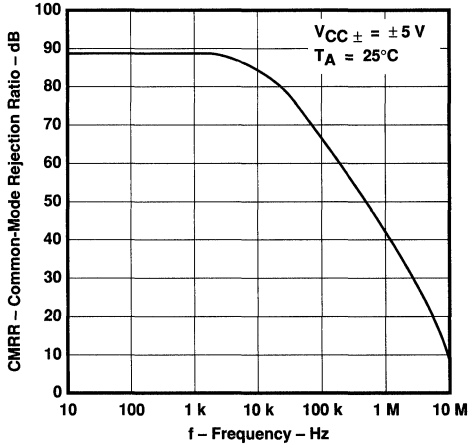


FIGURE 24

COMMON-MODE REJECTION RATIO
VS
FREQUENCY

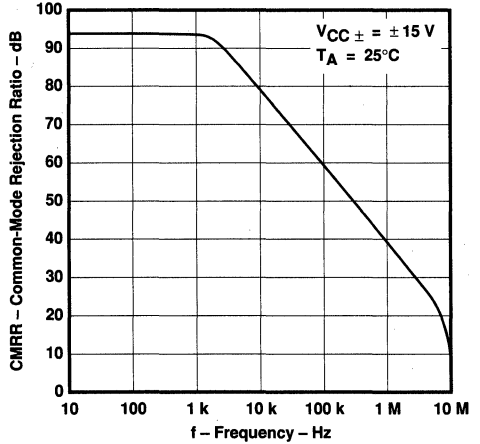


FIGURE 25

2

Operational Amplifiers

TL031, TL031A ENHANCED JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

COMMON-MODE REJECTION RATIO
VS
FREE-AIR TEMPERATURE

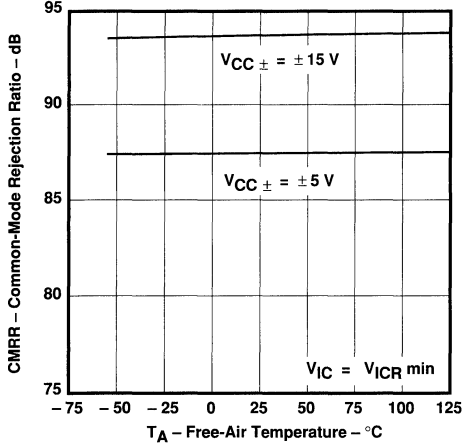


FIGURE 26

SUPPLY-VOLTAGE REJECTION RATIO
VS
FREE-AIR TEMPERATURE

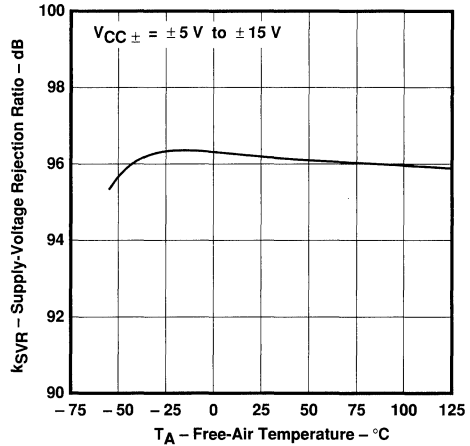


FIGURE 27

SHORT-CIRCUIT OUTPUT CURRENT
VS
SUPPLY VOLTAGE

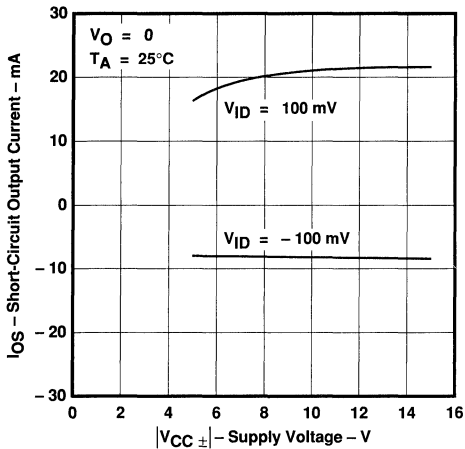


FIGURE 28

SHORT-CIRCUIT OUTPUT CURRENT
VS
TIME

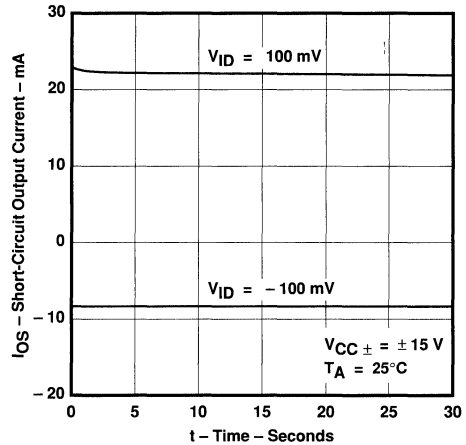


FIGURE 29

2
Operational Amplifiers

TL031, TL031A
ENHANCED JFET LOW-POWER LOW-OFFSET
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

SHORT-CIRCUIT OUTPUT CURRENT
 VS
 FREE-AIR TEMPERATURE

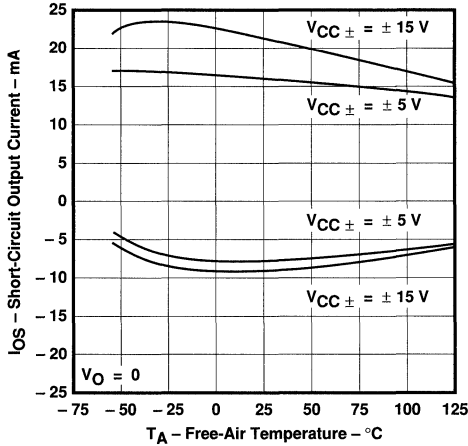


FIGURE 30

EQUIVALENT INPUT NOISE VOLTAGE
 VS
 FREQUENCY

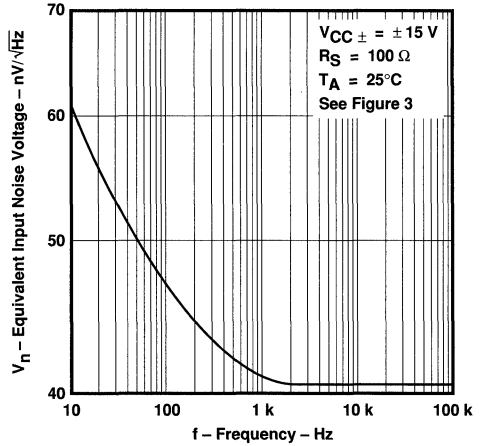


FIGURE 31

SUPPLY CURRENT
 VS
 SUPPLY VOLTAGE

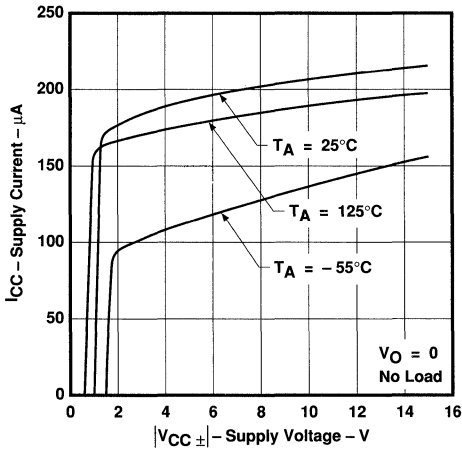


FIGURE 32

SUPPLY CURRENT
 VS
 FREE-AIR TEMPERATURE

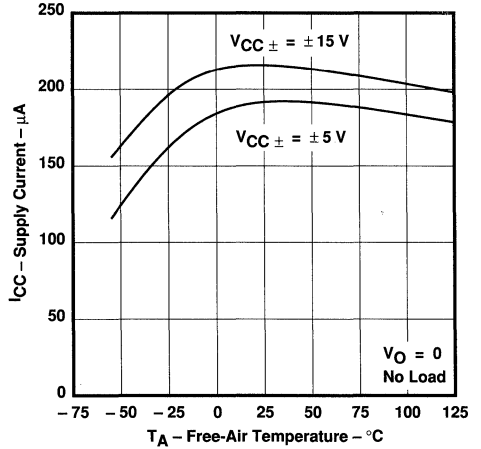


FIGURE 33

2
Operational Amplifiers

TYPICAL CHARACTERISTICS

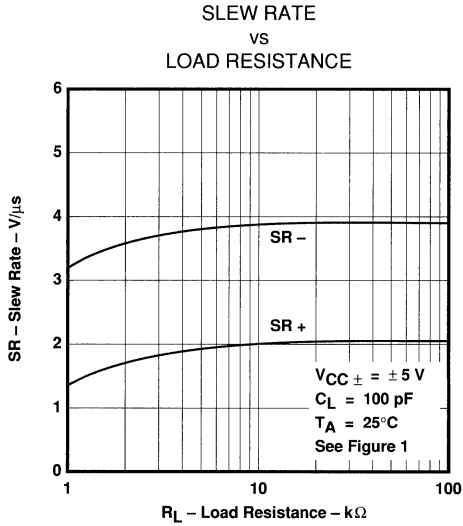


FIGURE 34

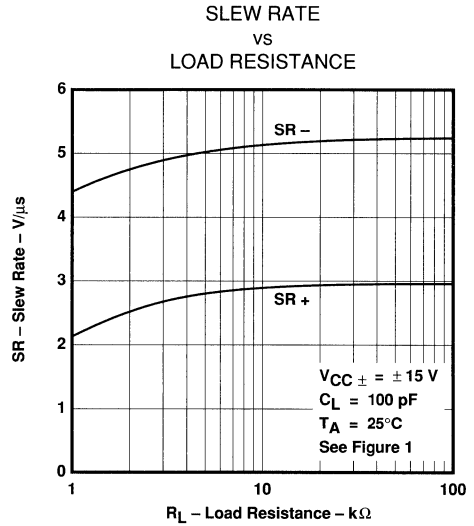


FIGURE 35

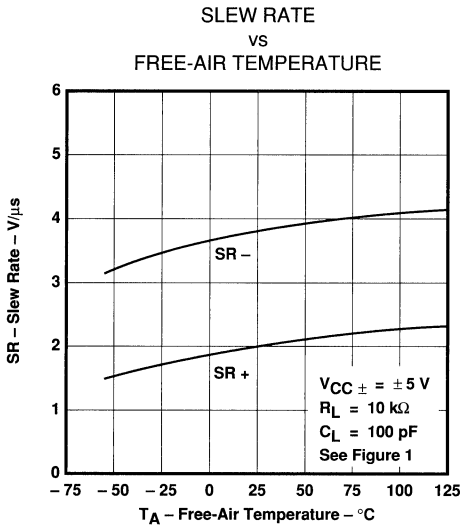


FIGURE 36

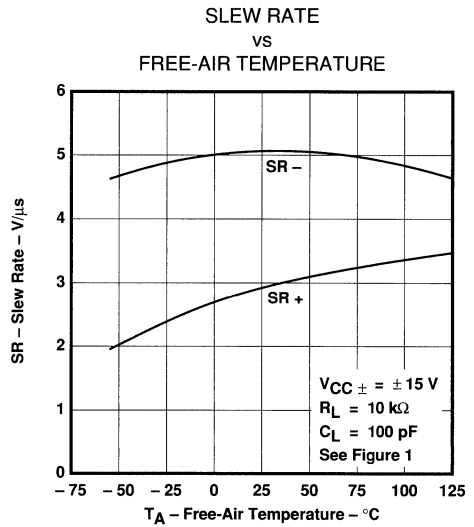


FIGURE 37

TL031, TL031A
ENHANCED JFET LOW-POWER LOW-OFFSET
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

**OVERSHOOT FACTOR
 VS
 LOAD CAPACITANCE**

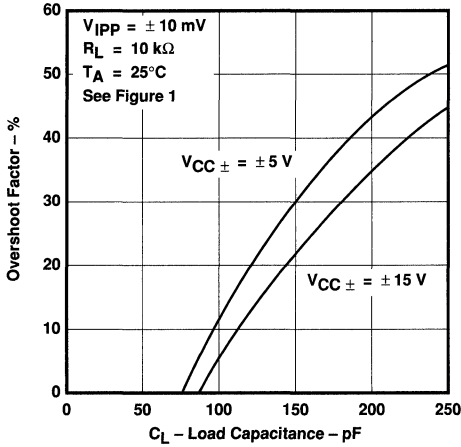


FIGURE 38

**TOTAL HARMONIC DISTORTION
 VS
 FREQUENCY**

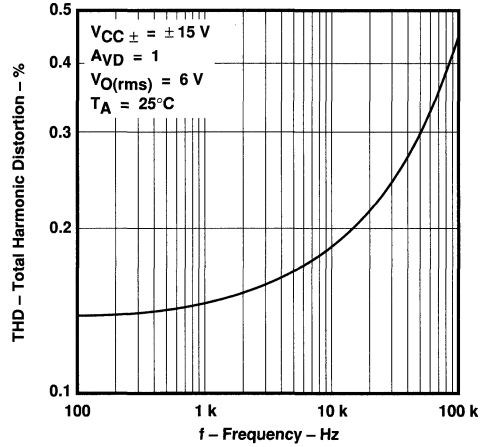


FIGURE 39

**UNITY-GAIN BANDWIDTH
 VS
 SUPPLY VOLTAGE**

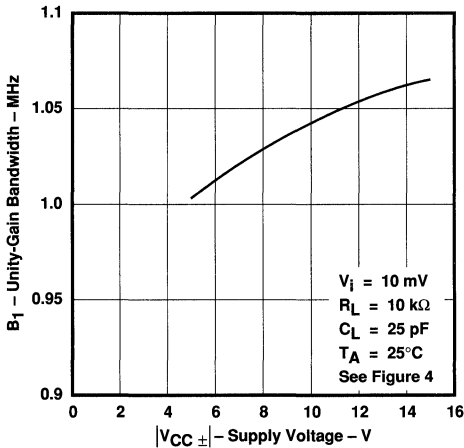


FIGURE 40

**UNITY-GAIN BANDWIDTH
 VS
 FREE-AIR TEMPERATURE**

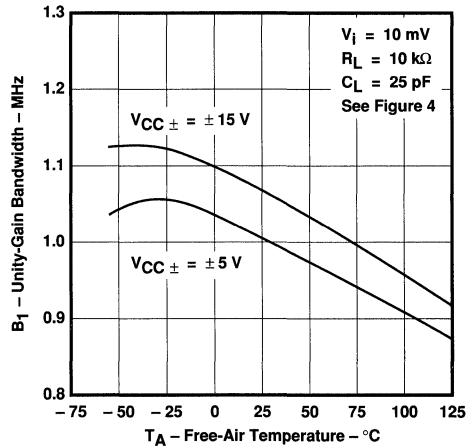


FIGURE 41

2
Operational Amplifiers

TYPICAL CHARACTERISTICS

PHASE MARGIN
 VS
 SUPPLY VOLTAGE

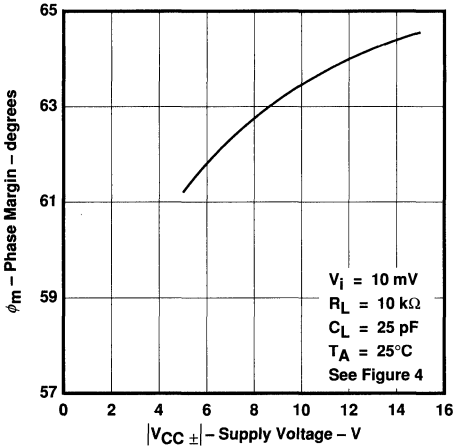


FIGURE 42

PHASE MARGIN
 VS
 LOAD CAPACITANCE

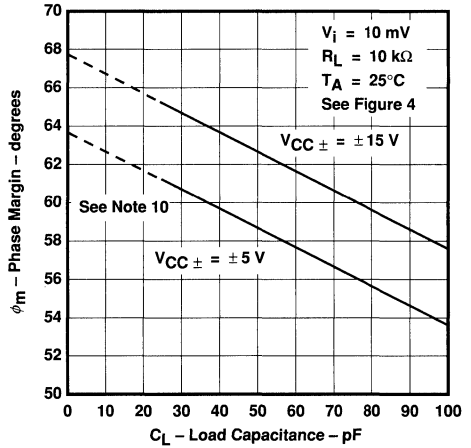


FIGURE 43

PHASE MARGIN
 VS
 FREE-AIR TEMPERATURE

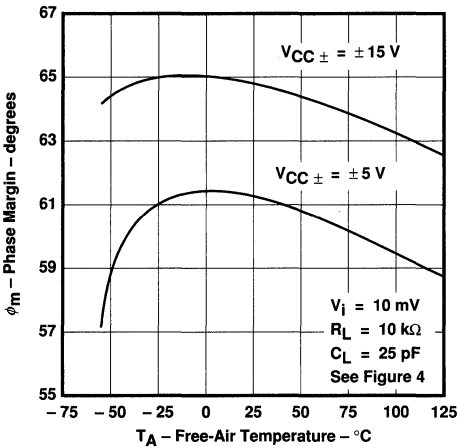


FIGURE 44

VOLTAGE-FOLLOWER
 SMALL-SIGNAL
 PULSE RESPONSE

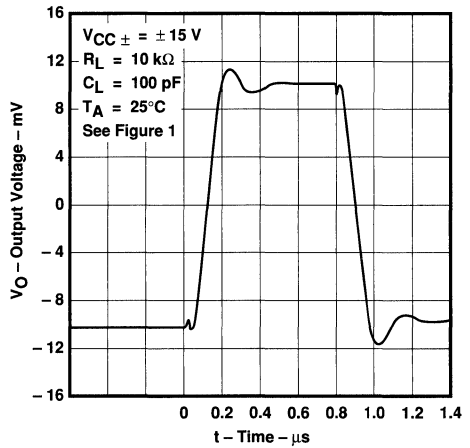


FIGURE 45

2
 Operational Amplifiers

NOTE 10: Values of phase margin below a load capacitance of 25 pF were estimated.

TL031, TL031A ENHANCED JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

VOLTAGE-FOLLOWER
LARGE-SIGNAL
PULSE RESPONSE

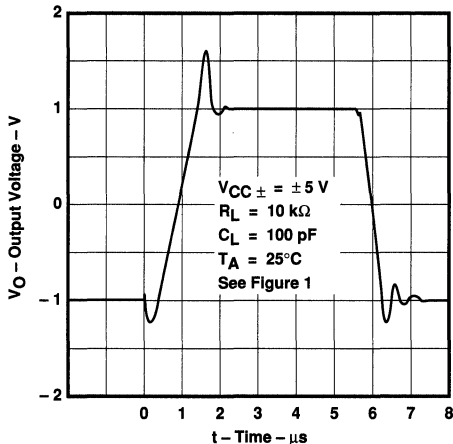


FIGURE 46

VOLTAGE-FOLLOWER
LARGE-SIGNAL
PULSE RESPONSE

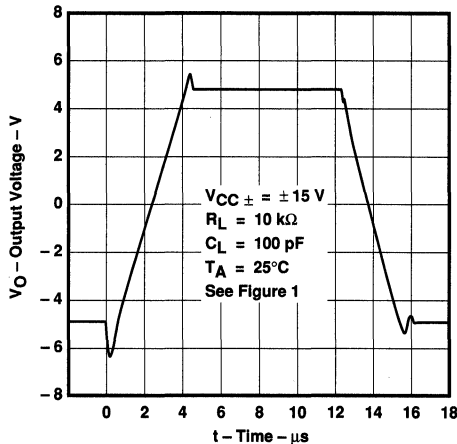


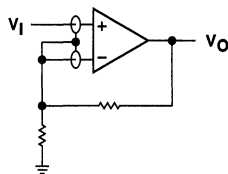
FIGURE 47

TYPICAL APPLICATION DATA

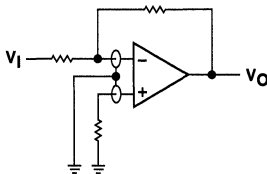
input characteristics

The TL031 and TL031A are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction.

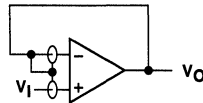
Because of the extremely high input impedance and resulting low bias current requirements, the TL031 and TL031A are well-suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause degradation in system performance. It is a good practice to include guard rings around inputs (see Figure 48). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.



(a) NONINVERTING AMPLIFIER



(b) INVERTING AMPLIFIER



(c) UNITY-GAIN AMPLIFIER

FIGURE 48. USE OF GUARD RINGS

TYPICAL APPLICATION DATA

output characteristics

All operating characteristics (except bandwidth and phase margin) are specified with 100-pF load capacitance. The TL031 and TL031A will drive higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem. Capacitive loads of 1000 pF and larger may be driven if enough resistance is added in series with the output (see Figure 49).

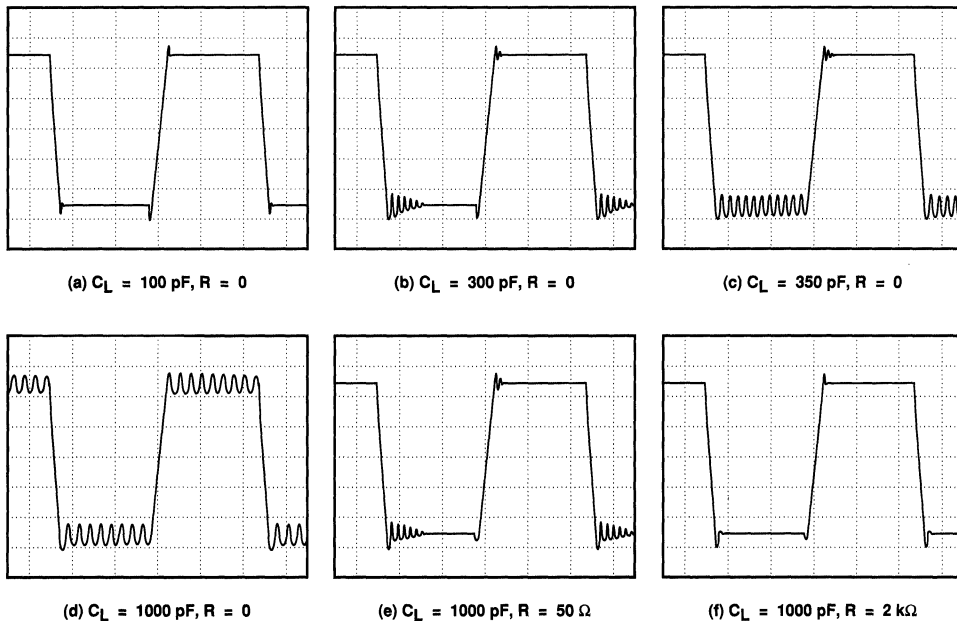
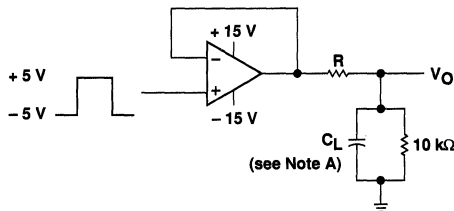


FIGURE 49. EFFECT OF CAPACITIVE LOADS



NOTE A: C_L includes fixture capacitance.

FIGURE 50. TEST CIRCUIT FOR OUTPUT CHARACTERISTICS

TL031, TL031A
ENHANCED JFET LOW-POWER LOW-OFFSET
OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

high-Q notch filter

In general, Texas Instruments enhanced JFET operational amplifiers serve as excellent filters. The circuit in Figure 51 provides a narrow notch at a specific frequency. Notch filters are designed to eliminate frequencies that are interfering with the operation of an application. For this filter, the center frequency can be calculated as:

$$f_0 = \frac{1}{2\pi R_1 C_1}$$

With the resistors and capacitors shown in Figure 51, the center frequency is 1 kHz. Note that $C_1 = C_3 = C_2 + 2$ and also that $R_1 = R_3 = 2 \times R_2$. The center frequency can be modified by varying these values. When adjusting the center frequency, be sure that the operational amplifier still has sufficient gain at the frequency of interest.

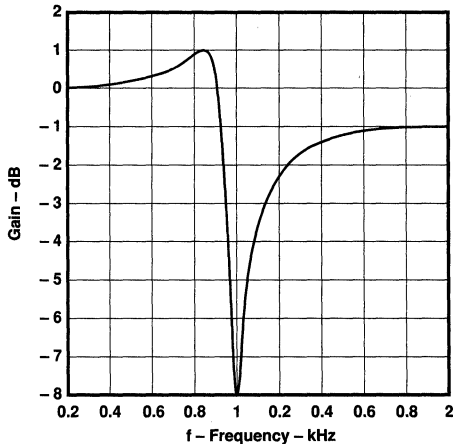
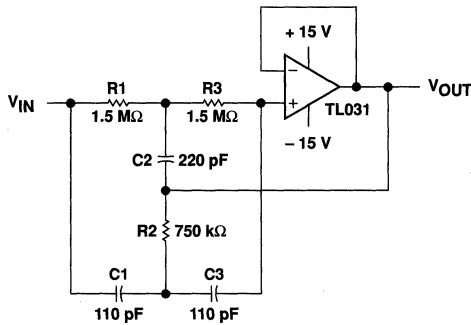


FIGURE 51. HIGH-Q NOTCH FILTER

TYPICAL APPLICATION DATA

transimpedance amplifier

The low-power precision TL031 allows accurate measurement of low currents. The high input impedance and low offset voltage of the TL031A greatly simplify the design of a transimpedance amplifier. At room temperature, this design achieves ten-bit accuracy with an error of less than 1/2 LSB.

Assuming that R2 is much less than R1 and ignoring error terms, the output voltage can be expressed as:

$$V_O = -I_{IN} \times R_F \left(\frac{R_1 + R_2}{R_2} \right)$$

Using the resistor values shown in the schematic, for a 1-nA input current, the output voltage equals -0.1 V. If the V_O limit for the TL031A is measured to be ± 12 V, the maximum input current for these resistor values is ± 120 nA. Similarly, one LSB on a ten-bit scale corresponds to 12 mV of output voltage or 120 pA of input current.

The following equation shows the effect of input offset voltage and input bias current on the output voltage:

$$V_O = -[V_{IO} + R_F(I_{IN} + I_{IB})] \left(\frac{R_1 + R_2}{R_2} \right)$$

If the application requires input protection for the transimpedance amplifier, do not use standard PN diodes. Instead, use low-leakage Siliconix SN4117 JFETs (or equivalent) connected as diodes across the TL031A inputs as shown in Figure 52.

As with all precision applications, special care must be taken to eliminate external sources of leakage and interference. Other precautions include using high-quality insulation, cleaning insulating surfaces to remove fluxes and other residue, and enclosing the application within a protective box.

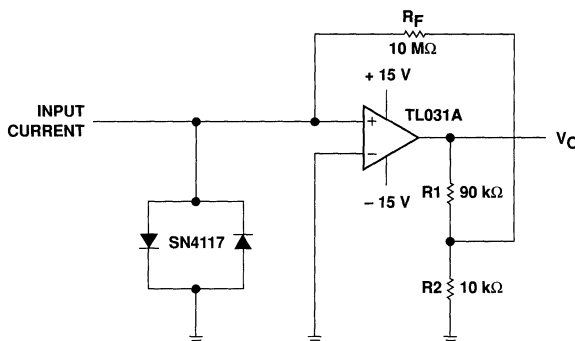


FIGURE 52. TRANSIMPEDANCE AMPLIFIER

TL031, TL031A ENHANCED JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

4- to 20-mA current loops

Often information from an analog sensor must be sent over a distance to the receiving circuitry. For many applications, the most feasible method involves converting voltage information to a current before transmission. The following circuits give two variations of low-power current loops. The circuit in Figure 53 requires three wires from the transmitting to receiving circuitry while the second variation in Figure 54 requires only two wires but includes an extra integrated circuit. Both circuits benefit from the high input impedance of the TL031A since many inexpensive sensors do not have low output impedance.

Assuming that the voltage at the noninverting input of the TL031A is zero, the following equation determines the output current:

$$I_O = V_{IN} \left(\frac{R_3}{R_1 \times R_S} \right) + 5 V \left(\frac{R_3}{R_2 \times R_S} \right) = 0.16 \times V_{IN} + 4 \text{ mA}$$

The circuits presently provide 4-to 20-mA output for an input voltage of 0 to 100 mV. By modifying R1, R2, and R3, the input voltage range or the output current range can be adjusted.

Including the offset voltage of the operational amplifier in the above equation clearly illustrates why the low offset TL031A was chosen:

$$I_O = V_{IN} \left(\frac{R_3}{R_1 \times R_S} \right) + 5 V \left(\frac{R_3}{R_2 \times R_S} \right) - V_{IO} \left(\frac{R_3}{R_1 \times R_S} + \frac{R_3}{R_2 \times R_S} + \frac{R_1}{R_S} \right) = 0.16 \times V_{IN} + 4 \text{ mA} - 0.17 \times V_{IO}$$

For example, an offset voltage of 1 mV decreases the output current by 0.17 mA.

Thanks to the low power consumption of the TL031A, both circuits have at least 2 mA available to drive the actual sensor from the 5-V reference node.

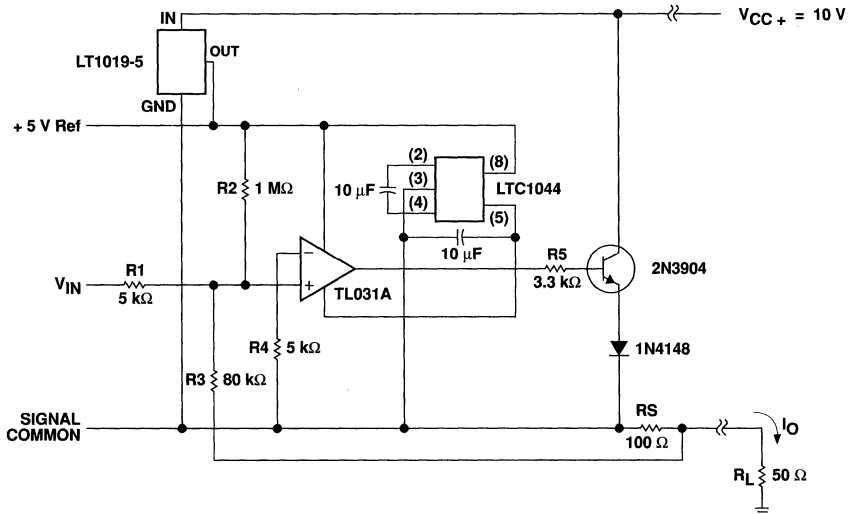


FIGURE 53. 2-WIRE 4- TO 20-mA CURRENT LOOP

TL031, TL031A
ENHANCED JFET LOW-POWER LOW-OFFSET
OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

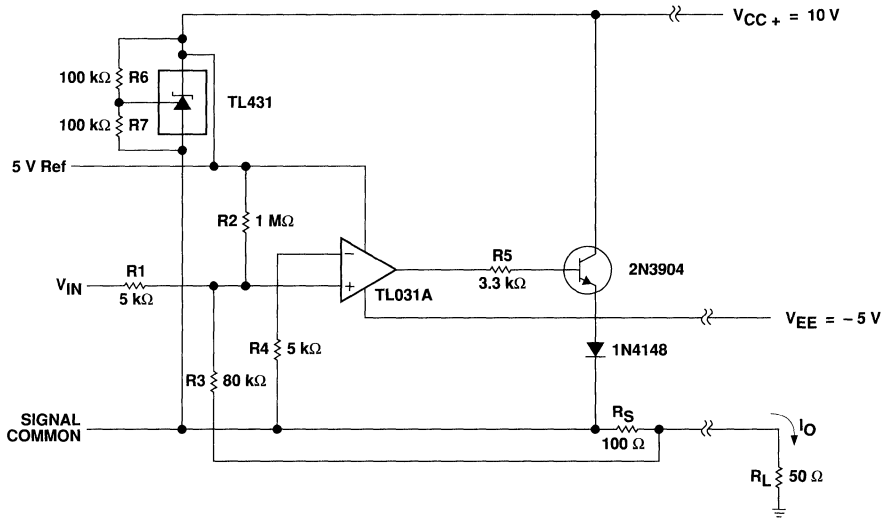


FIGURE 54. 3-WIRE 4- TO 20-mA CURRENT LOOP

2

Operational Amplifiers

2

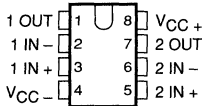
Operational Amplifiers

TL032, TL032A ENHANCED JFET LOW-POWER LOW-OFFSET DUAL OPERATIONAL AMPLIFIERS

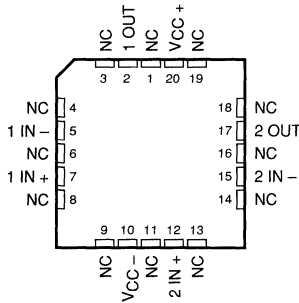
D3152, JULY 1988 – REVISED JANUARY 1989

- Maximum Offset Voltage ... 800 μV
- High Slew Rate ... 2.9 $\text{V}/\mu\text{s}$ Typ
- Low Input Bias Current ... 2 pA Typ
- Very Low Power Consumption ... 13 mW Typ
- Output Short-Circuit Protection

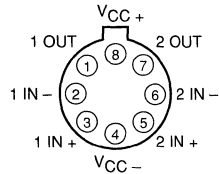
**D, JG, or P PACKAGE
(TOP VIEW)**



**FK PACKAGE
(TOP VIEW)**



**L PACKAGE
(TOP VIEW)**



Pin 4 (L Package) is in electrical contact with the case

NC – No internal connection

description

The TL032 and TL032A dual operational amplifiers incorporate well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. These devices offer the significant advantages of Texas Instruments new enhanced JFET process. This process affords not only low initial offset voltage due to the on-chip zener trim capability but also stable offset voltage over time and temperature. In comparison, traditional JFET processes are plagued by significant offset voltage drift.

This new enhanced process still maintains the traditional JFET advantages of fast slew rates and low input bias and offset currents. These advantages coupled with low power consumption make the TL032 well-suited for new state-of-the-art designs as well as existing design upgrades. The TL032 has been designed to be

2

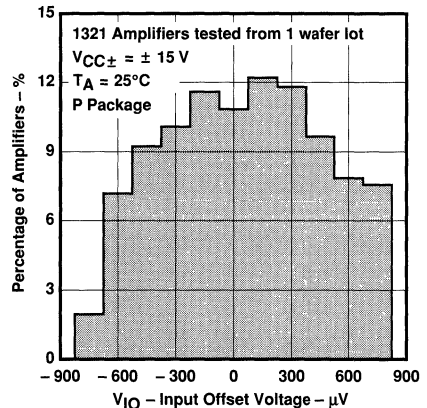
Operational Amplifiers

AVAILABLE OPTIONS

T _A	V _{IO} max AT 25°C	PACKAGE				
		SMALL- OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)
0°C to 70°C	0.8 mV	TL032ACD	—	TL032ACJG	TL032ACL	TL032ACP
	1.5 mV	TL032CD	—	TL032CJG	TL032CL	TL032CP
-40°C to 85°C	0.8 mV	TL032AID	—	TL032AIJG	TL032AIL	TL032AIP
	1.5 mV	TL032ID	—	TL032IJG	TL032IL	TL032IP
-55°C to 125°C	0.8 mV	TL032AMD	TL032AMFK	TL032AMJG	TL032AML	TL032AMP
	1.5 mV	TL032MD	TL032MFK	TL032MJG	TL032ML	TL032MP

D packages are available taped-and-reeled. Add "R" suffix to device type when ordering (e.g., TL032CDR).

DISTRIBUTION OF TL032A
INPUT OFFSET VOLTAGE



PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS
INSTRUMENTS**

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2-219

TL032, TL032A

ENHANCED JFET LOW-POWER LOW-OFFSET

DUAL OPERATIONAL AMPLIFIERS

description (continued)

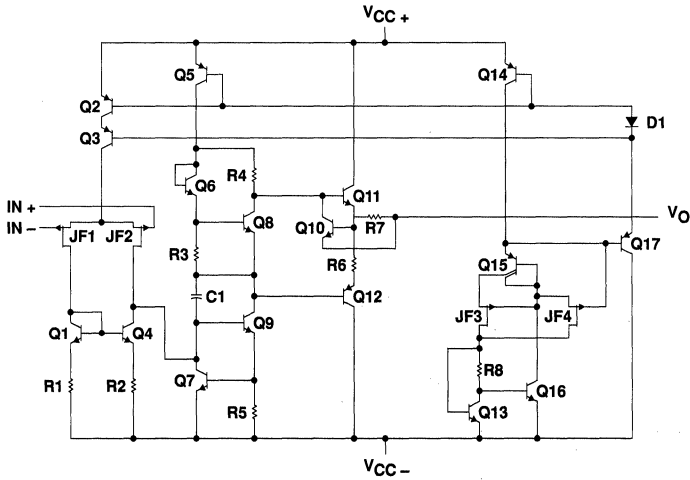
functionally compatible and pin compatible with the TL062. Two offset voltage grades are available: TL032 (1.5 mV max) and TL032A (800 μ V max).

A variety of available packaging options includes small-outline and chip carrier versions for high density system applications.

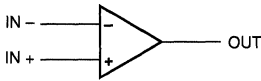
The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I-suffix devices are characterized for operation from -40°C to 85°C . The C-suffix devices are characterized for operation from 0°C to 70°C .

2 equivalent schematic (each amplifier)

Operational Amplifiers



symbol (each amplifier)



TL032, TL032A ENHANCED JFET LOW-POWER LOW-OFFSET DUAL OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	18 V
Supply voltage, V_{CC-} (see Note 1)	-18 V
Differential input voltage (see Note 2)	± 30 V
Input voltage range, V_I (any input, see Notes 1 and 3)	± 15 V
Input current, I_I (each input)	± 1 mA
Output current, I_O (each output)	± 40 mA
Total current into V_{CC+} terminal	160 mA
Total current out of V_{CC-} terminal	160 mA
Duration of short-circuit current at (or below) 25°C (see Note 4)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, T_A : M-suffix	-55°C to 125°C
I-suffix	-40°C to 85°C
C-suffix	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG or L package	300°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input with respect to the inverting input.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
 4. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
L	825 mW	6.6 mW/°C	528 mW	429 mW	165 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW

recommended operating conditions

		M-SUFFIX		I-SUFFIX		C-SUFFIX		UNIT
		MIN	NOM MAX	MIN	NOM MAX	MIN	NOM MAX	
Supply voltage, V_{CC}		± 5	± 15	± 5	± 15	± 5	± 15	V
Common-mode input voltage, V_{IC}	$V_{CC} \pm \pm 5$ V	-1.5	4	-1.5	4	-1.5	4	V
	$V_{CC} \pm \pm 15$ V	-11.5	14	-11.5	14	-11.5	14	
Operating free-air temperature, T_A		-55	125	-40	85	0	70	°C

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Operational Amplifiers

TL032M, TL032AM ENHANCED JFET LOW-POWER LOW-OFFSET DUAL OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL032M	25°C	0.69	3.5	0.57	1.5	mV	
			Full range	6.5		4.5			
		TL032AM	25°C	0.53	2.8	0.39	0.8		
			Full range	5.8		3.8			
α _{VIO} Temperature coefficient of input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL032M	25°C to 125°C	9.7		9.7		μV/°C	
		TL032AM	25°C to 125°C	9.7		9.7			
Input offset voltage long-term drift (see Note 5)		25°C	0.04		0.04		μV/mo		
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	1	100	1	100	pA		
		125°C	0.2	10	0.2	10	nA		
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	2	200	2	200	pA		
		125°C	7	20	8	20	nA		
V _{ICR} Common-mode input voltage range		25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4	V		
		Full range	-1.5 to 4		-11.5 to 14				
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.3	13	14	V		
		-55°C	3	4.1	13	14			
		125°C	3	4.4	13	14			
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-3	-4.2	-12.5	-13.9	V		
		-55°C	-3	-4	-12.5	-13.8			
		125°C	-3	-4.3	-12.5	-14			
A _{VD} Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	4	12	5	14.3	V/mV		
		-55°C	3	7.1	4	10.4			
		125°C	3	12.9	4	15			
r _i Input resistance		25°C	10 ¹²		10 ¹²		Ω		
C _i Input capacitance		25°C	5		4		pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	70	87	75	94	dB		
		-55°C	70	87	70	94			
		125°C	70	87	70	94			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75	96	75	96	dB		
		-55°C	75	95	75	95			
		125°C	75	96	75	96			
P _D Total power dissipation (two amplifiers)	No load, V _O = 0	25°C	3.8	5	13	17	mW		
		-55°C	2.3	5	9.4	17			
		125°C	3.6	5	11.8	17			
I _{CC} Supply current (two amplifiers)	No load, V _O = 0	25°C	384	500	434	560	μA		
		-55°C	228	500	312	560			
		125°C	356	500	394	560			
V _{O1} /V _{O2} Crosstalk attenuation	A _{VD} = 100	25°C	120		120		dB		

[†] Full range is -55°C to 125°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. At V_{CC} ± = ± 5 V, V_O = ± 2.3 V; at V_{CC} ± = ± 15 V, V_O = ± 10 V.

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Operational Amplifiers

TL032M, TL032AM
ENHANCED JFET LOW-POWER LOW-OFFSET
DUAL OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER		TEST CONDITIONS		T _A	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT	
					MIN	TYP	MAX	MIN	TYP	MAX		
SR +	Positive slew rate at unity gain	R _L = 10 kΩ, C _L = 100 pF, See Figure 1, See Note 7		25°C	2			2			V/μs	
				-55°C	1.4			1.2				
				125°C	2.4			1.2				
SR -	Negative slew rate at unity gain	See Figure 1, See Note 7		25°C	3.9			3			V/μs	
				-55°C	3.2			2.5				
				125°C	4.1			2.5				
t _r	Rise time	V _I PP = ± 10 mV, R _L = 10 kΩ, C _L = 100 pF, See Figures 1 and 2		25°C	138			132			ns	
				-55°C	142			123				
				125°C	166			158				
t _f	Fall time	See Figures 1 and 2		25°C	138			132			ns	
				-55°C	142			123				
				125°C	166			158				
Overshoot factor				25°C	11%			5%				
				-55°C	16%			6%				
				125°C	14%			8%				
V _n	Equivalent input noise voltage	TL032M	R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C	49			49			nV/√Hz
				f = 1 kHz	41			41				
		TL032AM		f = 10 Hz	25°C	49			49			
				f = 1 kHz	41			41				
I _n	Equivalent input noise current	f = 1 kHz		25°C	0.003			0.003			pA/√Hz	
B1	Unity-gain bandwidth	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	1			1.1			MHz	
				-55°C	1			1.1				
				125°C	0.9			0.9				
φ _m	Phase margin at unity gain	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	61°			65°				
				-55°C	57°			64°				
				125°C	59°			62°				

NOTE 7: For V_{CC} ± = ± 5 V, V_IPP = ± 1 V; for V_{CC} ± = ± 15 V, V_IPP = ± 5 V.

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Operational Amplifiers

TL032I, TL032AI

ENHANCED JFET LOW-POWER LOW-OFFSET

DUAL OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage		25°C	0.69 3.5			0.57 1.5			mV
			Full range			3.3			
		Full range	0.53 2.8			0.39 0.8			
			4.6			2.6			
α _{VIO} Temperature coefficient of input offset voltage (see Note 8)	V _O = 0, V _{IC} = 0, R _S = 50 Ω	25°C to 85°C	11.4			10.8			μV/°C
		25°C to 85°C	11.4			10.8 25			
Input offset voltage long-term drift (see Note 5)		25°C	0.04			0.04			μV/mo
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	1 100			1 100			pA
		85°C	0.02 0.45			0.02 0.45			nA
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	2 200			2 200			pA
		85°C	0.2 0.9			0.3 0.9			nA
V _{ICR} Common-mode input voltage range		25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4			V
		Full range	-1.5 to 4		-11.5 to 14				
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3 4.3		13 14				V
		-40°C	3 4.1		13 14				
		85°C	3 4.4		13 14				
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-3 -4.2		-12.5 -13.9				V
		-40°C	-3 -4.1		-12.5 -13.8				
		85°C	-3 -4.2		-12.5 -14				
A _{VD} Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	4 12		5 14.3				V/mV
		-40°C	3 8.4		4 11.6				
		85°C	4 13.5		5 15.3				
r _i Input resistance		25°C	10 ¹²			10 ¹²			Ω
C _i Input capacitance		25°C	5			4			pF
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	70 87		75 94				dB
		-40°C	70 87		75 94				
		85°C	70 87		75 94				
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75 96		75 96				dB
		-40°C	75 96		75 96				
		85°C	75 96		75 96				
P _D Total power dissipation (two amplifiers)	No load, V _O = 0	25°C	3.8 5		13 17				mW
		-40°C	2.9 5		10.9 17				
		85°C	3.7 5		12.4 17				
I _{CC} Supply current (two amplifiers)	No load, V _O = 0	25°C	384 500		434 560				μA
		-40°C	288 500		362 560				
		85°C	372 500		414 560				
V _{O1} /V _{O2} Crosstalk attenuation	A _{VD} = 100	25°C	120			120			dB

[†] Full range is -40°C to 85°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. At V_{CC} ± = ± 5 V, V_O = ± 2.3 V; at V_{CC} ± = ± 15 V, V_O = ± 10 V.

8. This parameter is tested on a sample basis for the TL032A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

TL032I, TL032AI
ENHANCED JFET LOW-POWER LOW-OFFSET
DUAL OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER		TEST CONDITIONS		T _A	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
					MIN	TYP	MAX	MIN	TYP	MAX	
SR +	Positive slew rate at unity gain	R _L = 10 kΩ, C _L = 100 pF, See Figure 1, See Note 7		25°C	2			2			V/μs
				-40°C	1.6			1.5			
				85°C	2.3			2			
SR -	Negative slew rate at unity gain	See Figure 1, See Note 7		25°C	3.9			3.5			V/μs
				-40°C	3.3			3.2			
				85°C	4.1			3.2			
t _r	Rise time	V _I PP = ± 10 mV, R _L = 10 kΩ, C _L = 100 pF, See Figures 1 and 2		25°C	138			132			ns
				-40°C	132			123			
				85°C	154			146			
t _f	Fall time	See Figures 1 and 2		25°C	138			132			ns
				-40°C	132			123			
				85°C	154			146			
Overshoot factor				25°C	11%			5%			
				-40°C	12%			5%			
				85°C	13%			7%			
V _n	Equivalent input noise voltage (see Note 9)	TL032I	R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C			49			nV/√Hz
				f = 1 kHz	41			41			
		TL032AI		f = 10 Hz	49			49			
				f = 1 kHz	41			41			
I _n		Equivalent input noise current		f = 1 kHz	25°C			0.003			pA/√Hz
B1	Unity-gain bandwidth	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	1			1.1			MHz
				-40°C	1			1.1			
				85°C	0.9			1			
φ _m	Phase margin at unity gain	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	61°			65°			
				-40°C	60°			65°			
				85°C	60°			64°			

NOTES: 7. For V_{CC} ± = ± 5 V, V_IPP = ± 1 V; for V_{CC} ± = ± 15 V, V_IPP = ± 5 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

TL032C, TL032AC

ENHANCED JFET LOW-POWER LOW-OFFSET

DUAL OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC±} = ±5 V			V _{CC±} = ±15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL032C	25°C	0.69	3.5	0.57	1.5	mV	
			Full range	4.5			2.5		
		TL032AC	25°C	0.53	2.8	0.39	0.8		
			Full range	3.8			1.8		
α _{VIO} Temperature coefficient of input offset voltage (see Note 8)	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL032C	25°C to 70°C	11.5			10.8	μV/°C	
		TL032AC	25°C to 70°C	11.5			10.8		25
Input offset voltage long-term drift (see Note 5)		25°C	0.04			0.04	μV/mo		
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	1 100			1 100	pA		
		70°C	9 200			12 200			
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	2 200			2 200	pA		
		70°C	50 400			80 400			
V _{ICR} Common-mode input voltage range		25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4	V		
		Full range	-1.5 to 4		-11.5 to 14				
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.3	13	14	V		
		0°C	3	4.2	13	14			
		70°C	3	4.3	13	14			
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-3	-4.2	-12.5	-13.9	V		
		0°C	-3	-4.1	-12.5	-13.9			
		70°C	-3	-4.2	-12.5	-14			
A _{VD} Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	4	12	5	14.3	V/mV		
		0°C	3	11.1	4	13.5			
		70°C	4	13.3	5	15.2			
r _i Input resistance		25°C	10 ¹²			10 ¹²	Ω		
C _i Input capacitance		25°C	5			4	pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	70	87	75	94	dB		
		0°C	70	87	75	94			
		70°C	70	87	75	94			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC±} / ΔV _{IO})	V _{CC±} = ±5 V to ±15 V, V _O = 0, R _S = 50 Ω	25°C	75	96	75	96	dB		
		0°C	75	96	75	96			
		70°C	75	96	75	96			
P _D Total power dissipation (two amplifiers)	No load, V _O = 0	25°C	3.8	5	13	17	mW		
		0°C	3.7	5	12.7	17			
		70°C	3.8	5	12.6	17			
I _{CC} Supply current (two amplifiers)	No load, V _O = 0	25°C	384	500	434	560	μA		
		0°C	368	500	422	560			
		70°C	378	500	420	560			
V _{O1} /V _{O2} Crosstalk attenuation	A _{VD} = 100	25°C	120			120	dB		

[†] Full range is 0°C to 70°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. At V_{CC±} = ±5 V, V_O = ±2.3 V; at V_{CC±} = ±15 V, V_O = ±10 V.

8. This parameter is tested on a sample basis for the TL032A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

TL032C, TL032AC ENHANCED JFET LOW-POWER LOW-OFFSET DUAL OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER		TEST CONDITIONS		T _A	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
					MIN	TYP	MAX	MIN	TYP	MAX	
SR +	Positive slew rate at unity gain	R _L = 10 kΩ, C _L = 100 pF, See Figure 1, See Note 7		25°C	2			2			V/μs
				0°C	1.8			1.5			
				70°C	2.2			2			
SR -	Negative slew rate at unity gain	See Figure 1, See Note 7		25°C	3.9			3.5			V/μs
				0°C	3.7			3.2			
				70°C	4			3.2			
t _r	Rise time	V _{Ipp} = ± 10 mV, R _L = 10 kΩ, C _L = 100 pF, See Figures 1 and 2		25°C	138			132			ns
				0°C	134			127			
				70°C	150			142			
t _f	Fall time	See Figures 1 and 2		25°C	138			132			ns
				0°C	134			127			
				70°C	150			142			
Overshoot factor				25°C	11%			5%			
				0°C	10%			4%			
				70°C	12%			6%			
V _n	Equivalent input noise voltage (see Note 9)	TL032C	R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C			49			nV/√Hz
				f = 1 kHz	41			41			
		TL032AC		f = 10 Hz	25°C			49			
				f = 1 kHz	41			41			
I _n		Equivalent input noise current		f = 1 kHz	25°C	0.003			pA/√Hz		
B1	Unity-gain bandwidth	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	1			1.1			MHz
				0°C	1			1.1			
				70°C	1			1			
φ _m	Phase margin at unity gain	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	61°			65°			
				0°C	61°			65°			
				70°C	60°			64°			

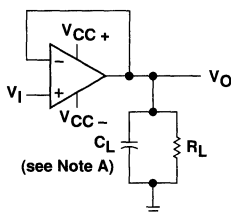
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Operational Amplifiers

NOTES: 7. For V_{CC} ± = ± 5 V, V_{Ipp} = ± 1 V; for V_{CC} ± = ± 15 V, V_{Ipp} = ± 5 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

PARAMETER MEASUREMENT INFORMATION



NOTE A: C_L includes fixture capacitance.

FIGURE 1. SLEW RATE, RISE/FALL TIME, AND OVERSHOOT TEST CIRCUIT

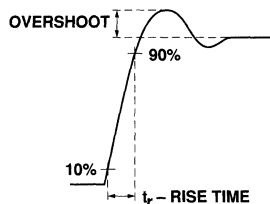


FIGURE 2. RISE TIME AND OVERSHOOT WAVEFORM

**TL032, TL032A
ENHANCED JFET LOW-POWER LOW-OFFSET
DUAL OPERATIONAL AMPLIFIERS**

PARAMETER MEASUREMENT INFORMATION

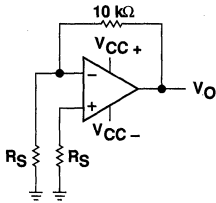
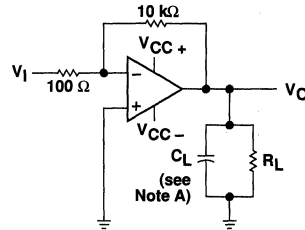
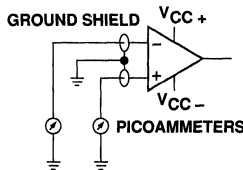


FIGURE 3. NOISE VOLTAGE TEST CIRCUIT



NOTE A: C_L includes fixture capacitance.

**FIGURE 4. UNITY-GAIN BANDWIDTH AND
PHASE MARGIN TEST CIRCUIT**



**FIGURE 5. INPUT BIAS AND OFFSET
CURRENT TEST CIRCUIT**

2

Operational Amplifiers

typical values

Typical values as presented in this data sheet represent the median (50% point) of device parametric performance.

input bias and offset current

At the picoamp bias current level typical of the TL032 and TL032A, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To accurately measure these small currents Texas Instruments uses a two step process. The socket leakage is measured using picoammeters with bias voltages applied but with no device in the socket. The device is then inserted into the socket and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

noise

Because of the increasing emphasis on low noise levels in many of today's applications, the input noise voltage density is sample-tested at $f = 1$ kHz. Texas Instruments also has additional noise testing capability to meet specific application requirements. Please contact the factory for details.

TL032, TL032A
ENHANCED JFET LOW-POWER LOW-OFFSET
DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

table of graphs

		FIGURE	
V_{IO}	Input offset voltage	Distribution	6
αV_{IO}	Temperature coefficient of input offset voltage	Distribution	7
I_{IO}	Input offset current	vs Temperature	8
I_{IB}	Input bias current	vs V_{IC}	9
		vs Temperature	8
V_I	Input voltage range	vs V_{CC}	10
		vs Temperature	11
V_{ID}	Differential input voltage	vs Output voltage	12, 13
		vs V_{CC}	14
V_{OM}	Maximum peak output voltage swing	vs Output current	16, 17
		vs Frequency	15
		vs Temperature	18, 19
A_{VD}	Differential voltage amplification	vs R_L	20
		vs Frequency	21
		vs Temperature	22
z_o	Output impedance	vs Frequency	23
CMRR	Common-mode rejection ratio	vs Frequency	24, 25
		vs Temperature	26
ksVR	Supply-voltage rejection ratio	vs Temperature	27
		vs V_{CC}	28
I_{OS}	Short-circuit output current	vs Time	29
		vs Temperature	30
I_{CC}	Supply current	vs V_{CC}	32
		vs Temperature	33
SR	Slew rate	vs R_L	34, 35
		vs Temperature	36, 37
	Overshoot factor	vs C_L	38
V_n	Equivalent input noise voltage	vs Frequency	31
THD	Total harmonic distortion	vs Frequency	39
B_1	Unity-gain bandwidth	vs V_{CC}	40
		vs Temperature	41
		vs V_{CC}	42
ϕ_m	Phase margin	vs C_L	43
		vs Temperature	44
	Phase shift	vs Frequency	21
	Pulse response	Small-signal	45
		Large-signal	46, 47

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Operational Amplifiers

TL032, TL032A
ENHANCED JFET LOW-POWER LOW-OFFSET
DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

DISTRIBUTION OF TL032
INPUT OFFSET VOLTAGE

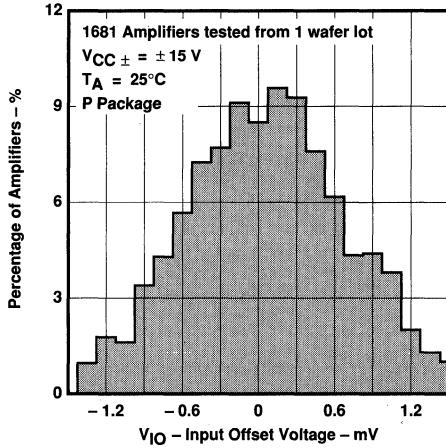


FIGURE 6

DISTRIBUTION OF TL032
INPUT OFFSET VOLTAGE
TEMPERATURE COEFFICIENT

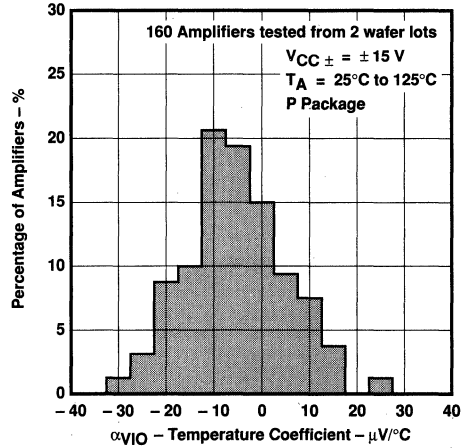


FIGURE 7

INPUT BIAS CURRENT AND
INPUT OFFSET CURRENT
vs
FREE-AIR TEMPERATURE

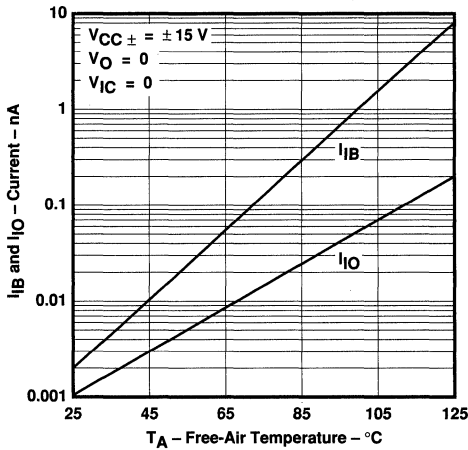


FIGURE 8

INPUT BIAS CURRENT
vs
COMMON-MODE INPUT VOLTAGE

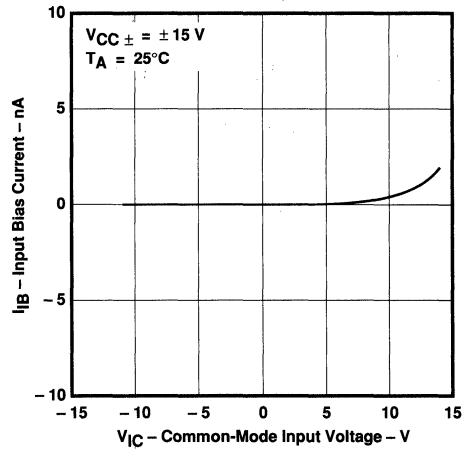


FIGURE 9

2
Operational Amplifiers

TL032, TL032A
ENHANCED JFET LOW-POWER LOW-OFFSET
DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

COMMON-MODE
 INPUT VOLTAGE RANGE LIMITS
 vs
 SUPPLY VOLTAGE

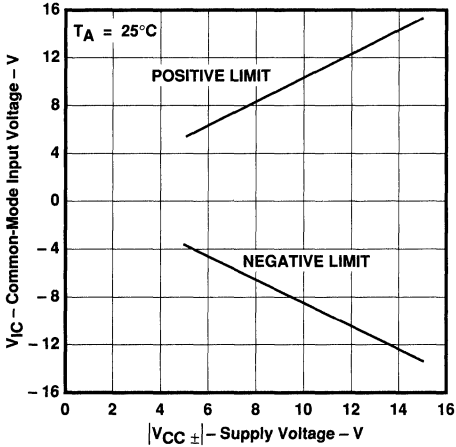


FIGURE 10

COMMON-MODE
 INPUT VOLTAGE RANGE LIMITS
 vs
 FREE-AIR TEMPERATURE

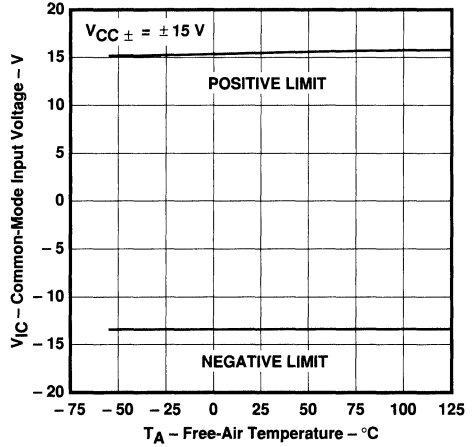


FIGURE 11

DIFFERENTIAL INPUT VOLTAGE
 vs
 OUTPUT VOLTAGE

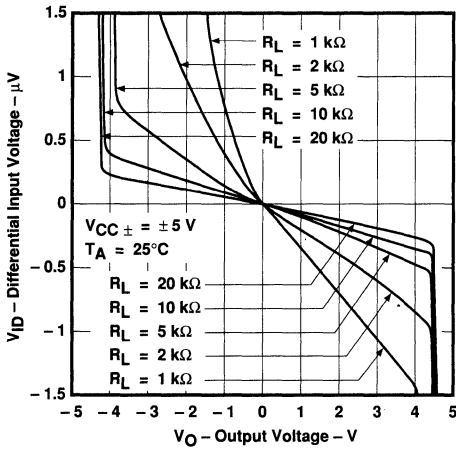


FIGURE 12

DIFFERENTIAL INPUT VOLTAGE
 vs
 OUTPUT VOLTAGE

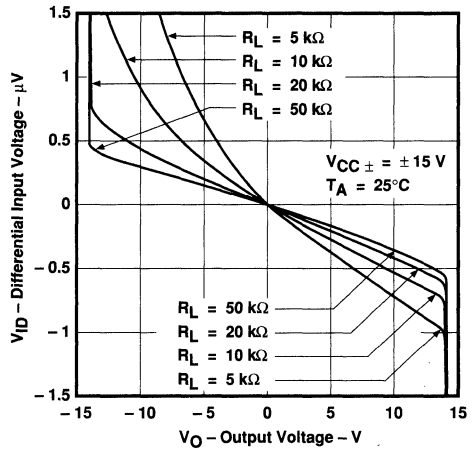


FIGURE 13

TL032, TL032A
ENHANCED JFET LOW-POWER LOW-OFFSET
DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

2

Operational Amplifiers

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 SUPPLY VOLTAGE

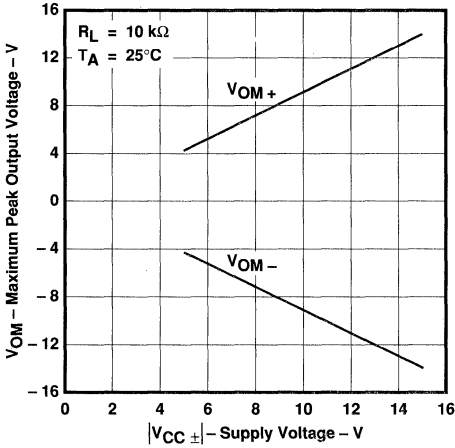


FIGURE 14

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 VS
 FREQUENCY

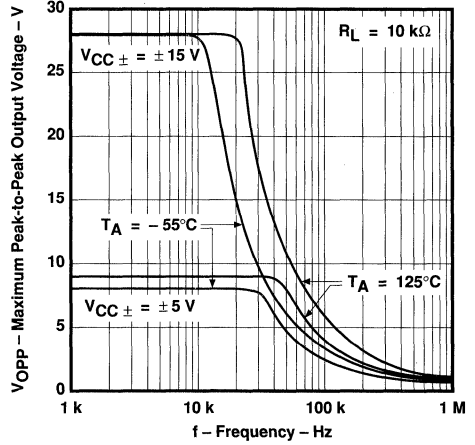


FIGURE 15

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 OUTPUT CURRENT

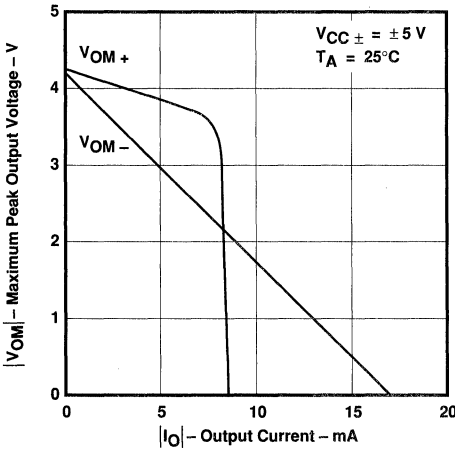


FIGURE 16

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 OUTPUT CURRENT

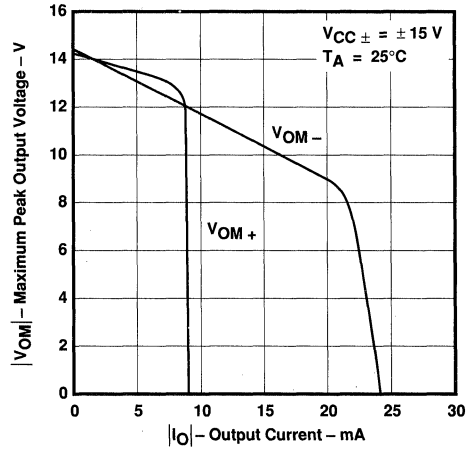


FIGURE 17

TL032, TL032A ENHANCED JFET LOW-POWER LOW-OFFSET DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

MAXIMUM PEAK OUTPUT VOLTAGE
VS
FREE-AIR TEMPERATURE

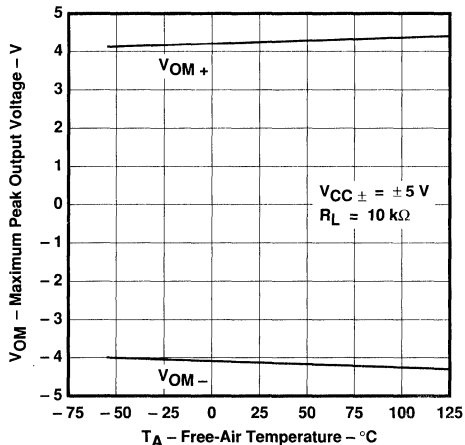


FIGURE 18

MAXIMUM PEAK OUTPUT VOLTAGE
VS
FREE-AIR TEMPERATURE

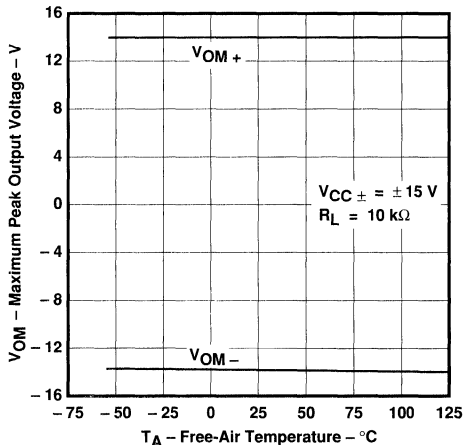


FIGURE 19

LARGE-SIGNAL VOLTAGE AMPLIFICATION
VS
LOAD RESISTANCE

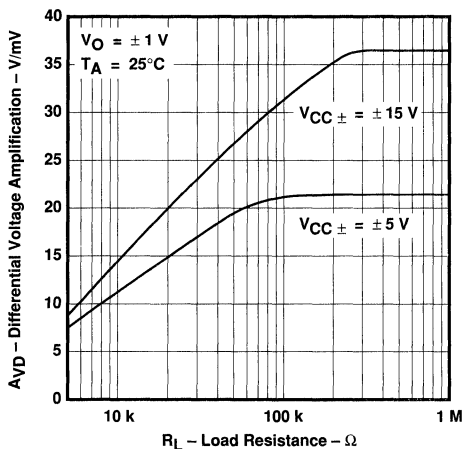


FIGURE 20

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
VS
FREQUENCY

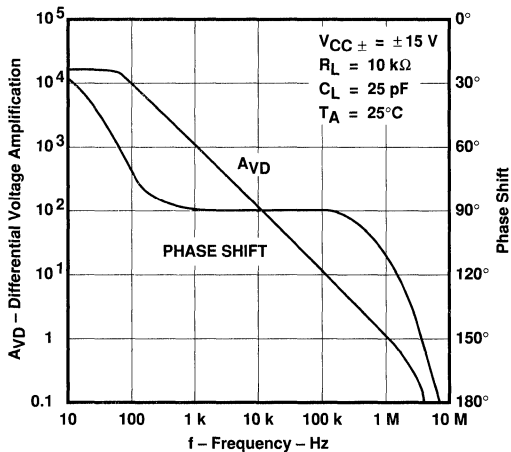


FIGURE 21

TL032, TL032A
ENHANCED JFET LOW-POWER LOW-OFFSET
DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

LARGE-SIGNAL VOLTAGE AMPLIFICATION
VS
FREE-AIR TEMPERATURE

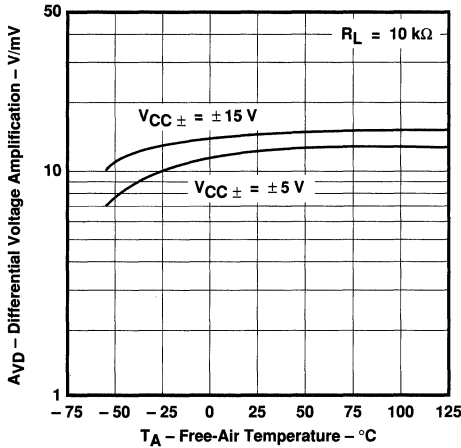


FIGURE 22

OUTPUT IMPEDANCE
VS
FREQUENCY

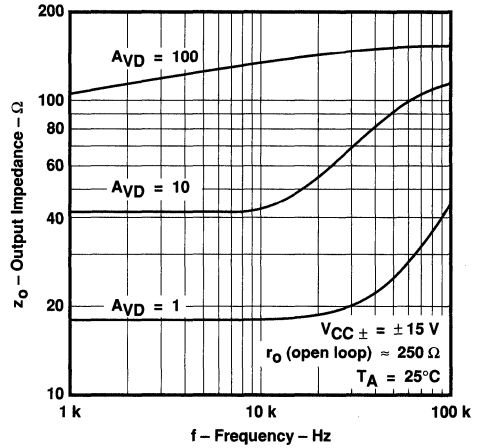


FIGURE 23

COMMON-MODE REJECTION RATIO
VS
FREQUENCY

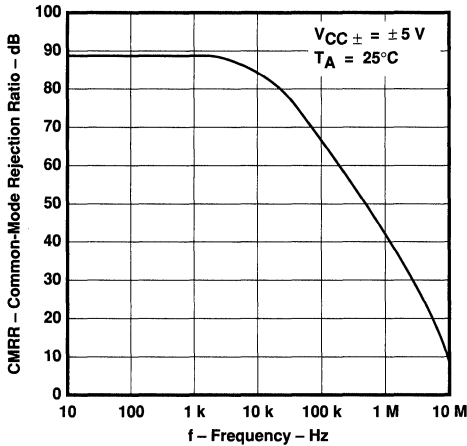


FIGURE 24

COMMON-MODE REJECTION RATIO
VS
FREQUENCY

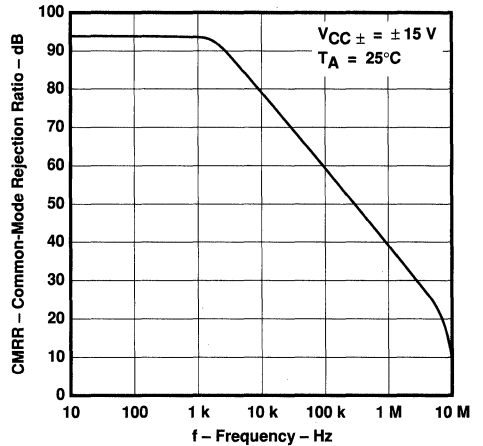


FIGURE 25

TL032, TL032A
ENHANCED JFET LOW-POWER LOW-OFFSET
DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

COMMON-MODE REJECTION RATIO
 VS
 FREE-AIR TEMPERATURE

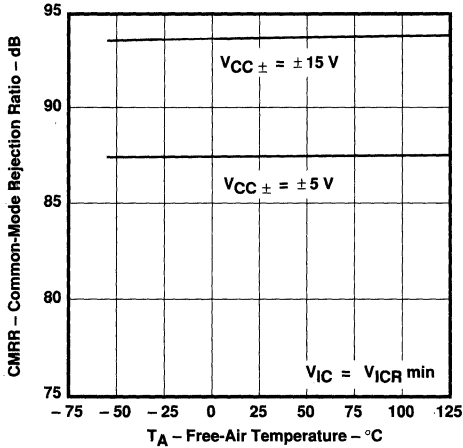


FIGURE 26

SUPPLY-VOLTAGE REJECTION RATIO
 VS
 FREE-AIR TEMPERATURE

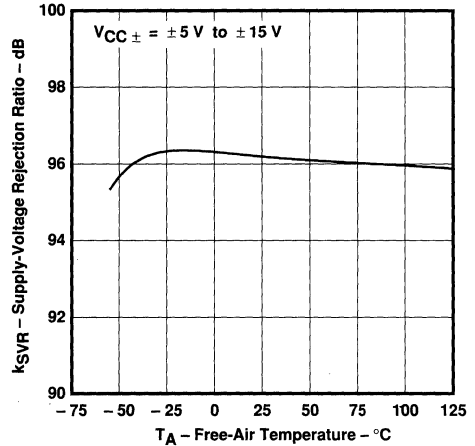


FIGURE 27

SHORT-CIRCUIT OUTPUT CURRENT
 VS
 SUPPLY VOLTAGE

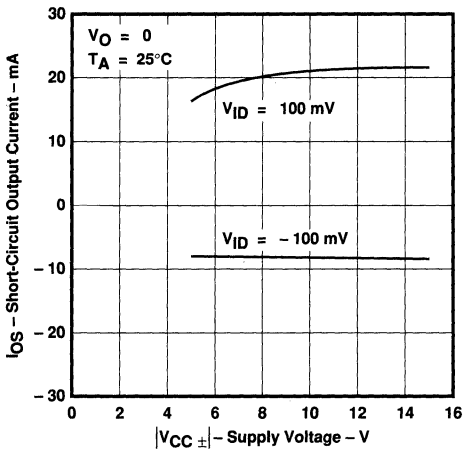


FIGURE 28

SHORT-CIRCUIT OUTPUT CURRENT
 VS
 TIME

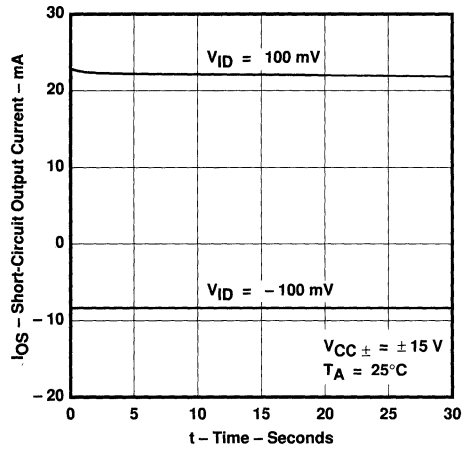


FIGURE 29

2
 Operational Amplifiers

TL032, TL032A
ENHANCED JFET LOW-POWER LOW-OFFSET
DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

SHORT-CIRCUIT OUTPUT CURRENT
VS
FREE-AIR TEMPERATURE

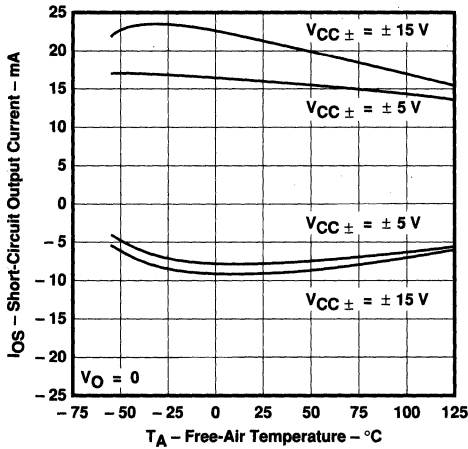


FIGURE 30

EQUIVALENT INPUT NOISE VOLTAGE
VS
FREQUENCY

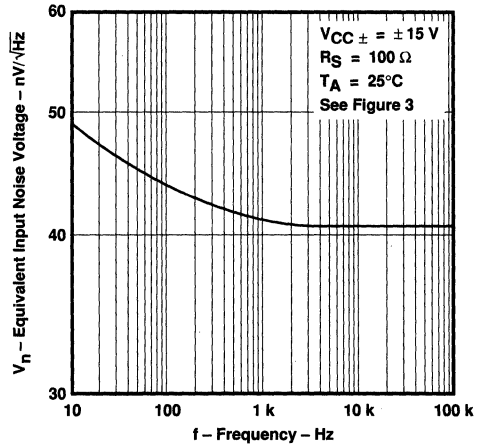


FIGURE 31

SUPPLY CURRENT
VS
SUPPLY VOLTAGE

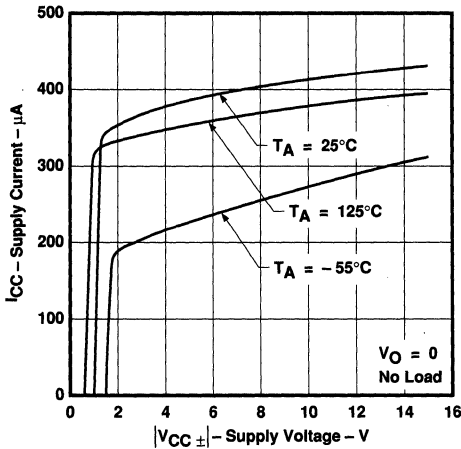


FIGURE 32

SUPPLY CURRENT
VS
FREE-AIR TEMPERATURE

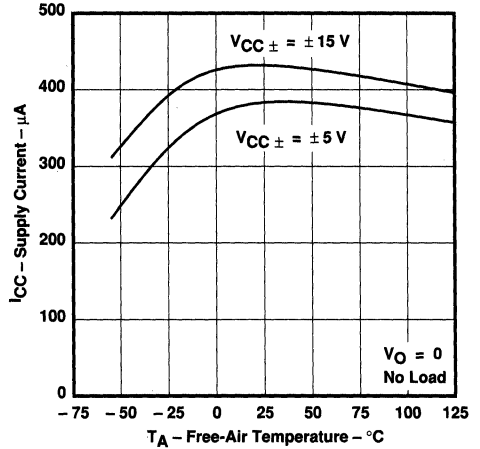
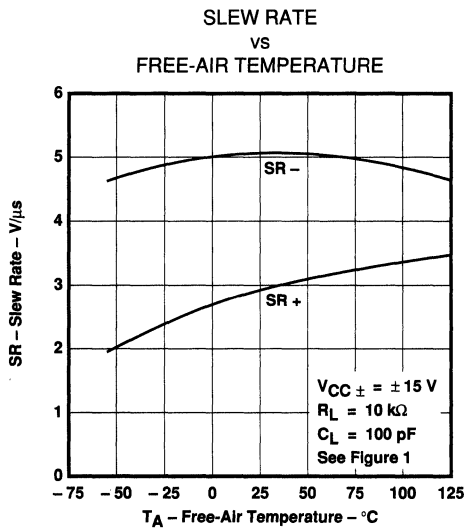
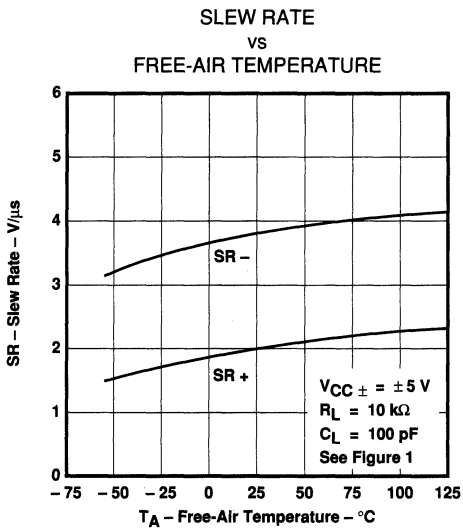
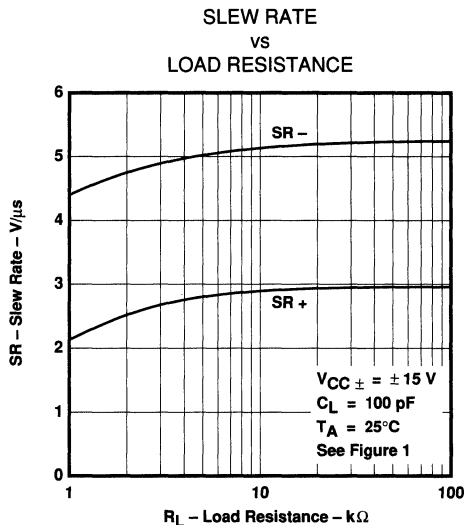
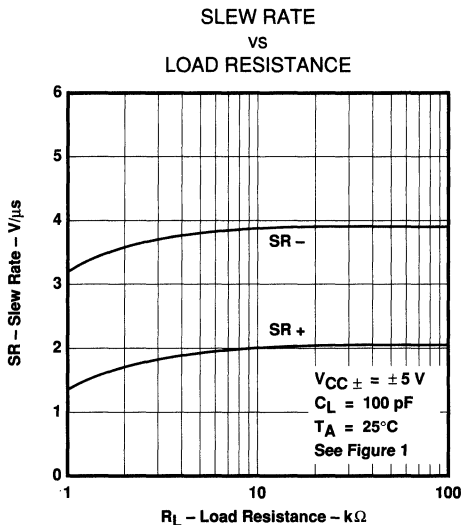


FIGURE 33

2
Operational Amplifiers

TL032, TL032A ENHANCED JFET LOW-POWER LOW-OFFSET DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS



TL032, TL032A
ENHANCED JFET LOW-POWER LOW-OFFSET
DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

OVERSHOOT FACTOR
VS
LOAD CAPACITANCE

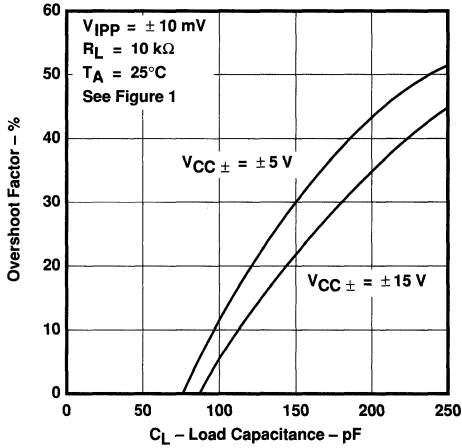


FIGURE 38

TOTAL HARMONIC DISTORTION
VS
FREQUENCY

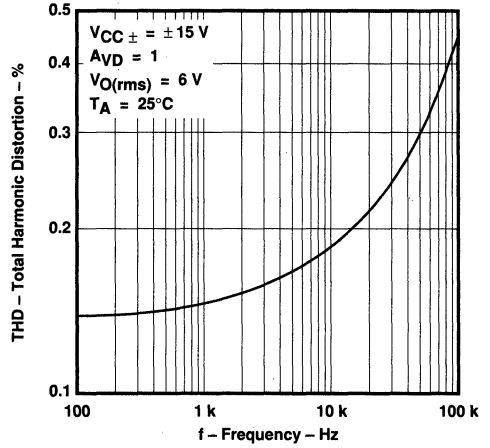


FIGURE 39

UNITY-GAIN BANDWIDTH
VS
SUPPLY VOLTAGE

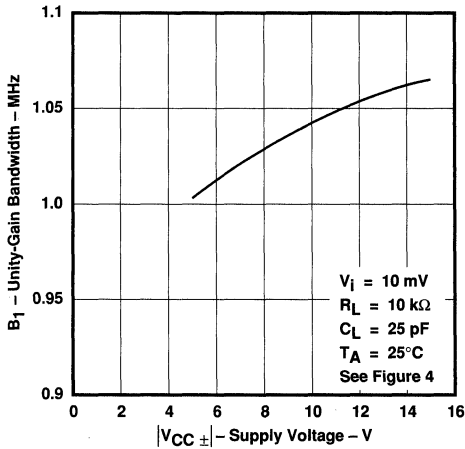


FIGURE 40

UNITY-GAIN BANDWIDTH
VS
FREE-AIR TEMPERATURE

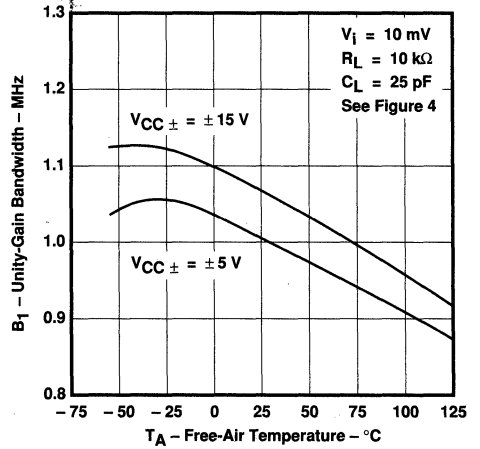


FIGURE 41

2
Operational Amplifiers

TL032, TL032A ENHANCED JFET LOW-POWER LOW-OFFSET DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

PHASE MARGIN
VS
SUPPLY VOLTAGE

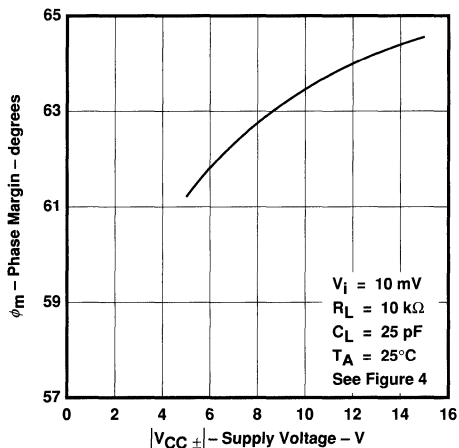


FIGURE 42

PHASE MARGIN
VS
LOAD CAPACITANCE

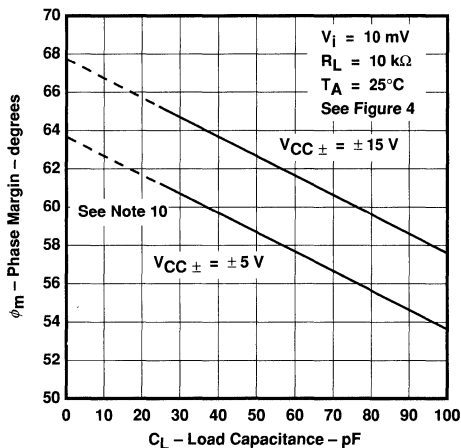


FIGURE 43

PHASE MARGIN
VS
FREE-AIR TEMPERATURE

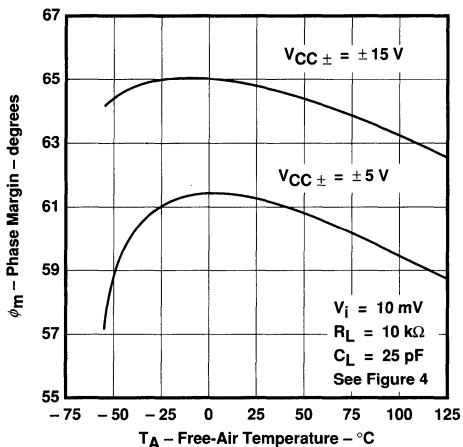


FIGURE 44

VOLTAGE-FOLLOWER
SMALL-SIGNAL
PULSE RESPONSE

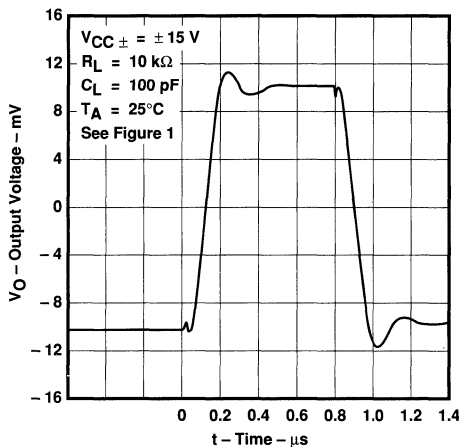


FIGURE 45

2
Operational Amplifiers

NOTE 10: Values of phase margin below a load capacitance of 25 pF were estimated.

TL032, TL032A ENHANCED JFET LOW-POWER LOW-OFFSET DUAL OPERATIONAL AMPLIFIERS

2

Operational Amplifiers

TYPICAL CHARACTERISTICS

VOLTAGE-FOLLOWER
LARGE-SIGNAL
PULSE RESPONSE

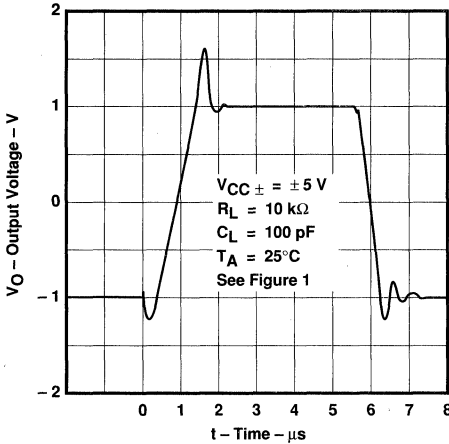


FIGURE 46

VOLTAGE-FOLLOWER
LARGE-SIGNAL
PULSE RESPONSE

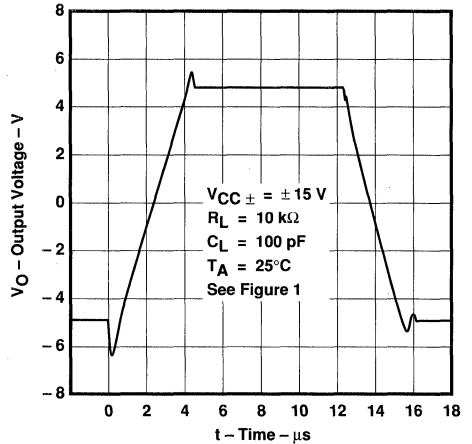


FIGURE 47

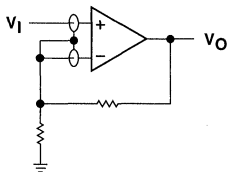
TYPICAL APPLICATION DATA

input characteristics

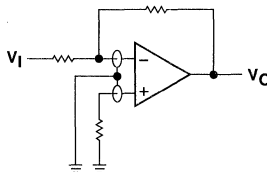
The TL032 and TL032A are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction.

Because of the extremely high input impedance and resulting low bias current requirements, the TL032 and TL032A are well-suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause degradation in system performance. It is a good practice to include guard rings around inputs (see Figure 48). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.

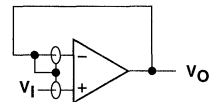
The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.



(a) NONINVERTING AMPLIFIER



(b) INVERTING AMPLIFIER



(c) UNITY-GAIN AMPLIFIER

FIGURE 48. USE OF GUARD RINGS

TYPICAL APPLICATION DATA

output characteristics

All operating characteristics (except bandwidth and phase margin) are specified with 100 pF load capacitance. The TL032 and TL032A will drive higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem. Capacitive loads of 1000 pF and larger may be driven if enough resistance is added in series with the output (see Figure 49).

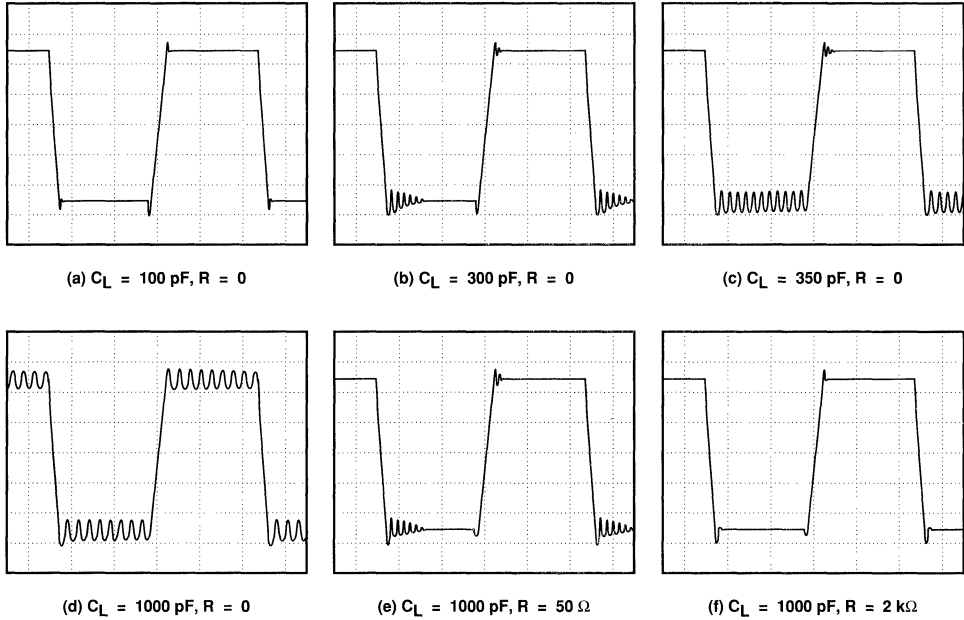


FIGURE 49. EFFECT OF CAPACITIVE LOADS

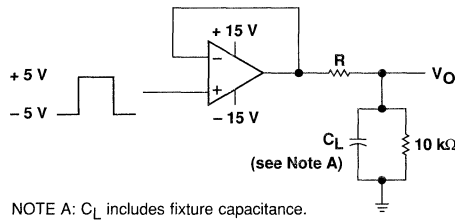


FIGURE 50. TEST CIRCUIT FOR OUTPUT CHARACTERISTICS

**TL032, TL032A
ENHANCED JFET LOW-POWER LOW-OFFSET
DUAL OPERATIONAL AMPLIFIERS**

TYPICAL APPLICATION DATA

high-Q notch filter

In general, Texas Instruments enhanced JFET operational amplifiers serve as excellent filters. This circuit provides a narrow notch at a specific frequency. Notch filters are designed to eliminate frequencies that are interfering with the operation of an application. For this filter, the center frequency can be calculated as:

$$f_0 = \frac{1}{2\pi R_1 C_1}$$

With the resistors and capacitors shown below, the center frequency is 1 kHz. Note that $C_1 = C_3 = C_2 \div 2$ and also that $R_1 = R_3 = 2 \times R_2$. The center frequency can be modified by varying these values. When adjusting the center frequency, be sure that the operational amplifier still has sufficient gain at the frequency of interest.

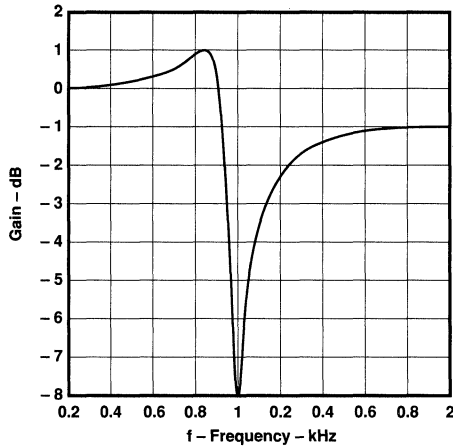
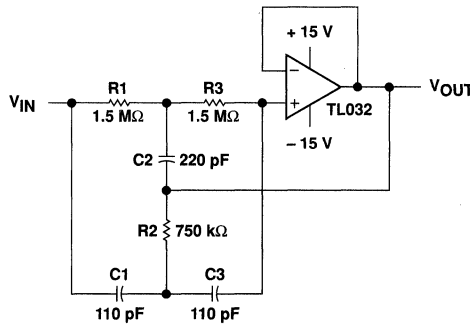


FIGURE 51. HIGH-Q NOTCH FILTER

2

Operational Amplifiers

TYPICAL APPLICATION DATA

2-wire 4- to 20-mA current loop

Often information from an analog sensor must be sent over a distance to the receiving circuitry. For many applications, the most feasible method involves converting voltage information to a current before transmission. The following circuit benefits from the high input impedance of the TL032A since many inexpensive sensors do not have low output impedance.

Assuming that the voltage at the TL032A's non-inverting input is zero, the following equation determines the output current:

$$I_O = V_{IN} \left(\frac{R_3}{R_1 \times R_S} \right) + 5 V \left(\frac{R_3}{R_2 \times R_S} \right) = 0.16 \times V_{IN} + 4 \text{ mA}$$

The current presently provides 4 to 20 mA output for an input voltage of 0 to 100 mV. By modifying R1, R2, and R3, the input voltage range or the output current range can be adjusted.

Including the offset voltage of the operational amplifier in the above equation clearly illustrates why the low offset TL032A was chosen:

$$I_O = V_{IN} \left(\frac{R_3}{R_1 \times R_S} \right) + 5 V \left(\frac{R_3}{R_2 \times R_S} \right) - V_{IO} \left(\frac{R_3}{R_1 \times R_S} + \frac{R_3}{R_2 \times R_S} + \frac{R_1}{R_S} \right) = 0.16 \times V_{IN} + 4 \text{ mA} - 0.17 \times V_{IO}$$

For example, an offset voltage of 1 mV decreases the output current by 0.17 mA.

Thanks to the low-power consumption of the TL032A, this circuit has at least 2 mA available to drive the actual sensor from the 5-V reference node.

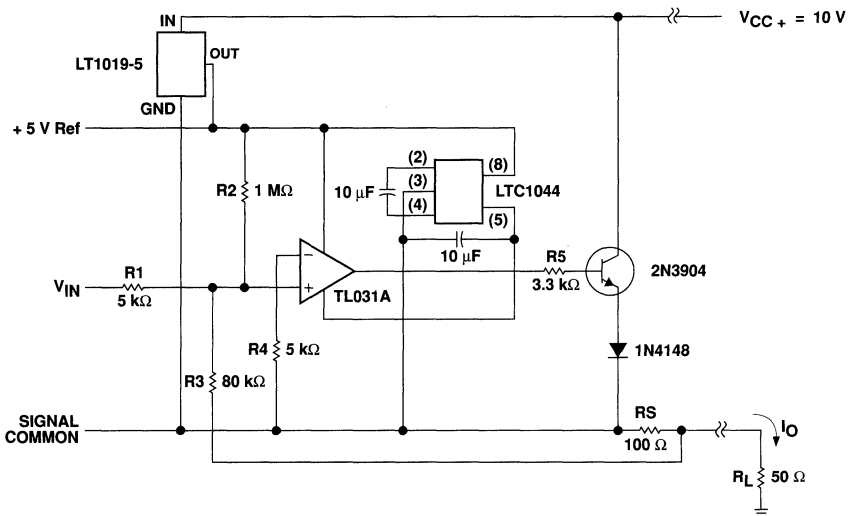


FIGURE 52. 2-WIRE 4- TO 20-mA CURRENT LOOP

TL032, TL032A ENHANCED JFET LOW-POWER LOW-OFFSET DUAL OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

low-level light detector preamplifier

Applications that need to detect small currents require high input impedance operational amplifiers; otherwise, the bias currents of the operational amplifier camouflage the current being monitored. Phototransistors provide a current that is proportional to the light reaching the transistor. The TL032 allows even the small currents resulting from low-level light to be detected.

In this circuit, if there is no light, the phototransistor is off and the output is high. As light is detected, the operational amplifier output begins pulling low. Adjusting R41 both compensates for offset voltage of the amplifier and adjusts the point of light detection by the amplifier.

2

Operational Amplifiers

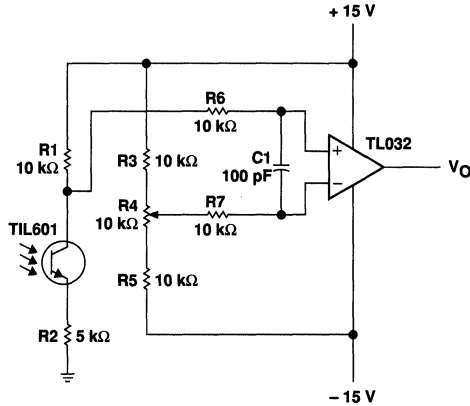


FIGURE 53. LOW-LEVEL LIGHT DETECTOR PREAMPLIFIER

TL034, TL034A

ENHANCED JFET LOW-POWER LOW-OFFSET

QUAD OPERATIONAL AMPLIFIERS

D3153, JULY 1988 – REVISED JANUARY 1989

- Maximum Offset Voltage ... 1.5 mV
- High Slew Rate ... 2.9 V/ μ s Typ
- Low Input Bias Current ... 2 pA Typ
- Very Low Power Consumption ... 26 mW Typ
- Output Short-Circuit Protection
- Monolithic Construction

description

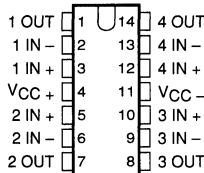
The TL034 and TL034A quadruple operational amplifiers incorporate well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. These devices offer the significant advantages of Texas Instruments new enhanced JFET process. This process affords not only low initial offset voltage due to the on-chip zener trim capability but also stable offset voltage over time and temperature. In comparison, traditional JFET processes are plagued by significant offset voltage drift.

This new enhanced process still maintains the traditional JFET advantages of fast slew rates and low input bias and offset currents. These advantages, coupled with low power consumption, make the TL034 well-suited for new state-of-the-art designs as well as existing design upgrades. The TL034 has been designed to be functionally compatible and pin compatible with the TL064.

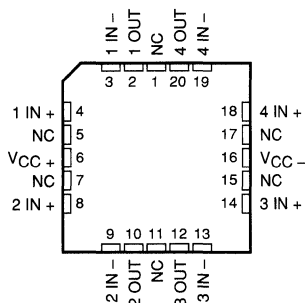
Two offset voltage grades are available: TL034 (4 mV max) and TL034A (1.5 mV max).

A variety of available packaging options includes small-outline and chip carrier versions for high density system applications.

D, J, or N PACKAGE
(TOP VIEW)



FK PACKAGE
(TOP VIEW)



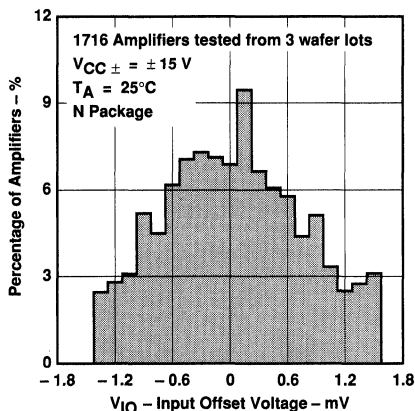
NC – No internal connection

AVAILABLE OPTIONS

T _A	V _{IO} max AT 25°C	PACKAGE			
		SMALL- OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	1.5 mV	TL034ACD	—	TL034ACJ	TL034ACN
	4 mV	TL034CD	—	TL034CJ	TL034CN
-40°C to 85°C	1.5 mV	TL034AID	—	TL034AIJ	TL034AIN
	4 mV	TL034ID	—	TL034IJ	TL034IN
-55°C to 125°C	1.5 mV	TL034AMD	TL034AMFK	TL034AMJ	TL034AMN
	4 mV	TL034MD	TL034MFK	TL034MJ	TL034MN

D packages are available taped-and-reeled. Add "R" suffix to device type (e.g., TL034CDR).

DISTRIBUTION OF TL034A
INPUT OFFSET VOLTAGE



PRODUCTION DATA documents contain information current as of publication date. Products conform to these specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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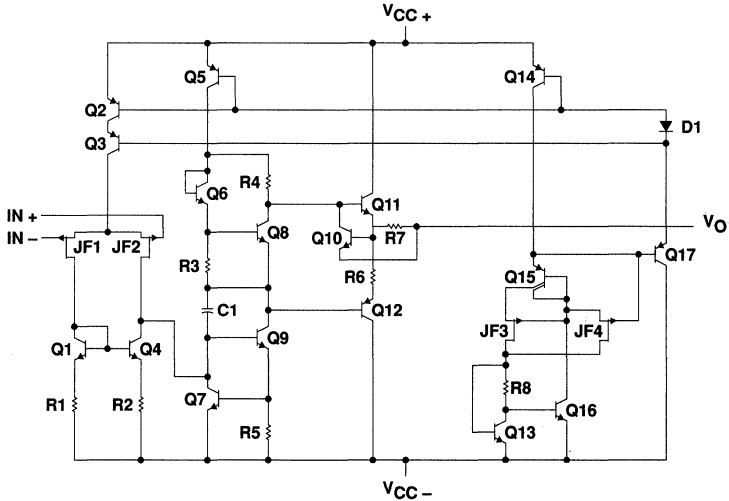
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TL034, TL034A ENHANCED JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

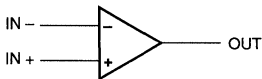
description (continued)

The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I-suffix devices are characterized for operation from -40°C to 85°C . The C-suffix devices are characterized for operation from 0°C to 70°C .

equivalent schematic (each amplifier)



symbol (each amplifier)



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Operational Amplifiers

TL034M, TL034AM

ENHANCED JFET LOW-POWER LOW-OFFSET

QUAD OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A †	V _{CC} ± ± 5 V			V _{CC} ± ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL034M	25°C	0.91 6		0.78 4		mV	
			Full range	11		9			
		TL034AM	25°C	0.7 3.5		0.58 1.5		mV	
			Full range	8.5		6.5			
α _{VIO} Temperature coefficient of input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL034M	25°C to 125°C	10.6		10.9		μV/°C	
		TL034AM	25°C to 125°C	10.6		10.9		μV/°C	
Input offset voltage long-term drift (see Note 5)		25°C	0.04		0.04		μV/mo		
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	1 100		1 100		pA		
		125°C	0.2 10		0.2 10		nA		
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	2 200		2 200		pA		
		125°C	7 20		8 20		nA		
V _{ICR} Common-mode input voltage range		25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4	V		
		Full range	-1.5 to 4		-11.5 to 14		V		
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3 4.3		13 14		V		
		-55°C	3 4.1		13 14				
		125°C	3 4.4		13 14				
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-3 -4.2		-12.5 -13.9		V		
		-55°C	-3 -4		-12.5 -13.8				
		125°C	-3 -4.3		-12.5 -14				
A _{VD} Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	4 12		5 14.3		V/mV		
		-55°C	3 7.1		4 10.4				
		125°C	3 12.9		4 15				
r _i Input resistance		25°C	10 ¹²			Ω			
C _i Input capacitance		25°C	5			pF			
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	70 87		75 94		dB		
		-55°C	70 87		70 94				
		125°C	70 87		70 94				
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75 96		75 96		dB		
		-55°C	75 95		75 95				
		125°C	75 96		75 96				
P _D Total power dissipation (four amplifiers)	No load, V _O = 0	25°C	7.7 10		26 34		mW		
		-55°C	4.6 10		18.7 34				
		125°C	7.1 10		23.6 34				
I _{CC} Supply current (four amplifiers)	No load, V _O = 0	25°C	0.77 1		0.87 1.12		mA		
		-55°C	0.46 1		0.62 1.12				
		125°C	0.71 1		0.79 1.12				
V _{O1} /V _{O2} Crosstalk attenuation	A _{VD} = 100	25°C	120		120		dB		

† Full range is -55°C to 125°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. At V_{CC} ± = ± 5 V, V_O = ± 2.3 V; at V_{CC} ± = ± 15 V, V_O = ± 10 V.

2

Operational Amplifiers

TL034M, TL034AM ENHANCED JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER		TEST CONDITIONS		T _A	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
					MIN	TYP	MAX	MIN	TYP	MAX	
SR +	Positive slew rate at unity gain	R _L = 10 kΩ, C _L = 100 pF, See Figure 1, See Note 7		25°C		2		2	2.9	V/μs	
				-55°C		1.4		1.2	1.9		
				125°C		2.4		1.2	3.5		
SR -	Negative slew rate at unity gain			25°C		3.9		3	5.1	V/μs	
				-55°C		3.2		2.5	4.6		
				125°C		4.1		2.5	4.7		
t _r	Rise time	V _{IPP} = ± 10 mV, R _L = 10 kΩ, C _L = 100 pF, See Figures 1 and 2		25°C		138		132	ns		
t _f	Fall time			-55°C		142		123			
				125°C		166		158			
				25°C		138		132			
Overshoot factor						-55°C		142		123	ns
						125°C		166		158	
		25°C				11%		5%			
V _n	Equivalent input noise voltage	R _S = 100 Ω, See Figure 3	f = 10 Hz f = 1 kHz f = 10 Hz f = 1 kHz	25°C		83		83	nV/√Hz		
						43		43			
				25°C		83		83			
						43		43			
I _n	Equivalent input noise current	f = 1 kHz		25°C		0.003		0.003	pA/√Hz		
B ₁	Unity-gain bandwidth	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C		1		1.1	MHz		
				-55°C		1		1.1			
				125°C		0.9		0.9			
φ _m	Phase margin at unity gain	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C		61°		65°			
				-55°C		57°		64°			
				125°C		59°		62°			

NOTE 7: For V_{CC} ± = ± 5 V, V_{IPP} = ± 1 V; for V_{CC} ± = ± 15 V, V_{IPP} = ± 5 V.

TL034I, TL034AI

ENHANCED JFET LOW-POWER LOW-OFFSET

QUAD OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL034I	25°C	0.91		6		0.79		mV
			Full range			9.3		7.3		
		TL034AI	25°C	0.7		3.5		0.58		mV
			Full range			6.8		4.8		
αV _{IO} Temperature coefficient of input offset voltage (see Note 8)	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL034I	25°C to 85°C		11.5		11.6		μV/°C	
		TL034AI	25°C to 85°C		11.5		11.6			
Input offset voltage long-term drift (see Note 5)		25°C	0.04		0.04		0.04		μV/mo	
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	1		100		1		pA	
		85°C	0.02		0.45		0.02			
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	2		200		2		pA	
		85°C	0.2		0.9		0.3			
V _{ICR} Common-mode input voltage range		25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14		-13.4 to 15.4		V	
		Full range	-1.5 to 4		-11.5 to 14					
V _{OM} + Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3		4.3		13		V	
		-40°C	3		4.1		13			
		85°C	3		4.4		13			
V _{OM} - Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-3		-4.2		-12.5		V	
		-40°C	-3		-4.1		-12.5			
		85°C	-3		-4.2		-12.5			
A _{VD} Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	4		12		5		V/mV	
		-40°C	3		8.4		4			
		85°C	4		13.5		5			
r _i Input resistance		25°C	10 ¹²		10 ¹²				Ω	
C _i Input capacitance		25°C	5		4				pF	
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	70		87		75		dB	
		-40°C	70		87		75			
		85°C	70		87		75			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75		96		75		dB	
		-40°C	75		96		75			
		85°C	75		96		75			
P _D Total power dissipation (four amplifiers)	No load, V _O = 0	25°C	7.7		10		26		mW	
		-40°C	5.8		10		21.7			
		85°C	7.4		10		24.8			
I _{CC} Supply current (four amplifiers)	No load, V _O = 0	25°C	0.77		1		0.87		mA	
		-40°C	0.58		1		0.72			
		85°C	0.74		1		0.83			
V _{O1} /V _{O2} Crosstalk attenuation	A _{VD} = 100	25°C	120		120		120		dB	

[†] Full range is -40°C to 85°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. At V_{CC} ± = ± 5 V, V_O = ± 2.3 V; at V_{CC} ± = ± 15 V, V_O = ± 10 V.

8. This parameter is tested on a sample basis for the TL034A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

2

Operational Amplifiers

TL034I, TL034AI

ENHANCED JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER		TEST CONDITIONS		T _A	V _{CC±} = ± 5 V			V _{CC±} = ± 15 V			UNIT
					MIN	TYP	MAX	MIN	TYP	MAX	
SR +	Positive slew rate at unity gain	R _L = 10 kΩ, C _L = 100 pF, See Figure 1, See Note 7		25°C	2			2			V/μs
				-40°C	1.6			1.5			
				85°C	2.3			2			
SR -	Negative slew rate at unity gain			25°C	3.9			3.5			V/μs
				-40°C	3.3			3.2			
				85°C	4.1			3.2			
t _r	Rise time	V _{Ipp} = ± 10 mV, R _L = 10 kΩ, C _L = 100 pF, See Figures 1 and 2		25°C	138			132			ns
				-40°C	132			123			
				85°C	154			146			
t _f	Fall time			25°C	138			132			ns
				-40°C	132			123			
				85°C	154			146			
Overshoot factor				25°C	11%			5%			
				-40°C	12%			5%			
				85°C	13%			7%			
V _n	Equivalent input noise voltage (see Note 9)	TL034I	R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C			83			nV/√Hz
				f = 1 kHz	25°C			43			
		TL034AI	f = 10 Hz	25°C			83				
			f = 1 kHz	25°C			43				
I _n	Equivalent input noise current	f = 1 kHz		25°C	0.003			0.003			pA/√Hz
B1	Unity-gain bandwidth	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	1			1.1			MHz
				-40°C	1			1.1			
				85°C	0.9			1			
φ _m	Phase margin at unity gain	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	61°			65°			
				-40°C	60°			65°			
				85°C	60°			64°			

NOTES: 7. For V_{CC±} = ± 5 V, V_{Ipp} = ± 1 V; for V_{CC±} = ± 15 V, V_{Ipp} = ± 5 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

2

Operational Amplifiers

TL034C, TL034AC

ENHANCED JFET LOW-POWER LOW-OFFSET

QUAD OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL034C	25°C	0.91	6	0.79	4	mV	
			Full range		8.2		6.2		
		TL034AC	25°C		0.7	3.5	0.58		1.5
			Full range			5.7			3.7
α _{VIO} Temperature coefficient of input offset voltage (see Note 8)	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL034C	25°C to 70°C	11.6		12		μV/°C	
		TL034AC	25°C to 70°C	11.6		12	25		
Input offset voltage long-term drift (see Note 5)			25°C	0.04		0.04		μV/mo	
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C		1	100	1	100	pA	
			70°C		9	200	12		200
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C		2	200	2	200	pA	
			70°C		50	400	80		400
V _{ICR} Common-mode input voltage range		25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4	V		
			Full range	-1.5 to 4		-11.5 to 14			
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.3	13	14	V		
		0°C	3	4.2	13	14			
		70°C	3	4.3	13	14			
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-3	-4.2	-12.5	-13.9	V		
		0°C	-3	-4.1	-12.5	-13.9			
		70°C	-3	-4.2	-12.5	-14			
A _{VD} Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	4	12	5	14.3	V/mV		
		0°C	3	11.1	4	13.5			
		70°C	4	13.3	5	15.2			
r _i Input resistance		25°C	10 ¹²			10 ¹²	Ω		
C _i Input capacitance		25°C	5			4	pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	70	87	75	94	dB		
		0°C	70	87	75	94			
		70°C	70	87	75	94			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75	96	75	96	dB		
		0°C	75	96	75	96			
		70°C	75	96	75	96			
P _D Total power dissipation (four amplifiers)	No load, V _O = 0	25°C	7.7	10	26	34	mW		
		0°C	7.4	10	25.3	34			
		70°C	7.6	10	25.2	34			
I _{CC} Supply current (four amplifiers)	No load, V _O = 0	25°C	0.77	1	0.87	1.12	mA		
		0°C	0.74	1	0.85	1.12			
		70°C	0.76	1	0.84	1.12			
V _{O1} /V _{O2} Crosstalk attenuation	A _{VD} = 100	25°C	120		120		dB		

[†] Full range is 0°C to 70°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. At V_{CC} ± = ± 5 V, V_O = ± 2.3 V; at V_{CC} ± = ± 15 V, V_O = ± 10 V.

8. This parameter is tested on a sample basis for the TL034A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

2

Operational Amplifiers

TL034C, TL034AC ENHANCED JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

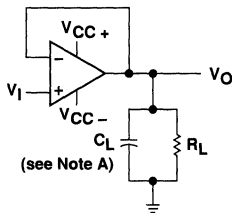
operating characteristics

PARAMETER		TEST CONDITIONS		T _A	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT	
					MIN	TYP	MAX	MIN	TYP	MAX		
SR +	Positive slew rate at unity gain	R _L = 10 kΩ, C _L = 100 pF, See Figure 1, See Note 7		25°C	2			2			V/μs	
				0°C	1.8			1.5				
				70°C	2.2			2				
SR -	Negative slew rate at unity gain	See Figure 1, See Note 7		25°C	3.9			3.5			V/μs	
				0°C	3.7			3.2				
				70°C	4			3.2				
t _r	Rise time	V _I PP = ± 10 mV, R _L = 10 kΩ, C _L = 100 pF, See Figures 1 and 2		25°C	138			132			ns	
				0°C	134			127				
				70°C	150			142				
t _f	Fall time	See Figures 1 and 2		25°C	138			132			ns	
				0°C	134			127				
				70°C	150			142				
Overshoot factor				25°C	11%			5%				
				0°C	10%			4%				
				70°C	12%			6%				
V _n	Equivalent input noise voltage (see Note 9)	TL034C	R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C	83			83			nV/√Hz
				f = 1 kHz	43			43				
		TL034AC		f = 10 Hz	25°C	83			83			
				f = 1 kHz	43			43			60	
I _n	Equivalent input noise current	f = 1 kHz		25°C	0.003			0.003			pA/√Hz	
B1	Unity-gain bandwidth	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	1			1.1			MHz	
				0°C	1			1.1				
				70°C	1			1				
φ _m	Phase margin at unity gain	V _i = 10 mV, R _L = 10 kΩ, C _L = 25 pF, See Figure 4		25°C	61°			65°				
				0°C	61°			65°				
				70°C	60°			64°				

NOTES: 7. For V_{CC} ± = ± 5 V, V_IPP = ± 1 V; for V_{CC} ± = ± 15 V, V_IPP = ± 5 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

PARAMETER MEASUREMENT INFORMATION



NOTE A: C_L includes fixture capacitance.

FIGURE 1. SLEW RATE, RISE/FALL TIME, AND OVERSHOOT TEST CIRCUIT

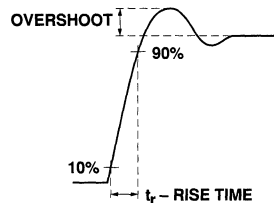


FIGURE 2. RISE TIME AND OVERSHOOT WAVEFORM

2
Operational Amplifiers

TL034, TL034A ENHANCED JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

PARAMETER MEASUREMENT INFORMATION

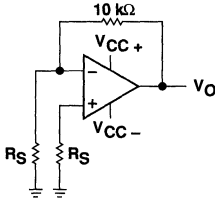
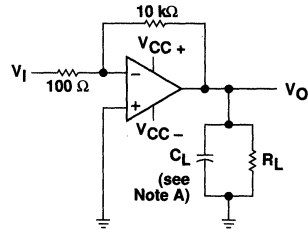


FIGURE 3. NOISE VOLTAGE TEST CIRCUIT



NOTE A: C_L includes fixture capacitance.

FIGURE 4. UNITY-GAIN BANDWIDTH AND
PHASE MARGIN TEST CIRCUIT

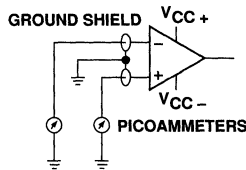


FIGURE 5. INPUT BIAS AND OFFSET
CURRENT TEST CIRCUIT

2

Operational Amplifiers

typical values

Typical values as presented in this data sheet represent the median (50% point) of device parametric performance.

input bias and offset current

At the picoamp bias current level typical of the TL034 and TL034A, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To accurately measure these small currents, Texas Instruments uses a two step process. The socket leakage is measured using picoammeters with bias voltages applied but with no device in the socket. The device is then inserted into the socket and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

noise

Because of the increasing emphasis on low noise levels in many of today's applications, the input noise voltage density is sample-tested at $f = 1$ kHz. Texas Instruments also has additional noise testing capability to meet specific application requirements. Please contact the factory for details.

TL034, TL034A
ENHANCED JFET LOW-POWER LOW-OFFSET
QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

table of graphs

		FIGURE	
V_{IO}	Input offset voltage	Distribution	6
α_{VIO}	Temperature coefficient of input offset voltage	Distribution	7
I_{IO}	Input offset current	vs Temperature	8
I_{IB}	Input bias current	vs V_{IC}	9
		vs Temperature	8
V_I	Input voltage range	vs V_{CC}	10
		vs Temperature	11
V_{ID}	Differential input voltage	vs Output voltage	12, 13
		vs V_{CC}	14
V_{OM}	Maximum peak output voltage swing	vs Output current	16, 17
		vs Frequency	15
		vs Temperature	18, 19
		vs R_L	20
A_{VD}	Differential voltage amplification	vs Frequency	21
		vs Temperature	22
		vs Frequency	23
z_o	Output impedance	vs Frequency	23
CMRR	Common-mode rejection ratio	vs Frequency	24, 25
		vs Temperature	26
k_{SVR}	Supply-voltage rejection ratio	vs Temperature	27
I_{OS}	Short-circuit output current	vs V_{CC}	28
		vs Time	29
		vs Temperature	30
I_{CC}	Supply current	vs V_{CC}	32
		vs Temperature	33
SR	Slew rate	vs R_L	34, 35
		vs Temperature	36, 37
	Overshoot factor	vs C_L	38
V_n	Equivalent input noise voltage	vs Frequency	31
THD	Total harmonic distortion	vs Frequency	39
B_1	Unity-gain bandwidth	vs V_{CC}	40
		vs Temperature	41
ϕ_m	Phase margin	vs V_{CC}	42
		vs C_L	43
		vs Temperature	44
	Phase shift	vs Frequency	21
	Pulse response	Small-signal	45
		Large-signal	46, 47

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Operational Amplifiers

TL034, TL034A
ENHANCED JFET LOW-POWER LOW-OFFSET
QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

DISTRIBUTION OF TL034
 INPUT OFFSET VOLTAGE

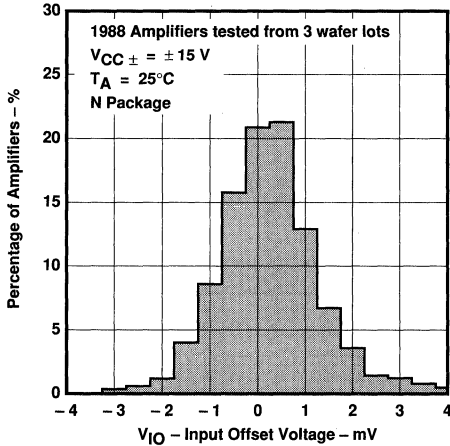


FIGURE 6

DISTRIBUTION OF TL034
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT

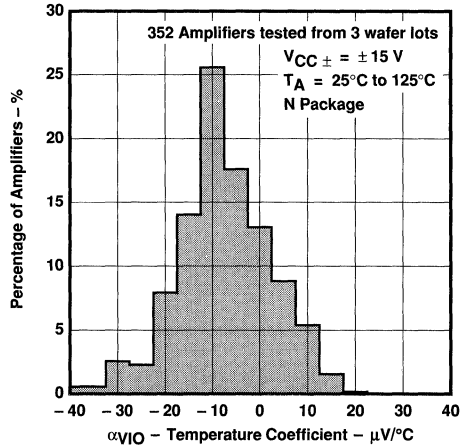


FIGURE 7

INPUT BIAS CURRENT AND
 INPUT OFFSET CURRENT
 vs
 FREE-AIR TEMPERATURE

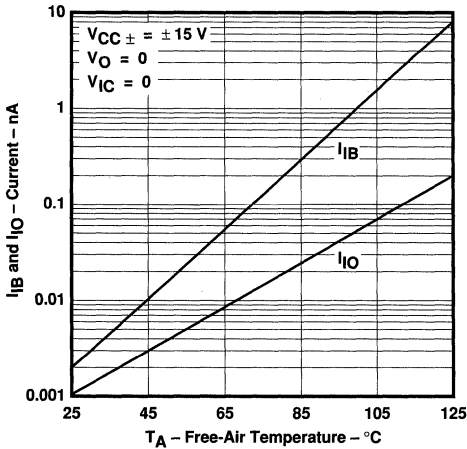


FIGURE 8

INPUT BIAS CURRENT
 vs
 COMMON-MODE INPUT VOLTAGE

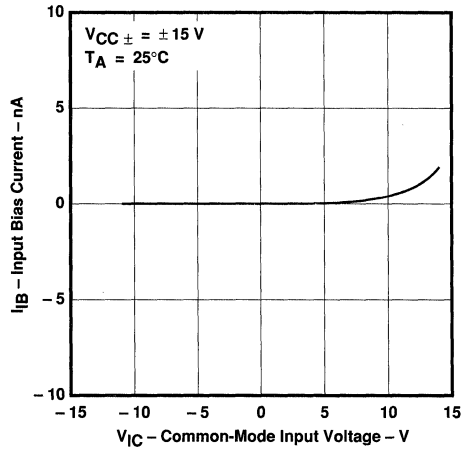


FIGURE 9

TL034, TL034A ENHANCED JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

COMMON-MODE
INPUT VOLTAGE RANGE LIMITS
VS
SUPPLY VOLTAGE

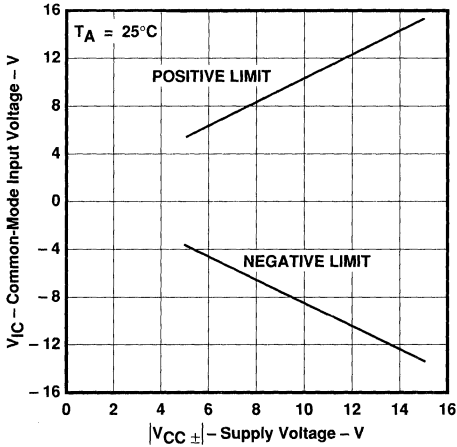


FIGURE 10

COMMON-MODE
INPUT VOLTAGE RANGE LIMITS
VS
FREE-AIR TEMPERATURE

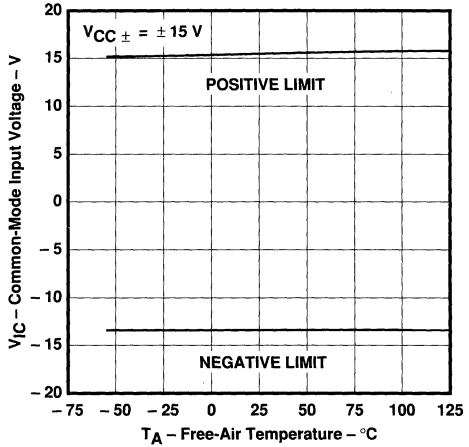


FIGURE 11

DIFFERENTIAL INPUT VOLTAGE
VS
OUTPUT VOLTAGE

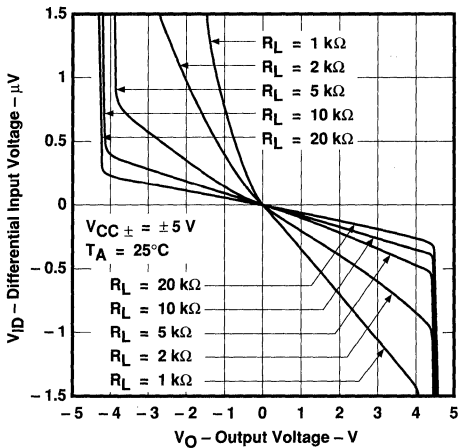


FIGURE 12

DIFFERENTIAL INPUT VOLTAGE
VS
OUTPUT VOLTAGE

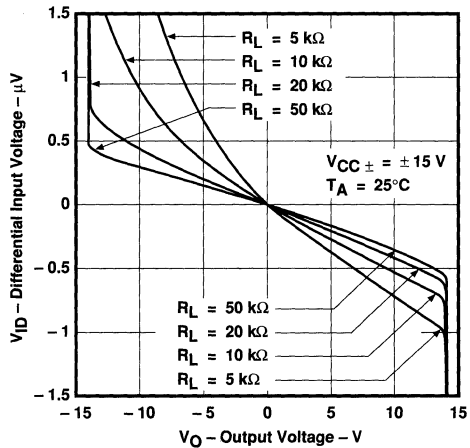


FIGURE 13

2
Operational Amplifiers

TL034, TL034A
ENHANCED JFET LOW-POWER LOW-OFFSET
QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

MAXIMUM PEAK OUTPUT VOLTAGE
 vs
 SUPPLY VOLTAGE

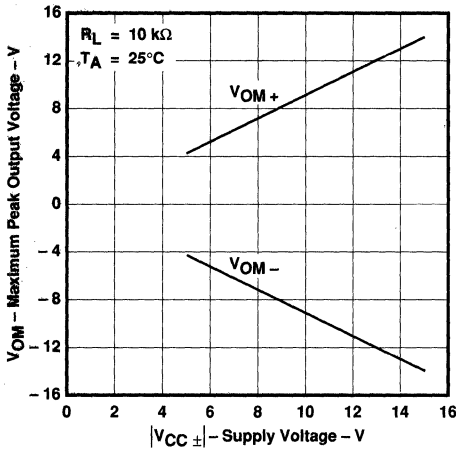


FIGURE 14

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 vs
 FREQUENCY

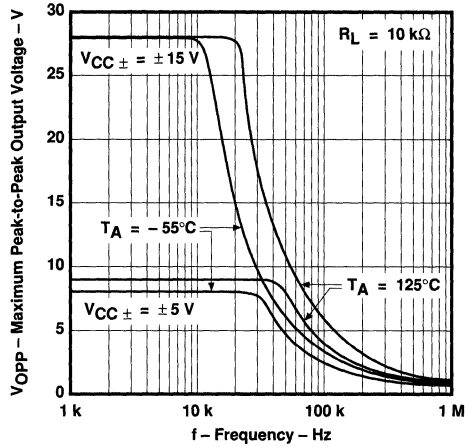


FIGURE 15

MAXIMUM PEAK OUTPUT VOLTAGE
 vs
 OUTPUT CURRENT

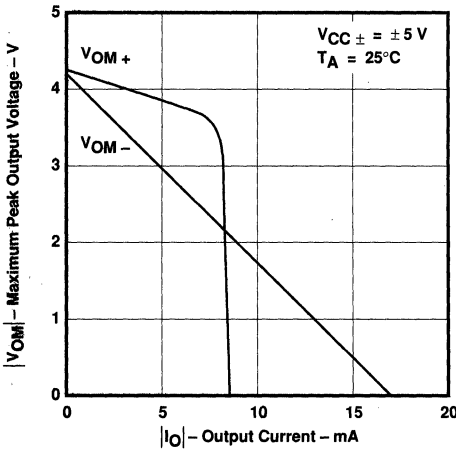


FIGURE 16

MAXIMUM PEAK OUTPUT VOLTAGE
 vs
 OUTPUT CURRENT

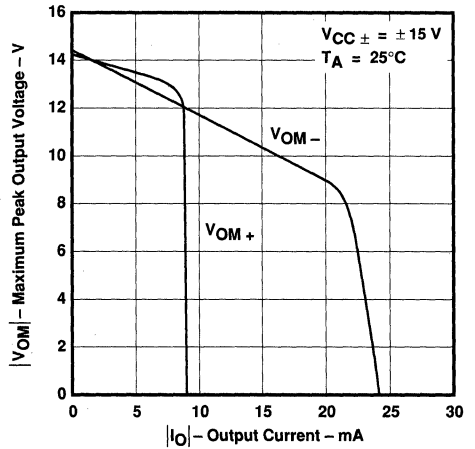


FIGURE 17

TL034, TL034A ENHANCED JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

MAXIMUM PEAK OUTPUT VOLTAGE
VS
FREE-AIR TEMPERATURE

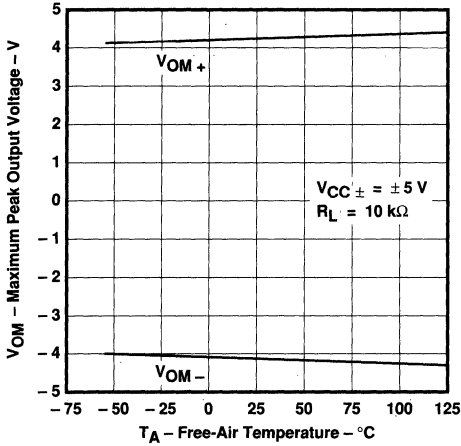


FIGURE 18

MAXIMUM PEAK OUTPUT VOLTAGE
VS
FREE-AIR TEMPERATURE

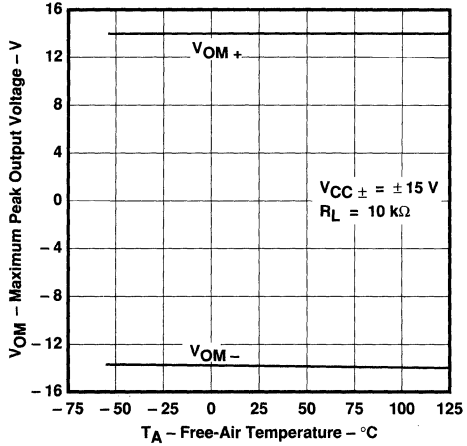


FIGURE 19

LARGE-SIGNAL VOLTAGE AMPLIFICATION
VS
LOAD RESISTANCE

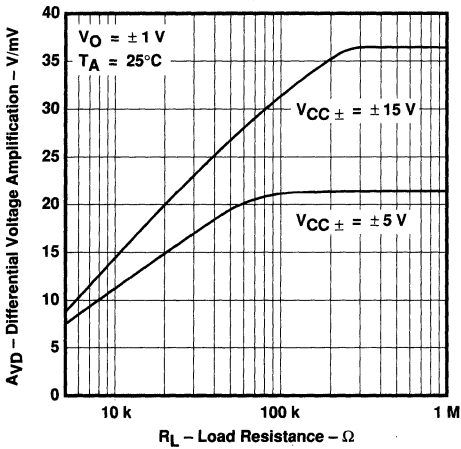


FIGURE 20

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
VS
FREQUENCY

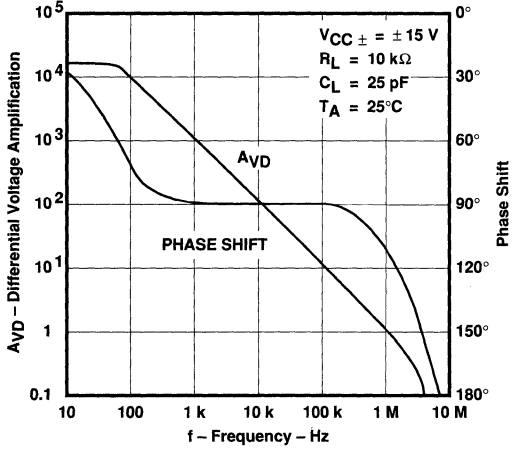


FIGURE 21

2
Operational Amplifiers

TL034, TL034A
ENHANCED JFET LOW-POWER LOW-OFFSET
QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

LARGE-SIGNAL VOLTAGE AMPLIFICATION
VS
FREE-AIR TEMPERATURE

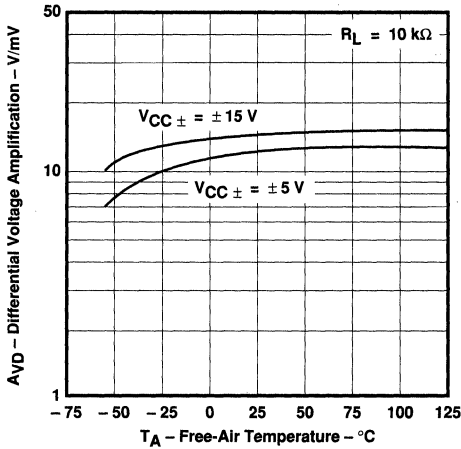


FIGURE 22

OUTPUT IMPEDANCE
VS
FREQUENCY

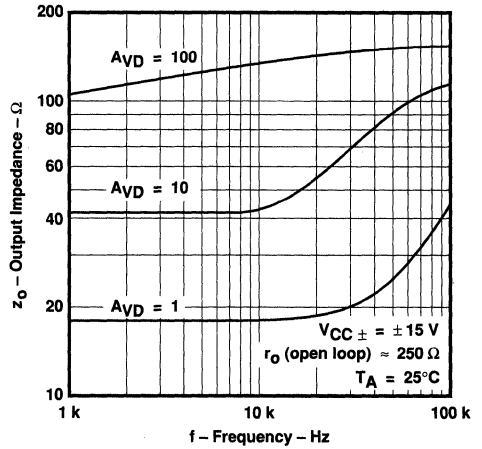


FIGURE 23

COMMON-MODE REJECTION RATIO
VS
FREQUENCY

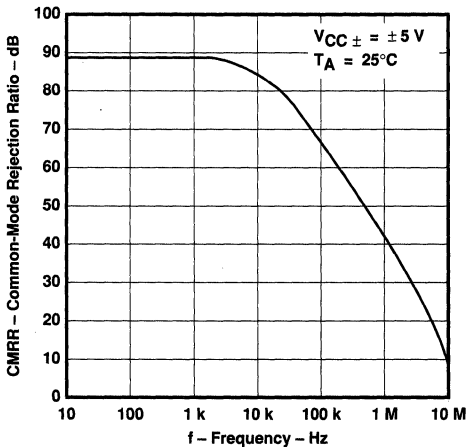


FIGURE 24

COMMON-MODE REJECTION RATIO
VS
FREQUENCY

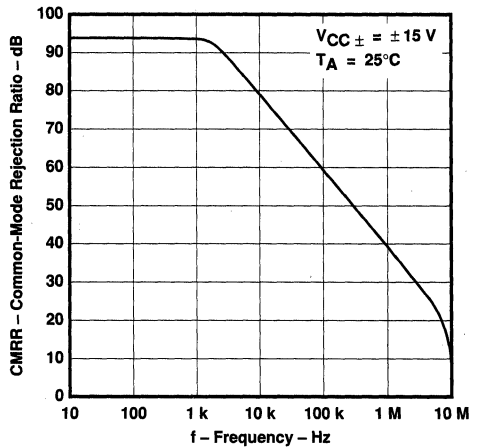


FIGURE 25

TL034, TL034A
ENHANCED JFET LOW-POWER LOW-OFFSET
QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

COMMON-MODE REJECTION RATIO
 VS
 FREE-AIR TEMPERATURE

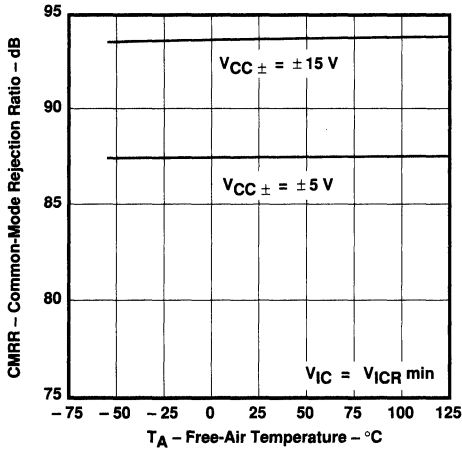


FIGURE 26

SUPPLY-VOLTAGE REJECTION RATIO
 VS
 FREE-AIR TEMPERATURE

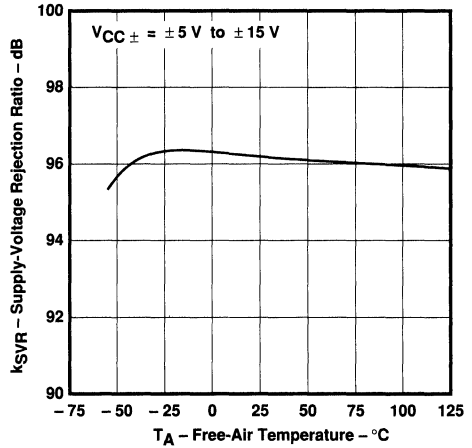


FIGURE 27

SHORT-CIRCUIT OUTPUT CURRENT
 VS
 SUPPLY VOLTAGE

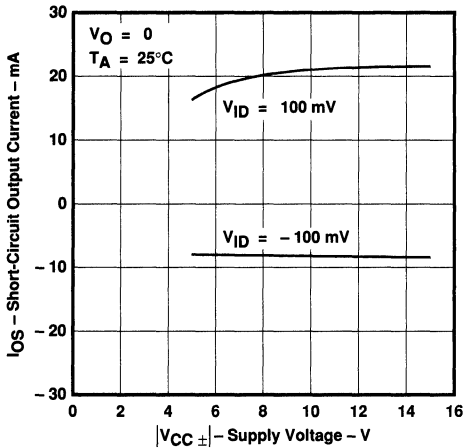


FIGURE 28

SHORT-CIRCUIT OUTPUT CURRENT
 VS
 TIME

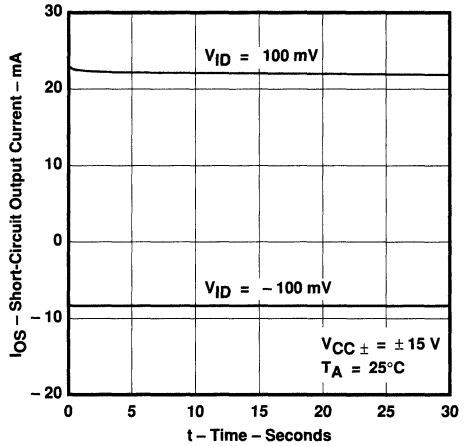


FIGURE 29

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Operational Amplifiers

TL034, TL034A
ENHANCED JFET LOW-POWER LOW-OFFSET
QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

2
Operational Amplifiers

SHORT-CIRCUIT OUTPUT CURRENT
VS
FREE-AIR TEMPERATURE

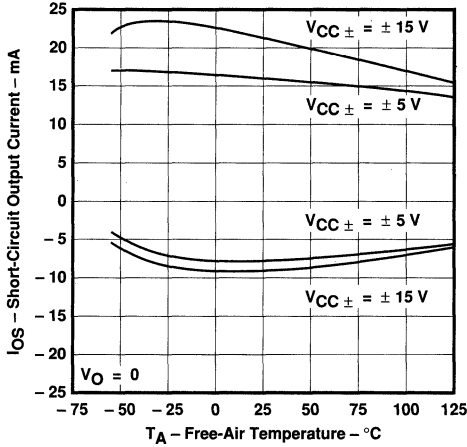


FIGURE 30

EQUIVALENT INPUT NOISE VOLTAGE
VS
FREQUENCY

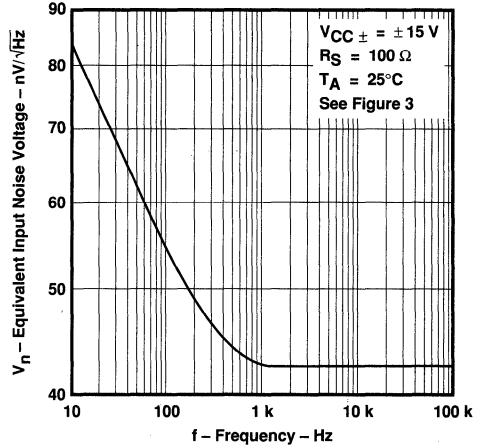


FIGURE 31

SUPPLY CURRENT
VS
SUPPLY VOLTAGE

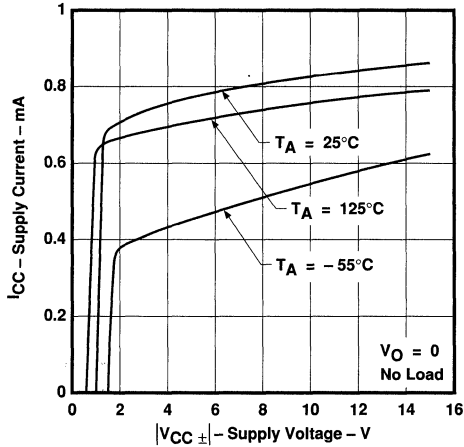


FIGURE 32

SUPPLY CURRENT
VS
FREE-AIR TEMPERATURE

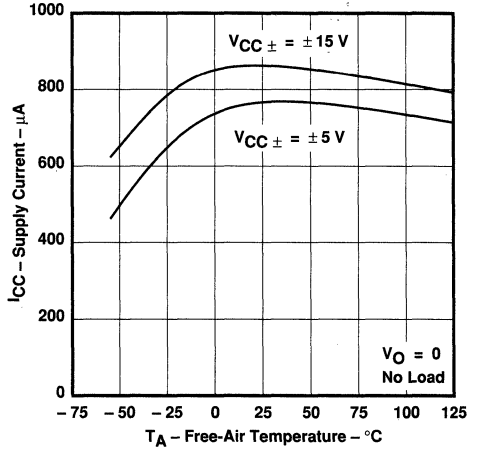


FIGURE 33

TL034, TL034A ENHANCED JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

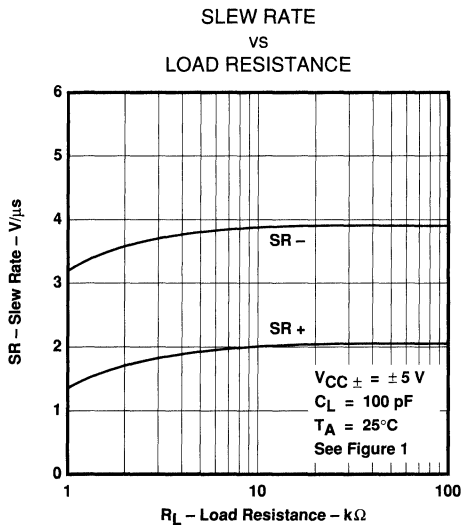


FIGURE 34

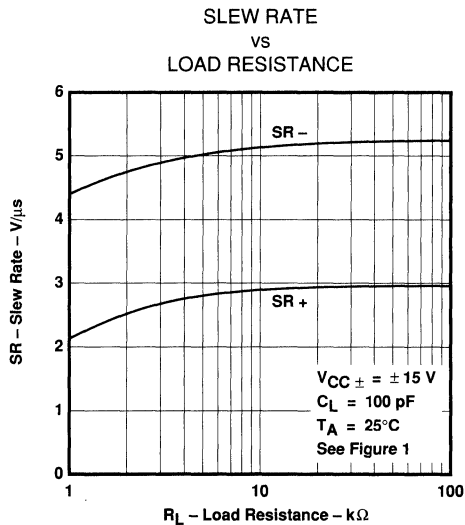


FIGURE 35

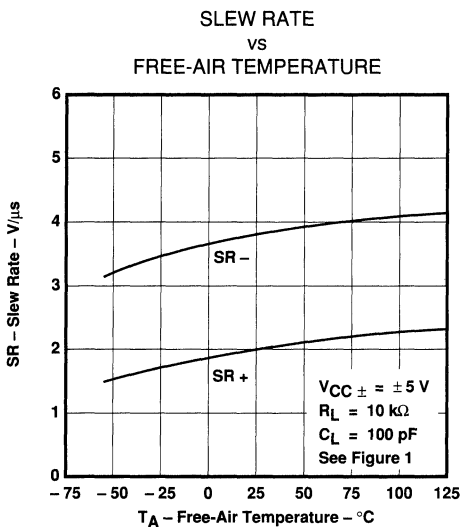


FIGURE 36

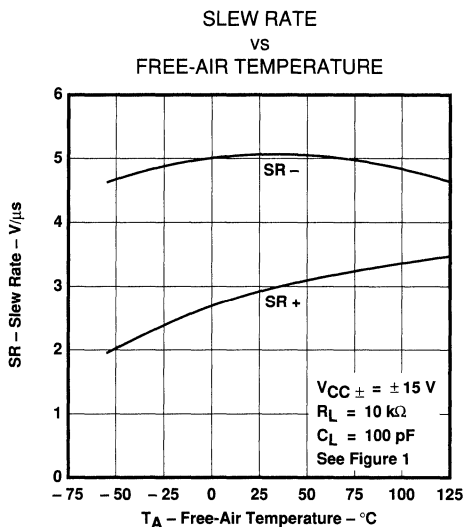


FIGURE 37

2
Operational Amplifiers

TL034, TL034A
ENHANCED JFET LOW-POWER LOW-OFFSET
QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

OVERSHOOT FACTOR
VS
LOAD CAPACITANCE

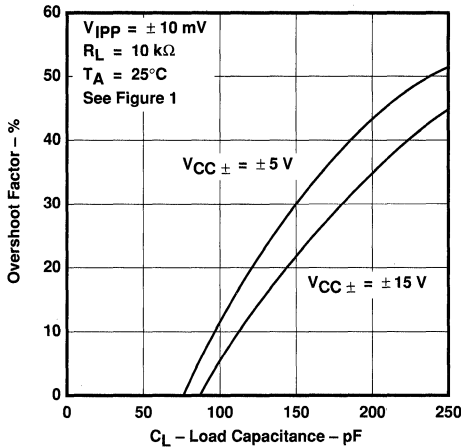


FIGURE 38

TOTAL HARMONIC DISTORTION
VS
FREQUENCY

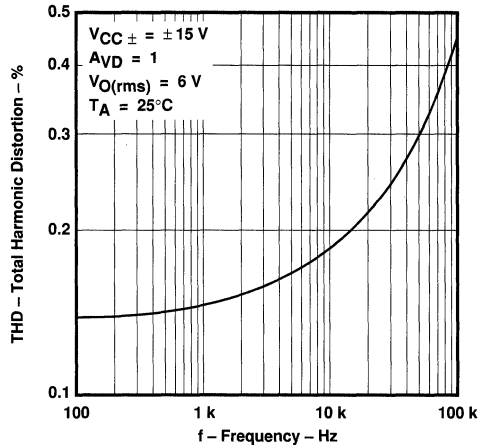


FIGURE 39

UNITY-GAIN BANDWIDTH
VS
SUPPLY VOLTAGE

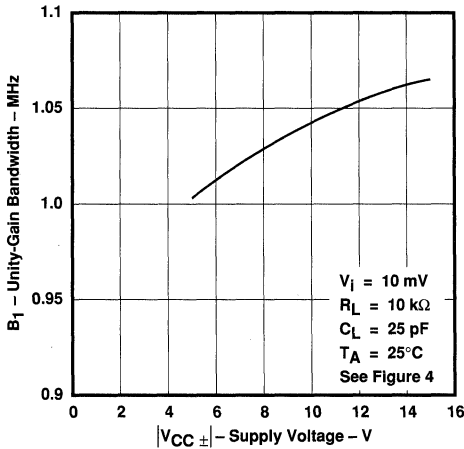


FIGURE 40

UNITY-GAIN BANDWIDTH
VS
FREE-AIR TEMPERATURE

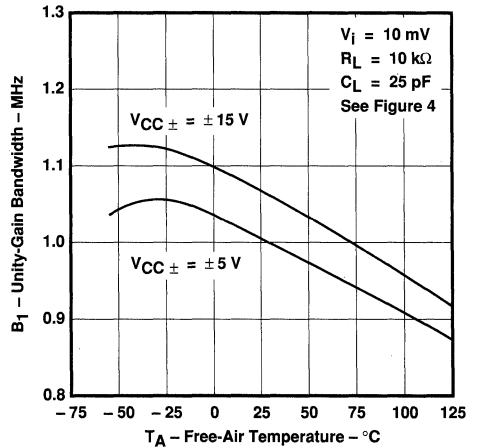


FIGURE 41

2

Operational Amplifiers

TL034, TL034A ENHANCED JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

PHASE MARGIN
vs
SUPPLY VOLTAGE

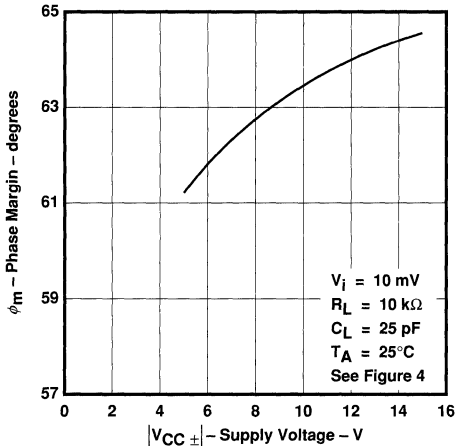


FIGURE 42

PHASE MARGIN
vs
LOAD CAPACITANCE

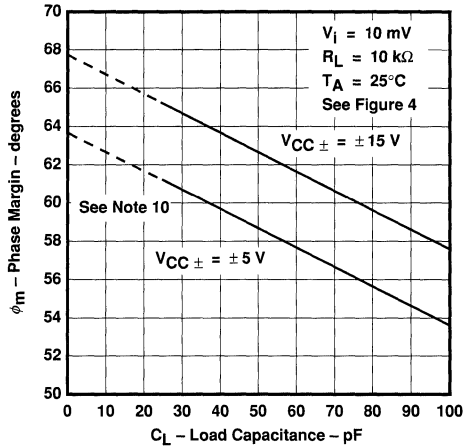


FIGURE 43

PHASE MARGIN
vs
FREE-AIR TEMPERATURE

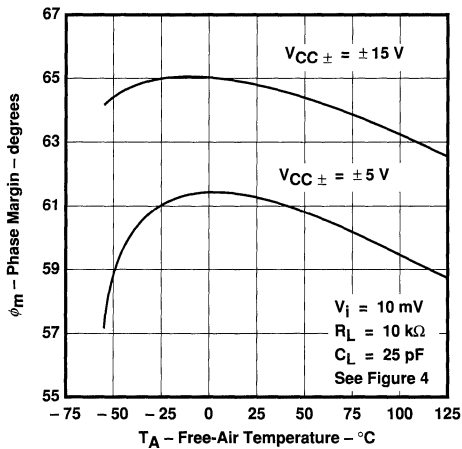


FIGURE 44

VOLTAGE-FOLLOWER
SMALL-SIGNAL
PULSE RESPONSE

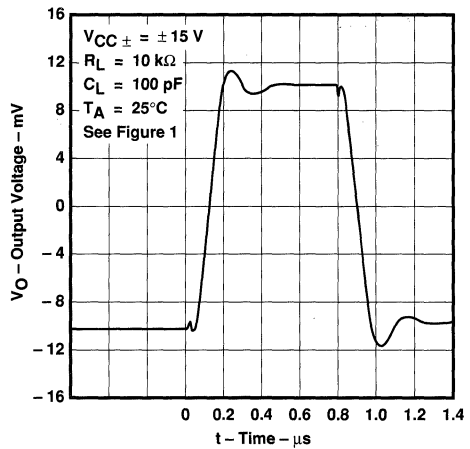


FIGURE 45

NOTE 10: Values of phase margin below a load capacitance of 25 pF were estimated.

TL034, TL034A ENHANCED JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

VOLTAGE-FOLLOWER
LARGE-SIGNAL
PULSE RESPONSE

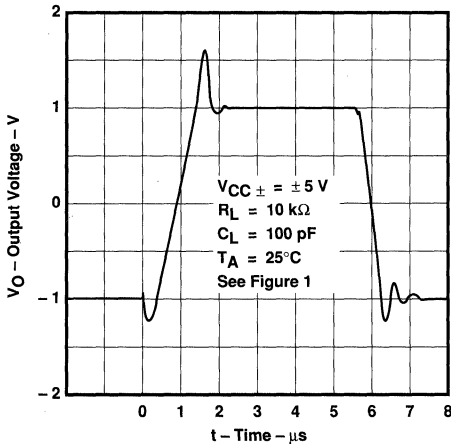


FIGURE 46

VOLTAGE-FOLLOWER
LARGE-SIGNAL
PULSE RESPONSE

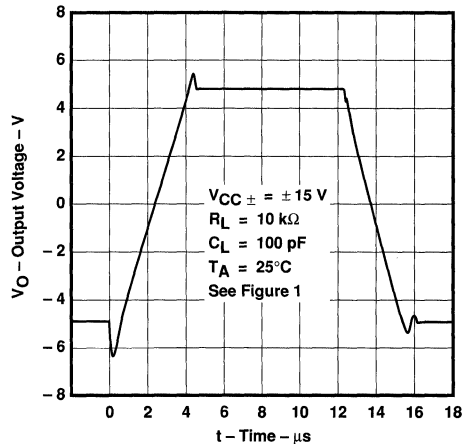


FIGURE 47

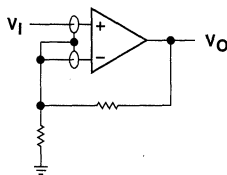
TYPICAL APPLICATION DATA

input characteristics

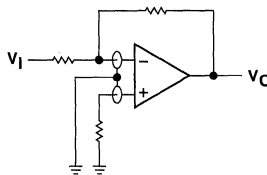
The TL034 and TL034A are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction.

Because of the extremely high input impedance and resulting low bias current requirements, the TL034 and TL034A are well-suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause degradation in system performance. It is a good practice to include guard rings around inputs (see Figure 48). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.

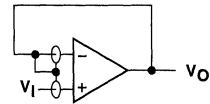
The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.



(a) NONINVERTING AMPLIFIER



(b) INVERTING AMPLIFIER



(c) UNITY-GAIN AMPLIFIER

FIGURE 48. USE OF GUARD RINGS

TL034, TL034A ENHANCED JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

output characteristics

All operating characteristics (except bandwidth and phase margin) are specified with 100 pF load capacitance. The TL034 and TL034A will drive higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem. Capacitive loads of 1000 pF and larger may be driven if enough resistance is added in series with the output (see Figure 49).

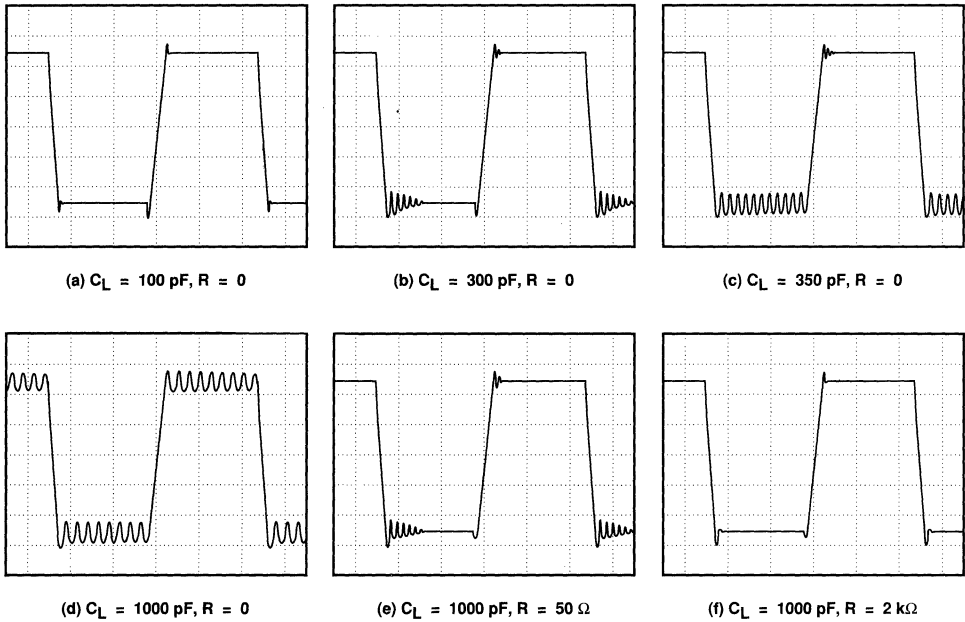
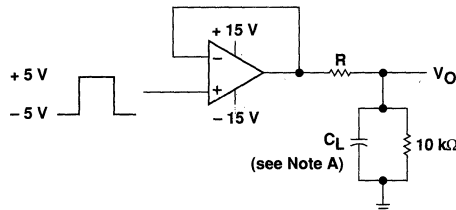


FIGURE 49. EFFECT OF CAPACITIVE LOADS



NOTE A: C_L includes fixture capacitance.

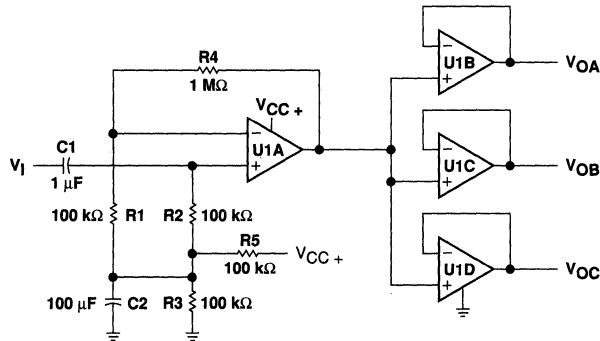
FIGURE 50. TEST CIRCUIT FOR OUTPUT CHARACTERISTICS

TL034, TL034A ENHANCED JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

audio distribution amplifier

This audio distribution amplifier feeds the input signal to three separate output channels. U1A amplifies the input signal with a gain of 10 while U1B, U1C, and U1D serve as buffers to the output channels. The gain response of this circuit is very flat from 20 Hz to 20 kHz. The TL034 allows quick response to the input signal while maintaining low power consumption.



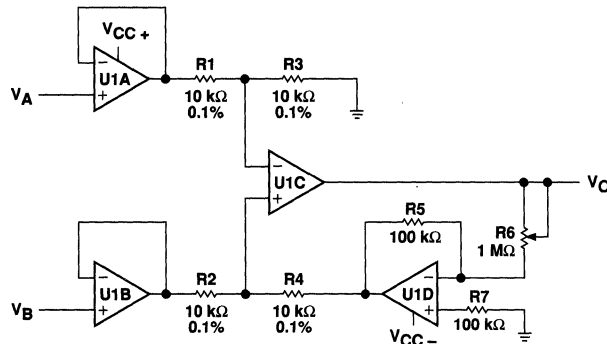
NOTE: U1A through U1D = TL034; $V_{CC+} = 5\text{ V}$.

instrumentation amplifier with linear gain adjust

The TL034 low-offset voltage and low-power consumption provides an accurate but inexpensive instrumentation amplifier. This particular configuration offers the advantage that the gain can be linearly set by one resistor:

$$V_O = \frac{R_6}{R_5} \times (V_B - V_A)$$

Adjusting R_6 varies the gain. The value of R_6 should always be greater or equal to the value of R_5 in order to ensure stability. The disadvantage of this instrumentation amplifier topology is the high degree of CMRR degradation resulting from mismatches between R_1 , R_2 , R_3 , and R_4 . For this reason, these four resistors should be 0.1% tolerance resistors.



NOTE: U1A through U1D = TL034; $V_{CC\pm} = \pm 15\text{ V}$.

TL044M, TL044C QUAD LOW-POWER OPERATIONAL AMPLIFIERS

D1662, SEPTEMBER 1973—REVISED JUNE 1988

- Very Low Power Consumption
- Typical Power Dissipation with $\pm 2\text{-V}$ Supplies . . . $340 \mu\text{W}$
- Low Input Bias and Offset Currents
- Output Short-Circuit Protection
- Low Input Offset Voltage
- Internal Frequency Compensation
- Latch-Up-Free Operation
- Power Applied in Pairs

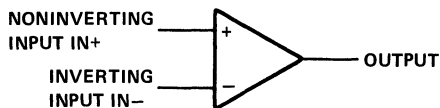
TL044M IS NOT RECOMMENDED FOR NEW DESIGNS.

description

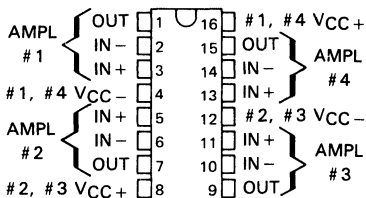
The TL044 is a quad low-power operational amplifier designed to replace higher-power devices in many applications without sacrificing system performance. High input impedance, low supply currents, and low equivalent input noise voltage over a wide range of operating supply voltages result in an extremely versatile operational amplifier for use in a variety of analog applications including battery-operated circuits. Internal frequency compensation, absence of latch-up, high slew rate, and output short-circuit protection assure ease of use. Power may be applied separately to Section A (amplifiers 1 and 4) or Section B (amplifiers 2 and 3) while the other pair remains unpowered.

The TL044M is characterized for operation over the full military temperature range of -55°C to 125°C ; the TL044C is characterized for operation from 0°C to 70°C .

symbol (each amplifier)

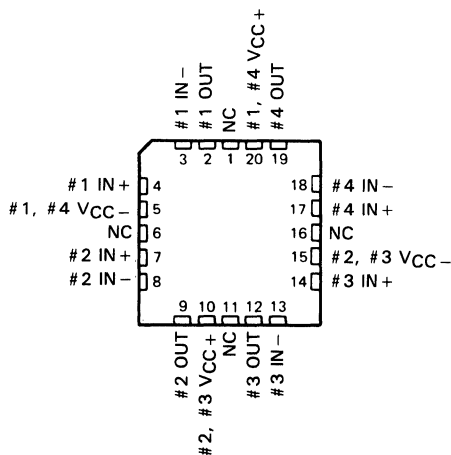


TL044M . . . J OR W DUAL-IN-LINE PACKAGE
TL044C . . . J OR N PACKAGE
(TOP VIEW)



Pins 4 and 12 are internally connected together in the N package only.

TL044M . . . FK PACKAGE
(TOP VIEW)



NC—No internal connection

2

Operational Amplifiers

TL044M, TL044C QUAD LOW-POWER OPERATIONAL AMPLIFIERS

AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE			
		CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	FLAT PACK (W)
0°C to 70°C	5 mV	—	TL044CJ	TL044CN	—
-55°C to 125°C	5 mV	TL044MFK	TL044MJ	—	TL044MW

2 absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Operational Amplifiers

	TL044M	TL044C	UNIT
Supply voltage V _{CC+} (see Note 1)	22	18	V
Supply voltage V _{CC-} (see Note 1)	-22	-18	V
Differential input voltage (see Note 2)	±30	±30	V
Input voltage (any input, see Notes 1 and 3)	±15	±15	V
Duration of output short-circuit (see Note 4)	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Case temperature for 60 seconds	FK package	260	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J or W package	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	N package	260	°C

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-}.
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
 4. The output may be shorted to ground or either power supply. For the TL044M only, the unlimited duration of the short-circuit applies at (or below) 125°C case temperature or 85°C free-air temperature.

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C	DERATING FACTOR	DERATE ABOVE T _A	T _A = 70°C	T _A = 125°C
	POWER RATING			POWER RATING	POWER RATING
FK	680 mW	11.0 mW/°C	88°C	680 mW	275 mW
J (TL044M)	680 mW	11.0 mW/°C	88°C	680 mW	275 mW
J (TL044C)	680 mW	8.2 mW/°C	67°C	656 mW	—
N	680 mW	N/A	N/A	680 mW	—
W	680 mW	8.0 mW/°C	65°C	640 mW	200 mW

TL044M, TL044C

QUAD LOW-POWER OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$

PARAMETER	TEST CONDITIONS†	TL044M			TL044C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0$, $R_S = 50\ \Omega$	25°C	1	5	1	5	mV	
		Full range		6		7.5		
I_{IO} Input offset current	$V_O = 0$	25°C	5	40	15	80	nA	
		Full range		100		200		
I_{IB} Input bias current	$V_O = 0$	25°C	50	100	100	250	nA	
		Full range		250		400		
V_{ICR} Common-mode input voltage range		25°C	±12	±13	±12	±13	V	
		Full range	±12		±12			
V_{OPP} Maximum peak-to-peak output voltage swing	$R_L = 10\text{ k}\Omega$	25°C	20	26	20	26	V	
	$R_L \geq 10\text{ k}\Omega$	Full range	20		20			
AVD Large-signal differential voltage amplification	$R_L \geq 10\text{ k}\Omega$, $V_O = \pm 10\text{ V}$	25°C	72	86	60	80	dB	
		Full range	66		60			
B_1 Unity-gain bandwidth		25°C	0.5		0.5		MHz	
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\ min}$, $V_O = 0$, $R_S = 50\ \Omega$	25°C	60	72	60	72	dB	
		Full range	60		60			
k_{SVS} Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	$V_{CC} = \pm 9\text{ V to } \pm 15\text{ V}$, $V_O = 0$, $R_S = 50\ \Omega$	25°C	30		30	200	$\mu\text{V/V}$	
		Full range				200		
V_n Equivalent input noise voltage	AVD = 20 dB, B = 1 Hz, f = 1 kHz	25°C	50		50		nV/ $\sqrt{\text{Hz}}$	
I_{OS} Short-circuit output current		25°C	±6		±6		mA	
I_{CC} Supply current (four amplifiers)	No load, $V_O = 0\text{ V}$	25°C	250	400	250	500	μA	
		Full range	400		500			
P_D Total dissipation (four amplifiers)	No load, $V_O = 0\text{ V}$	25°C	7.5	12	7.5	15	mW	
		Full range	12		15			

†All characteristics are measured under open-loop conditions with zero common-mode input voltage, unless otherwise specified. Full range for TL044M is -55°C to 125°C and for TL044C is 0°C to 70°C .

operating characteristics, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TL044M			TL044C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
t_r Rise time	$V_I = 20\text{ mV}$, $R_L = 10\text{ k}\Omega$,	0.3			0.3			μs
Overshoot factor	$C_L = 100\text{ pF}$, See Figure 1	5%			5%			
SR Slew rate at unity gain	$V_I = 10\text{ V}$, $C_L = 100\text{ pF}$, See Figure 1	0.5			0.5			V/ μs

2

Operational Amplifiers

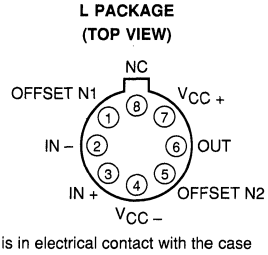
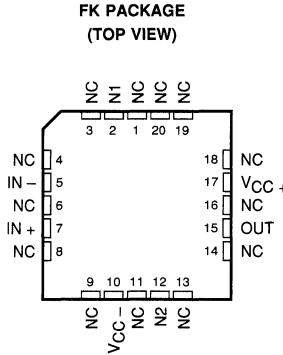
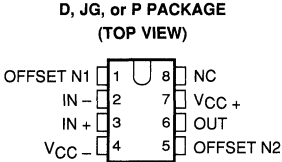
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Operational Amplifiers

TL051, TL051A ENHANCED JFET PRECISION OPERATIONAL AMPLIFIERS

D3234, JUNE 1988 - REVISED FEBRUARY 1989

- Maximum Offset Voltage ... 800 μV (TL051A)
- High Slew Rate ... 19.8 $\text{V}/\mu\text{s}$ Typ at 25°C
- Low Total Harmonic Distortion ... 0.003% Typ at $R_L = 2 \text{ k}\Omega$
- Low Noise Voltage ... 18 $\text{nV}/\sqrt{\text{Hz}}$ Typ at $f = 1 \text{ kHz}$
- Low Input Bias Currents ... 30 pA Typ



NC - No internal connection

description

The TL051 and TL051A operational amplifiers incorporate well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. These devices offer the significant advantages of Texas Instruments new enhanced JFET process. This process affords not only low initial offset voltage due to the on-chip zener trim capability but also stable offset voltage over time and temperature. In comparison, traditional JFET processes are plagued by significant offset voltage drift.

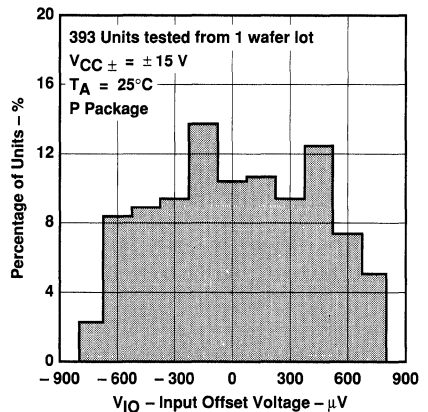
This new enhanced process still maintains the traditional JFET advantages of fast slew rates and low input bias and offset currents. These advantages coupled with low noise and low harmonic distortion make the TL051 well-suited for new state-of-the-art designs as well as existing design upgrades. The TL051 has been

AVAILABLE OPTIONS

T _A	V _{IO} max AT 25°C	PACKAGE				
		SMALL- OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)
0°C to 70°C	800 μV 1500 μV	TL051ACD TL051CD	—	TL051ACJG TL051CJG	TL051ACL TL051CL	TL051ACP TL051CP
-40°C to 85°C	800 μV 1500 μV	TL051AID TL051ID	—	TL051AIJG TL051IJG	TL051AIL TL051IL	TL051AIP TL051IP
-55°C to 125°C	800 μV 1500 μV	TL051AMD TL051MD	TL051AMFK TL051MFK	TL051AMJG TL051MJG	TL051AML TL051ML	TL051AMP TL051MP

D packages are available taped-and-reeled. Add "R" suffix to device type (e.g., TL051CDR).

DISTRIBUTION OF TL051A INPUT OFFSET VOLTAGE



2

Operational Amplifiers

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS
INSTRUMENTS**

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TL051, TL051A ENHANCED JFET PRECISION OPERATIONAL AMPLIFIERS

description (continued)

designed to be functionally compatible, as well as pin compatible, with the TL071 and TL081. Two offset voltage grades are available: TL051 (1.5 mV max) and TL051A (800 μ V max).

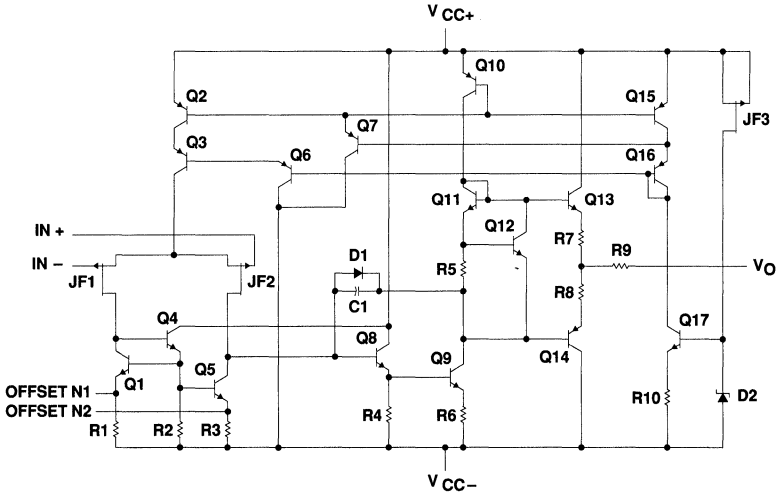
A variety of available packaging options includes small-outline and chip carrier versions for high-density system applications.

The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I-suffix devices are characterized for operation from -40°C to 85°C , and the C-suffix devices are characterized for operation from 0°C to 70°C .

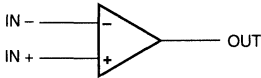
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equivalent schematic (each amplifier)

Operational Amplifiers



symbol (each amplifier)



TL051M, TL051AM ENHANCED JFET PRECISION OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL051M	25°C	0.75	3.5	0.59	1.5	mV	
			Full range	6.5			4.5		
		TL051AM	25°C	0.55	2.8	0.35	0.8		
			Full range	5.8			3.8		
α _{VIO} Temperature coefficient of input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL051M	25°C to 125°C	8			8	μV/°C	
		TL051AM	25°C to 125°C	8			8		
Input offset voltage long-term drift (see Note 5)		25°C	0.04			0.04	μV/mo		
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	4	100	5	100	pA		
		125°C	1	20	2	20	nA		
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	20	200	30	200	pA		
		125°C	10	50	20	50	nA		
V _{ICR} Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	V		
		Full range	-1 to 4		-11 to 11				
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.2	13	13.9	V		
		Full range	3			13			
	R _L = 2 kΩ	25°C	2.5	3.8	11.5	12.7			
		Full range	2.5			11.5			
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-2.5	-3.5	-12	-13.2	V		
		Full range	-2.5			-12			
	R _L = 2 kΩ	25°C	-2.3	-3.2	-11	-12			
		Full range	-2.3			-11			
A _{VD} Large-signal differential voltage amplification	R _L = 2 kΩ, See Note 6	25°C	25	59	50	105	V/mV		
		-55°C	30	76	60	149			
		125°C	10	32	15	49			
r _i Input resistance		25°C	10 ¹²			10 ¹²	Ω		
C _i Input capacitance		25°C	10			12	pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	65	85	75	93	dB		
		-55°C	65	83	75	92			
		125°C	65	84	75	94			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75	99	75	99	dB		
		-55°C	75	98	75	98			
		125°C	75	100	75	100			
I _{CC} Supply current (four amplifiers)	No load, V _O = 0	25°C	2.6	3.2	2.7	3.2	mA		
		-55°C	2.3	3.2	2.4	3.2			
		125°C	2.4	3.2	2.5	3.2			

[†] Full range is -55°C to 125°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. For V_{CC} ± = ± 5 V, V_O = ± 2.3 V; at V_{CC} ± = ± 15 V, V_O = ± 10 V.

2

Operational Amplifiers

TL051M, TL051AM
ENHANCED JFET PRECISION
OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR + Positive slew rate at unity gain	R _L = 2 kΩ, C _L = 100 pF, See Figure 1 and Note 7	25°C		18.2		15	23.7	V/μs	
		-55°C		17.5		20			
		125°C		15		21.2			
SR - Negative slew rate at unity gain		25°C		16.5		15	19.8		
		-55°C		15.1		17			
		125°C		14.8		18.2			
t _r Rise time	V _I PP = ± 10 mV, R _L = 2 kΩ, C _L = 100 pF, See Figures 1 and 2	25°C		55		56	ns		
		-55°C		51		52			
		125°C		68		68			
t _f Fall time		25°C		55		57			
		-55°C		51		52			
		125°C		68		69			
Overshoot factor		25°C		24%		19%			
		-55°C		25%		19%			
		125°C		25%		19%			
V _n Equivalent input noise voltage		R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C	75			75	nV/√Hz
			f = 1 kHz	25°C	18			19	
V _{NPP} Peak-to-peak equivalent input noise voltage			f = 10 Hz to 10 kHz	25°C	4			4	μV
I _n Equivalent input noise current	f = 1 kHz	25°C		0.01		0.01	pA/√Hz		
THD Total harmonic distortion	R _S = 1 kΩ, R _L = 2 kΩ, f = 1 kHz, See Note 8	25°C		0.003%		0.003%			
B ₁ Unity-gain bandwidth	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C		3		3.1	MHz		
		-55°C		3.6		3.7			
		125°C		2.3		2.4			
φ _m Phase margin at unity gain	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C		59°		62°			
		-55°C		57°		61°			
		125°C		59°		62°			

2
Operational Amplifiers

[†] Full range is -55°C to 125°C.

NOTES: 7. For V_{CC} ± = ± 5 V, V_IPP = ± 1 V; for V_{CC} ± = ± 15 V, V_IPP = ± 5 V.

8. For V_{CC} ± = ± 5 V, V_{o(rms)} = 1 V; for V_{CC} ± = ± 15 V, V_{o(rms)} = 6 V.

TL051I, TL051AI ENHANCED JFET PRECISION OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL051I	25°C	0.75	3.5	0.59	1.5	mV	
			Full range	5.3			3.3		
		TL051AI	25°C	0.55	2.8	0.35	0.8		
			Full range	4.6			2.6		
α _{VIO} Temperature coefficient of input offset voltage (see Note 9)	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL051I	25°C to 85°C	7			8	μV/°C	
		TL051AI	25°C to 85°C	8			8 25		
Input offset voltage long-term drift (see Note 5)		25°C	0.04			0.04	μV/mo		
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	4 100			5 100	pA		
		85°C	0.06 10			0.07 10	nA		
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	20 200			30 200	pA		
		85°C	0.6 20			0.7 20	nA		
V _{ICR} Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	V		
		Full range	-1 to 4			-11 to 11			
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3 4.2			13 13.9	V		
		Full range	3			13			
	R _L = 2 kΩ	25°C	2.5 3.8			11.5 12.7			
		Full range	2.5			11.5			
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-2.5 -3.5			-12 -13.2	V		
		Full range	-2.5			-12			
	R _L = 2 kΩ	25°C	-2.3 -3.2			-11 -12			
		Full range	-2.3			-11			
A _{VD} Large-signal differential voltage amplification	R _L = 2 kΩ, See Note 6	25°C	25 59			50 105	V/mV		
		-40°C	30 74			60 145			
		85°C	20 43			30 76			
r _i Input resistance		25°C	10 ¹²			10 ¹²	Ω		
C _i Input capacitance		25°C	10			12	pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	65 85			75 93	dB		
		-40°C	65 83			75 90			
		85°C	65 84			75 93			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC±} / ΔV _{IO})	V _{CC±} = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75 99			75 99	dB		
		-40°C	75 98			75 98			
		85°C	75 99			75 99			
I _{CC} Supply current (four amplifiers)	No load, V _O = 0	25°C	2.6 3.2			2.7 3.2	mA		
		-40°C	2.4 3.2			2.6 3.2			
		85°C	2.5 3.2			2.6 3.2			

[†] Full range is -40°C to 85°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. For V_{CC±} = ± 5 V, V_O = ± 2.3 V; at V_{CC±} = ± 15 V, V_O = ± 10 V.

9. This parameter is tested on a sample basis for the TL051A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

2

Operational Amplifiers

TL0511, TL051A1 ENHANCED JFET PRECISION OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER	TEST CONDITIONS	T_A^\dagger	$V_{CC} \pm = \pm 5 \text{ V}$			$V_{CC} \pm = \pm 15 \text{ V}$			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR + Positive slew rate at unity gain	$R_L = 2 \text{ k}\Omega$, $C_L = 100 \text{ pF}$, See Figure 1 and Note 7	25°C	18.2		15	23.7			V/ μ s
		-40°C	20.1		13	23			
		85°C	16.1		13	21.9			
SR - Negative slew rate at unity gain		25°C	16.5		15	19.8			
		-40°C	16.6		13	19.4			
		85°C	15.7		13	19.1			
t_r Rise time	$V_{Ipp} = \pm 10 \text{ mV}$, $R_L = 2 \text{ k}\Omega$, $C_L = 100 \text{ pF}$, See Figures 1 and 2	25°C		55		56		ns	
t_f Fall time		-40°C		52		53			
		85°C		64		65			
		25°C		55		57			
		-40°C		51		53			
		85°C		64		65			
		Overshoot factor	25°C		24%		19%		
-40°C				24%		19%			
85°C				24%		19%			
V_n Equivalent input noise voltage (see Note 10)	$R_S = 100 \Omega$, See Figure 3	f = 10 Hz	25°C		75		nV/ $\sqrt{\text{Hz}}$		
f = 1 kHz		25°C		18	30				
V_{NPP} Peak-to-peak equivalent input noise voltage		f = 10 Hz to 10 kHz	25°C		4		μ V		
I_n Equivalent input noise current	f = 1 kHz	25°C		0.01		0.01	pA/ $\sqrt{\text{Hz}}$		
THD Total harmonic distortion	$R_S = 1 \text{ k}\Omega$, $R_L = 2 \text{ k}\Omega$, f = 1 kHz, See Note 8	25°C		0.003%		0.003%			
B_1 Unity-gain bandwidth	$V_i = 10 \text{ mV}$, $R_L = 2 \text{ k}\Omega$, $C_L = 25 \text{ pF}$, See Figure 4	25°C		3		3.1	MHz		
		-40°C		3.5		3.6			
		85°C		2.6		2.7			
ϕ_m Phase margin at unity gain	$V_i = 10 \text{ mV}$, $R_L = 2 \text{ k}\Omega$, $C_L = 25 \text{ pF}$, See Figure 4	25°C		59°		62°			
		-40°C		58°		61°			
		85°C		59°		62°			

† Full range is -40°C to 85°C.

NOTES: 7. For $V_{CC} \pm = \pm 5 \text{ V}$, $V_{Ipp} = \pm 1 \text{ V}$; for $V_{CC} \pm = \pm 15 \text{ V}$, $V_{Ipp} = \pm 5 \text{ V}$.

8. For $V_{CC} \pm = \pm 5 \text{ V}$, $V_o(\text{rms}) = 1 \text{ V}$; for $V_{CC} \pm = \pm 15 \text{ V}$, $V_o(\text{rms}) = 6 \text{ V}$.

10. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

TL051C, TL051AC ENHANCED JFET PRECISION OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL051C	25°C	0.75	3.5		0.59	1.5	mV
			Full range		4.5		2.5		
TL051AC		25°C	0.55	2.8		0.35	0.8		
		Full range		3.8		1.8			
α _{VIO} Temperature coefficient of input offset voltage (see Note 9)		TL051C	25°C to 70°C	8		8		μV/°C	
		TL051AC	25°C to 70°C	8		8	25		
Input offset voltage long-term drift (see Note 5)		25°C	0.04		0.04		μV/mo		
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	4		100	5	100	pA	
		70°C	0.02		1	0.025	1	nA	
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	20		200	30	200	pA	
		70°C	0.15		4	0.2	4	nA	
V _{ICR} Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6		-11 to 11	-12.3 to 15.6	V	
		Full range	-1 to 4			-11 to 11			
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.2		13	13.9	V	
		Full range	3		13				
	R _L = 2 kΩ	25°C	2.5	3.8		11.5	12.7		
		Full range	2.5		11.5				
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-2.5	-3.5		-12	-13.2	V	
		Full range	-2.5		-12				
	R _L = 2 kΩ	25°C	-2.3	-3.2		-11	-12		
		Full range	-2.3		-11				
A _{VD} Large-signal differential voltage amplification	R _L = 2 kΩ, See Note 6	25°C	25	59		50	105	V/mV	
		0°C	30	65		60	129		
		70°C	20	46		30	85		
r _i Input resistance		25°C	10 ¹²			10 ¹²		Ω	
C _i Input capacitance		25°C	10			12		pF	
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	65	85		75	93	dB	
		0°C	65	84		75	92		
		70°C	65	84		75	91		
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75	99		75	99	dB	
		0°C	75	98		75	98		
		70°C	75	97		75	97		
I _{CC} Supply current (four amplifiers)	No load, V _O = 0	25°C	2.6	3.2		2.7	3.2	mA	
		0°C	2.7	3.2		2.8	3.2		
		70°C	2.6	3.2		2.7	3.2		

[†] Full range is 0°C to 70°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. For V_{CC} ± = ± 5 V, V_O = ± 2.3 V; at V_{CC} ± = ± 15 V, V_O = ± 10 V.

9. This parameter is tested on a sample basis for the TL051A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

TL051C, TL051AC ENHANCED JFET PRECISION OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER	TEST CONDITIONS	T _A †	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR + Positive slew rate at unity gain	R _L = 2 kΩ, C _L = 100 pF, See Figure 1 and Note 7	25°C		18.2		15	23.7	V/μs	
		0°C		19.5		13	24.1		
		70°C		16.4		13	22.6		
SR - Negative slew rate at unity gain		25°C		16.5		15	19.8		
		0°C		16.8		13	19.9		
		70°C		16		13	19.3		
t _r Rise time	V _{Ipp} = ± 10 mV, R _L = 2 kΩ, C _L = 100 pF, See Figures 1 and 2	25°C		55		56	ns		
		0°C		54		55			
		70°C		63		63			
t _f Fall time		25°C		55		57			
		0°C		54		56			
		70°C		62		64			
Overshoot factor		25°C		24%		19%			
		0°C		24%		19%			
		70°C		24%		19%			
V _n Equivalent input noise voltage (see Note 10)	R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C	75		75	nV/√Hz		
		f = 1 kHz	25°C	18		18		30	
V _{NPP} Peak-to-peak equivalent input noise voltage	f = 10 Hz to 10 kHz	25°C	4		4	μV			
I _n Equivalent input noise current	f = 1 kHz	25°C	0.01		0.01	pA/√Hz			
THD Total harmonic distortion	R _S = 1 kΩ, R _L = 2 kΩ, f = 1 kHz, See Note 8	25°C	0.003%		0.003%				
B ₁ Unity-gain bandwidth	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C		3		3.1	MHz		
		0°C		3.2		3.3			
		70°C		2.7		2.8			
φ _m Phase margin at unity gain	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C		59°		62°			
		0°C		58°		62°			
		70°C		59°		62°			

† Full range is 0°C to 70°C.

NOTES: 7. For V_{CC} ± = ± 5 V, V_{Ipp} = ± 1 V; for V_{CC} ± = ± 15 V, V_{Ipp} = ± 5 V.

8. For V_{CC} ± = ± 5 V, V_{o(rms)} = 1 V; for V_{CC} ± = ± 15 V, V_{o(rms)} = 6 V.

10. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

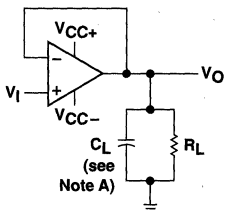
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Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

2

Operational Amplifiers



NOTE A: C_L includes fixture capacitance.

FIGURE 1. SLEW RATE, RISE/FALL TIME, AND OVERSHOOT TEST CIRCUIT

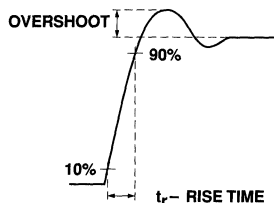


FIGURE 2. RISE TIME AND OVERSHOOT WAVEFORM

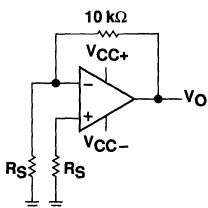
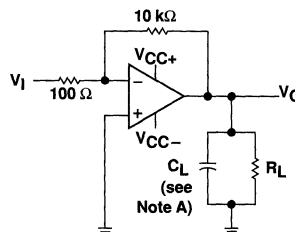


FIGURE 3. NOISE VOLTAGE TEST CIRCUIT



NOTE A: C_L includes fixture capacitance.

FIGURE 4. UNITY-GAIN BANDWIDTH AND PHASE MARGIN TEST CIRCUIT

typical values

Typical values as presented in this data sheet represent the median (50% point) of device parametric performance.

input bias and offset current

At the picoamp-bias-current level typical of the TL051 and TL051A, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To accurately measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied but with no device in the socket. The device is then inserted in the socket and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

noise

Because of the increasing emphasis on low noise levels in many of today's applications, the input noise voltage density is sample-tested at $f = 1$ kHz. Texas Instruments also has additional noise testing capability to meet specific application requirements. Please contact the factory for details.

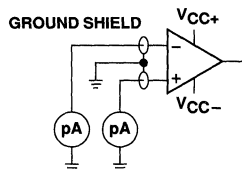


FIGURE 5. INPUT BIAS AND OFFSET CURRENT TEST CIRCUIT

TL051, TL051A
ENHANCED JFET PRECISION
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

table of graphs

		FIGURE	
V_{IO}	Input offset voltage	Distribution	6
α_{VIO}	Temperature coefficient of input offset voltage	Distribution	7
I_{IO}	Input offset current	vs Temperature	8
I_{IB}	Input bias current	vs V_{IC}	9
		vs Temperature	8
V_I	Input voltage range	vs V_{CC}	10
		vs Temperature	11
V_O	Output voltage	vs Differential input voltage	12, 13
		vs V_{CC}	14
V_{OM}	Maximum peak output voltage swing	vs Output current	18, 19
		vs Frequency	15, 16, 17
		vs Temperature	20, 21
		vs R_L	22
A_{VD}	Differential voltage amplification	vs Frequency	23
		vs Temperature	24, 25
		vs Frequency	29
z_o	Output impedance	vs Frequency	29
CMRR	Common-mode rejection ratio	vs Frequency	26, 27
		vs Temperature	28
k_{SVR}	Supply-voltage rejection ratio	vs Temperature	30
I_{OS}	Short-circuit output current	vs V_{CC}	31
		vs Time	32
		vs Temperature	33
I_{CC}	Supply current	vs V_{CC}	34
		vs Temperature	35
SR	Slew rate	vs R_L	36, 37
		vs Temperature	38, 39
	Overshoot factor	vs C_L	40
V_n	Equivalent input noise voltage	vs Frequency	41
THD	Total harmonic distortion	vs Frequency	42
B_1	Unity-gain bandwidth	vs V_{CC}	43
		vs Temperature	44
ϕ_m	Phase margin	vs V_{CC}	45
		vs C_L	46
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Operational Amplifiers

TL051, TL051A
ENHANCED JFET PRECISION
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

2
Operational Amplifiers

DISTRIBUTION OF TL051
 INPUT OFFSET VOLTAGE

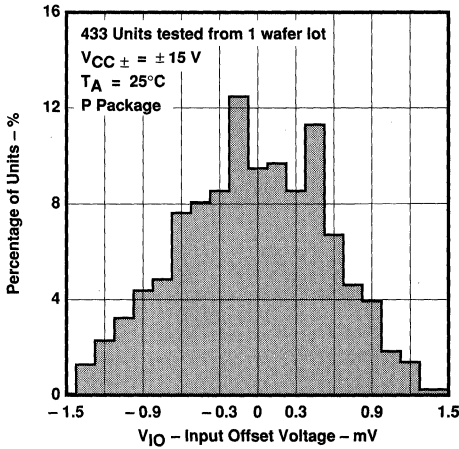


FIGURE 6

DISTRIBUTION OF TL051
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT

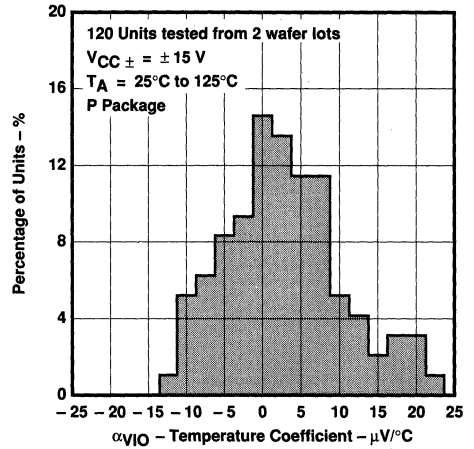


FIGURE 7

INPUT BIAS CURRENT AND
 INPUT OFFSET CURRENT
 VS
 FREE-AIR TEMPERATURE

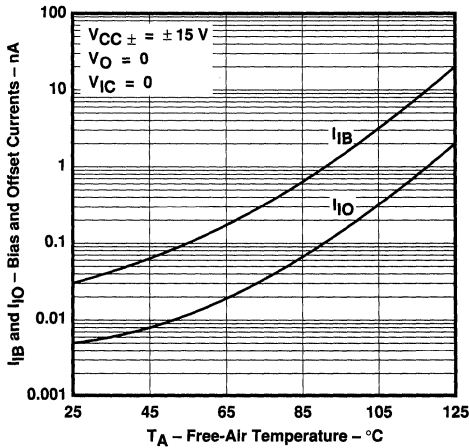


FIGURE 8

INPUT BIAS CURRENT
 VS
 COMMON-MODE INPUT VOLTAGE

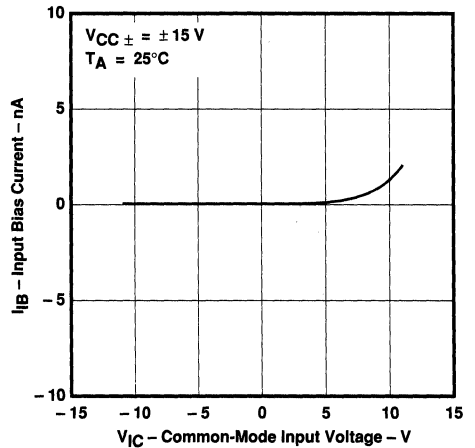


FIGURE 9

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

COMMON-MODE
 INPUT VOLTAGE RANGE LIMITS
 VS
 SUPPLY VOLTAGE

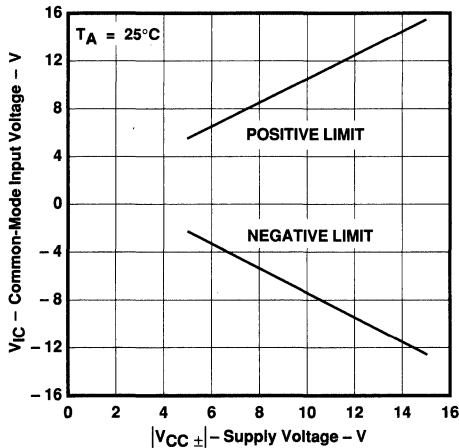


FIGURE 10

COMMON-MODE
 INPUT VOLTAGE RANGE LIMITS
 VS
 FREE-AIR TEMPERATURE

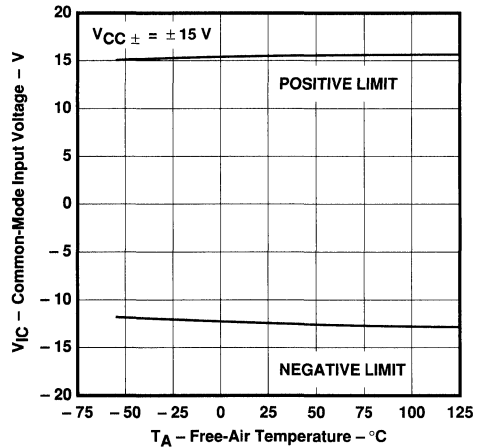


FIGURE 11

OUTPUT VOLTAGE
 VS
 DIFFERENTIAL INPUT VOLTAGE

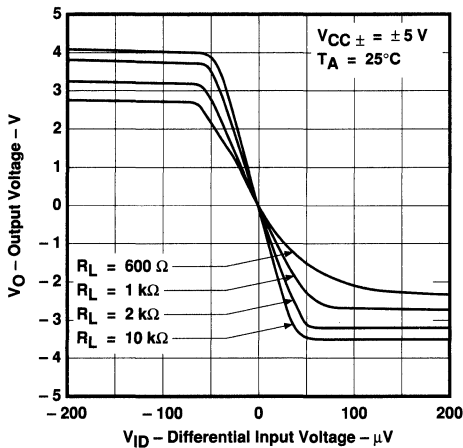


FIGURE 12

OUTPUT VOLTAGE
 VS
 DIFFERENTIAL INPUT VOLTAGE

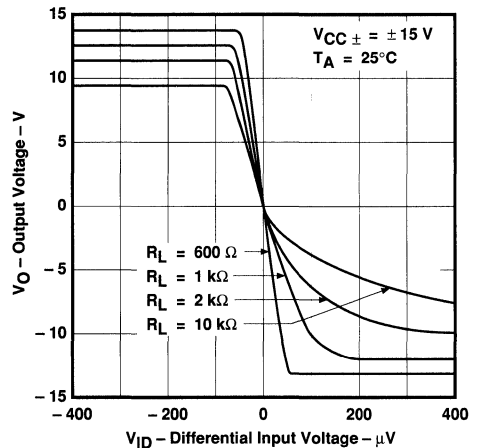


FIGURE 13

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE
vs
SUPPLY VOLTAGE

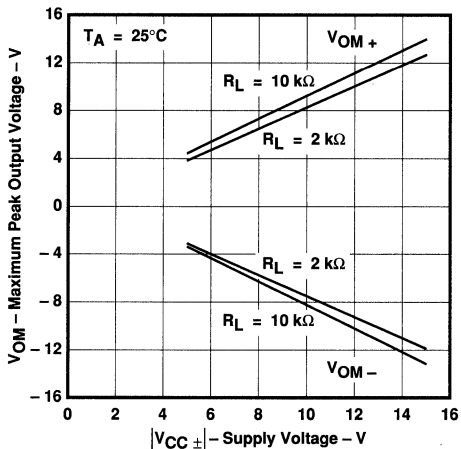


FIGURE 14

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
vs
FREQUENCY

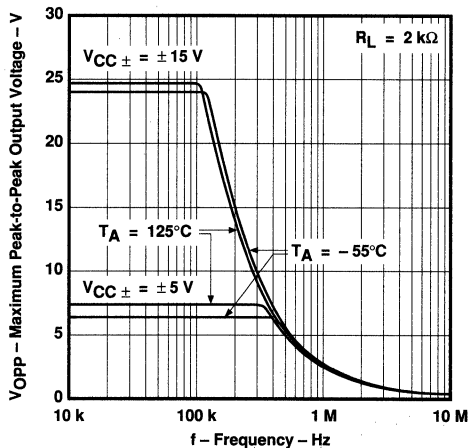


FIGURE 15

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
vs
FREQUENCY

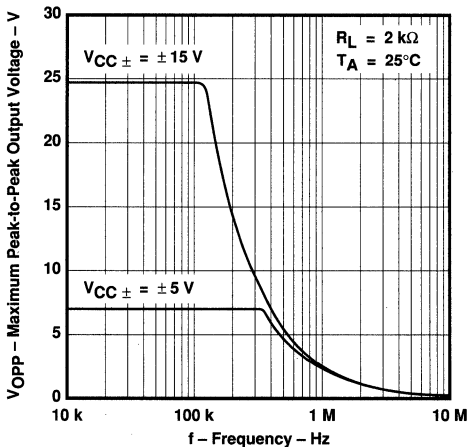


FIGURE 16

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
vs
FREQUENCY

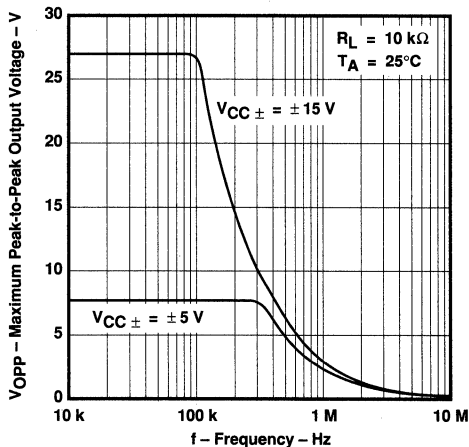


FIGURE 17

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 OUTPUT CURRENT

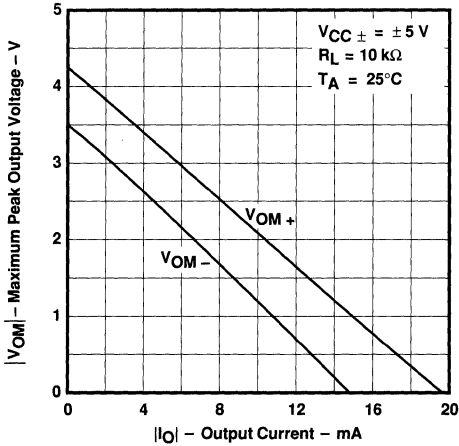


FIGURE 18

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 OUTPUT CURRENT

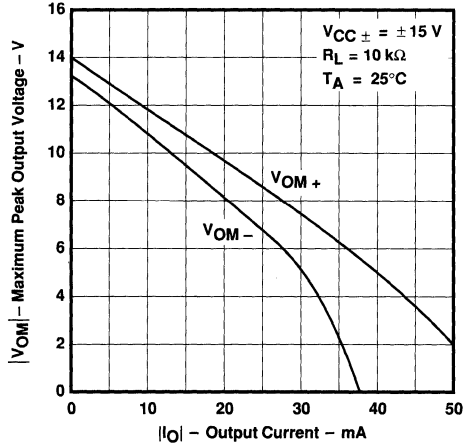


FIGURE 19

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

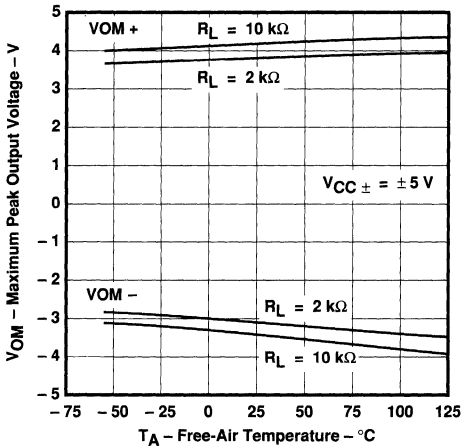


FIGURE 20

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

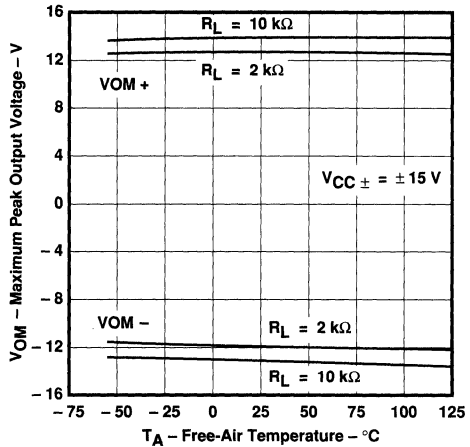


FIGURE 21

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL051, TL051A ENHANCED JFET PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

LARGE-SIGNAL VOLTAGE AMPLIFICATION
vs
LOAD RESISTANCE

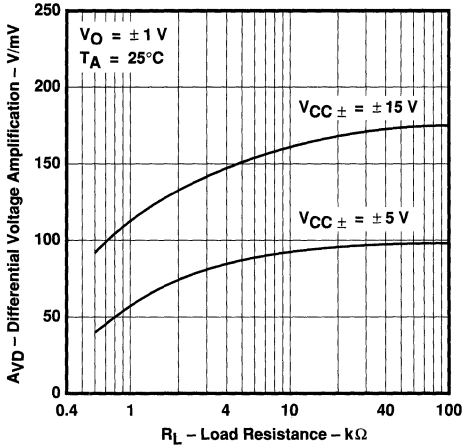


FIGURE 22

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
vs
FREQUENCY

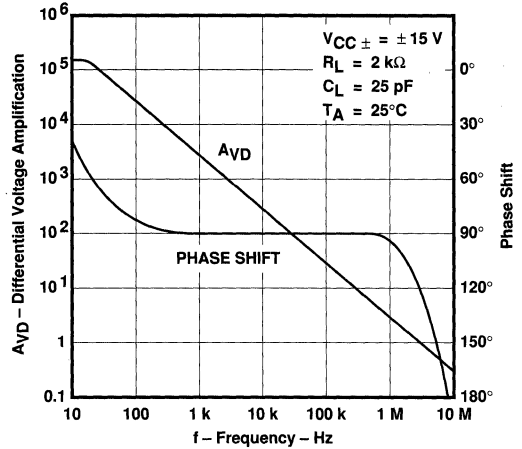


FIGURE 23

LARGE-SIGNAL VOLTAGE AMPLIFICATION
vs
FREE-AIR TEMPERATURE

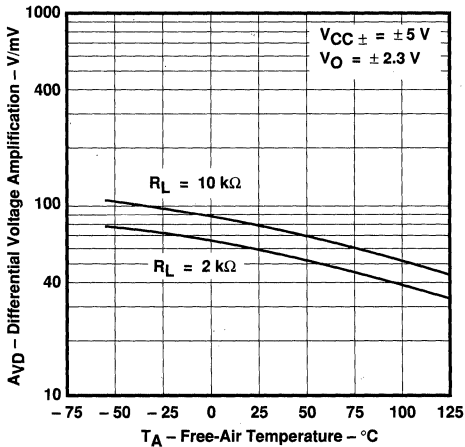


FIGURE 24

LARGE-SIGNAL VOLTAGE AMPLIFICATION
vs
FREE-AIR TEMPERATURE

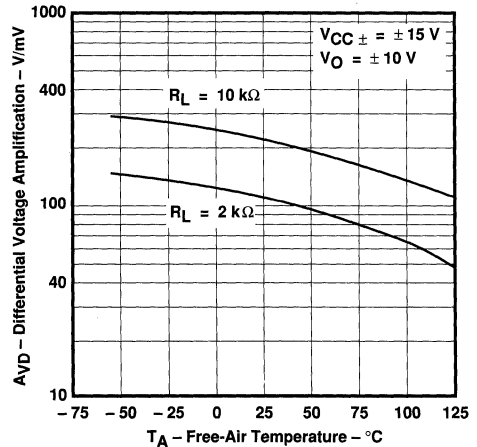


FIGURE 25

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

**COMMON-MODE REJECTION RATIO
VS
FREQUENCY**

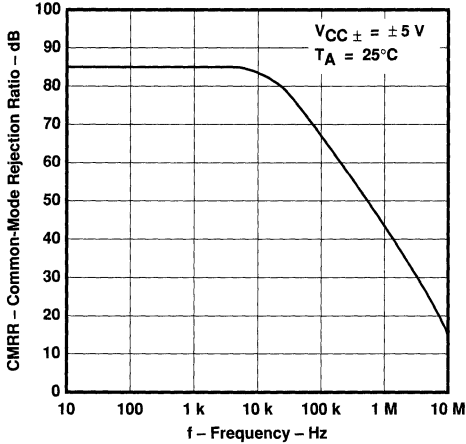


FIGURE 26

**COMMON-MODE REJECTION RATIO
VS
FREQUENCY**

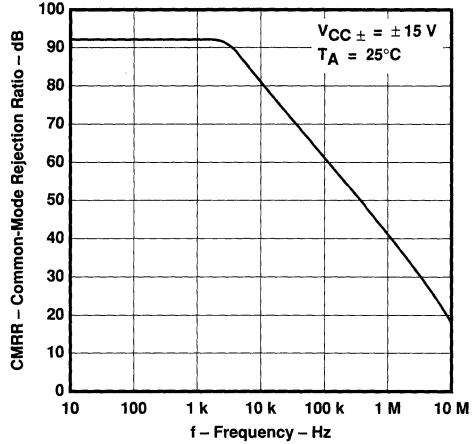


FIGURE 27

**COMMON-MODE REJECTION RATIO
VS
FREE-AIR TEMPERATURE**

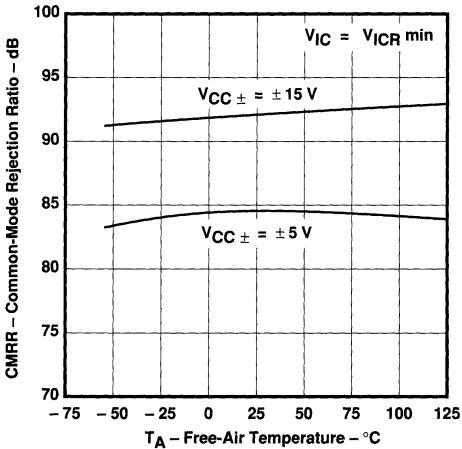


FIGURE 28

**OUTPUT IMPEDANCE
VS
FREQUENCY**

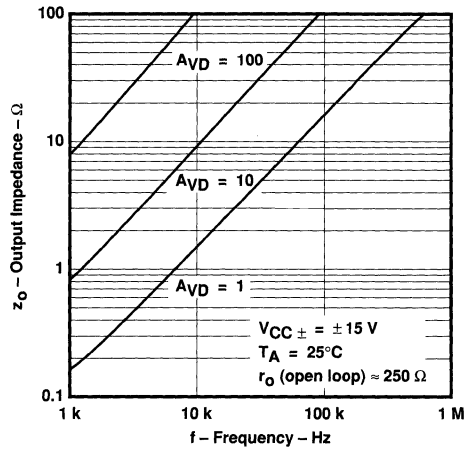


FIGURE 29

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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TYPICAL CHARACTERISTICS†

**SUPPLY-VOLTAGE REJECTION RATIO
VS
FREE-AIR TEMPERATURE**

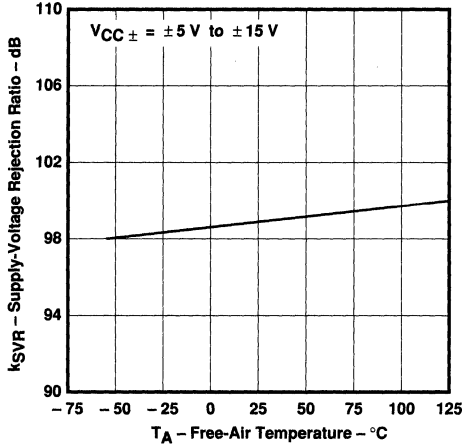


FIGURE 30

**SHORT-CIRCUIT OUTPUT CURRENT
VS
SUPPLY VOLTAGE**

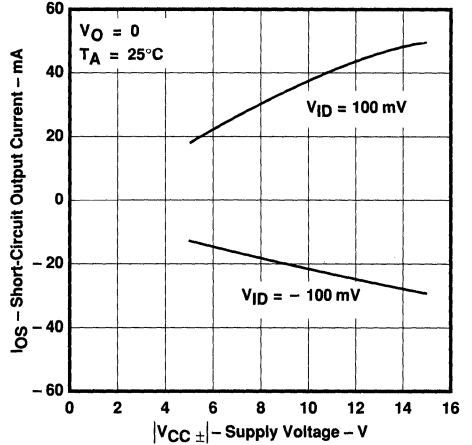


FIGURE 31

**SHORT-CIRCUIT OUTPUT CURRENT
VS
TIME**

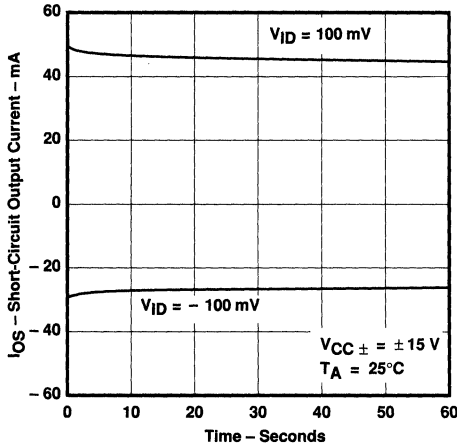


FIGURE 32

**SHORT-CIRCUIT OUTPUT CURRENT
VS
FREE-AIR TEMPERATURE**

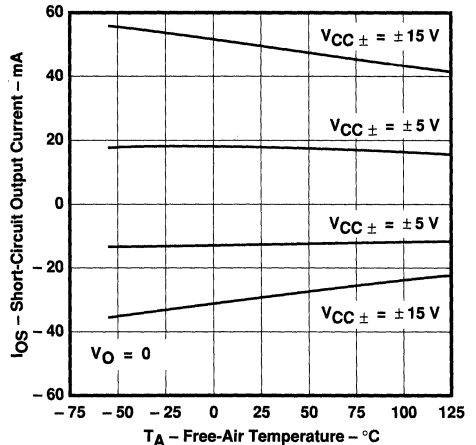


FIGURE 33

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL051, TL051A ENHANCED JFET PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

SUPPLY CURRENT
VS
SUPPLY VOLTAGE

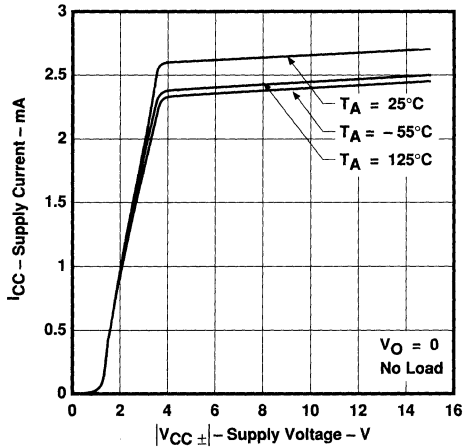


FIGURE 34

SUPPLY CURRENT
VS
FREE-AIR TEMPERATURE

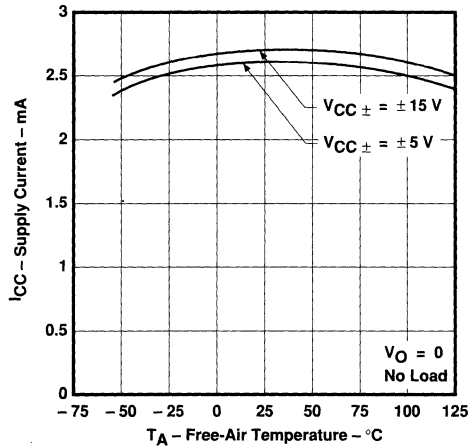


FIGURE 35

SLEW RATE
VS
LOAD RESISTANCE

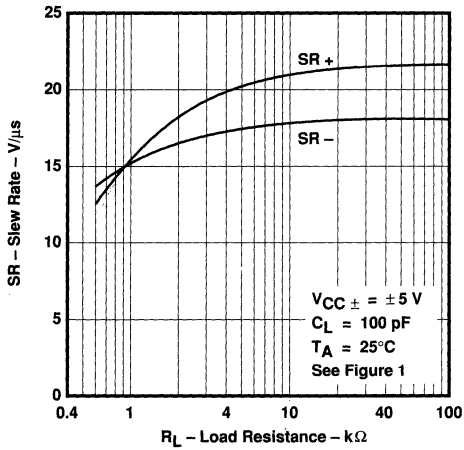


FIGURE 36

SLEW RATE
VS
LOAD RESISTANCE

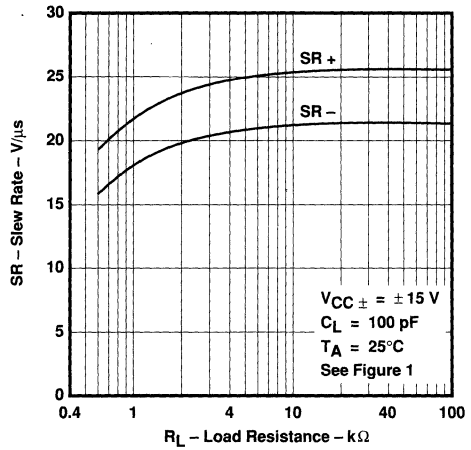


FIGURE 37

2
Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

SLEW RATE
vs
FREE-AIR TEMPERATURE

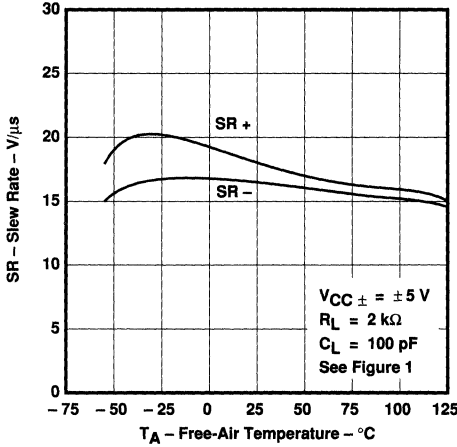


FIGURE 38

SLEW RATE
vs
FREE-AIR TEMPERATURE

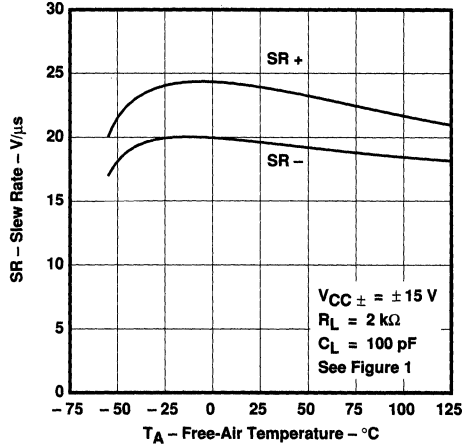


FIGURE 39

OVERSHOOT FACTOR
vs
LOAD CAPACITANCE

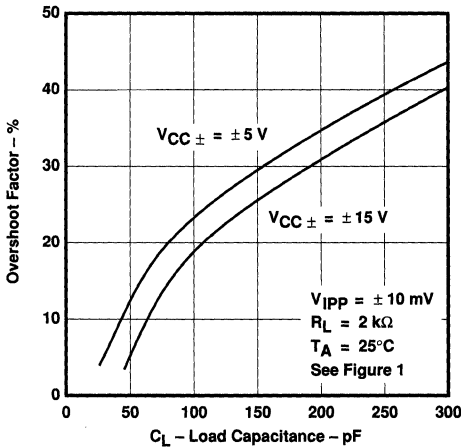


FIGURE 40

EQUIVALENT INPUT NOISE VOLTAGE
vs
FREQUENCY

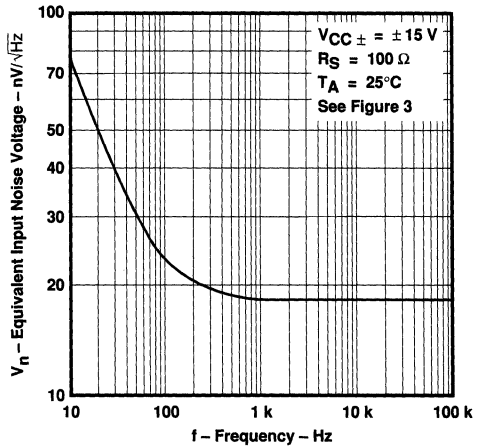


FIGURE 41

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

**TOTAL HARMONIC DISTORTION
VS
FREQUENCY**

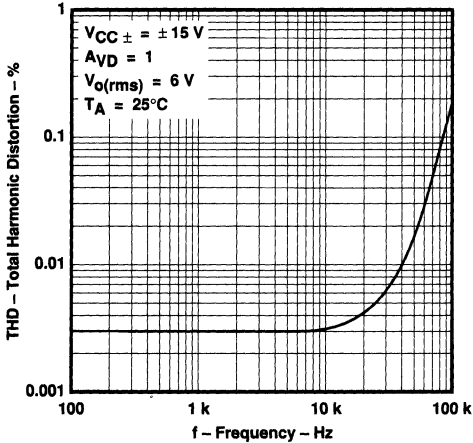


FIGURE 42

**UNITY-GAIN BANDWIDTH
VS
SUPPLY VOLTAGE**

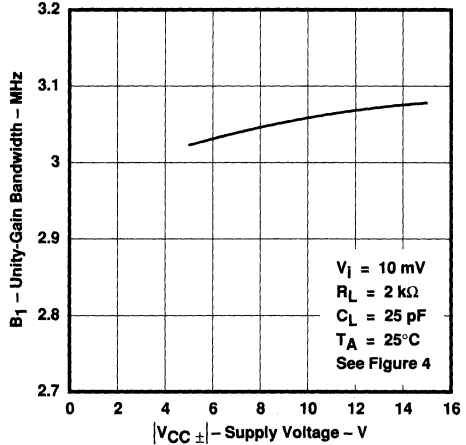


FIGURE 43

**UNITY-GAIN BANDWIDTH
VS
FREE-AIR TEMPERATURE**

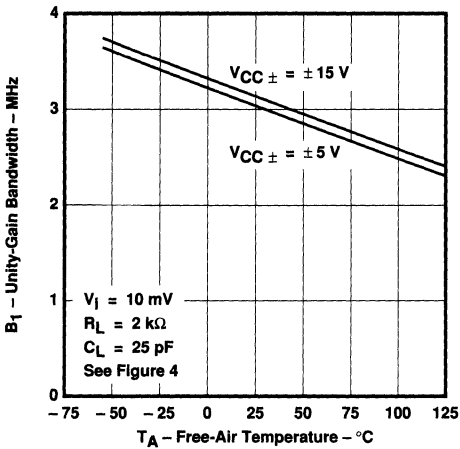


FIGURE 44

**PHASE MARGIN
VS
SUPPLY VOLTAGE**

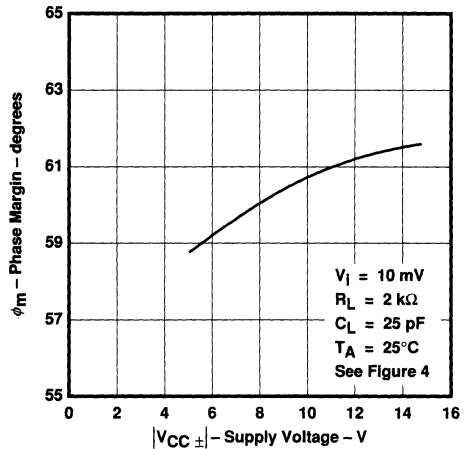


FIGURE 45

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL051, TL051A ENHANCED JFET PRECISION OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

2 Operational Amplifiers

PHASE MARGIN
VS
LOAD CAPACITANCE

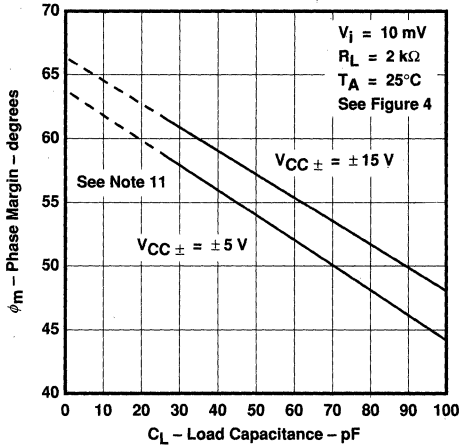


FIGURE 46

PHASE MARGIN
VS
FREE-AIR TEMPERATURE

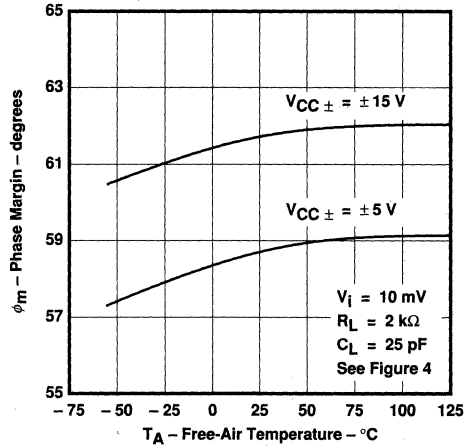


FIGURE 47

VOLTAGE-FOLLOWER
SMALL-SIGNAL
PULSE RESPONSE

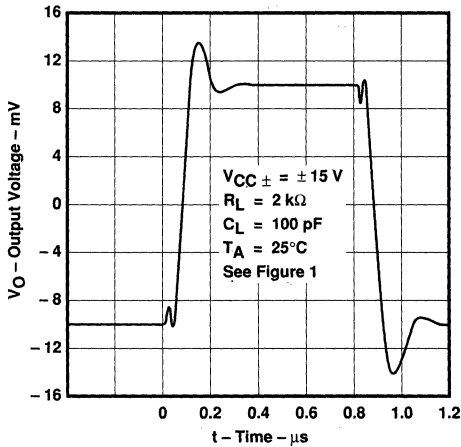


FIGURE 48

VOLTAGE-FOLLOWER
LARGE-SIGNAL
PULSE RESPONSE

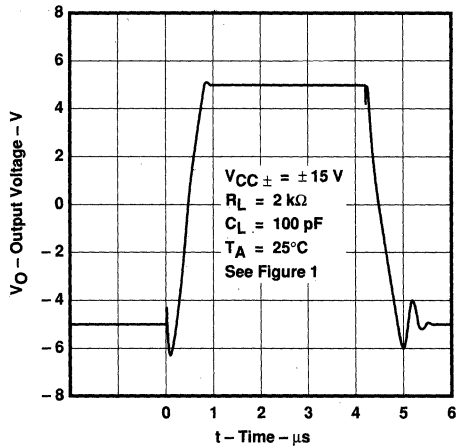


FIGURE 49

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.
NOTE 11: Values of phase margin below a load capacitance of 25 pF were estimated.

TYPICAL APPLICATION DATA

output characteristics

All operating characteristics (except bandwidth and phase margin) are specified with 100-pF load capacitance. The TL051 and TL051A will drive higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem. Capacitive loads of 1000 pF and larger may be driven if enough resistance is added in series with the output (see Figure 50).

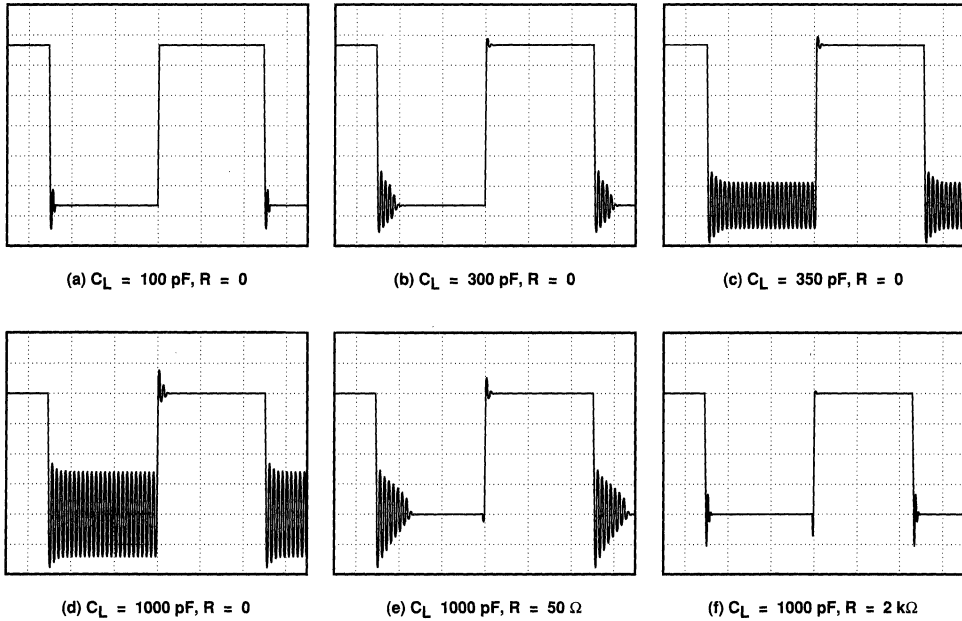
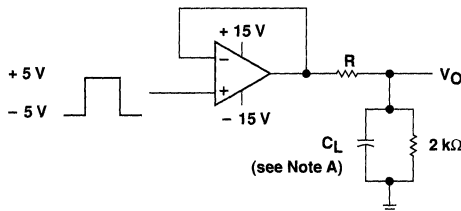


FIGURE 50. EFFECT OF CAPACITIVE LOADS



NOTE A: C_L includes fixture capacitance.

FIGURE 51. TEST CIRCUIT FOR OUTPUT CHARACTERISTICS

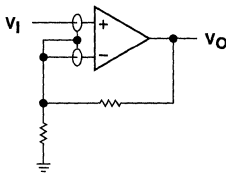
TL051, TL051A ENHANCED JFET PRECISION OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

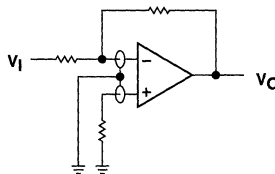
input characteristics

The TL051 and TL051A are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction.

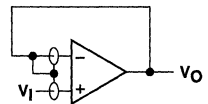
Because of the extremely high input impedance and resulting low bias current requirements, the TL051 and TL051A are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause degradation in system performance. It is good practice to include guard rings around inputs (see Figure 52). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.



(a) NONINVERTING AMPLIFIER



(b) INVERTING AMPLIFIER



(c) UNITY-GAIN AMPLIFIER

FIGURE 52. USE OF GUARD RINGS

noise performance

The noise specifications in op amp circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TL051 and TL051A result in a very low current noise. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than 50 k Ω .

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Operational Amplifiers

TYPICAL APPLICATION DATA

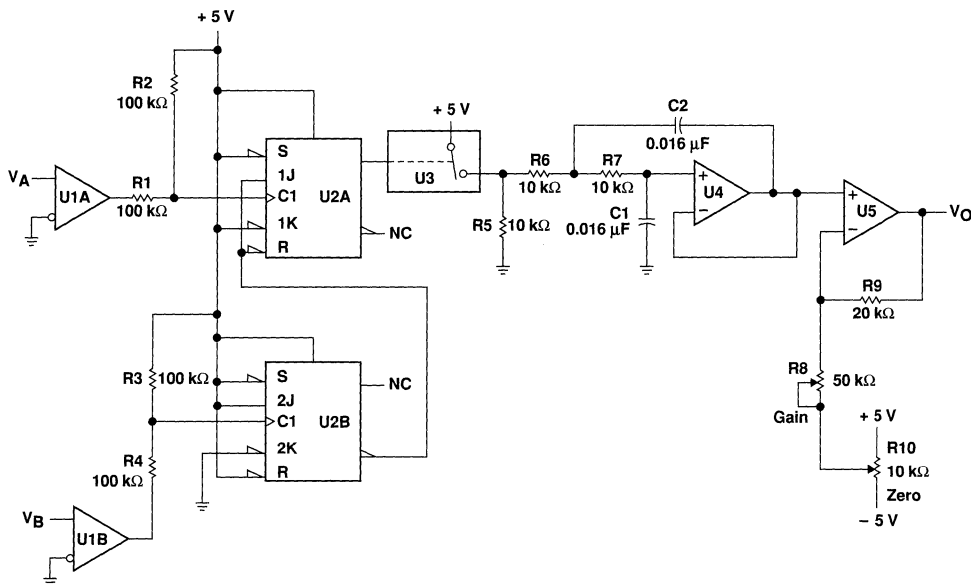
phase meter

The phase meter in Figure 53 produces an output voltage of 10 mV per degree of phase delay between the two input signals V_A and V_B . The reference signal V_A must be the same frequency as V_B . The TLC3702 comparators (U1) convert these two input sine waves into ± 5 -V square waves. Then R1 and R4 provide level shifting prior to the SN74HC109 dual J-K flip flop.

Flip-flop U2B is connected as a toggle flip-flop and generates a square wave at half the frequency of V_B . Flip-flop U2A also produces a square wave at half the input frequency. The pulse duration of U2A varies from zero to half the period, where zero corresponds to zero phase delay between V_A and V_B , and half the period corresponds to V_B lagging V_A by 360 degrees.

The output pulse from U2A causes the TLC4066 (U3) switch to charge the TL051 (U4) integrator capacitors C1 and C2. As the phase delay approaches 360 degrees, the output of U2A approximates a square wave, and U4 has an output of almost 2.5 V. U5 acts as a noninverting amplifier with a gain of 1.44 in order to scale the 0- to 2.5-V integrator output to a 0- to 3.6-V output range.

R8 and R10 provide output gain and zero-level calibration. This circuit operates over a 100-Hz to 10-kHz frequency range.



- NOTES: U1 = TLC3702; $V_{CC} \pm = \pm 5$ V.
 U2 = SN74HC109.
 U3 = TLC4066.
 U4, U5 = TL051; $V_{CC} \pm = \pm 5$ V.

FIGURE 53. PHASE METER

2
 Operational Amplifiers

TL051, TL051A ENHANCED JFET PRECISION OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

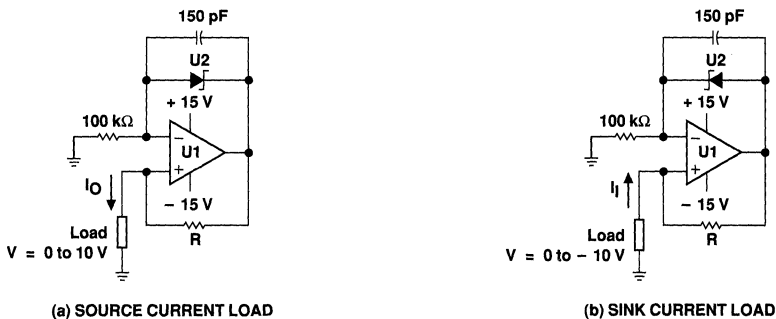
precision constant-current source over temperature

A precision current source benefits from the high input impedance and stability of Texas Instruments enhanced JFET process. A low-current shunt regulator maintains 2.5 V between the inverting input and the output of the TL051. The negative feedback then forces 2.5 V across the current setting resistor R; therefore, the current to the load is simply 2.5 V divided by R.

Possible choices for the shunt regulator include the LT1004, LT1009, and LM385. Note that if the regulator's cathode connects to the op amp output, this circuit will source load current. Similarly, if the cathode connects to the inverting input, the circuit will sink current from the load. To minimize output current change with temperature, R should be a metal film resistor with a low temperature coefficient. Also, this circuit must be operated with split voltage supplies.

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Operational Amplifiers



NOTES: U1 = TL051.
 U2 = LM385, LT1004, or LT1009 voltage reference.
 $I = \frac{2.5 \text{ V}}{R}$, R = Low temperature coefficient metal film resistor.

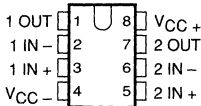
FIGURE 54. PRECISION CONSTANT-CURRENT SOURCE

TL052, TL052A ENHANCED JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

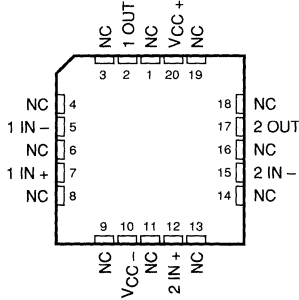
D3235, JUNE 1988 - REVISED FEBRUARY 1989

- Maximum Offset Voltage ... 800 μV (TL052A)
- High Slew Rate ... 17.8 $\text{V}/\mu\text{s}$ Typ at 25°C
- Low Total Harmonic Distortion ... 0.003% Typ at $R_L = 2 \text{ k}\Omega$
- Low Noise Voltage ... 19 $\text{nV}/\sqrt{\text{Hz}}$ Typ at $f = 1 \text{ kHz}$
- Low Input Bias Currents ... 30 pA Typ

D, JG, or P PACKAGE
(TOP VIEW)

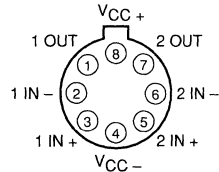


FK PACKAGE
(TOP VIEW)



NC - No internal connection

L PACKAGE
(TOP VIEW)



Pin 4 is in electrical contact with the case

description

The TL052 and TL052A dual operational amplifiers incorporate well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. These devices offer the significant advantages of Texas Instruments new enhanced JFET process. This process affords not only low initial offset voltage due to the on-chip zener trim capability but also stable offset voltage over time and temperature. In comparison, traditional JFET processes are plagued by significant offset voltage drift.

This new enhanced process still maintains the traditional JFET advantages of fast slew rates and low input bias and offset currents. These advantages coupled with low noise and low harmonic distortion make the TL052 well-suited for new state-of-the-art designs as well as existing design upgrades. The TL052 has been

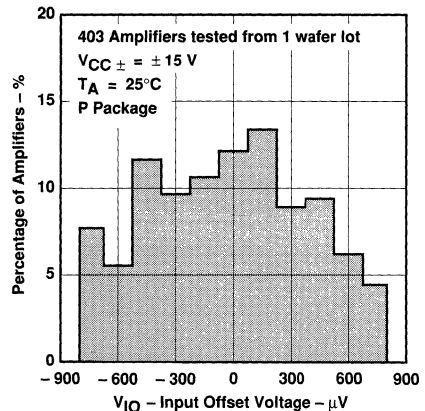
2
Operational Amplifiers

AVAILABLE OPTIONS

T_A	$V_{IO} \text{ max}$ AT 25°C	PACKAGE					
		SMALL- OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)	
0°C to 70°C	800 μV	TL052ACD	—	TL052ACJG	TL052ACL	TL052ACP	
	1500 μV	TL052CD	—	TL052CJG	TL052CL	TL052CP	
-40°C to 85°C	800 μV	TL052AID	—	TL052AJG	TL052AIL	TL052AIP	
	1500 μV	TL052ID	—	TL052IJG	TL052IL	TL052IP	
-55°C to 125°C	800 μV	TL052AMD	TL052AMFK	TL052AMJG	TL052AML	TL052AMP	
	1500 μV	TL052MD	TL052MFK	TL052MJG	TL052ML	TL052MP	

D packages are available taped-and-reeled. Add "R" suffix to device type (e.g., TL052CDR).

DISTRIBUTION OF TL052A
INPUT OFFSET VOLTAGE



PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS
INSTRUMENTS

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TL052, TL052A ENHANCED JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

description (continued)

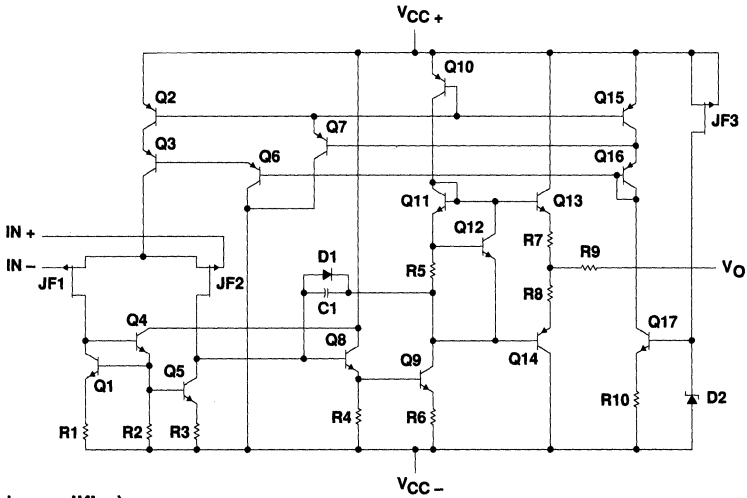
designed to be functionally compatible, as well as pin compatible, with the TL072 and TL082. Two offset voltage grades are available: TL052 (1.5 mV max) and TL052A (800 μ V max).

A variety of available packaging options includes small-outline and chip carrier versions for high-density system applications.

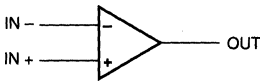
The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I-suffix devices are characterized for operation from -40°C to 85°C , and the C-suffix devices are characterized for operation from 0°C to 70°C .

2

equivalent schematic (each amplifier)



symbol (each amplifier)



Operational Amplifiers

TL052, TL052A ENHANCED JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

2

Operational Amplifiers

absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	18 V
Supply voltage, V_{CC-} (see Note 1)	- 18 V
Differential input voltage (see Note 2)	± 30 V
Input voltage range, V_I (any input, see Notes 1 and 3)	± 15 V
Input current, I_I (each input)	± 1 mA
Output current, I_O (each output)	± 80 mA
Total current into V_{CC+} terminal	160 mA
Total current out of V_{CC-} terminal	160 mA
Duration of short-circuit current at (or below) 25°C (see Note 4)	Unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, T_A : M-suffix	- 55°C to 125°C
I-suffix	- 40°C to 85°C
C-suffix	0°C to 70°C
Storage temperature range	- 65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG or L package	300°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
2. Differential voltages are at the noninverting input with respect to the inverting input.
3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
4. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
L	825 mW	6.6 mW/°C	528 mW	429 mW	165 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW

recommended operating conditions

		M-SUFFIX		I-SUFFIX		C-SUFFIX		UNIT
		MIN	NOM MAX	MIN	NOM MAX	MIN	NOM MAX	
Supply voltage, V_{CC}		± 5	± 15	± 5	± 15	± 5	± 15	V
Common-mode input voltage, V_{IC}	$V_{CC} \pm \pm 5$ V	- 1	4	- 1	4	- 1	4	V
	$V_{CC} \pm \pm 15$ V	- 11	11	- 11	11	- 11	11	
Operating free-air temperature, T_A		- 55	125	- 40	85	0	70	°C

TL052M, TL052AM ENHANCED JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL052M	25°C	0.73		3.5		0.65		1.5
			Full range			6.5		4.5		
		TL052AM	25°C	0.51		2.8		0.4		0.8
			Full range			5.8		3.8		
α _{VIO} Temperature coefficient of input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL052M	25°C to 125°C		10		9		μV/°C	
		TL052AM	25°C to 125°C		9		8			
Input offset voltage long-term drift (see Note 5)		25°C	0.04		0.04				μV/mo	
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	4		100		5		100	pA
		125°C	1		20		2			
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	20		200		30		200	pA
		125°C	10		50		20			
V _{ICR} Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6	-11 to 11		-12.3 to 15.6		V	
		Full range	-1 to 4		-11 to 11					
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3		4.2		13		13.9	V
		Full range	3				13			
	R _L = 2 kΩ	25°C	2.5		3.8		11.5		12.7	
		Full range	2.5				11.5			
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-2.5		-3.5		-12		-13.2	V
		Full range	-2.5				-12			
	R _L = 2 kΩ	25°C	-2.3		-3.2		-11		-12	
		Full range	-2.3				-11			
A _{VD} Large-signal differential voltage amplification	R _L = 2 kΩ, See Note 6	25°C	25		59		50		105	V/mV
		-55°C	30		76		60			
		125°C	10		32		15			
r _i Input resistance		25°C	10 ¹²			10 ¹²			Ω	
C _i Input capacitance		25°C	10			12			pF	
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	65		85		75		93	dB
		-55°C	65		83		75			
		125°C	65		84		75			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75		99		75		99	dB
		-55°C	75		98		75			
		125°C	75		100		75			
I _{CC} Supply current (four amplifiers)	No load, V _O = 0	25°C	4.6		5.6		4.8		5.6	mA
		-55°C	4.4		6.4		4.5			
		125°C	4.2		6.4		4.4			
V _{O1} /V _{O2} Crosstalk attenuation	A _{VD} = 100	25°C	120			120			dB	

[†] Full range is -55°C to 125°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. For V_{CC} ± = ± 5 V, V_O = ± 2.3 V; at V_{CC} ± = ± 15 V, V_O = ± 10 V.

2

Operational Amplifiers

TL052M, TL052AM ENHANCED JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR + Positive slew rate at unity gain	R _L = 2 kΩ, C _L = 100 pF, See Figure 1 and Note 7	25°C		17.8		13	20.7	V/μs	
		-55°C		18.8		20.3			
		125°C		14.5		20.2			
SR - Negative slew rate at unity gain		25°C		15.4		13	17.8		
		-55°C		15.7		17.6			
		125°C		13.8		16.5			
t _r Rise time	V _{Ipp} = ± 10 mV, R _L = 2 kΩ, C _L = 100 pF, See Figures 1 and 2	25°C		55		56	ns		
		-55°C		51		52			
		125°C		68		68			
t _f Fall time		25°C		55		57			
		-55°C		51		52			
		125°C		68		69			
Overshoot factor		25°C		24%		19%			
		-55°C		25%		19%			
		125°C		25%		19%			
V _n Equivalent input noise voltage		R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C	71			71	nV/√Hz
			f = 1 kHz	25°C	19			19	
V _{NPP} Peak-to-peak equivalent input noise voltage		f = 10 Hz to 10 kHz	25°C		4			4	μV
I _n Equivalent input noise current	f = 1 kHz	25°C		0.01		0.01	pA/√Hz		
THD Total harmonic distortion	R _S = 1 kΩ, R _L = 2 kΩ, f = 1 kHz, See Note 8	25°C		0.003%		0.003%			
B ₁ Unity-gain bandwidth	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C		3		3	MHz		
		-55°C		3.6		3.7			
		125°C		2.3		2.4			
φ _m Phase margin at unity gain	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C		60°		63°			
		-55°C		57°		61°			
		125°C		60°		63°			

[†] Full range is -55°C to 125°C.

NOTES: 7. For V_{CC} ± = ± 5 V, V_{Ipp} = ± 1 V; for V_{CC} ± = ± 15 V, V_{Ipp} = ± 5 V.

8. For V_{CC} ± = ± 5 V, V_{0(rms)} = 1 V; for V_{CC} ± = ± 15 V, V_{0(rms)} = 6 V.

2

Operational Amplifiers

TL052I, TL052AI

ENHANCED JFET PRECISION

DUAL OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL052I	25°C	0.73	3.5	0.65	1.5	mV	
			Full range	5.3			3.3		
		TL052AI	25°C	0.51	2.8	0.4	0.8		
			Full range	4.6			2.6		
α _{VIO} Temperature coefficient of input offset voltage (see Note 9)	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL052I	25°C to 85°C	7		6		μV/°C	
		TL052AI	25°C to 85°C	6		6	25		
Input offset voltage long-term drift (see Note 5)		25°C	0.04		0.04		μV/mo		
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	4		100	5	100	pA	
		85°C	0.06		10	0.07	10	nA	
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	20		200	30	200	pA	
		85°C	0.6		20	0.7	20	nA	
V _{ICR} Common-mode input voltage range		25°C	-1	-2.3	-11	-12.3	V		
			to	to	to	to			
			4	5.6	11	15.6			
		Full range	-1		-11				
			to		to				
			4		11				
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.2	13	13.9	V		
		Full range	3			13			
	R _L = 2 kΩ	25°C	2.5	3.8	11.5	12.7			
		Full range	2.5			11.5			
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-2.5	-3.5	-12	-13.2	V		
		Full range	-2.5			-12			
	R _L = 2 kΩ	25°C	-2.3	-3.2	-11	-12			
		Full range	-2.3			-11			
A _{VD} Large-signal differential voltage amplification	R _L = 2 kΩ, See Note 6	25°C	25	59	50	105	V/mV		
		-40°C	30	74	60	145			
		85°C	20	43	30	76			
r _i Input resistance		25°C	10 ¹²			10 ¹²	Ω		
C _i Input capacitance		25°C	10			12	pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	65	85	75	93	dB		
		-40°C	65	83	75	90			
		85°C	65	84	75	93			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75	99	75	99	dB		
		-40°C	75	98	75	98			
		85°C	75	99	75	99			
I _{CC} Supply current (four amplifiers)	No load, V _O = 0	25°C	4.6	5.6	4.8	5.6	mA		
		-40°C	4.5	6.4	4.7	6.4			
		85°C	4.4	6.4	4.6	6.4			
V _{O1} /V _{O2} Crosstalk attenuation	A _{VD} = 100	25°C	120		120		dB		

[†] Full range is -40°C to 85°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. For V_{CC} ± = ± 5 V, V_O = ± 2.3 V; at V_{CC} ± = ± 15 V, V_O = ± 10 V.

9. This parameter is tested on a sample basis for the TL052A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

TL052I, TL052AI
ENHANCED JFET PRECISION
DUAL OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER	TEST CONDITIONS		T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
SR + Positive slew rate at unity gain	R _L = 2 kΩ, C _L = 100 pF, See Figure 1 and Note 7		25°C		17.8		13	20.7	V/μs	
			-40°C		18.8		11	20.6		
			85°C		16		11	20.7		
SR - Negative slew rate at unity gain			25°C		15.4		13	17.8		
			-40°C		16		11	17.8		
			85°C		14.5		11	17.2		
t _r Rise time	V _I PP = ± 10 mV, R _L = 2 kΩ, C _L = 100 pF, See Figures 1 and 2		25°C		55		56	ns		
t _f Fall time			-40°C		52		53			
			85°C		64		65			
			25°C		55		57			
Overshoot factor			-40°C		51		53			
			85°C		64		65			
	25°C		24%		19%					
V _n Equivalent input noise voltage (see Note 10)	R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C		71		71	nV/√Hz		
		f = 1 kHz	25°C		19		19		30	
		f = 10 Hz to 10 kHz	25°C		4		4			
V _{NPP} Peak-to-peak equivalent input noise voltage			25°C		4		4	μV		
I _n Equivalent input noise current	f = 1 kHz		25°C		0.01		0.01	pA/√Hz		
THD Total harmonic distortion	R _S = 1 kΩ, R _L = 2 kΩ, f = 1 kHz, See Note 8	25°C			0.003%		0.003%			
B ₁ Unity-gain bandwidth	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4		25°C		3		3	MHz		
			-40°C		3.5		3.6			
			85°C		2.5		2.6			
φ _m Phase margin at unity gain	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4		25°C		60°		63°			
			-40°C		58°		61°			
			85°C		60°		63°			

[†] Full range is -40°C to 85°C.

NOTES: 7. For V_{CC} ± = ± 5 V, V_IPP = ± 1 V; for V_{CC} ± = ± 15 V, V_IPP = ± 5 V.

8. For V_{CC} ± = ± 5 V, V_o(rms) = 1 V; for V_{CC} ± = ± 15 V, V_o(rms) = 6 V.

10. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

TL052C, TL052AC ENHANCED JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL052C	25°C	0.73	3.5	0.65	1.5	mV	
			Full range	4.5		2.5			
		TL052AC	25°C	0.51	2.8	0.4	0.8		
			Full range	3.8		1.8			
α _{VIO} Temperature coefficient of input offset voltage (see Note 9)	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL052C	25°C to 70°C	8		8		μV/°C	
		TL052AC	25°C to 70°C	8		6	25		
Input offset voltage long-term drift (see Note 5)		25°C	0.04		0.04		μV/mo		
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	4	100	5	100	pA		
		70°C	0.02	1	0.025	1	nA		
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	20	200	30	200	pA		
		70°C	0.15	4	0.2	4	nA		
V _{ICR} Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	V		
		Full range	-1 to 4		-11 to 11				
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.2	13	13.9	V		
		Full range	3		13				
	R _L = 2 kΩ	25°C	2.5	3.8	11.5	12.7			
		Full range	2.5		11.5				
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-2.5	-3.5	-12	-13.2	V		
		Full range	-2.5		-12				
	R _L = 2 kΩ	25°C	-2.3	-3.2	-11	-12			
		Full range	-2.3		-11				
A _{VD} Large-signal differential voltage amplification	R _L = 2 kΩ, See Note 6	25°C	25	59	50	105	V/mV		
		0°C	30	65	60	129			
		70°C	20	46	30	85			
r _i Input resistance		25°C	10 ¹²		10 ¹²		Ω		
C _i Input capacitance		25°C	10		12		pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	65	85	75	93	dB		
		0°C	65	84	75	92			
		70°C	65	84	75	91			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75	99	75	99	dB		
		0°C	75	98	75	98			
		70°C	75	97	75	97			
I _{CC} Supply current (four amplifiers)	No load, V _O = 0	25°C	4.6	5.6	4.8	5.6	mA		
		0°C	4.7	6.4	4.8	6.4			
		70°C	4.4	6.4	4.6	6.4			
V _{O1} /V _{O2} Crosstalk attenuation	A _{VD} = 100	25°C	120		120		dB		

[†] Full range is 0°C to 70°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. For V_{CC} ± = ± 5 V, V_O = ± 2.3 V; at V_{CC} ± = ± 15 V, V_O = ± 10 V.

9. This parameter is tested on a sample basis for the TL052A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

TL052C, TL052AC ENHANCED JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR + Positive slew rate at unity gain	R _L = 2 kΩ, C _L = 100 pF, See Figure 1 and Note 7	25°C	17.8			13 20.7			V/μs
		0°C	18.5			11 20.9			
		70°C	16.5			11 20.8			
SR - Negative slew rate at unity gain		25°C	15.4			13 17.8			
		0°C	15.7			11 18.5			
		70°C	14.7			11 16.5			
t _r Rise time	V _I PP = ± 10 mV, R _L = 2 kΩ, C _L = 100 pF, See Figures 1 and 2	25°C	55			56			ns
		0°C	54			55			
		70°C	63			63			
t _f Fall time		25°C	55			57			
		0°C	54			56			
		70°C	62			64			
Overshoot factor		25°C	24%			19%			
		0°C	24%			19%			
		70°C	24%			19%			
V _n Equivalent input noise voltage (see Note 10)		R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C			71		
	f = 1 kHz		25°C			19 30			
V _{NPP} Peak-to-peak equivalent input noise voltage		f = 10 Hz to 10 kHz	25°C			4			μV
I _n Equivalent input noise current	f = 1 kHz	25°C	0.01			0.01			pA/√Hz
THD Total harmonic distortion	R _S = 1 kΩ, R _L = 2 kΩ, f = 1 kHz, See Note 8	25°C	0.003%			0.003%			
B ₁ Unity-gain bandwidth	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C	3			3			MHz
		0°C	3.2			3.2			
		70°C	2.6			2.7			
φ _m Phase margin at unity gain	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C	60°			63°			
		0°C	59°			63°			
		70°C	60°			63°			

[†] Full range is 0°C to 70°C.

NOTES: 7. For V_{CC} ± = ± 5 V, V_IPP = ± 1 V; for V_{CC} ± = ± 15 V, V_IPP = ± 5 V.

8. For V_{CC} ± = ± 5 V, V_O(rms) = 1 V; for V_{CC} ± = ± 15 V, V_O(rms) = 6 V.

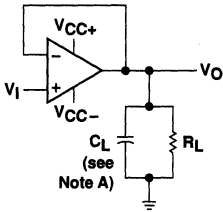
10. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

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Operational Amplifiers



NOTE A: C_L includes fixture capacitance.

FIGURE 1. SLEW RATE, RISE/FALL TIME, AND OVERSHOOT TEST CIRCUIT

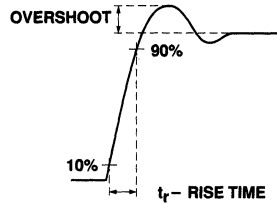


FIGURE 2. RISE TIME AND OVERSHOOT WAVEFORM

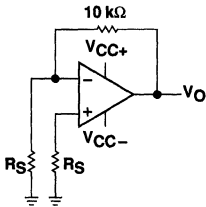
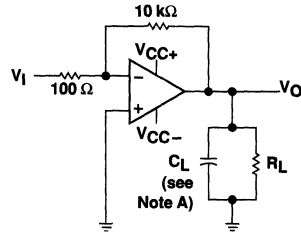


FIGURE 3. NOISE VOLTAGE TEST CIRCUIT



NOTE A: C_L includes fixture capacitance.

FIGURE 4. UNITY-GAIN BANDWIDTH AND PHASE MARGIN TEST CIRCUIT

typical values

Typical values as presented in this data sheet represent the median (50% point) of device parametric performance.

input bias and offset current

At the picoamp-bias-current level typical of the TL052 and TL052A, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To accurately measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied, but with no device in the socket. The device is then inserted in the socket and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

noise

Because of the increasing emphasis on low noise levels in many of today's applications, the input noise voltage density is sample-tested at $f = 1$ kHz. Texas Instruments also has additional noise testing capability to meet specific application requirements. Please contact the factory for details.

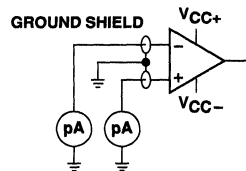


FIGURE 5. INPUT BIAS AND OFFSET CURRENT TEST CIRCUIT

TL052, TL052A
ENHANCED JFET PRECISION
DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

table of graphs

		FIGURE	
V_{IO}	Input offset voltage	Distribution	6
α_{VIO}	Temperature coefficient of input offset voltage	Distribution	7
I_{IO}	Input offset current	vs Temperature	8
I_{IB}	Input bias current	vs V_{IC}	9
		vs Temperature	8
V_I	Input voltage range	vs V_{CC}	10
		vs Temperature	11
V_O	Output voltage	vs Differential input voltage	12, 13
		vs V_{CC}	14
V_{OM}	Maximum peak output voltage swing	vs Output current	18, 19
		vs Frequency	15, 16, 17
		vs Temperature	20, 21
		vs R_L	22
A_{VD}	Differential voltage amplification	vs Frequency	23
		vs Temperature	24, 25
		vs Frequency	29
z_o	Output impedance	vs Frequency	29
CMRR	Common-mode rejection ratio	vs Frequency	26, 27
		vs Temperature	28
k_{SVR}	Supply-voltage rejection ratio	vs Temperature	30
I_{OS}	Short-circuit output current	vs V_{CC}	31
		vs Time	32
		vs Temperature	33
I_{CC}	Supply current	vs V_{CC}	34
		vs Temperature	35
SR	Slew rate	vs R_L	36, 37
		vs Temperature	38, 39
	Overshoot factor	vs C_L	40
V_n	Equivalent input noise voltage	vs Frequency	41
THD	Total harmonic distortion	vs Frequency	42
B_1	Unity-gain bandwidth	vs V_{CC}	43
		vs Temperature	44
ϕ_m	Phase margin	vs V_{CC}	45
		vs C_L	46
		vs Temperature	47
	Phase shift	vs Frequency	23
	Pulse response	Small-signal	48
		Large-signal	49

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Operational Amplifiers

TL052, TL052A ENHANCED JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

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Operational Amplifiers

DISTRIBUTION OF TL052
INPUT OFFSET VOLTAGE

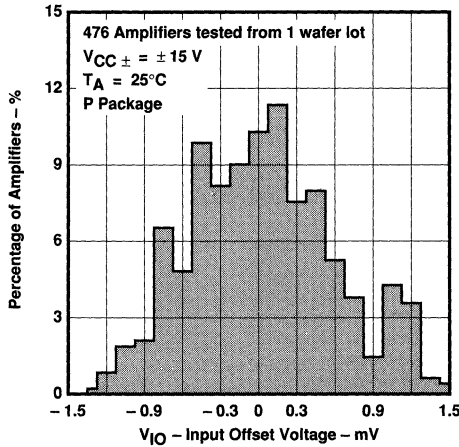


FIGURE 6

DISTRIBUTION OF TL052
INPUT OFFSET VOLTAGE
TEMPERATURE COEFFICIENT

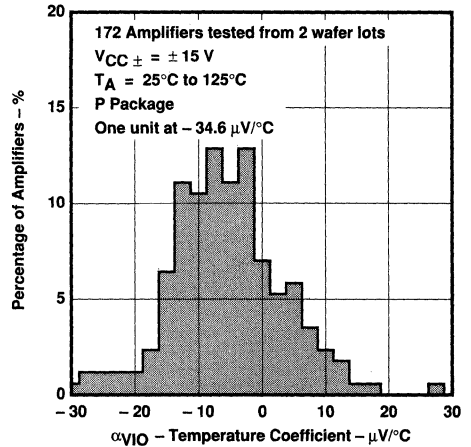


FIGURE 7

INPUT BIAS CURRENT AND
INPUT OFFSET CURRENT
VS
FREE-AIR TEMPERATURE

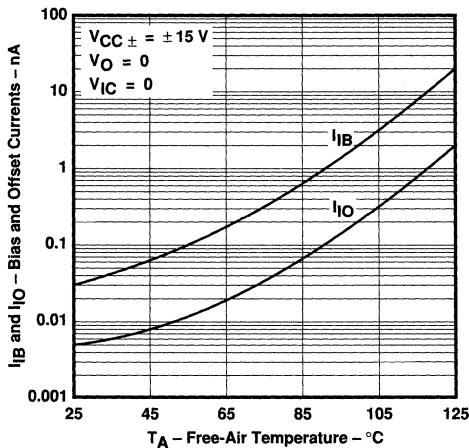


FIGURE 8

INPUT BIAS CURRENT
VS
COMMON-MODE INPUT VOLTAGE

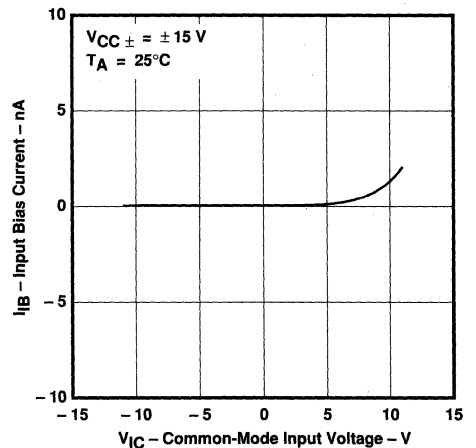


FIGURE 9

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

**COMMON-MODE
 INPUT VOLTAGE RANGE LIMITS
 VS
 SUPPLY VOLTAGE**

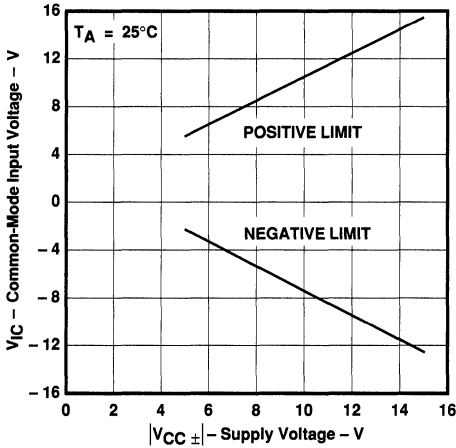


FIGURE 10

**COMMON-MODE
 INPUT VOLTAGE RANGE LIMITS
 VS
 FREE-AIR TEMPERATURE**

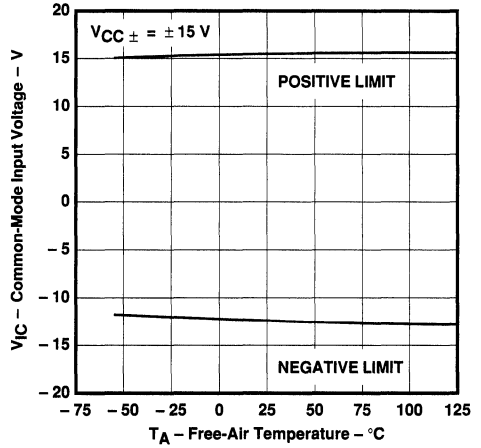


FIGURE 11

**OUTPUT VOLTAGE
 VS
 DIFFERENTIAL INPUT VOLTAGE**

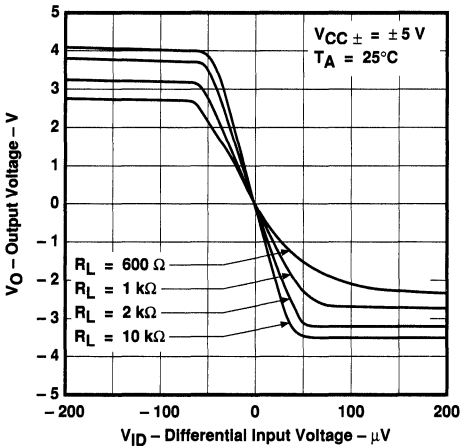


FIGURE 12

**OUTPUT VOLTAGE
 VS
 DIFFERENTIAL INPUT VOLTAGE**

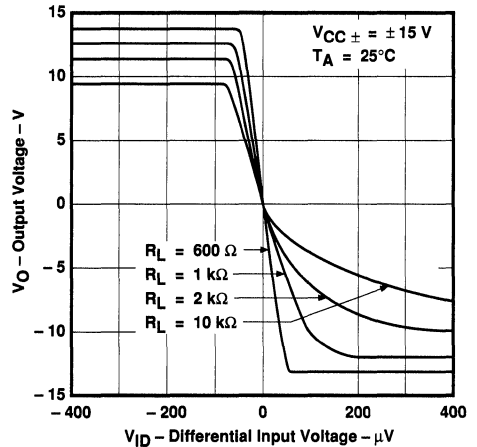


FIGURE 13

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Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL052, TL052A ENHANCED JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE
VS
SUPPLY VOLTAGE

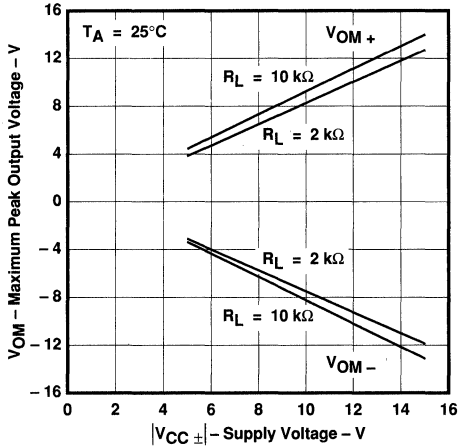


FIGURE 14

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
VS
FREQUENCY

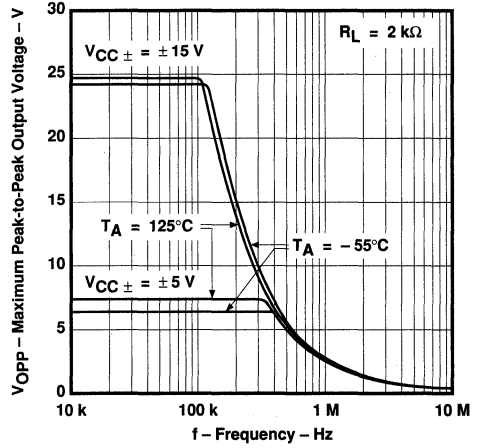


FIGURE 15

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
VS
FREQUENCY

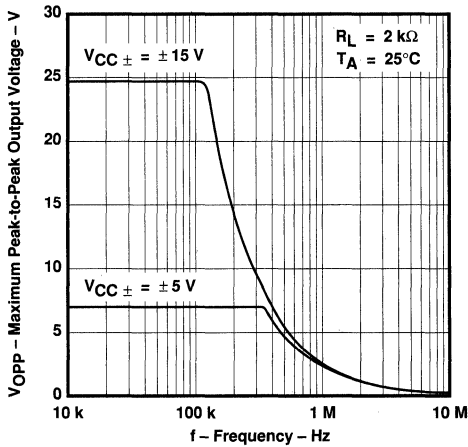


FIGURE 16

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
VS
FREQUENCY

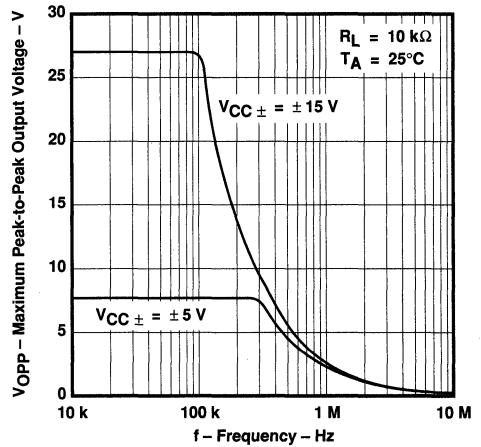


FIGURE 17

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL052, TL052A
ENHANCED JFET PRECISION
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TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 OUTPUT CURRENT

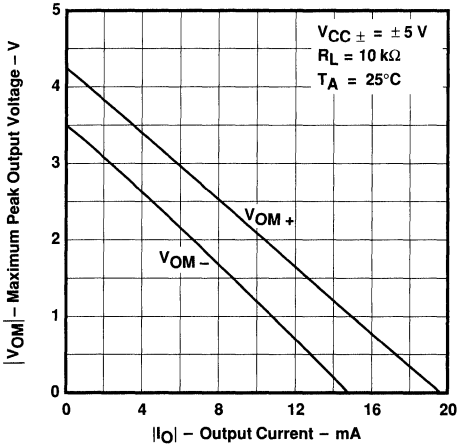


FIGURE 18

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 OUTPUT CURRENT

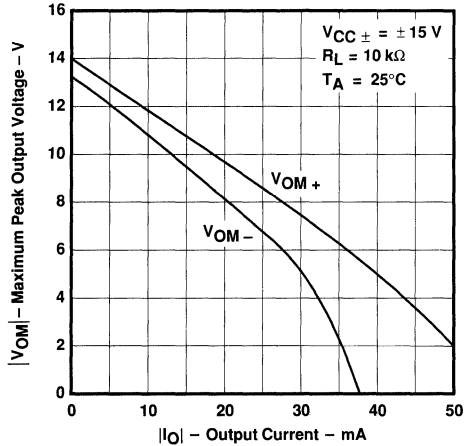


FIGURE 19

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

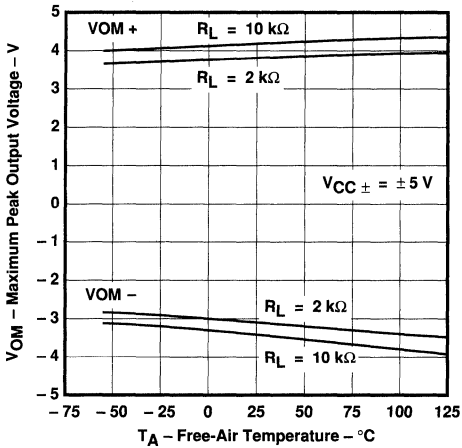


FIGURE 20

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

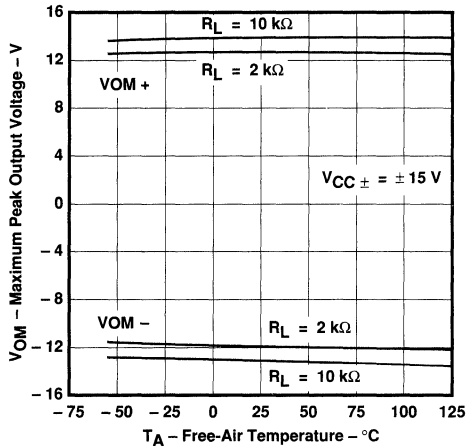


FIGURE 21

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Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL052, TL052A
ENHANCED JFET PRECISION
DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

LARGE-SIGNAL VOLTAGE AMPLIFICATION
VS
LOAD RESISTANCE

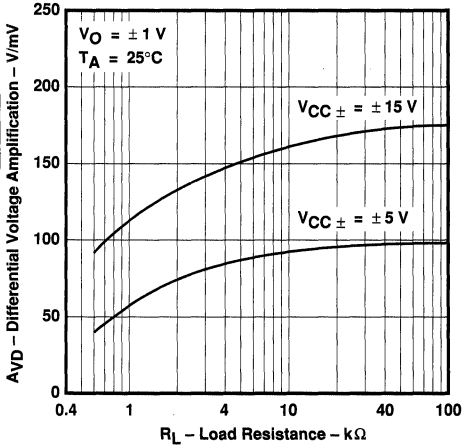


FIGURE 22

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
VS
FREQUENCY

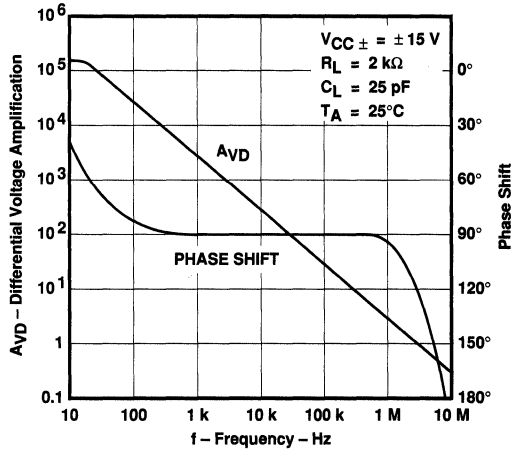


FIGURE 23

LARGE-SIGNAL VOLTAGE AMPLIFICATION
VS
FREE-AIR TEMPERATURE

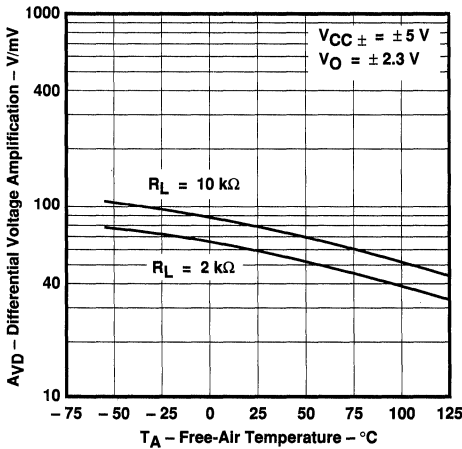


FIGURE 24

LARGE-SIGNAL VOLTAGE AMPLIFICATION
VS
FREE-AIR TEMPERATURE

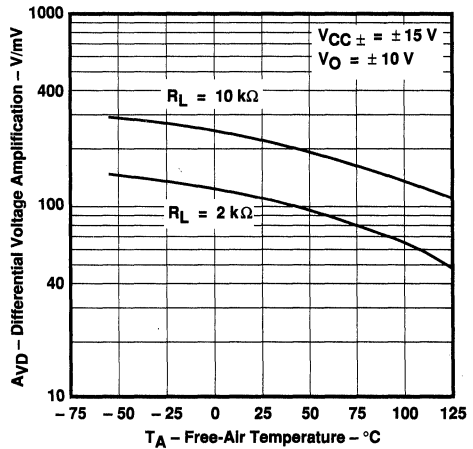


FIGURE 25

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

2
Operational Amplifiers

TL052, TL052A ENHANCED JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

COMMON-MODE REJECTION RATIO
vs
FREQUENCY

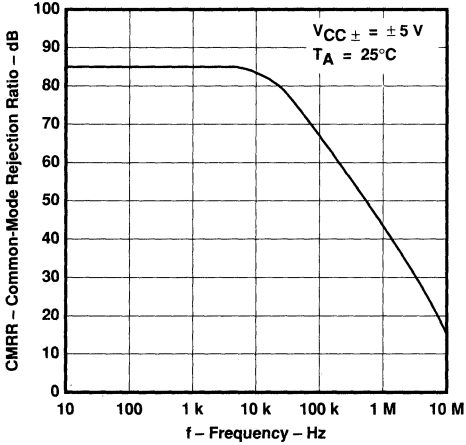


FIGURE 26

COMMON-MODE REJECTION RATIO
vs
FREQUENCY

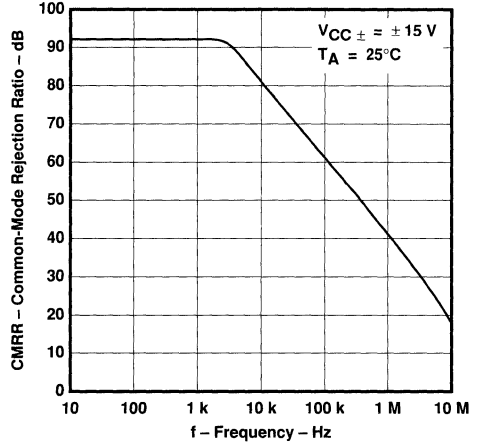


FIGURE 27

COMMON-MODE REJECTION RATIO
vs
FREE-AIR TEMPERATURE

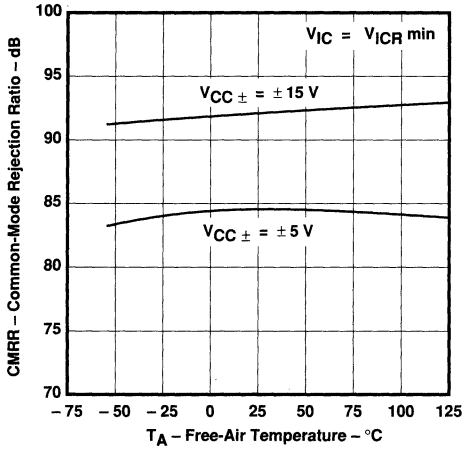


FIGURE 28

OUTPUT IMPEDANCE
vs
FREQUENCY

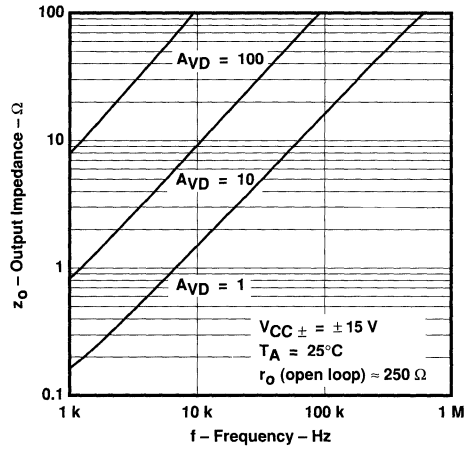


FIGURE 29

2
Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL052, TL052A
ENHANCED JFET PRECISION
DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

SUPPLY-VOLTAGE REJECTION RATIO
vs
FREE-AIR TEMPERATURE

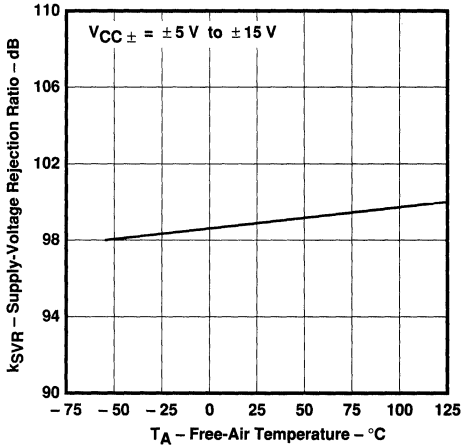


FIGURE 30

SHORT-CIRCUIT OUTPUT CURRENT
vs
SUPPLY VOLTAGE

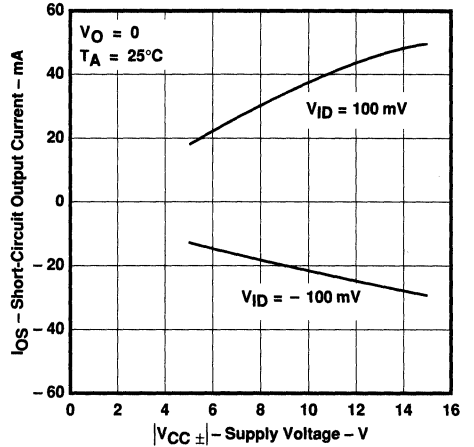


FIGURE 31

SHORT-CIRCUIT OUTPUT CURRENT
vs
TIME

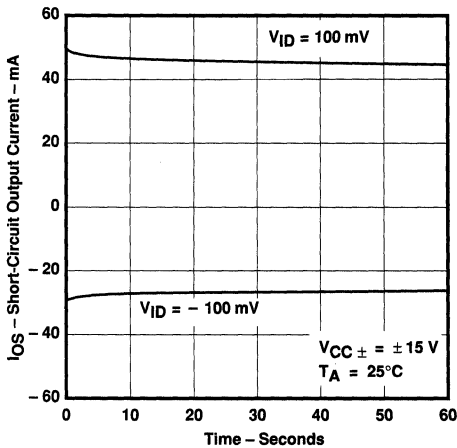


FIGURE 32

SHORT-CIRCUIT OUTPUT CURRENT
vs
FREE-AIR TEMPERATURE

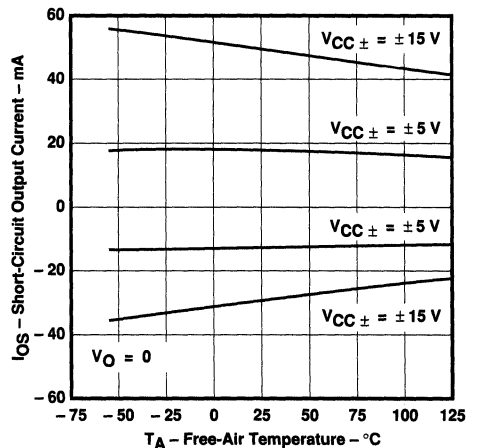


FIGURE 33

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL052, TL052A ENHANCED JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

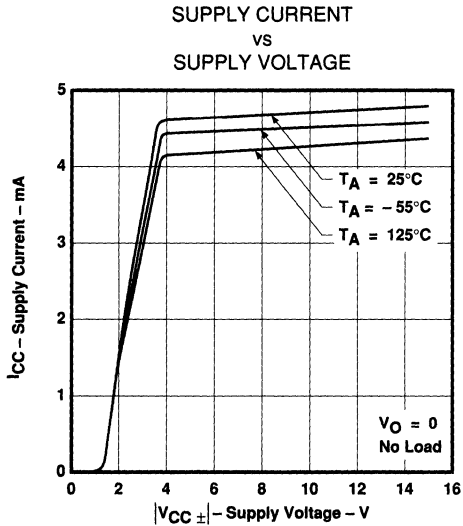


FIGURE 34

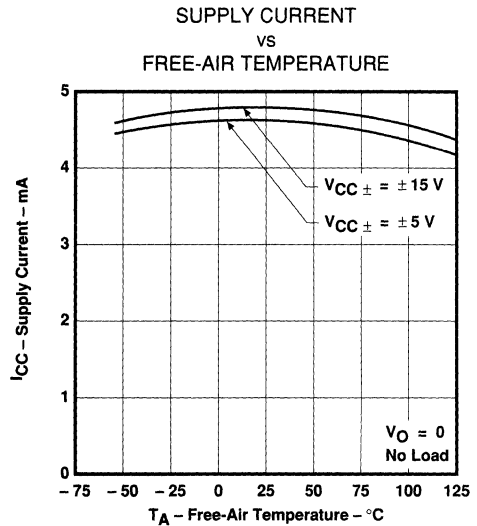


FIGURE 35

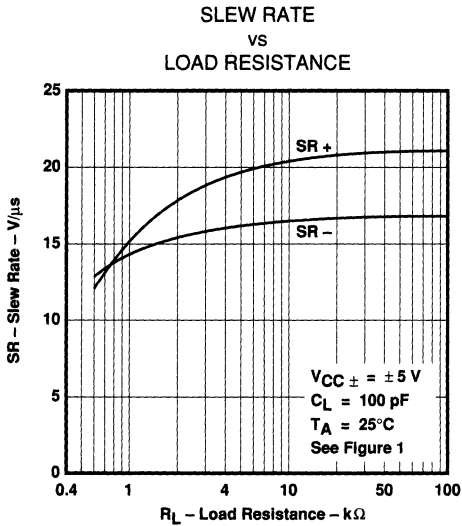


FIGURE 36

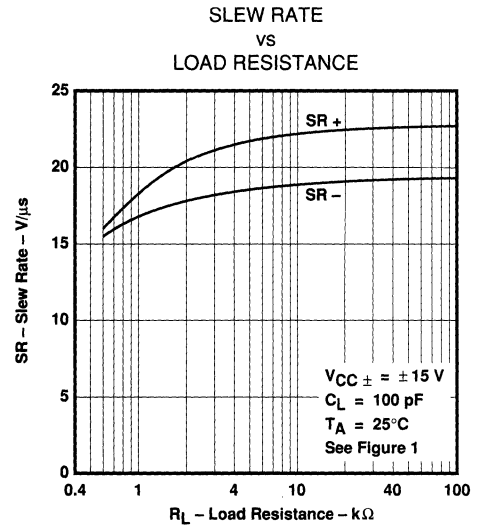


FIGURE 37

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

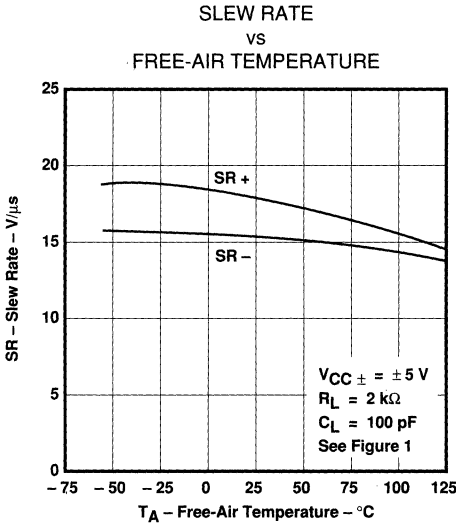


FIGURE 38

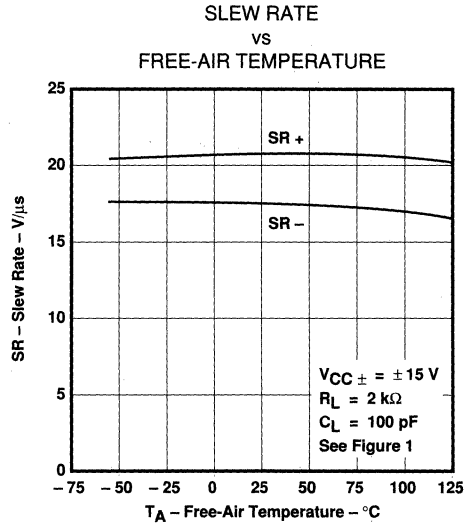


FIGURE 39

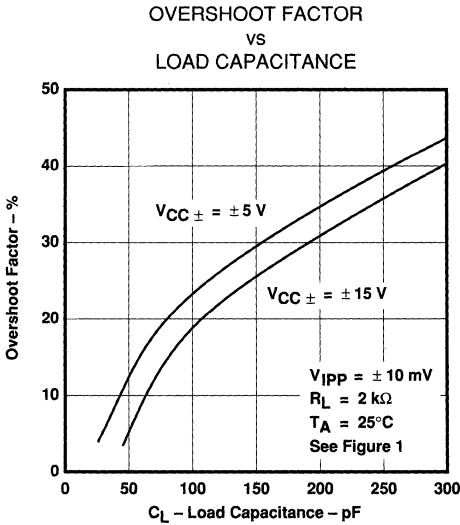


FIGURE 40

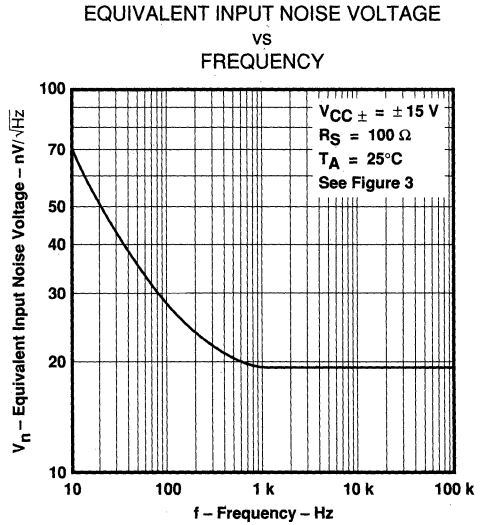


FIGURE 41

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL052, TL052A
ENHANCED JFET PRECISION
DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

TOTAL HARMONIC DISTORTION
VS
FREQUENCY

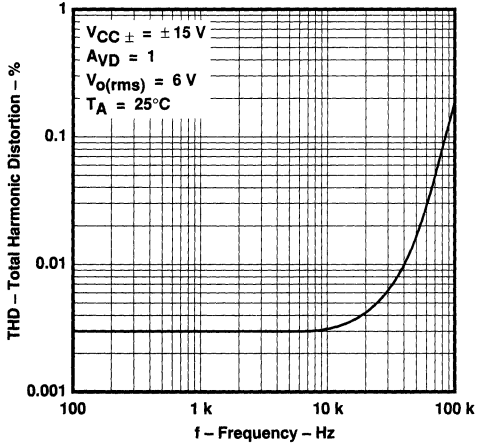


FIGURE 42

UNITY-GAIN BANDWIDTH
VS
SUPPLY VOLTAGE

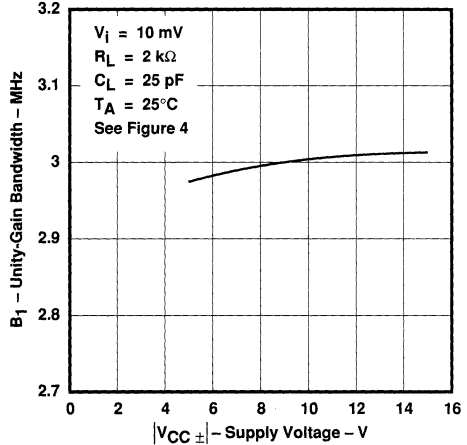


FIGURE 43

UNITY-GAIN BANDWIDTH
VS
FREE-AIR TEMPERATURE

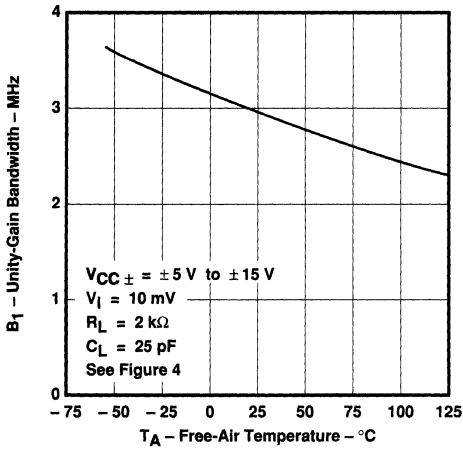


FIGURE 44

PHASE MARGIN
VS
SUPPLY VOLTAGE

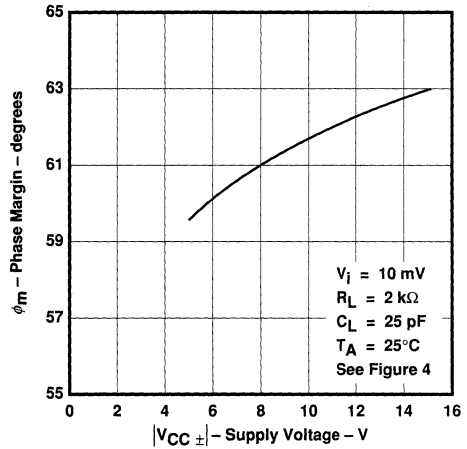


FIGURE 45

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL052, TL052A
ENHANCED JFET PRECISION
DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

2
Operational Amplifiers

PHASE MARGIN
vs
LOAD CAPACITANCE

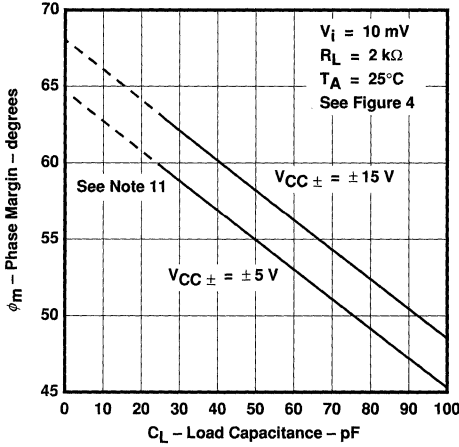


FIGURE 46

PHASE MARGIN
vs
FREE-AIR TEMPERATURE

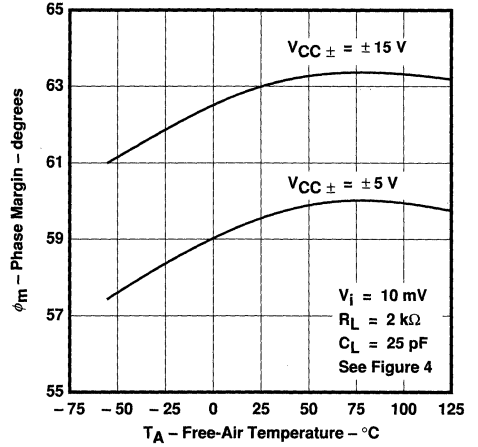


FIGURE 47

VOLTAGE-FOLLOWER
SMALL-SIGNAL
PULSE RESPONSE

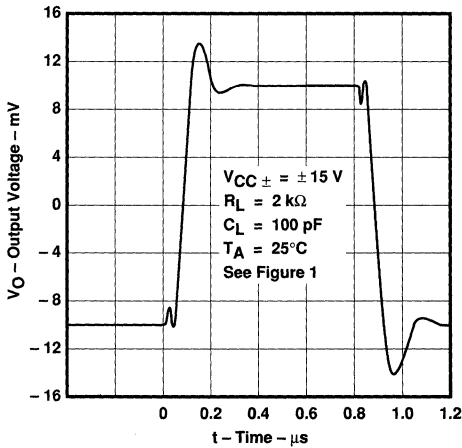


FIGURE 48

VOLTAGE-FOLLOWER
LARGE-SIGNAL
PULSE RESPONSE

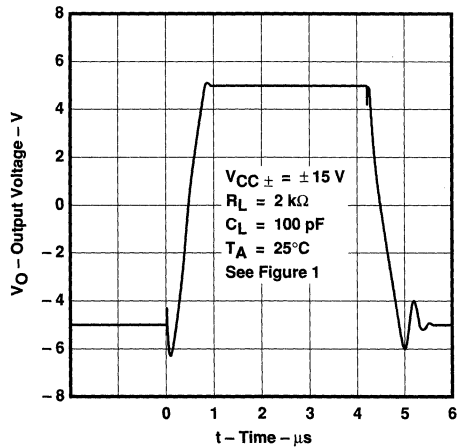


FIGURE 49

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.
 NOTE 11: Values of phase margin below a load capacitance of 25 pF were estimated.

TYPICAL APPLICATION DATA

output characteristics

All operating characteristics (except bandwidth and phase margin) are specified with 100-pF load capacitance. The TL052 and TL052A will drive higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem. Capacitive loads of 1000 pF and larger may be driven if enough resistance is added in series with the output (see Figure 50).

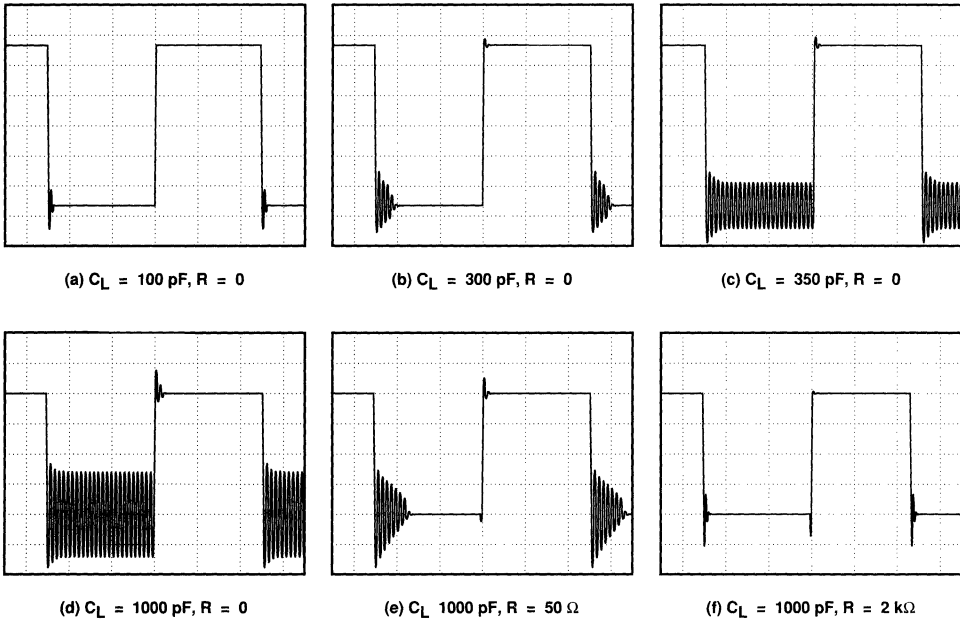
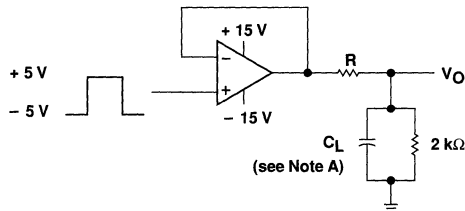


FIGURE 50. EFFECT OF CAPACITIVE LOADS



NOTE A: C_L includes fixture capacitance.

FIGURE 51. TEST CIRCUIT FOR OUTPUT CHARACTERISTICS

TL052, TL052A ENHANCED JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

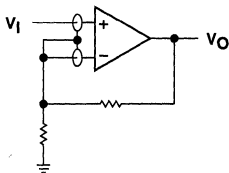
TYPICAL APPLICATION DATA

input characteristics

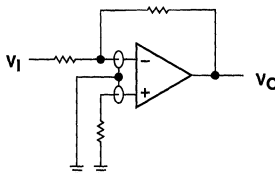
The TL052 and TL052A are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction.

Because of the extremely high input impedance and resulting low bias current requirements, the TL052 and TL052A are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause degradation in system performance. It is good practice to include guard rings around inputs (see Figure 52). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.

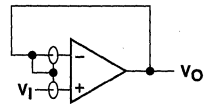
The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.



(a) NONINVERTING AMPLIFIER



(b) INVERTING AMPLIFIER



(c) UNITY-GAIN AMPLIFIER

FIGURE 52. USE OF GUARD RINGS

noise performance

The noise specifications in op amp circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TL052 and TL052A result in a very low current noise. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than 50 k Ω .

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Operational Amplifiers

TYPICAL APPLICATION DATA

instrumentation amplifier with adjustable gain/null

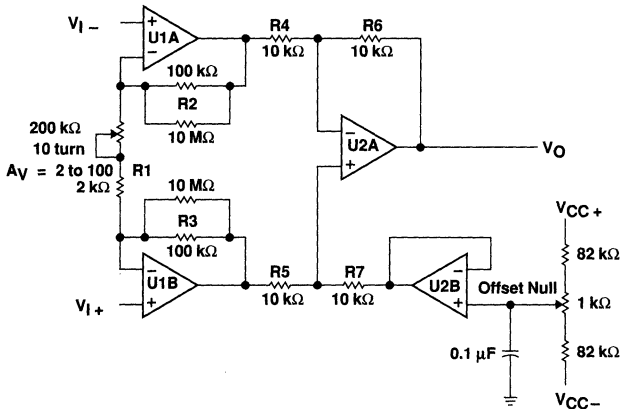
The instrumentation amplifier in Figure 53 benefits greatly from the high input impedance and stable input offset voltage of the TL052A. Amplifiers U1A, U1B, and U2A form the actual instrumentation amplifier, while U2B provides offset null. Potentiometer R1 provides gain adjust. With $R1 = 2 \text{ k}\Omega$, the circuit gain equals 100, while with $R1 = 200 \text{ k}\Omega$, the circuit gain equals two. The following equation shows the instrumentation amplifier gain as a function of $R1$:

$$A_V = 1 + \left(\frac{R2 + R3}{R1} \right)$$

Readjusting the offset null is necessary whenever the circuit gain is changed. Note that if U2B is needed for another application, $R7$ can be terminated at ground. The low input offset voltage of the TL052A minimizes the dc error of the circuit. For best matching, all resistors should be one percent tolerance. The matching between $R4$, $R5$, $R6$, and $R7$ controls the CMRR of this application.

The following equation shows the output voltages when the input voltage equals zero. This dc error can be nulled by adjusting the offset null potentiometer; however, any change in offset voltage over time or temperature also creates an error. To calculate the error from changes in offset, consider the three offset components in the equation as delta offsets rather than initial offsets. The improved stability of Texas Instruments enhanced JFETs minimizes the error resulting from change in input offset voltage with time. Assuming V_{IN} equals zero, V_O can be shown as a function of the offset voltage:

$$V_O = V_{IO2} \left[\left(1 + \frac{R3}{R1} \right) \left(\frac{R7}{R5 + R7} \right) \left(1 + \frac{R6}{R4} \right) + \frac{R2}{R1} \left(\frac{R6}{R4} \right) \right] - V_{IO1} \left[\frac{R3}{R1} \left(\frac{R7}{R5 + R7} \right) \left(1 + \frac{R6}{R4} \right) + \frac{R6}{R4} \left(1 + \frac{R2}{R1} \right) \right] + V_{IO3} \left(1 + \frac{R6}{R4} \right)$$



NOTE: U1A through U2B = TL052A; $V_{CC \pm} = \pm 15 \text{ V}$.

FIGURE 53. INSTRUMENTATION AMPLIFIER

TL052, TL052A
ENHANCED JFET PRECISION
DUAL OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

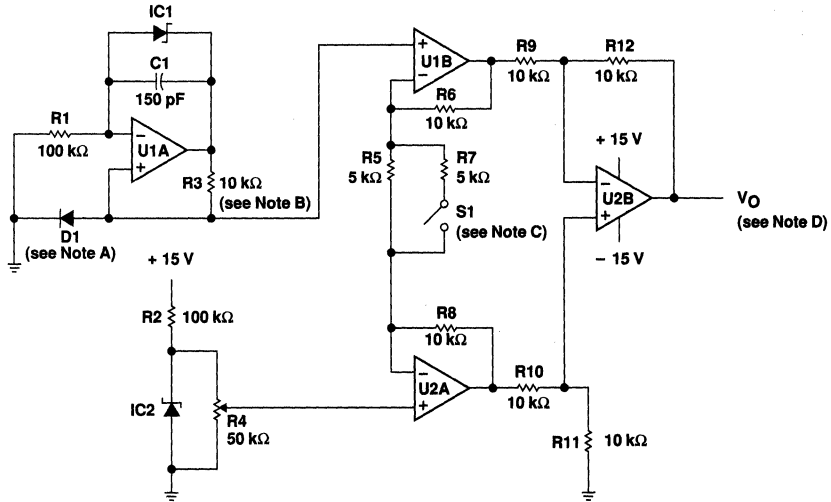
analog thermometer

By combining a current source that does not vary over temperature with an instrumentation amplifier, a precise analog thermometer can be built (see Figure 54). Amplifier U1A and IC1 establish a constant current through the temperature sensing diode D1. For this section of the circuit to operate correctly, the TL052 must use split supplies and R3 must be a metal film resistor with a low temperature coefficient.

The temperature-sensitive voltage from the diode is compared to a temperature-stable voltage reference set by IC2. R4 should be adjusted to provide the correct output voltage when the diode is at a known temperature. Although this potentiometer resistance varies with temperature, the divider ratio of the potentiometer remains constant.

Amplifiers U1B, U2A, and U2B form the instrumentation amplifier that converts the difference between the diode and reference voltage to a voltage proportional to the temperature. With switch S1 closed, the amplifier gain equals 5, and the output voltage is proportional to temperature in degrees Celsius. With S1 open, the amplifier gain is 9, and the output is proportional to temperature in degrees Fahrenheit. Every time that S1 is changed, R4 must be recalibrated. By setting S1 correctly, the output voltage equals 10 mV per degree (C or F).

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Operational Amplifiers



- NOTES: A. Temperature sensing diode = (-2 mV/°C).
 B. Metal film (low temperature coefficient).
 C. Switch open for °F and closed for °C.
 D. V_O ∝ Temperature; 10 mV/°C or 10 mV/°F.
 E. U1A through U2B = TL052. IC1, IC2 = LM385, LT1004, or LT1009 voltage reference.

FIGURE 54. ANALOG THERMOMETER

TYPICAL APPLICATION DATA

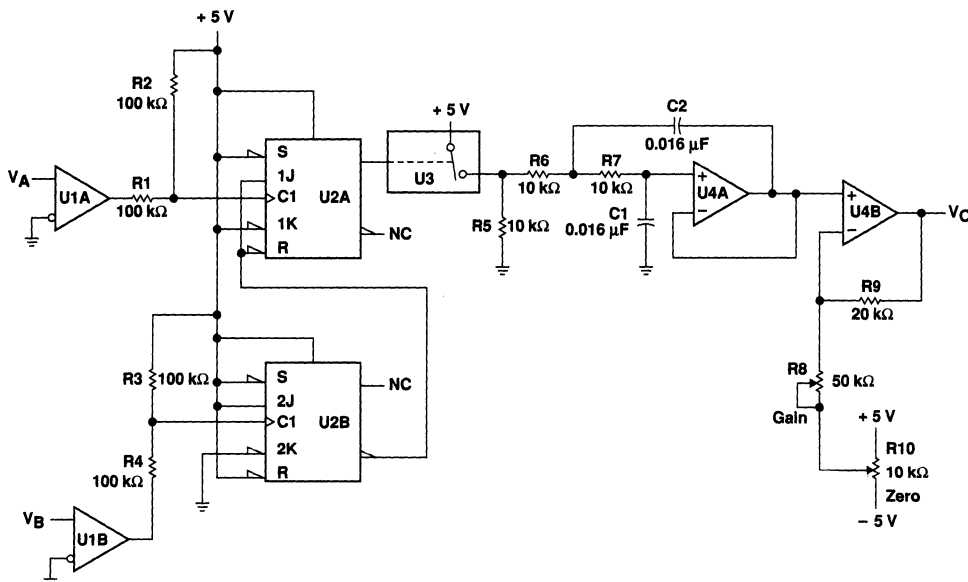
phase meter

The phase meter in Figure 55 produces an output voltage of 10 mV per degree of phase delay between the two input signals V_A and V_B . The reference signal V_A must be the same frequency as V_B . The TLC3702 comparators (U1) convert these two input sine waves into ± 5 V square waves. Then R1 and R4 provide level shifting prior to the SN74HC109 dual J-K flip flops.

Flip-flop U2B is connected as a toggle flip-flop and generates a square wave at half the frequency of V_B . Flip-flop U2A also produces a square wave at half the input frequency. The pulse duration of U2A varies from zero to half the period, where zero corresponds to zero phase delay between V_A and V_B , and half the period corresponds to V_B lagging V_A by 360 degrees.

The output pulse from U2A causes the TLC4066 (U3) switch to charge the TL052 (U4) integrator capacitors C1 and C2. As the phase delay approaches 360 degrees, the output of U4A approximates a square wave, and U2A has an output of almost 2.5 V. U4B acts as a noninverting amplifier with a gain of 1.44 in order to scale the 0- to 2.5-V integrator output to a 0-to 3.6-V output range.

R8 and R10 provide output gain and zero-level calibration. This circuit operates over a 100-Hz to 10-kHz frequency range.



- NOTES: U1 = TLC3702; $V_{CC} \pm = \pm 5$ V.
 U2 = SN74HC109.
 U3 = TLC4066.
 U4 = TL052; $V_{CC} \pm = \pm 5$ V.

FIGURE 55. PHASE METER

**TL052, TL052A
ENHANCED JFET PRECISION
DUAL OPERATIONAL AMPLIFIERS**

TYPICAL APPLICATION DATA

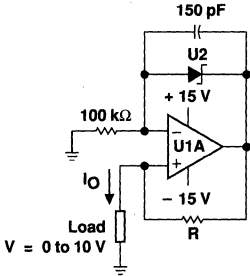
precision constant-current source over temperature

A precision current source benefits from the high input impedance and stability of Texas Instruments enhanced JFET process. A low-current shunt regulator maintains 2.5 V between the inverting input and the output of the TL052. The negative feedback then forces 2.5 V across the current setting resistor R; therefore, the current to the load is simply 2.5 V divided by R.

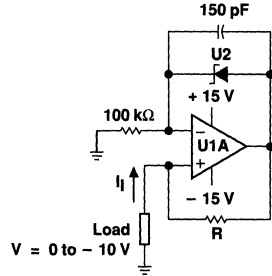
Possible choices for the shunt regulator include the LT1004, LT1009, and LM385. Note that if the regulator's cathode connects to the op amp output, this circuit will source load current. Similarly, if the cathode connects to the inverting input, the circuit will sink current from the load. To minimize output current change with temperature, R should be a metal film resistor with a low temperature coefficient. Also, this circuit must be operated with split voltage supplies.

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Operational Amplifiers



(a) SOURCE CURRENT LOAD



(b) SINK CURRENT LOAD

NOTES: IC1 = LM385, LT1004, or LT1009 voltage reference.

U1A = TL052.

$I = \frac{2.5V}{R}$, R = Low temperature coefficient metal film resistor.

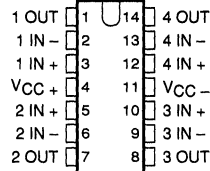
FIGURE 56. PRECISION CONSTANT-CURRENT SOURCE

TL054, TL054A ENHANCED JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

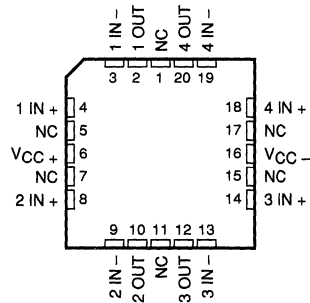
D3236, JUNE 1988 – REVISED JANUARY 1989

- **Maximum Offset Voltage** ... 1.5 mV (TL054A)
- **High Slew Rate** ... 15.9 V/ μ s Typ at 25°C
- **Low Total Harmonic Distortion** ... 0.003% Typ at $R_L = 2$ k Ω
- **Low Noise Voltage** ... 21 nV/ $\sqrt{\text{Hz}}$ Typ at $f = 1$ kHz
- **Low Input Bias Currents** ... 30 pA Typ
- **Monolithic Construction**

**D, J, or N PACKAGE
(TOP VIEW)**



**FK PACKAGE
(TOP VIEW)**



NC – No internal connection

description

The TL054 and TL054A quad operational amplifiers incorporate well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. These devices offer the significant advantages of Texas Instruments new enhanced JFET process. This process affords not only low initial offset voltage due to the on-chip zener trim capability but also stable offset voltage over time and temperature. In comparison, traditional JFET processes are plagued by significant offset voltage drift.

This new enhanced process still maintains the traditional JFET advantages of fast slew rates and low input bias and offset currents. These advantages coupled with low noise and low harmonic distortion make the TL054 well-suited for new state-of-the-art designs as well as existing design upgrades. The TL054 has been designed to be functionally compatible, as well as pin compatible, with the TL074 and TL084.

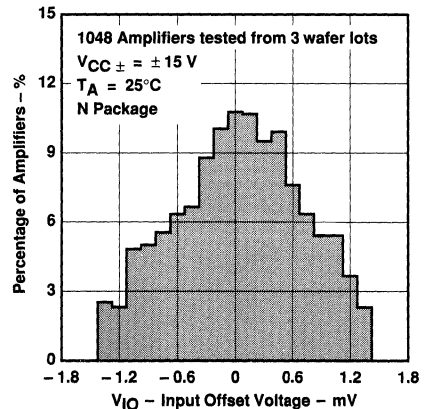
Two offset voltage grades are available: TL054 (4 mV max) and TL054A (1.5 mV max).

AVAILABLE OPTIONS

T _A	V _{IO} max AT 25°C	PACKAGE			
		SMALL-OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	1.5 mV	TL054ACD	—	TL054ACJ	TL054ACN
	4 mV	TL054CD	—	TL054CJ	TL054CN
-40°C to 85°C	1.5 mV	TL054AID	—	TL054AIJ	TL054AIN
	4 mV	TL054ID	—	TL054IJ	TL054IN
-55°C to 125°C	1.5 mV	TL054AMD	TL054AMFK	TL054AMJ	TL054AMN
	4 mV	TL054MD	TL054MFK	TL054MJ	TL054MN

D packages are available taped-and-reeled. Add "R" suffix to device type (e.g., TL054CDR).

**DISTRIBUTION OF TL054A
INPUT OFFSET VOLTAGE**



PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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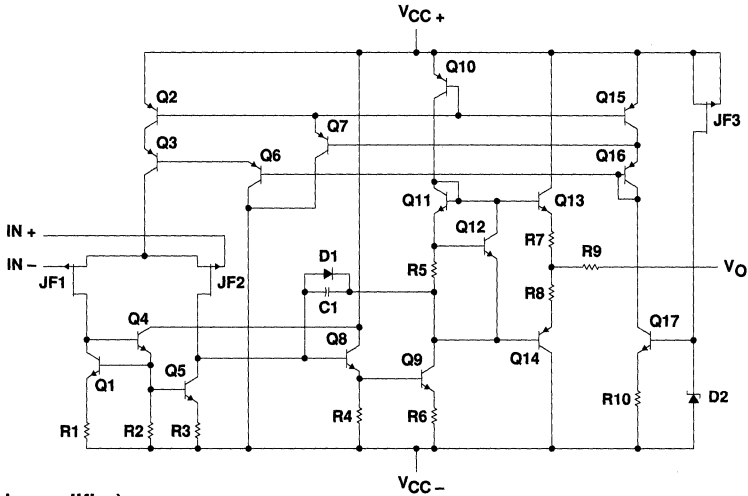
TL054, TL054A ENHANCED JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

description (continued)

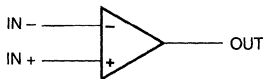
A variety of available packaging options includes small-outline and chip carrier versions for high-density system applications.

The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I-suffix devices are characterized for operation from -40°C to 85°C , and the C-suffix devices are characterized for operation from 0°C to 70°C .

equivalent schematic (each amplifier)



symbol (each amplifier)



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Operational Amplifiers

TL054, TL054A ENHANCED JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	18 V
Supply voltage, V_{CC-} (see Note 1)	-18 V
Differential input voltage (see Note 2)	± 30 V
Input voltage range, V_I (any input, see Notes 1 and 3)	± 15 V
Input current, I_I (each input)	± 1 mA
Output current, I_O (each output)	± 80 mA
Total current into V_{CC+} terminal	160 mA
Total current out of V_{CC-} terminal	160 mA
Duration of short-circuit current at (or below) 25°C (see Note 4)	Unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, T_A : M-suffix	-55°C to 125°C
I-suffix	-40°C to 85°C
C-suffix	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

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Operational Amplifiers

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input with respect to the inverting input.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
 4. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	190 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
N	1575 mW	12.6 mW/°C	1008 mW	819 mW	230 mW

recommended operating conditions

		M-SUFFIX		I-SUFFIX		C-SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, V_{CC}		± 5	± 15	± 5	± 15	± 5	± 15	V
Common-mode input voltage, V_{IC}	$V_{CC} \pm \pm 5$ V	-1	4	-1	4	-1	4	V
	$V_{CC} \pm \pm 15$ V	-11	11	-11	11	-11	11	
Operating free-air temperature, T_A		-55	125	-40	85	0	70	°C

TL054M, TL054AM ENHANCED JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ±5 V			V _{CC} ± = ±15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL054M	25°C	0.64	5.5	0.56	4	mV	
			Full range	10.5		9			
		TL054AM	25°C	0.57	3.5	0.5	1.5		
			Full range	8.5		6.5			
α _{VIO} Temperature coefficient of input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL054M	25°C to 125°C	21		20		μV/°C	
		TL054AM	25°C to 125°C	21		20			
Input offset voltage long-term drift (see Note 5)		25°C	0.04		0.04		μV/mo		
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	4 100		5 100		pA		
		125°C	1 20		2 20		nA		
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	20 200		30 200		pA		
		125°C	10 50		20 50		nA		
V _{ICR} Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	V		
		Full range	-1 to 4		-11 to 11				
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.2	13	13.9	V		
		Full range	3		13				
	R _L = 2 kΩ	25°C	2.5	3.8	11.5	12.7			
		Full range	2.5		11.5				
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-2.5	-3.5	-12	-13.2	V		
		Full range	-2.5		-12				
	R _L = 2 kΩ	25°C	-2.3	-3.2	-11	-12			
		Full range	-2.3		-11				
A _{VD} Large-signal differential voltage amplification	R _L = 2 kΩ, See Note 6	25°C	25	72	50	133	V/mV		
		-55°C	30	99	60	209			
		125°C	10	35	15	35			
r _i Input resistance		25°C	10 ¹²		10 ¹²		Ω		
C _i Input capacitance		25°C	10		12		pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	65	84	75	92	dB		
		-55°C	65	83	75	92			
		125°C	65	84	75	93			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ±5 V to ±15 V, V _O = 0, R _S = 50 Ω	25°C	75	99	75	99	dB		
		-55°C	75	98	75	98			
		125°C	75	100	75	100			
I _{CC} Supply current (four amplifiers)	No load, V _O = 0	25°C	8.1	11.2	8.4	11.2	mA		
		-55°C	7.8	12.8	8.1	12.8			
		125°C	7.1	11.2	7.5	11.2			
V _{O1} /V _{O2} Crosstalk attenuation	A _{VD} = 100	25°C	120		120		dB		

[†] Full range is -55°C to 125°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. For V_{CC} ± = ±5 V, V_O = ±2.3 V; at V_{CC} ± = ±15 V, V_O = ±10 V.

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Operational Amplifiers

TL054M, TL054AM ENHANCED JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC±} = ± 5 V			V _{CC±} = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR + Positive slew rate at unity gain	R _L = 2 kΩ, C _L = 100 pF, See Figure 1 and Note 7	25°C		15.4		10	17.8	V/μs	
		-55°C		16.7		18.3			
		125°C		12.9		16.7			
SR - Negative slew rate at unity gain		25°C		13.9		10	15.9		
		-55°C		14.7		16.3			
		125°C		12.2		14.5			
t _r Rise time	V _{Ipp} = ± 10 mV, R _L = 2 kΩ, C _L = 100 pF, See Figures 1 and 2	25°C		55		56	ns		
		-55°C		51		52			
		125°C		68		68			
t _f Fall time		25°C		55		57			
		-55°C		51		52			
		125°C		68		69			
Overshoot factor		25°C		24%		19%			
		-55°C		25%		19%			
		125°C		25%		19%			
V _n Equivalent input noise voltage	R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C	75		75	nV/√Hz		
		f = 1 kHz	25°C	21		21			
V _{NPP} Peak-to-peak equivalent input noise voltage	f = 10 Hz to 10 kHz	25°C	4		4	μV			
I _n Equivalent input noise current	f = 1 kHz	25°C	0.01		0.01	pA/√Hz			
THD Total harmonic distortion	R _S = 1 kΩ, R _L = 2 kΩ, f = 1 kHz, See Note 8	25°C	0.003%		0.003%				
B ₁ Unity-gain bandwidth	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C		2.7		2.7	MHz		
		-55°C		3.4		3.4			
		125°C		2.1		2.1			
φ _m Phase margin at unity gain	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C		61°		64°			
		-55°C		58°		62°			
		125°C		60°		64°			

[†] Full range is -55°C to 125°C.

NOTES: 7. For V_{CC±} = ± 5 V, V_{Ipp} = ± 1 V; for V_{CC±} = ± 15 V, V_{Ipp} = ± 5 V.

8. For V_{CC±} = ± 5 V, V_{o(rms)} = 1 V; for V_{CC±} = ± 15 V, V_{o(rms)} = 6 V.

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Operational Amplifiers

TL054I, TL054AI ENHANCED JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL054I	25°C	0.64 5.5		0.56 4		mV	
			Full range	8.8		7.3			
		TL054AI	25°C	0.57 3.5		0.5 1.5			
			Full range	6.8		4.8			
αV _{IO} Temperature coefficient of input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL054I	25°C to 85°C	25		24		μV/°C	
		TL054AI	25°C to 85°C	25		23			
Input offset voltage long-term drift (see Note 5)		25°C	0.04		0.04		μV/mo		
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	4 100		5 100		pA		
		85°C	0.06 10		0.07 10		nA		
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	20 200		30 200		pA		
		85°C	0.6 20		0.7 20		nA		
V _{ICR} Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	V		
		Full range	-1 to 4	to 4	-11 to 11	to 11			
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3 4.2		13 13.9		V		
		Full range	3		13				
	R _L = 2 kΩ	25°C	2.5 3.8		11.5 12.7				
		Full range	2.5		11.5				
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-2.5 -3.5		-12 -13.2		V		
		Full range	-2.5		-12				
	R _L = 2 kΩ	25°C	-2.3 -3.2		-11 -12				
		Full range	-2.3		-11				
A _{VD} Large-signal differential voltage amplification	R _L = 2 kΩ, See Note 6	25°C	25 72		50 133		V/mV		
		-40°C	30 101		60 212				
		85°C	20 50		30 70				
r _i Input resistance		25°C	10 ¹²		10 ¹²		Ω		
C _i Input capacitance		25°C	10		12		pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	65 84		75 92		dB		
		-40°C	65 83		75 92				
		85°C	65 84		75 93				
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± = ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75 99		75 99		dB		
		-40°C	75 98		75 99				
		85°C	75 99		75 99				
I _{CC} Supply current (four amplifiers)	No load, V _O = 0	25°C	8.1 11.2		8.4 11.2		mA		
		-40°C	7.9 12.8		8.2 12.8				
		85°C	7.6 11.2		7.9 11.2				
V _{O1} /V _{O2} Crosstalk attenuation	A _{VD} = 100	25°C	120		120		dB		

[†] Full range is -40°C to 85°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. For V_{CC} ± = ± 5 V, V_O = ± 2.3 V; at V_{CC} ± = ± 15 V, V_O = ± 10 V.

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Operational Amplifiers

TL054I, TL054AI
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QUAD OPERATIONAL AMPLIFIERS

operating characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± ± 5 V			V _{CC} ± ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR + Positive slew rate at unity gain	R _L = 2 kΩ, C _L = 100 pF, See Figure 1 and Note 7	25°C		15.4		10	17.8	V/μs	
		-40°C		16.4		8	18		
		85°C		14		8	17.3		
SR - Negative slew rate at unity gain		25°C		13.9		10	15.9		
		-40°C		14.7		8	16.1		
		85°C		13		8	15.3		
t _r Rise time	V _{Ipp} = ± 10 mV, R _L = 2 kΩ, C _L = 100 pF, See Figures 1 and 2	25°C		55		56	ns		
		-40°C		52		53			
		85°C		64		65			
t _f Fall time		25°C		55		57			
		-40°C		51		53			
		85°C		64		65			
Overshoot factor		25°C		24%		19%			
		-40°C		24%		19%			
		85°C		24%		19%			
V _n Equivalent input noise voltage (see Note 9)	R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C	75		75	nV/√Hz		
		f = 1 kHz	25°C	21		21		45	
V _{NPP} Peak-to-peak equivalent input noise voltage	See Figure 3	f = 10 Hz to 10 kHz	25°C	4		4	μV		
I _n Equivalent input noise current	f = 1 kHz	25°C		0.01		0.01	pA/√Hz		
THD Total harmonic distortion	R _S = 1 kΩ, R _L = 2 kΩ, f = 1 kHz, See Note 8	25°C		0.003%		0.003%			
B ₁ Unity-gain bandwidth	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C		2.7		2.7	MHz		
		-40°C		3.3		3.3			
		85°C		2.3		2.4			
φ _m Phase margin at unity gain	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C		61°		64°			
		-40°C		59°		62°			
		85°C		61°		64°			

[†] Full range is -40°C to 85°C.

NOTES: 7. For V_{CC} ± ± 5 V, V_{Ipp} = ± 1 V; for V_{CC} ± ± 15 V, V_{Ipp} = ± 5 V.

8. For V_{CC} ± ± 5 V, V_{O(rms)} = 1 V; for V_{CC} ± ± 15 V, V_{O(rms)} = 6 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

TL054C, TL054AC

ENHANCED JFET PRECISION

QUAD OPERATIONAL AMPLIFIERS

electrical characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± ± 5 V			V _{CC} ± ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL054C	25°C	0.64	5.5	0.56	4	mV	
			Full range	7.7		6.2			
TL054AC		25°C	0.57	3.5	0.5	1.5			
		Full range	5.7		3.7				
α _{VIO} Temperature coefficient of input offset voltage	TL054C	25°C to 70°C	25			23			μV/°C
		TL054AC	25°C to 70°C	24			23		
Input offset voltage long-term drift (see Note 5)		25°C	0.04			0.04			μV/mo
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	4	100	5	100	pA		
		70°C	0.02	1	0.025	1	nA		
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	20	200	30	200	pA		
		70°C	0.15	4	0.2	4	nA		
V _{ICR} Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	V		
		Full range	-1 to 4		-11 to 11				
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.2	13	13.9	V		
		Full range	3		13				
	R _L = 2 kΩ	25°C	2.5	3.8	11.5	12.7			
		Full range	2.5		11.5				
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-2.5	-3.5	-12	-13.2	V		
		Full range	-2.5		-12				
	R _L = 2 kΩ	25°C	-2.3	-3.2	-11	-12			
		Full range	-2.3		-11				
A _{VD} Large-signal differential voltage amplification	R _L = 2 kΩ, See Note 6	25°C	25	72	50	133	V/mV		
		0°C	30	88	60	173			
		70°C	20	57	30	85			
r _i Input resistance		25°C	10 ¹²			10 ¹²	Ω		
C _i Input capacitance		25°C	10			12	pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min, V _O = 0, R _S = 50 Ω	25°C	65	84	75	92	dB		
		0°C	65	84	75	92			
		70°C	65	84	75	93			
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} ± / ΔV _{IO})	V _{CC} ± ± 5 V to ± 15 V, V _O = 0, R _S = 50 Ω	25°C	75	99	75	99	dB		
		0°C	75	99	75	99			
		70°C	75	99	75	99			
I _{CC} Supply current (four amplifiers)	No load, V _O = 0	25°C	8.1	11.2	8.4	11.2	mA		
		0°C	8.2	12.8	8.5	12.8			
		70°C	7.9	11.2	8.2	11.2			
V _{O1} /V _{O2} Crosstalk attenuation	A _{VD} = 100	25°C	120			120	dB		

[†] Full range is 0°C to 70°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C extrapolated to T_A = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. For V_{CC} ± ± 5 V, V_O = ± 2.3 V; at V_{CC} ± ± 15 V, V_O = ± 10 V.

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Operational Amplifiers

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operating characteristics

PARAMETER	TEST CONDITIONS	T _A [†]	V _{CC} ± = ± 5 V			V _{CC} ± = ± 15 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR + Positive slew rate at unity gain	R _L = 2 kΩ, C _L = 100 pF, See Figure 1 and Note 7	25°C	15.4			10	17.8	V/μs	
		0°C	15.7			8	17.9		
		70°C	14.4			8	17.5		
SR - Negative slew rate at unity gain		25°C	13.9			10	15.9		
		0°C	14.3			8	16.1		
		70°C	13.3			8	15.5		
t _r Rise time	V _{Ipp} = ± 10 mV, R _L = 2 kΩ, C _L = 100 pF, See Figures 1 and 2	25°C	55		56		ns		
		0°C	54		55				
		70°C	63		63				
t _f Fall time		25°C	55		57				
		0°C	54		56				
		70°C	62		64				
Overshoot factor		25°C	24%		19%				
		0°C	24%		19%				
		70°C	24%		19%				
V _n Equivalent input noise voltage (see Note 9)		R _S = 100 Ω, See Figure 3	f = 10 Hz	25°C		75		nV/√Hz	
			f = 1 kHz	25°C		21			
V _{NPP} Peak-to-peak equivalent input noise voltage		See Figure 3	f = 10 Hz to 10 kHz	25°C		4		μV	
			25°C		4				
I _n Equivalent input noise current	f = 1 kHz	25°C	0.01		0.01		pA/√Hz		
THD Total harmonic distortion	R _S = 1 kΩ, R _L = 2 kΩ, f = 1 kHz, See Note 8	25°C	0.003%		0.003%				
B ₁ Unity-gain bandwidth	V _i = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C	2.7		2.7		MHz		
		0°C	3		3				
		70°C	2.4		2.4				
φ _m Phase margin at unity gain		25°C	61°		64°				
	0°C	60°		64°					
	70°C	61°		63°					

[†] Full range is 0°C to 70°C.

NOTES: 7. For V_{CC} ± = ± 5 V, V_{Ipp} = ± 1 V; for V_{CC} ± = ± 15 V, V_{Ipp} = ± 5 V.

8. For V_{CC} ± = ± 5 V, V_{o(rms)} = 1 V; for V_{CC} ± = ± 15 V, V_{o(rms)} = 6 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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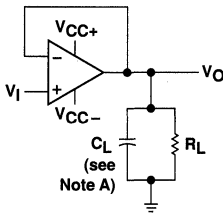
Operational Amplifiers

TL054, TL054A ENHANCED JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

PARAMETER MEASUREMENT INFORMATION

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Operational Amplifiers



NOTE A: C_L includes fixture capacitance.

FIGURE 1. SLEW RATE, RISE/FALL TIME, AND OVERSHOOT TEST CIRCUIT

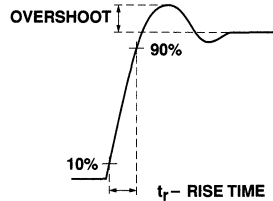


FIGURE 2. RISE TIME AND OVERSHOOT WAVEFORM

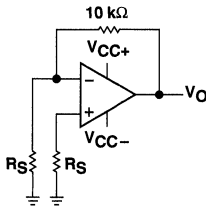


FIGURE 3. NOISE VOLTAGE TEST CIRCUIT

typical values

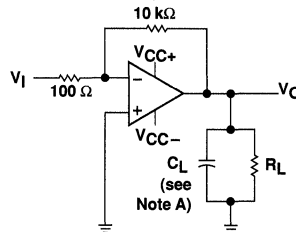
Typical values as presented in this data sheet represent the median (50% point) of device parametric performance.

input bias and offset current

At the picoamp-bias-current level typical of the TL054 and TL054A, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To accurately measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied, but with no device in the socket. The device is then inserted in the socket and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

noise

Because of the increasing emphasis on low noise levels in many of today's applications, the input noise voltage density is sample-tested at $f = 1$ kHz. Texas Instruments also has additional noise testing capability to meet specific application requirements. Please contact the factory for details.



NOTE A: C_L includes fixture capacitance.

FIGURE 4. UNITY-GAIN BANDWIDTH AND PHASE MARGIN TEST CIRCUIT

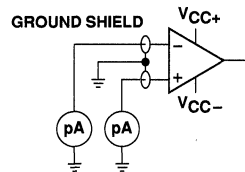


FIGURE 5. INPUT BIAS AND OFFSET CURRENT TEST CIRCUIT

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TYPICAL CHARACTERISTICS

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		vs Temperature	8
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		vs Temperature	20, 21
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		vs Frequency	23
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		vs Temperature	28
k_{SVR}	Supply-voltage rejection ratio	vs Temperature	30
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		vs Temperature	33
I_{CC}	Supply current	vs V_{CC}	34
		vs Temperature	35
SR	Slew rate	vs R_L	36, 37
		vs Temperature	38, 39
	Overshoot factor	vs C_L	40
V_n	Equivalent input noise voltage	vs Frequency	41
THD	Total harmonic distortion	vs Frequency	42
B_1	Unity-gain bandwidth	vs V_{CC}	43
		vs Temperature	44
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		vs C_L	46
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TL054, TL054A
ENHANCED JFET PRECISION
QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

**DISTRIBUTION OF TL054
 INPUT OFFSET VOLTAGE**

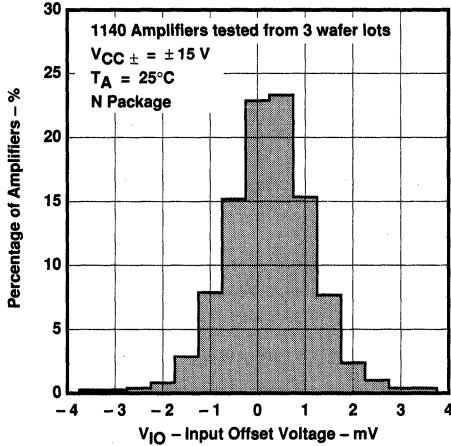


FIGURE 6

**DISTRIBUTION OF TL054
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT**

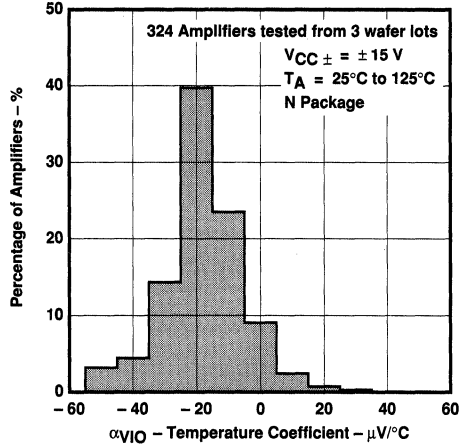


FIGURE 7

**INPUT BIAS CURRENT AND
 INPUT OFFSET CURRENT
 VS
 FREE-AIR TEMPERATURE**

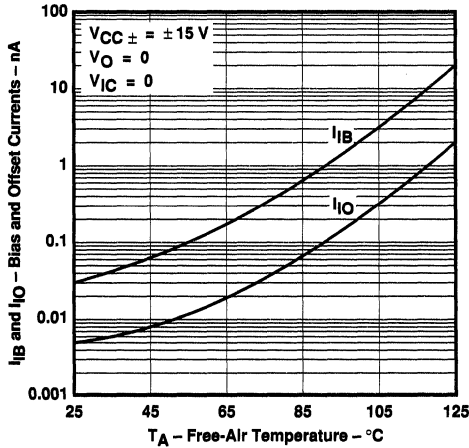


FIGURE 8

**INPUT BIAS CURRENT
 VS
 COMMON-MODE INPUT VOLTAGE**

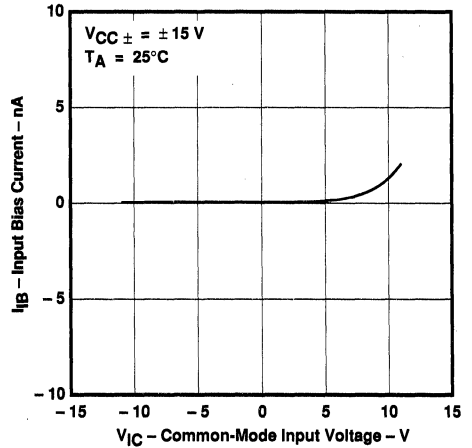


FIGURE 9

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

2
Operational Amplifiers

TL054, TL054A ENHANCED JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

COMMON-MODE
INPUT VOLTAGE RANGE LIMITS
VS
SUPPLY VOLTAGE

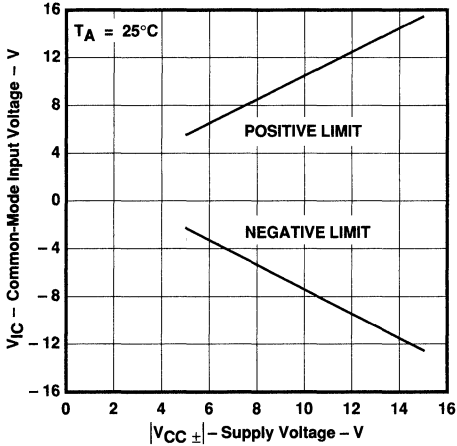


FIGURE 10

COMMON-MODE
INPUT VOLTAGE RANGE LIMITS
VS
FREE-AIR TEMPERATURE

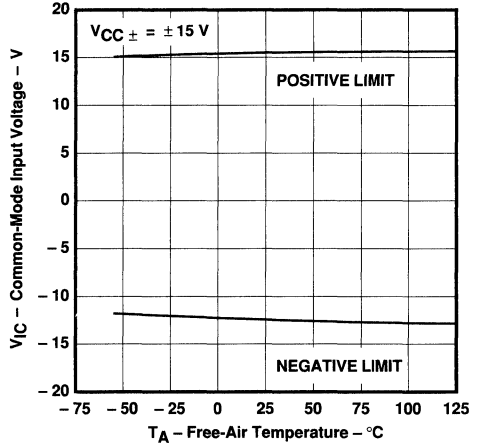


FIGURE 11

OUTPUT VOLTAGE
VS
DIFFERENTIAL INPUT VOLTAGE

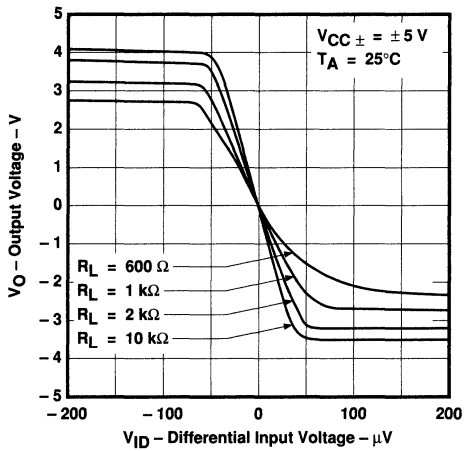


FIGURE 12

OUTPUT VOLTAGE
VS
DIFFERENTIAL INPUT VOLTAGE

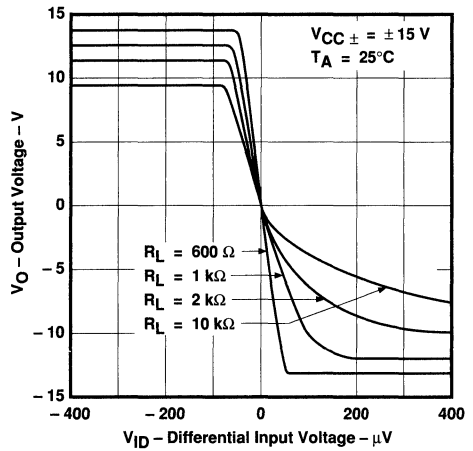


FIGURE 13

2

Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL054, TL054A
ENHANCED JFET PRECISION
QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE
VS
SUPPLY VOLTAGE

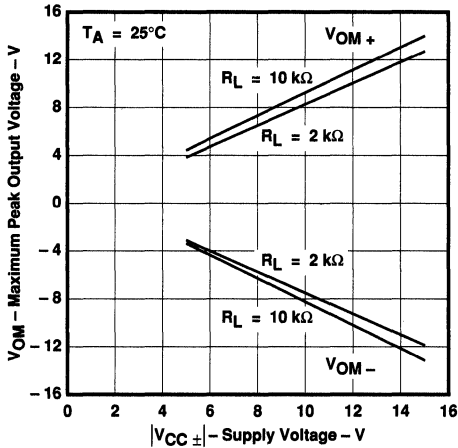


FIGURE 14

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
VS
FREQUENCY

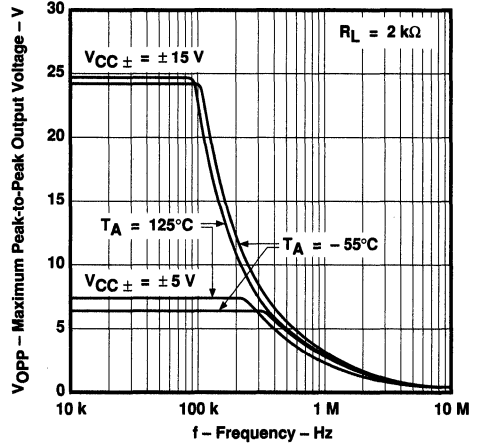


FIGURE 15

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
VS
FREQUENCY

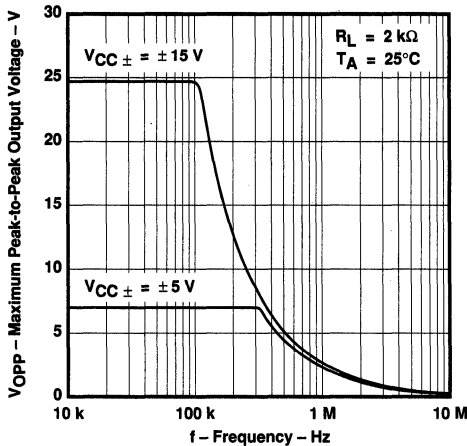


FIGURE 16

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
VS
FREQUENCY

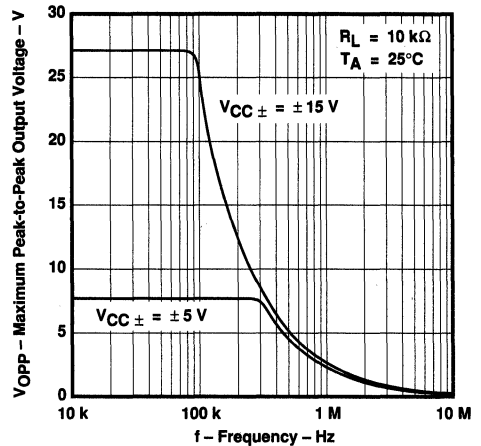


FIGURE 17

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 OUTPUT CURRENT

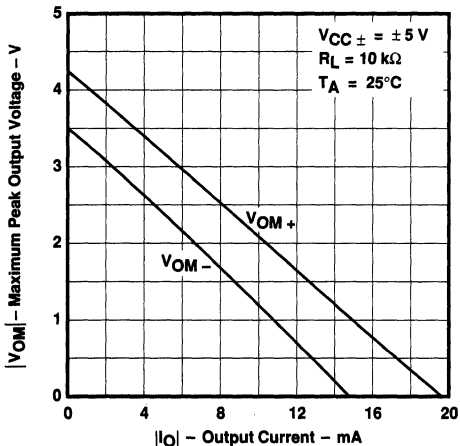


FIGURE 18

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 OUTPUT CURRENT

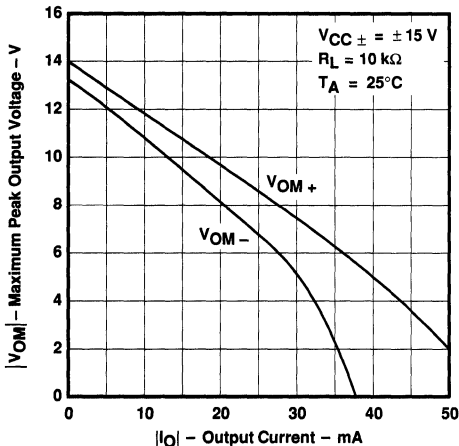


FIGURE 19

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

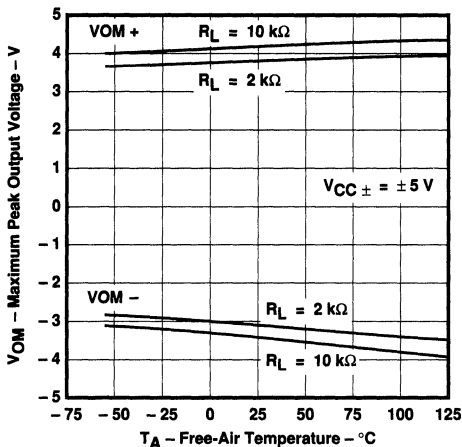


FIGURE 20

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

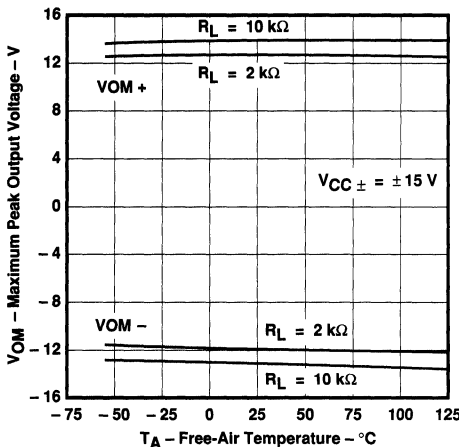


FIGURE 21

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL054, TL054A
ENHANCED JFET PRECISION
QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

LARGE-SIGNAL VOLTAGE AMPLIFICATION
 VS
 LOAD RESISTANCE

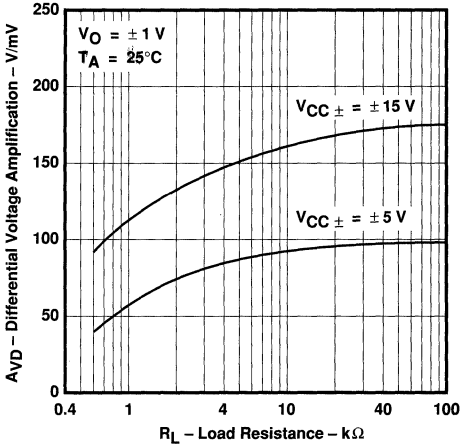


FIGURE 22

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 VS
 FREQUENCY

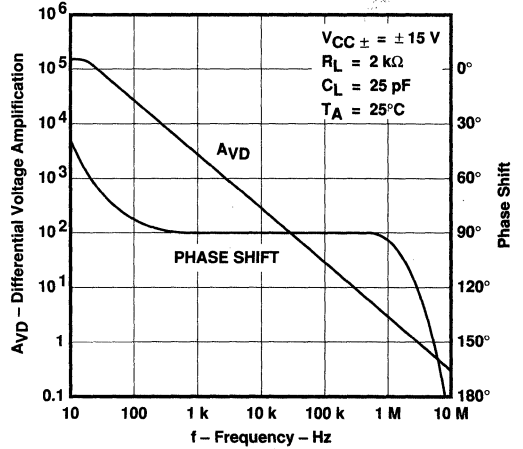


FIGURE 23

LARGE-SIGNAL VOLTAGE AMPLIFICATION
 VS
 FREE-AIR TEMPERATURE

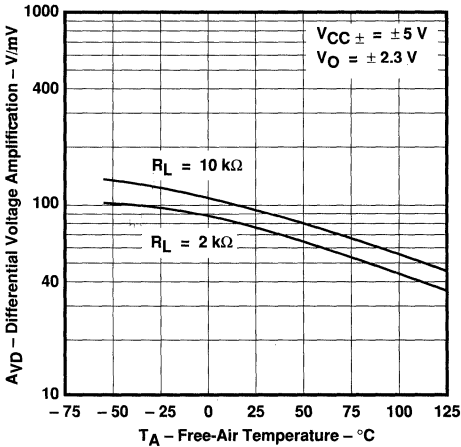


FIGURE 24

LARGE-SIGNAL VOLTAGE AMPLIFICATION
 VS
 FREE-AIR TEMPERATURE

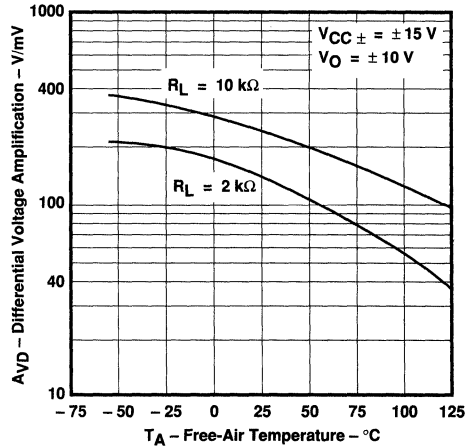


FIGURE 25

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

2
Operational Amplifiers

TYPICAL CHARACTERISTICS†

COMMON-MODE REJECTION RATIO
 VS
 FREQUENCY

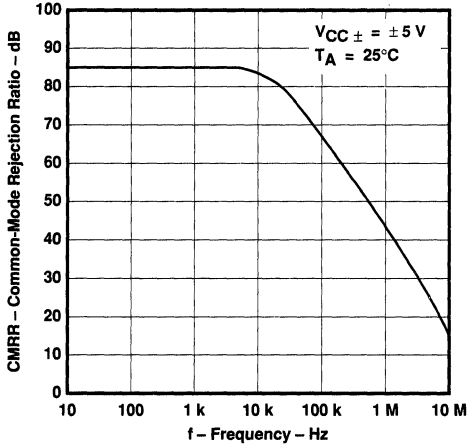


FIGURE 26

COMMON-MODE REJECTION RATIO
 VS
 FREQUENCY

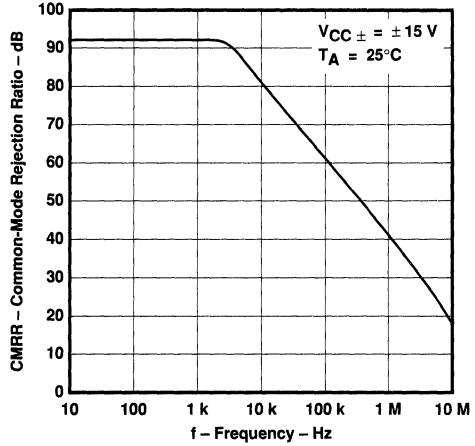


FIGURE 27

COMMON-MODE REJECTION RATIO
 VS
 FREE-AIR TEMPERATURE

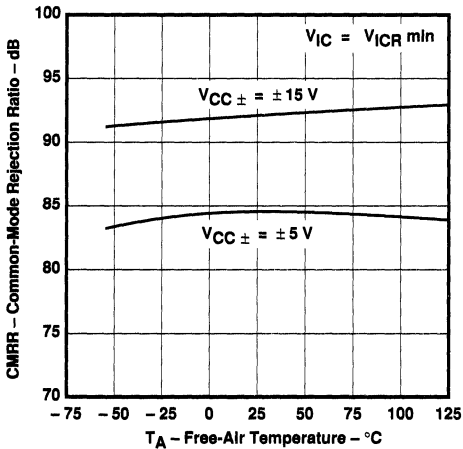


FIGURE 28

OUTPUT IMPEDANCE
 VS
 FREQUENCY

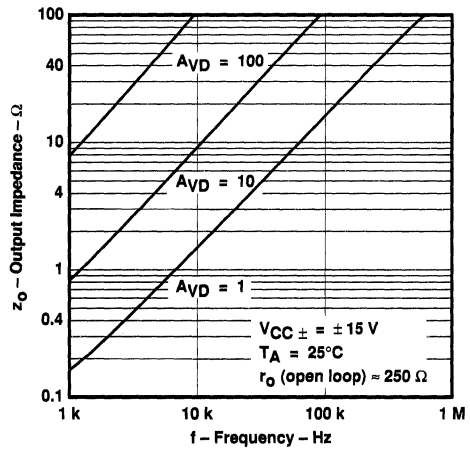


FIGURE 29

2
 Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL054, TL054A
ENHANCED JFET PRECISION
QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

SUPPLY-VOLTAGE REJECTION RATIO
VS
FREE-AIR TEMPERATURE

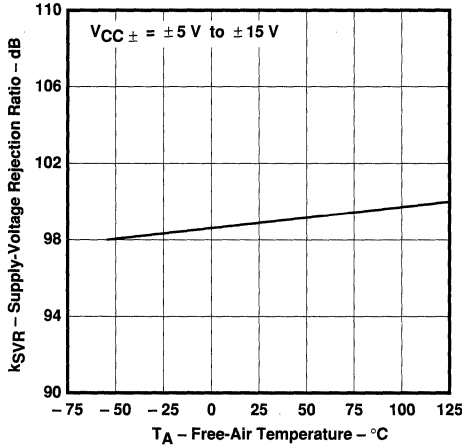


FIGURE 30

SHORT-CIRCUIT OUTPUT CURRENT
VS
SUPPLY VOLTAGE

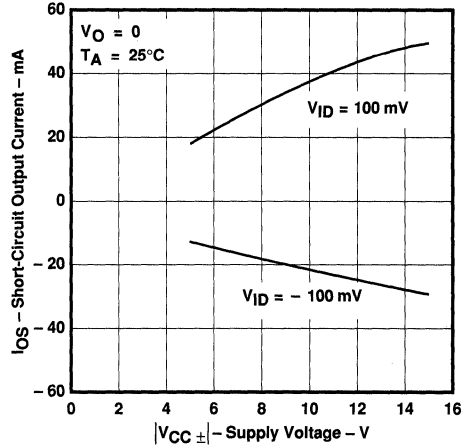


FIGURE 31

SHORT-CIRCUIT OUTPUT CURRENT
VS
TIME

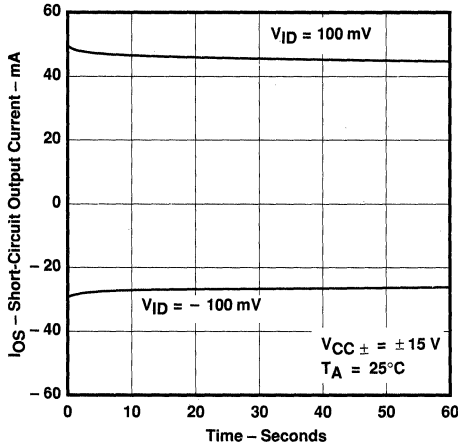


FIGURE 32

SHORT-CIRCUIT OUTPUT CURRENT
VS
FREE-AIR TEMPERATURE

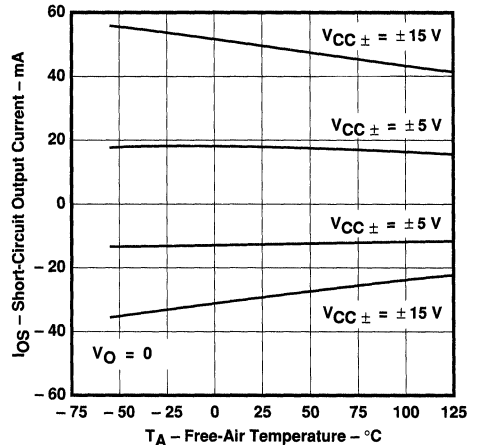


FIGURE 33

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL054, TL054A ENHANCED JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

SUPPLY CURRENT
vs
SUPPLY VOLTAGE

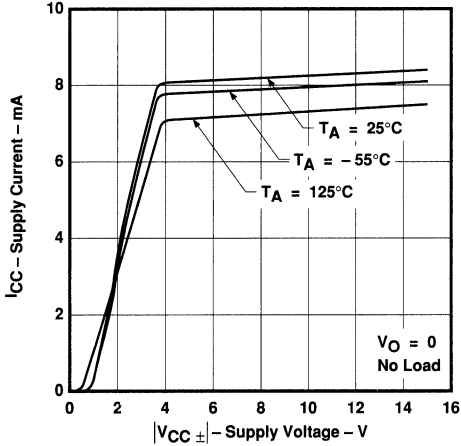


FIGURE 34

SUPPLY CURRENT
vs
FREE-AIR TEMPERATURE

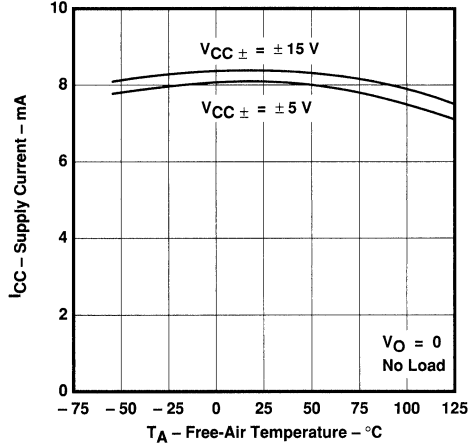


FIGURE 35

SLEW RATE
vs
LOAD RESISTANCE

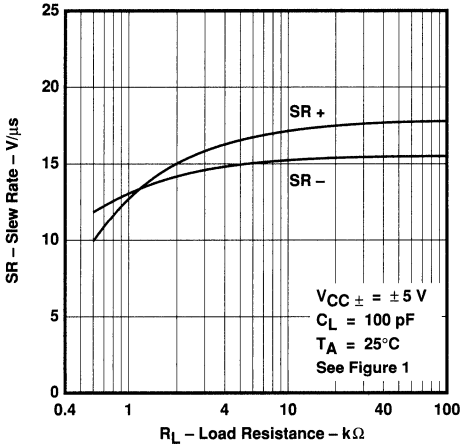


FIGURE 36

SLEW RATE
vs
LOAD RESISTANCE

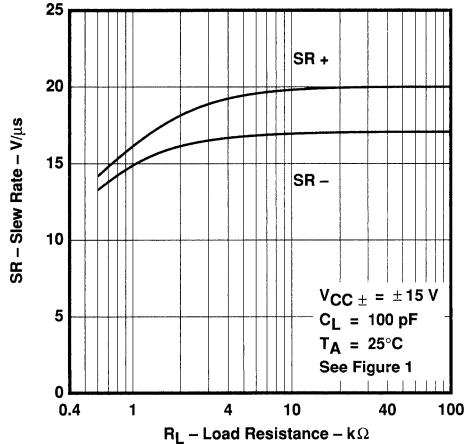


FIGURE 37

2

Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL054, TL054A
ENHANCED JFET PRECISION
QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

SLEW RATE
VS
FREE-AIR TEMPERATURE

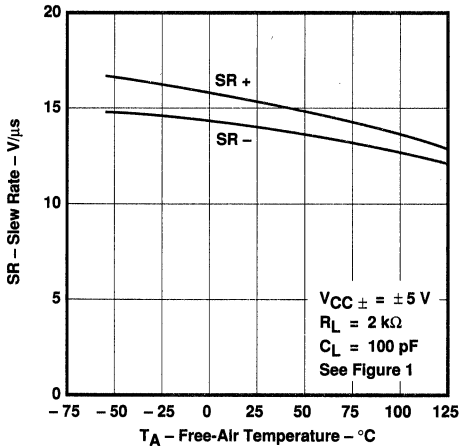


FIGURE 38

SLEW RATE
VS
FREE-AIR TEMPERATURE

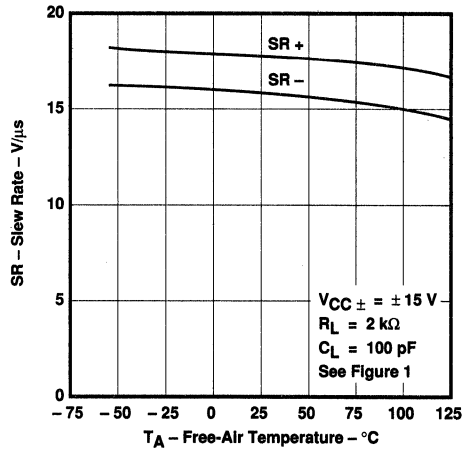


FIGURE 39

OVERSHOOT FACTOR
VS
LOAD CAPACITANCE

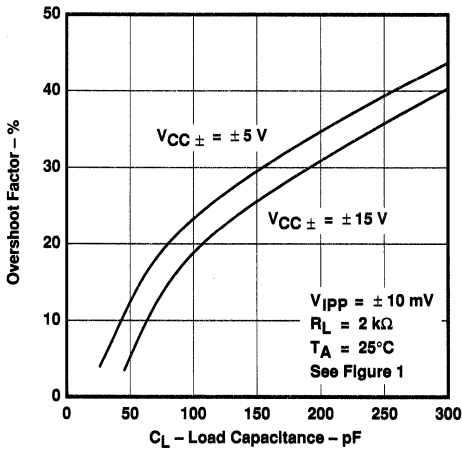


FIGURE 40

EQUIVALENT INPUT NOISE VOLTAGE
VS
FREQUENCY

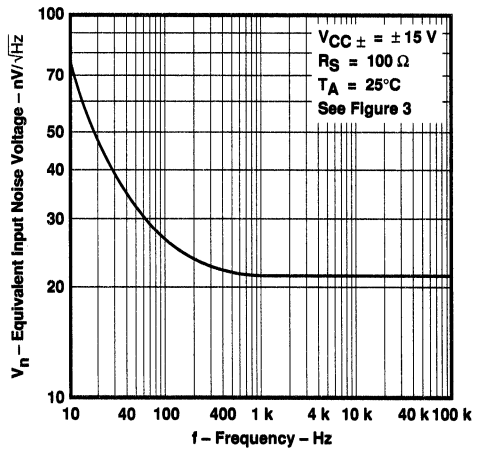


FIGURE 41

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

TOTAL HARMONIC DISTORTION
 VS
 FREQUENCY

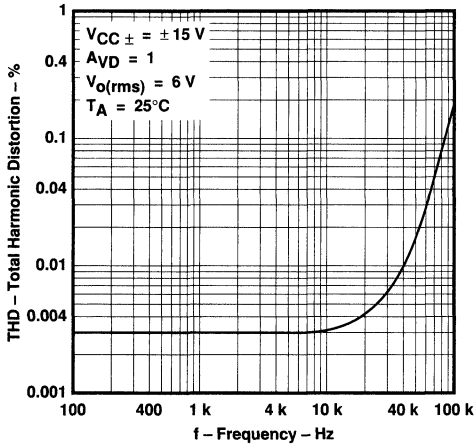


FIGURE 42

UNITY-GAIN BANDWIDTH
 VS
 SUPPLY VOLTAGE

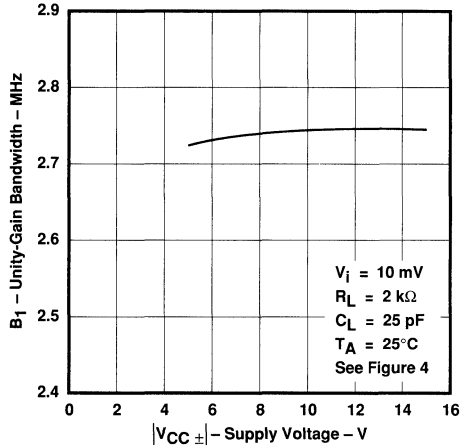


FIGURE 43

UNITY-GAIN BANDWIDTH
 VS
 FREE-AIR TEMPERATURE

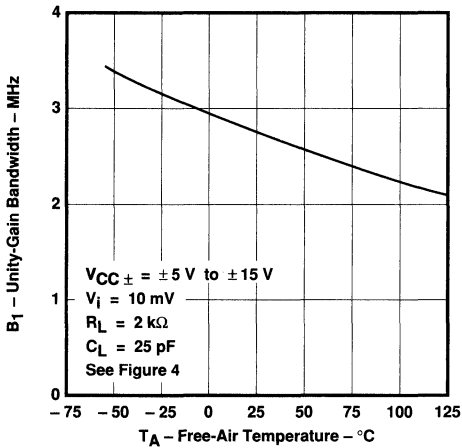


FIGURE 44

PHASE MARGIN
 VS
 SUPPLY VOLTAGE

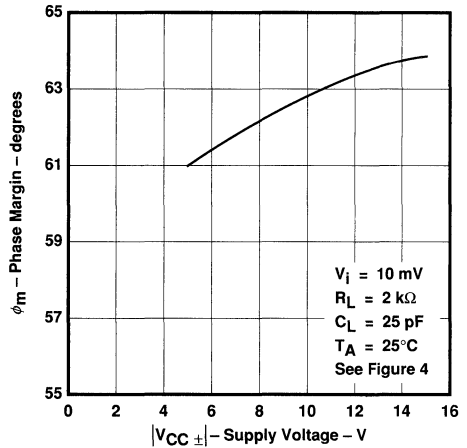


FIGURE 45

2
 Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL054, TL054A
ENHANCED JFET PRECISION
QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

PHASE MARGIN
VS
LOAD CAPACITANCE

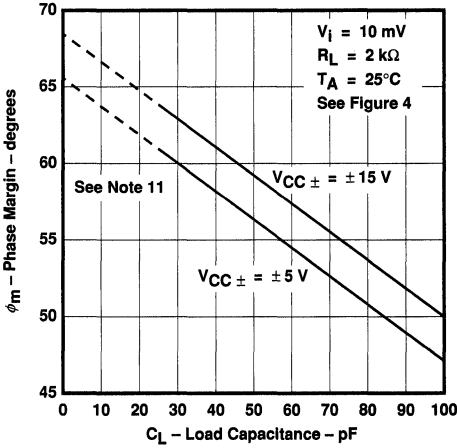


FIGURE 46

PHASE MARGIN
VS
FREE-AIR TEMPERATURE

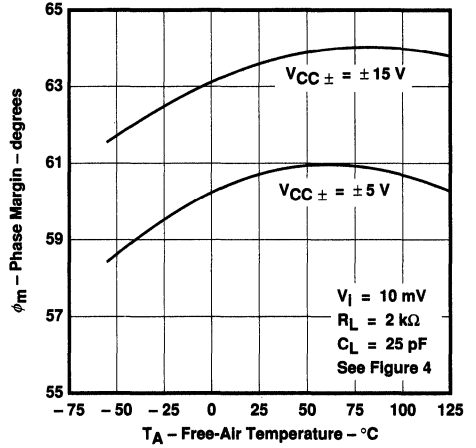


FIGURE 47

VOLTAGE-FOLLOWER
SMALL-SIGNAL
PULSE RESPONSE

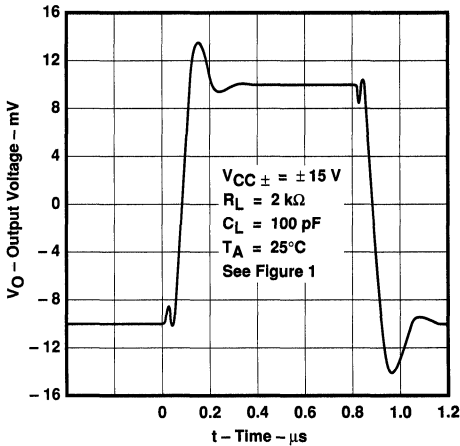


FIGURE 48

VOLTAGE-FOLLOWER
LARGE-SIGNAL
PULSE RESPONSE

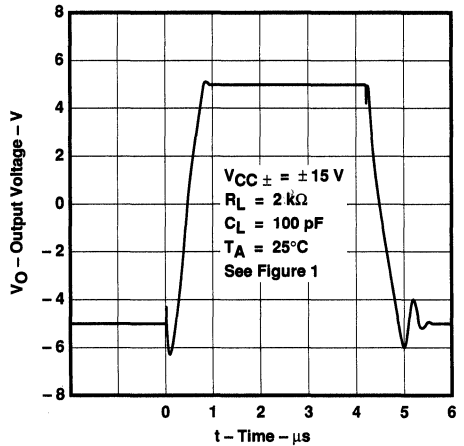


FIGURE 49

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.
 NOTE 11: Values of phase margin below a load capacitance of 25 pF were estimated.

TYPICAL APPLICATION DATA

output characteristics

All operating characteristics (except bandwidth and phase margin) are specified with 100-pF load capacitance. The TL054 and TL054A will drive higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem. Capacitive loads of 1000 pF and larger may be driven if enough resistance is added in series with the output (see Figure 50).

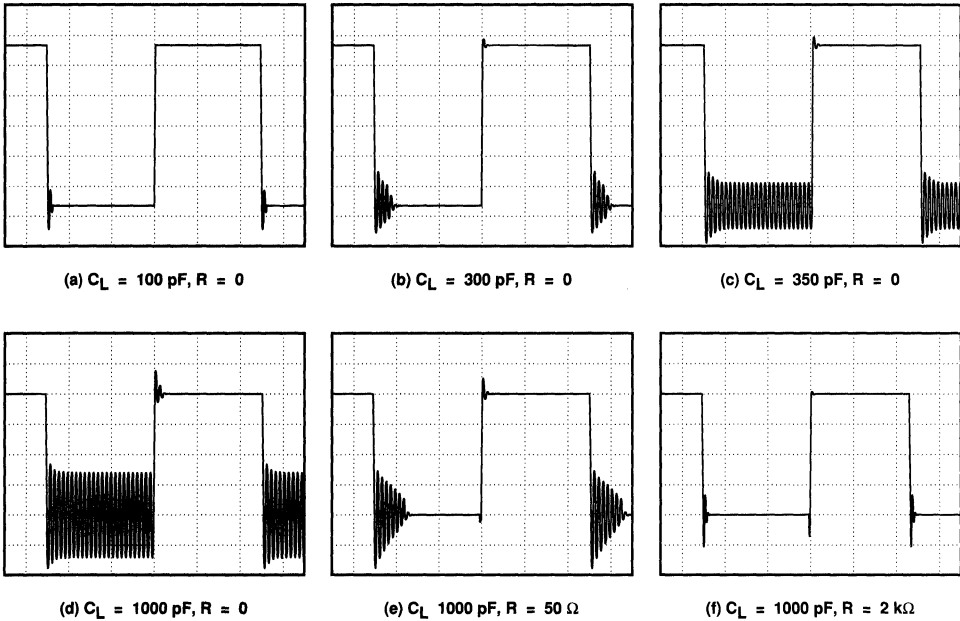
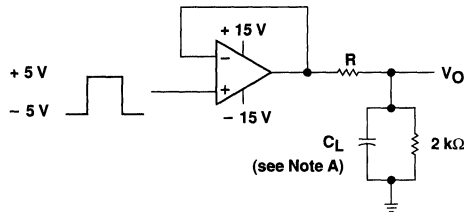


FIGURE 50. EFFECT OF CAPACITIVE LOADS



NOTE A: C_L includes fixture capacitance.

FIGURE 51. TEST CIRCUIT FOR OUTPUT CHARACTERISTICS

TL054, TL054A ENHANCED JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

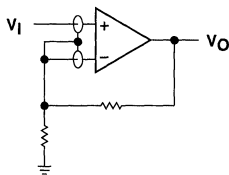
TYPICAL APPLICATION DATA

input characteristics

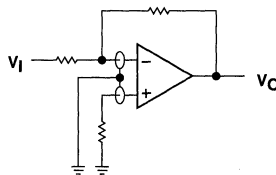
The TL054 and TL054A are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction.

Because of the extremely high input impedance and resulting low bias current requirements, the TL054 and TL054A are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause degradation in system performance. It is a good practice to include guard rings around inputs (see Figure 52). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.

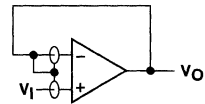
The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.



(a) NONINVERTING AMPLIFIER



(b) INVERTING AMPLIFIER



(c) UNITY-GAIN AMPLIFIER

FIGURE 52. USE OF GUARD RINGS

noise performance

The noise specifications in op amp circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TL054 and TL054A result in very low current noise. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than 50 k Ω .

TYPICAL APPLICATION DATA

Instrumentation amplifier with adjustable gain/null

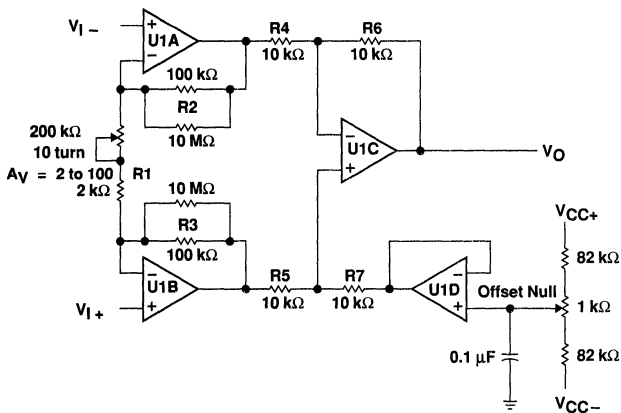
The instrumentation amplifier in Figure 53 benefits greatly from the high input impedance and stable input offset voltage of the TL054A. Amplifiers U1A, U1B, and U1C form the actual instrumentation amplifier, while U1D provides offset null. Potentiometer R1 provides gain adjust. With R1 = 2 kΩ, the circuit gain equals 100, while with R1 = 200 kΩ, the circuit gain equals two. The following equation shows the instrumentation amplifier gain as a function of R1:

$$A_V = 1 + \left(\frac{R_2 + R_3}{R_1} \right)$$

Readjusting the offset null is necessary whenever the circuit gain is changed. Note that if U1D is needed for another application, R7 can be terminated at ground. The low input offset voltage of the TL054A minimizes the dc error of the circuit. For best matching, all resistors should be one percent tolerance. The matching between R4, R5, R6, and R7 controls the CMRR of this application.

The following equation shows the output voltages when the input voltage equals zero. This dc error can be nulled by adjusting the offset null potentiometer; however, any change in offset voltage over time or temperature also creates an error. To calculate the error from changes in offset, consider the three offset components in the equation as delta offsets rather than initial offsets. The improved stability of Texas Instruments enhanced JFETs minimizes the error resulting from change in input offset voltage with time. Assuming V_{IN} equals zero, V_O can be shown as a function of the offset voltage:

$$V_O = V_{IO2} \left[\left(1 + \frac{R_3}{R_1} \right) \left(\frac{R_7}{R_5 + R_7} \right) \left(1 + \frac{R_6}{R_4} \right) + \frac{R_2}{R_1} \left(\frac{R_6}{R_4} \right) \right] - V_{IO1} \left[\frac{R_3}{R_1} \left(\frac{R_7}{R_5 + R_7} \right) \left(1 + \frac{R_6}{R_4} \right) + \frac{R_6}{R_4} \left(1 + \frac{R_2}{R_1} \right) \right] + V_{IO3} \left(1 + \frac{R_6}{R_4} \right)$$



NOTE A: U1A through U1D = TL054A; $V_{CC} \pm = \pm 15$ V.

FIGURE 53. INSTRUMENTATION AMPLIFIER

2
Operational Amplifiers

TL054, TL054A ENHANCED JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

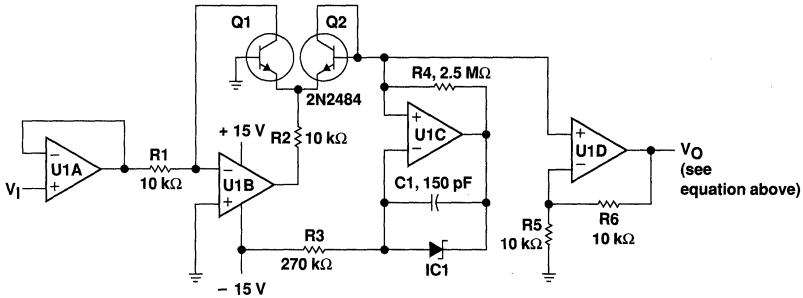
TYPICAL APPLICATION DATA

high input impedance log amplifier

The low input offset voltage and high input impedance of the TL054A create a precision log amplifier (see figure 54). IC1 is a 2.5-V, low-current precision, shunt regulator. Transistors Q1 and Q2 must be a closely matched NPN pair. For best performance over temperature, R4 should be a metal film resistor with a low temperature coefficient.

In this circuit, U1A serves as a high-impedance unity-gain buffer. Amplifier U1B converts the input voltage to a current through R1 and Q1. Amplifier U1C, IC1, and R4 form a 1 μ A temperature-stable current source that sets the base-emitter voltage of Q2. Amplifier U1D then amplifies the difference between the base-emitter voltage of Q1 and Q2. The output voltage is given by the following equation:

$$V_O = - \left[1 + \frac{R_6}{R_5} \right] \frac{kT}{q} \ln \left[\frac{V_I}{(R_1 \times 1 \times 10^{-6})} \right] \quad \text{where } k = 1.38 \times 10^{-23}, q = 1.602 \times 10^{-19}, \text{ and } T \text{ is in kelvins.}$$



NOTES: U1A thru U1D = TL054A.
IC1 = LM385, LT1004, or LT1009 voltage reference.

FIGURE 54. LOG AMPLIFIER

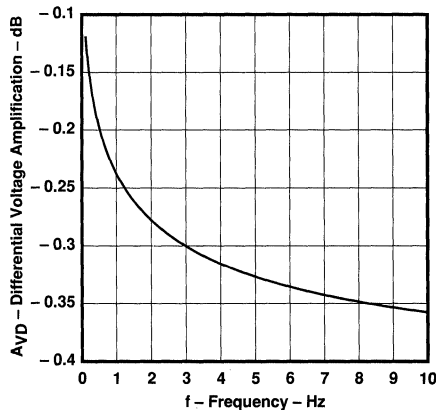


FIGURE 55. OUTPUT VOLTAGE vs INPUT VOLTAGE FOR LOG AMPLIFIER

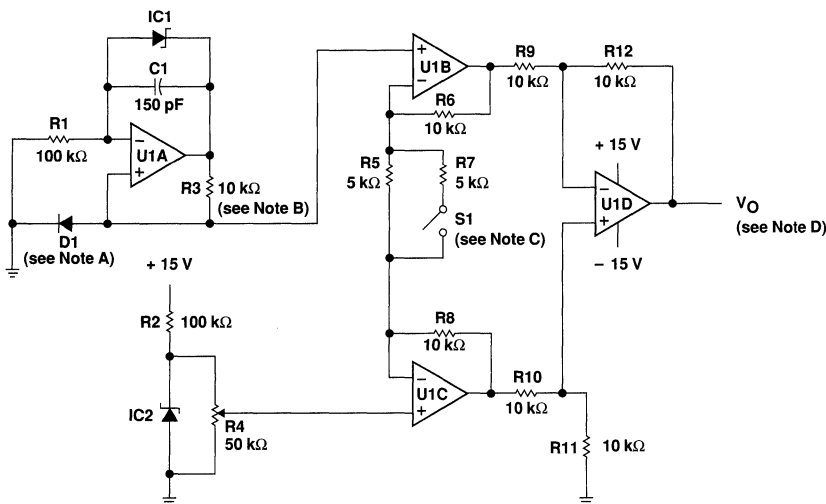
TYPICAL APPLICATION DATA

analog thermometer

By combining a current source that does not vary over temperature with an instrumentation amplifier, a precise analog thermometer can be built (see Figure 56). Amplifier U1A and IC1 establish a constant current through the temperature sensing diode D1. For this section of the circuit to operate correctly, the TL054 must use split supplies and R3 must be a metal film resistor with a low temperature coefficient.

The temperature-sensitive voltage from the diode is compared to a temperature-stable voltage reference set by IC2. R4 should be adjusted to provide the correct output voltage when the diode is at a known temperature. Although this potentiometer resistance varies with temperature, the divider ratio of the potentiometer remains constant.

Amplifiers U1B, U1C, and U1D form the instrumentation amplifier that converts the difference between the diode and reference voltage to a voltage proportional to the temperature. With switch S1 closed, the amplifier gain equals 5, and the output voltage is proportional to temperature in degrees Celsius. With S1 open, the amplifier gain is 9, and the output is proportional to temperature in degrees Fahrenheit. Every time that S1 is changed, R4 must be recalibrated. By setting S1 correctly, the output voltage equals 10 mV per degree (C or F).



- NOTES: A. Temperature sensing diode $\approx (-2 \text{ mV}/^\circ\text{C})$.
 B. Metal film (low temperature coefficient).
 C. Switch open for $^\circ\text{F}$ and closed for $^\circ\text{C}$.
 D. $V_O \propto \text{Temperature}$; 10 mV/ $^\circ\text{C}$ or 10 mV/ $^\circ\text{F}$.
 E. U1A thru U1D = TL054. IC1, IC2 = LM385, LT1004, or LT1009 voltage reference.

FIGURE 56. ANALOG THERMOMETER

TYPICAL APPLICATION DATA

voltage-ratio-to-dB converter

The application in Figure 57 measures the amplitude ratio of two signals and then converts the ratio to decibels. The output voltage provides a resolution of 100 mV/dB. The two inputs can be either dc or sinusoidal ac signals. When using ac signals, both signals should be the same frequency or output glitches will occur. For measuring two input signals of different frequencies, extra filtering should be added after the rectifiers.

The circuit contains three low-offset TL054A devices. Two of these devices provide the rectification and logarithmic conversion of the inputs. The third TL054A forms an instrumentation amplifier. The stage performing the logarithmic conversion also requires two well-matched NPN transistors.

The input signal first passes through a high impedance unity-gain buffer U1A (U2A). Then U1B (U2B) rectifies the input signal at a gain of 0.5, and U1C (U2C) provides a noninverting gain of 2 so that the system gain is still one. U1D (U2D), R6 (R13), and Q1 (Q2) perform the logarithmic conversion of the rectified input signal. The instrumentation amplifier formed by U3A, U3B, U3D scales the difference of the two logarithmic voltages by a gain of 33.6. As a result, the output voltage equals 100 mV/dB. The 1-kΩ potentiometer on the input of U3C calibrates the zero dB reference level. The following equations are used to derive the relationship between the input voltage ratio expressed in decibels and the output voltage.

$$X \text{ dB} = 20 \log \left[\frac{V_A}{V_B} \right] = 20 \left[\frac{\ln(V_A) - \ln(V_B)}{\ln(10)} \right]$$

$$X \text{ dB} = 8.686 \left[\ln(V_A) - \ln(V_B) \right]$$

$$V_{BE(Q1)} = \frac{kT}{q} \ln \left[\frac{V_A}{R \times I_S} \right] \quad V_{BE(Q2)} = \frac{kT}{q} \ln \left[\frac{V_B}{R \times I_S} \right]$$

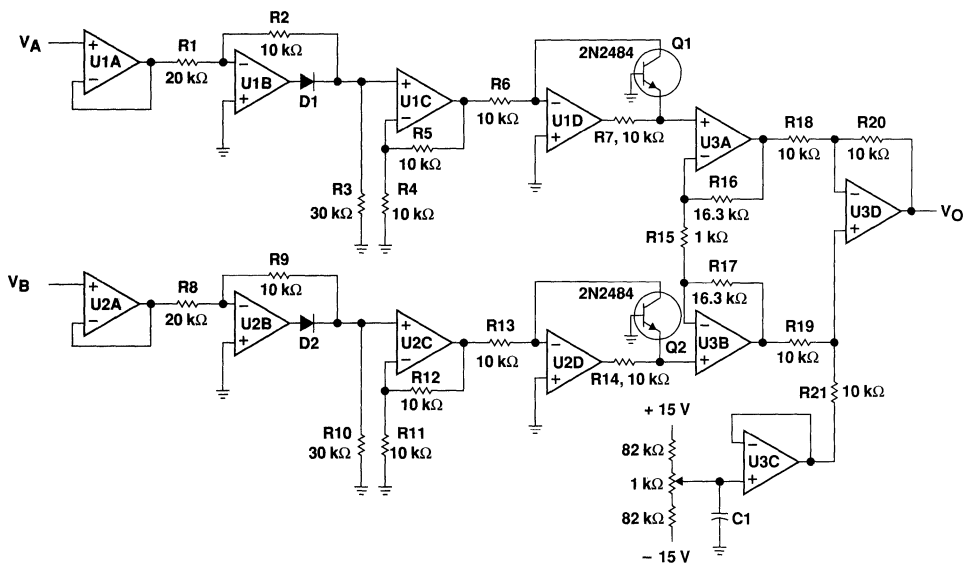
$$\Delta V_{BE} = V_{BE(Q1)} - V_{BE(Q2)} = \frac{kT}{q} \left[\ln(V_A) - \ln(V_B) \right]$$

$$X \text{ dB} = \frac{8.686}{kT/q} \left[V_{BE(Q1)} - V_{BE(Q2)} \right] = 336 \left[V_{BE(Q1)} - V_{BE(Q2)} \right] \text{ at } 25^\circ\text{C.}$$

where $k = 1.38 \times 10^{-23}$, $q = 1.602 \times 10^{-19}$, and T is in kelvins.

This would give a resolution of 1 V/dB. Therefore, the gain of the instrumentation amplifier is set at 33.6 to obtain 100 mV/dB.

TYPICAL APPLICATION DATA



NOTES: U1A through U3D = TL054A, $V_{CC} \pm = \pm 15$ V.
 D1 and D2 = 1N914.

FIGURE 57. VOLTAGE-RATIO-TO-dB CONVERTER

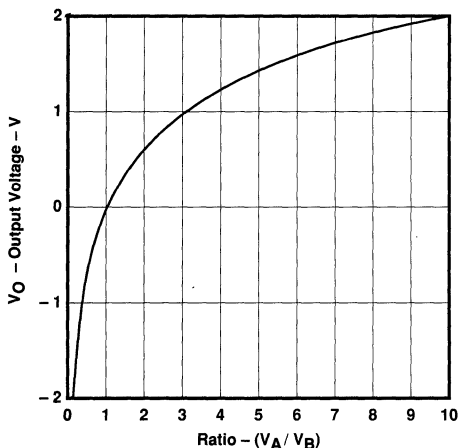


FIGURE 58. OUTPUT VOLTAGE vs THE RATIO OF THE INPUT VOLTAGES FOR VOLTAGE-TO-dB CONVERTER

2

Operational Amplifiers

TL060, TL060A, TL060B, TL061, TL061A, TL061B TL062, TL062A, TL062B, TL064, TL064A, TL064B LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

D2392, NOVEMBER 1978—REVISED NOVEMBER 1988

20 DEVICES COVER MILITARY, INDUSTRIAL, AND COMMERCIAL TEMPERATURE RANGES

- Very Low Power Consumption
- Typical Supply Current . . . 200 μ A (per Amplifier)
- Wide Common-Mode and Differential Voltage Ranges
- Low Input Bias and Offset Currents
- Common-Mode Input Voltage Range Includes V_{CC+}
- Output Short-Circuit Protection
- High Input Impedance . . . JFET-Input Stage
- Internal Frequency Compensation (Except TL060)
- Latch-Up-Free Operation
- High Slew Rate . . . 3.5 V/ μ s Typ

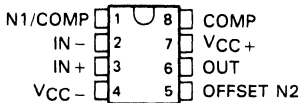
2

description

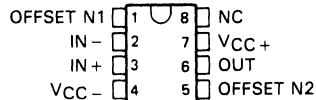
The JFET-input operational amplifiers of the TL061 series are designed as low-power versions of the TL081 series amplifiers. They feature high input impedance, wide bandwidth, high slew rate, and low input offset and bias currents. The TL061 series features the same terminal assignments as the TL071 and TL081 series. Each of these JFET-input operational amplifiers incorporates well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit.

M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . I-suffix devices are characterized for operation from -40°C to 85°C , and C-suffix devices are characterized for operation from 0°C to 70°C .

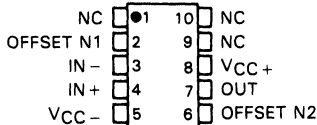
TL060, TL060A, TL060B
D, JG, OR P PACKAGE
(TOP VIEW)



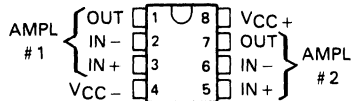
TL061, TL061A, TL061B
D, JG, OR P PACKAGE
(TOP VIEW)



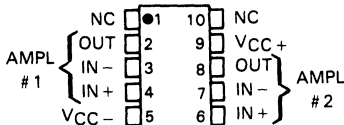
TL061 . . . U PACKAGE
(TOP VIEW)



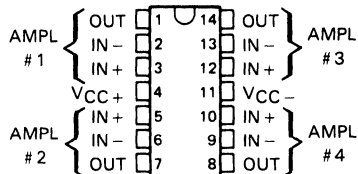
TL062, TL062A, TL062B
D, JG, OR P PACKAGE
(TOP VIEW)



TL062 . . . U PACKAGE
(TOP VIEW)



TL064 . . . D, J, N, OR W PACKAGE
TL064A, TL064B . . . D, J, OR N PACKAGE
(TOP VIEW)



Operational Amplifiers

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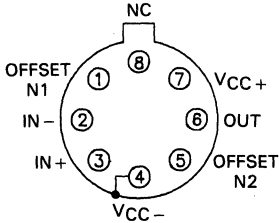
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**TL060, TL060A, TL060B, TL061, TL061A, TL061B
TL062, TL062A, TL064, TL064A, TL064B
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

2

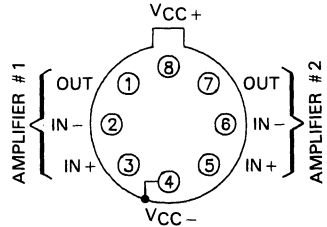
Operational Amplifiers

**TL061 . . . L PACKAGE
(TOP VIEW)**



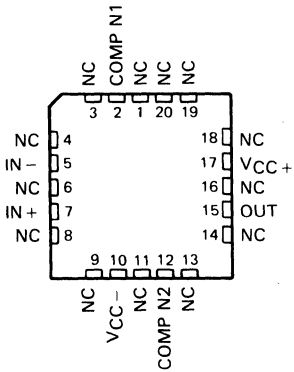
PIN 4 IS IN ELECTRICAL CONTACT WITH THE CASE

**TL062 . . . L PACKAGE
(TOP VIEW)**

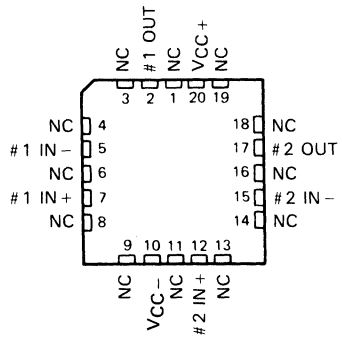


PIN 4 IS IN ELECTRICAL CONTACT WITH THE CASE

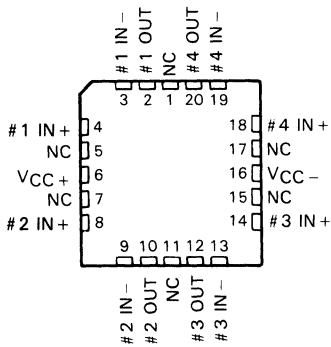
**TL061 . . . FK PACKAGE
(TOP VIEW)**



**TL062 . . . FK PACKAGE
(TOP VIEW)**



**TL064 . . . FK PACKAGE
(TOP VIEW)**



NC—No internal connection

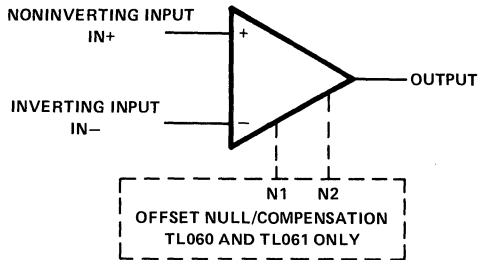
**TL060, TL060A, TL060B, TL061, TL061A, TL061B
TL062, TL062A, TL062B, TL064, TL064A, TL064B
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

AVAILABLE OPTIONS

T _A	V _{IO} MAX at 25°C	PACKAGE								
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	CERAMIC DIP (JG)	PLUG- IN (L)	PLASTIC DIP (N)	PLASTIC DIP (P)	FLAT PACK (U)	FLAT PACK (W)
0°C to 70°C	15 mV	TL060CD			TL060CJG			TL060CP		
	6 mV	TL060ACD			TL060ACJG			TL060ACP		
	3 mV	TL060BCD			TL060BCJG			TL060BCP		
	15 mV	TL061CD			TL061CJG			TL061CP		
	6 mV	TL061ACD			TL061ACJG			TL061ACP		
	3 mV	TL061BCD			TL061BCJG			TL061BCP		
	15 mV	TL062CD			TL062CJG			TL062CP		
	6 mV	TL062ACD			TL062ACJG			TL062ACP		
	3 mV	TL062BCD			TL062BCJG			TL062BCP		
	15 mV	TL064CD		TL064CJ			TL064CN			
	6 mV	TL064ACD		TL064ACJ			TL064ACN			
	3 mV	TL064BCD		TL064BCJ			TL064BCN			
-40°C to 85°C	6 mV	TL060ID			TL060IJG			TL060IP		
	6 mV	TL061ID			TL061IJG			TL061IP		
	6 mV	TL062ID			TL062IJG			TL062IP		
	6 mV	TL064ID		TL064IJ			TL064IN			
-55°C to 125°C	6 mV		TL061MFK		TL061MJG	TL061ML			TL061MU	
	6 mV		TL062MFK		TL062MJG	TL062ML			TL062MU	
	9 mV		TL064MFK	TL064MJ						TL064MW

The D package is available taped and reeled. Add the suffix R to the device type (e.g., TL061CDR).

symbol (each amplifier)



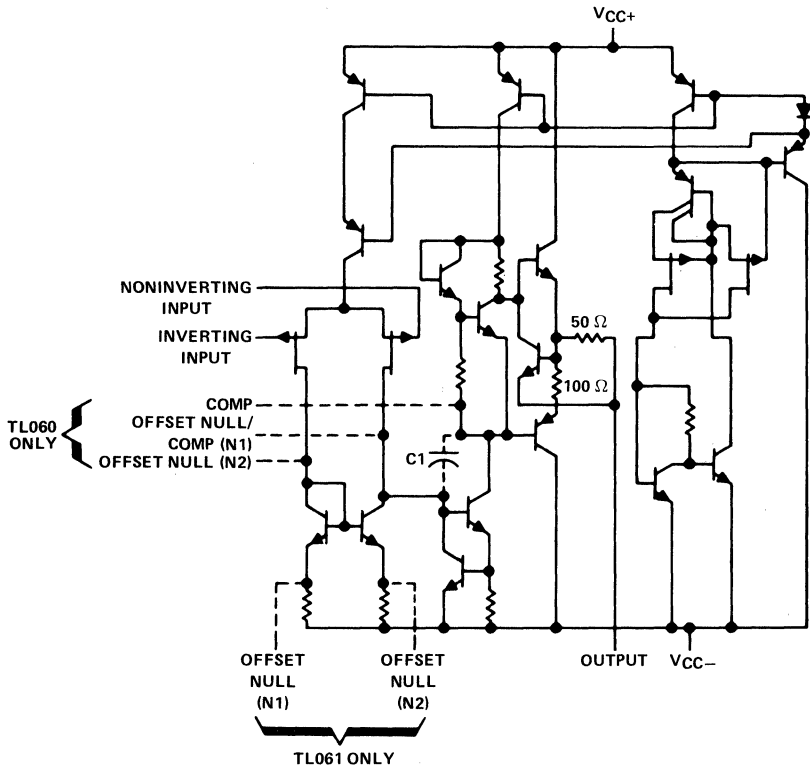
2
Operational Amplifiers

**TL060, TL060A, TL060B, TL061, TL061A, TL061B
 TL062, TL062A, TL064, TL064A, TL064B
 LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

schematic (each amplifier)

2

Operational Amplifiers



C1 = 10 pF ON TL061, TL062, AND TL064 ONLY
 COMPONENT VALUES SHOWN ARE NOMINAL

**TL060, TL060A, TL060B, TL061, TL061A, TL061B
TL062, TL062A, TL062B, TL064, TL064A, TL064B
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	TL06_M	TL06_I	TL06_C, TL06_AC, TL06_BC	UNIT
Supply voltage, V_{CC+} (see Note 1)	18	18	18	V
Supply voltage, V_{CC-} (see Note 1)	-18	-18	-18	V
Differential input voltage (see Note 2)	± 30	± 30	± 30	V
Input voltage (see Notes 1 and 3)	± 15	± 15	± 15	V
Voltage between power-control terminal and V_{CC-}	± 0.5	± 0.5	± 0.5	V
Duration of output short circuit (see Note 4)	unlimited	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table			
Operating free-air temperature range	-55 to 125	-40 to 85	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	°C
Case temperature for 60 seconds	FK package	260		°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J, JG, U or W package	300	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D, N or P package		260	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	L package	300		°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
 4. The output may be shorted to ground or to either supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D (8-pin)	680 mW	5.8 mW/°C	33°C	464 mW	377 mW	N/A
D (14-pin)	680 mW	7.6 mW/°C	60°C	608 mW	494 mW	N/A
FK	680 mW	11.0 mW/°C	88°C	680 mW	680 mW	275 mW
J (TL06_M)	680 mW	11.0 mW/°C	88°C	680 mW	680 mW	275 mW
J (all others)	680 mW	8.2 mW/°C	67°C	656 mW	533 mW	N/A
JG (TL06_M)	680 mW	8.4 mW/°C	69°C	672 mW	546 mW	210 mW
JG (all others)	680 mW	6.6 mW/°C	47°C	528 mW	429 mW	N/A
L	680 mW	6.6 mW/°C	47°C	528 mW	429 mW	165 mW
N	680 mW	9.2 mW/°C	76°C	680 mW	598 mW	N/A
P	680 mW	8.0 mW/°C	65°C	640 mW	520 mW	N/A
U	675 mW	5.4 mW/°C	25°C	432 mW	351 mW	135 mW
W	680 mW	8.0 mW/°C	65°C	640 mW	520 mW	200 mW

TL061M, TL062M, TL064M
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TL061M TL062M			TL064M			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0,$ $R_S = 50\ \Omega,$ $T_A = 25^\circ\text{C}$ $T_A = -55^\circ\text{C to } 125^\circ\text{C}$	3 6			3 9			mV
					15			
αV_{IO} Temperature coefficient of input offset voltage	$V_O = 0,$ $R_S = 50\ \Omega,$ $T_A = -55^\circ\text{C to } 125^\circ\text{C}$	10			10			$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current‡	$V_O = 0$	$T_A = 25^\circ\text{C}$ $T_A = -55^\circ\text{C to } 125^\circ\text{C}$			5 100 20			pA nA
I_{IB} Input bias current	$V_O = 0$	$T_A = 25^\circ\text{C}$ $T_A = -55^\circ\text{C to } 125^\circ\text{C}$			30 200 50			pA nA
V_{ICR} Common-mode input voltage range	$T_A = 25^\circ\text{C}$	-12 to +15			-12 to +15			V
V_{OM} Maximum peak output voltage swing	$R_L = 10\ \text{k}\Omega,$ $T_A = 25^\circ\text{C}$ $R_L \geq 10\ \text{k}\Omega,$ $T_A = -55^\circ\text{C to } 125^\circ\text{C}$	$\pm 10 \pm 13.5$			$\pm 10 \pm 13.5$			V
		± 10			± 10			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 10\ \text{V},$ $R_L \geq 10\ \text{k}\Omega$	$T_A = 25^\circ\text{C}$ $T_A = -55^\circ\text{C to } 125^\circ\text{C}$			4 6 4			V/mV
B_1 Unity-gain bandwidth	$R_L = 10\ \text{k}\Omega,$ $T_A = 25^\circ\text{C}$							MHz
r_i Input resistance	$T_A = 25^\circ\text{C}$	10^{12}			10^{12}			Ω
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR}\ \text{min},$ $V_O = 0,$ $R_S = 50\ \Omega,$ $T_A = 25^\circ\text{C}$	80 86			80 86			dB
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC\pm}/\Delta V_{IO}$)	$V_{CC} = \pm 15\ \text{V to } \pm 9\ \text{V},$ $V_O = 0,$ $R_S = 50\ \Omega,$ $T_A = 25^\circ\text{C}$	80 95			80 95			dB
P_D Total power dissipation (each amplifier)	No load, $V_O = 0,$ $T_A = 25^\circ\text{C}$	6 7.5			6 7.5			mW
I_{CC} Supply current (each amplifier)	No load, $V_O = 0,$ $T_A = 25^\circ\text{C}$	200 250			200 250			μA
V_{O1}/V_{O2} Crosstalk attenuation	$A_{VD} = 100,$ $T_A = 25^\circ\text{C}$	120			120			dB

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 17. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

2

Operational Amplifiers

electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS [†]		TL060J			TL060C			TL060AC			TL060BC			UNIT
			TL061I			TL061C			TL061AC			TL061BC			
			TL062I			TL062C			TL062AC			TL062BC			
TL064I			TL064C			TL064AC			TL064BC						
	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX			
V_{IO} Input offset voltage	$V_O = 0,$ $R_S = 50\ \Omega$	$T_A = 25^\circ\text{C}$	3	6	3	15	3	6	2	3			mV		
		$T_A = \text{full range}$		9		20		7.5		5					
α_{VIO} Temperature coefficient of input offset voltage	$V_O = 0,$ $R_S = 50\ \Omega,$ $T_A = \text{full range}$		10		10		10		10		10		$\mu\text{V}/^\circ\text{C}$		
I_{IO} Input offset current [‡]	$V_O = 0$	$T_A = 25^\circ\text{C}$	5	100	5	200	5	100	5	100	5	100	pA		
		$T_A = \text{full range}$		10		5		3		3		3	nA		
I_{IB} Input bias current [‡]	$V_O = 0$	$T_A = 25^\circ\text{C}$	30	200	30	400	30	200	30	200	30	200	pA		
		$T_A = \text{full range}$		20		10		7		7		7	nA		
V_{ICR} Common-mode input voltage range	$T_A = 25^\circ\text{C}$		± 11.5	-12 to $+15$	± 11	-12 to $+15$	± 11.5	-12 to $+15$	± 11.5	-12 to $+15$	± 11.5	-12 to $+15$	V		
V_{OM} Maximum peak output voltage swing	$R_L = 10\ \text{k}\Omega,$ $T_A = 25^\circ\text{C}$		± 10	± 13.5	± 10	± 13.5	± 10	± 13.5	± 10	± 13.5	± 10	± 13.5	V		
	$R_L \geq 10\ \text{k}\Omega,$ $T_A = \text{full range}$		± 10		± 10		± 10		± 10		± 10				
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 10\ \text{V},$ $R_L \geq 10\ \text{k}\Omega$	$T_A = 25^\circ\text{C}$	4	6	3	6	4	6	4	6	4	6	V/mV		
		$T_A = \text{full range}$		4		3		4		4		4			
B_1 Unity-gain bandwidth	$R_L = 10\ \text{k}\Omega,$ $T_A = 25^\circ\text{C}$		1		1		1		1		1		MHz		
r_i Input resistance	$T_A = 25^\circ\text{C}$		10^{12}		10^{12}		10^{12}		10^{12}		10^{12}		Ω		
$CMRR$ Common-mode rejection ratio	$V_{IC} = V_{ICR\ \text{min}},$ $V_O = 0,$ $R_S = 50\ \Omega,$ $T_A = 25^\circ\text{C}$		80	86	70	86	80	86	80	86	80	86	dB		
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC\pm}/\Delta V_{IO}$)	$V_{CC} = \pm 15\ \text{V}$ to $\pm 9\ \text{V},$ $V_O = 0,$ $R_S = 50\ \Omega,$ $T_A = 25^\circ\text{C}$		80	95	70	95	80	95	80	95	80	95	dB		
P_D Total power dissipation (each amplifier)	No load, $V_O = 0,$ $T_A = 25^\circ\text{C}$		6	7.5	6	7.5	6	7.5	6	7.5	6	7.5	mW		
I_{CC} Supply current (each amplifier)	No load, $V_O = 0,$ $T_A = 25^\circ\text{C}$		200	250	200	250	200	250	200	250	200	250	μA		
V_{O1}/V_{O2} Crosstalk attenuation	$A_{VD} = 100,$ $T_A = 25^\circ\text{C}$		120		120		120		120		120		dB		

[†] All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified. Full range for T_A is -40°C to 85°C for TL06_1I and 0°C to 70°C for TL06_C, TL06_AC, and TL06_BC.

[‡] Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 17. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

**TL060, TL060A, TL060B, TL061, TL061A, TL061B
TL062, TL062A, TL062B, TL064, TL064A, TL064B
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

operating characteristics, $V_{CC\pm} = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	$V_I = 10\text{ V}$, $R_L = 10\text{ k}\Omega$, $C_L = 100\text{ pF}$, See Figure 1	1.5	3.5		$\text{V}/\mu\text{s}$
t_r Rise time	$V_I = 20\text{ mV}$, $R_L = 10\text{ k}\Omega$		0.2		μs
Overshoot factor	$C_L = 100\text{ pF}$, See Figure 1		10%		
V_n Equivalent input noise voltage	$R_S = 100\ \Omega$, $f = 1\text{ kHz}$		42		$\text{nV}/\sqrt{\text{Hz}}$

PARAMETER MEASUREMENT INFORMATION

2

Operational Amplifiers

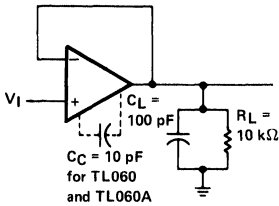


FIGURE 1. UNITY-GAIN AMPLIFIER

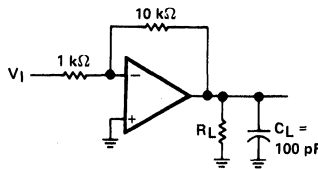


FIGURE 2. GAIN-OF-10
INVERTING AMPLIFIER

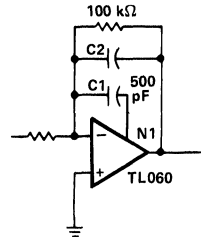
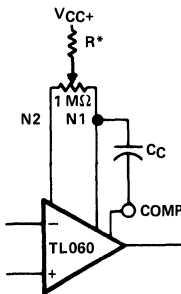


FIGURE 3. FEED-FORWARD
COMPENSATION

INPUT OFFSET VOLTAGE NULL CIRCUITS



*For best results use $R = 20\text{ M}\Omega$ for $V_{CC\pm} = \pm 15\text{ V}$ to $R = 5\text{ M}\Omega$ for $V_{CC\pm} = \pm 3\text{ V}$.

FIGURE 4

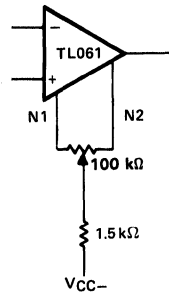


FIGURE 5

TL060, TL060A, TL060B, TL061, TL061A, TL061B TL062, TL062A, TL062B, TL064, TL064A, TL064B LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE
vs
SUPPLY VOLTAGE

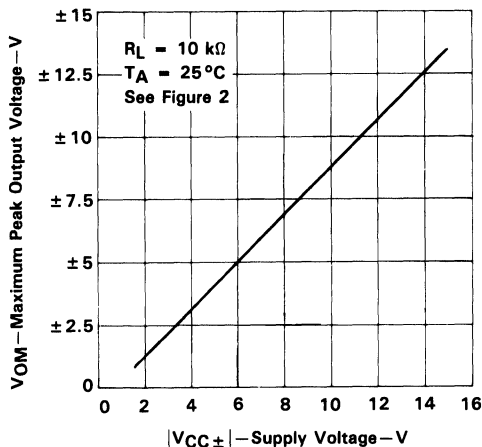


FIGURE 6

MAXIMUM PEAK OUTPUT VOLTAGE
vs
FREE-AIR TEMPERATURE

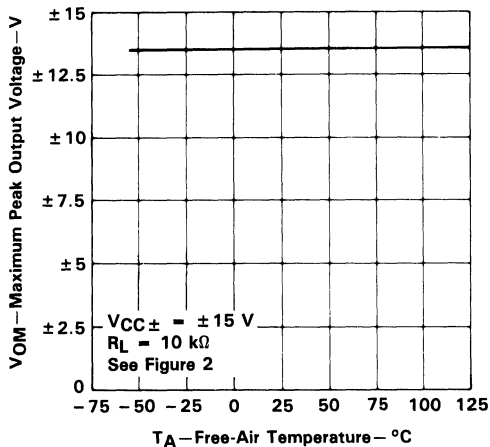


FIGURE 7

MAXIMUM PEAK OUTPUT VOLTAGE
vs
LOAD RESISTANCE

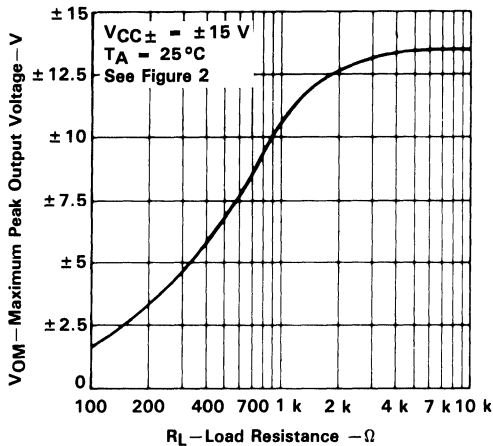


FIGURE 8

MAXIMUM PEAK OUTPUT VOLTAGE
vs
FREQUENCY

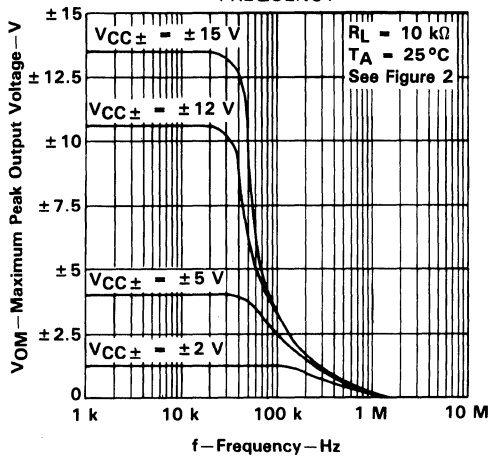


FIGURE 9

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. A 10-pF compensation capacitor is used with TL060 and TL060A.

**TL060, TL060A, TL060B, TL061, TL061A, TL062B
TL062, TL062A, TL062B, TL064, TL064A, TL064B
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS†

**DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREE-AIR TEMPERATURE**

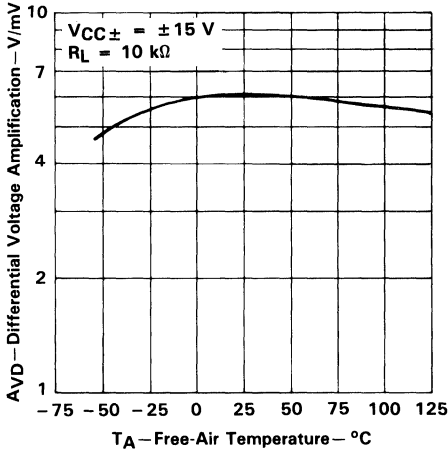


FIGURE 10

**LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
AND PHASE SHIFT
vs
FREQUENCY**

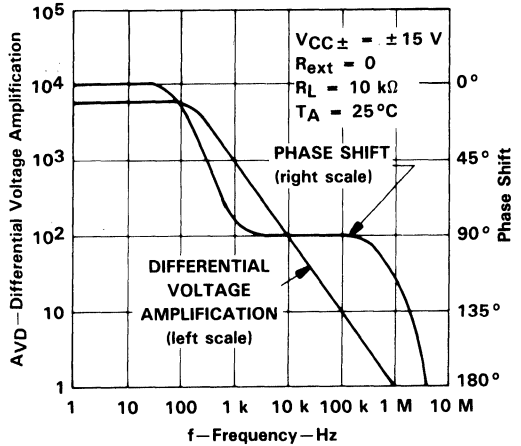


FIGURE 11

**SUPPLY CURRENT
vs
SUPPLY VOLTAGE**

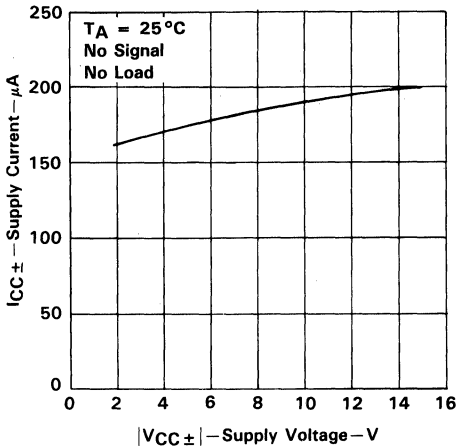


FIGURE 12

**SUPPLY CURRENT
vs
FREE-AIR TEMPERATURE**

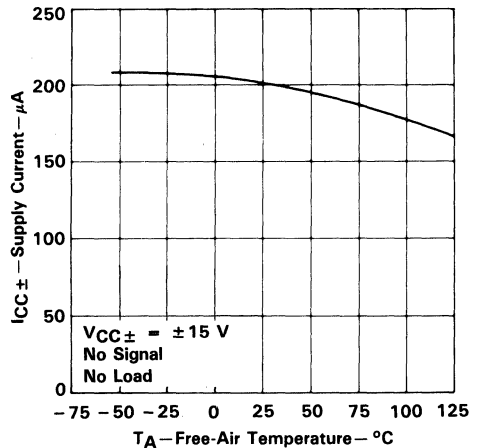


FIGURE 13

†A 10-pF compensation capacitor is used with TL060 and TL060A.

TL060, TL060A, TL060B, TL061, TL061A, TL061B TL062, TL062A, TL062B, TL064, TL064A, TL064B LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

TOTAL POWER DISSIPATED
vs
FREE-AIR TEMPERATURE

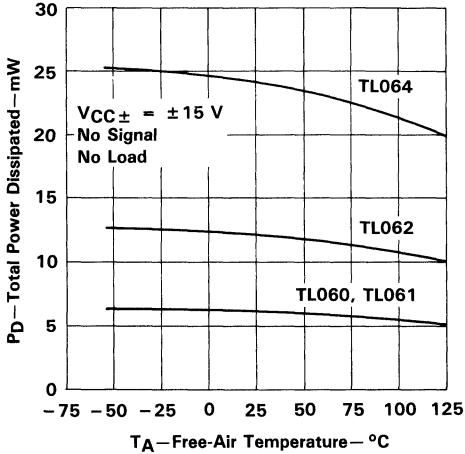


FIGURE 14

ALL EXCEPT TL06_C
COMMON-MODE REJECTION RATIO
vs
FREE-AIR TEMPERATURE

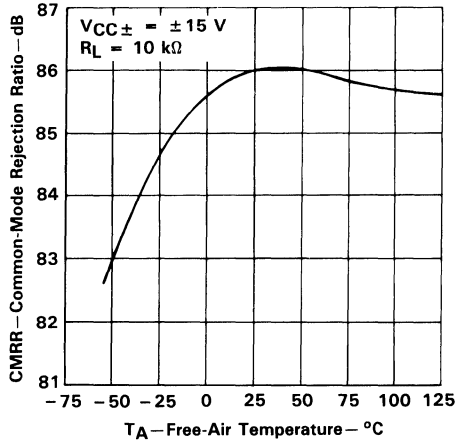


FIGURE 15

NORMALIZED UNITY GAIN BANDWIDTH
SLEW RATE, AND PHASE SHIFT
vs
FREE-AIR TEMPERATURE

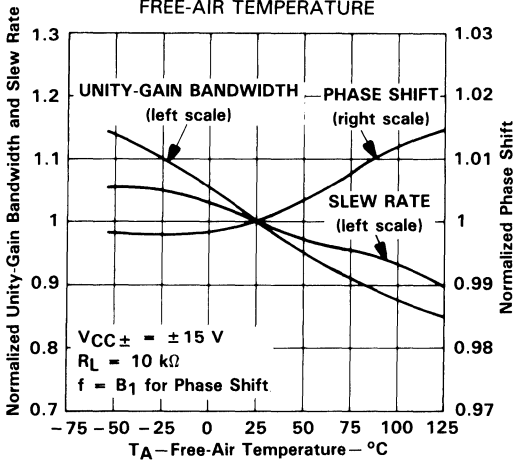


FIGURE 16

INPUT BIAS CURRENT
vs
FREE-AIR TEMPERATURE

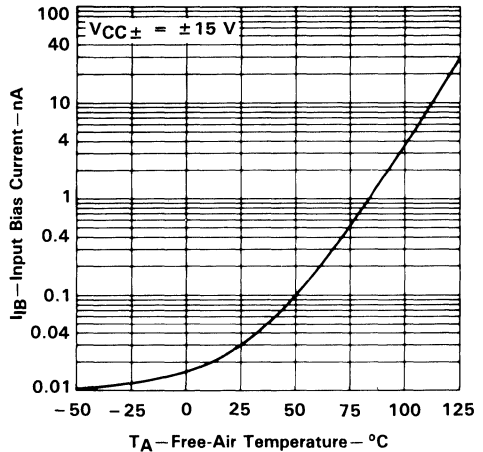


FIGURE 17

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. A 10-pF compensation capacitor is used with TL060 and TL060A.

**TL060, TL060A, TL060B, TL061, TL061A, TL061B
TL062, TL062A, TL062B, TL064, TL064A, TL064B
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS†

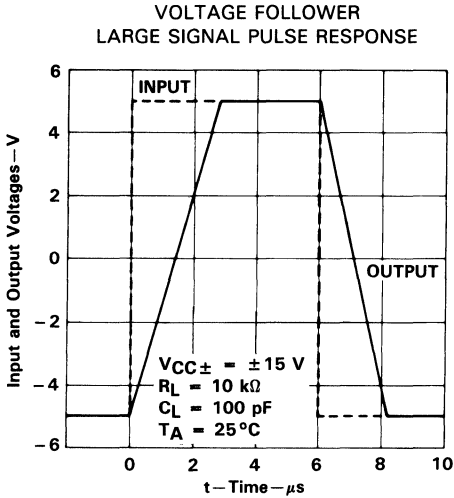


FIGURE 18

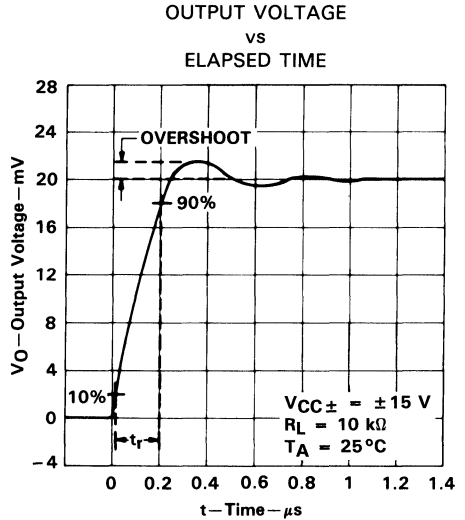


FIGURE 19

EQUIVALENT INPUT NOISE VOLTAGE
vs
FREQUENCY

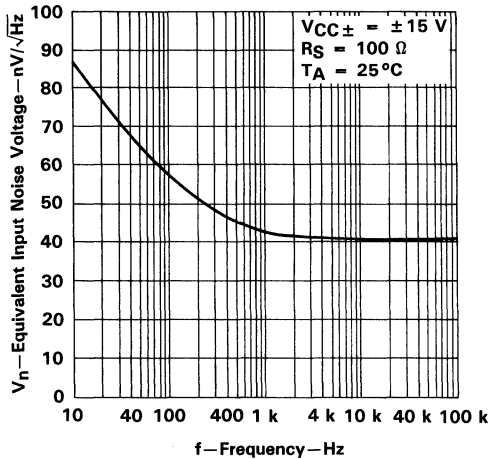


FIGURE 20

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. A 10-pF compensation capacitor is used with TL060 and TL060A.

**TL060, TL060A, TL060B, TL061, TL061A, TL061B
TL062, TL062A, TL062B, TL064, TL064A, TL064B
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL APPLICATION DATA

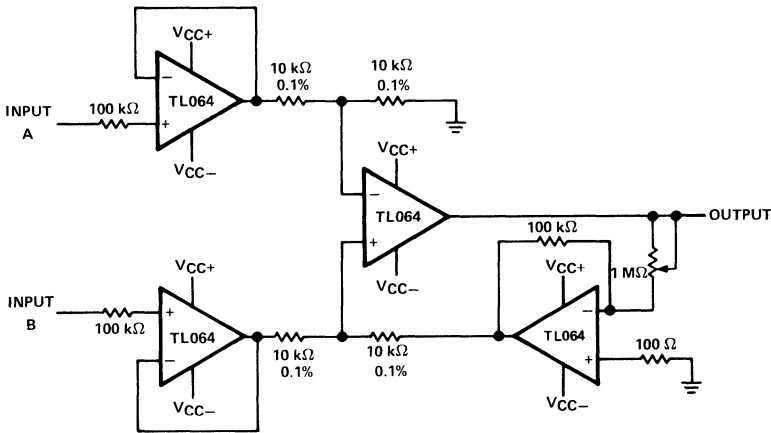


FIGURE 21. INSTRUMENTATION AMPLIFIER

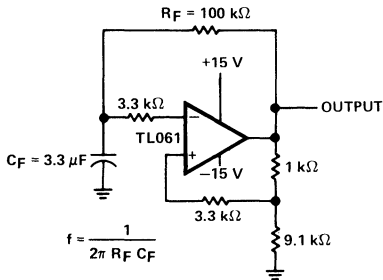


FIGURE 22. 0.5-Hz SQUARE-WAVE OSCILLATOR

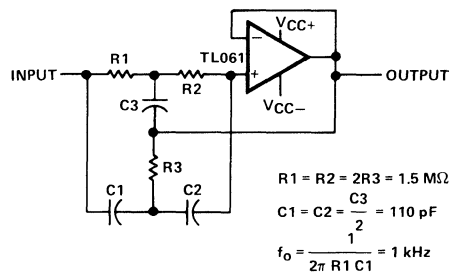


FIGURE 23. HIGH-Q NOTCH FILTER

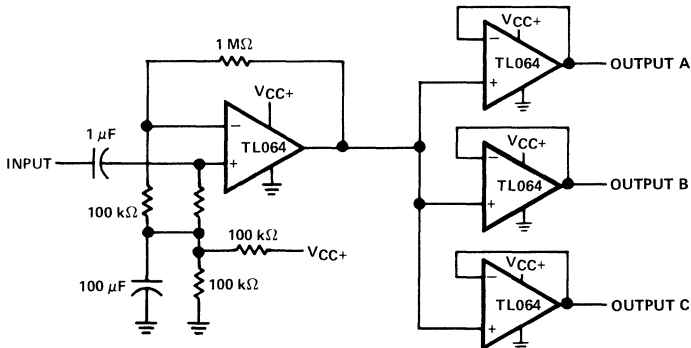


FIGURE 24. AUDIO DISTRIBUTION AMPLIFIER

**TL060, TL060A, TL060B, TL061, TL061A, TL062B
TL062, TL062A, TL062B, TL064, TL064A, TL064B
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL APPLICATION DATA

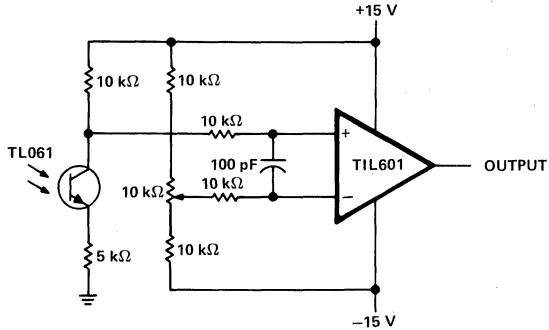


FIGURE 25. LOW-LEVEL LIGHT DETECTOR PREAMPLIFIER

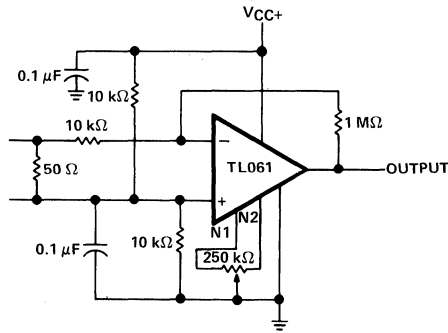


FIGURE 26. AC AMPLIFIER

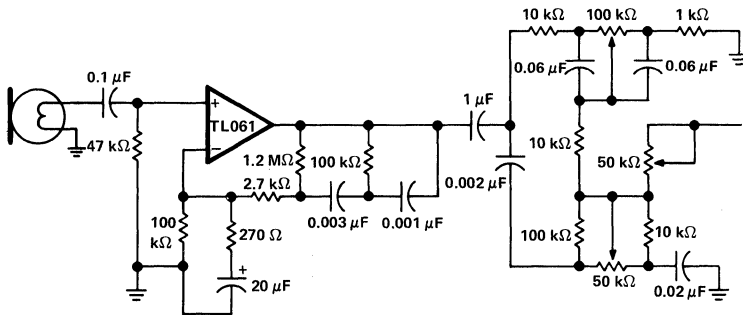


FIGURE 27. MICROPHONE PREAMPLIFIER WITH TONE CONTROL

**TL060, TL060A, TL060B, TL061, TL061A, TL061B
TL062, TL062A, TL062B, TL064, TL064A, TL064B
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL APPLICATION DATA

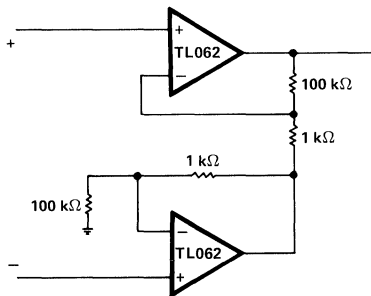


FIGURE 28. INSTRUMENTATION AMPLIFIER

IC PREAMPLIFIER RESPONSE CHARACTERISTICS

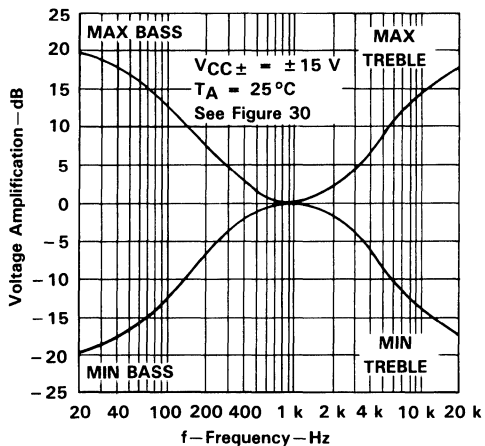


FIGURE 29

**TL060, TL060A, TL060B, TL061, TL061A, TL061B
 TL062, TL062A, TL062B, TL064, TL064A, TL064B
 LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL APPLICATION DATA

2

Operational Amplifiers

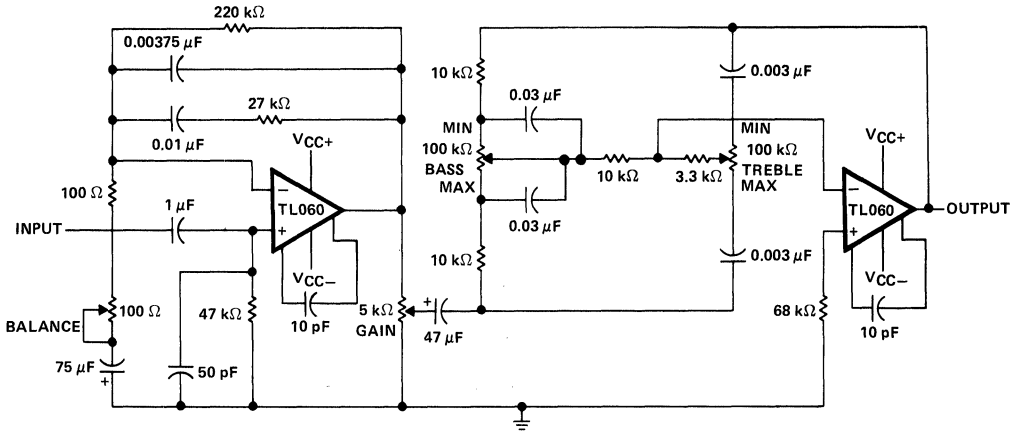


FIGURE 30. IC PREAMPLIFIER

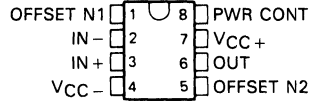
TL066M, TL066I, TL066C, TL066AC, TL066BC ADJUSTABLE LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

D2494, FEBRUARY 1979—REVISED NOVEMBER 1988

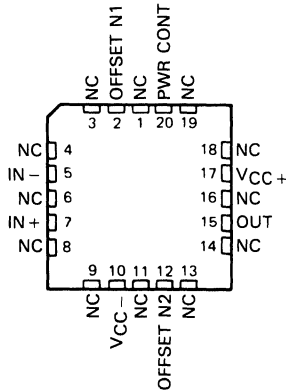
5 DEVICES COVER COMMERCIAL, INDUSTRIAL, AND MILITARY TEMPERATURE RANGES

- Very Low, Adjustable ("Programmable") Power Consumption
- Adjustable Supply Current . . . 5 μA to 200 μA
- Very Low Input Bias and Offset Currents
- Wide Supply Range . . . $\pm 1.2\text{ V}$ to $\pm 18\text{ V}$
- Wide Common-Mode and Differential Voltage Range
- Output Short-Circuit Protection
- High Input Impedance . . . JFET-Input Stage
- Unity-Gain Bandwidth . . . 1 MHz Typ (100 kHz at 25 μW)
- High Slew Rate . . . 3.5 V/ μs Typ
- Internal Frequency Compensation
- Latch-Up-Free Operation
- Common-Mode Input Voltage Range Includes V_{CC+}

TL066M . . . JG PACKAGE
TL066I, TL066C, TL066AC, TL066BC . . . D, JG, OR P PACKAGE
(TOP VIEW)



TL066M . . . FK PACKAGE
(TOP VIEW)



NC—No internal connection

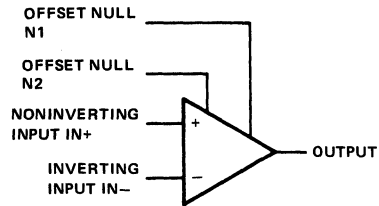
TL066M IS NOT RECOMMENDED FOR NEW DESIGNS

description

The TL066, TL066A, and TL066B are JFET-input operational amplifiers similar to the TL061 with the additional feature of being power-adjustable. They feature very low input offset and bias currents, high input impedance, wide bandwidth, and high slew rate. The power-control feature permits the amplifiers to be adjusted to require as little as 25 μW of power. This type of amplifier, which provides for changing several characteristics by varying one external element, is sometimes referred to as being "programmable." The JFET-input stage combined with the adjustable-low-power feature results in superior bandwidth and slew-rate performance compared to low-power bipolar-input devices.

The TL066M is characterized for operation over the full military temperature range of -55°C to 125°C ; the TL066I is characterized for operation from -40°C to 85°C ; the TL066C, TL066AC, and TL066BC are characterized for operation from 0°C to 70°C .

symbol



PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS
INSTRUMENTS

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TL066M, TL066I, TL066C, TL066AC, TL066BC ADJUSTABLE LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

AVAILABLE OPTIONS

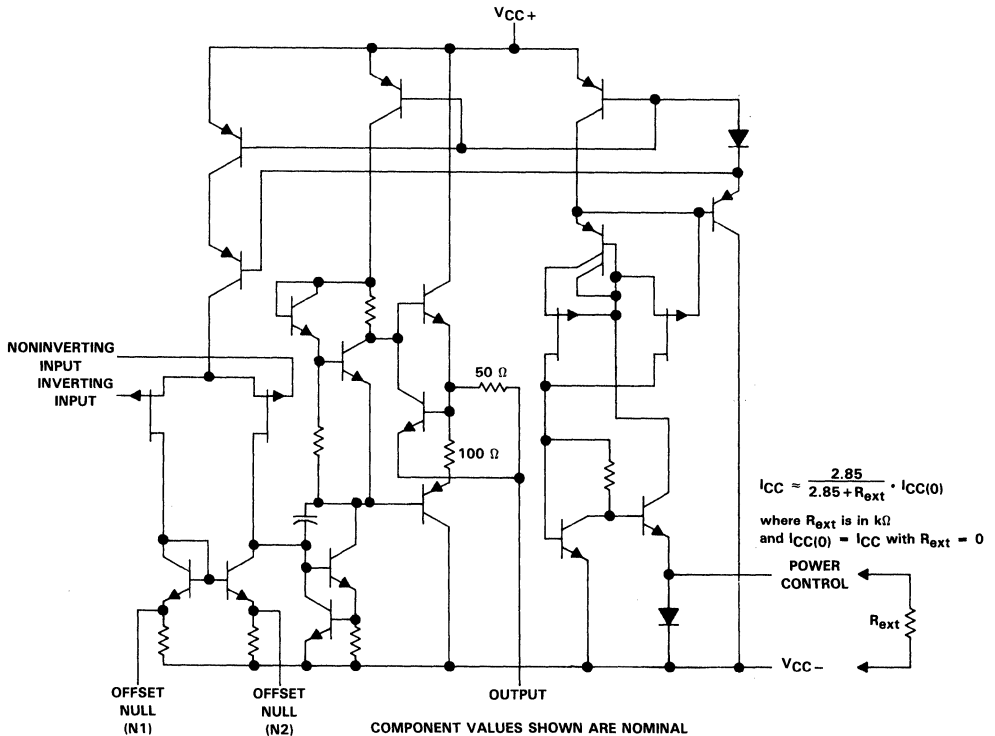
T _A	V _{IO} MAX AT 25°C	PACKAGE			
		SMALL-OUTLINE (D)	CHIP-CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	15 mV 6 mV 3 mV	TL066CD TL066ACD TL066BCD		TL066CJG TL066ACJG TL066BCJG	TL066CP TL066ACP TL066BCP
-40°C to 85°C	6 mV	TL066ID		TL066IJG	TL066IP
-55°C to 125°C	6 mV		TL066MFK	TL066MJG	

The D package is available taped and reeled. Add the suffix "R" to the device type (e.g., TL066CDR).

2

Operational Amplifiers

schematic



TL066M, TL066I, TL066C, TL066AC, TL066BC

ADJUSTABLE LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	TL066M	TL066I	TL066C, TL066AC, TL066BC	UNIT
Supply voltage, V_{CC+} (see Note 1)	18	18	18	V
Supply voltage, V_{CC-} (see Note 1)	-18	-18	-18	V
Differential input voltage (see Note 2)	± 30	± 30	± 30	V
Input voltage (see Notes 1 and 3)	± 15	± 15	± 15	V
Voltage between power-control terminal and V_{CC-}	± 0.5	± 0.5	± 0.5	V
Duration of output short circuit (see Note 4)	unlimited	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table			
Operating free-air temperature range	-55 to 125	-40 to 85	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	°C
Case temperature for 60 seconds	FK package	260		°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	JG package	300	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package		260	°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
 4. The output may be shorted to ground or to either supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING			POWER RATING	POWER RATING	POWER RATING
D	680 mW	5.8 mW/°C	33°C	464 mW	377 mW	N/A
FK	680 mW	11.0 mW/°C	88°C	680 mW	680 mW	275 mW
JG (TL066M)	680 mW	8.4 mW/°C	69°C	672 mW	546 mW	210 mW
JG (all others)	680 mW	6.6 mW/°C	47°C	528 mW	429 mW	N/A
P	680 mW	8.0 mW/°C	65°C	640 mW	520 mW	N/A

2

Operational Amplifiers

electrical characteristics, $V_{CC} = \pm 15 \text{ V}$

PARAMETER	TEST CONDITIONS†	TL066M			TL066I			TL066C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0$, $R_S = 50 \Omega$, $T_A = 25^\circ\text{C}$		3	6		3	6		3	15	mV
	$V_O = 0$, $R_S = 50 \Omega$, $T_A = \text{full range}$			9			9			20	
αV_{IO} Temperature coefficient of input offset voltage	$V_O = 0$, $R_S = 50 \Omega$, $T_A = \text{full range}$		10			10			10		$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current‡	$V_O = 0$, $T_A = 25^\circ\text{C}$		5	100		5	100		5	200	pA
	$V_O = 0$, $T_A = \text{full range}$			20			10			5	nA
I_{IB} Input bias current‡	$V_O = 0$, $T_A = 25^\circ\text{C}$		30	200		30	200		30	400	pA
	$V_O = 0$, $T_A = \text{full range}$			50			20			10	nA
V_{ICR} Common-mode input voltage range	$T_A = 25^\circ\text{C}$	± 11.5	-12 to +15		± 11.5	-12 to +15		± 11	-12 to +15		V
V_{OM} Maximum peak output voltage swing	$T_A = 25^\circ\text{C}$, $R_L \geq 10 \text{ k}\Omega$		± 10	± 13.5		± 10	± 13.5		± 10	± 13.5	V
	$T_A = \text{full range}$, $R_L \geq 10 \text{ k}\Omega$		± 10	± 13.5		± 10	± 13.5		± 10	± 13.5	
A_{VD} Large-signal differential voltage amplification	$R_L \geq 10 \text{ k}\Omega$, $V_O = \pm 10 \text{ V}$, $T_A = 25^\circ\text{C}$		4	6		4	6		3	6	V/mV
	$R_L \geq 10 \text{ k}\Omega$, $V_O = \pm 10 \text{ V}$, $T_A = \text{full range}$		4						3		
B_1 Unity-gain bandwidth	$T_A = 25^\circ\text{C}$, $R_L = 10 \text{ k}\Omega$		1			1			1		MHz
r_i Input resistance	$T_A = 25^\circ\text{C}$		10	12		10	12		10	12	Ω
r_o Output resistance	$T_A = 25^\circ\text{C}$, $f = 1 \text{ kHz}$		220			220			220		Ω
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR \text{ min}}$, $V_O = 0$, $R_S = 50 \Omega$, $T_A = 25^\circ\text{C}$		80	86		80	86		70	76	dB
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC} \pm / \Delta V_{IO}$)	$V_{CC} = \pm 9 \text{ V}$ to $\pm 15 \text{ V}$, $V_O = 0$, $R_S = 50 \Omega$, $T_A = 25^\circ\text{C}$		80	95		80	95		70	95	dB
P_D Total power dissipation	$V_O = 0$, No load, $T_A = 25^\circ\text{C}$		6	7.5		6	7.5		6	7.5	mW
I_{CC} Supply current	$V_O = 0$, No load, $T_A = 25^\circ\text{C}$		200	250		200	250		200	250	μA

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range of T_A is -55°C to 125°C for TL066M; -40°C to 85°C for TL066I; and 0°C to 70°C for TL066C. The electrical parameters are measured with the power-control terminal (pin 8) connected to $V_{CC} -$.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature-sensitive. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as is possible.

TL066AC, TL066BC ADJUSTABLE LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

electrical characteristics, $V_{CC} = \pm 15 \text{ V}$

PARAMETER	TEST CONDITIONS†	TL066AC			TL066BC			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0,$ $T_A = 25^\circ\text{C}$ $R_S = 50 \Omega,$		3	6		2	3	mV
	$V_O = 0,$ $T_A = \text{full range}$ $R_S = 50 \Omega,$			7.5			5	
α_{VIO} Temperature coefficient of input offset voltage	$V_O = 0,$ $T_A = \text{full range}$ $R_S = 50 \Omega,$		10			10		$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current‡	$V_O = 0,$ $T_A = 25^\circ\text{C}$		5	100		5	100	pA
	$V_O = 0,$ $T_A = \text{full range}$			3			3	nA
I_{IB} Input bias current‡	$V_O = 0,$ $T_A = 25^\circ\text{C}$		30	200		30	200	pA
	$V_O = 0,$ $T_A = \text{full range}$			7			7	nA
V_{ICR} Common-mode input voltage range	$T_A = 25^\circ\text{C}$		-12 to ± 11.5 to ± 15			-12 to ± 11.5 to ± 15		V
V_{OM} Maximum peak output voltage swing	$T_A = 25^\circ\text{C},$ $R_L \geq 10 \text{ k}\Omega,$		± 10	± 13.5		± 10	± 13.5	V
	$T_A = \text{full range},$ $R_L \geq 10 \text{ k}\Omega,$		± 10	± 13.5		± 10	± 13.5	
A_{VD} Large-signal differential voltage amplification	$R_L \geq 10 \text{ k}\Omega,$ $T_A = 25^\circ\text{C}$ $V_O = \pm 10 \text{ V},$		4	6		4	6	V/mV
	$R_L \geq 10 \text{ k}\Omega,$ $T_A = \text{full range}$ $V_O = \pm 10 \text{ V},$		4			4		
B_1 Unity-gain bandwidth	$T_A = 25^\circ\text{C},$ $R_L = 10 \text{ k}\Omega$		1			1		MHz
r_i Input resistance	$T_A = 25^\circ\text{C}$		10^{12}			10^{12}		Ω
r_o Output resistance	$T_A = 25^\circ\text{C},$ $f = 1 \text{ kHz}$		220			220		Ω
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR} \text{ min},$ $R_S = 50 \Omega,$ $T_A = 25^\circ\text{C}$ $V_O = 0,$		80	86		80	86	dB
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC} \pm \Delta V_{IO}$)	$V_{CC} = \pm 9 \text{ V to } \pm 15 \text{ V},$ $R_S = 50 \Omega,$ $T_A = 25^\circ\text{C}$ $V_O = 0,$		80	95		80	95	dB
P_D Total power dissipation	No load, $T_A = 25^\circ\text{C}$ $V_O = 0,$		6	7.5		6	7.5	mW
I_{CC} Supply current	No load, $T_A = 25^\circ\text{C}$ $V_O = 0,$		200	250		200	250	μA

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range of T_A is -55°C to 125°C for TL066M; -40°C to 85°C for TL066I; and 0°C to 70°C for TL066C, TL066AC, and TL066BC. The electrical parameters are measured with the power-control terminal (pin 8) connected to V_{CC-} .

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature-sensitive. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as is possible.

2
Operational Amplifiers

TL066M, TL066I, TL066C, TL066C, TL066BC
ADJUSTABLE LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

operating characteristics, $V_{CC\pm} = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$, $R_{ext} = 0$

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$V_I = 10\text{ V}$, $C_L = 100\text{ pF}$,	$R_L = 10\text{ k}\Omega$, See Figure 1	1.5	3.5		$\text{V}/\mu\text{s}$
t_r	Rise time	$V_I = 20\text{ mV}$,	$R_L = 10\text{ k}\Omega$		0.2		μs
	Overshoot factor	$C_L = 100\text{ pF}$,	See Figure 1		10%		
V_n	Equivalent input noise voltage	$R_S = 100\ \Omega$,	$f = 1\text{ kHz}$		42		$\text{nV}/\sqrt{\text{Hz}}$

PARAMETER MEASUREMENT INFORMATION

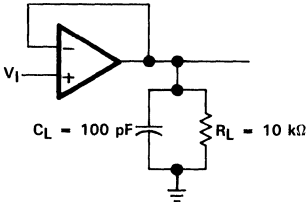


FIGURE 1. UNITY-GAIN AMPLIFIER

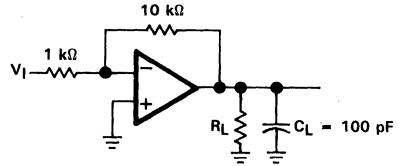


FIGURE 2. GAIN-OF-10 INVERTING AMPLIFIER

INPUT OFFSET VOLTAGE NULL CIRCUIT

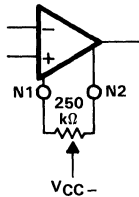


FIGURE 3

2

Operational Amplifiers

TLO66M, TLO66I, TLO66C, TLO66AC, TLO66BC
ADJUSTABLE LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

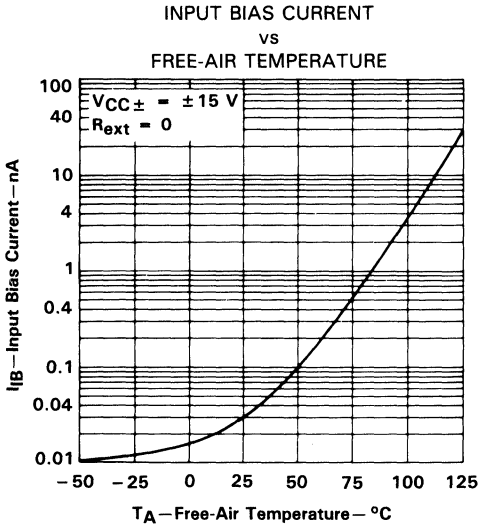


FIGURE 4

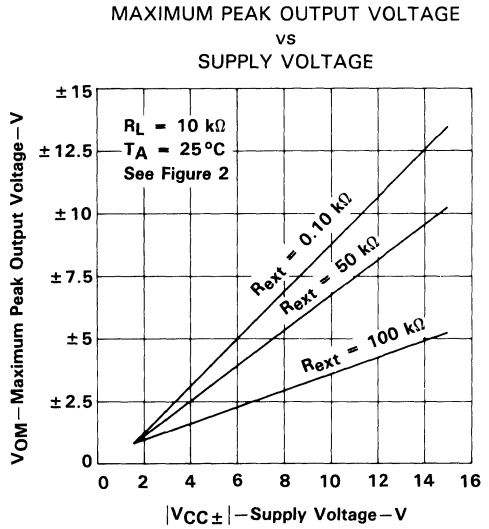


FIGURE 5

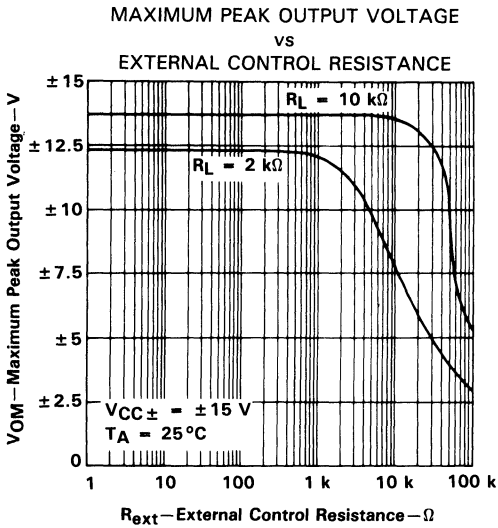


FIGURE 6

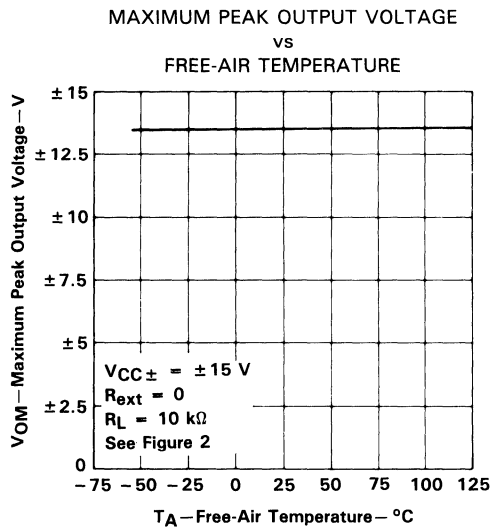


FIGURE 7

† Data at high and low temperatures are applicable only within the rated free-air temperature ranges of the various devices.

TL066M, TL066I, TL066C, TL066AC, TL066BC
ADJUSTABLE LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

**MAXIMUM PEAK OUTPUT VOLTAGE
 vs
 LOAD RESISTANCE**

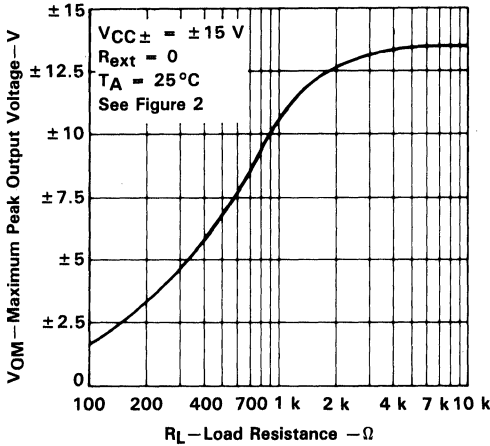


FIGURE 8

**MAXIMUM PEAK OUTPUT VOLTAGE
 vs
 FREQUENCY**

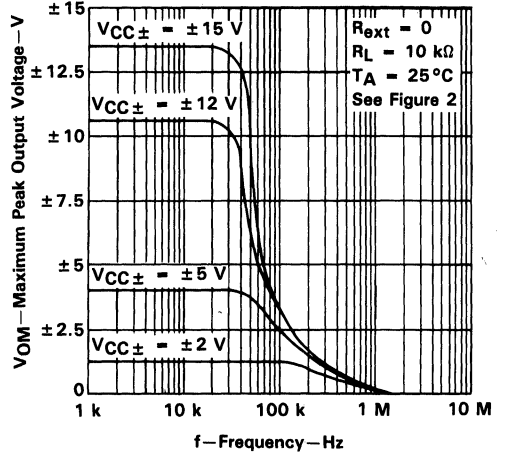


FIGURE 9

**DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 EXTERNAL CONTROL RESISTANCE**

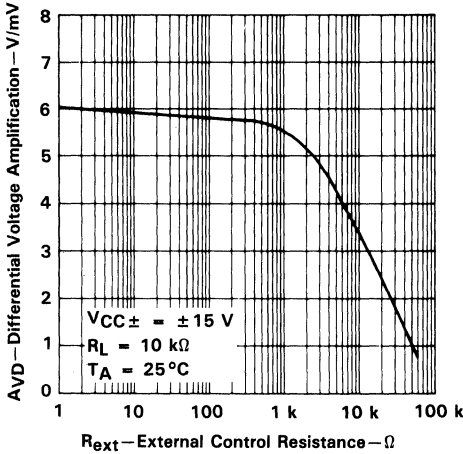


FIGURE 10

**DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 FREE-AIR TEMPERATURE**

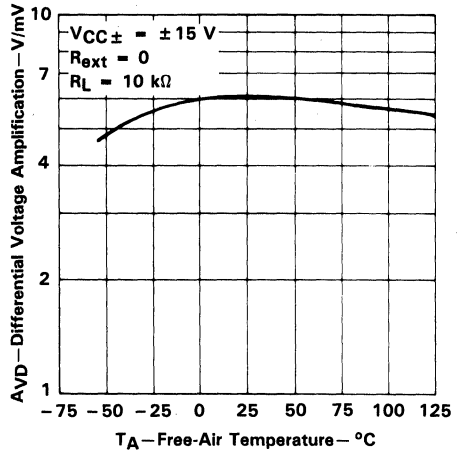


FIGURE 11

†Data at high and low temperatures are applicable only within the rated free-air temperature ranges of the various devices.

2
Operational Amplifiers

TL066M, TL066I, TL066C, TL066AC, TL066BC

ADJUSTABLE LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

LARGE SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
AND PHASE SHIFT
vs
FREQUENCY

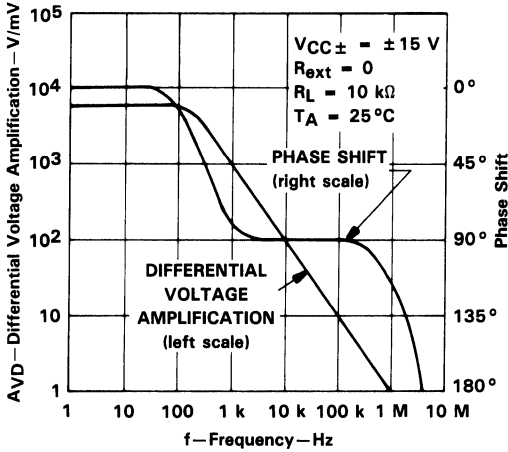


FIGURE 12

SUPPLY CURRENT
vs
SUPPLY VOLTAGE

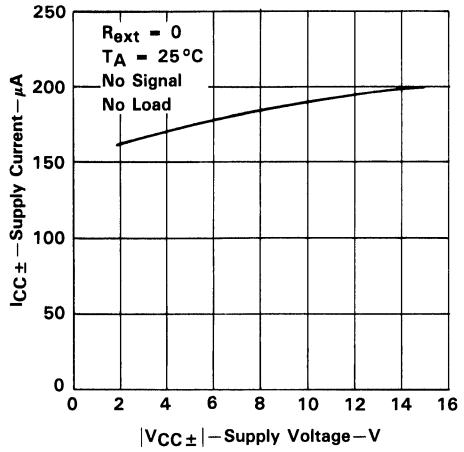


FIGURE 13

SUPPLY CURRENT
vs
EXTERNAL CONTROL RESISTANCE

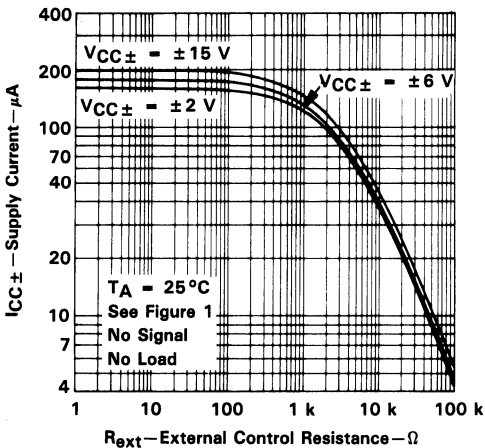


FIGURE 14

SUPPLY CURRENT
vs
FREE-AIR TEMPERATURE

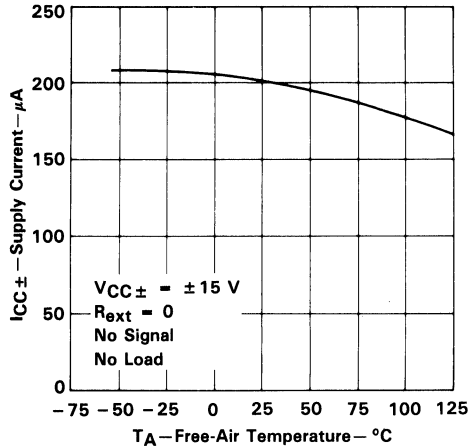


FIGURE 15

†Data at high and low temperatures are applicable only within the rated free-air temperature ranges of the various devices.

TL066M, TL066I, TL066C, TL066AC, TL066BC
ADJUSTABLE LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

TOTAL POWER DISSIPATION
 vs
 EXTERNAL CONTROL RESISTANCE

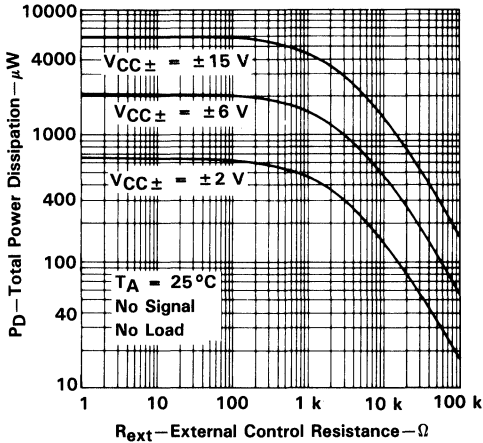


FIGURE 16

ALL EXCEPT TL066C
 COMMON-MODE REJECTION RATIO
 vs
 EXTERNAL CONTROL RESISTANCE

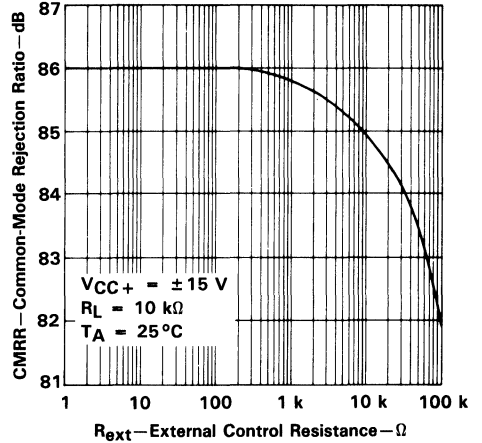


FIGURE 17

ALL EXCEPT TL066C
 COMMON-MODE REJECTION RATIO
 vs
 FREE-AIR TEMPERATURE

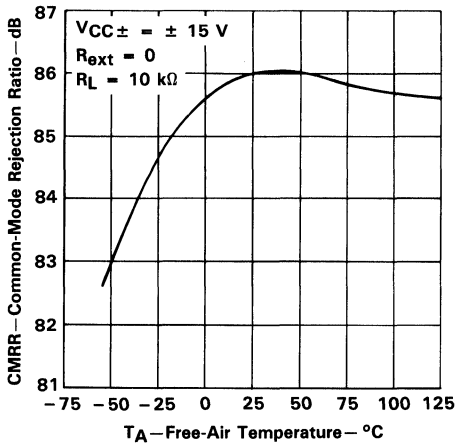


FIGURE 18

EQUIVALENT INPUT NOISE VOLTAGE
 vs
 FREQUENCY

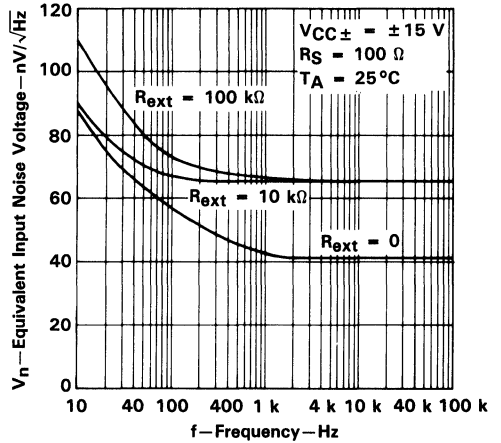


FIGURE 19

†Data at high and low temperatures are applicable only within the rated free-air temperature ranges of the various devices.

TLO66M, TLO66I, TLO66C, TLO66AC, TLO66BC
ADJUSTABLE LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

EQUIVALENT INPUT NOISE VOLTAGE
 VS
 SOURCE RESISTANCE

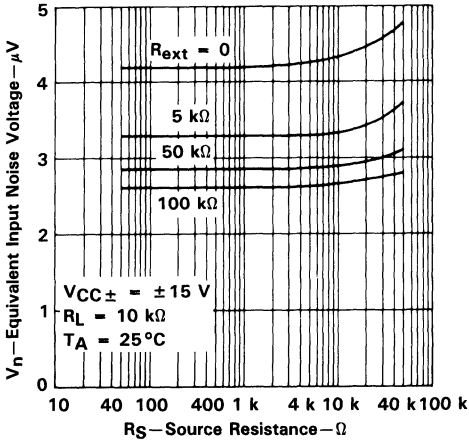


FIGURE 20

UNITY GAIN BANDWIDTH
 VS
 EXTERNAL CONTROL RESISTANCE

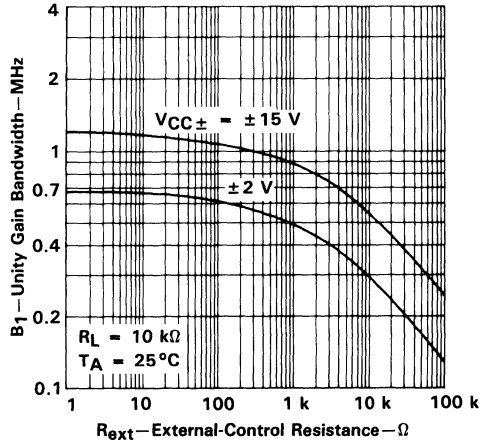


FIGURE 21

SLEW RATE
 VS
 EXTERNAL CONTROL RESISTANCE

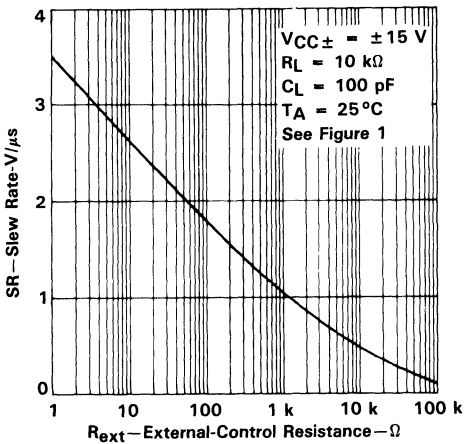


FIGURE 22

NORMALIZED UNITY GAIN BANDWIDTH
 SLEW RATE, AND PHASE SHIFT
 VS
 FREE-AIR TEMPERATURE

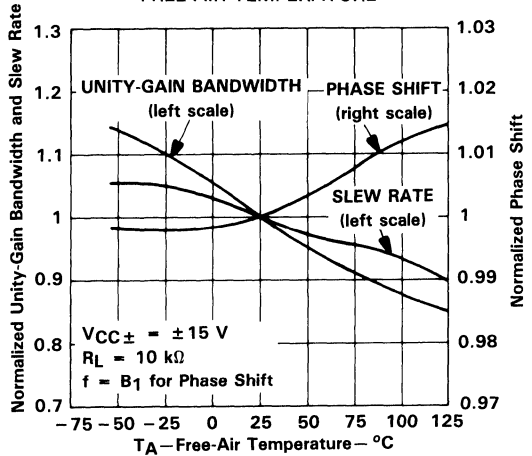


FIGURE 23

†Data at high and low temperatures are applicable only within the rated free-air temperature ranges of the various devices.

TL066M, TL066I, TL066C, TL066AC, TL066BC
ADJUSTABLE LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

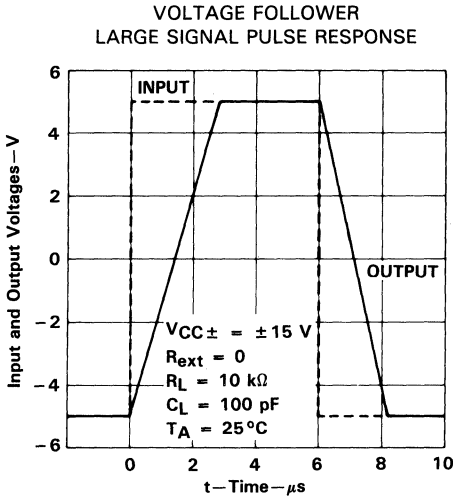


FIGURE 24

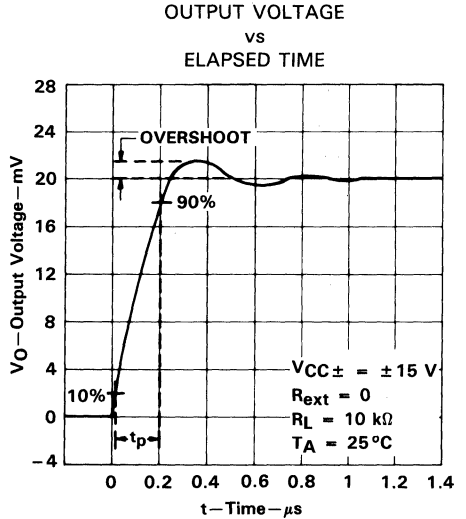


FIGURE 25

†Data at high and low temperatures are applicable only within the rated free-air temperature ranges of the various devices.

TYPICAL APPLICATION DATA

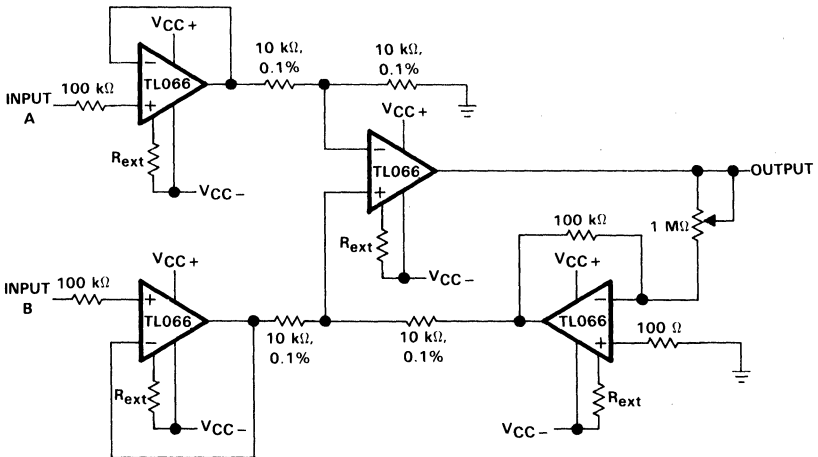


FIGURE 26. INSTRUMENTATION AMPLIFIER

TL066M, TL066I, TL066C, TL066AC, TL066BC
ADJUSTABLE LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

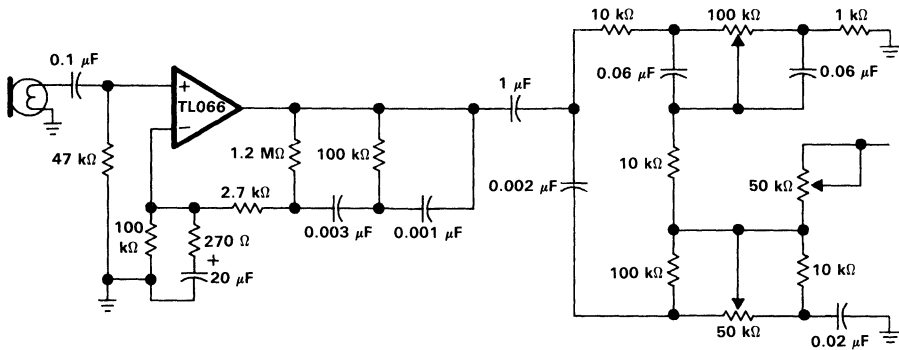


FIGURE 27. MICROPHONE PREAMPLIFIER WITH TONE CONTROL

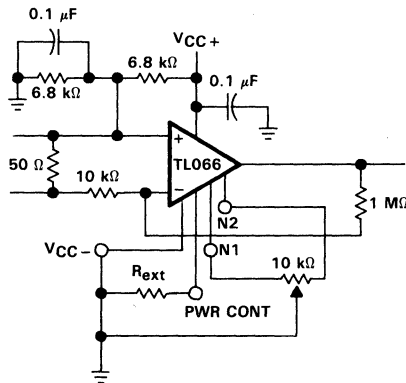


FIGURE 28. AC AMPLIFIER

TYPICAL APPLICATION DATA

IC PREAMPLIFIER RESPONSE CHARACTERISTICS

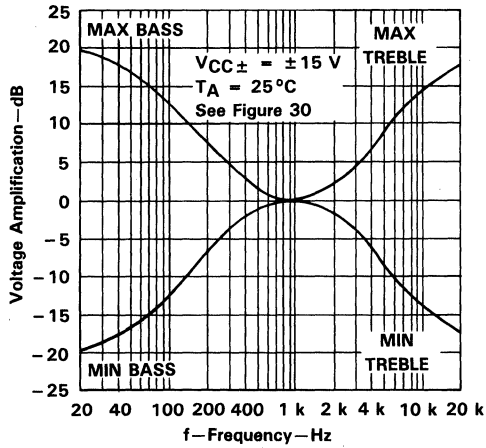


FIGURE 29

2

Operational Amplifiers

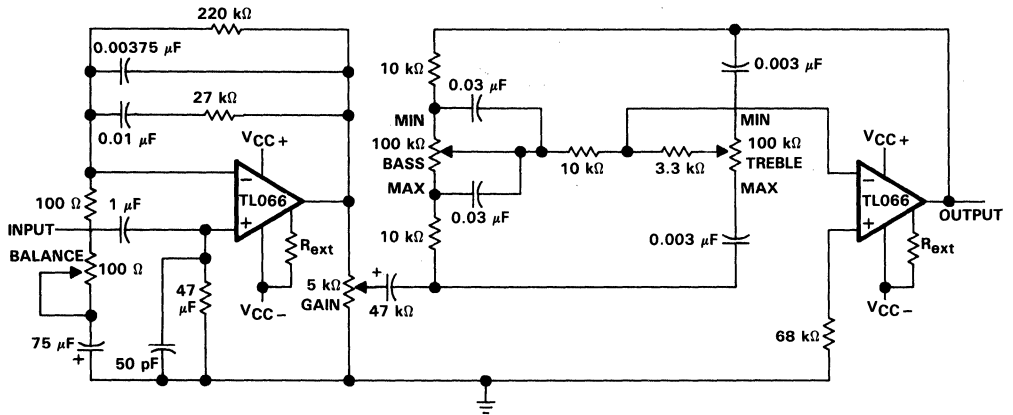


FIGURE 30. IC PREAMPLIFIER

TL070, TL070A, TL071, TL071A, TL071B, TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL075 LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS

D2393, SEPTEMBER 1978—REVISED JANUARY 1989

19 DEVICES COVER COMMERCIAL, INDUSTRIAL, AND MILITARY TEMPERATURE RANGES

- Low Power Consumption
- Wide Common-Mode and Differential Voltage Ranges
- Low Input Bias and Offset Currents
- Output Short-Circuit Protection
- Low Total Harmonic Distortion . . . 0.003% Typ
- Common-Mode Input Voltage Range Includes V_{CC+}
- Low Noise . . . $V_n = 18 \text{ nV}/\sqrt{\text{Hz}}$ Typ
- High Input Impedance . . . JFET-Input Stage
- Internal Frequency Compensation (Except TL070, TL070A)
- Latch-Up-Free Operation
- High Slew Rate . . . $13 \text{ V}/\mu\text{s}$ Typ

description

The JFET-input operational amplifiers in the TL07__ series are designed as low-noise versions of the TL08__ series amplifiers with low input bias and offset currents and fast slew rate. The low harmonic distortion and low noise make the TL07__ series ideally suited as amplifiers for high-fidelity and audio preamplifier applications. Each amplifier features JFET-inputs (for high input impedance) coupled with bipolar output stages all integrated on a single monolithic chip.

The M suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I suffix devices are characterized for operation from -40°C to 85°C , and the C suffix devices are characterized for operation from 0°C to 70°C .

AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE							
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (N)	PLASTIC DIP (P)	FLAT PACK (W)
0°C to 70°C	10 mV	TL070CD			TL070CJG			TL070CP	
	6 mV	TL070ACD			TL070ACJG			TL070ACP	
	10 mV	TL071CD			TL071CJG			TL071CP	
	6 mV	TL071ACD			TL071ACJG			TL071ACP	
	3 mV	TL071BCD			TL071BCJG			TL071BCP	
	10 mV	TL072CD			TL072CJG			TL072CP	
	6 mV	TL072ACD			TL072ACJG			TL072ACP	
	3 mV	TL072BCD			TL072BCJG			TL072BCP	
	10 mV	TL074CD		TL074CJ			TL074CN		
	6 mV	TL074ACD		TL074ACJ			TL074ACN		
	3 mV	TL074BCD		TL074BCJ			TL074BCN		
	10 mV						TL075CN		
-40°C to 85°C	6 mV	TL070ID			TL070IJG			TL070IP	
	6 mV	TL071ID			TL071IJG			TL071IP	
	6 mV	TL072ID			TL072IJG			TL072IP	
	6 mV	TL074ID		TL074IJ			TL074IN		
-55°C to 125°C	6 mV		TL071MFK		TL071MJG	TL071ML			
	6 mV		TL072MFK		TL072MJG	TL072ML			
	9 mV		TL074MFK	TL074MJ					TL074MW

The D package is available taped and reeled. Add the suffix R to the device type (e.g., TL071CDR).

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

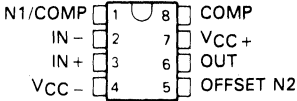


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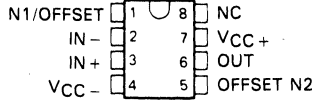
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**TL070, TL070A, TL071, TL071A, TL071B,
TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL075**
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS

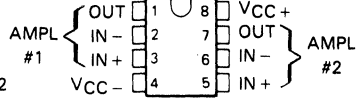
TL070, TL070A
D, JG, OR P PACKAGE
(TOP VIEW)



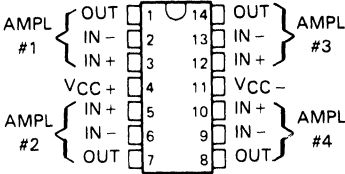
TL071, TL071A, TL071B
D, JG, OR P PACKAGE
(TOP VIEW)



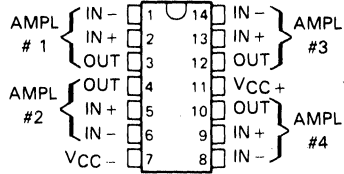
TL072, TL072A, TL072B
D, JG, OR P PACKAGE
(TOP VIEW)



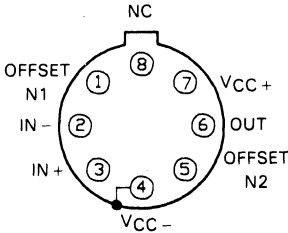
TL074, TL074A, TL074B
D, J, OR N PACKAGE
TL074 . . . W PACKAGE
(TOP VIEW)



TL075
N PACKAGE
(TOP VIEW)

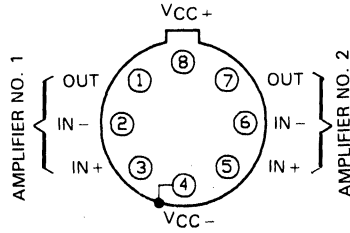


TL071 . . . L PACKAGE
(TOP VIEW)



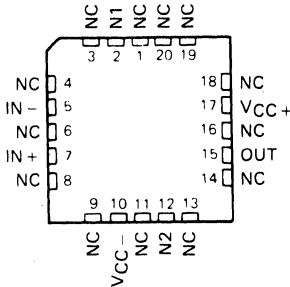
PIN 4 IS IN ELECTRICAL CONTACT WITH THE CASE

TL072 . . . L PACKAGE
(TOP VIEW)

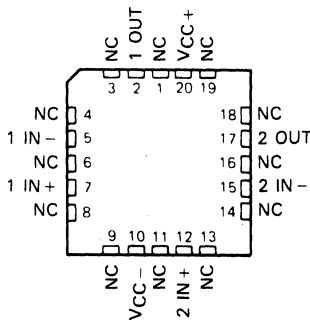


PIN 4 IS IN ELECTRICAL CONTACT WITH THE CASE

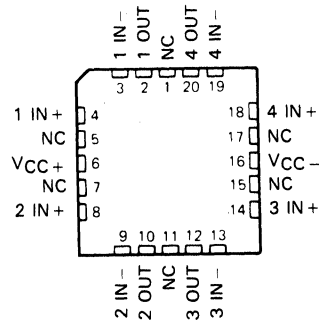
TL071
FK PACKAGE
(TOP VIEW)



TL072
FK PACKAGE
(TOP VIEW)



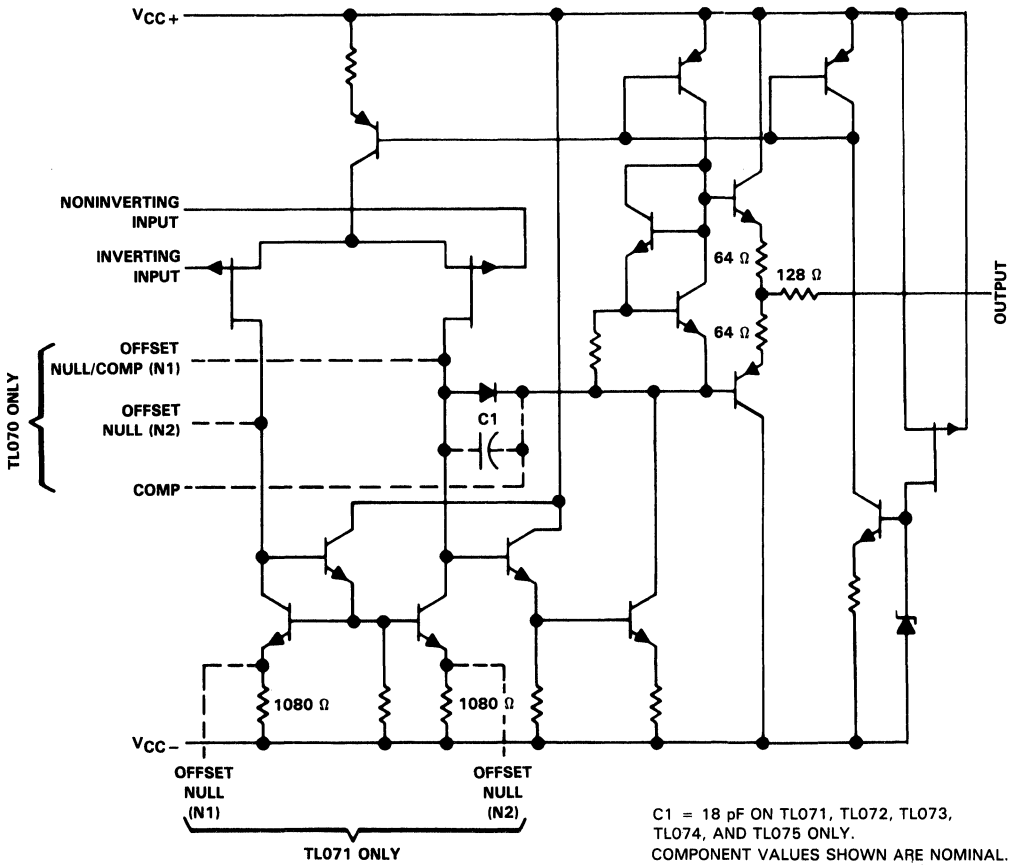
TL074
FK PACKAGE
(TOP VIEW)



NC—No internal connection.

**TL070, TL070A, TL071, TL071A, TL071B,
TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL075
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**

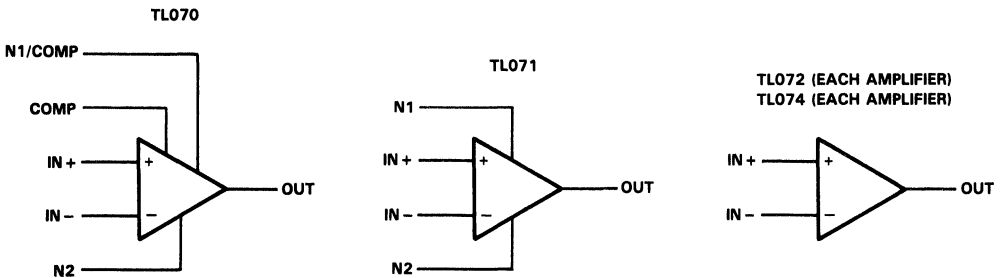
schematic (each amplifier)



2

Operational Amplifiers

symbols



**TL070, TL070A, TL071, TL071A, TL071B,
TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL075**
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	TL07__M	TL07__I	TL07__C TL07__AC TL07__BC	UNIT	
Supply voltage, V_{CC+} (see Note 1)	18	18	18	V	
Supply voltage, V_{CC-} (see Note 1)	-18	-18	-18	V	
Differential input voltage (see Note 2)	± 30	± 30	± 30	V	
Input voltage (see Notes 1 and 3)	± 15	± 15	± 15	V	
Duration of output short circuit (see Note 4)	unlimited	unlimited	unlimited		
Continuous total dissipation	See Dissipation Rating Table				
Operating free-air temperature range	-55 to 125	-40 to 85	0 to 70	°C	
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	°C	
Case temperature for 60 seconds	FK package	260		°C	
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J, JG, or W package	300	300	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D, N, or P package		260	260	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	L package	300			°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
 4. The output may be shorted to ground or to either supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING			POWER RATING	POWER RATING	POWER RATING
D (8-pin)	680 mW	5.8 mW/°C	33°C	464 mW	377 mW	N/A
D (14-pin)	680 mW	7.6 mW/°C	60°C	608 mW	494 mW	N/A
FK	680 mW	11.0 mW/°C	88°C	680 mW	680 mW	275 mW
J (TL07__M)	680 mW	11.0 mW/°C	88°C	680 mW	680 mW	275 mW
J (all others)	680 mW	8.2 mW/°C	67°C	656 mW	533 mW	N/A
JG (TL07__M)	680 mW	8.4 mW/°C	69°C	672 mW	546 mW	210 mW
JG (all others)	680 mW	6.6 mW/°C	47°C	528 mW	429 mW	N/A
L	680 mW	6.6 mW/°C	25°C	528 mW	429 mW	165 mW
N	680 mW	9.2 mW/°C	76°C	680 mW	598 mW	N/A
P	680 mW	8.0 mW/°C	65°C	640 mW	520 mW	N/A
W	680 mW	8.0 mW/°C	65°C	640 mW	520 mW	200 mW

2
Operational Amplifiers

TL071M, TL072M, TL074M LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS

electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS [†]		TL071M TL072M			TL074M			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0,$ $R_S = 50\ \Omega,$	$T_A = 25\ ^\circ\text{C}$	3 6			3 9			mV
		$T_A = -55\ ^\circ\text{C}$ to $125\ ^\circ\text{C}$	9						
αV_{IO} Temperature coefficient of input offset voltage	$V_O = 0,$ $T_A = -55\ ^\circ\text{C}$ to $125\ ^\circ\text{C}$	$R_S = 50\ \Omega,$	18			18			$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current [‡]	$V_O = 0$	$T_A = 25\ ^\circ\text{C}$	5 100			5 100			pA
		$T_A = -55\ ^\circ\text{C}$ to $125\ ^\circ\text{C}$	20						nA
I_{IB} Input bias current	$V_O = 0$	$T_A = 25\ ^\circ\text{C}$	65 200			65 200			pA
		$T_A = -55\ ^\circ\text{C}$ to $125\ ^\circ\text{C}$	50						nA
V_{ICR} Common-mode input voltage range	$T_A = 25\ ^\circ\text{C}$		-12 ±11 to +15			-12 ±11 to +15			V
V_{OM} Maximum peak output voltage swing	$R_L = 10\ \text{k}\Omega$	$T_A = 25\ ^\circ\text{C}$	±12 ±13.5			±12 ±13.5			V
	$R_L \geq 10\ \text{k}\Omega$	$T_A = -55\ ^\circ\text{C}$ to $125\ ^\circ\text{C}$	±12			±12			
	$R_L \geq 2\ \text{k}\Omega$		±10			±10			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 10\ \text{V},$ $R_L \geq 2\ \text{k}\Omega$	$T_A = 25\ ^\circ\text{C}$	35 200			35 200			V/mV
		$T_A = -55\ ^\circ\text{C}$ to $125\ ^\circ\text{C}$	15						
B_1 Unity-gain bandwidth	$T_A = 25\ ^\circ\text{C}$		3			3			MHz
r_i Input resistance	$T_A = 25\ ^\circ\text{C}$		10^{12}			10^{12}			Ω
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR}\ \text{min}, V_O = 0,$ $R_S = 50\ \Omega, T_A = 25\ ^\circ\text{C}$		80 86			80 86			dB
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC\pm}/\Delta V_{IO}$)	$V_{CC} = \pm 15\ \text{V}$ to $\pm 9\ \text{V}, V_O = 0,$ $R_S = 50\ \Omega, T_A = 25\ ^\circ\text{C}$		80 86			80 86			dB
I_{CC} Supply current (each amplifier)	No load, $V_O = 0,$ $T_A = 25\ ^\circ\text{C}$		1.4 2.5			1.4 2.5			mA
V_{O1}/V_{O2} Crosstalk attenuation	$A_{VD} = 100, T_A = 25\ ^\circ\text{C}$		120			120			dB

[†]All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified.

[‡]Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 6. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

2

Operational Amplifiers

electrical characteristics, $V_{CC\pm} = \pm 15$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONS [†]		TL070I TL071I TL072I TL074I			TL070C TL071C TL072C TL074C			TL070AC TL071AC TL072AC TL074AC			TL070BC TL071BC TL072BC TL074BC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = \Omega$, $R_S = 50 \Omega$	$T_A = 25^\circ\text{C}$	3		6	3		10	3		6	2		3	mV
		$T_A = \text{full range}$			8			13			7.5			5	
αV_{IO} Temperature coefficient of input offset voltage	$V_O = 0$, $T_A = \text{full range}$	$R_S = 50 \Omega$	18			18			18			18			$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current [‡]	$V_O = 0$	$T_A = 25^\circ\text{C}$	5	100		5	100		5	100		5	100	μA	
		$T_A = \text{full range}$			10			2		2		2		2	nA
I_{IB} Input bias current [‡]	$V_O = 0$	$T_A = 25^\circ\text{C}$	65	200		65	200		65	200		65	200	μA	
		$T_A = \text{full range}$			20			7		7		7		7	nA
V_{ICR} Common-mode input voltage range	$T_A = 25^\circ\text{C}$		-12 ± 11 to +15			-12 ± 11 to +15			-12 ± 11 to +15			-12 ± 11 to +15		V	
V_{OM} Maximum peak output voltage swing	$R_L = 10 \text{ k}\Omega$ $R_L \geq 10 \text{ k}\Omega$ $R_L \geq 2 \text{ k}\Omega$	$T_A = 25^\circ\text{C}$	± 12	± 13.5		± 12	± 13.5		± 12	± 13.5		± 12	± 13.5	V	
		$T_A = \text{full range}$	± 12			± 12			± 12			± 12			
			± 10			± 10			± 10			± 10			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 10 \text{ V}$, $R_L \geq 2 \text{ k}\Omega$	$T_A = 25^\circ\text{C}$	50	200		25	200		50	200		50	200	V/mV	
		$T_A = \text{full range}$	25			15			25			25			
B_1 Unity-gain bandwidth	$T_A = 25^\circ\text{C}$		3			3			3			3			MHz
r_i Input resistance	$T_A = 25^\circ\text{C}$		10^{12}			10^{12}			10^{12}			10^{12}			Ω
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR} \text{ min}$, $V_O = 0$, $R_S = 50 \Omega$, $T_A = 25^\circ\text{C}$		80	100		70	100		80	100		80	100	dB	
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC\pm}/\Delta V_{IO}$)	$V_{CC} = \pm 15 \text{ V}$ to $\pm 9 \text{ V}$, $V_O = 0$, $R_S = 50 \Omega$, $T_A = 25^\circ\text{C}$		80	100		70	100		80	100		80	100	dB	
I_{CC} Supply current (each amplifier)	No load, $T_A = 25^\circ\text{C}$		$V_O = 0$	1.4	2.5		1.4	2.5		1.4	2.5		1.4	2.5	mA
V_{O1}/V_{O2} Crosstalk attenuation	$A_{VD} = 100$, $T_A = 25^\circ\text{C}$		120			120			120			120			dB

[†]All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified. Full range for T_A is -40°C to 85°C for TL07_1 and 0°C to 70°C for TL07_C, TL07_AC, and TL07_BC.

[‡]Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 6. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

TL070, TL070A, TL071, TL071A, TL071B, TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL074C, TL074D, TL074E, TL074F, TL074G, TL074H, TL074I, TL074J, TL074K, TL074L, TL074M, TL074N, TL074P, TL074Q, TL074R, TL074S, TL074T, TL074U, TL074V, TL074W, TL074X, TL074Y, TL074Z, TL074AA, TL074AB, TL074AC, TL074AD, TL074AE, TL074AF, TL074AG, TL074AH, TL074AI, TL074AJ, TL074AK, TL074AL, TL074AM, TL074AN, TL074AO, TL074AP, TL074AQ, TL074AR, TL074AS, TL074AT, TL074AU, TL074AV, TL074AW, TL074AX, TL074AY, TL074AZ, TL074BA, TL074BB, TL074BC, TL074BD, TL074BE, TL074BF, TL074BG, TL074BH, TL074BI, TL074BJ, TL074BK, TL074BL, TL074BM, TL074BN, TL074BO, TL074BP, TL074BQ, TL074BR, TL074BS, TL074BT, TL074BU, TL074BV, TL074BW, TL074BX, TL074BY, TL074BZ, TL074CA, TL074CB, TL074CC, TL074CD, TL074CE, TL074CF, TL074CG, TL074CH, TL074CI, TL074CJ, TL074CK, TL074CL, TL074CM, TL074CN, TL074CO, TL074CP, TL074CQ, TL074CR, TL074CS, TL074CT, TL074CU, TL074CV, TL074CW, TL074CX, TL074CY, TL074CZ, TL074DA, TL074DB, TL074DC, 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TL070, TL070A, TL071, TL071A, TL071B, TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL075 LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS

operating characteristics, $V_{CC\pm} = \pm 15 \text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TL07_M			ALL OTHERS			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain $V_I = 10 \text{ V}$, $C_L = 100 \text{ pF}$, $R_L = 2 \text{ k}\Omega$, See Figure 1	8	13		8	13		$\text{V}/\mu\text{s}$
t_r	Rise time overshoot factor $V_I = 20 \text{ mV}$, $C_L = 100 \text{ pF}$, $R_L = 2 \text{ k}\Omega$, See Figure 1	0.1			0.1			μs
		20			20			%
V_n	Equivalent input noise voltage $R_S = 100 \Omega$	$f = 1 \text{ kHz}$			18			$\text{nV}/\sqrt{\text{Hz}}$
		$f = 10 \text{ Hz to } 10 \text{ kHz}$			4			μV
I_n	Equivalent input noise current $R_S = 100 \Omega$, $f = 1 \text{ kHz}$	0.01			0.01			$\text{pA}/\sqrt{\text{Hz}}$
THD	Total harmonic distortion $V_{O(\text{rms})} = 10 \text{ V}$, $R_L \geq 2 \text{ k}\Omega$, $R_S \leq 1 \text{ k}\Omega$, $f = 1 \text{ kHz}$	0.003			0.003			%

2

Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

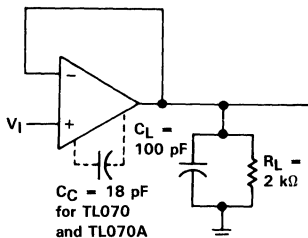


FIGURE 1. UNITY-GAIN AMPLIFIER

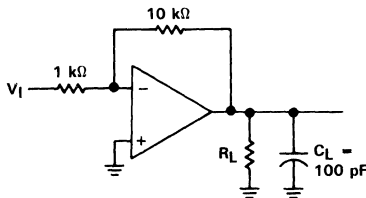


FIGURE 2. GAIN-OF-10
INVERTING AMPLIFIER

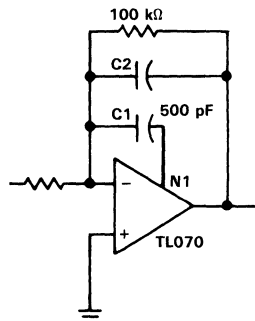


FIGURE 3. FEED-FORWARD
COMPENSATION

INPUT OFFSET VOLTAGE NULL CIRCUITS

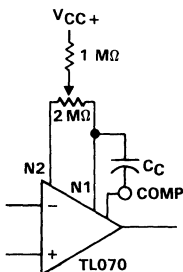


FIGURE 4

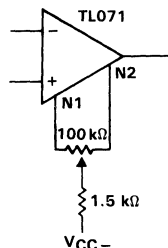


FIGURE 5

**TL070, TL070A, TL071, TL071A, TL071B,
TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL075**
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

**INPUT BIAS CURRENT
vs
FREE-AIR TEMPERATURE**

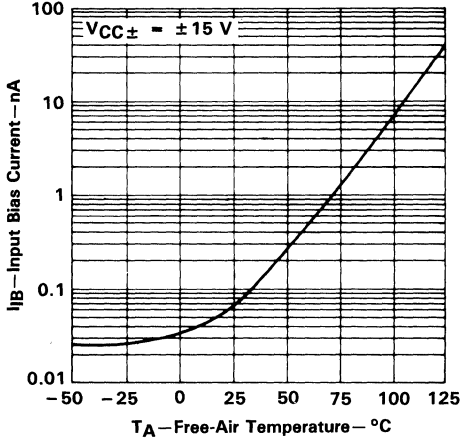


FIGURE 6

**MAXIMUM PEAK OUTPUT VOLTAGE
vs
FREQUENCY**

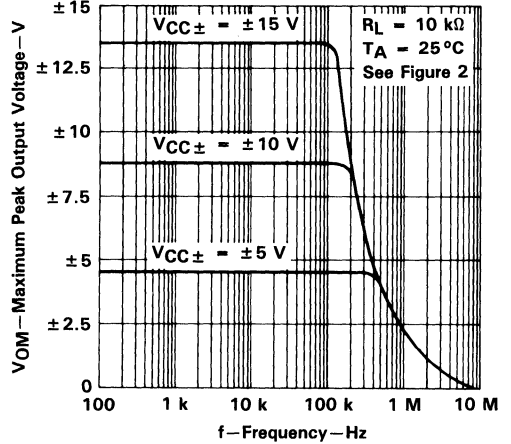


FIGURE 7

**MAXIMUM PEAK OUTPUT VOLTAGE
vs
FREQUENCY**

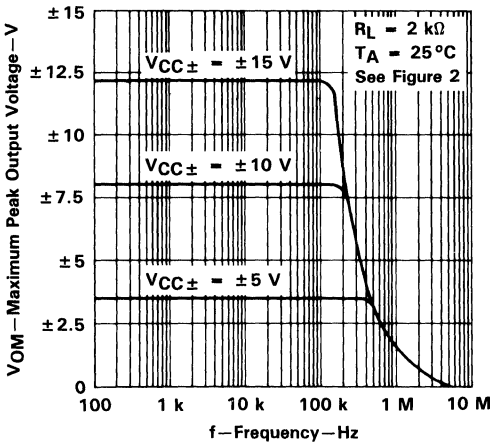


FIGURE 8

**MAXIMUM PEAK OUTPUT VOLTAGE
vs
FREQUENCY**

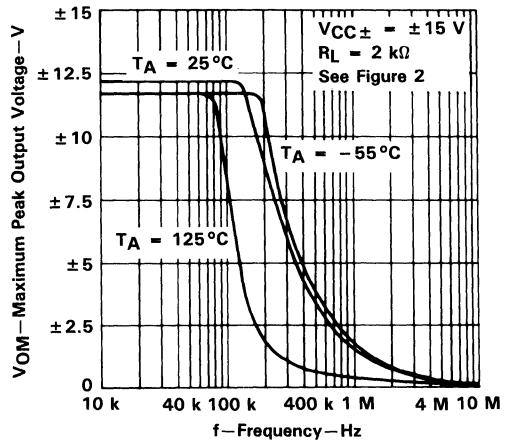


FIGURE 9

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. An 18-pF compensation capacitor is used with TL070 and TL070A.

**TL070, TL070A, TL071, TL071A, TL071B
TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL075
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE
vs
FREE-AIR TEMPERATURE

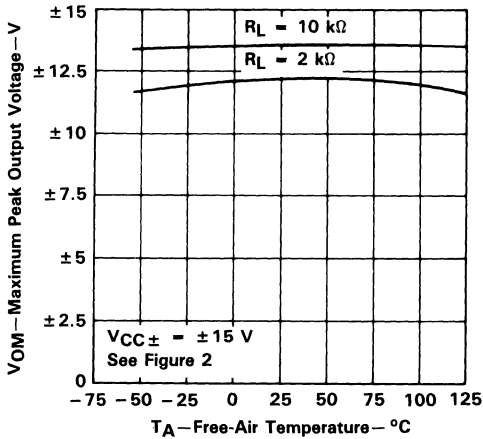


FIGURE 10

MAXIMUM PEAK OUTPUT VOLTAGE
vs
LOAD RESISTANCE

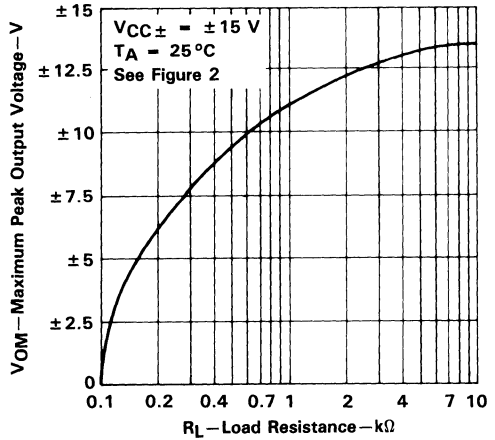


FIGURE 11

MAXIMUM PEAK OUTPUT VOLTAGE
vs
SUPPLY VOLTAGE

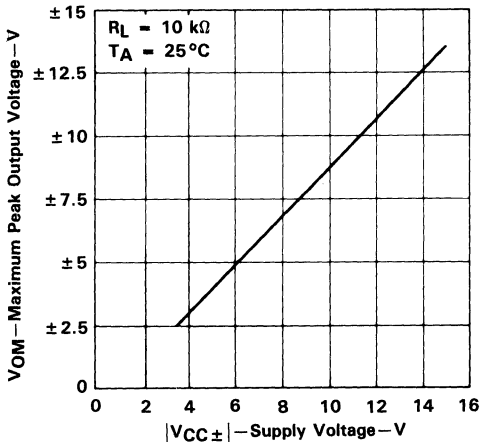


FIGURE 12

LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREE-AIR TEMPERATURE

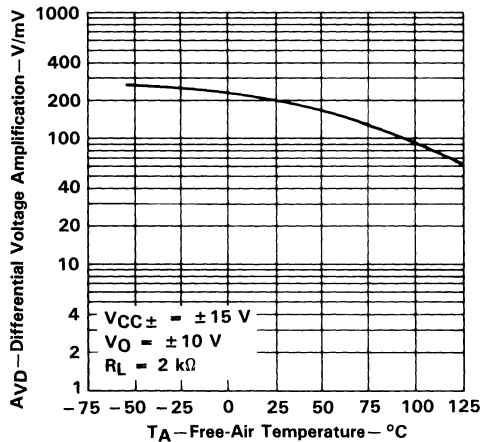


FIGURE 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. An 18-pF compensation capacitor is used with TL070 and TL070A.

**TL070, TL070A, TL071, TL071A, TL071B,
TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL075**
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

TL070
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREQUENCY WITH FEED-FORWARD
COMPENSATION

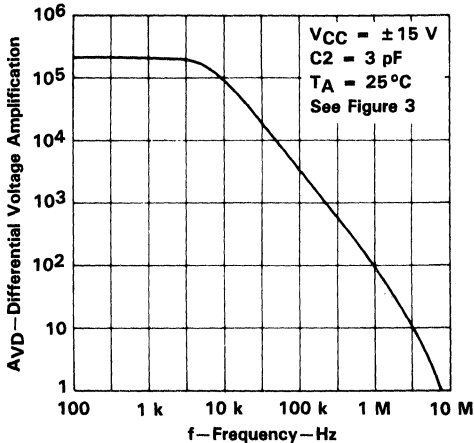


FIGURE 14

LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
and PHASE SHIFT
vs
FREQUENCY

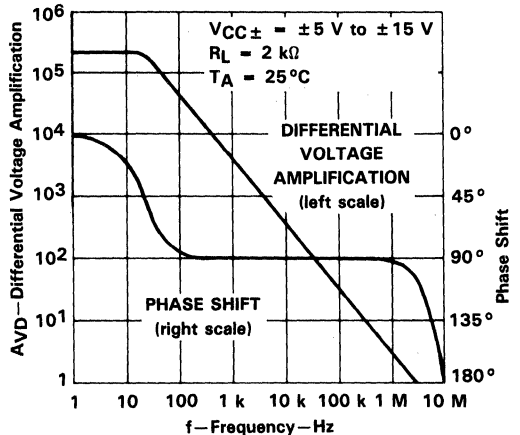


FIGURE 15

NORMALIZED UNITY-GAIN BANDWIDTH
and PHASE SHIFT
vs
FREE-AIR TEMPERATURE

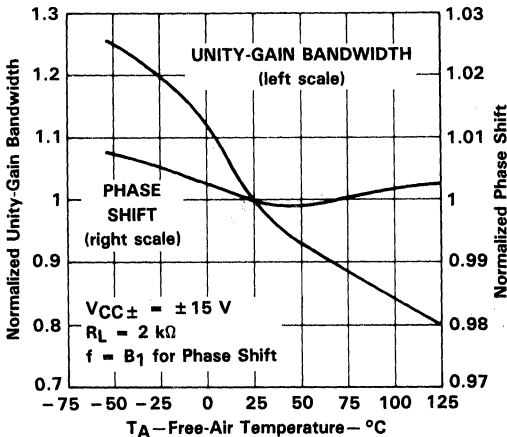


FIGURE 16

COMMON-MODE REJECTION RATIO
vs
FREE-AIR TEMPERATURE

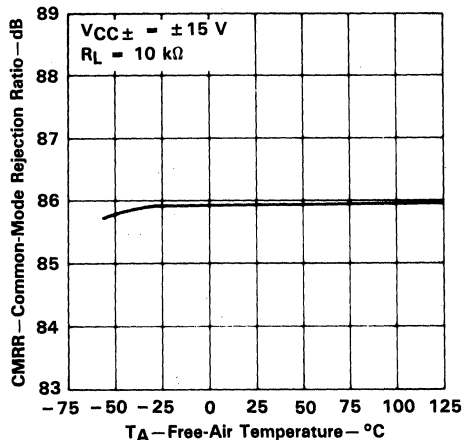


FIGURE 17

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. An 18-pF compensation capacitor is used with TL070 and TL070A.

**TL070, TL070A, TL071, TL071A, TL071B
TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL075
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS†

SUPPLY CURRENT PER AMPLIFIER
vs
SUPPLY VOLTAGE

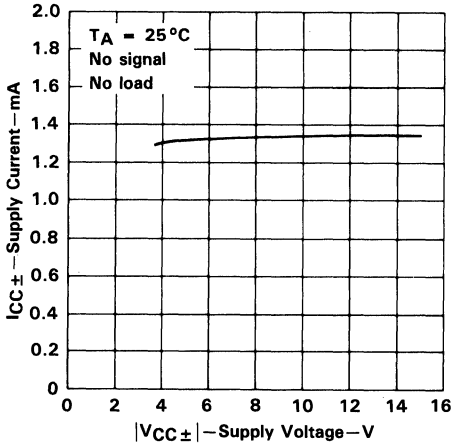


FIGURE 18

SUPPLY CURRENT PER AMPLIFIER
vs
FREE-AIR TEMPERATURE

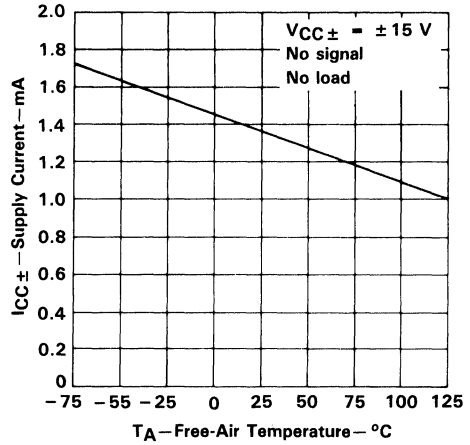


FIGURE 19

TOTAL POWER DISSIPATED
vs
FREE-AIR TEMPERATURE

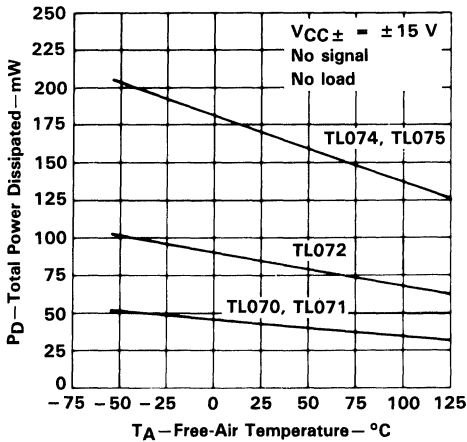


FIGURE 20

NORMALIZED SLEW RATE
vs
FREE-AIR TEMPERATURE

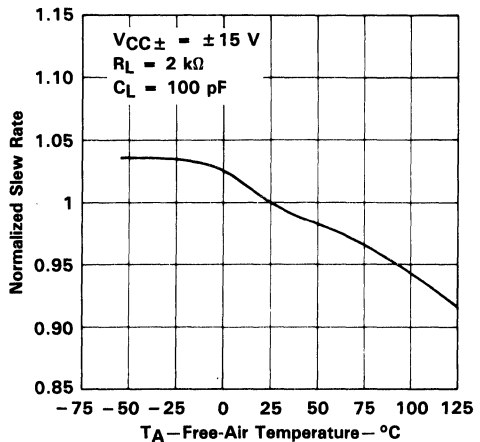


FIGURE 21

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. An 18-pF compensation capacitor is used with TL070 and TL070A.

**TL070, TL070A, TL071, TL071A, TL071B,
TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL075**
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

**EQUIVALENT INPUT NOISE VOLTAGE
vs
FREQUENCY**

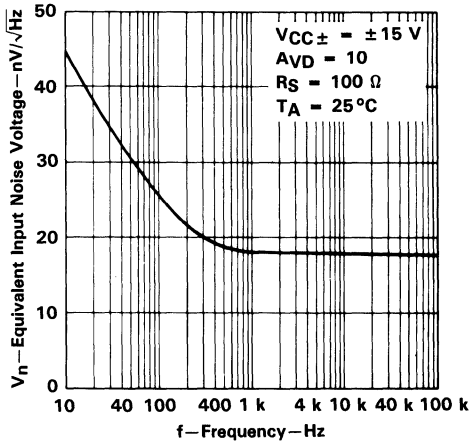


FIGURE 22

**TOTAL HARMONIC DISTORTION
vs
FREQUENCY**

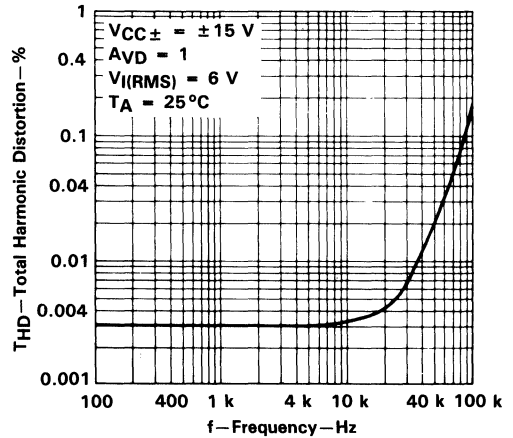


FIGURE 23

**VOLTAGE-FOLLOWER
LARGE-SIGNAL PULSE RESPONSE**

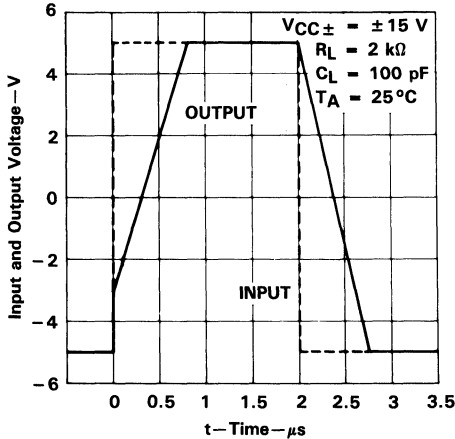


FIGURE 24

**OUTPUT VOLTAGE
vs
ELAPSED TIME**

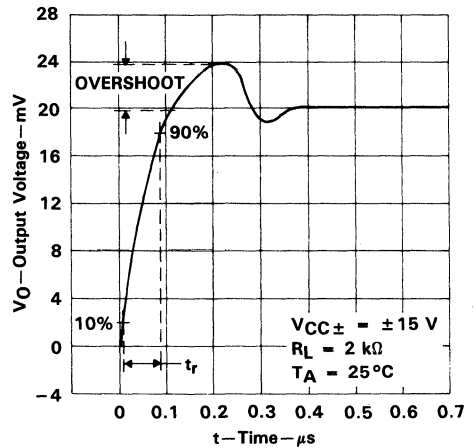


FIGURE 25

**TL070, TL070A, TL071, TL071A, TL071B
TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL075
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL APPLICATION DATA

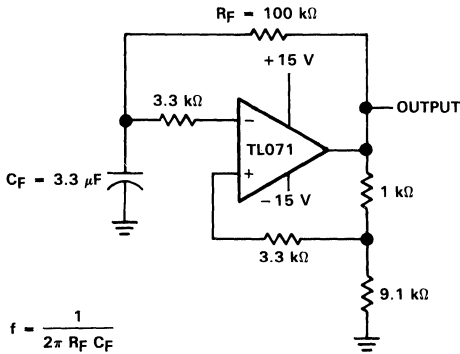


FIGURE 26. 0.5-Hz SQUARE-WAVE OSCILLATOR

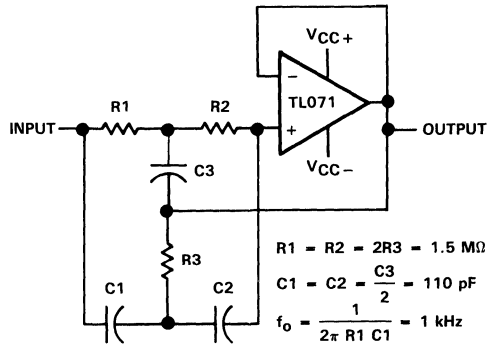
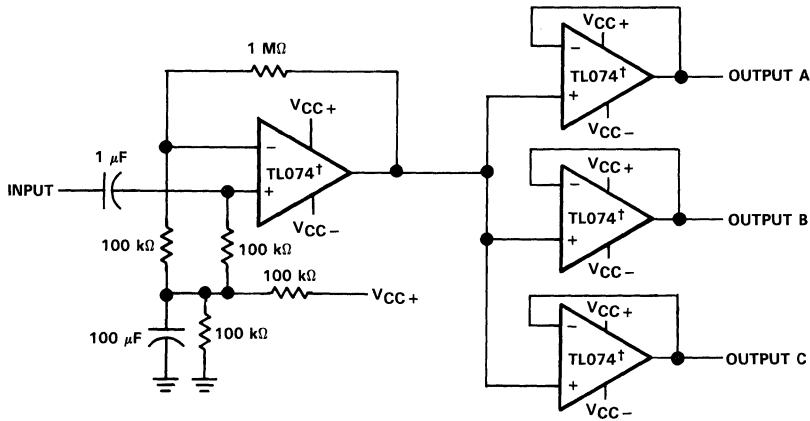


FIGURE 27. HIGH-Q NOTCH FILTER



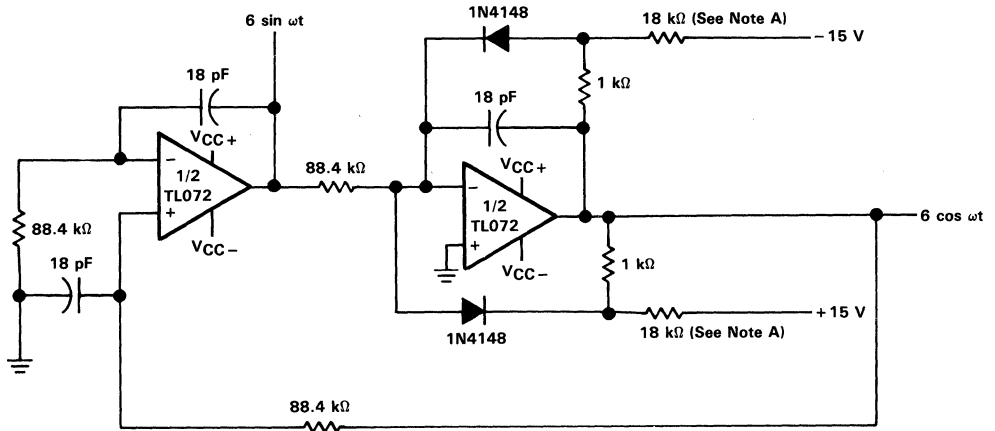
[†]or TL075

FIGURE 28. AUDIO DISTRIBUTION AMPLIFIER

**TL070, TL070A, TL071, TL071A, TL071B,
TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL075**
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

2
Operational Amplifiers



Note A: These resistor values may be adjusted for a symmetrical output.

FIGURE 29. 100-kHz QUADRATURE OSCILLATOR

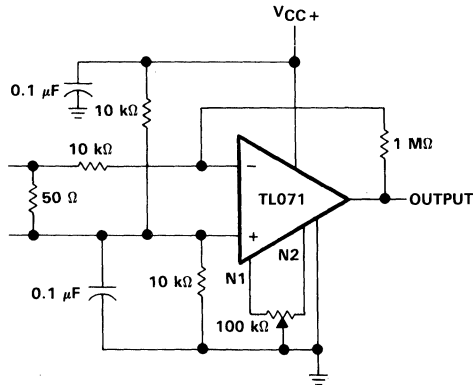


FIGURE 30. AC AMPLIFIER

**TL070, TL070A, TL071, TL071A, TL071B
TL072, TL072A, TL072B, TL074, TL074A, TL074B, TL075
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL APPLICATION DATA

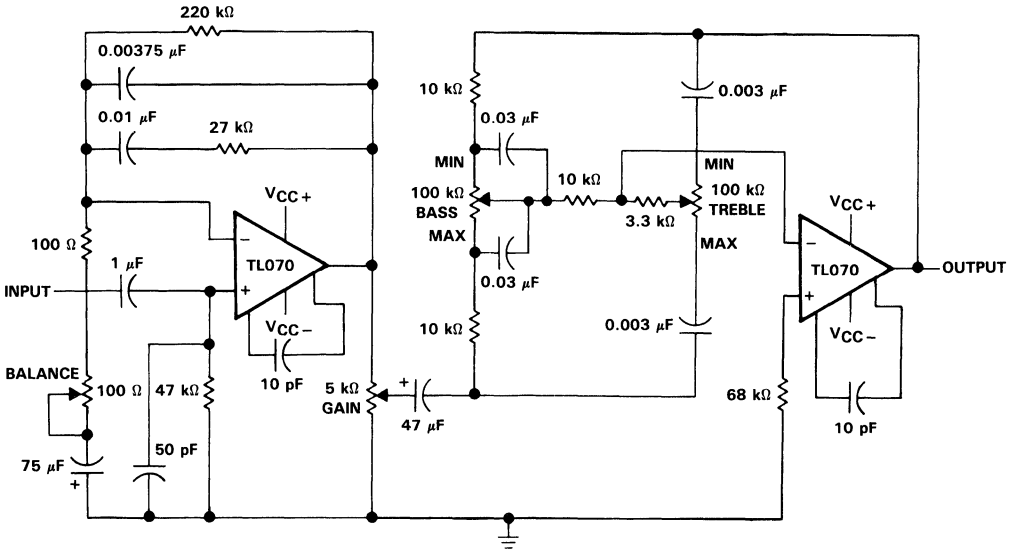


FIGURE 31. IC PREAMPLIFIER

**IC PREAMPLIFIER
RESPONSE CHARACTERISTICS**

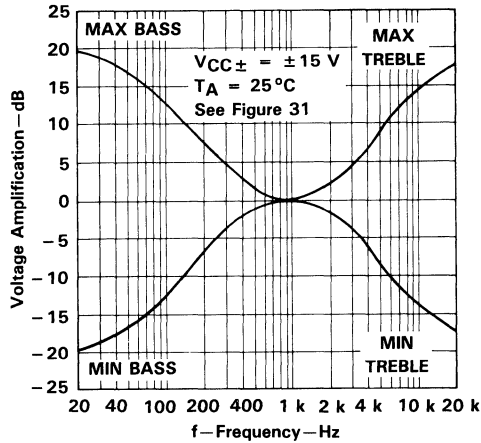


FIGURE 32

2
Operational Amplifiers

2

Operational Amplifiers

TL080 THRU TL085, TL080A THRU TL084A TL081B, TL082B, TL084B JFET-INPUT OPERATIONAL AMPLIFIERS

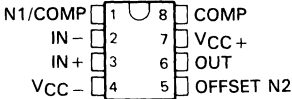
D2297, FEBRUARY 1977—REVISED NOVEMBER 1988

24 DEVICES COVER MILITARY, INDUSTRIAL AND COMMERCIAL TEMPERATURE RANGES

- Low-Power Consumption
- High Input Impedance . . . JFET-Input Stage
- Wide Common-Mode and Differential Voltage Ranges
- Internal Frequency Compensation (Except TL080, TL080A)
- Low Input Bias and Offset Currents
- Latch-Up-Free Operation
- Output Short-Circuit Protection
- High Slew Rate . . . 13 V/ μ s Typ
- Low Total Harmonic Distortion . . . 0.003% Typ
- Common-Mode Input Voltage Range Includes $V_{CC}+$

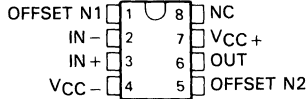
TL080, TL080A
D, JG, OR P PACKAGE

(TOP VIEW)



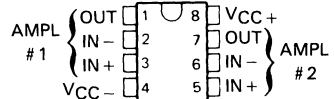
TL081, TL081A, TL081B
D, JG, OR P PACKAGE

(TOP VIEW)



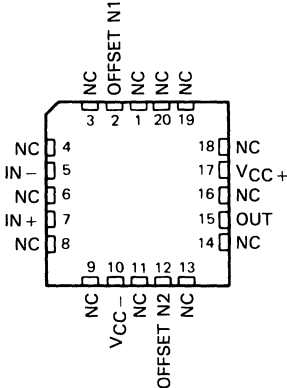
TL082, TL082A, TL082B
D, JG, OR P PACKAGE

(TOP VIEW)



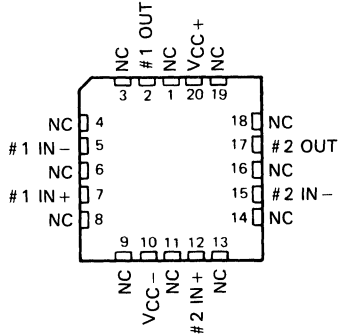
TL081M . . . FK CHIP CARRIER PACKAGE

(TOP VIEW)



TL082M . . . FK CHIP CARRIER PACKAGE

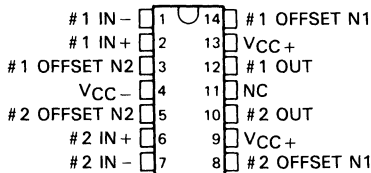
(TOP VIEW)



TL083, TL083A

D, J, OR N PACKAGE

(TOP VIEW)

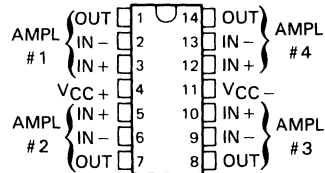


Pins 9 and 13 are internally interconnected

TL084, TL084A, TL084B

D, J, OR N PACKAGE

(TOP VIEW)



2

Operational Amplifiers

NC—No internal connection

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS
INSTRUMENTS**

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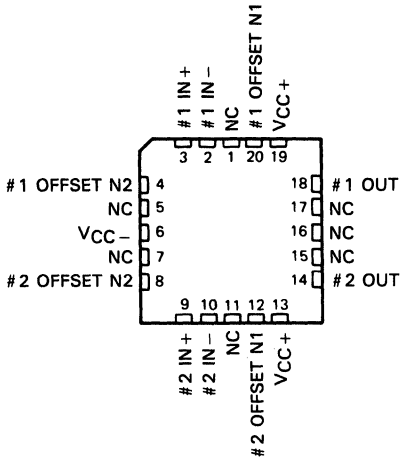
Copyright © 1983, Texas Instruments Incorporated

2-403

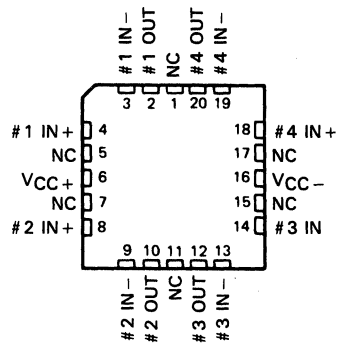
**TL080 THRU TL085, TL080A THRU TL084A
TL081B, TL082B, TL084B
JFET-INPUT OPERATIONAL AMPLIFIERS**

**2
Operational Amplifiers**

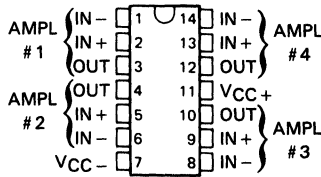
**TL083M . . . FK CHIP CARRIER PACKAGE
(TOP VIEW)**



**TL084M . . . FK CHIP CARRIER PACKAGE
(TOP VIEW)**

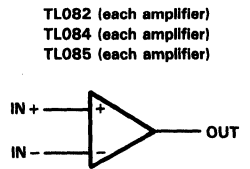
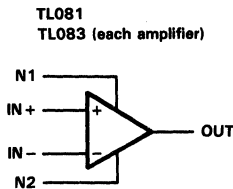
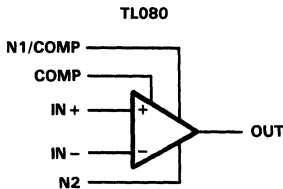


**TL085
N PACKAGE
(TOP VIEW)**



NC—No internal connection

symbols



TL080 THRU TL085, TL080A THRU TL084A TL081B, TL082B, TL084B JFET-INPUT OPERATIONAL AMPLIFIERS

description

The TL08__ JFET-input operational amplifier family is designed to offer a wider selection than any previously developed operational amplifier family. Each of these JFET-input operational amplifiers incorporates well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. The devices feature high slew rates, low input bias and offset currents, and low offset voltage temperature coefficient. Offset adjustment and external compensation options are available within the TL08__ family.

Device types with an "M" suffix are characterized for operation over the full military temperature range of -55°C to 125°C, those with an "I" suffix are characterized for operation from -40°C to 85°C, and those with a "C" suffix are characterized for operation from 0°C to 70°C.

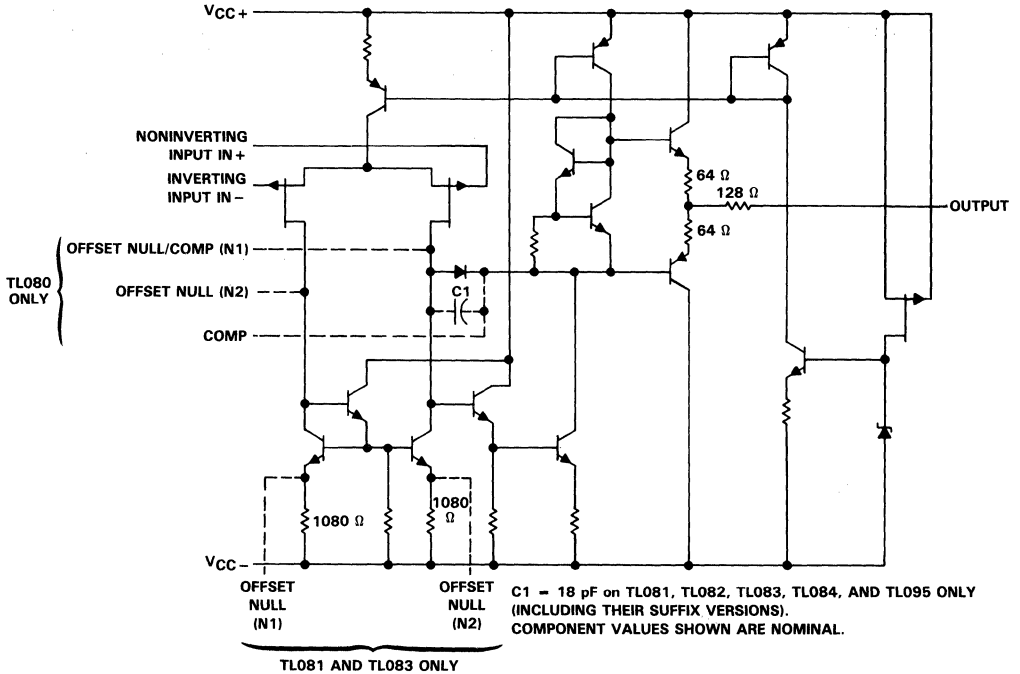
AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE					
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	CERAMIC DIP (JG)	PLASTIC DIP (N)	PLASTIC DIP (P)
0°C TO 70°C	15 mV	TL080CD			TL080CJG		TL080CP
	6 mV	TL080ACD			TL080ACJG		TL080ACP
	15 mV	TL081CD			TL081CJG		TL081CP
	6 mV	TL081ACD			TL081ACJG		TL081ACP
	3 mV	TL081BCD			TL081BCJG		TL081BCP
	15 mV	TL082CD			TL082CJG		TL082CP
	6 mV	TL082ACD			TL082ACJG		TL082ACP
	3 mV	TL082BCD			TL082BCJG		TL082BCP
	15 mV	TL083CD		TL083CJ		TL083CN	
	6 mV	TL083ACD		TL083ACJ		TL083ACN	
	15 mV	TL084CD		TL084CJ		TL084CN	
	6 mV	TL084ACD		TL084ACJ		TL084ACN	
3 mV	TL084BCD		TL084BCJ		TL084BCN		
15 mV					TL085CN		
-40°C TO 85°C	6 mV	TL080ID			TL080IJG		TL080IP
	6 mV	TL081ID			TL081IJG		TL081IP
	6 mV	TL082ID			TL082IJG		TL082IP
	6 mV	TL083ID		TL083IJ		TL083IN	
	6 mV	TL084ID		TL084IJ		TL084IN	
-55°C TO 125°C	6 mV		TL081MFK		TL080MJG		
	6 mV		TL082MFK		TL081MJG		
	6 mV		TL083MFK	TL083MJ	TL082MJG		
	6 mV		TL084MFK	TL084MJ			
	9 mV						

The D package is available taped and reeled. Add "R" suffix to device type (e.g., TL080CDR).

TL080 THRU TL085, TL080A THRU TL084A TL081B, TL082B, TL084B JFET-INPUT OPERATIONAL AMPLIFIERS

schematic (each amplifier)



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	TL08_M	TL08_I	TL08_C TL08_AC TL08_BC	UNIT
Supply voltage, V_{CC+} (see Note 1)	18	18	18	V
Supply voltage, V_{CC-} (see Note 1)	-18	-18	-18	V
Differential input voltage (see Note 2)	±30	±30	±30	V
Input voltage (see Notes 1 and 3)	±15	±15	±15	V
Duration of output short circuit (see Note 4)	unlimited	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table			
Operating free-air temperature range	-55 to 125	-40 to 85	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	°C
Case temperature for 60 seconds	FK package	260		°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J or JG package	300	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D, N, or P package		260	°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
 4. The output may be shorted to ground or to either supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

TL080 THRU TL085, TL080A THRU TL084A TL081B, TL082B, TL084B JFET-INPUT OPERATIONAL AMPLIFIERS

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$		DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$		$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	POWER RATING			POWER RATING	POWER RATING		
D (8 Pin)	680 mW	5.8 mW/ $^\circ\text{C}$		32 $^\circ\text{C}$	464 mW	377 mW	N/A	
D (14 Pin)	680 mW	7.6 mW/ $^\circ\text{C}$		60 $^\circ\text{C}$	608 mW	494 mW	N/A	
FK	680 mW	11.0 mW/ $^\circ\text{C}$		88 $^\circ\text{C}$	680 mW	680 mW	275 mW	
J (TL08_M)	680 mW	11.0 mW/ $^\circ\text{C}$		88 $^\circ\text{C}$	680 mW	680 mW	275 mW	
J (all others)	680 mW	8.2 mW/ $^\circ\text{C}$		67 $^\circ\text{C}$	656 mW	533 mW	N/A	
JG (TL08_M)	680 mW	8.4 mW/ $^\circ\text{C}$		69 $^\circ\text{C}$	672 mW	546 mW	210 mW	
JG (all others)	680 mW	6.6 mW/ $^\circ\text{C}$		47 $^\circ\text{C}$	528 mW	429 mW	N/A	
N	680 mW	9.2 mW/ $^\circ\text{C}$		76 $^\circ\text{C}$	680 mW	598 mW	N/A	
P	680 mW	8.0 mW/ $^\circ\text{C}$		65 $^\circ\text{C}$	640 mW	520 mW	N/A	

2

Operational Amplifiers

electrical characteristics, $V_{CC} \pm = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS [†]		TL080M, TL081M TL082M, TL083M			TL084M			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0,$ $R_S = 50\ \Omega$	$T_A = 25^\circ\text{C}$	3 6		3 9				mV
		$T_A = -55^\circ\text{C}$ to 125°C	9		15				
α_{VIO} Temperature coefficient of input offset voltage	$V_O = 0,$ $T_A = -55^\circ\text{C}$ to 125°C	$R_S = 50\ \Omega,$	18		18				$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current [‡]	$V_O = 0$	$T_A = 25^\circ\text{C}$	5	100	5	100			pA
		$T_A = 125^\circ\text{C}$	20		20				nA
I_{IB} Input bias current [‡]	$V_O = 0$	$T_A = 25^\circ\text{C}$	30 200		30 200				pA
		$T_A = 125^\circ\text{C}$	50		50				nA
V_{ICR} Common-mode input voltage range	$T_A = 25^\circ\text{C}$		-12 ± 11 to 15		-12 ± 11 to 15				V
V_{OM} Maximum peak output voltage swing	$T_A = 25^\circ\text{C},$		± 12 ± 13.5		± 12 ± 13.5				V
	$T_A = -55^\circ\text{C}$ to 125°C		$R_L \geq 10\ \text{k}\Omega$		± 12				
			$R_L \geq 2\ \text{k}\Omega$		± 10 ± 12		± 10 ± 12		
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 10\ \text{V},$ $T_A = 25^\circ\text{C}$		$R_L \geq 2\ \text{k}\Omega,$		25 200		25 200		V/mV
	$V_O = \pm 10\ \text{V},$ $T_A = -55^\circ\text{C}$ to 125°C		$R_L \geq 2\ \text{k}\Omega,$		15		15		
B_1 Unity-gain bandwidth	$T_A = 25^\circ\text{C}$		3		3				MHz
r_i Input resistance	$T_A = 25^\circ\text{C}$		10^{12}		10^{12}				Ω
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR}\ \text{min},$ $R_S = 50\ \Omega,$	$V_O = 0,$ $T_A = 25^\circ\text{C}$	80	86	80	86			dB
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC} \pm / \Delta V_{IO}$)	$V_{CC} = \pm 15\ \text{V}$ to $\pm 9\ \text{V},$ $V_O = 0,$ $R_S = 50\ \Omega,$ $T_A = 25^\circ\text{C}$		80	86	80	86			dB
I_{CC} Supply current (per amplifier)	No load, $T_A = 25^\circ\text{C}$		$V_O = 0,$ 1.4 2.8		1.4 2.8				mA
V_{O1}/V_{O2} Crosstalk attenuation	$A_{VD} = 100,$ $T_A = 25^\circ\text{C}$		120		120				dB

[†] All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified.

[‡] Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 18. Pulse techniques must be used that will maintain the junction temperatures as close to the ambient temperature as is possible.

electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†		TL080I			TL080C			TL080AC			TL081BC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0,$ $R_S = 50\ \Omega$	$T_A = 25^\circ\text{C}$	3 6			3 15			3 6			2 3			mV
		$T_A = \text{full range}$	9			20			7.5			5			
α_{VIO} Temperature coefficient of input offset voltage	$V_O = 0,$ $T_A = \text{full range}$	$R_S = 50\ \Omega,$	18			18			18			18			$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current‡	$V_O = 0$	$T_A = 25^\circ\text{C}$	5 100			5 200			5 100			5 100			pA
		$T_A = \text{full range}$	10			2			2			2			nA
I_{IB} Input bias current‡	$V_O = 0$	$T_A = 25^\circ\text{C}$	30 200			30 400			30 200			30 200			pA
		$T_A = \text{full range}$	20			10			7			7			nA
V_{ICR} Common-mode input voltage range	$T_A = 25^\circ\text{C}$		-12 to 15			-12 to 15			-12 to 15			-12 to 15			V
			± 11			± 11			± 11			± 11			
V_{OM} Maximum peak output voltage swing	$T_A = 25^\circ\text{C},$ $T_A = \text{full range}$	$R_L = 10\ \text{k}\Omega$	$\pm 12 \pm 13.5$			$\pm 12 \pm 13.5$			$\pm 12 \pm 13.5$			$\pm 12 \pm 13.5$			V
		$R_L \geq 10\ \text{k}\Omega$	± 12			± 12			± 12			± 12			
		$R_L \geq 2\ \text{k}\Omega$	$\pm 10 \pm 12$			$\pm 10 \pm 12$			$\pm 10 \pm 12$			$\pm 10 \pm 12$			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 10\ \text{V},$ $T_A = 25^\circ\text{C}$	$R_L \geq 2\ \text{k}\Omega,$	50 200			25 200			50 200			50 200			V/mV
		$T_A = \text{full range}$	25			15			25			25			
B_1 Unity-gain bandwidth	$T_A = 25^\circ\text{C}$		3			3			3			3			MHz
r_i Input resistance	$T_A = 25^\circ\text{C}$		10^{12}			10^{12}			10^{12}			10^{12}			Ω
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\ \text{min}}, V_O = 0,$ $R_S = 50\ \Omega, T_A = 25^\circ\text{C}$		80 86			70 86			80 86			80 86			dB
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC\pm}/\Delta V_{IO}$)	$V_{CC} = \pm 15\ \text{V to } \pm 9\ \text{V}, V_O = 0,$ $R_S = 50\ \Omega, T_A = 25^\circ\text{C}$		80 86			70 86			80 86			80 86			dB
I_{CC} Supply current (per amplifier)	No load, $T_A = 25^\circ\text{C}$	$V_O = 0,$	1.4 2.8			1.4 2.8			1.4 2.8			1.4 2.8			mA
V_{O1}/V_{O2} Crosstalk attenuation	$AVD = 100, T_A = 25^\circ\text{C}$		120			120			120			120			dB

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for T_A is -40°C to 85°C for TL080_I and 0°C to 70°C for TL080_C, TL080_AC, and TL080_BC.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 18. Pulse techniques must be used that will maintain the junction temperatures as close to the ambient temperature as is possible.

TL080 THRU TL085, TL080A THRU TL084A
TL081B, TL082B, TL084B
JFET-INPUT OPERATIONAL AMPLIFIERS

TL080 THRU TL085, TL080A THRU TL084A TL081B, TL082B, TL084B JFET-INPUT OPERATIONAL AMPLIFIERS

operating characteristics, $V_{CC} \pm = \pm 15 \text{ V}$, $T_A = 25^\circ \text{C}$

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$V_I = 10 \text{ V}$, $C_L = 100 \text{ pF}$,	$R_L = 2 \text{ k}\Omega$, See Figure 1	8	13		$\text{V}/\mu\text{s}$
t_r	Rise time	$V_I = 20 \text{ mV}$,	$R_L = 2 \text{ k}\Omega$,		0.05		μs
	Overshoot factor	$C_L = 100 \text{ pF}$,	See Figure 1		20%		
V_n	Equivalent input noise voltage	$R_S = 100 \Omega$	$f = 1 \text{ kHz}$		18		$\text{nV}/\sqrt{\text{Hz}}$
			$f = 10 \text{ Hz to } 10 \text{ kHz}$		4		μV
I_n	Equivalent input noise current	$R_S = 100 \Omega$,	$f = 1 \text{ kHz}$		0.01		$\text{pA}/\sqrt{\text{Hz}}$
THD	Total harmonic distortion	$V_{O(\text{rms})} = 10 \text{ V}$, $R_L \geq 2 \text{ k}\Omega$,	$R_S \leq 1 \text{ k}\Omega$, $f = 1 \text{ kHz}$		0.003%		

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Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

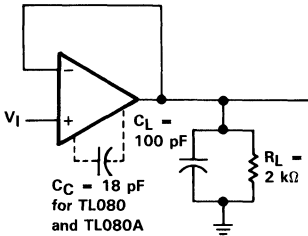


FIGURE 1. UNITY-GAIN AMPLIFIER

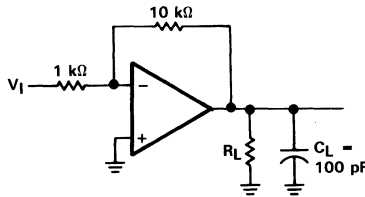


FIGURE 2. GAIN-OF-10
INVERTING AMPLIFIER

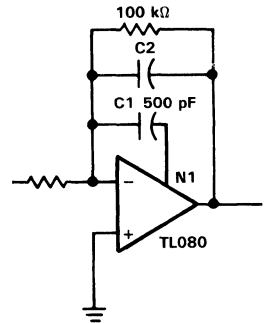


FIGURE 3. FEED-FORWARD
COMPENSATION

INPUT OFFSET VOLTAGE NULL CIRCUITS

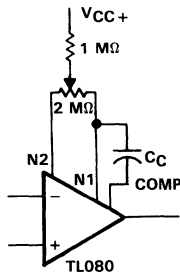


FIGURE 4

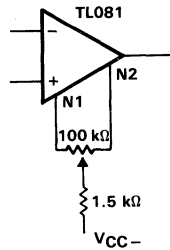


FIGURE 5

**TL080 THRU TL085, TL080A THRU TL084A
TL081B, TL082B, TL084B
JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS†

**MAXIMUM PEAK OUTPUT VOLTAGE
vs
FREQUENCY**

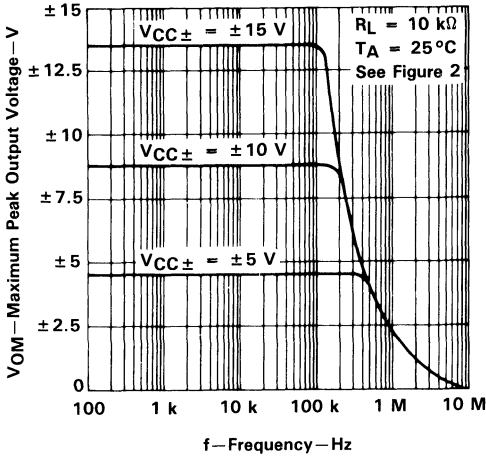


FIGURE 6

**MAXIMUM PEAK OUTPUT VOLTAGE
vs
FREQUENCY**

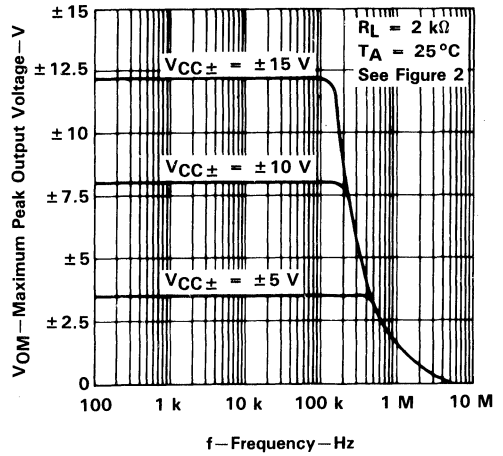


FIGURE 7

**MAXIMUM PEAK OUTPUT VOLTAGE
vs
FREQUENCY**

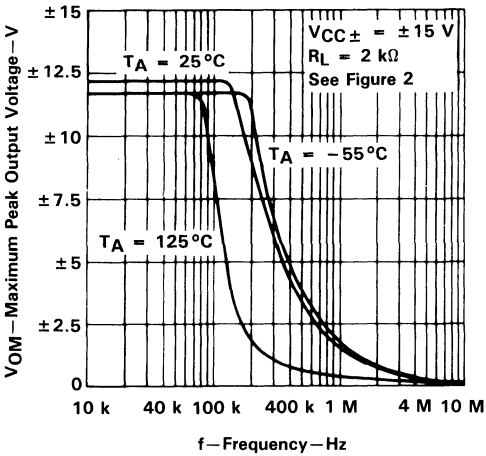


FIGURE 8

**MAXIMUM PEAK OUTPUT VOLTAGE
vs
FREE-AIR TEMPERATURE**

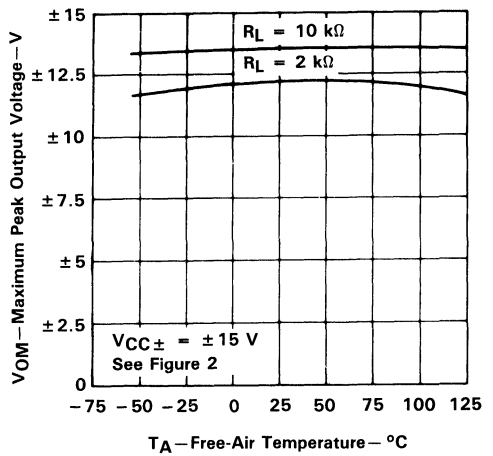


FIGURE 9

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. A 12-pF compensation capacitor is used with TL080 and TL080A.

**TL080 THRU TL085, TL080A THRU TL084A
TL081B, TL082B, TL084B
JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS†

**MAXIMUM PEAK OUTPUT VOLTAGE
vs
LOAD RESISTANCE**

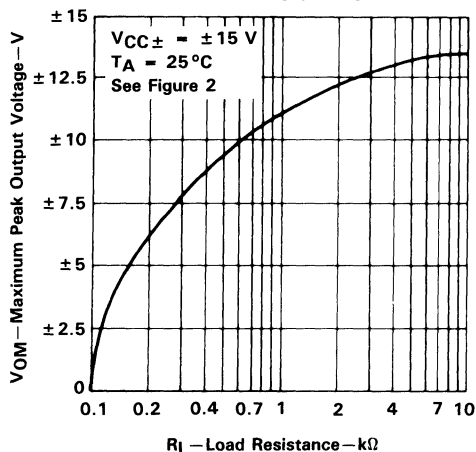


FIGURE 10

**MAXIMUM PEAK OUTPUT VOLTAGE
vs
SUPPLY VOLTAGE**

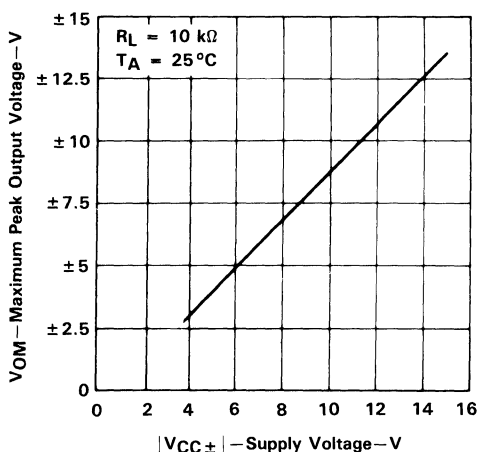


FIGURE 11

**LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREE-AIR TEMPERATURE**

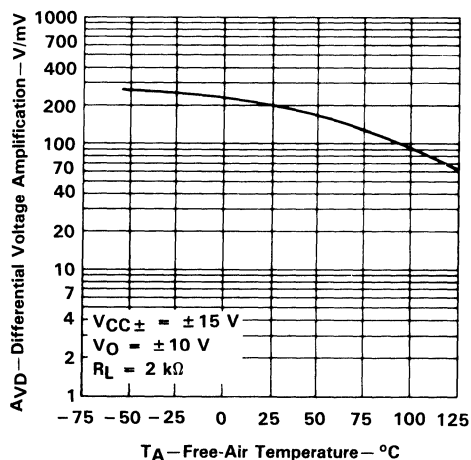


FIGURE 12

**LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
and PHASE SHIFT
vs
FREQUENCY**

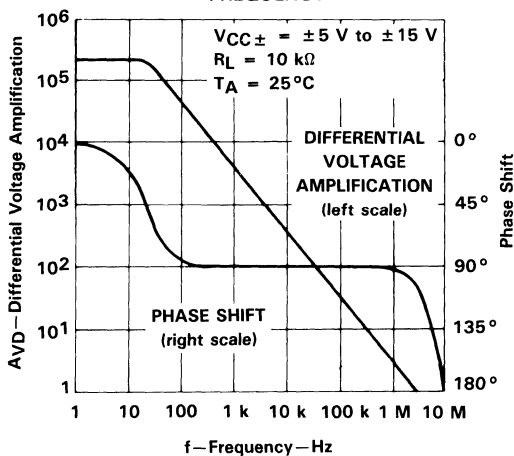


FIGURE 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. A 12-pF compensation capacitor is used with TL080 and TL080A.

**TL080 THRU TL085, TL080A THRU TL084A
TL081B, TL082B, TL084B
JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS†

**TL080, TL080A
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREQUENCY WITH FEED-FORWARD COMPENSATION**

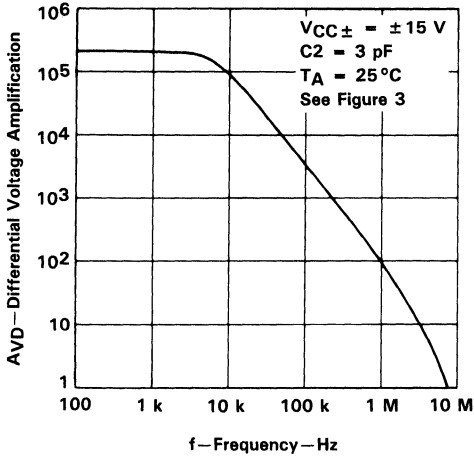


FIGURE 14

**TOTAL POWER DISSIPATED
vs
FREE-AIR TEMPERATURE**

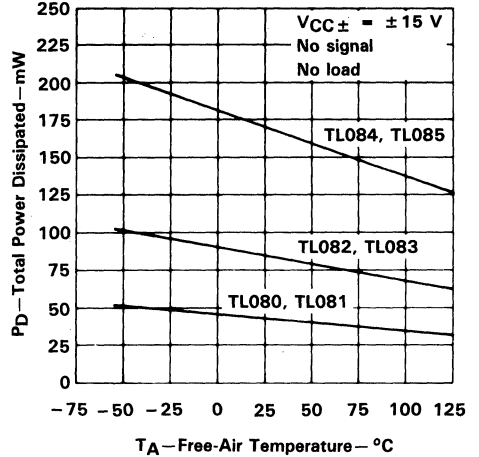


FIGURE 15

**SUPPLY CURRENT PER AMPLIFIER
vs
FREE-AIR TEMPERATURE**

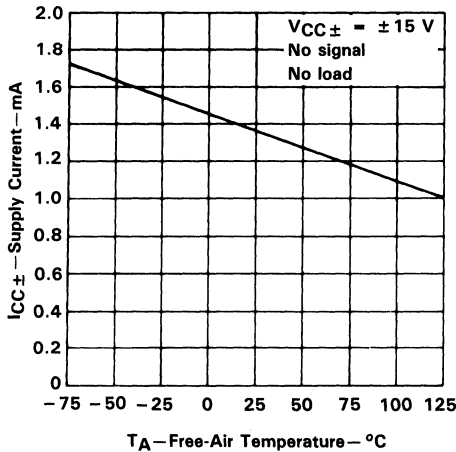


FIGURE 16

**SUPPLY CURRENT
vs
SUPPLY VOLTAGE**

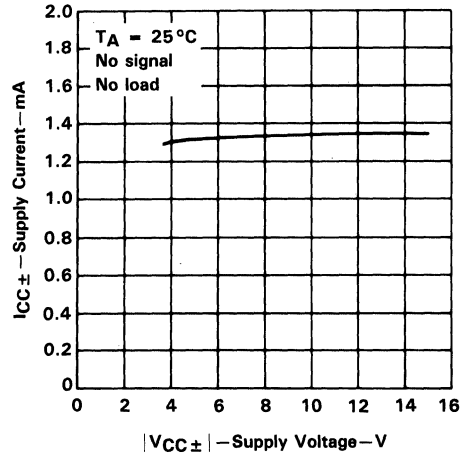


FIGURE 17

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. A 12-pF compensation capacitor is used with TL080 and TL080A.

**TL080 THRU TL085, TL080A THRU TL084A
TL081B, TL082B, TL084B
JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS†

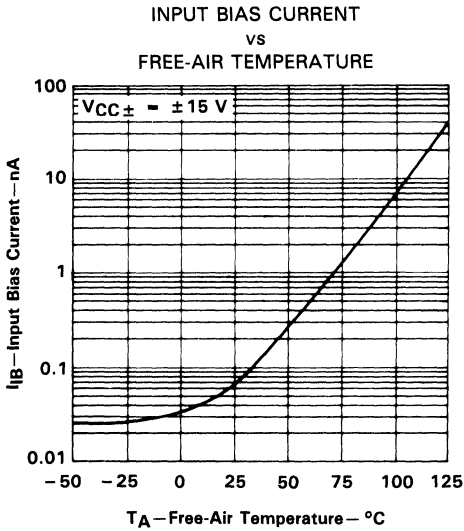


FIGURE 18

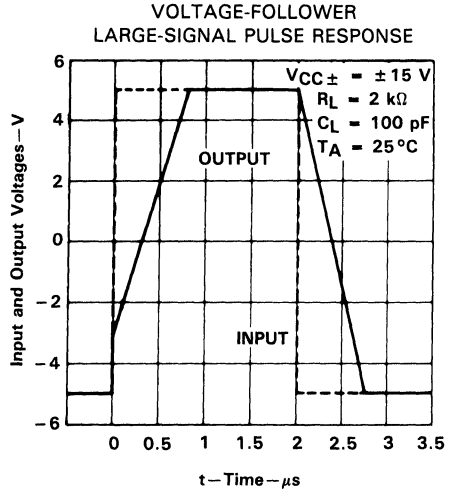


FIGURE 19

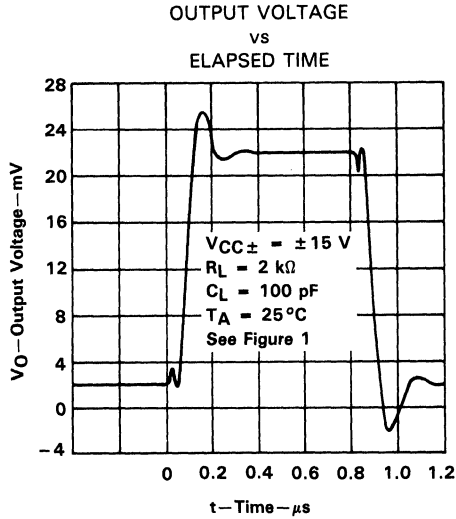


FIGURE 20

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. A 12-pF compensation capacitor is used with TL080 and TL080A.

**TL080 THRU TL085, TL080A THRU TL084A
TL081B, TL082B, TL084B
JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL CHARACTERISTICS†

**COMMON-MODE REJECTION RATIO
vs
FREE-AIR TEMPERATURE**

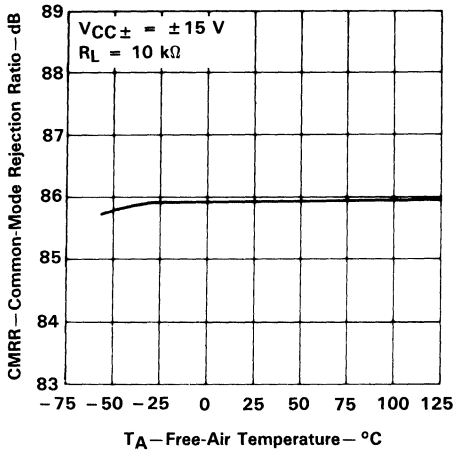


FIGURE 21

**EQUIVALENT INPUT NOISE VOLTAGE
vs
FREQUENCY**

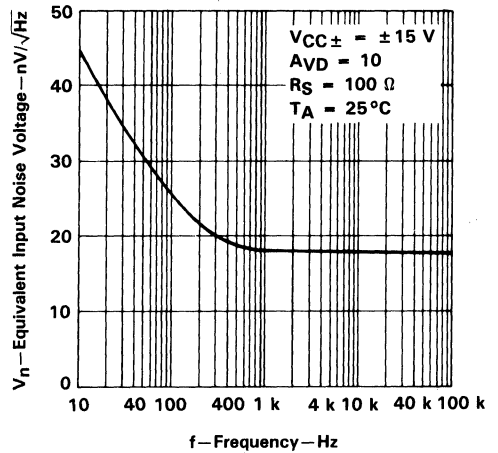


FIGURE 22

**TOTAL HARMONIC DISTORTION
vs
FREQUENCY**

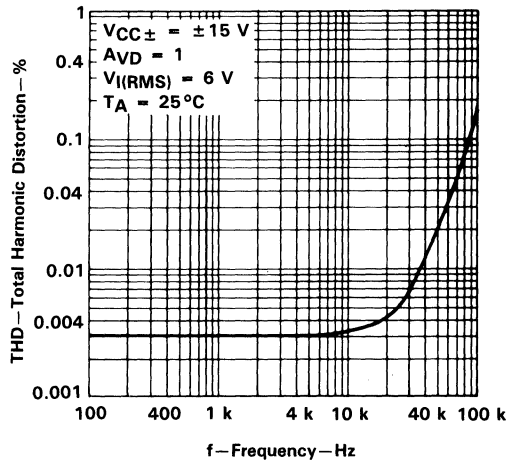


FIGURE 23

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. A 12-pF compensation capacitor is used with TL080 and TL080A.

**TL080 THRU TL085, TL080A THRU TL084A
TL081B, TL082B, TL084B
JFET-INPUT OPERATIONAL AMPLIFIERS**

TYPICAL APPLICATION DATA

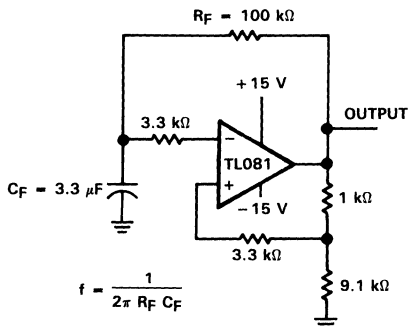


FIGURE 24. 0.5-Hz SQUARE-WAVE OSCILLATOR

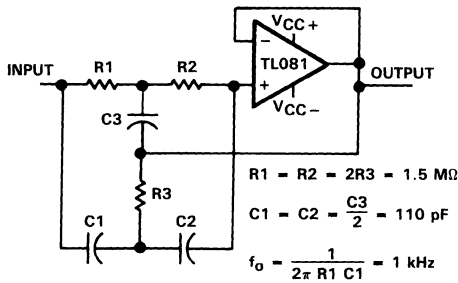


FIGURE 25. HIGH-Q NOTCH FILTER

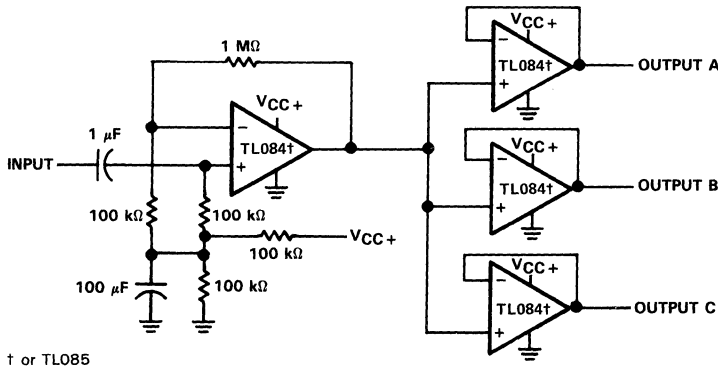
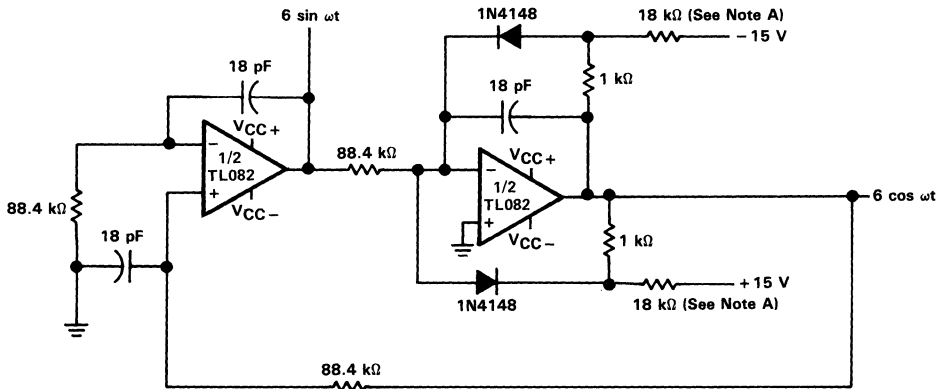


FIGURE 26. AUDIO DISTRIBUTION AMPLIFIER



NOTE A: These resistor values may be adjusted for a symmetrical output.

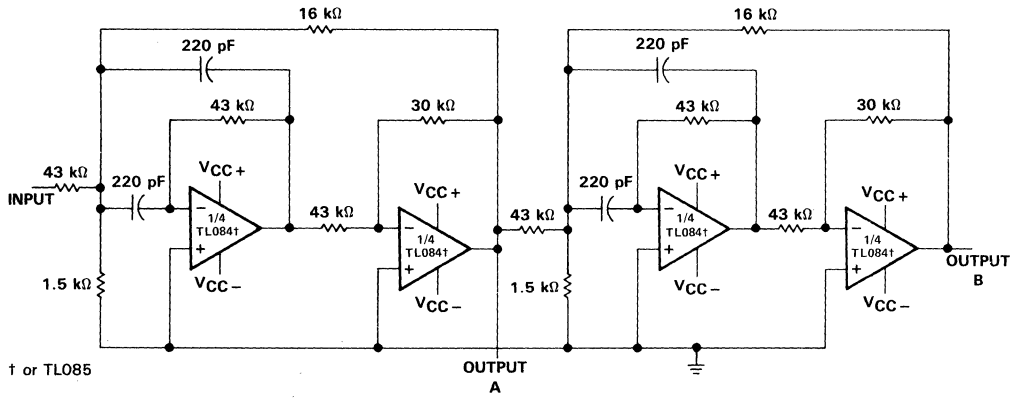
FIGURE 27. 100-kHz QUADRATURE OSCILLATOR

**TL080 THRU TL085, TL080A THRU TL084A
TL081B, TL082B, TL084B
JFET-INPUT OPERATIONAL AMPLIFIERS**

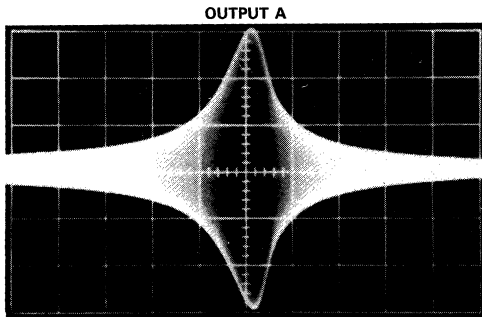
TYPICAL APPLICATION DATA

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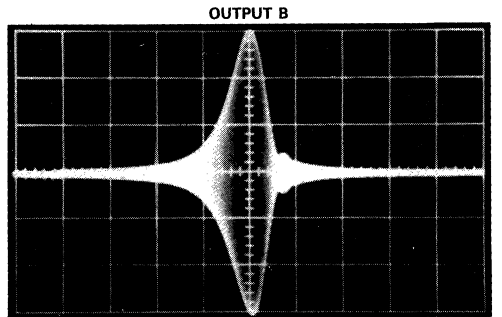
† or TL085



OUTPUT A

2 kHz/div

SECOND-ORDER BANDPASS FILTER
 $f_o = 100 \text{ kHz}$, $Q = 30$, $GAIN = 4$



OUTPUT B

2 kHz/div

CASCADED BANDPASS FILTER
 $f_o = 100 \text{ kHz}$, $Q = 69$, $GAIN = 16$

FIGURE 28. POSITIVE-FEEDBACK BANDPASS FILTER

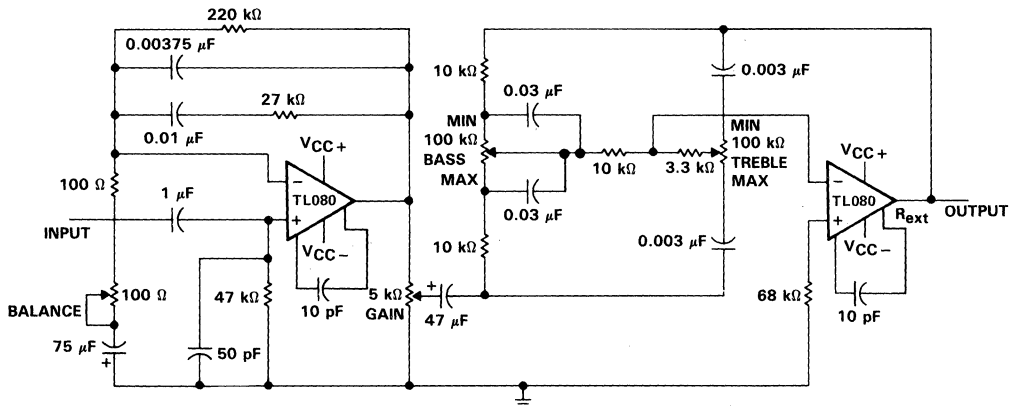


FIGURE 29. IC PREAMPLIFIER

TL087, TL088, TL287, TL288 JFET-INPUT OPERATIONAL AMPLIFIERS

D2484, MARCH 1979—REVISED MARCH 1989

- Low Input Offset Voltage . . . 0.5 mV Max
- Low Power Consumption
- Wide Common-Mode and Differential Voltage Ranges
- Low Input Bias and Offset Currents
- High Input Impedance . . . JFET-Input Stage
- Internal Frequency Compensation
- Latch-Up-Free Operation
- High Slew Rate . . . 18 V/ μ s Typ
- Low Total Harmonic Distortion . . . 0.003% Typ

description

These JFET-input operational amplifiers incorporate well-matched high-voltage JFET and bipolar transistors in a monolithic integrated circuit. They feature low input offset voltage, high slew rate, low input bias and offset currents, and low temperature coefficient of input offset voltage. Offset-voltage adjustment is provided for the TL087 and TL088.

The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I-suffix devices are characterized for operation from -40°C to 85°C , and the C-suffix devices are characterized for operation from 0°C to 70°C .

AVAILABLE OPTIONS

T _A	TYPE	V _{IO} MAX AT 25°C	PACKAGE				
			SMALL OUTLINE (D)	CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)	FLAT (U)
0°C to 70°C	Single	0.5 mV 1 mV	TL087CD TL088CD	TL087CJG TL088CJG	TL087CL TL088CL	TL087CP TL088CP	
	Dual	0.5 mV 1 mV	TL287CD TL288CD	TL287CJG TL288CJG	TL287CL TL288CL	TL287CP TL288CP	
-40°C to 85°C	Single	0.5 mV 1 mV	TL087ID TL088ID	TL087IJG TL088IJG	TL087IL TL088IL	TL087IP TL088IP	
	Dual	0.5 mV 1 mV	TL287ID TL288ID	TL287IJG TL288IJG	TL287IL TL288IL	TL287IP TL288IP	
-55°C to 125°C	Single	1 mV		TL088MJG	TL088ML		TL088MU
	Dual	1 mV		TL288MJG	TL288ML		TL288MU

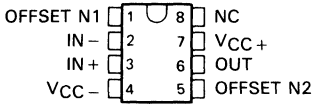
The D package is available taped and reeled. Add the suffix R to the device type (e.g., TL087CDR).

2

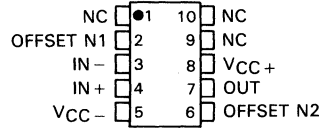
Operational Amplifiers

TL087, TL088, TL287, TL288 JFET-INPUT OPERATIONAL AMPLIFIERS

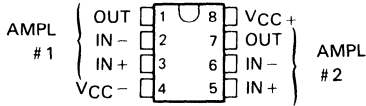
TL087, TL088
D, JG, OR P PACKAGE
(TOP VIEW)



TL088M
U PACKAGE
(TOP VIEW)

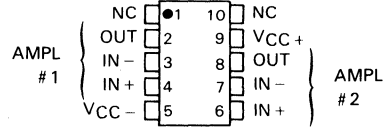


TL287, TL288
D, JG, OR P PACKAGE
(TOP VIEW)



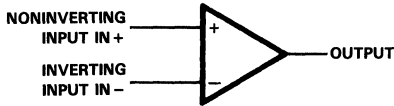
NC—No internal connection

TL288M
U PACKAGE
(TOP VIEW)

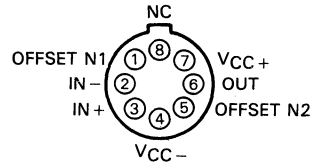


2 Operational Amplifiers

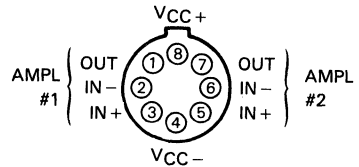
symbol (each amplifier)



TL087, TL088
L PACKAGE
(TOP VIEW)



TL287, TL288
L PACKAGE
(TOP VIEW)



Pin 4 (L Package) is in electrical contact with the case
NC—No internal connection

TL087, TL088, TL287, TL288 JFET-INPUT OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	TL088M TL288M	TL087I TL088I TL287I TL288I	TL087C TL088C TL287C TL288C	UNIT
Supply voltage, V_{CC+} (see Note 1)	18	18	18	V
Supply voltage, V_{CC-} (see Note 1)	-18	-18	-18	V
Differential input voltage (see Note 2)	± 30	± 30	± 30	V
Input voltage (see Notes 1 and 3)	± 15	± 15	± 15	V
Input current, I_I (each input)	± 1	± 1	± 1	mA
Output current, I_O (each output)	± 80	± 80	± 80	mA
Total V_{CC+} terminal current	160	160	160	mA
Total V_{CC-} terminal current	-160	-160	-160	mA
Duration of output short circuit (see Note 4)	unlimited	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table			
Operating free-air temperature range	-55 to 125	-25 to 85	0 to 70	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	JG, L, or U package	300	300	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package		260	$^{\circ}\text{C}$

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
 4. The output may be shorted to ground or to either supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$	DERATING FACTOR	$T_A = 70^{\circ}\text{C}$	$T_A = 85^{\circ}\text{C}$	$T_A = 125^{\circ}\text{C}$
	POWER RATING	ABOVE $T_A = 25^{\circ}\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/ $^{\circ}\text{C}$	464 mW	377 mW	N/A
JG	1050 mW	8.4 mW/ $^{\circ}\text{C}$	672 mW	546 mW	210 mW
L	650 mW	5.2 mW/ $^{\circ}\text{C}$	416 mW	338 mW	130 mW
P	1000 mW	8.0 mW/ $^{\circ}\text{C}$	640 mW	520 mW	N/A
U	675 mW	5.4 mW/ $^{\circ}\text{C}$	432 mW	351 mW	135 mW

recommended operating conditions

		M-SUFFIX		I-SUFFIX		C-SUFFIX		UNIT			
		MIN	NOM	MAX	MIN	NOM	MAX		MIN	NOM	MAX
Supply voltage, V_{CC}		± 5		± 15	± 5		± 15	± 5		± 15	V
Common-mode input voltage, V_{IC}	$V_{CC\pm} = \pm 5\text{ V}$	-1		4	-1		4	-1		4	V
	$V_{CC\pm} = \pm 15\text{ V}$	-11		11	-11		11	-11		11	V
Input voltage, V_I	$V_{CC\pm} = \pm 5\text{ V}$	-1		4	-1		4	-1		4	V
	$V_{CC\pm} = \pm 15\text{ V}$	-11		11	-11		11	-11		11	V
Operating free-air temperature, T_A		-55		125	-40		85	0		70	$^{\circ}\text{C}$

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electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$

PARAMETER	TEST CONDITIONS†		TL088M TL288M			TL087I TL088I TL287I TL288I			TL087C TL088C TL287C TL288C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$R_S = 50\ \Omega$, $V_O = 0$, $T_A = 25^\circ\text{C}$	TL087, TL287				0.1	0.5		0.1	0.5	mV	
		TL088, TL288	0.1	3		0.1	1		0.1	1		
	$R_S = 50\ \Omega$, $V_O = 0$, $T_A = \text{full range}$	TL087, TL287					2			1.5		
		TL088, TL288		6			3			2.5		
αV_{IO} Temperature coefficient of input offset voltage	$R_S = 50\ \Omega$, $T_A = 25^\circ\text{C}$ to MAX		10			8			8	$\mu\text{V}/^\circ\text{C}$		
I_{IO} Input offset current	$T_A = 25^\circ\text{C}$		5			5	100		5	100	pA	
	$T_A = \text{full range}$			25			3			2	nA	
I_{IB} Input bias current‡	$T_A = 25^\circ\text{C}$		30			30	200		30	200	pA	
	$T_A = \text{full range}$			100			20			7	nA	
V_{ICR} Common-mode input voltage range	$T_A = 25^\circ\text{C}$		$V_{CC-} + 4$ to $V_{CC+} - 4$			$V_{CC-} + 4$ to $V_{CC+} - 4$			$V_{CC-} + 4$ to $V_{CC+} - 4$		V	
V_{OPP} Maximum-peak-to-peak output voltage swing	$T_A = 25^\circ\text{C}$, $T_A = \text{full range}$	$R_L = 10\ \text{k}\Omega$	24	27		24	27		24	27	V	
		$R_L \geq 10\ \text{k}\Omega$	24			24			24			
	$R_L \geq 2\ \text{k}\Omega$	20			20			20				
A_{VD} Large-signal differential voltage amplification	$R_L \geq 2\ \text{k}\Omega$, $T_A = 25^\circ\text{C}$	$V_O = \pm 10\ \text{V}$,	50	105		50	105		50	105	V/mV	
		$V_O = \pm 10\ \text{V}$,	25			25			25			
	$T_A = \text{full range}$											
B_1 Unity-gain bandwidth	$T_A = 25^\circ\text{C}$		3			3			3	MHz		
r_i Input resistance	$T_A = 25^\circ\text{C}$		10^{12}			10^{12}			10^{12}		Ω	
CMRR Common-mode rejection ratio	$R_S = 50\ \Omega$, $V_O = 0\ \text{V}$, $V_{IC} = V_{ICR\ \text{min}}$, $T_A = 25^\circ\text{C}$		80	93		80	93		80	93	dB	
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC\pm} / \Delta V_{IO}$)	$R_S = 50\ \Omega$, $V_O = 0\ \text{V}$, $V_{CC\pm} = \pm 9\ \text{V}$ to $\pm 15\ \text{V}$, $T_A = 25^\circ\text{C}$		80	99		80	99		80	99	dB	
I_{CC} Supply current (per amplifier)	No load, $T_A = 25^\circ\text{C}$, $V_O = 0$,		2.6	2.8		2.6	2.8		2.6	2.8	mA	

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for T_A is -55°C to 125°C for TL_88M; -40°C to 85°C for TL_8_I; and 0°C to 70°C for TL_8_C.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

TL087, TL088, TL287, TL288 JFET-INPUT OPERATIONAL AMPLIFIERS

operating characteristics $V_{CC} = \pm 15 \text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TL088M, TL288M			TL087I, TL087C TL088I, TL088C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_I = 10 \text{ V}$, $R_L = 2 \text{ k}\Omega$, $C_L = 100 \text{ pF}$, $A_{VD} = 1$	18			8	18		$\text{V}/\mu\text{s}$
t_r Rise time	$V_I = 20 \text{ mV}$, $R_L = 2 \text{ k}\Omega$, $C_L = 100 \text{ pF}$, $A_{VD} = 1$	55			55			ns
Overshoot factor	$C_L = 100 \text{ pF}$, $A_{VD} = 1$	25%			25%			
V_n Equivalent input noise voltage	$R_S = 100 \Omega$, $f = 1 \text{ kHz}$	19			19			$\text{nV}/\sqrt{\text{Hz}}$

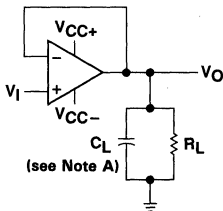
2

Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

2

Operational Amplifiers



NOTE A: C_L includes fixture capacitance.

FIGURE 1. SLEW RATE, RISE/FALL TIME, AND OVERSHOOT TEST CIRCUIT

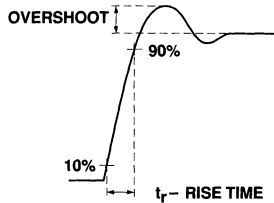


FIGURE 2. RISE TIME AND OVERSHOOT WAVEFORM

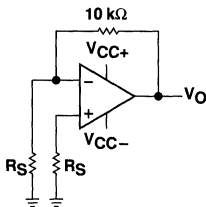


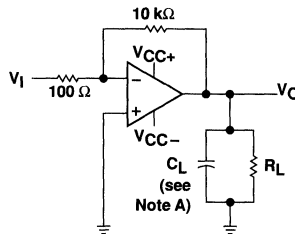
FIGURE 3. NOISE VOLTAGE TEST CIRCUIT

typical values

Typical values as presented in this data sheet represent the median (50% point) of device parametric performance.

input bias and offset current

At the picoamp bias current level typical of these JFET operational amplifiers, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To accurately measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied, but with no device in the socket. The device is then inserted in the socket and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.



NOTE A: C_L includes fixture capacitance.

FIGURE 4. UNITY-GAIN BANDWIDTH AND PHASE MARGIN TEST CIRCUIT

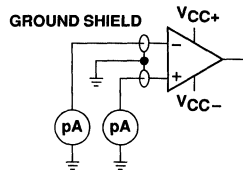


FIGURE 5. INPUT BIAS AND OFFSET CURRENT TEST CIRCUIT

TYPICAL CHARACTERISTICS

table of graphs

		FIGURE	
αV_{IO}	Temperature coefficient of input offset voltage	Distribution	6, 7
I_{IO}	Input offset current	vs Temperature	8
I_{IB}	Input bias current	vs V_{IC}	9
		vs Temperature	8
V_I	Common-mode input voltage range limits	vs V_{CC}	10
		vs Temperature	11
V_{ID}	Differential input voltage	vs Output voltage	12
		vs V_{CC}	13
V_{OM}	Maximum peak output voltage swing	vs Output current	17
		vs Frequency	14, 15, 16
		vs Temperature	18
A_{VD}	Differential voltage amplification	vs R_L	19
		vs Frequency	20
		vs Temperature	21
z_o	Output impedance	vs Frequency	24
$CMRR$	Common-mode rejection ratio	vs Frequency	22
		vs Temperature	23
k_{SVR}	Supply-voltage rejection ratio	vs Temperature	25
		vs V_{CC}	26
I_{OS}	Short-circuit output current	vs Time	27
		vs Temperature	28
I_{CC}	Supply current	vs V_{CC}	29
		vs Temperature	30
SR	Slew Rate	vs R_L	31
		vs Temperature	32
	Overshoot factor	vs C_L	33
V_n	Equivalent input noise voltage	vs Frequency	34
THD	Total harmonic distortion	vs Frequency	35
B_1	Unity-gain bandwidth	vs V_{CC}	36
		vs Temperature	37
		vs V_{CC}	38
ϕ_m	Phase margin	vs C_L	39
		vs Temperature	40
	Phase shift	vs Frequency	20
	Pulse response	Small-signal	41
		Large-signal	42

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Operational Amplifiers

TL087, TL088, TL287, TL288
JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

DISTRIBUTION OF TL088
INPUT OFFSET VOLTAGE
TEMPERATURE COEFFICIENT

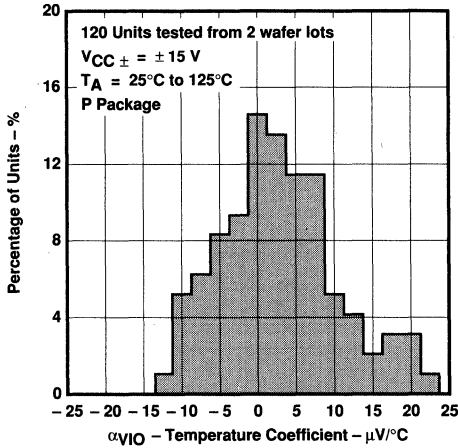


FIGURE 6

DISTRIBUTION OF TL288
INPUT OFFSET VOLTAGE
TEMPERATURE COEFFICIENT

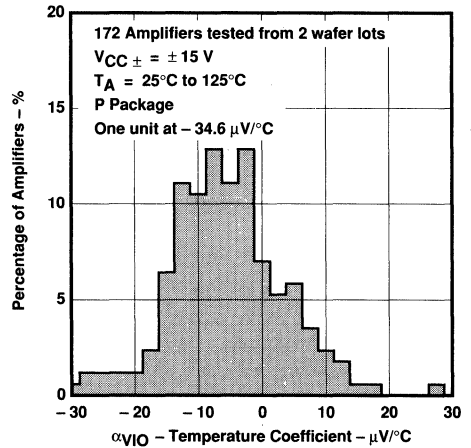


FIGURE 7

INPUT BIAS CURRENT AND
INPUT OFFSET CURRENT
VS
FREE-AIR TEMPERATURE

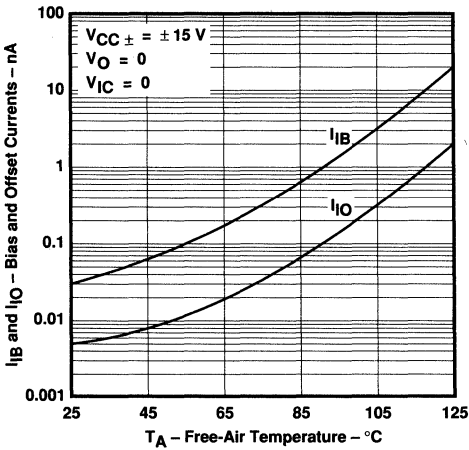


FIGURE 8

INPUT BIAS CURRENT
VS
COMMON-MODE INPUT VOLTAGE

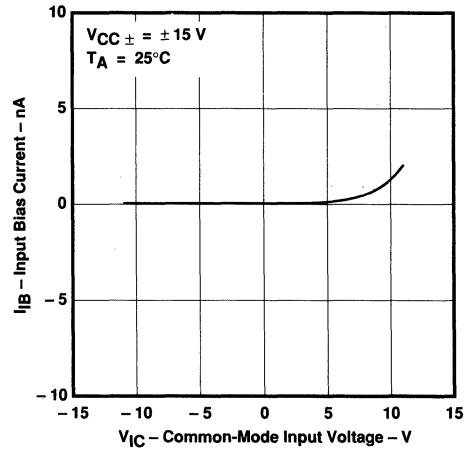


FIGURE 9

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

COMMON-MODE
 INPUT VOLTAGE RANGE LIMITS
 VS
 SUPPLY VOLTAGE

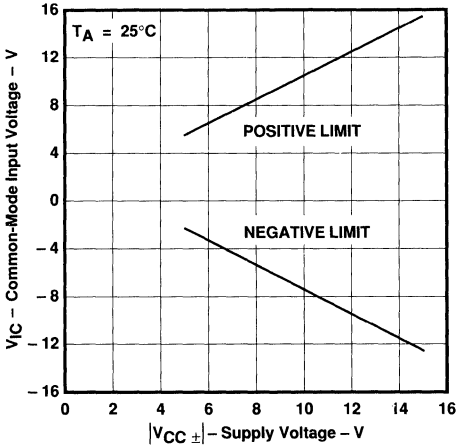


FIGURE 10

COMMON-MODE
 INPUT VOLTAGE RANGE LIMITS
 VS
 FREE-AIR TEMPERATURE

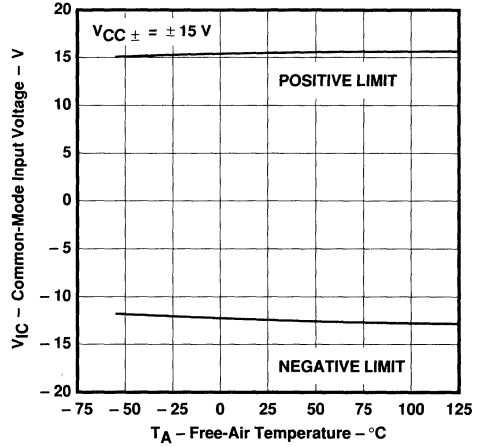


FIGURE 11

OUTPUT VOLTAGE
 VS
 DIFFERENTIAL INPUT VOLTAGE

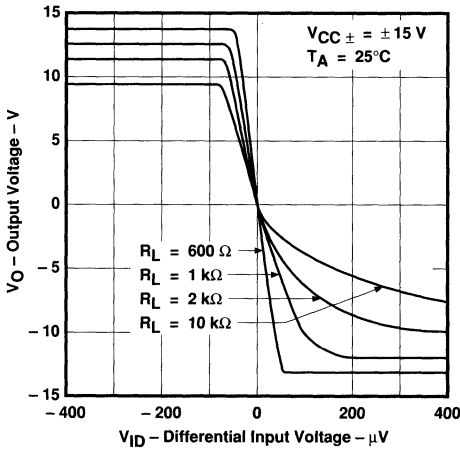


FIGURE 12

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 SUPPLY VOLTAGE

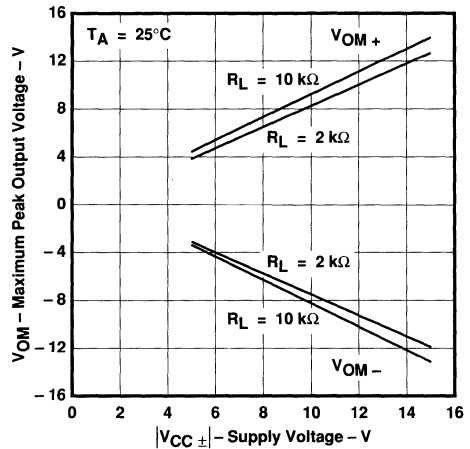


FIGURE 13

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 Operational Amplifiers

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

**MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 VS
 FREQUENCY**

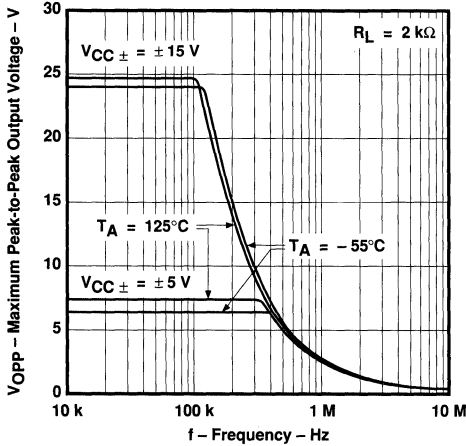


FIGURE 14

**MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 VS
 FREQUENCY**

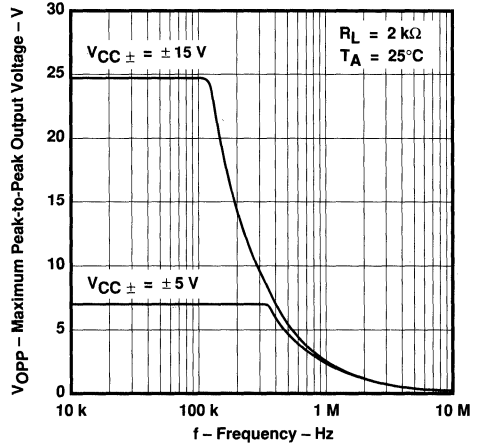


FIGURE 15

**MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 VS
 FREQUENCY**

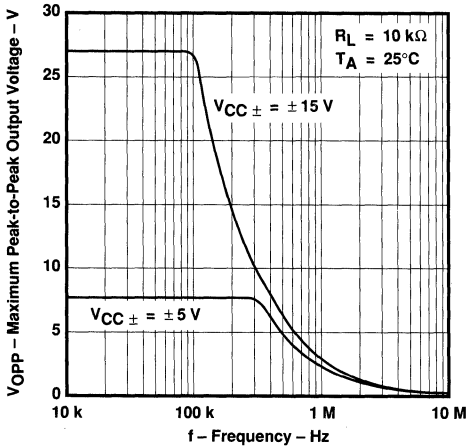


FIGURE 16

**MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 OUTPUT CURRENT**

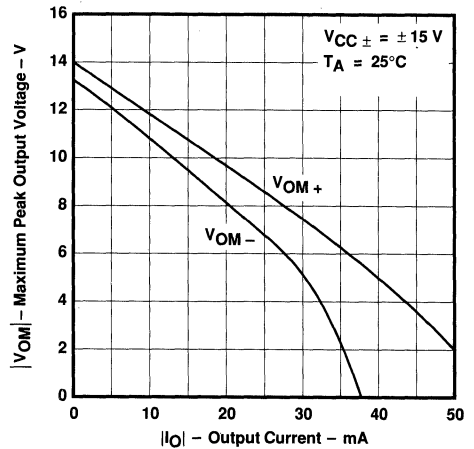


FIGURE 17

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
FREE-AIR TEMPERATURE

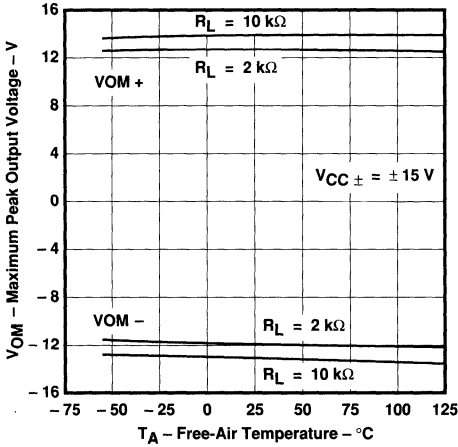


FIGURE 18

LARGE-SIGNAL VOLTAGE AMPLIFICATION
 VS
LOAD RESISTANCE

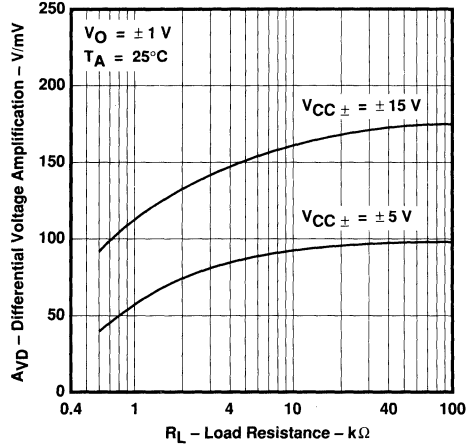


FIGURE 19

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
 VS
FREQUENCY

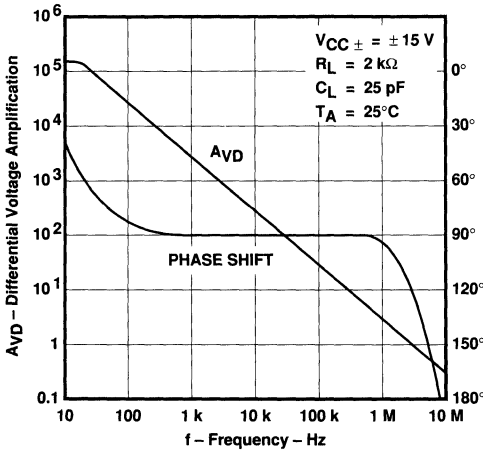


FIGURE 20

LARGE-SIGNAL VOLTAGE AMPLIFICATION
 VS
FREE-AIR TEMPERATURE

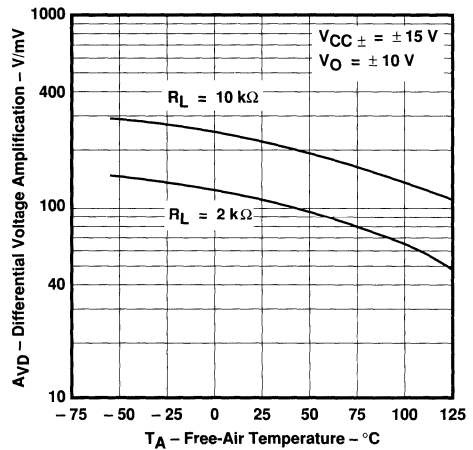


FIGURE 21

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

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Operational Amplifiers

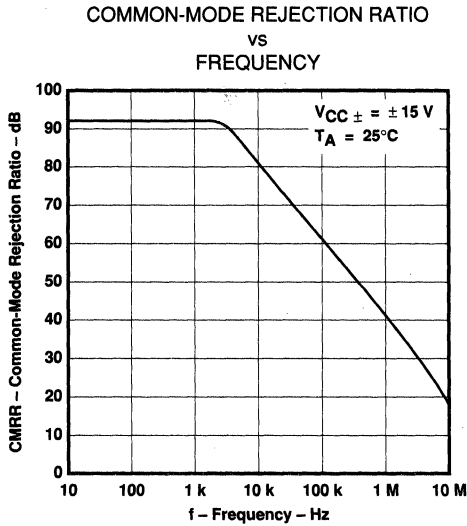


FIGURE 22

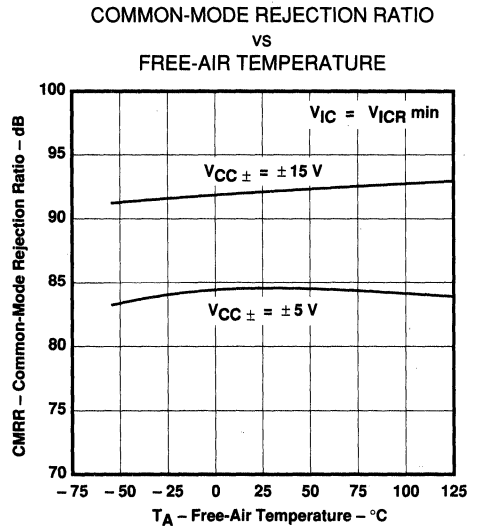


FIGURE 23

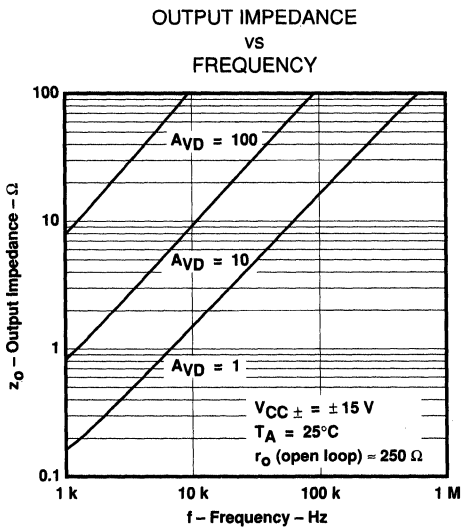


FIGURE 24

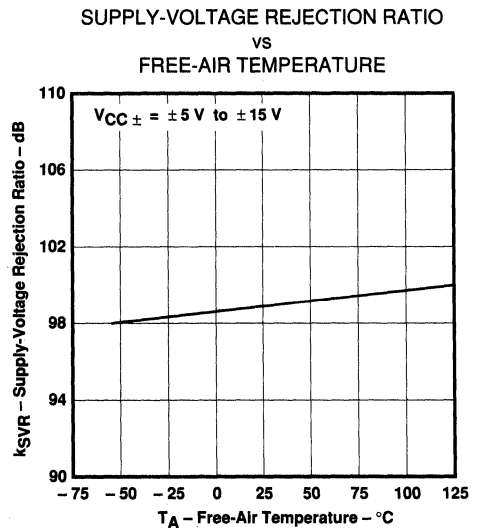


FIGURE 25

† Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

TL087, TL088, TL287, TL288
JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

SHORT-CIRCUIT OUTPUT CURRENT
 VS
 SUPPLY VOLTAGE

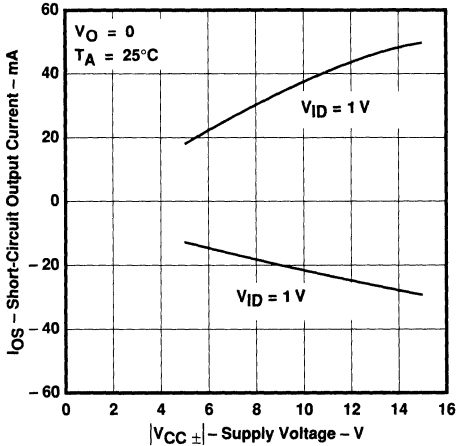


FIGURE 26

SHORT-CIRCUIT OUTPUT CURRENT
 VS
 TIME

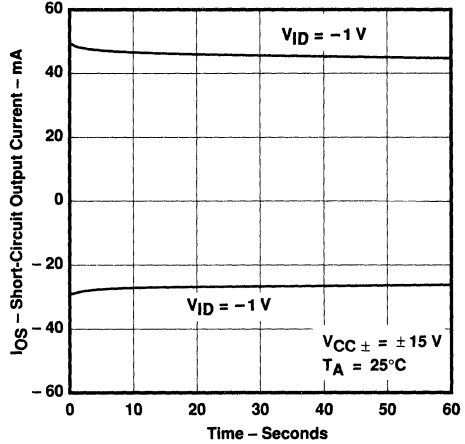


FIGURE 27

SHORT-CIRCUIT OUTPUT CURRENT
 VS
 FREE-AIR TEMPERATURE

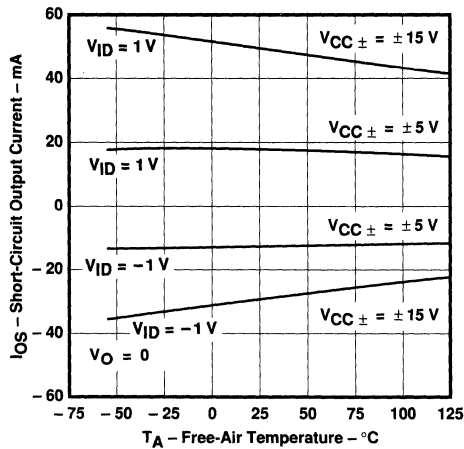


FIGURE 28

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

TL087, TL088, TL287, TL288
JFET-INPUT OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

2
Operational Amplifiers

SUPPLY CURRENT
 VS
 SUPPLY VOLTAGE

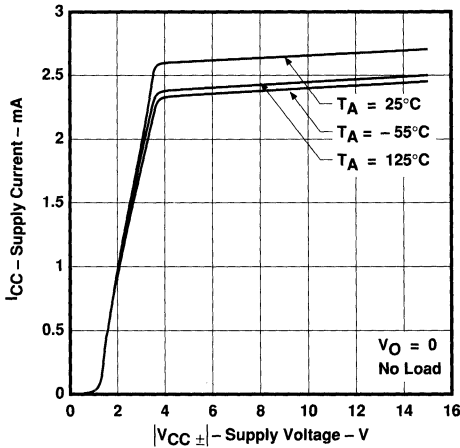


FIGURE 29

SUPPLY CURRENT
 VS
 FREE-AIR TEMPERATURE

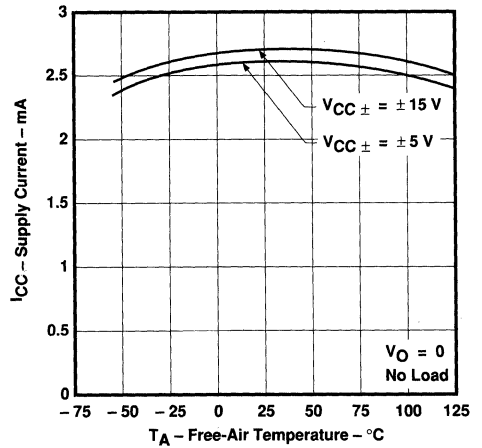


FIGURE 30

SLEW RATE
 VS
 LOAD RESISTANCE

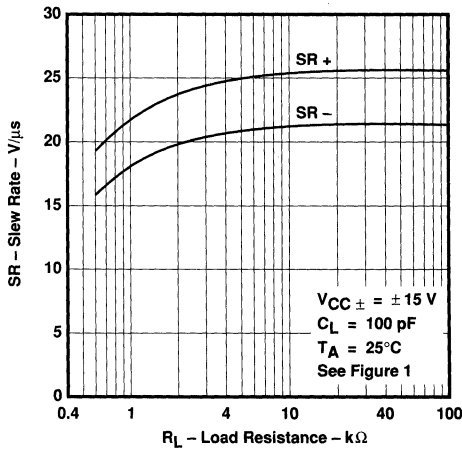


FIGURE 31

SLEW RATE
 VS
 FREE-AIR TEMPERATURE

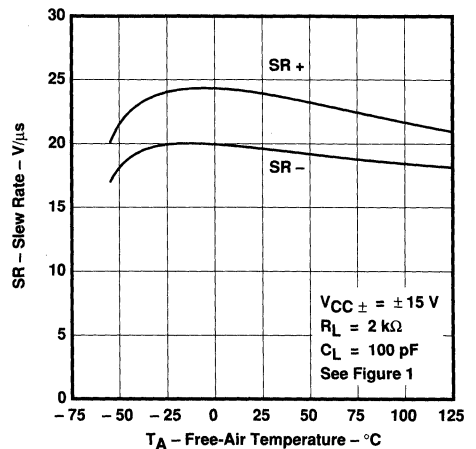


FIGURE 32

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

OVERSHOOT FACTOR
 VS
 LOAD CAPACITANCE

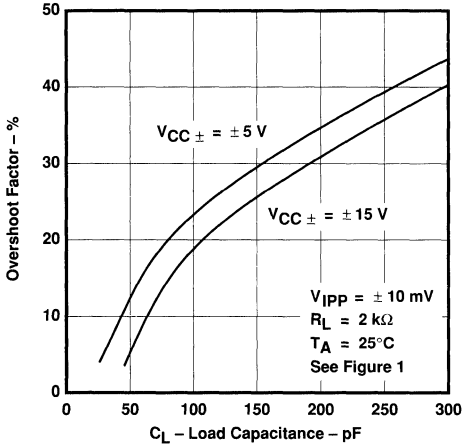


FIGURE 33

EQUIVALENT INPUT NOISE VOLTAGE
 VS
 FREQUENCY

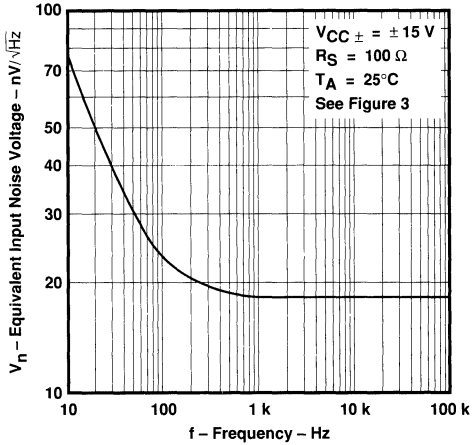


FIGURE 34

TOTAL HARMONIC DISTORTION
 VS
 FREQUENCY

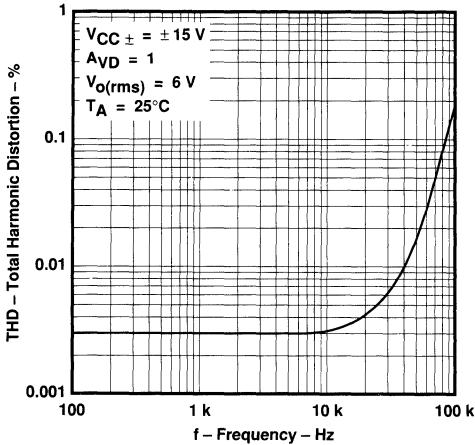


FIGURE 35

UNITY-GAIN BANDWIDTH
 VS
 SUPPLY VOLTAGE

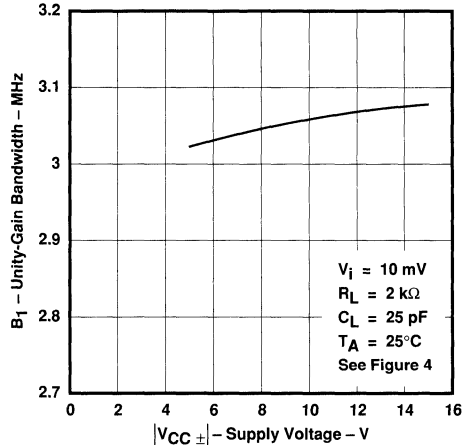


FIGURE 36

†Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

UNITY-GAIN BANDWIDTH
 vs
FREE-AIR TEMPERATURE

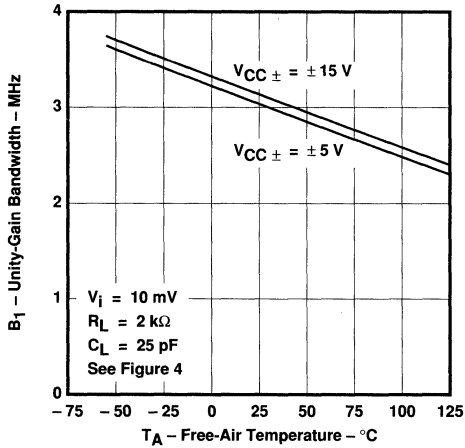


FIGURE 37

PHASE MARGIN
 vs
SUPPLY VOLTAGE

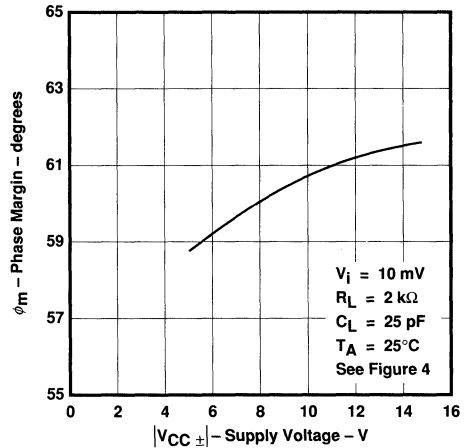


FIGURE 38

PHASE MARGIN
 vs
LOAD CAPACITANCE

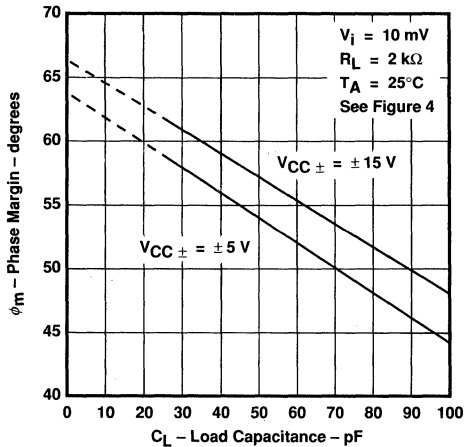


FIGURE 39

PHASE MARGIN
 vs
FREE-AIR TEMPERATURE

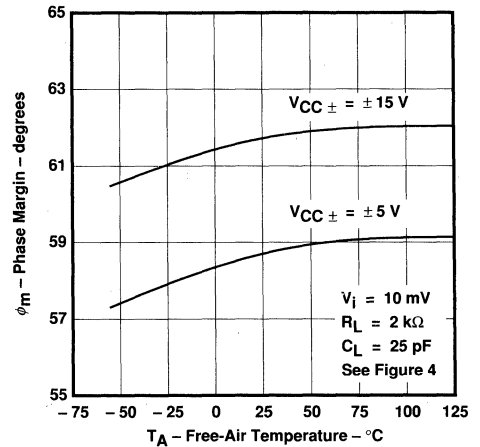


FIGURE 40

† Data at high and low temperatures are applicable within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

VOLTAGE-FOLLOWER
SMALL-SIGNAL
PULSE RESPONSE

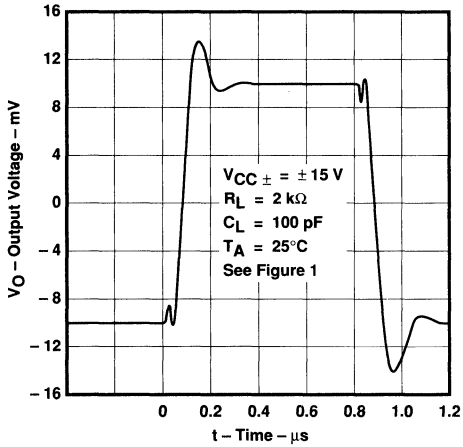


FIGURE 41

VOLTAGE-FOLLOWER
LARGE-SIGNAL
PULSE RESPONSE

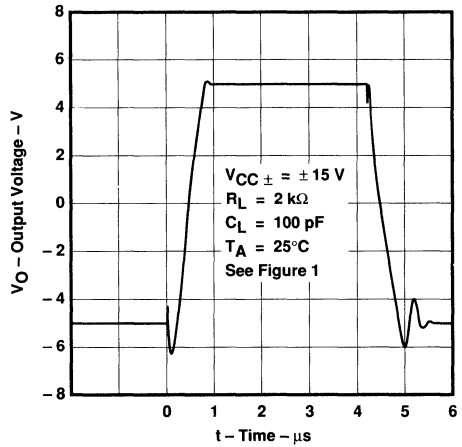


FIGURE 42

TYPICAL APPLICATION DATA

output characteristics

All operating characteristics are specified with 100-pF load capacitance. These amplifiers will drive higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem. Capacitive loads of 1000 pF and larger may be driven if enough resistance is added in series with the output (see Figure 43).

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Operational Amplifiers

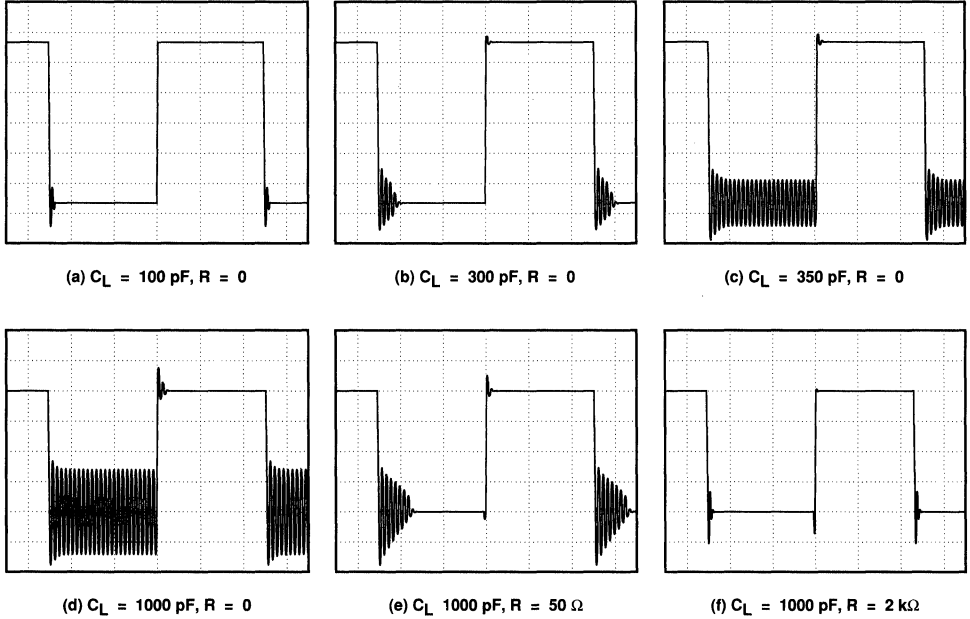
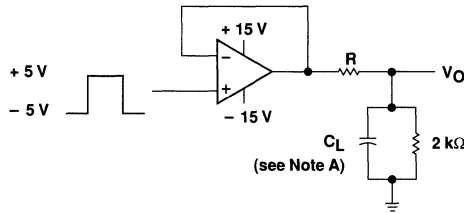


FIGURE 43. EFFECT OF CAPACITIVE LOADS



NOTE A: C_L includes fixture capacitance.

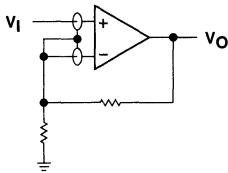
FIGURE 44. TEST CIRCUIT FOR OUTPUT CHARACTERISTICS

TYPICAL APPLICATION DATA

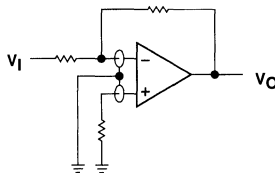
input characteristics

These amplifiers are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction.

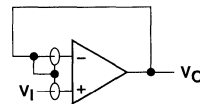
Because of the extremely high input impedance and resulting low bias current requirements, these amplifiers are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause degradation in system performance. It is good practice to include guard rings around inputs (see Figure 45). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.



(a) NONINVERTING AMPLIFIER



(b) INVERTING AMPLIFIER



(c) UNITY-GAIN AMPLIFIER

FIGURE 45. USE OF GUARD RINGS

noise performance

The noise specifications in op amp circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of these amplifiers result in a very low current noise. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than 50 k Ω .

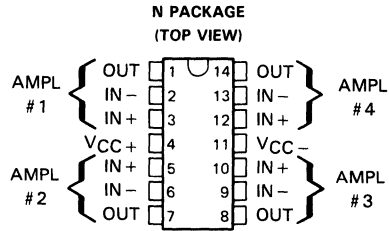
2

Operational Amplifiers

TL136C QUAD HIGH-PERFORMANCE OPERATIONAL AMPLIFIER

D2604, NOVEMBER 1981 – MAY 1988

- Continuous-Short Circuit Protection
- Wide Common-Mode and Differential Voltage Ranges
- No Frequency Compensation Required
- Low Power Consumption
- No Latch-Up
- Unity-Gain Bandwidth 3 MHz Typical
- Gain and Phase Match Between Amplifiers



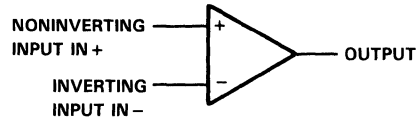
description

The TL136C is a quad high-performance operational amplifier with each amplifier electrically similar to the uA741 except that offset null capability is not provided.

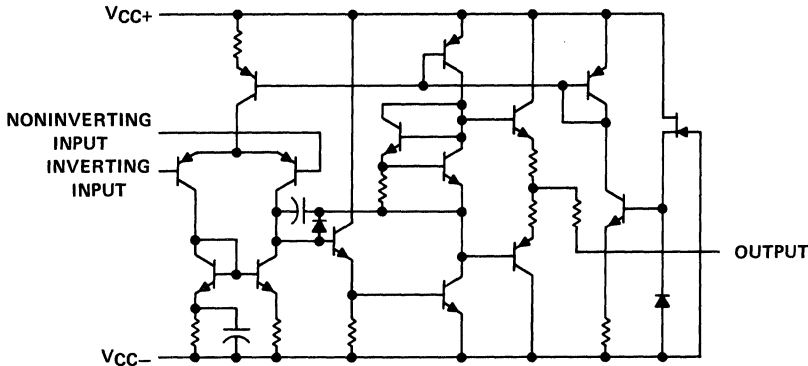
The high common-mode input voltage range and the absence of latch-up make these amplifiers ideal for voltage-follower applications. The devices are short-circuit protected and the internal frequency compensation ensures stability without external components.

The TL136C is characterized for operation from 0°C to 70°C.

symbol (each amplifier)



schematic (each amplifier)



2

Operational Amplifiers

TL136C
QUAD HIGH-PERFORMANCE OPERATIONAL AMPLIFIER

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage V_{CC+} (see Note 1)	18 V
Supply voltage V_{CC-} (see Note 1)	-18 V
Differential input voltage (see Note 2)	± 30 V
Input voltage (any input, see Notes 1 and 3)	± 15 V
Duration of output short-circuit to ground, one amplifier at a time (see Note 4)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 volts, whichever is less.
 4. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	$T_A = 70^\circ\text{C}$ POWER RATING
N	800 mW	9.2 mW/°C	736 mW

2
Operational Amplifiers

TL136C

QUAD HIGH-PERFORMANCE OPERATIONAL AMPLIFIER

electrical characteristics at specified free-air temperature, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$

PARAMETER		TEST CONDITIONS†		MIN	TYP	MAX	UNIT	
V_{IO}	Input offset voltage	$V_O = 0$,	$R_S = 50\ \Omega$	25°C	0.5	6	mV	
				0°C to 70°C		7.5		
I_{IO}	Input offset current	$V_O = 0$		25°C	5	200	nA	
				0°C to 70°C		300		
I_{IB}	Input bias current	$V_O = 0$		25°C	40	500	nA	
				0°C to 70°C		800		
V_{ICR}	Common-mode input voltage range			25°C	±12	±14	V	
V_{OPP}	Maximum peak-to-peak output voltage swing	$R_L = 10\text{ k}\Omega$		25°C	24	28	V	
				$R_L = 2\text{ k}\Omega$	25°C	20		26
				$R_L \geq 2\text{ k}\Omega$	0°C to 70°C	20		
A_{VD}	Large-signal differential voltage amplification	$R_L \geq 2\text{ k}\Omega$,	$V_O = \pm 10\text{ V}$	25°C	20	300	V/mV	
				0°C to 70°C	15			
B_1	Unity-gain bandwidth			25°C		3	MHz	
r_i	Input resistance			25°C	0.3	5	M Ω	
CMRR	Common-mode rejection ratio	$V_C = V_{ICR}\text{ min.}$		25°C	70	90	dB	
k_{SVS}	Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	$V_{CC\pm} = \pm 9\text{ V to } \pm 15\text{ V}$,		25°C		30	150	$\mu\text{V/V}$
V_n	Equivalent input noise voltage (closed-loop)	$A_{VD} = 100$,	$R_S = 100\ \Omega$,	25°C		7.5	nV/ $\sqrt{\text{Hz}}$	
I_{CC}	Supply current (All four amplifiers)	No load,	$V_O = 0\text{ V}$	25°C	5	11.3	mA	
				0°C	6	13.7		
				70°C	4.5	11.3		
P_D	Total power dissipation (All four amplifiers)	No load,	$V_O = 0\text{ V}$	25°C	150	340	mW	
				0°C	180	400		
				70°C	135	300		
$V_{O1}V_{O2}$	Crosstalk attenuation Open loop A_{VD}	$R_S = 1\text{ k}\Omega$,	$f = 10\text{ kHz}$	25°C		105	dB	
				25°C		105		

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified.

operating characteristics, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
t_r	Rise time	$V_I = 20\text{ mV}$,	$R_L = 2\text{ k}\Omega$,		0.13		μs
SR	Slew rate at unity gain	$V_I = 10\text{ V}$,	$R_L = 2\text{ k}\Omega$,		2.0		V/ μs

2

Operational Amplifiers

2

Operational Amplifiers

TL3211, TL321C OPERATIONAL AMPLIFIERS

D2343, APRIL 1977—REVISED OCTOBER 1988

- Wide Range of Supply Voltages Single Supply . . . 3 V to 30 V or Dual Supplies
- Low Supply Current Drain Independent of Supply Voltage . . . 0.8 mA Typ
- Common-Mode Input Voltage Range Includes Ground Allowing Direct Sensing near Ground
- Low Input Bias and Offset Parameters Input Input Offset Voltage . . . 2 mV Typ Input Offset Current . . . 3 nA Typ (TL3211) Input Bias Current . . . 45 nA Typ
- Differential Input Voltage Range Equal to Maximum-Rated Supply Voltage . . . ± 3 V
- Open-Loop Differential Voltage Amplification . . . 100 V/mV Typ
- Internal Frequency Compensation

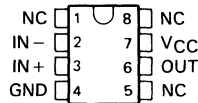
description

The TL321 is a high-gain, frequency-compensated operational amplifier that was designed specifically to operate from a single supply over a wide range of voltages. Operation from split supplies is also possible as long as the difference between the two supplies is 3 V to 30 V and pin 7 is at least 1.5 V more positive than the input common-mode voltage. The low supply current drain is independent of the magnitude of the supply voltage.

Applications include transducer amplifiers, d-c amplification blocks, and all the conventional operational amplifier circuits that now can be more easily implemented in single-supply-voltage systems. For example, the TL321 can be operated directly off of the standard 5-V supply that is used in digital systems and will easily provide the required interface electronics without requiring additional ± 15 -V supplies.

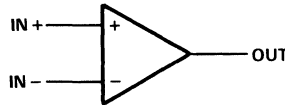
The TL3211 is characterized for operation from -25°C to 85°C . The TL321C is characterized for operation from 0°C to 70°C .

TL3211, TL321C . . . D, JG OR P PACKAGE
(TOP VIEW)



NC—No internal connection

symbol



AVAILABLE OPTIONS

T _A	V _{IO} MAX at 25°C	PACKAGE		
		SMALL- OUTLINE (D)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	7 mV	TL321CD	TL321CJG	TL321CP
-25°C to 85°C	5 mV	TL3211D	TL3211JG	TL3211P

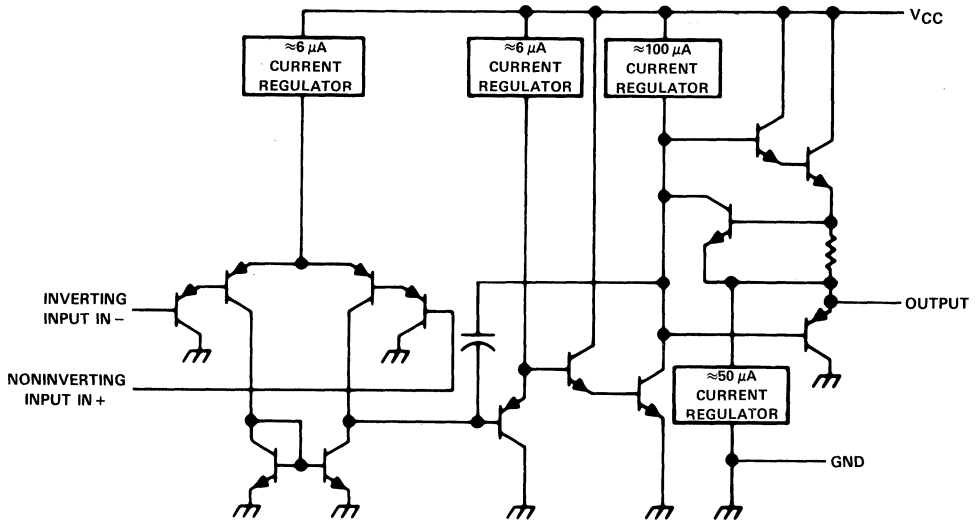
The D packages are available taped and reeled. Add the suffix R to the device type. (e.g., TL321CDR)

TL3211, TL321C OPERATIONAL AMPLIFIERS

schematic

2

Operational Amplifiers



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	32 V
Differential input voltage (see Note 2)	± 32 V
Input voltage range (either input)	-0.3 V to 32 V
Duration of output short-circuit to ground at (or below) 25°C free-air temperature ($V_{CC} \leq 15$ V) (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: TL3211	-25°C to 85°C
TL321C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the network ground terminal.
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. Short circuits from the output to V_{CC} can cause excessive heating and eventual destruction.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
D	680 mW	5.8 mW/°C	33°C	464 mW	377 mW
JG	680 mW	6.6 mW/°C	47°C	528 mW	429 mW
P	680 mW	8.0 mW/°C	65°C	640 mW	520 mW

TL321I, TL321C OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{CC} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TL321I			TL321C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = V_{ICR}\text{ min,}$ $V_{CC} = 5\text{ V to }30\text{ V,}$ $V_O = 1.4\text{ V,}$ $R_S = 50\text{ k}\Omega$	25°C	2	5	2	7	mV	
		Full range		7		9		
I_{IO} Input offset current	$V_O = 1.4\text{ V}$	25°C	3	30	5	50	nA	
		Full range		100		150		
I_{IB} Input bias current	$V_O = 1.4\text{ V}$	25°C	-45	-150	-45	-250	nA	
		Full range		-300		-500		
V_{ICR} Common-mode input voltage range	$V_{CC} = 5\text{ V to }30\text{ V}$	25°C	0 to $V_{CC}-1.5$		0 to $V_{CC}-1.5$		V	
		Full range	0 to $V_{CC}-2$		0 to $V_{CC}-2$			
V_{OH} High-level output voltage	$V_{CC} = 30\text{ V,}$ $R_L = 2\text{ k}\Omega$	Full range	26		26	V		
		Full range	27	28	27		28	
		25°C	3.5		3.5			
V_{OL} Low-level output voltage	$R_L \leq 10\text{ k}\Omega$	Full range	5	20	5	20	mV	
A_{VD} Large-signal differential voltage amplification	$V_{CC} = 15\text{ V,}$ $V_O = 1\text{ V to }11\text{ V,}$ $R_L \geq 2\text{ k}\Omega$	25°C	50	100	25	100	V/mV	
		Full range	25		15			
CMRR	$V_{IC} = V_{ICR}\text{ min,}$ $R_S = 50\ \Omega$	25°C	70	85	65	85	dB	
k_{SVR}	$V_{CC} = 5\text{ V to }30\text{ V,}$ $R_S = 50\ \Omega$	25°C	65	100	65	100	dB	
I_O Output current	Source	25°C	-25	-40	-20	-40	mA	
		Full range	-10	-20	-10	-20		
	Sink	25°C	10	20	10	20		
		Full range	5	8	5	8		
	$V_{ID} = -1\text{ V,}$ $V_O = 200\text{ mV}$	25°C	12	50	12	50	μA	
I_{CC} Supply current	No load $V_O = 15\text{ V,}$ $V_{CC} = 30\text{ V}$	Full range		2		2	mA	
	No load $V_O = 2.5\text{ V,}$ $V_{CC} = 5\text{ V}$	Full range		0.4	1	1		

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified. Full range is -25°C to 85°C for TL321I, and 0°C to 70°C for TL321C.

2

Operational Amplifiers

2

Operational Amplifiers

TL322I, TL322C DUAL LOW-POWER OPERATIONAL AMPLIFIERS

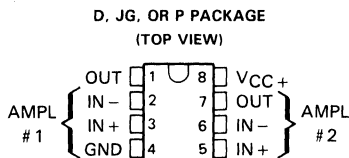
D2567, OCTOBER 1979—REVISED OCTOBER 1988

- Wide Range of Supply Voltages
Single Supply . . . 3 V to 36 V or
Dual Supplies
- Class AB Output Stage
- True Differential Input Stage
- Low Input Bias Current
- Internal Frequency Compensation
- Short-Circuit Protection

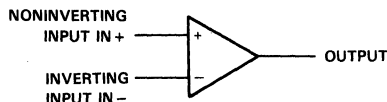
description

The TL322I and the TL322C are dual operational amplifiers similar in performance to the uA741 but with several distinct advantages. They are designed to operate from a single supply over a range of voltages from 3 V to 36 V. Operation from split supplies is also possible provided the difference between the two supplies is 3 V to 36 V. The common-mode input range includes the negative supply. Output range is from the negative supply to $V_{CC} - 1.5$ V. Quiescent supply currents per amplifier are typically less than one-half those of the uA741.

The TL322I is characterized for operation from -40°C to 85°C . The TL322C is characterized for operation from 0°C to 70°C .



symbol (each amplifier)

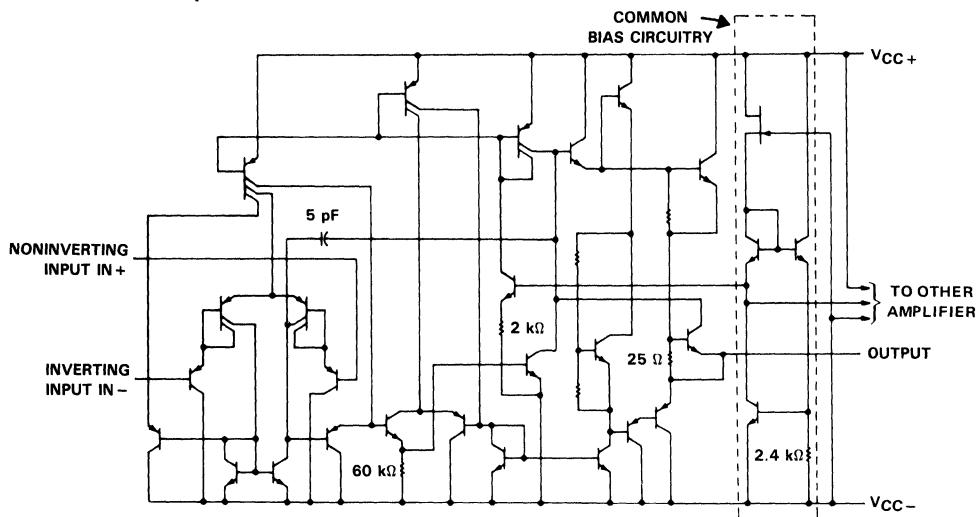


AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE		
		SMALL-OUTLINE (D)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	10 mV	TL322CD	TL322CJG	TL322CP
-40°C to 85°C	8 mV	TL322ID	TL322IJG	TL322IP

D packages are available taped and reeled. Add "R" suffix to device type, (e.g. TL322CDR)

schematic (each amplifier)



All component values shown are nominal

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS
INSTRUMENTS

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2-445

TL322I, TL322C DUAL LOW-POWER OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	TL322I	TL322C	UNIT
Supply voltage V_{CC+} (see Note 1)	18	18	V
Supply voltage V_{CC-} (see Note 1)	-18	-18	V
Supply voltage V_{CC+} with respect to V_{CC-}	36	36	V
Differential input voltage (see Note 2)	± 36	± 36	V
Input voltage (see Notes 1 and 3)	± 18	± 18	V
Continuous total power dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-40 to 85	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	JG package	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package	260	°C

- NOTES: 1. These voltage values are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. Neither input must ever be more positive than V_{CC+} or more negative than V_{CC-} .

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$
	POWER RATING			POWER RATING	POWER RATING
D	680 mW	5.8 mW/°C	33°C	464 mW	377 mW
JG	680 mW	6.6 mW/°C	47°C	528 mW	429 mW
P	680 mW	8.0 mW/°C	65°C	640 mW	520 mW

2

Operational Amplifiers

TL3221, TL322C DUAL LOW-POWER OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature; $V_{CC\pm} = 15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†		TL3221			TL322C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 0,$ $R_S = 50\ \Omega$	25°C	2		8	2		10	mV
		Full range	10			12			
α_{VIO} Temperature coefficient of input offset voltage	$V_O = 0,$ $R_S = 50\ \Omega$	25°C	10			10			$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current	$V_O = 0$	25°C	30		75	30		50	nA
		Full range	250			200			
α_{IIO} Temperature coefficient of input offset current	$V_O = 0$	25°C	50			50			$\text{pA}/^\circ\text{C}$
I_{IB} Input bias current	$V_O = 0$	25°C	-0.2	-0.5		-0.2	-0.5		μA
		Full range	-1			-0.8			
V_{ICR} Common-mode input voltage range‡		25°C	V_{CC-} to 13		V_{CC-} to 13.5	V_{CC-} to 13		V_{CC-} to 13.5	V
V_{OM} Peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	± 12		± 12.5	± 12		± 13.5	V
		25°C	± 10		± 12	± 10		± 13	
		Full range	± 10			± 10			
AVD Large-signal differential voltage amplification	$V_O = \pm 10\ \text{V},$ $R_L = 2\ \text{k}\Omega$	25°C	20	200		20	200		V/mV
		Full range	15			15			
B_{OM} Maximum-output-swing bandwidth	$V_{OPP} = 20\ \text{V},$ $AVD = 1,$ $THD \leq 5\%,$ $R_L = 2\ \text{k}\Omega$	25°C	9			9			kHz
B_1 Unity-gain bandwidth	$V_O = 50\ \text{mV},$ $R_L = 10\ \text{k}\Omega$	25°C	1			1			MHz
ϕ_m Phase margin	$R_L = 2\ \text{k}\Omega,$ $C_L = 200\ \text{pF}$	25°C	60°			60°			
r_i Input resistance	$f = 20\ \text{Hz}$	25°C	0.3	1		0.3	1		M Ω
r_o Output resistance	$f = 20\ \text{Hz}$	25°C	75			75			Ω
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\ \text{min}},$ $R_S = 50\ \Omega$	25°C	70	90		70	90		dB
k_{SVS} Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	$V_{CC} = \pm 2.5\ \text{V}$ to $\pm 15\ \text{V},$ $R_S = 50\ \Omega$	25°C	30		150	30		150	$\mu\text{V}/\text{V}$
I_{OS} Short-circuit output current [§]	$V_O = 0$	25°C	± 10	± 30	± 45	± 10	± 30	± 45	mA
I_{CC} Total supply current	$V_O = 0,$ No load	25°C	1.4		4	1.4		4	mA

† All characteristics are noted under open-loop conditions unless otherwise noted. Full range for T_A is -40°C to 85°C for TL3221, and 0°C to 70°C for TL322C.

‡ The V_{ICR} limits are directly linked volt-for-volt to supply voltage; the positive limit is 2 V less than V_{CC+} .

§ Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

2
Operational Amplifiers

TL322I, TL322C DUAL LOW-POWER OPERATIONAL AMPLIFIERS

2
Operational Amplifiers

electrical characteristics, $V_{CC+} = 5\text{ V}$, $V_{CC-} = 0\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS [†]	TL322I			TL322C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_O = 2.5\text{ V}$, $R_S = 50\ \Omega$			8		2	10	mV
I_{IO} Input offset current	$V_O = 2.5\text{ V}$			75		30	50	nA
I_{IB} Input bias current				-0.5		-0.2	-0.5	pA
V_{OM} Peak output voltage swing [‡]	$R_L = 10\text{ k}\Omega$ $R_L = 10\text{ k}\Omega$, $V_{CC+} = 5\text{ V to } 30\text{ V}$		3.3	3.5		3.3	3.5	V
A_{VD} Large-signal differential voltage amplification	$V_O = 1.7\text{ V to } 3.3\text{ V}$, $R_L = 2\text{ k}\Omega$		20	200		20	200	V/mV
k_{SVS} Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC\pm}$)	$V_{CC} = \pm 2.5\text{ V to } \pm 15\text{ V}$			150			150	$\mu\text{V/V}$
I_{CC} Supply current	$V_O = 2.5\text{ V}$, No load		1.2	4		1.2	4	mA
V_{O1}/V_{O2} Crosstalk attenuation	$A_{VD} = 100$, $f = 1\text{ kHz to } 20\text{ kHz}$		120			120		dB

[†]All characteristics are specified under open-loop conditions.

[‡]Output will swing essentially to ground.

switching characteristics: $V_{CC+} = \pm 15\text{ V}$, $A_{VD} = 1$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	$V_I = \pm 10\text{ V}$, $C_L = 100\text{ pF}$, See Figure 1		0.6		V/ μs
t_r Rise time	$\Delta V_O = 50\text{ mV}$, $C_L = 100\text{ pF}$, $R_L = 10\text{ k}\Omega$, See Figure 1		0.35		μs
t_f Fall time			0.35		μs
Overshoot factor			20%		
Crossover distortion	$V_{Ipp} = 30\text{ mV}$, $V_{Opp} = 2\text{ V}$, $f = 10\text{ kHz}$		1%		

PARAMETER MEASUREMENT INFORMATION

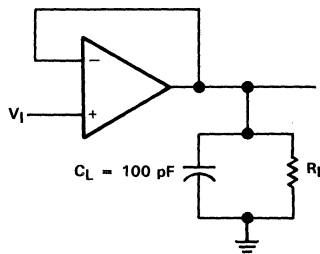


FIGURE 1. UNITY-GAIN AMPLIFIER

TYPICAL CHARACTERISTICS†

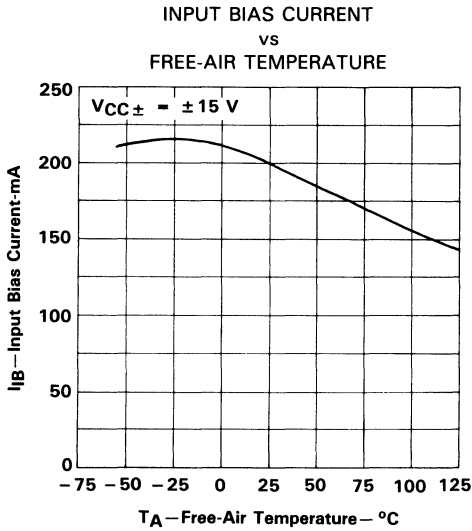


FIGURE 2

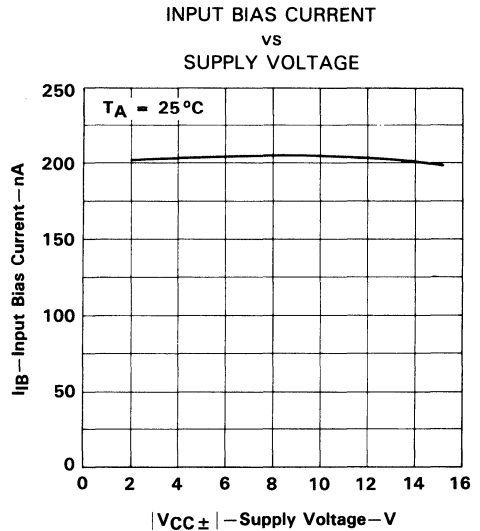


FIGURE 3

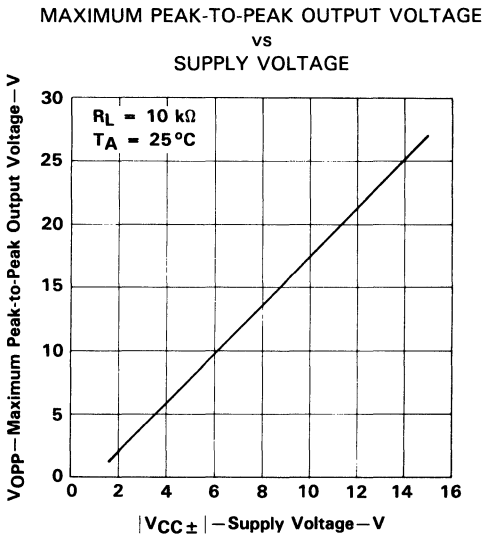


FIGURE 4

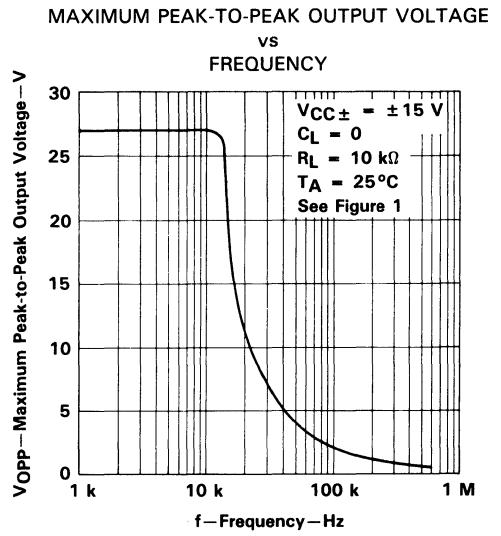


FIGURE 5

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL3221, TL322C
DUAL LOW-POWER OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

**LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREQUENCY**

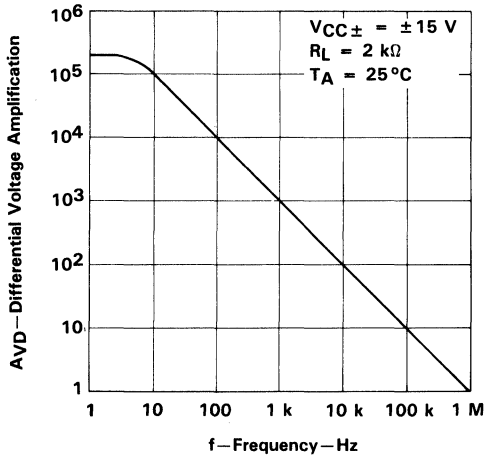


FIGURE 6

**VOLTAGE-FOLLOWER
LARGE-SIGNAL PULSE RESPONSE**

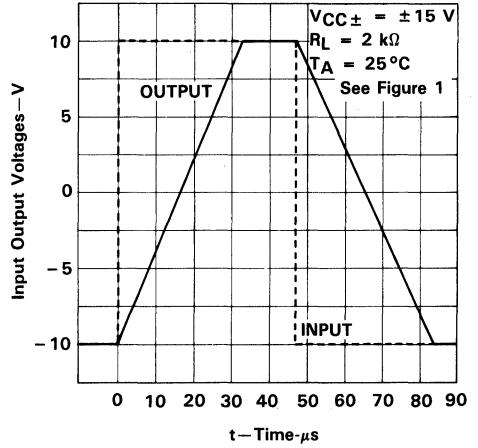


FIGURE 7

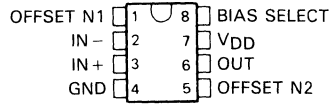
†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC251C, TLC251AC, TLC251BC PROGRAMMABLE LOW-POWER LinCMOS™ OPERATIONAL AMPLIFIERS

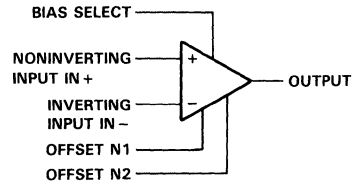
D2751, JULY 1983—REVISED SEPTEMBER 1988

- **Wide Range of Supply Voltages**
1.4 V to 16 V
- **True Single-Supply Operation**
- **Common-Mode Input Voltage Range**
Includes the Negative Rail
- **Low Noise . . . 30 nV $\sqrt{\text{Hz}}$ Typ at 1 kHz**
(High Bias)

D, JG, OR P PACKAGE
(TOP VIEW)



symbol



description

The TLC251C, TLC251AC, and TLC251BC are low-cost, low-power programmable operational amplifiers designed to operate with single or dual supplies. Unlike traditional metal-gate CMOS op amps, these devices utilize Texas Instruments silicon-gate LinCMOS™ process, giving them stable input offset voltages without sacrificing the advantages of metal-gate CMOS. This series of parts is available in selected grades of input offset voltage and can be nulled with one external potentiometer. Because the input common-mode range extends to the negative rail and the power consumption is extremely low, this family is ideally suited for battery-powered or energy-conserving applications. A bias-select pin can be used to program one of three ac performance and power-dissipation levels to suit the application. The series features operation down to a 1.4-V supply and is stable at unity gain.

The TLC251C series is characterized for operation from 0°C to 70°C.

These devices have internal electrostatic discharge (ESD) protection circuits that will prevent catastrophic failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.1. However, care should be exercised in handling these devices as exposure to ESD may result in a degradation of the device parametric performance.

Because of the extremely high input impedance and low input bias and offset currents, applications for the TLC251C series include many areas that have previously been limited to BIFET and NFET product types. Any circuit using high-impedance elements and requiring small offset errors is a good candidate for cost-effective use of these devices. Many features associated with bipolar technology are available with LinCMOS™ operational amplifiers without the power penalties of traditional bipolar devices. Remote and

AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE		
		SMALL-OUTLINE (D)	PLASTIC DIP (P)	CERAMIC DIP (JG)
0°C	10 mV	TLC251CD	TLC251CP	TLC251CJG
to	5 mV	TLC251ACD	TLC251ACP	TLC251ACJG
70°C	2 mV	TLC251BCD	TLC251BCP	TLC251BCJG

D packages are available taped-and-reeled. Add "R" suffix to device type when ordering (e.g., TLC251CDR).

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Operational Amplifiers

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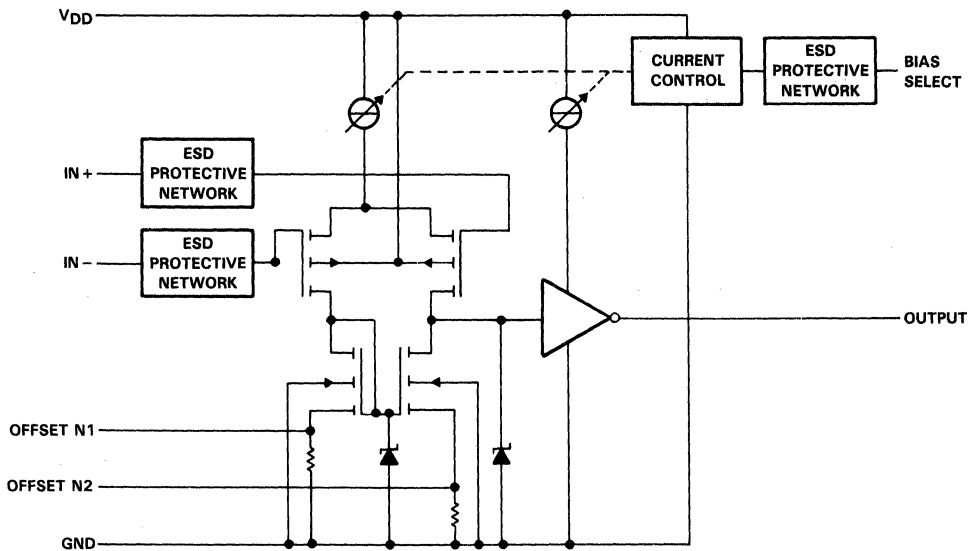
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TLC251C, TLC251AC, TLC251BC PROGRAMMABLE LOW-POWER LinCMOS™ OPERATIONAL AMPLIFIERS

description (continued)

inaccessible equipment applications are possible using the low-voltage and low-power capabilities of the TLC251C series. In addition, by driving the bias-select input with a logic signal from a microprocessor, these operational amplifiers can have software-controlled performance and power consumption. The TLC251C series is well suited to solve the difficult problems associated with single battery and solar cell-powered applications.

schematic



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Operational Amplifiers

TLC251C, TLC251AC, TLC251BC
PROGRAMMABLE LOW-POWER LinCMOS™ OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage, V_{DD} (see Note 1)	18 V
Differential input voltage (see Note 2)	± 18 V
Input voltage range (any input)	-0.3 V to 18 V
Duration of short-circuit at (or below) 25°C free-air temperature (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16) inch from the case for 60 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16) inch from case for 10 seconds: D or P package	260°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground terminal.
 2. Differential voltages are at the noninverting input terminal, with respect to the inverting input terminal.
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A = 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING
D	725 mW	5.8 mW/°C	464 mW
JG	825 mW	6.6 mW/°C	528 mW
P	1000 mW	8.0 mW/°C	640 mW

recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}		1.4		16	V
Common-mode input voltage, V_{IC}	$V_{DD} = 1.4$ V		0	0.2	V
	$V_{DD} = 5$ V		-0.2	4	
	$V_{DD} = 10$ V		-0.2	9	
	$V_{DD} = 16$ V		-0.2	14	
Operating free-air temperature, T_A			0	70	°C
Bias Select pin voltage		See Application Information			

TLC251C, TLC251AC, TLC251BC
PROGRAMMABLE LOW-POWER LinCMOS™ OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		BIAS	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC251C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	Any	25°C		10	mV
					0°C to 70°C		12	
				Any	25°C		5	
					0°C to 70°C		6.5	
				Any	25°C		2	
					0°C to 70°C		3	
α_{VIO}	Average temperature coefficient of input offset voltage		25°C to 70°C	Low		0.7	$\mu\text{V}/^\circ\text{C}$	
				Medium		2		
				High		5		
I_{IO}	Input offset current	$V_{IC} = 5\text{ V}$, $V_O = 5\text{ V}$	25°C	Any		1	pA	
					0°C to 70°C			300
I_{IB}	Input bias current	$V_{IC} = 5\text{ V}$, $V_O = 5\text{ V}$	25°C	Any		1	pA	
					0°C to 70°C			600
V_{ICR}	Common-mode input voltage range		25°C	Any	-0.2		9	V
V_{OM}	Peak output voltage range‡	$V_{ID} = 100\text{ mV}$	25°C	Any	8	8.6	V	
			0°C to 70°C		7.8			
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ to }6\text{ V}$, $R_S = 50\ \Omega$	25°C	Low	30	500	V/mV	
				Medium	20	280		
				High	10	40		
			0°C to 70°C	Low	25			
				Medium	15			
				High	7.5			
$CMRR$	Common-mode rejection ratio	$V_O = 1.4\text{ V}$, $V_{IC} = V_{ICR\text{ min}}$	25°C	Any	65	88	dB	
k_{SVR}	Supply voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IO}$)	$V_{DD} = 5\text{ to }10\text{ V}$, $V_O = 1.4\text{ V}$	25°C	Low	70	88	dB	
				Medium	70	88		
				High	65	82		
I_{OS}	Short-circuit output current	$V_O = 0$, $V_{ID} = 100\text{ mV}$	25°C	Any		-55	mA	
		$V_O = V_{DD}$, $V_{ID} = -100\text{ mV}$				15		
$I_{IH(SEL)}$	High-level input current to bias select	$V_{I(SEL)} = 0\text{ V}$	25°C	High		10.5	μA	
$I_{IL(SEL)}$	Low-level input current to bias select	$V_{I(SEL)} = 10\text{ V}$	25°C	Low		1.3	μA	
I_{DD}	Supply current	No load, $V_O = 5\text{ V}$, $V_{IC} = 5\text{ V}$	25°C	Low	10	23	μA	
				Medium	150	300		
				High	1000	2000		
			0°C to 70°C	Low		33		
				Medium		400		
				High		2200		

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Unless otherwise noted, an output load resistor is connected from the output to ground and has the following values: for low bias $R_L = 1\text{ M}\Omega$, for medium bias $R_L = 100\text{ k}\Omega$, and for high bias $R_L = 10\text{ k}\Omega$.

‡ The output will swing to the potential of the ground pin.

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Operational Amplifiers

TLC251C, TLC251AC, TLC251BC

PROGRAMMABLE LOW-POWER LinCMOS™ OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 1.4\text{ V}$

PARAMETER		TEST CONDITIONS†		BIAS	MIN	TYP	MAX	UNIT	
V_{IO}	Input offset voltage	TLC251C	$V_O = 0.2\text{ V}$, $R_S = 50\ \Omega$	25°C	Any		10	mV	
				0°C to 70°C			12		
				TLC251AC	25°C	Any			5
					0°C to 70°C				6.5
				TLC251BC	25°C	Any			2
					0°C to 70°C				3
α_{VIO}	Average temperature coefficient of input offset voltage		25°C to 70°C	Any		1		$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current	$V_O = 0.2\text{ V}$	25°C	Any		1		pA	
			0°C to 70°C			300			
I_{IB}	Input bias current	$V_O = 0.2\text{ V}$	25°C	Any		1		pA	
			0°C to 70°C			600			
V_{ICR}	Common-mode input voltage range		25°C	Any	0 to 0.2			V	
V_{OM}	Peak output voltage swing‡	$V_{ID} = 100\text{ mV}$	25°C	Any	450	700		mV	
A_{VD}	Large-signal differential voltage amplification	$V_O = 100\text{ to }300\text{ mV}$, $R_S = 50\ \Omega$	25°C	Low		20		V/mV	
				High		10			
$CMRR$	Common-mode rejection ratio	$R_S = 50\ \Omega$, $V_O = 0.2\text{ V}$, $V_{IC} = V_{IC\text{ min}}$	25°C	Any	60	77		dB	
I_{DD}	Supply current	$V_O = 0.2\text{ V}$, No load	25°C	Low		5	17	μA	
				High		150	190		

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Unless otherwise noted, an output load resistor is connected from the output to ground and has the following values: for low bias $R_L = 1\text{ M}\Omega$, for medium bias $R_L = 100\text{ k}\Omega$, and for high bias $R_L = 10\text{ k}\Omega$.

‡ The output will swing to the potential of the ground pin.

operating characteristics, $V_{DD} = 1.4\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	BIAS	MIN	TYP	MAX	UNIT
B_1	Unity-gain bandwidth	$C_L = 100\text{ pF}$	Low		12	kHz
			High		75	
SR	Slew rate at unity gain	See Figure 1	Low		0.001	$\text{V}/\mu\text{s}$
			High		0.01	
Overshoot factor	See Figure 1	Low		35%		
		High		30%		

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Operational Amplifiers

TLC251C, TLC251AC, TLC251BC
PROGRAMMABLE LOW-POWER LinCMOS™ OPERATIONAL AMPLIFIERS

operating characteristics, $V_{DD} = 10\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	BIAS	MIN	TYP	MAX	UNIT
B_1 Unity-gain bandwidth	$A_V = 40\text{ dB}$, $C_L = 100\text{ pF}$, $R_S = 50\ \Omega$	Low		0.1		MHz
		Medium		0.7		
		High		2.3		
SR Slew rate at unity gain	See Figure 1	Low		0.04		$\text{V}/\mu\text{s}$
		Medium		0.6		
		High		4.5		
Overshoot factor	See Figure 1	Low		30%		
		Medium		35%		
		High		35%		
ϕ_m Phase margin at unity gain	$A_V = 40\text{ dB}$, $R_S = 100\ \Omega$, $C_L = 100\text{ pF}$	Low		43°		
		Medium		43°		
		High		50°		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, $R_S = 100\ \Omega$	Low		70		$\text{nV}/\sqrt{\text{Hz}}$
		Medium		38		
		High		30		

PARAMETER MEASUREMENT INFORMATION

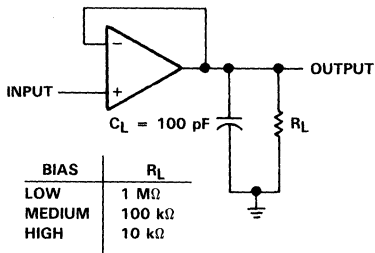


FIGURE 1. UNITY-GAIN AMPLIFIER

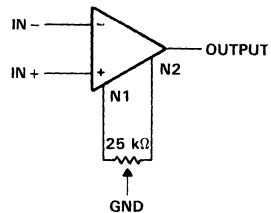


FIGURE 2. INPUT OFFSET VOLTAGE NULL CIRCUIT

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Operational Amplifiers

TLC251C, TLC251AC, TLC251BC
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TYPICAL CHARACTERISTICS

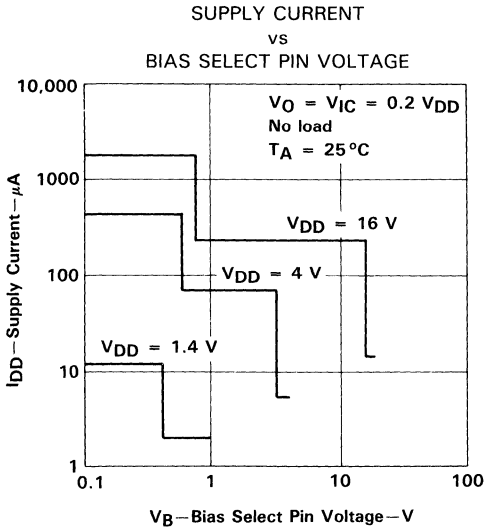


FIGURE 3

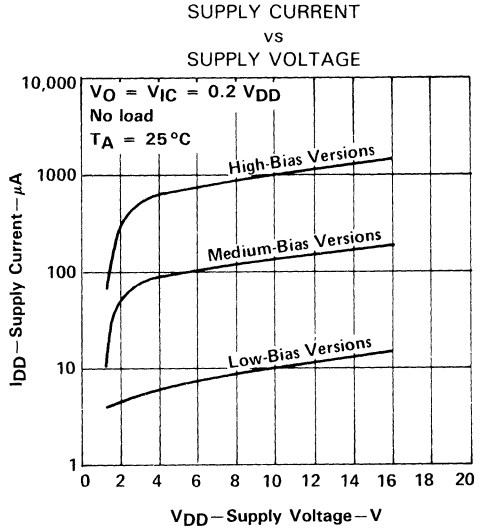


FIGURE 4

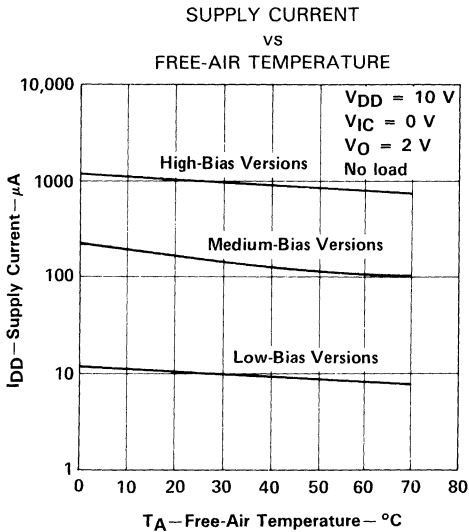


FIGURE 5

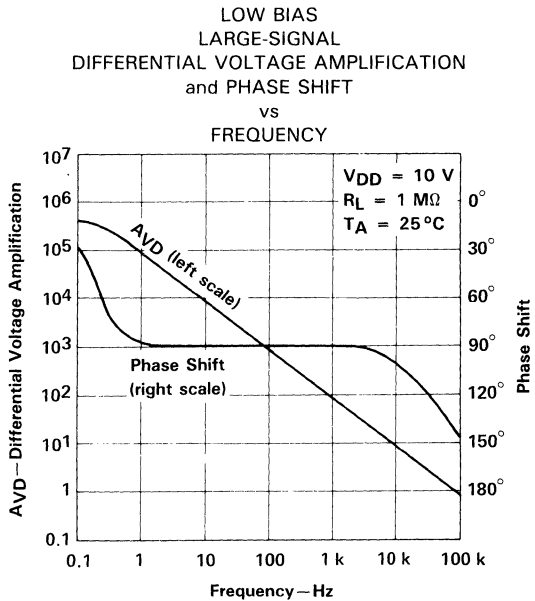


FIGURE 6

TYPICAL CHARACTERISTICS

MEDIUM BIAS
 LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 and PHASE SHIFT
 vs
 FREQUENCY

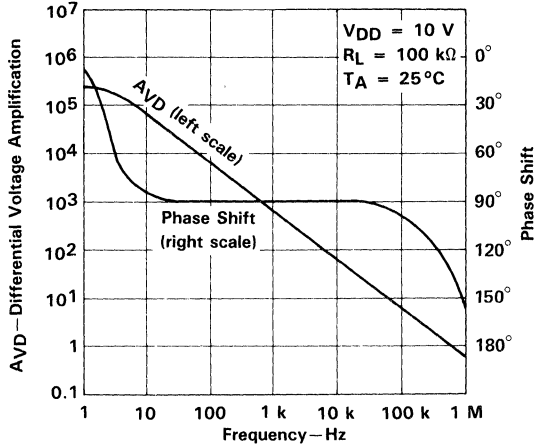


FIGURE 7

HIGH BIAS
 LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 and PHASE SHIFT
 vs
 FREQUENCY

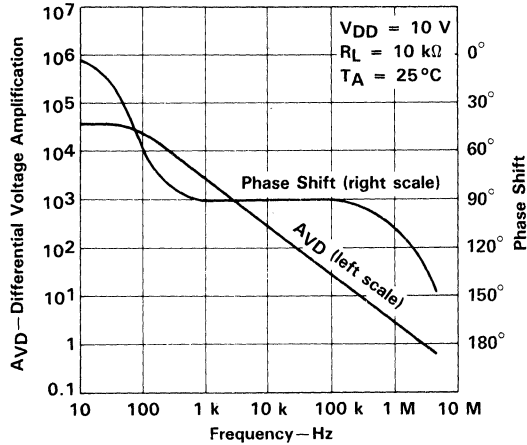


FIGURE 8

TYPICAL APPLICATION INFORMATION

latchup avoidance

Junction-isolated CMOS circuits have an inherent parasitic PNP structure that can function as an SCR. Under certain conditions, this SCR may be triggered into a low-impedance state, resulting in excessive supply current. To avoid such conditions, no voltage greater than 0.3 V beyond the supply rails should be applied to any pin. In general, the op amp supplies should be applied simultaneously with, or before, application of any input signals.

using the bias select pin

The TLC251C series has a bias select pin that allows the selection of one of three I_{DD} conditions (10, 150, and 1000 μA typical). This allows the user to trade-off power and ac performance. As shown in the typical supply current (I_{DD}) versus supply voltage (V_{DD}) curves (Figure 4), the I_{DD} varies only slightly from 4 V to 16 V. Below 4 V, the I_{DD} varies more significantly. Note that the I_{DD} values in the medium and low-bias modes at $V_{DD} = 1.4$ V are typically 2 μA , and in the high mode are typically 12 μA . The following table shows the recommended bias select pin connections at $V_{DD} = 10$ V:

BIAS MODE	AC PERFORMANCE	BIAS SELECT CONNECTION†	TYPICAL I_{DD} ‡
Low	Low	V_{DD}	10 μA
Medium	Medium	0.8 V to 9.2 V	150 μA
High	High	Ground pin	1000 μA

†The Bias Select pin may also be controlled by external circuitry to conserve power, etc. For information regarding the bias select pin, see Figure 3 in the typical characteristics curves.

‡For I_{DD} characteristics at voltages other than 10 V, see Figure 4 in the typical characteristics curves.

output stage considerations

The amplifier's output stage consists of a source-follower-connected pullup transistor and an open-drain pulldown transistor. The high-level output voltage (V_{OH}) is virtually independent of the I_{DD} selection, and increases with higher values of V_{DD} and reduced output loading. The low-level output voltage (V_{OL}) decreases with reduced output current and higher input common-mode voltage. With no load, V_{OL} is essentially equal to the GND pin potential.

input offset nulling

The TLC251C series offers external offset null control. Nulling may be achieved by adjusting a 25-k Ω potentiometer connected between the offset null terminals with the wiper connected to the device GND pin as shown in Figure 2. The amount of nulling range varies with the bias selection. At an I_{DD} setting of 1000 μA (high bias), the nulling range will allow the maximum offset specified to be trimmed to zero. In low or medium bias or when the amplifier is used below 4 V, total nulling may not be possible for all units.

supply configurations

Even though the TLC251C series is characterized for single-supply operation, it can be used effectively in a split-supply configuration when the input common-mode voltage (V_{ICR}), output swing (V_{OL} and V_{OH}), and supply voltage limits are not exceeded.

circuit layout precautions

The user is cautioned that when ever extremely high circuit impedances are used, care must be exercised in layout, construction, board cleanliness, and supply filtering to avoid hum and noise pickup, as well as excessive dc leakages.

2

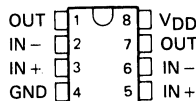
Operational Amplifiers

TLC252C, TLC25L2C, TLC25M2C LinCMOS™ DUAL OPERATIONAL AMPLIFIERS

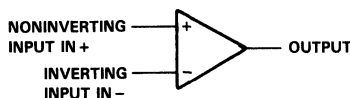
D2752, JUNE 1983—REVISED SEPTEMBER 1988

- A Suffix Versions Offer 5-mV V_{IO}
- B Suffix Versions Offer 2-mV V_{IO}
- Wide Range of Supply Voltages
1.4 V to 16 V
- True Single-Supply Operation
- Common-Mode Input Voltage Includes the Negative Rail
- Low Noise . . . 30 nV/√Hz Typ at
f = 1 kHz (High-Bias Versions)

D, JG, OR P PACKAGE
(TOP VIEW)



symbol (each amplifier)



description

The TLC252C, TLC25L2C, and TLC25M2C are low-cost, low-power dual operational amplifiers designed to operate with single or dual supplies. These devices utilize the Texas Instruments silicon gate LinCMOS™ process, giving them stable input offset voltages that are available in selected grades of 2, 5 or 10 mV maximum, very high input impedances, and extremely low input offset and bias currents. Because the input common-mode range extends to the negative rail and the power consumption is extremely low, this series is ideally suited for battery-powered or energy-conserving applications. The series offers operation down to a 1.4-V supply, is stable at unity gain, and has excellent noise characteristics.

The TLC252C series is characterized for operation from 0°C to 70°C

These devices have internal electrostatic discharge (ESD) protection circuits that will prevent catastrophic failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.1. However, care should be exercised in handling these devices as exposure to ESD may result in a degradation of the device parametric performance.

Because of the extremely high input impedance and low input bias and offset currents, applications for the TLC252C series include many areas that have previously been limited to BIFET and NFET product types. Any circuit using high-impedance elements and requiring small offset errors is a good candidate for cost-effective use of these devices. Many features associated with bipolar technology are available with LinCMOS™ operational amplifiers without the power penalties of traditional bipolar devices. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are all easily designed with the TLC252C series devices. Remote and inaccessible equipment

AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE		
		SMALL-OUTLINE (D)	PLASTIC DIP (P)	CERAMIC DIP (JG)
0°C to 70°C	10 mV	TLC252CD	TLC252CP	TLC252CJG
	5 mV	TLC252ACD	TLC252ACP	TLC252ACJG
	2 mV	TLC252BCD	TLC252BCP	TLC252BCJG
	10 mV	TLC25L2CD	TLC25L2CP	TLC25L2CJG
	5 mV	TLC25L2ACD	TLC25L2ACP	TLC25L2ACJG
	2 mV	TLC25L2BCD	TLC25L2BCP	TLC25L2BCJG
70°C	10 mV	TLC25M2CD	TLC25M2CP	TLC25M2CJG
	5 mV	TLC25M2ACD	TLC25M2ACP	TLC25M2ACJG
	2 mV	TLC25M2BCD	TLC25M2BCP	TLC25M2BCJG

D packages are available taped-and-reeled. Add "R" suffix to device type when ordering (e.g., TLC252CDR).

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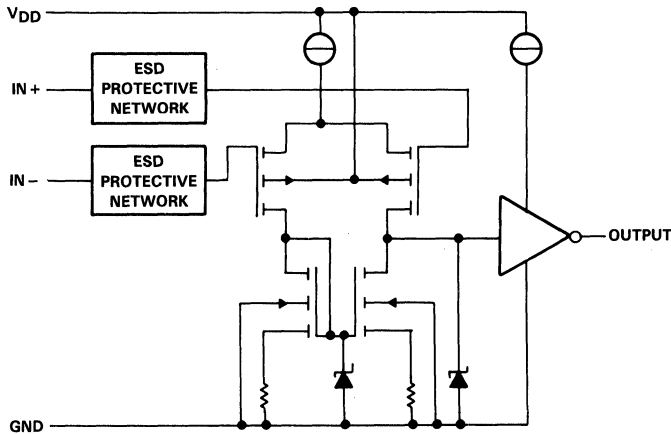
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TLC252C, TLC25L2C, TLC25M2C LinCMOS™ DUAL OPERATIONAL AMPLIFIERS

description (continued)

applications are possible using their low-voltage and low-power capabilities. The TLC252C series is well suited to solve the difficult problems associated with single-battery and solar-cell-powered applications. This series includes devices that are characterized for the commercial temperature range and are available in 8-pin plastic and ceramic dual-in-line (DIP) packages and the small outline (D) package.

schematic (each amplifier)



absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage, V_{DD} (see Note 1)	18 V
Differential input voltage (see Note 2)	± 18 V
Input voltage range (any input)	-0.3 V to 18 V
Duration of short-circuit at (or below) 25°C free-air temperature (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16) inch from the case for 60 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16) inch from case for 10 seconds: D or P package	260°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground terminal.
 2. Differential voltages are at the noninverting input terminal, with respect to the inverting input terminal.
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$
	POWER RATING		POWER RATING
D	725 mW	5.8 mW/°C	464 mW
JG	825 mW	6.6 mW/°C	528 mW
P	1000 mW	8.0 mW/°C	640 mW

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Operational Amplifiers

TLC252C, TLC25L2C, TLC25M2C LinCMOS™ DUAL OPERATIONAL AMPLIFIERS

recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}		1.4		16	V
Common-mode input voltage, V_{IC}	$V_{DD} = 1.4$ V		0	0.2	V
	$V_{DD} = 5$ V		-0.2	4	
	$V_{DD} = 10$ V		-0.2	9	
	$V_{DD} = 16$ V		-0.2	14	
Operating free-air temperature, T_A		0		70	°C

electrical characteristics at specified free-air temperature, $V_{DD} = 10$ V (unless otherwise noted)

PARAMETER		TEST CONDITIONS†	TLC252_C			TLC25L2_C			TLC25M2_C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_O	Input offset voltage	TLC25_2C TLC25_2AC TLC25_2BC	$V_O = 1.4$ V, $R_S = 50$ Ω	25°C		10		10		10	mV	
				0°C to 70°C		12		12				
				25°C		5		5				
				0°C to 70°C		6.5		6.5				
α_{VIO}	Average temperature coefficient of input offset voltage		25°C to 70°C		5		0.7		2	μ V/°C		
				I_{IO}	Input offset current	$V_{IC} = 5$ V, $V_O = 5$ V	25°C		1		1	pA
				0°C to 70°C				300		300		
				I_{IB}	Input bias current	$V_{IC} = 5$ V, $V_O = 5$ V	25°C		1		1	pA
0°C to 70°C		600		600								
V_{ICR}	Common-mode input voltage range		25°C		-0.2 to 9		-0.2 to 9		-0.2 to 9	V		
V_{OM}	Peak output voltage swing‡	$V_{ID} = 100$ mV	25°C	8	8.6	8	8.6	8	8.6	V		
			0°C to 70°C	7.8		7.8		7.8				
A_{VD}	Large-signal differential voltage amplification	$V_O = 1$ to 6 V, $R_S = 50$ Ω	25°C	10	40	50	500	25	280	V/mV		
$CMRR$	Common-mode rejection ratio	$V_O = 1.4$ V, $V_{IC} = V_{ICR}$ min	25°C	65	88	65	88	65	88	dB		
k_{SVR}	Supply voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IO}$)	$V_{DD} = 5$ to 10 V, $V_O = 1.4$ V	25°C	65	82	70	88	70	88	dB		
I_{OS}	Short-circuit output current	$V_O = 0$, $V_{ID} = 100$ mV	25°C		-55		-55		-55	mA		
				$V_O = V_{DD}$, $V_{ID} = -100$ mV		15		15			15	
I_{DD}	Supply current	No load, $V_O = 5$ V, $V_{IC} = 5$ V	25°C	2000	4000	20	46	300	600	μ A		
			0°C to 70°C		4400		66		800			

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Unless otherwise noted, an output load resistor is connected from the output to ground and has the following value: For low bias $R_L = 1$ M Ω ; for medium bias $R_L = 100$ k Ω ; and for high bias $R_L = 10$ k Ω .

‡ The output will swing to the potential of the ground pin.

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Operational Amplifiers

TLC252C, TLC25L2C, TLC25M2C LinCMOS™ DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 1.4\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		TLC252_C			TLC25L2_C			TLC25M2_C			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO}	Input offset voltage	TLC25_2C	$V_O = 0.2\text{ V}$, $R_S = 50\ \Omega$	25°C		10		10		10		mV	
				0°C to 70°C		12		12					
		TLC25_2AC		25°C		5		5					
				0°C to 70°C		6.5		6.5					
		TLC25_2BC		25°C		2		2					
0°C to 70°C			3		3								
α_{VIO}	Average temperature coefficient of input offset voltage		25°C to 70°C		1		1		1		$\mu\text{V}/^\circ\text{C}$		
I_{IO}	Input offset current		$V_O = 0.2\text{ V}$	25°C		1		1		1	pA		
				0°C to 70°C		300		300					
I_{IB}	Input bias current		$V_O = 0.2\text{ V}$	25°C		1		1		1	pA		
				0°C to 70°C		600		600					
V_{ICR}	Common-mode input voltage range		25°C		0 to 0.2		0 to 0.2		0 to 0.2		V		
V_{OM}	Peak output voltage swing‡	$V_{ID} = 100\text{ mV}$	25°C	450	700	450	700	450	700		mV		
A_{VD}	Large-signal differential voltage amplification	$V_O = 100\text{ to }300\text{ mV}$, $R_S = 50\ \Omega$	25°C		10		20		20		V/mV		
CMRR	Common-mode rejection ratio	$V_O = 0.2\text{ V}$, $V_{IC} = V_{ICR}\text{ min}$	25°C	60	77	60	77	60	77		dB		
I_{DD}	Supply current	No load, $V_O = 0.2\text{ V}$	25°C		300 375		25 34		200 250		μA		

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Unless otherwise noted, an output load resistor is connected from the output to ground and has the following value: For low bias $R_L = 1\text{ M}\Omega$; for medium bias $R_L = 100\text{ k}\Omega$; and for high bias $R_L = 10\text{ k}\Omega$.

‡ The output will swing to the potential of the ground pin.

operating characteristics, $V_{DD} = 10\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS	TLC252_C			TLC25L2_C			TLC25M2_C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
B_1	Unity-gain bandwidth	$A_V = 40\text{ dB}$, $C_L = 100\text{ pF}$, $R_S = 50\ \Omega$		2.2			0.11			0.635		MHz
SR	Slew rate at unity gain	See Figure 1		4.6			0.05			0.56		$\text{V}/\mu\text{s}$
	Overshoot factor	See Figure 1		35%			30%			35%		
ϕ_m	Phase margin at unity gain	$A_V = 40\text{ dB}$, $R_S = 100\ \Omega$, $C_L = 100\text{ pF}$		49°			38°			43°		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, $R_S = 100\ \Omega$		25			68			32		$\text{nV}\sqrt{\text{Hz}}$
V_{o1}/V_{o2}	Cross talk attenuation	$A_V = 100$		120			120			120		dB

TLC252C, TLC25L2C, TLC25M2C LinCMOS™ DUAL OPERATIONAL AMPLIFIERS

operating characteristics, $V_{DD} = 1.4\text{ V}$, $T_A = 25^\circ\text{C}$

TEST CONDITIONS	TEST CONDITIONS	TLC252_C			TLC25L2_C			TLC25M2_C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
B_1 Unity-gain bandwidth	$A_V = 40\text{ dB}$, $C_L = 10\text{ pF}$, $R_S = 50\ \Omega$	75			12			12			kHz
SR Slow rate at unity gain	See Figure 1	0.01			0.001			0.001			$\text{V}/\mu\text{s}$
Overshoot factor	See Figure 1	30%			35%			35%			

PARAMETER MEASUREMENT INFORMATION

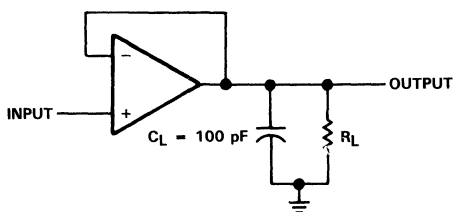


FIGURE 1. UNITY-GAIN AMPLIFIER

TYPICAL CHARACTERISTICS

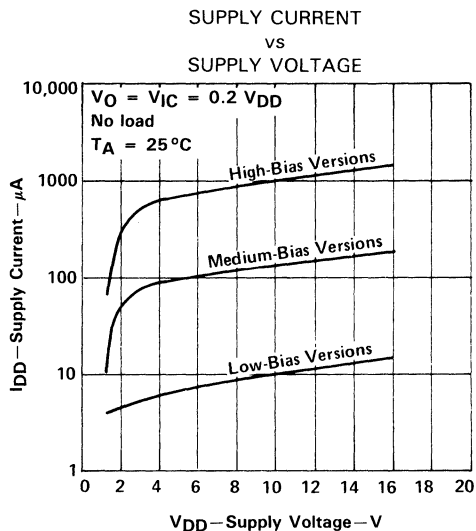


FIGURE 2

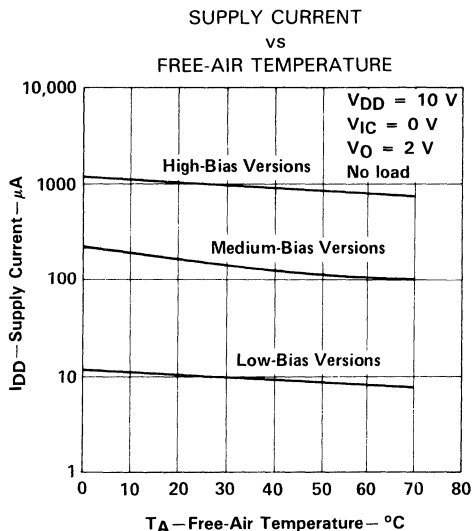


FIGURE 3

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Operational Amplifiers

TYPICAL CHARACTERISTICS

LOW-BIAS VERSIONS
 LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 and PHASE SHIFT
 vs FREQUENCY

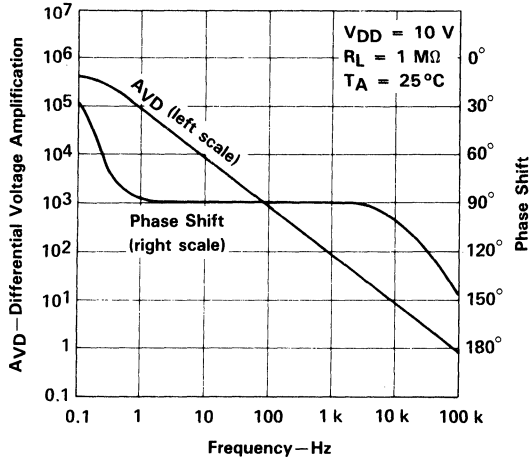


FIGURE 4

MEDIUM-BIAS VERSIONS
 LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 and PHASE SHIFT
 vs FREQUENCY

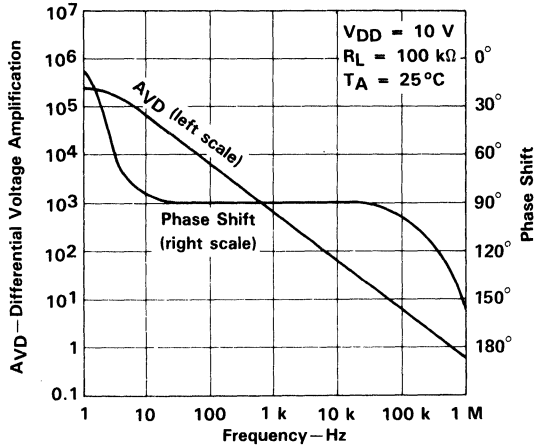


FIGURE 5

TYPICAL CHARACTERISTICS

HIGH-BIAS VERSIONS
 LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 and PHASE SHIFT
 vs FREQUENCY

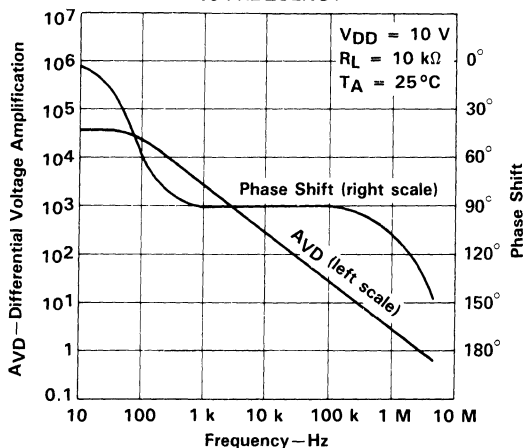


FIGURE 6

TYPICAL APPLICATION INFORMATION

latchup avoidance

Junction-isolated CMOS circuits have an inherent parasitic PNP structure that can function as an SCR. Under certain conditions, this SCR may be triggered into a low-impedance state, resulting in excessive supply current. To avoid such conditions, no voltage greater than 0.3 V beyond the supply rails should be applied to any pin. In general, the op amp supplies should be established simultaneously with, or before, application of any input signals.

output stage considerations

The amplifier's output stage consists of a source follower connected pullup transistor and an open drain pulldown transistor. The high-level output voltage (V_{OH}) is virtually independent of the I_{DD} selection, and increases with higher values of V_{DD} and reduced output loading. The low-level output voltage (V_{OL}) decreases with reduced output current and higher input common-mode voltage. With no load, V_{OL} is essentially equal to the GND pin potential.

supply configurations

Even though the TLC252C series is characterized for single-supply operation, it can be used effectively in a split supply configuration if the input common-mode voltage (V_{ICR}), output swing (V_{OL} and V_{OH}), and supply voltage limits are not exceeded.

circuit layout precautions

The user is cautioned that whenever extremely high circuit impedances are used, care must be exercised in layout, construction, board cleanliness, and supply filtering to avoid hum and noise pickup, as well as excessive DC leakages.

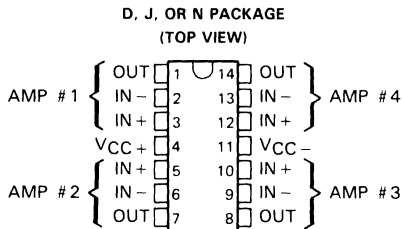
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Operational Amplifiers

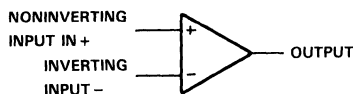
TLC254C, TLC25L4C, TLC25M4C LinCMOS™ QUAD OPERATIONAL AMPLIFIERS

D2753, JUNE 1983—REVISED OCTOBER 1988

- **A Suffix Versions Offer 5-mV V_{IO}**
- **B Suffix Versions Offer 2-mV V_{IO}**
- **Wide Range of Supply Voltages
1.4 V to 16 V**
- **True Single-Supply Operation**
- **Common-Mode Input Voltage Includes the
Negative Rail**
- **Low Noise . . . 30 nV $\sqrt{\text{Hz}}$ Typ at
 $f = 1 \text{ kHz}$ (High-Bias Versions)**



symbol (each amplifier)



description

The TLC254C, TLC25L4C, and TLC25M4C are low-cost, low-power quad operational amplifiers designed to operate with single or dual supplies. These devices utilize the Texas Instruments

silicon gate LinCMOS™ process, giving them stable input offset voltages that are available in selected grades of 2, 5, or 10 mV maximum, very high input impedances, and extremely low input offset and bias currents. Because the input common-mode range extends to the negative rail and the power consumption is extremely low, this series is ideally suited for battery-powered or energy-conserving applications. The series offers operation down to a 1.4-V supply, is stable at unity gain, and has excellent noise characteristics.

The TLC254C series is characterized for operation from 0°C to 70°C

These devices have internal electrostatic discharge (ESD) protection circuits that will prevent catastrophic failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.1. However, care should be exercised in handling these devices as exposure to ESD may result in a degradation of the device parametric performance.

Because of the extremely high input impedance and low input bias and offset currents, applications for the TLC254C series include many areas that have previously been limited to BIFET and NFET product types. Any circuit using high-impedance elements and requiring small offset errors is a good candidate for cost-effective use of these devices. Many features associated with bipolar technology are available with LinCMOS™ operational amplifiers without the power penalties of traditional bipolar devices. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal

AVAILABLE OPTIONS

T_A	V_{IO} MAX AT 25°C	PACKAGE		
		SMALL-OUTLINE (D)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C	10 mV	TLC254CD	TLC254CJ	TLC254CN
	5 mV	TLC254ACD	TLC254ACJ	TLC254ACN
	2 mV	TLC254BCD	TLC254BCJ	TLC254BCN
to 70°C	10 mV	TLC25L4CD	TLC25L4CJ	TLC25L4CN
	5 mV	TLC25L4ACD	TLC25L4ACJ	TLC25L4ACN
	2 mV	TLC25L4BCD	TLC25L4BCJ	TLC25L4BCN
	10 mV	TLC25M4CD	TLC25M4CJ	TLC25M4CN
	5 mV	TLC25M4ACD	TLC25M4ACJ	TLC25M4ACN
	2 mV	TLC25M4BCD	TLC25M4BCJ	TLC25M4BCN

D packages are available taped-and-reeled. Add "R" suffix to device type when ordering (e.g., TLC254CDR).

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Operational Amplifiers

TLC254C, TLC25L4C, TLC25M4C

LinCMOS™ QUAD OPERATIONAL AMPLIFIERS

description (continued)

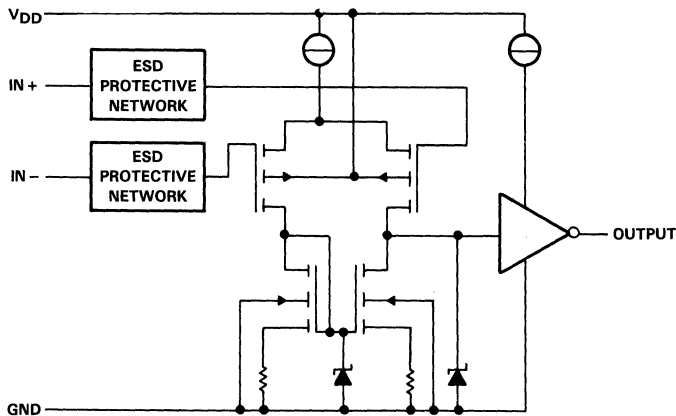
buffering are all easily designed with the TLC254C series devices. Remote and inaccessible equipment applications are possible using their low-voltage and low-power capabilities. The TLC254C series is well suited to solve the difficult problems associated with single-battery and solar-cell-powered applications. This series includes devices that are characterized for the commercial temperature range and are available in 14-pin plastic and ceramic dual-in-line (DIP) packages and the small outline (D) package.

DEVICE FEATURES

PARAMETER	TLC25L4C (LOW BIAS)	TLC25M4C (MEDIUM BIAS)	TLC254C (HIGH BIAS)
Supply current (Typ)	40 μA	600 μA	4000 μA
Slew rate (Typ)	0.04 $\text{V}/\mu\text{s}$	0.6 $\text{V}/\mu\text{s}$	4.5 $\text{V}/\mu\text{s}$
Input offset voltage (Max)			
TLC254C, TLC25L4C, TLC25M4C	10 mV	10 mV	10 mV
TLC254AC, TLC25L4AC, TLC25M4AC	5 mV	5 mV	5 mV
TLC254BC, TLC25L4BC, TLC25M4BC	2 mV	2 mV	2 mV
Offset voltage drift (Typ)	0.1 $\mu\text{V}/\text{month}^\dagger$	0.1 $\mu\text{V}/\text{month}^\dagger$	0.1 $\mu\text{V}/\text{month}^\dagger$
Offset voltage temperature coefficient (Typ)	0.7 $\mu\text{V}/^\circ\text{C}$	2 $\mu\text{V}/^\circ\text{C}$	5 $\mu\text{V}/^\circ\text{C}$
Input bias current (Typ)	1 pA	1 pA	1 pA
Input offset current (Typ)	1 pA	1 pA	1 pA

[†] The long-term drift value applies after the first month.

schematic (each amplifier)



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Operational Amplifiers

TLC254C, TLC25L4C, TLC25M4C LinCMOS™ QUAD OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage, V_{DD} (see Note 1)	18 V
Differential input voltage (see Note 2)	± 18 V
Input voltage range (any input)	-0.3 V to 18 V
Duration of short-circuit at (or below) 25°C free-air temperature (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16) inch from the case for 60 seconds: J package	300°C
Lead temperature 1,6 mm (1/16) inch from case for 10 seconds: D or N package	260°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground terminal.
 2. Differential voltages are at the noninverting input terminal, with respect to the inverting input terminal.
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A = 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING
D	725 mW	5.8 mW/°C	464 mW
J	1025 mW	8.2 mW/°C	656 mW
N	1150 mW	9.2 mW/°C	736 mW

recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}		1.4		16	V
Common-mode input voltage, V_{IC}	$V_{DD} = 1$ V	0		0.2	V
	$V_{DD} = 5$ V	-0.2		4	
	$V_{DD} = 10$ V	-0.2		9	
	$V_{DD} = 16$ V	-0.2		14	
Operating free-air temperature, T_A		0		70	°C

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Operational Amplifiers

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†	TLC254_C			TLC25L4_C			TLC25M4_C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_O	Input offset voltage	TLC25_4C $V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	25°C	10		10		10		10		mV
			0°C to 70°C	12		12		12		12		
			25°C	5		5		5		5		
			0°C to 70°C	6.5		6.5		6.5		6.5		
			25°C	2		2		2		2		
			0°C to 70°C	3		3		3		3		
αV_{IO}	Average temperature coefficient of input offset voltage		25°C to 70°C	5		0.7		2		$\mu\text{V}/^\circ\text{C}$		
I_{IO}	Input offset current	$V_{IC} = 5\text{ V}$, $V_O = 5\text{ V}$	25°C	1		1		1		1		pA
			0°C to 70°C	300		300		300		300		
I_{IB}	Input bias current	$V_{IC} = 5\text{ V}$, $V_O = 5\text{ V}$	25°C	1		1		1		1		pA
			0°C to 70°C	600		600		600		600		
V_{ICR}	Common-mode input voltage range		25°C	-0.2 to 9		-0.2 to 9		-0.2 to 9		V		
V_{OM}	Peak output voltage swing‡	$V_{ID} = 100\text{ mV}$	25°C	8 8.6		8 8.6		8 8.6		8 8.6		V
			0°C to 70°C	7.8		7.8		7.8		7.8		
AVD	Large-signal differential voltage amplification	$V_O = 1\text{ to }6\text{ V}$, $R_S = 50\ \Omega$	25°C	10 40		50 500		25 280		25 280		V/mV
			0°C to 70°C	7.5		50		15		15		
CMRR	Common-mode rejection ratio	$V_O = 1.4\text{ V}$, $V_{IC} = V_{ICR\text{ min}}$	25°C	65 88		65 88		65 88		65 88		dB
kSVR	Supply voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IO}$)	$V_{DD} = 5\text{ to }10\text{ V}$, $V_O = 1.4\text{ V}$	25°C	65 82		70 88		70 88		70 88		dB
I_{OS}	Short-circuit output current	$V_O = 0$, $V_{ID} = 100\text{ mV}$	25°C	-55		-55		-55		-55		mA
		$V_O = V_{DD}$, $V_{ID} = -100\text{ mV}$		15		15		15		15		
I_{DD}	Supply current	No load, $V_O = 5\text{ V}$, $V_{IC} = 5\text{ V}$	25°C	4000 8000		40 92		600 1200		600 1200		μA
			0°C to 70°C	8800		132		1600		1600		

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Unless otherwise noted, an output load resistor is connected from the output to ground and has the following value: For low bias $R_L = 1\text{ M}\Omega$; for medium bias $R_L = 100\text{ k}\Omega$; and for high bias $R_L = 10\text{ k}\Omega$.

‡ The output will swing to the potential of the ground pin.

electrical characteristics at specified free-air temperature, $V_{DD} = 1.4\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		TLC254_C			TLC25L4_C			TLC25M4_C			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO}	Input offset voltage	TLC25...4C	$V_O = 0.2\text{ V},$ $R_S = 50\ \Omega$	25°C		10		10		10	mV		
				0°C to 70°C		12		12		12			
		TLC25...4AC		25°C		5		5		5			
				0°C to 70°C		6.5		6.5		6.5			
		TLC25...4BC		25°C		2		2		2			
				0°C to 70°C		3		3		3			
α_{VIO}	Average temperature coefficient of input offset voltage		25°C to 70°C		1		1		1	$\mu\text{V}/^\circ\text{C}$			
I_{IO}	Input offset current	$V_O = 0.2\text{ V}$	25°C		1		1		1	pA			
			0°C to 70°C		300		300		300				
I_{IB}	Input bias current	$V_O = 0.2\text{ V}$	25°C		1		1		1	pA			
			0°C to 70°C		600		600		600				
V_{ICR}	Common-mode input voltage range		25°C	0 to 0.2		0 to 0.2		0 to 0.2		V			
V_{OM}	Peak output voltage swing‡	$V_{ID} = 100\text{ mV}$	25°C	450	700	450	700	450	700	mV			
A_{VD}	Large-signal differential voltage amplification	$V_O = 100\text{ to }300\text{ mV},$ $R_S = 50\ \Omega$	25°C		10		20		20	V/mV			
CMRR	Common-mode rejection ratio	$V_O = 0.2\text{ V},$ $V_{IC} = V_{ICR}\text{ min}$	25°C	60	77	60	77	60	77	dB			
I_{DD}	Supply current	No load, $V_O = 0.2\text{ V}$	25°C	600	750	50	68	400	500	μA			

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Unless otherwise noted, an output load resistor is connected from the output to ground and has the following value: For low bias $R_L = 1\text{ M}\Omega$; for medium bias $R_L = 100\text{ k}\Omega$; and for high bias $R_L = 10\text{ k}\Omega$.

‡ The output will swing to the potential of the ground pin.

TLC254C, TLC25L4C, TLC25M4C
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS

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Operational Amplifiers

operating characteristics, $V_{DD} = 10\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TLC254_C			TLC25L4_C			TLC25M4_C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
B_1	Unity-gain bandwidth $A_V = 40\text{ dB}$, $C_L = 100\text{ pF}$, $R_S = 50\ \Omega$	2.2			0.11			0.635			MHz
SR	Slew rate at unity gain See Figure 1	4.6			0.05			0.56			$\text{V}/\mu\text{s}$
	Overshoot factor See Figure 1	30%			30%			35%			
ϕ_m	Phase margin at unity gain $A_V = 40\text{ dB}$, $R_S = 100\ \Omega$, $C_L = 100\text{ pF}$	49°			38°			43°			
V_n	Equivalent input noise voltage $f = 1\text{ kHz}$, $R_S = 100\ \Omega$	25			68			32			$\text{nV}\sqrt{\text{Hz}}$
V_{O1}/V_{O2}	Cross talk attenuation $A_V = 100$	120			120			120			dB

operating characteristics, $V_{DD} = 1.4\text{ V}$, $T_A = 25^\circ\text{C}$

TEST CONDITIONS	TEST CONDITIONS	TLC254_C			TLC25L4_C			TLC25M4_C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
B_1	Unity-gain bandwidth $A_V = 40\text{ dB}$, $C_L = 10\text{ pF}$, $R_S = 50\ \Omega$	75			12			12			kHz
SR	Slew rate at unity gain See Figure 1	0.01			0.001			0.001			$\text{V}/\mu\text{s}$
	Overshoot factor See Figure 1	30%			35%			35%			

PARAMETER MEASUREMENT INFORMATION

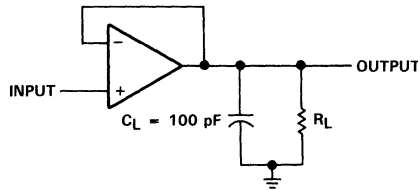


FIGURE 1. UNITY-GAIN AMPLIFIER

TLC254C, TLC25L4C, TLC25M4C
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

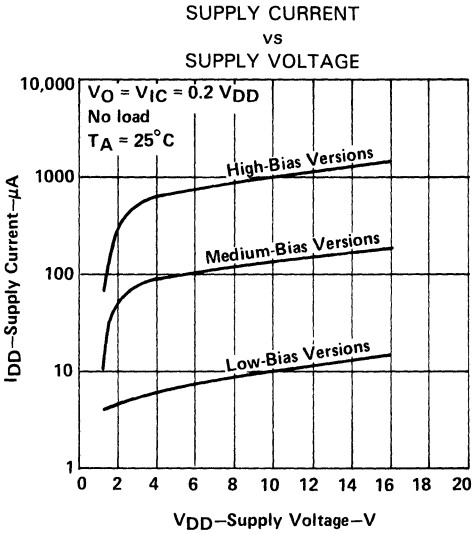


FIGURE 2

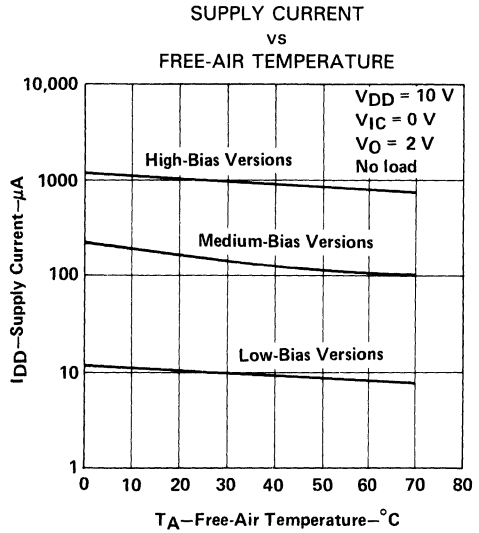


FIGURE 3

LOW-BIAS VERSIONS
 LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 and PHASE SHIFT
 vs
 FREQUENCY

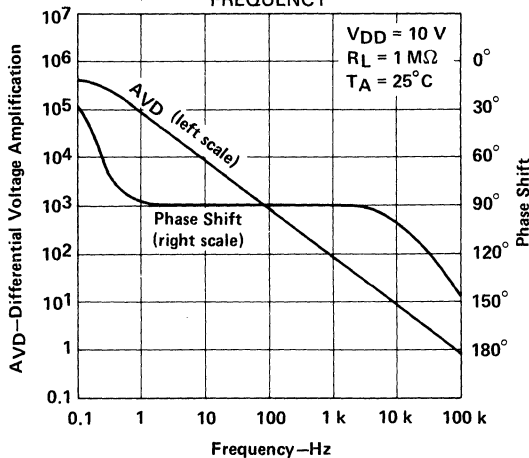


FIGURE 4

TYPICAL CHARACTERISTICS
 MEDIUM-BIAS VERSIONS
 LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 and PHASE SHIFT
 vs
 FREQUENCY

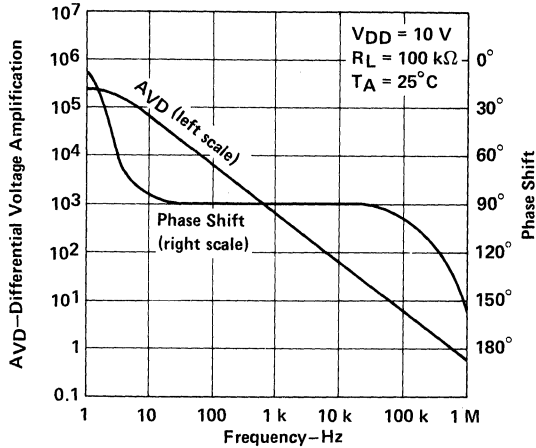


FIGURE 5

HIGH-BIAS VERSIONS
 LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 and PHASE SHIFT
 vs
 FREQUENCY

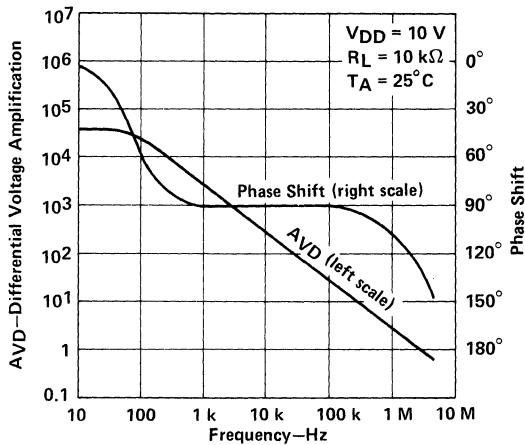


FIGURE 6

TYPICAL APPLICATION INFORMATION

latchup avoidance

Junction-isolated CMOS circuits have an inherent parasitic PNP structure that can function as an SCR. Under certain conditions, this SCR may be triggered into a low-impedance state, resulting in excessive supply current. To avoid such conditions, no voltage greater than 0.3 V beyond the supply rails should be applied to any pin. In general, the op amp supplies should be established simultaneously with, or before, application of any input signals.

output stage considerations

The amplifier's output stage consists of a source follower connected pullup transistor and an open drain pulldown transistor. The high-level output voltage (V_{OH}) is virtually independent of the I_{DD} selection, and increases with higher values of V_{DD} and reduced output loading. The low-level output voltage (V_{OL}) decreases with reduced output current and higher input common-mode voltage. With no load, V_{OL} is essentially equal to the GND pin potential.

supply configurations

Even though the TLC254C series is characterized for single-supply operation, it can be used effectively in a split supply configuration if the input common-mode voltage (V_{ICR}), output swing (V_{OL} and V_{OH}), and supply voltage limits are not exceeded.

circuit layout precautions

The user is cautioned that whenever extremely high circuit impedances are used, care must be exercised in layout, construction, board cleanliness, and supply filtering to avoid hum and noise pickup, as well as excessive DC leakages.

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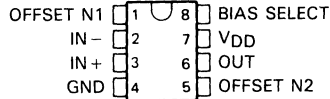
Operational Amplifiers

TLC271, TLC271A, TLC271B LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

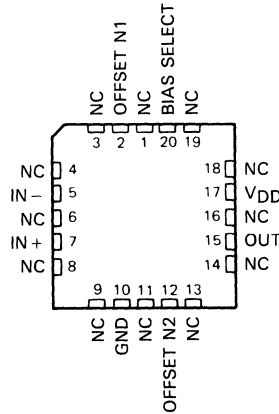
D3137, NOVEMBER 1987—REVISED MARCH 1989

- Input Offset Voltage Drift . . . Typically 0.1 $\mu\text{V}/\text{Month}$, Including the First 30 Days
- Wide Range of Supply Voltages over Specified Temperature Range:
–55°C to 125°C . . . 5 V to 16 V
–40°C to 85°C . . . 4 V to 16 V
0°C to 70°C . . . 3 V to 16 V
- Single-Supply Operation
- Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix types)
- Low Noise . . . Typically 25 nV $\sqrt{\text{Hz}}$ at $f = 1 \text{ kHz}$ (High-Bias Mode)
- Output Voltage Range Includes Negative Rail
- High Input Impedance . . . $10^{12} \Omega$ Typical
- ESD-Protection Circuitry
- Small-Outline Package Option Also Available in Tape and Reel
- Designed-In Latchup Immunity

D, JG, OR P PACKAGE
(TOP VIEW)



FK PACKAGE
(TOP VIEW)



NC—No internal connection

description

The TLC271 operational amplifier combines a wide range of input offset voltage grades with low offset voltage drift, and high input impedance. In addition, the TLC271 offers a bias select mode which allows the user to select the best combination of power dissipation and AC performance for a particular application. These devices use Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

AVAILABLE OPTIONS

T_A	V_{IOmax} at 25°C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	2 mV	TLC271BCD	—	TLC271BCJG	TLC271BCP
	5 mV	TLC271ACD	—	TLC271ACJG	TLC271ACP
	10 mV	TLC271CD	—	TLC271CJG	TLC271CP
–40°C to 85°C	2 mV	TLC271BID	—	TLC271BIJG	TLC271BIP
	5 mV	TLC271AID	—	TLC271AIJG	TLC271AIP
	10 mV	TLC271ID	—	TLC271IJG	TLC271IP
–55°C to 125°C	10 mV	—	TLC271MFK	TLC271MJG	—

	DEVICE FEATURES			UNIT
	TYPICAL at $V_{DD} = 5 \text{ V}$, $T_A = 25^\circ\text{C}$			
	BIAS-SELECT MODE			
	HIGH	MEDIUM	LOW	
P_D	3375	525	50	μW
SR	3.6	0.4	0.03	$\text{V}/\mu\text{s}$
V_n	25	32	68	$\text{nV}/\sqrt{\text{Hz}}$
B_1	1.7	0.5	0.09	MHz
A_{VD}	23	170	480	V/mV

The D package is available in tape and reel. Add R suffix to the device type (e.g., TLC271BCDR).

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Operational Amplifiers

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

description (continued)

Using the bias select option, these cost-effective devices can be “programmed” to span a wide range of applications which previously required BiFET, NFET or bipolar technology. Three offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC271 (10 mV) to the TLC271B (2 mV) low-offset version. The extremely high input impedance and low bias currents, in conjunction with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available in LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC271. The devices also exhibit low-voltage single-supply operation, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip carrier versions for high-density system applications.

The device inputs and output are designed to withstand –100-mA surge currents without sustaining latchup.

The TLC271 incorporates internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in degradation of device parametric performance.

The M-suffix devices are characterized for operation over the full military temperature range of –55 °C to 125 °C. The I-suffix devices are characterized for operation from –40 °C to 85 °C, and C-suffix devices are characterized for operation from 0 °C to 70 °C.

bias select feature

The TLC271 offers a bias select feature that allows the user to select any one of three bias levels, depending on the level of performance desired. The trade-offs between bias levels involve AC performance and power dissipation (see Table 1).

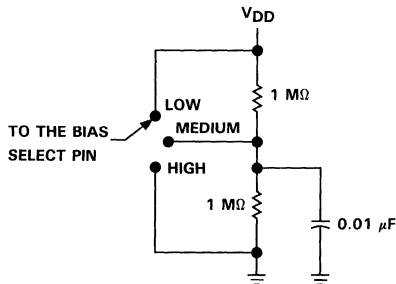
TABLE 1. EFFECT OF BIAS SELECTION ON PERFORMANCE

TYPICAL PARAMETER VALUES T _A = 25 °C, V _{DD} = 5 V		MODE			UNITS
		HIGH-BIAS R _L = 10 kΩ	MEDIUM-BIAS R _L = 100 kΩ	LOW-BIAS R _L = 1 MΩ	
P _D	Power dissipation	3.4	0.5	0.05	mW
SR	Slew rate	3.6	0.4	0.03	V/μs
V _n	Equivalent input noise voltage at f = 1 kHz	25	32	68	nV/√Hz
B ₁	Unity-gain bandwidth	1.7	0.5	0.09	MHz
φ _m	Phase margin	46°	40°	34°	
A _{VD}	Large-signal differential voltage amplification	23	170	480	V/mV

bias selection

Bias selection is achieved by connecting the bias select pin to one of three voltage levels (see Figure 1). For medium-bias applications, it is recommended that the bias select pin be connected to the mid-point between the supply rails. This procedure is simple in split-supply applications, since this point is ground. In single-supply applications, the medium-bias mode will necessitate using a voltage divider as indicated in Figure 1. The use of large-value resistors in the voltage divider will reduce the current drain of the divider from the supply line. However, large-value resistors used in conjunction with a large-value capacitor will require significant time to charge up to the supply midpoint after the supply is switched on. A voltage other than the mid-point may be used if it is within the voltages specified in the following table.

bias selection (continued)



BIAS MODE	BIAS SELECT VOLTAGE (Single Supply)
LOW	V_{DD}
MEDIUM	1 V to $V_{DD} - 1$ V
HIGH	GND

FIGURE 1. BIAS SELECTION FOR SINGLE-SUPPLY APPLICATIONS

high-bias mode

In the high-bias mode, the TLC271 series features low offset voltage drift, high input impedance, and low noise. Speed in this mode approaches that of BiFET devices, but at only a fraction of the power dissipation. Unity-gain bandwidth is typically greater than 1 MHz.

medium-bias mode

The TLC271 in the medium-bias mode features low offset voltage drift, high input impedance, and low noise. Speed in this mode is similar to general-purpose bipolar devices, but power dissipation is only a fraction of that consumed by bipolar devices.

low-bias mode

In the low-bias mode, the TLC271 features low offset voltage drift, high input impedance, extremely low power consumption, and high differential voltage gain.

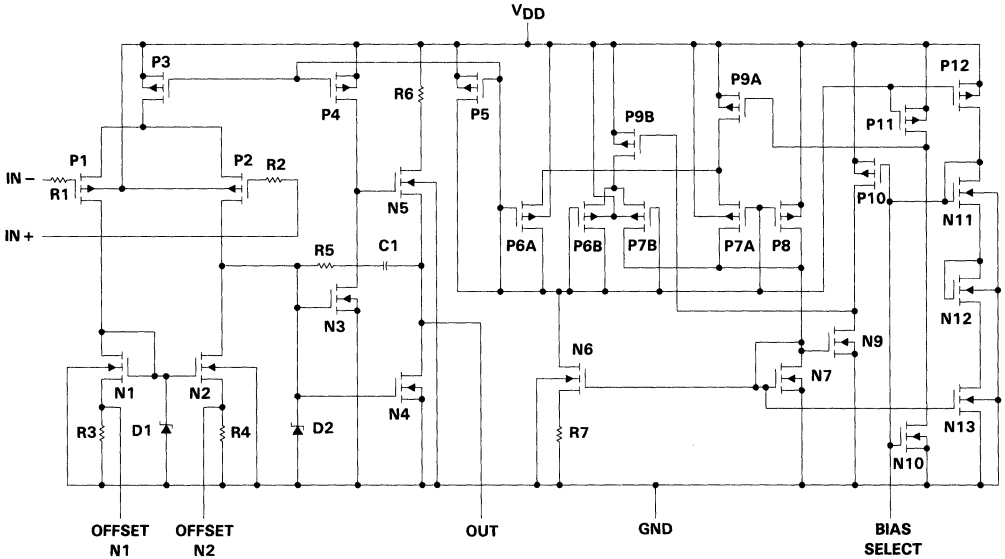
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electrical characteristics operating characteristics typical characteristics	medium (Figures 34–65)
electrical characteristics operating characteristics typical characteristics	low (Figures 66–97)
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TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

equivalent schematic

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Operational Amplifiers



TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{DD} (see Note 1)	18 V
Differential input voltage (see Note 2)	$\pm V_{DD}$
Input voltage range, V_I (any input)	-0.3 V to V_{DD}
Input current, I_I	± 5 mA
Output current, I_O	± 30 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	Unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, T_A : M-suffix	-55°C to 125°C
I-suffix	-40°C to 85°C
C-suffix	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D and P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.
 2. Differential voltages are at the noninverting input with respect to the inverting input.
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING		POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/ $^\circ\text{C}$	464 mW	377 mW	
FK	1375 mW	11 mW/ $^\circ\text{C}$	880 mW	715 mW	275 mW
JG (C-, I-suffix)	825 mW	6.6 mW/ $^\circ\text{C}$	528 mW	429 mW	
JG (M-suffix)	1050 mW	8.4 mW/ $^\circ\text{C}$	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/ $^\circ\text{C}$	640 mW	520 mW	

recommended operating conditions

		M-SUFFIX			I-SUFFIX			C-SUFFIX			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, V_{DD}		5		16	4		16	3		16	V
Common-mode input voltage, V_{IC}	$V_{DD} = 5$ V	0		3.5	-0.2		3.5	-0.2		3.5	V
	$V_{DD} = 10$ V	0		8.5	-0.2		8.5	-0.2		8.5	V
Operating free-air temperature, T_A		-55		125	-40		85	0		70	$^\circ\text{C}$

TLC271M
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

HIGH-BIAS MODE

electrical characteristics at specified free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T _A †	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 10 kΩ	25 °C		1.1	10		1.1	10	mV
		Full range			12			12	
αV _{IO} Average temperature coefficient of input offset voltage		25 °C to 125 °C		2.1			2.2		μV/°C
I _{IO} Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25 °C		0.1			0.1		pA
		125 °C		1.4	15		1.8	15	nA
I _{IB} Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25 °C		0.6			0.7		pA
		125 °C		9	35		10	35	nA
V _{ICR} Common-mode input voltage range (see Note 5)		25 °C	0	-0.3		0	-0.3		V
			to	to		to	to		
		Full range	4	4.2		9	9.2		V
			0			0			
V _{OH} High-level output voltage	V _{ID} = 100 mV, R _L = 10 kΩ	25 °C	3.2	3.8		8	8.5	V	
		-55 °C	3	3.8		7.8	8.5		
		125 °C	3	3.8		7.8	8.4		
V _{OL} Low-level output voltage	V _{ID} = -100 mV, I _{OL} = 0	25 °C		0	50		0	50	mV
		-55 °C		0	50		0	50	
		125 °C		0	50		0	50	
A _{VD} Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25 °C	5	23		10	36	V/mV	
		-55 °C	3.5	35		7	50		
		125 °C	3.5	16		7	27		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25 °C	65	80		65	85	dB	
		-55 °C	60	81		60	87		
		125 °C	60	84		60	86		
k _{SVR} Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V	25 °C	65	95		65	95	dB	
		-55 °C	60	90		60	90		
		125 °C	60	97		60	97		
I _{I(SEL)} Input current to bias select pin	V _{I(SEL)} = 0	25 °C		-1.4			-1.9	μA	
I _{DD} Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25 °C		675	1600		950	2000	μA
		-55 °C		1000	2500		1475	3000	
		125 °C		475	1100		625	1400	

† Full range is -55 °C to 125 °C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.
6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

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Operational Amplifiers

TLC271I, TLC271AI, TLC271BI
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

HIGH-BIAS MODE

electrical characteristics at specified free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A †	V _{DD} = 5 V			V _{DD} = 10 V			UNIT	
				MIN	TYP	MAX	MIN	TYP	MAX		
V _{IO}	Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 10 kΩ	25°C	1.1		10	1.1		10	mV	
			Full range	13			13				
			25°C	0.9		5	0.9		5		
			Full range	7			7				
			25°C	0.34		2	0.39		2		
Full range	3.5			3.5							
αV _{IO}	Average temperature coefficient of input offset voltage		25°C to 85°C	1.8			2			μV/°C	
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1			0.1			pA	
			85°C	24	1000		26	1000			
I _{IB}	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6			0.7			pA	
			85°C	200	2000		220	2000			
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	-0.2	-0.3		-0.2	-0.3		V	
				to	to		to	to			
			4	4.2			9	9.2			
			Full range	-0.2			-0.2			V	
				to			to				
				3.5			8.5				
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 10 kΩ	25°C	3.2	3.8		8	8.5		V	
			-40°C	3	3.8		7.8	8.5			
			85°C	3	3.8		7.8	8.5			
V _{OL}	Low-level output voltage	V _{ID} = -100 mV, I _{OL} = 0	25°C	0			50	0		50	mV
			-40°C	0			50	0		50	
			85°C	0			50	0		50	
A _{VD}	Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	5	23		10	36		V/mV	
			-40°C	3.5	32		7	46			
			85°C	3.5	19		7	31			
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65	80		65	85		dB	
			-40°C	60	81		60	87			
			85°C	60	86		60	88			
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V	25°C	65	95		65	95		dB	
			-40°C	60	92		60	92			
			85°C	60	96		60	96			
I _{I(SEL)}	Input current to bias select pin	V _{I(SEL)} = 0	25°C	-1.4			-1.9			μA	
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	675	1600		950	2000		μA	
			-40°C	950	2200		1375	2500			
			85°C	525	1200		725	1600			

†Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

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Operational Amplifiers

TLC271C, TLC271AC, TLC271BC
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

HIGH-BIAS MODE

electrical characteristics at specified free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A †	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	TLC271C V _O = 1.4 V; V _{IC} = 0 V, R _S = 50 Ω, R _L = 10 kΩ	25°C	1.1 10			1.1 10			mV
			Full range	12			12			
			25°C	0.9 5			0.9 5			
			Full range	6.5			6.5			
			25°C	0.34 2			0.39 2			
			Full range	3			3			
αV _{IO}	Average temperature coefficient of input offset voltage		25°C to 70°C	1.8			2			μV/°C
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1			0.1			pA
			70°C	7 300			7 300			
I _{IB}	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6			0.7			pA
			70°C	40 600			50 600			
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2	-0.2 to 9	-0.3 to 9.2			V
			Full range	-0.2 to 3.5		-0.2 to 8.5				V
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 10 kΩ	25°C	3.2 3.8		8 8.5				V
			0°C	3 3.8		7.8 8.5				
			70°C	3 3.8		7.8 8.4				
V _{OL}	Low-level output voltage	V _{ID} = -100 mV, I _{OL} = 0	25°C	0 50		0 50				mV
			0°C	0 50		0 50				
			70°C	0 50		0 50				
A _{VD}	Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	5 23		10 36				V/mV
			0°C	4 27		7.5 42				
			70°C	4 20		7.5 32				
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65 80		65 85				dB
			0°C	60 84		60 88				
			70°C	60 85		60 88				
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V	25°C	65 95		65 95				dB
			0°C	60 94		60 94				
			70°C	60 96		60 96				
I _{I(SEL)}	Input current to bias select pin	V _{I(SEL)} = 0	25°C	-1.4			-1.9			μA
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	675 1600		950 2000				μA
			0°C	775 1800		1125 2200				
			70°C	575 1300		750 1700				

† Full range is -0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

2

Operational Amplifiers

HIGH-BIAS MODE

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{Ipp} = 1\text{ V}$	25 °C		3.6		V/ μ s
				-55 °C		4.7		
				125 °C		2.3		
			$V_{Ipp} = 2.5\text{ V}$	25 °C		2.9		
				-55 °C		3.7		
				125 °C		2		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$,	25 °C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25 °C		320		kHz
				-55 °C		400		
				125 °C		230		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25 °C		1.7		MHz
				-55 °C		2.9		
				125 °C		1.1		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	25 °C		46 °		
				-55 °C		49 °		
				125 °C		41 °		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{Ipp} = 1\text{ V}$	25 °C		5.3		V/ μ s
				-55 °C		7.1		
				125 °C		3.1		
			$V_{Ipp} = 5.5\text{ V}$	25 °C		4.6		
				-55 °C		6.1		
				125 °C		2.7		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$,	25 °C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25 °C		200		kHz
				-55 °C		280		
				125 °C		110		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25 °C		2.2		MHz
				-55 °C		3.4		
				125 °C		1.6		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	25 °C		49 °		
				-55 °C		52 °		
				125 °C		44 °		

2

Operational Amplifiers

TLC271I, TLC271AI, TLC271BI
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

HIGH-BIAS MODE

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{IPP} = 1\text{ V}$	25°C		3.6		$V/\mu\text{s}$
				-40°C		4.5		
				85°C		2.8		
			$V_{IPP} = 2.5\text{ V}$	25°C		2.9		
				-40°C		3.5		
				85°C		2.3		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$,	25°C		25		$nV/\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C		320		kHz
				-40°C		380		
				85°C		250		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		1.7		MHz
				-40°C		2.6		
				85°C		1.2		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	25°C		46°		
				-40°C		49°		
				85°C		43°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{IPP} = 1\text{ V}$	25°C		5.3		$V/\mu\text{s}$
				-40°C		6.8		
				85°C		4		
			$V_{IPP} = 5.5\text{ V}$	25°C		4.6		
				-40°C		5.8		
				85°C		3.5		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$,	25°C		25		$nV/\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C		200		kHz
				-40°C		260		
				85°C		130		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		2.2		MHz
				-40°C		3.1		
				85°C		1.7		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	25°C		49°		
				-40°C		52°		
				85°C		46°		

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Operational Amplifiers

TLC271C, TLC271AC, TLC271BC
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

HIGH-BIAS MODE

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{IPP} = 1\text{ V}$	25°C		3.6		V/ μ s
				0°C		4		
				70°C		3		
			$V_{IPP} = 2.5\text{ V}$	25°C		2.9		
				0°C		3.1		
				70°C		2.5		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$	25°C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 98	25°C		320		kHz
				0°C		340		
				70°C		260		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		1.7		MHz
				0°C		2		
				70°C		1.3		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	25°C		46°		
				0°C		47°		
				70°C		44°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 99	$V_{IPP} = 1\text{ V}$	25°C		5.3		V/ μ s
				0°C		5.9		
				70°C		4.3		
			$V_{IPP} = 5.5\text{ V}$	25°C		4.6		
				0°C		5.1		
				70°C		3.8		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$	25°C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 98	25°C		200		kHz
				0°C		220		
				70°C		140		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		2.2		MHz
				0°C		2.5		
				70°C		1.8		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	25°C		49°		
				0°C		50°		
				70°C		46°		

2

Operational Amplifiers

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†

**DISTRIBUTION OF TLC271
 INPUT OFFSET VOLTAGE**

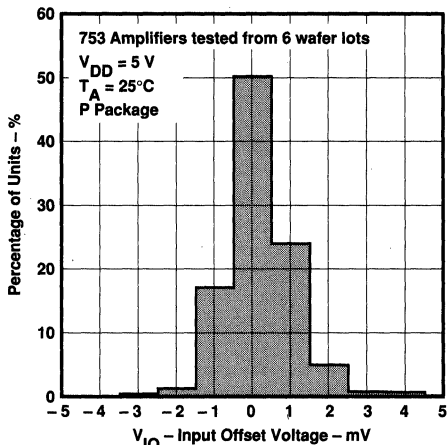


FIGURE 2

**DISTRIBUTION OF TLC271
 INPUT OFFSET VOLTAGE**

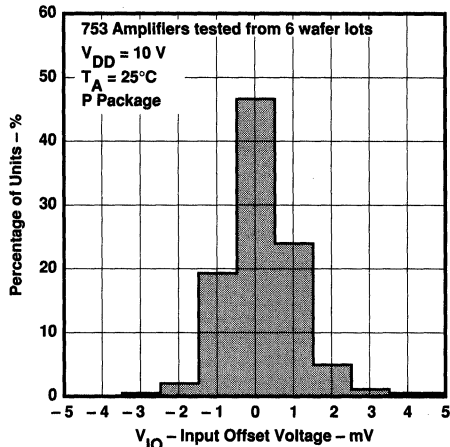


FIGURE 3

**DISTRIBUTION OF TLC271
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT**

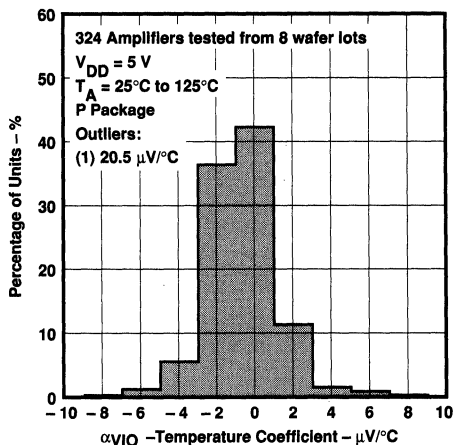


FIGURE 4

**DISTRIBUTION OF TLC271
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT**

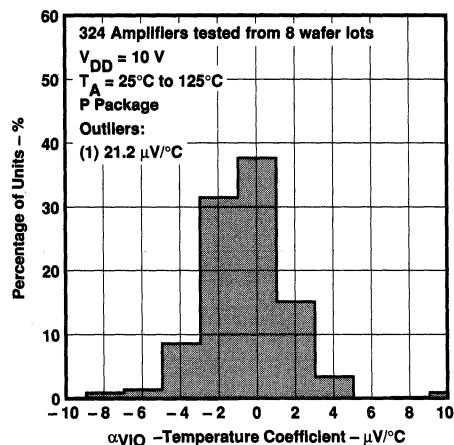


FIGURE 5

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

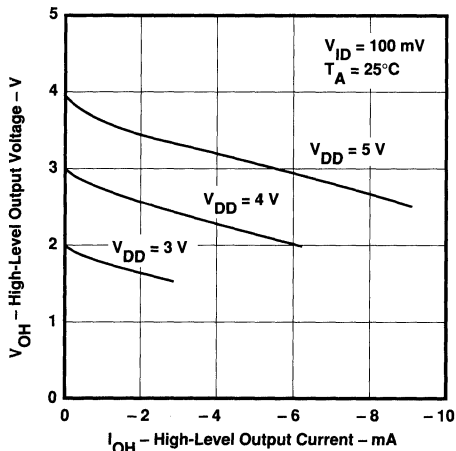


FIGURE 6

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

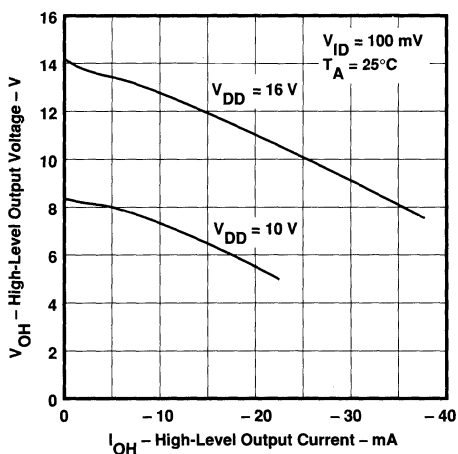


FIGURE 7

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 SUPPLY VOLTAGE

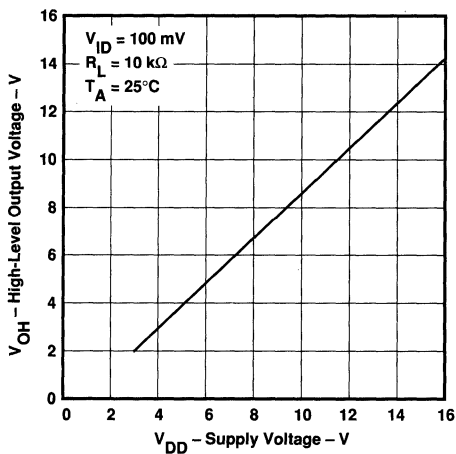


FIGURE 8

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

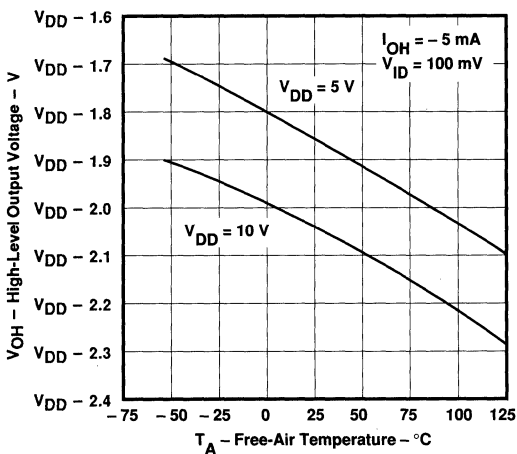


FIGURE 9

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

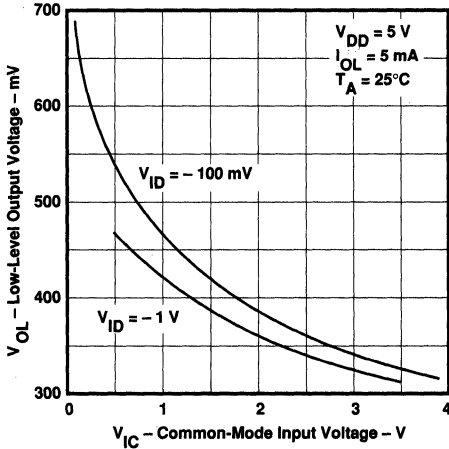


FIGURE 10

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

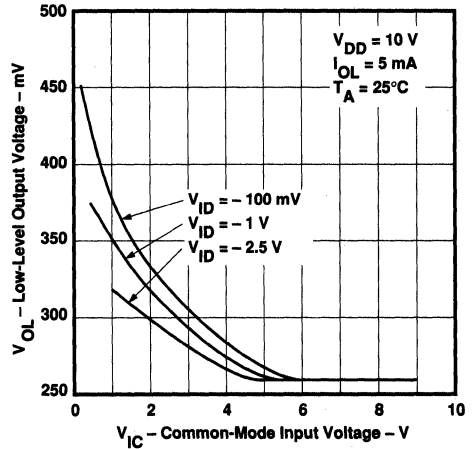


FIGURE 11

LOW-LEVEL OUTPUT VOLTAGE
 vs
 DIFFERENTIAL INPUT VOLTAGE

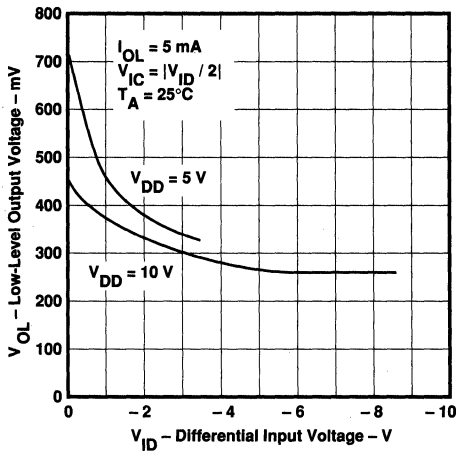


FIGURE 12

LOW-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

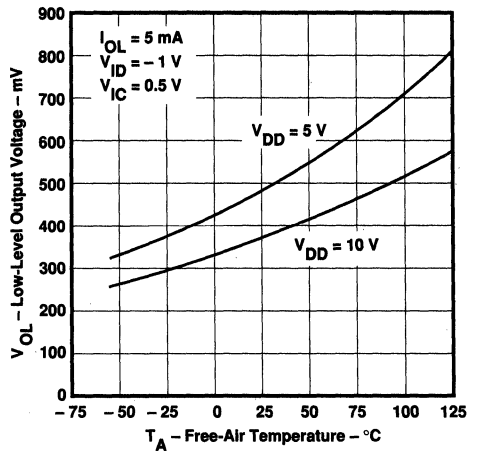


FIGURE 13

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†

LOW-LEVEL OUTPUT VOLTAGE
vs
LOW-LEVEL OUTPUT CURRENT

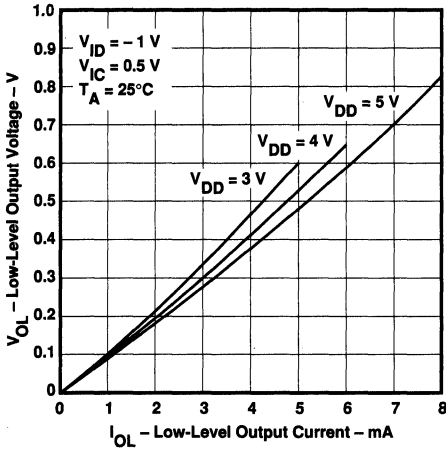


FIGURE 14

LOW-LEVEL OUTPUT VOLTAGE
vs
LOW-LEVEL OUTPUT CURRENT

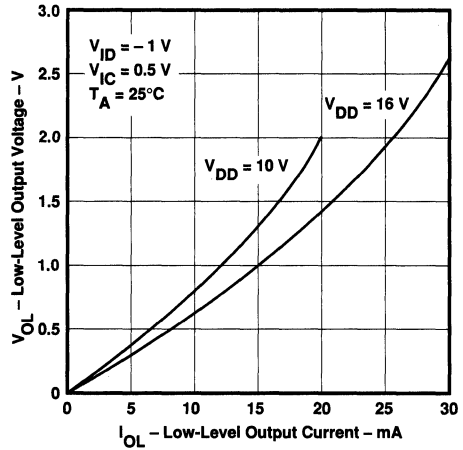


FIGURE 15

LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
SUPPLY VOLTAGE

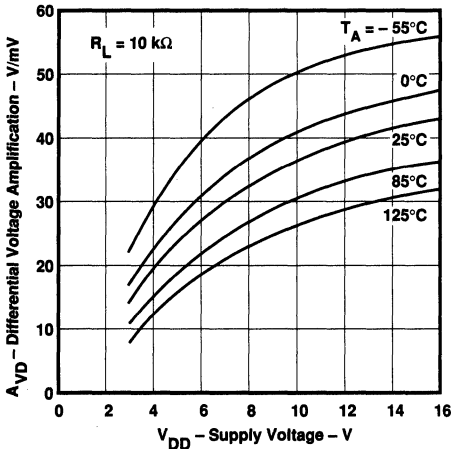


FIGURE 16

LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREE-AIR TEMPERATURE

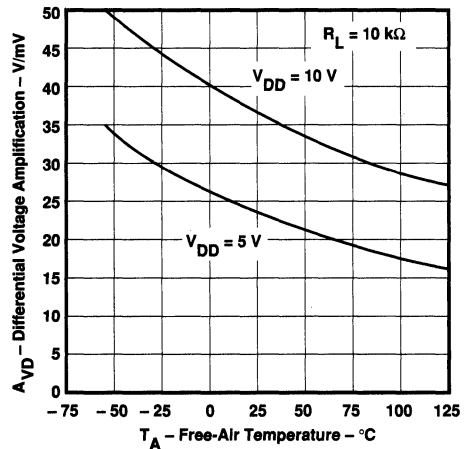


FIGURE 17

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Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†

INPUT BIAS CURRENT AND INPUT OFFSET CURRENT
vs
FREE-AIR TEMPERATURE

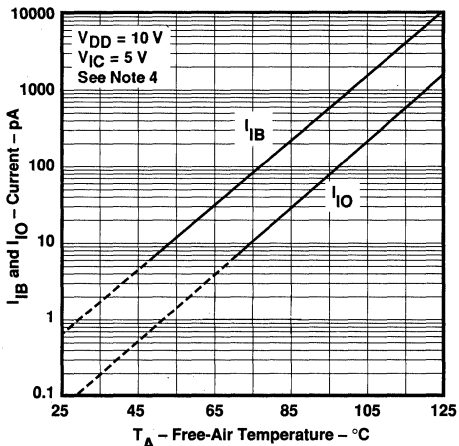


FIGURE 18

COMMON-MODE INPUT VOLTAGE
POSITIVE LIMIT
vs
SUPPLY VOLTAGE

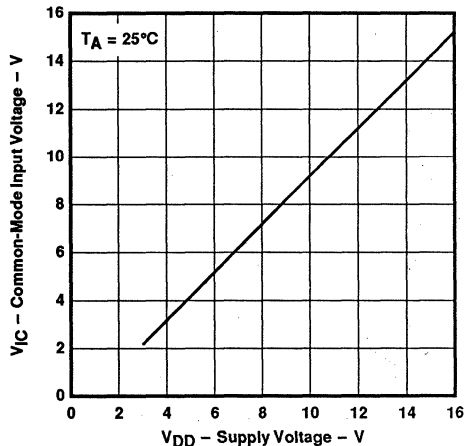


FIGURE 19

SUPPLY CURRENT
vs
SUPPLY VOLTAGE

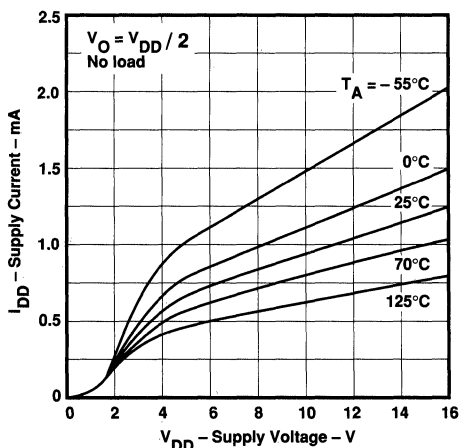


FIGURE 20

SUPPLY CURRENT
vs
FREE-AIR TEMPERATURE

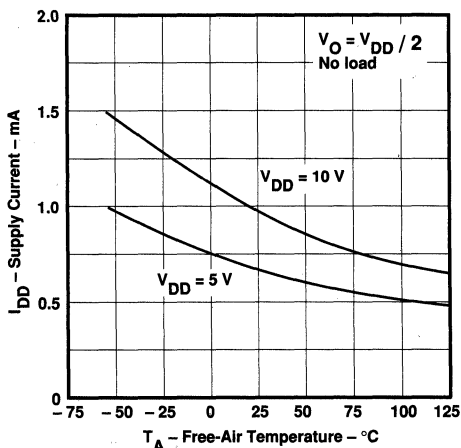


FIGURE 21

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

NOTE 4: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†

SLEW RATE
 vs
 SUPPLY VOLTAGE

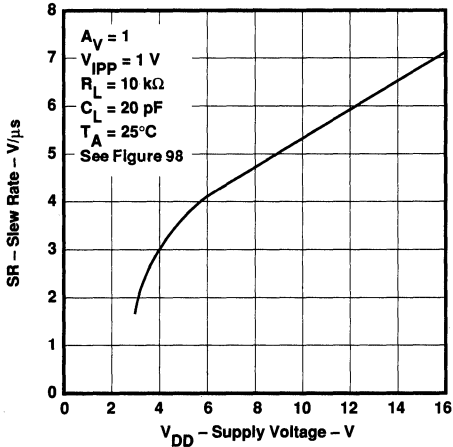


FIGURE 22

SLEW RATE
 vs
 FREE-AIR TEMPERATURE

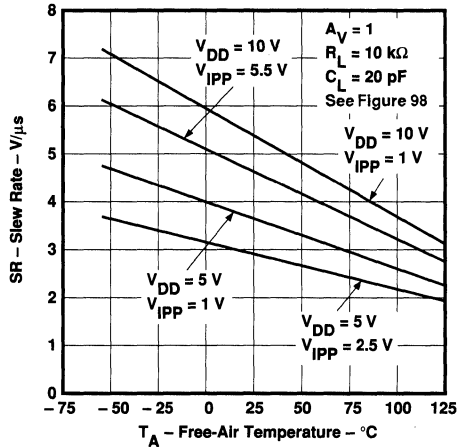


FIGURE 23

BIAS SELECT CURRENT
 vs
 SUPPLY VOLTAGE

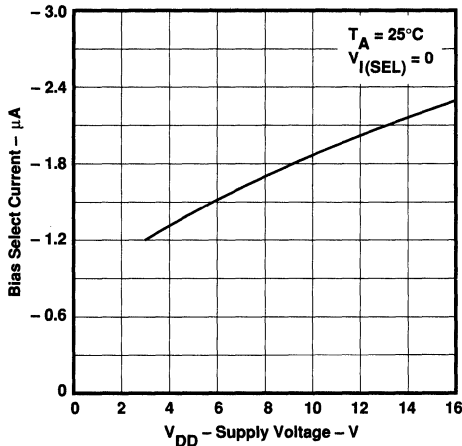


FIGURE 24

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 FREQUENCY

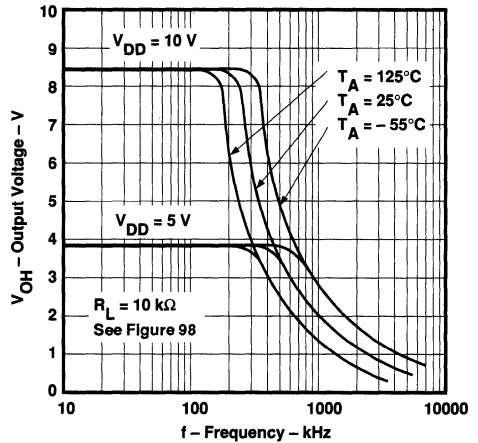


FIGURE 25

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 Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†

UNITY-GAIN BANDWIDTH
 vs
 FREE-AIR TEMPERATURE

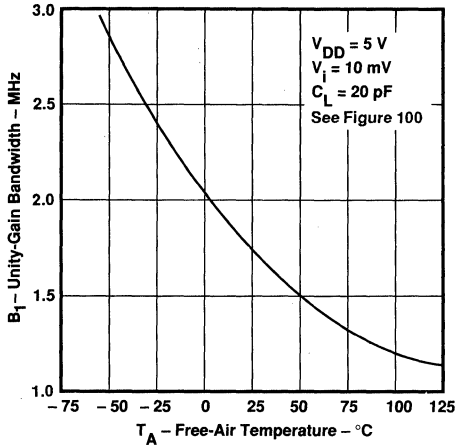


FIGURE 26

UNITY-GAIN BANDWIDTH
 vs
 SUPPLY VOLTAGE

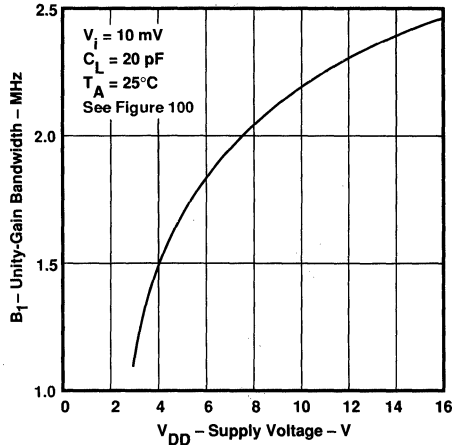


FIGURE 27

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

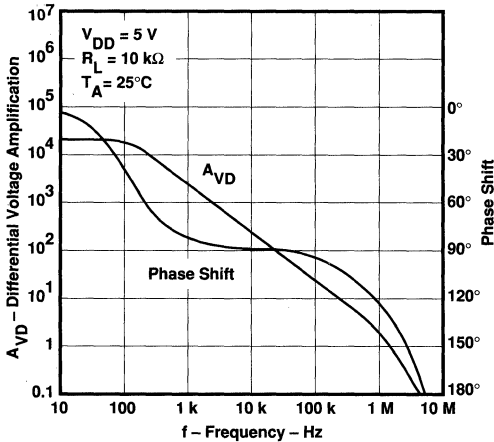


FIGURE 28

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

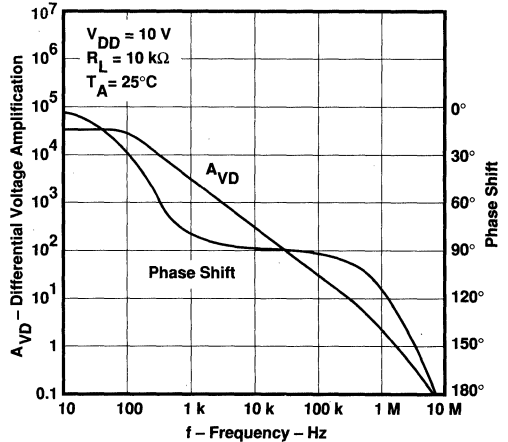


FIGURE 29

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†

PHASE MARGIN
 vs
 SUPPLY VOLTAGE

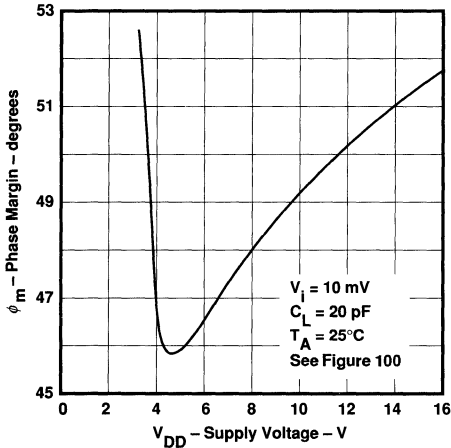


FIGURE 30

PHASE MARGIN
 vs
 FREE-AIR TEMPERATURE

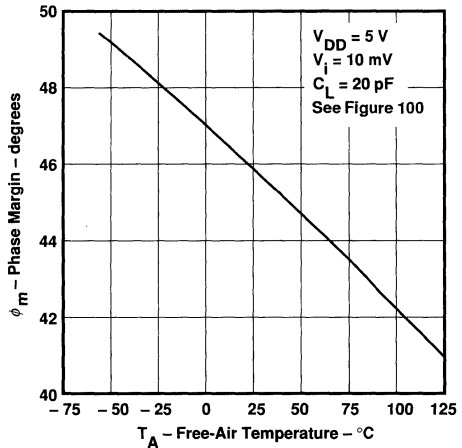


FIGURE 31

PHASE MARGIN
 vs
 CAPACITIVE LOAD

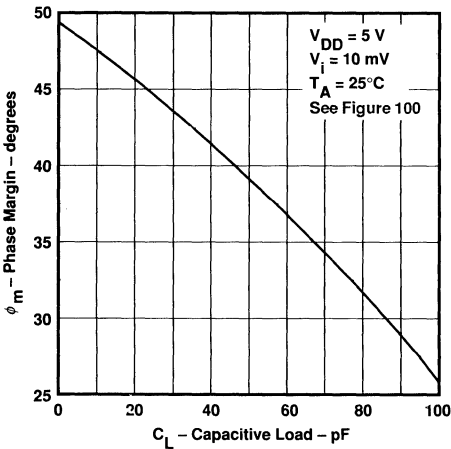


FIGURE 32

EQUIVALENT INPUT NOISE VOLTAGE
 vs
 FREQUENCY

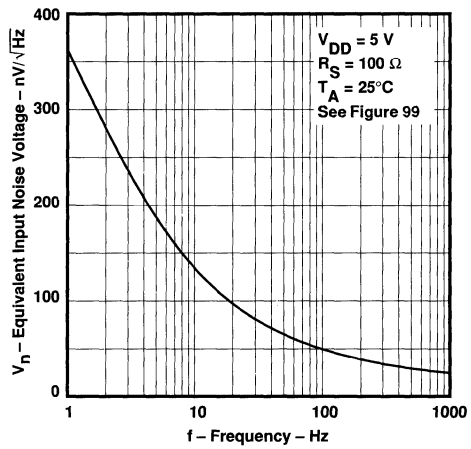


FIGURE 33

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TLC271M
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

MEDIUM-BIAS MODE

electrical characteristics at specified free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T _A [†]	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 100 kΩ	25°C	1.1 10			1.1 10			mV
		Full range	12			12			
αV _{IO} Average temperature coefficient of input offset voltage		25°C to 125°C	1.7			2.1			μV/°C
I _{IO} Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1			0.1			pA
		125°C	1.4	15		1.8	15		nA
I _{IB} Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6			0.7			pA
		125°C	9	35		10	35		nA
V _{ICR} Common-mode input voltage range (see Note 5)		25°C	0 to 4	-0.3 to 4.2		0 to 9	-0.3 to 9.2		V
		Full range	0 to 3.5	to		0 to 8.5	to		V
V _{OH} High-level output voltage	V _{ID} = 100 mV, R _L = 100 kΩ	25°C	3.2	3.9		8	8.7		V
		-55°C	3	3.9		7.8	8.6		
		125°C	3	4		7.8	8.8		
V _{OL} Low-level output voltage	V _{ID} = -100 mV, I _{OL} = 0	25°C	0 50			0 50			mV
		-55°C	0 50			0 50			
		125°C	0 50			0 50			
A _{VD} Large-signal differential voltage amplification	R _L = 100 kΩ, See Note 6	25°C	25	170		25	275		V/mV
		-55°C	15	290		15	420		
		125°C	15	120		15	190		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65	91		65	94		dB
		-55°C	60	89		60	93		
		125°C	60	91		60	93		
k _{SVR} Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V	25°C	70	93		70	93		dB
		-55°C	60	91		60	91		
		125°C	60	94		60	94		
I _{I(SEL)} Input current to bias select pin	V _{I(SEL)} = V _{DD} /2	25°C	-130			-160			nA
I _{DD} Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	105 280			143 300			μA
		-55°C	170 440			245 500			
		125°C	70 180			90 240			

[†]Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

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Operational Amplifiers

TLC2711, TLC271A1, TLC271B1
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

MEDIUM-BIAS MODE

electrical characteristics at specified free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A [†]	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 100 kΩ	25°C		1.1	10		1.1	10	mV
			Full range			13			13	
	TLC271A1		25°C		0.9	5		0.9	5	
			Full range			7			7	
	TLC271B1		25°C		0.25	2		0.26	2	
			Full range			3.5			3.5	
α _{VIO}	Average temperature coefficient of input offset voltage		25°C to 85°C		1.7		2.1		μV/°C	
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C		0.1			0.1	pA	
			85°C		24	1000		26		1000
I _{IB}	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C		0.6			0.7	pA	
			85°C		200	2000		220		2000
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2		-0.2 to 9	-0.3 to 9.2	V	
			Full range	-0.2 to 3.5			-0.2 to 8.5		V	
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 100 kΩ	25°C		3.2	3.9		8	8.7	V
			-40°C		3	3.9		7.8	8.7	
			85°C		3	4		7.8	8.7	
V _{OL}	Low-level output voltage	V _{ID} = -100 mV, I _{OL} = 0	25°C		0	50		0	50	mV
			-40°C		0	50		0	50	
			85°C		0	50		0	50	
A _{VD}	Large-signal differential voltage amplification	R _L = 100 kΩ, See Note 6	25°C		25	170		25	275	V/mV
			-40°C		15	270		15	390	
			85°C		15	130		15	220	
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C		65	91		65	94	dB
			-40°C		60	90		60	93	
			85°C		60	90		60	94	
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V	25°C		70	93		70	93	dB
			-40°C		60	91		60	91	
			85°C		60	94		60	94	
I _{I(SEL)}	Input current to bias select pin	V _{I(SEL)} = V _{DD} /2	25°C		-130			-160	nA	
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C		105	280		143	300	μA
			-40°C		158	400		225	450	
			85°C		80	200		103	260	

[†]Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

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Operational Amplifiers

TLC271C, TLC271AC, TLC271BC
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

MEDIUM-BIAS MODE

electrical characteristics at specified free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A [†]	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 100 kΩ	25°C	1.1	10		1.1	10	mV	
			Full range		12		12			
			25°C	0.9	5	0.9	5			
			Full range		6.5		6.5			
			25°C	0.25	2	0.26	2			
			Full range		3		3			
α _{VIO}	Average temperature coefficient of input offset voltage		25°C to 70°C	1.7		2.1		μV/°C		
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1		0.1		pA		
			70°C	7	300	7	300			
I _B	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6		0.7		pA		
			70°C	40	600	50	600			
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2	-0.2 to 9	-0.3 to 9.2	V		
			Full range	-0.2 to 3.5		-0.2 to 8.5		V		
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 100 kΩ	25°C	3.2	3.9	8	8.7	V		
			0°C	3	3.9	7.8	8.7			
			70°C	3	4	7.8	8.7			
V _{OL}	Low-level output voltage	V _{ID} = -100 mV, I _{OL} = 0	25°C		0 50		0 50	mV		
			0°C		0 50		0 50			
			70°C		0 50		0 50			
A _{VD}	Large-signal differential voltage amplification	R _L = 100 kΩ, See Note 6	25°C	25	170	25	275	V/mV		
			0°C	15	200	15	320			
			70°C	15	140	15	230			
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65	91	65	94	dB		
			0°C	60	91	60	94			
			70°C	60	92	60	94			
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V	25°C	70	93	70	93	dB		
			0°C	60	92	60	92			
			70°C	60	94	60	94			
I _{I(SEL)}	Input current to bias select pin	V _{I(SEL)} = V _{DD} /2	25°C	-130		-160		nA		
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	105	280	143	300	μA		
			0°C	125	320	173	400			
			70°C	85	220	110	280			

[†] Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.
6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

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Operational Amplifiers

MEDIUM-BIAS MODE

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{Ipp} = 1\text{ V}$	25°C		0.43		V/ μ s
			-55°C		0.54		
			125°C		0.29		
		$V_{Ipp} = 2.5\text{ V}$	25°C		0.40		
			-55°C		0.50		
			125°C		0.28		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$	25°C		32		nV/ $\sqrt{\text{Hz}}$
BOM Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C		55		kHz
			-55°C		80		
			125°C		40		
B ₁ Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$	25°C		525		kHz
			-55°C		850		
			125°C		330		
			25°C		40°		
ϕ_m Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$	$f = B_1$, See Figure 100	25°C		43°		
			-55°C		43°		
			125°C		36°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{Ipp} = 1\text{ V}$	25°C		0.62		V/ μ s
			-55°C		0.81		
			125°C		0.38		
		$V_{Ipp} = 5.5\text{ V}$	25°C		0.56		
			-55°C		0.73		
			125°C		0.35		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$	25°C		32		nV/ $\sqrt{\text{Hz}}$
BOM Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C		35		kHz
			-55°C		50		
			125°C		20		
B ₁ Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$	25°C		635		kHz
			-55°C		960		
			125°C		440		
			25°C		43°		
ϕ_m Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$	$f = B_1$, See Figure 100	25°C		43°		
			-55°C		47°		
			125°C		39°		

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LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

MEDIUM-BIAS MODE

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{IP} = 1\text{ V}$	25°C		0.43		V/ μ s
				-40°C		0.51		
				85°C		0.35		
			$V_{IP} = 2.5\text{ V}$	25°C		0.40		
				-40°C		0.48		
				85°C		0.32		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$,	25°C		32		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C		55		kHz
				-40°C		75		
				85°C		45		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		525		kHz
				-40°C		770		
				85°C		370		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	25°C		40°		
				-40°C		43°		
				85°C		38°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{IP} = 1\text{ V}$	25°C		0.62		V/ μ s
				-40°C		0.77		
				85°C		0.47		
			$V_{IP} = 5.5\text{ V}$	25°C		0.56		
				-40°C		0.70		
				85°C		0.44		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$,	25°C		32		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C		35		kHz
				-40°C		45		
				85°C		25		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		635		kHz
				-40°C		880		
				85°C		480		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	25°C		43°		
				-40°C		46°		
				85°C		41°		

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Operational Amplifiers

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LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

MEDIUM-BIAS MODE

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{IPP} = 1\text{ V}$	25°C		0.43		V/ μ s
				0°C		0.46		
				70°C		0.36		
			$V_{IPP} = 2.5\text{ V}$	25°C		0.40		
				0°C		0.43		
				70°C		0.34		
V_n	Equivalent input noise voltage	$f_i = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$	25°C		32		nV/ $\sqrt{\text{Hz}}$
B _{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C		55		kHz
				0°C		60		
				70°C		50		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, See Figure 100	25°C		525		kHz
				0°C		600		
				70°C		400		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$, See Figure 100	$f = B_1$, See Figure 100	25°C		40°		
				0°C		41°		
				70°C		39°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{IPP} = 1\text{ V}$	25°C		0.62		V/ μ s
				0°C		0.67		
				70°C		0.51		
			$V_{IPP} = 5.5\text{ V}$	25°C		0.56		
				0°C		0.61		
				70°C		0.46		
V_n	Equivalent input noise voltage	$f_i = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$	25°C		32		nV/ $\sqrt{\text{Hz}}$
B _{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C		35		kHz
				0°C		40		
				70°C		30		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, See Figure 100	25°C		635		kHz
				0°C		710		
				70°C		510		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$, See Figure 100	$f = B_1$, See Figure 100	25°C		43°		
				0°C		44°		
				70°C		42°		

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Operational Amplifiers

TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†

**DISTRIBUTION OF TLC271
 INPUT OFFSET VOLTAGE**

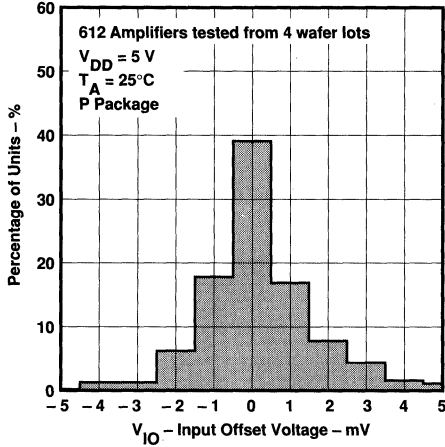


FIGURE 34

**DISTRIBUTION OF TLC271
 INPUT OFFSET VOLTAGE**

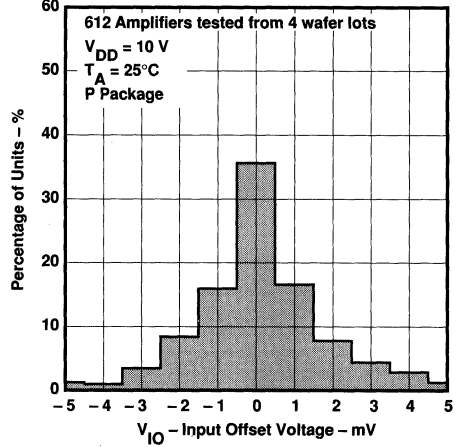


FIGURE 35

**DISTRIBUTION OF TLC271
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT**

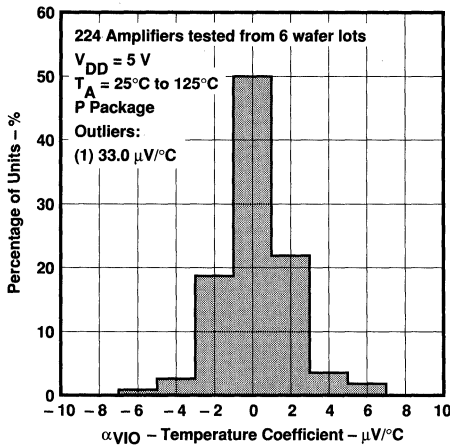


FIGURE 36

**DISTRIBUTION OF TLC271
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT**

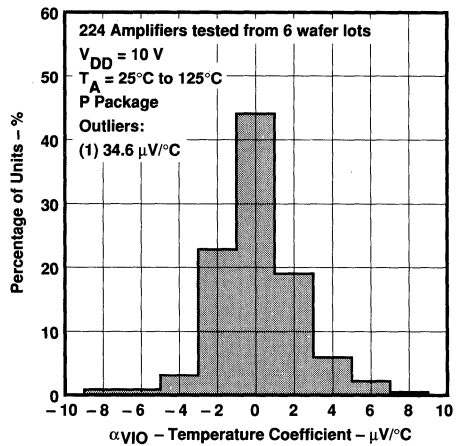


FIGURE 37

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

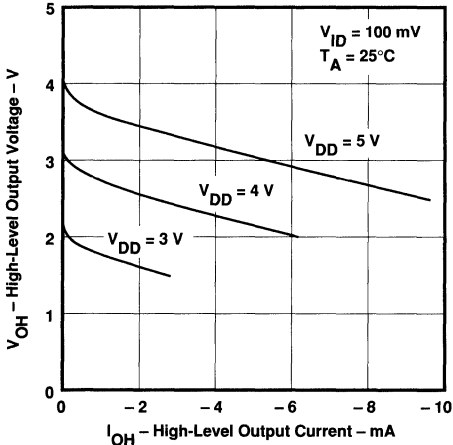


FIGURE 38

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

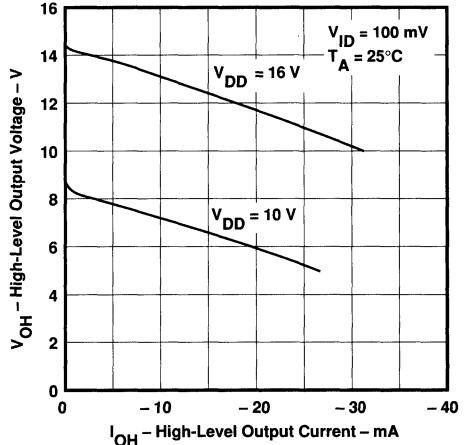


FIGURE 39

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 SUPPLY VOLTAGE

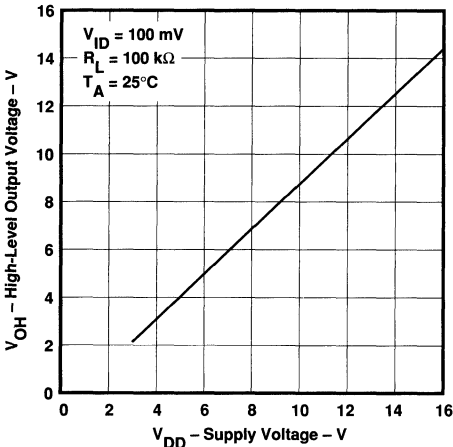


FIGURE 40

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

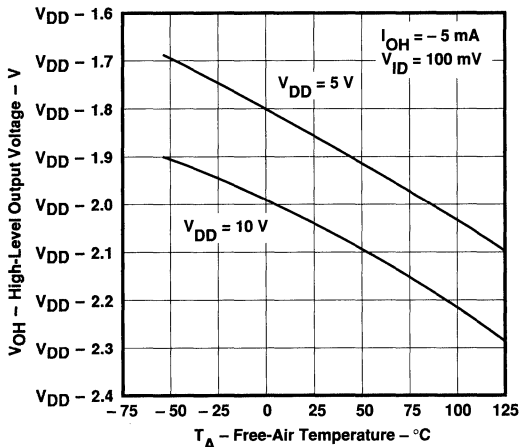


FIGURE 41

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

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TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†

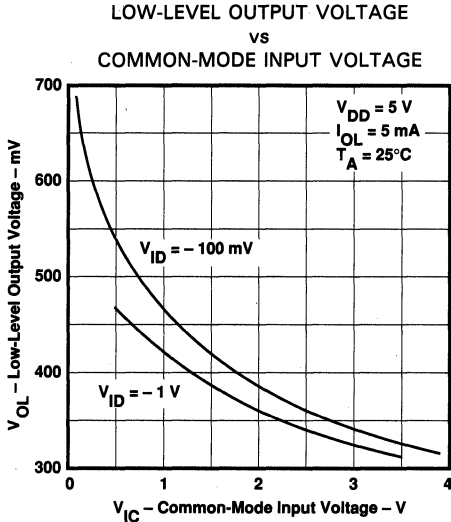


FIGURE 42

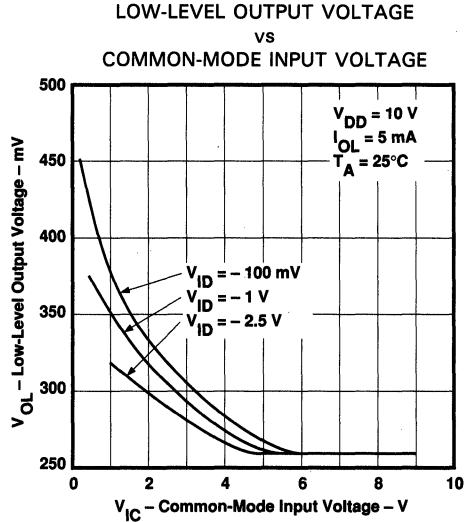


FIGURE 43

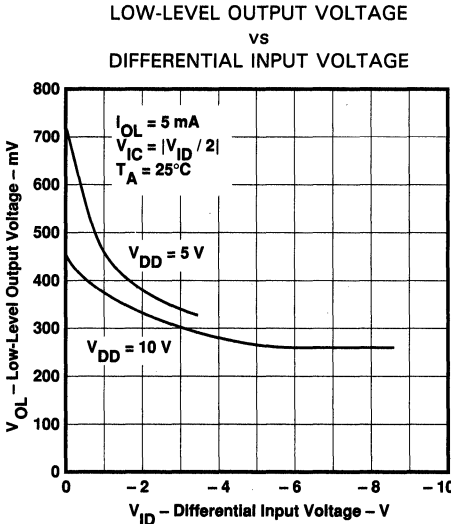


FIGURE 44

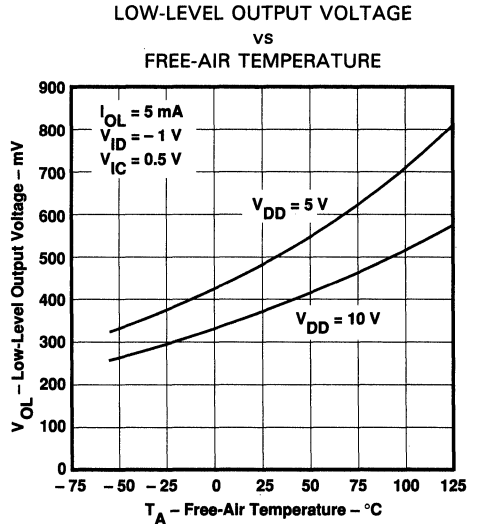


FIGURE 45

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

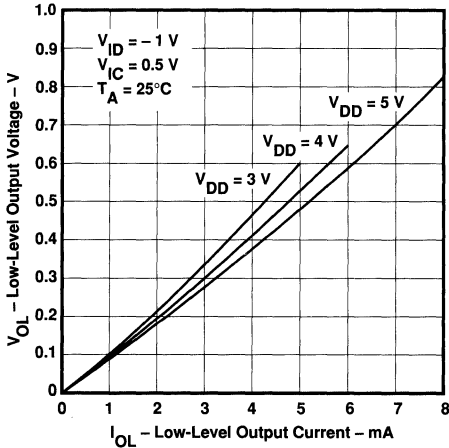


FIGURE 46

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

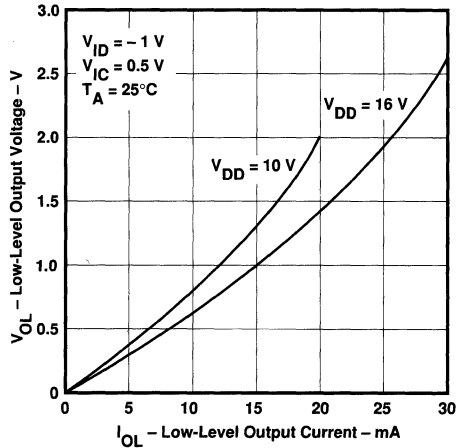


FIGURE 47

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 SUPPLY VOLTAGE

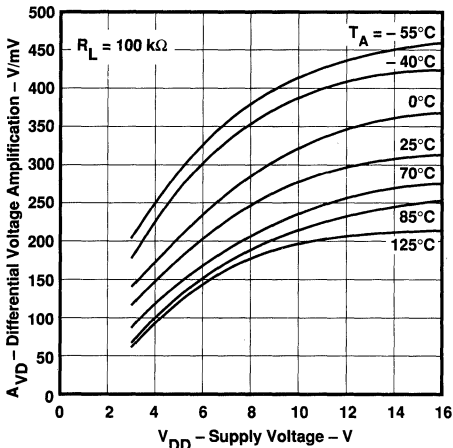


FIGURE 48

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 FREE-AIR TEMPERATURE

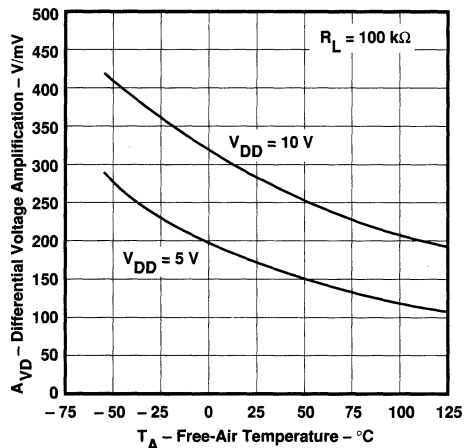


FIGURE 49

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

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OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†

INPUT BIAS CURRENT AND INPUT
 OFFSET CURRENT

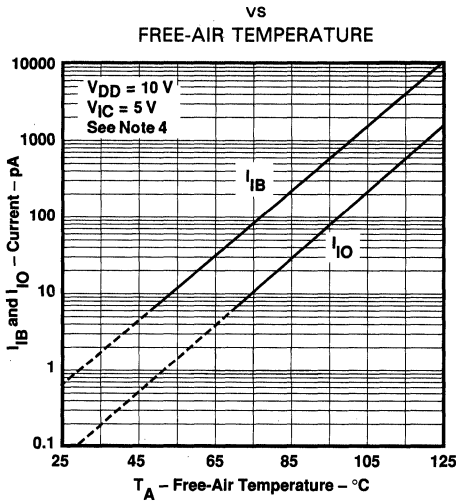


FIGURE 50

COMMON-MODE INPUT VOLTAGE
 POSITIVE LIMIT

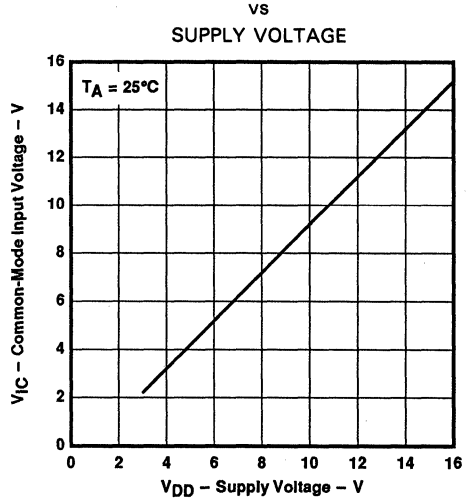


FIGURE 51

SUPPLY CURRENT
 vs
 SUPPLY VOLTAGE

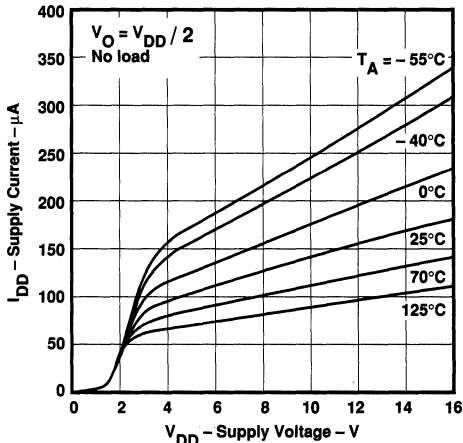


FIGURE 52

SUPPLY CURRENT
 vs
 FREE-AIR TEMPERATURE

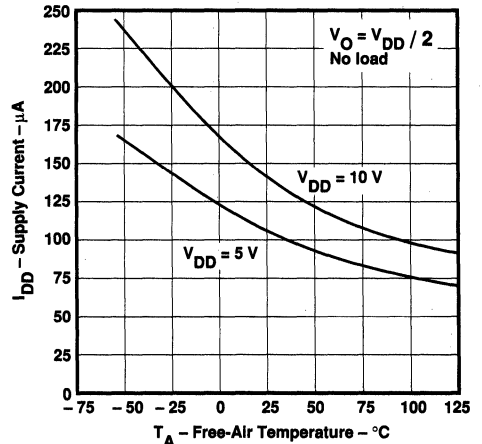


FIGURE 53

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.
 NOTE 4: The typical values of input bias current and input offset current and input offset current below 5 pA were determined mathematically.

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†

SLEW RATE
vs
SUPPLY VOLTAGE

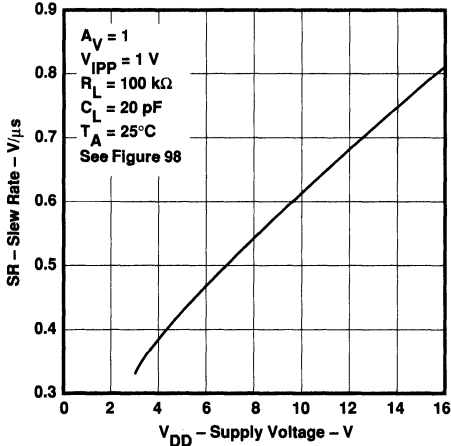


FIGURE 54

SLEW RATE
vs
FREE-AIR TEMPERATURE

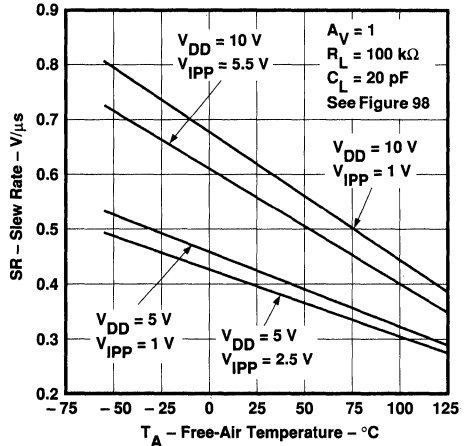


FIGURE 55

BIAS SELECT CURRENT
vs
SUPPLY VOLTAGE

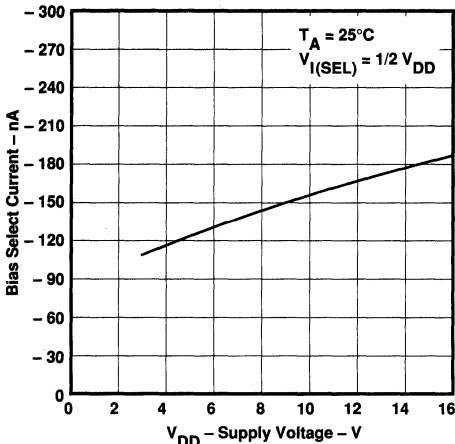


FIGURE 56

HIGH-LEVEL OUTPUT VOLTAGE
vs
FREQUENCY

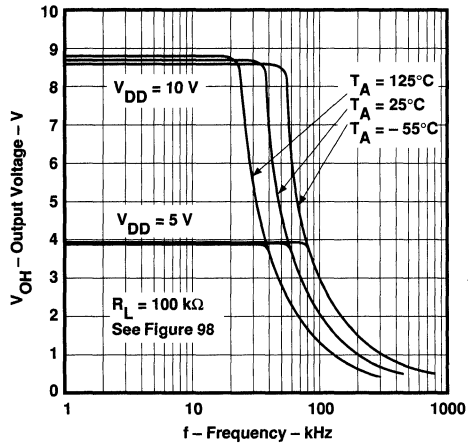


FIGURE 57

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TLC271, TLC271A, TLC271B
linCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†

UNITY-GAIN BANDWIDTH
 vs
 FREE-AIR TEMPERATURE

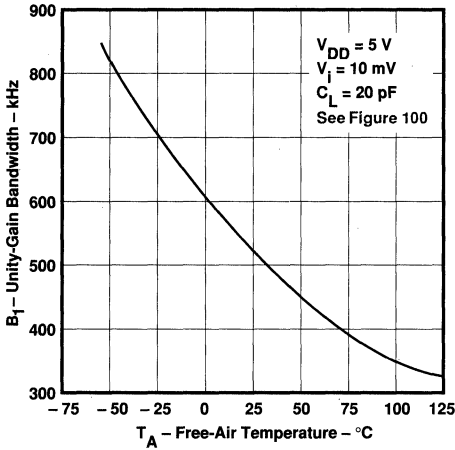


FIGURE 58

UNITY-GAIN BANDWIDTH
 vs
 SUPPLY VOLTAGE

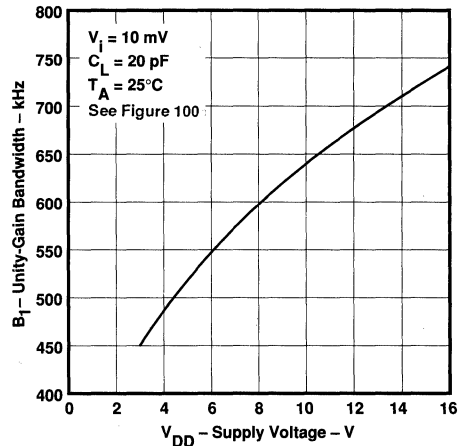


FIGURE 59

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

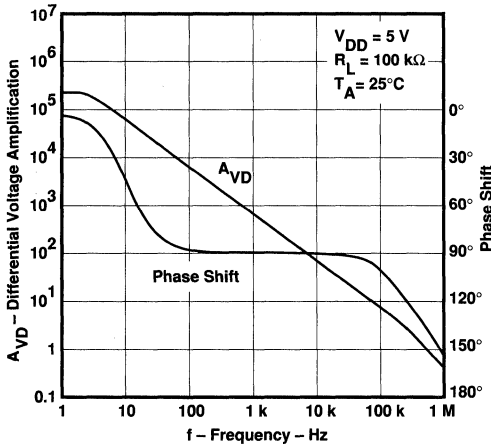


FIGURE 60

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

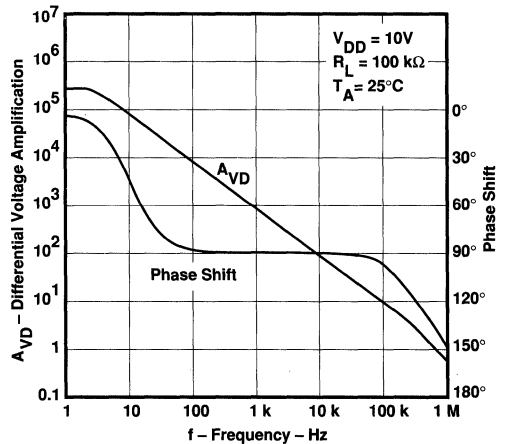


FIGURE 61

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†

PHASE MARGIN
 vs
 SUPPLY VOLTAGE

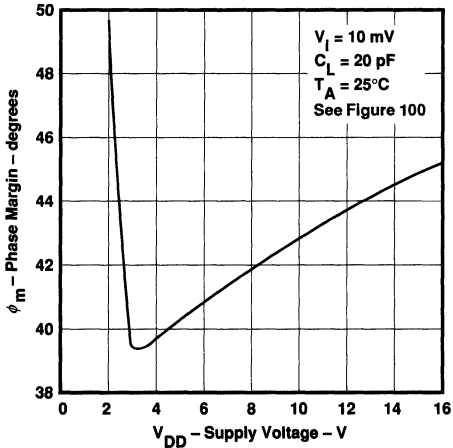


FIGURE 62

PHASE MARGIN
 vs
 FREE-AIR TEMPERATURE

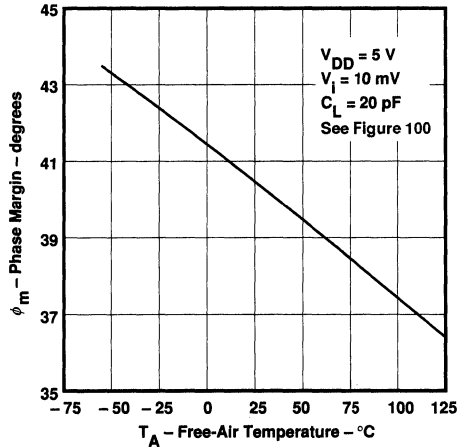


FIGURE 63

PHASE MARGIN
 vs
 CAPACITIVE LOAD

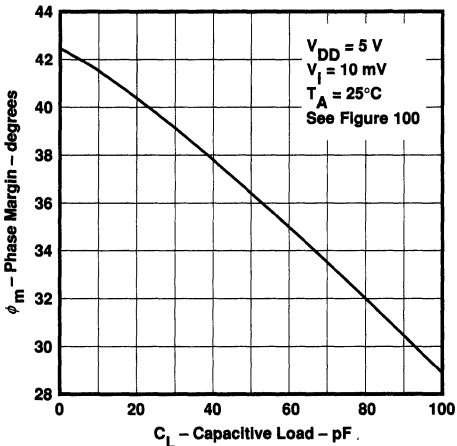


FIGURE 64

EQUIVALENT INPUT NOISE VOLTAGE
 vs
 FREQUENCY

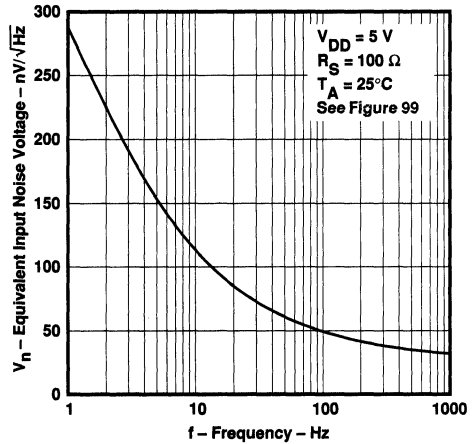


FIGURE 65

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TLC271M
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

LOW-BIAS MODE

electrical characteristics at specified free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T _A †	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 1 MΩ	25 °C		1.1	10		1.1	10	mV
		Full range						12	
αV _{IO} Average temperature coefficient of input offset voltage		25 °C to 125 °C		1.4			1.4		μV/°C
I _{IO} Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25 °C		0.1			0.1		pA
		125 °C		1.4	15		1.8	15	nA
I _{IB} Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25 °C		0.6			0.7		pA
		125 °C		9	35		10	35	nA
V _{ICR} Common-mode input voltage range (see Note 5)		25 °C	0	-0.3		0	-0.3		V
			to	to		to	to		
		4	4.2		9	9.2			
Full range			0			0		V	
			to			to			
3.5			8.5						
V _{OH} High-level output voltage	V _{ID} = 100 mV, R _L = 1 MΩ	25 °C	3.2	4.1		8	8.9	V	
		-55 °C	3	4.1		7.8	8.8		
		125 °C	3	4.2		7.8	9		
V _{OL} Low-level output voltage	V _{ID} = -100 mV, I _{OL} = 0	25 °C		0	50		0	50	mV
		-55 °C		0	50		0	50	
		125 °C		0	50		0	50	
A _{VD} Large-signal differential voltage amplification	R _L = 1 MΩ, See Note 6	25 °C	50	520		50	870	V/mV	
		-55 °C	25	1000		25	1775		
		125 °C	25	200		25	380		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25 °C	65	94		65	97	dB	
		-55 °C	60	95		60	97		
		125 °C	60	85		60	91		
k _{SVR} Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V	25 °C	70	97		70	97	dB	
		-55 °C	60	97		60	97		
		125 °C	60	98		60	98		
I _{I(SEL)} Input current to bias select pin	V _{I(SEL)} = V _{DD}	25 °C		65		95	nA		
I _{DD} Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25 °C		10	17		14	23	μA
		-55 °C		17	30		28	48	
		125 °C		7	12		9	15	

† Full range is -55 °C to 125 °C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

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Operational Amplifiers

TLC2711, TLC271AI, TLC271BI
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

LOW-BIAS MODE

electrical characteristics at specified free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A [†]	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 1 MΩ	25°C	1.1		10	1.1		10	mV
			Full range			13			13	
			25°C	0.9		5	0.9		5	
			Full range			7			7	
			25°C	0.24		2	0.26		2	
			Full range			3.5			3.5	
αV _{IO}	Average temperature coefficient of input offset voltage		25°C to 85°C	1.1			1		μV/°C	
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1			0.1		pA	
			85°C	24	1000	26	1000			
I _{IB}	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6			0.7		pA	
			85°C	200	2000	220	2000			
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2		-0.2 to 9	-0.3 to 9.2	V	
			Full range	-0.2 to 3.5			-0.2 to 8.5		V	
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 1 MΩ	25°C	3.2	4.1		8	8.9	V	
			-40°C	3	4.1		7.8	8.9		
			85°C	3	4.2		7.8	8.9		
V _{OL}	Low-level output voltage	V _{ID} = -100 mV, I _{OL} = 0	25°C		0	50		0	50	mV
			-40°C		0	50		0	50	
			85°C		0	50		0	50	
A _{VD}	Large-signal differential voltage amplification	R _L = 1 MΩ, See Note 6	25°C	50	520		50	870	V/mV	
			-40°C	50	900		50	1550		
			85°C	50	330		50	585		
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65	94		65	97	dB	
			-40°C	60	95		60	97		
			85°C	60	95		60	98		
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V	25°C	70	97		70	97	dB	
			-40°C	60	97		60	97		
			85°C	60	98		60	98		
I _{I(SEL)}	Input current to bias select pin	V _{I(SEL)} = V _{DD}	25°C	65			95		nA	
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	10	17		14	23	μA	
			-40°C	16	27		25	43		
			85°C	7	13		10	18		

[†]Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

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Operational Amplifiers

TLC271C, TLC271AC, TLC271BC
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

LOW-BIAS MODE

electrical characteristics at specified free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A [†]	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	TLC271C V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 1 MΩ	25°C	1.1 10			1.1 10			mV
			Full range	12			12			
			25°C	0.9 5			0.9 5			
			Full range	6.5			6.5			
			25°C	0.24 2			0.26 2			
			Full range	3			3			
α _{VIO}	Average temperature coefficient of input offset voltage		25°C to 70°C	1.1			1			μV/°C
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1			0.1			pA
			70°C	7 300			7 300			
I _{IB}	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6			0.7			pA
			70°C	40 600			50 600			
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2		-0.2 to 9	-0.3 to 9.2		V
			Full range	-0.2 to 3.5			-0.2 to 8.5			V
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 1 MΩ	25°C	3.2 4.1			8 8.9			V
			0°C	3 4.1			7.8 8.9			
			70°C	3 4.2			7.8 8.9			
V _{OL}	Low-level output voltage	V _{ID} = -100 mV, I _{OL} = 0	25°C	0 50			0 50			mV
			0°C	0 50			0 50			
			70°C	0 50			0 50			
A _{VD}	Large-signal differential voltage amplification	R _L = 1 MΩ, See Note 6	25°C	50 520			50 870			V/mV
			0°C	50 700			50 1030			
			70°C	50 380			50 660			
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65 94			65 97			dB
			0°C	60 95			60 97			
			70°C	60 95			60 97			
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V	25°C	70 97			70 97			dB
			0°C	60 97			60 97			
			70°C	60 98			60 98			
I _{I(SEL)}	Input current to bias select pin	V _{I(SEL)} = V _{DD}	25°C	65			95			nA
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	10 17			14 23			μA
			0°C	12 21			18 33			
			70°C	8 14			11 20			

[†]Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

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Operational Amplifiers

LOW-BIAS MODE

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{Ipp} = 1\text{ V}$	25 °C		0.03		V/ μ s
				-55 °C		0.04		
				125 °C		0.02		
			$V_{Ipp} = 2.5\text{ V}$	25 °C		0.03		
				-55 °C		0.04		
				125 °C		0.02		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$,	25 °C		68		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25 °C		5		kHz
				-55 °C		8		
				125 °C		3		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25 °C		85		kHz
				-55 °C		140		
				125 °C		45		
				25 °C		34 °		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	-55 °C		39 °		
				125 °C		25 °		
				25 °C				

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{Ipp} = 1\text{ V}$	25 °C		0.05		V/ μ s
				-55 °C		0.06		
				125 °C		0.03		
			$V_{Ipp} = 5.5\text{ V}$	25 °C		0.04		
				-55 °C		0.06		
				125 °C		0.03		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$,	25 °C		68		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25 °C		1		kHz
				-55 °C		1.5		
				125 °C		0.7		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25 °C		110		kHz
				-55 °C		165		
				125 °C		70		
				25 °C		38 °		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	-55 °C		43 °		
				125 °C		29 °		
				25 °C				

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Operational Amplifiers

TLC271I, TLC271AI, TLC271BI
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

LOW-BIAS MODE

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{IPP} = 1\text{ V}$	25°C		0.03		V/ μ s
				-40°C		0.04		
				85°C		0.03		
			$V_{IPP} = 2.5\text{ V}$	25°C		0.03		
				-40°C		0.04		
				85°C		0.02		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$,	25°C		68		nV/ $\sqrt{\text{Hz}}$
BOM	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C		5		kHz
				-40°C		7		
				85°C		4		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		85		kHz
				-40°C		130		
				85°C		55		
				25°C		34°		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	-40°C		38°		
				85°C		28°		
				25°C				

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{IPP} = 1\text{ V}$	25°C		0.05		V/ μ s
				-40°C		0.06		
				85°C		0.03		
			$V_{IPP} = 5.5\text{ V}$	25°C		0.04		
				-40°C		0.05		
				85°C		0.03		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$,	25°C		68		nV/ $\sqrt{\text{Hz}}$
BOM	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C		1		kHz
				-40°C		1.4		
				85°C		0.8		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		110		kHz
				-40°C		155		
				85°C		80		
				25°C		38°		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	-40°C		42°		
				85°C		32°		
				25°C				

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Operational Amplifiers

TLC271C, TLC271AC, TLC271BC
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

LOW-BIAS MODE

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{Ipp} = 1\text{ V}$	25°C		0.03		V/ μ s
				0°C		0.04		
				70°C		0.03		
			$V_{Ipp} = 2.5\text{ V}$	25°C		0.03		
				0°C		0.03		
				70°C		0.02		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$,	25°C		68		nV/ $\sqrt{\text{Hz}}$
B _{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C		5		kHz
				0°C		6		
				70°C		4.5		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		85		kHz
				0°C		100		
				70°C		65		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	25°C		34°		
				0°C		36°		
				70°C		30°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{Ipp} = 1\text{ V}$	25°C		0.05		V/ μ s
				0°C		0.05		
				70°C		0.04		
			$V_{Ipp} = 5.5\text{ V}$	25°C		0.04		
				0°C		0.05		
				70°C		0.04		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 100\ \Omega$,	25°C		68		nV/ $\sqrt{\text{Hz}}$
B _{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C		1		kHz
				0°C		1.3		
				70°C		0.9		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		110		kHz
				0°C		125		
				70°C		90		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	25°C		38°		
				0°C		40°		
				70°C		34°		

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Operational Amplifiers

TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

**DISTRIBUTION OF TLC271
 INPUT OFFSET VOLTAGE**

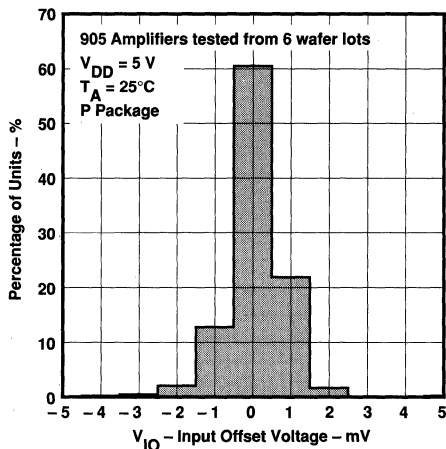


FIGURE 66

**DISTRIBUTION OF TLC271
 INPUT OFFSET VOLTAGE**

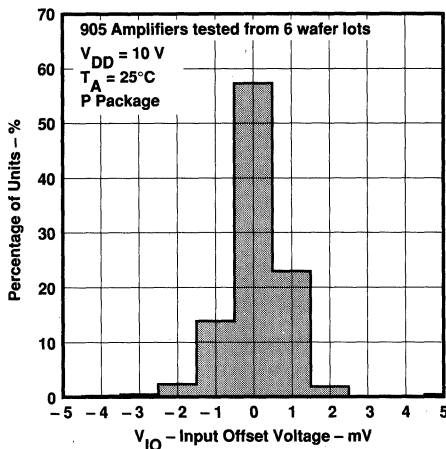


FIGURE 67

**DISTRIBUTION OF TLC271
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT**

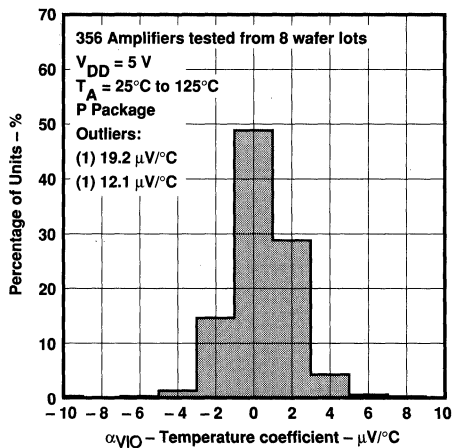


FIGURE 68

**DISTRIBUTION OF TLC271
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT**

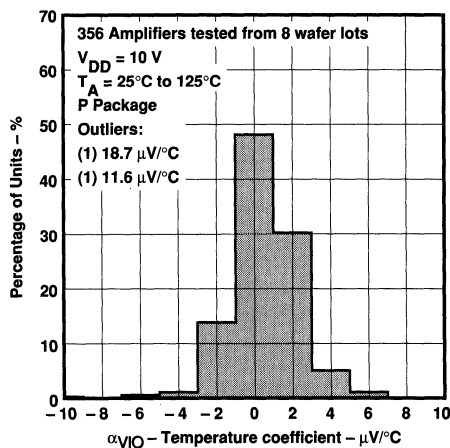


FIGURE 69

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

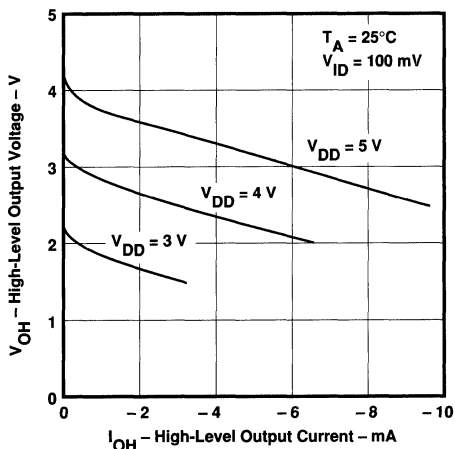


FIGURE 70

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

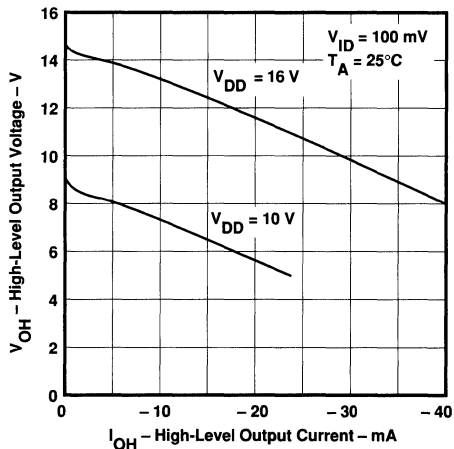


FIGURE 71

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 SUPPLY VOLTAGE

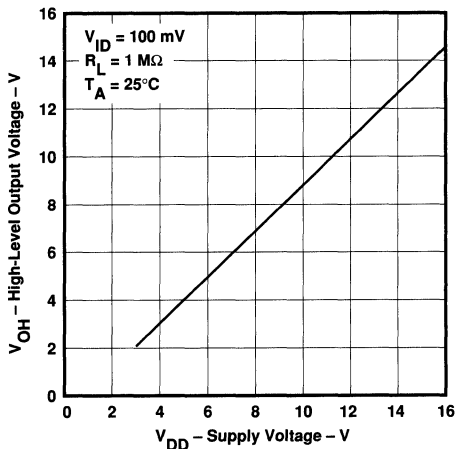


FIGURE 72

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

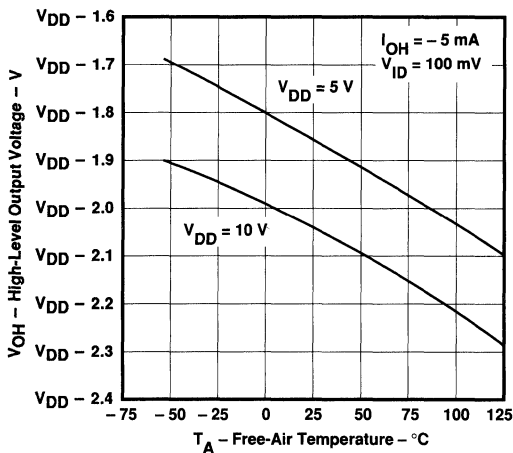


FIGURE 73

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

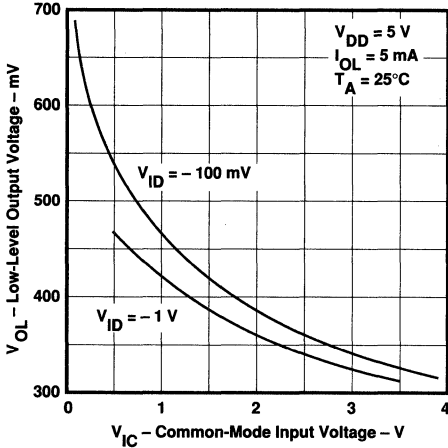


FIGURE 74

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

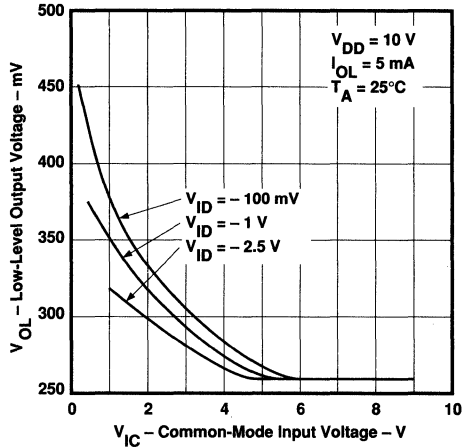


FIGURE 75

LOW-LEVEL OUTPUT VOLTAGE
 vs
 DIFFERENTIAL INPUT VOLTAGE

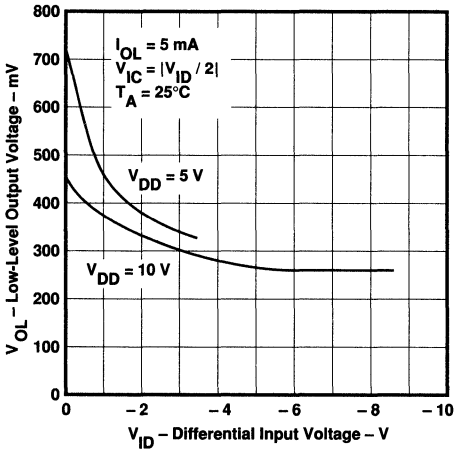


FIGURE 76

LOW-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

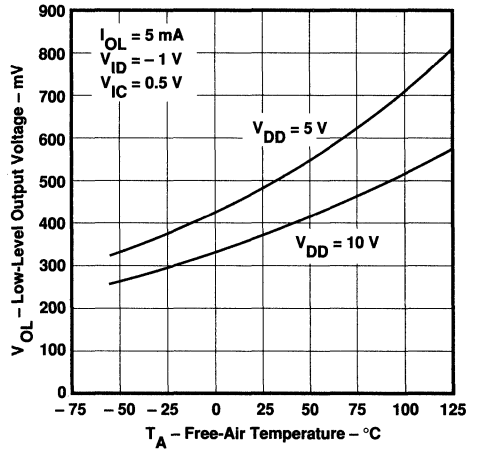


FIGURE 77

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

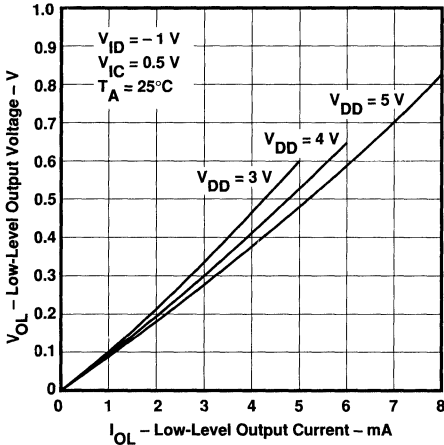


FIGURE 78

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

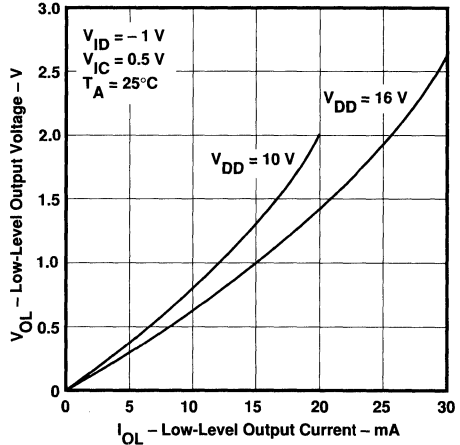


FIGURE 79

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 SUPPLY VOLTAGE

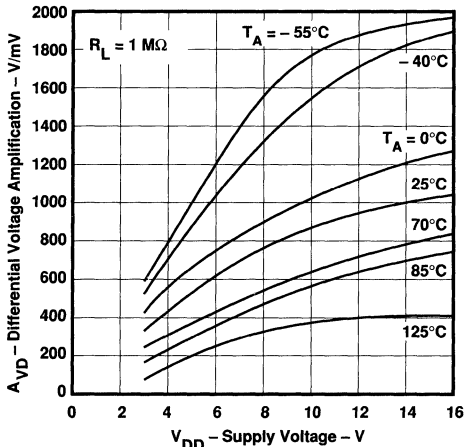


FIGURE 80

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 FREE-AIR TEMPERATURE

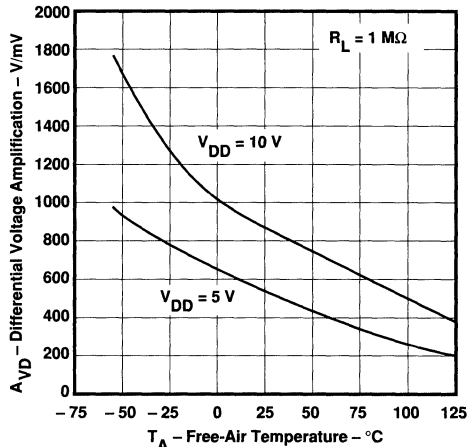


FIGURE 81

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

**INPUT BIAS CURRENT AND
 INPUT OFFSET CURRENT**
 vs
FREE-AIR TEMPERATURE

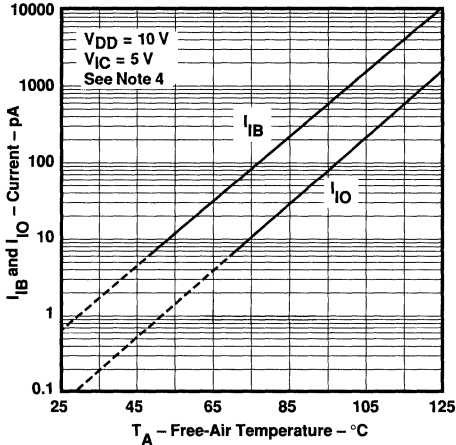


FIGURE 82

**COMMON-MODE INPUT VOLTAGE
 POSITIVE LIMIT**
 vs
SUPPLY VOLTAGE

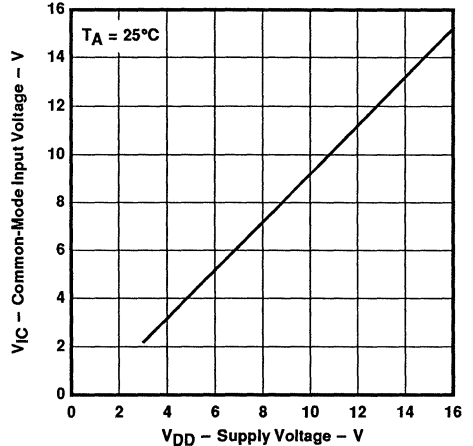


FIGURE 83

SUPPLY CURRENT
 vs
SUPPLY VOLTAGE

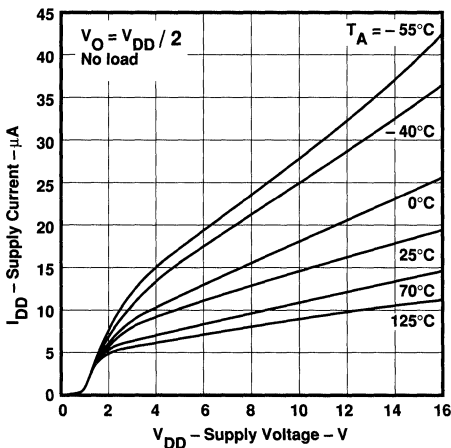


FIGURE 84

SUPPLY CURRENT
 vs
FREE-AIR TEMPERATURE

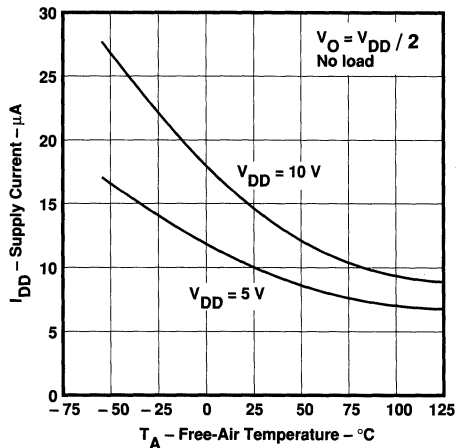


FIGURE 85

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.
 NOTE 4: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

**SLEW RATE
 vs
 SUPPLY VOLTAGE**

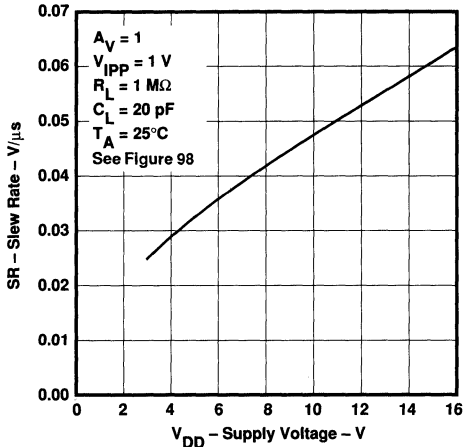


FIGURE 86

**SLEW RATE
 vs
 FREE-AIR TEMPERATURE**

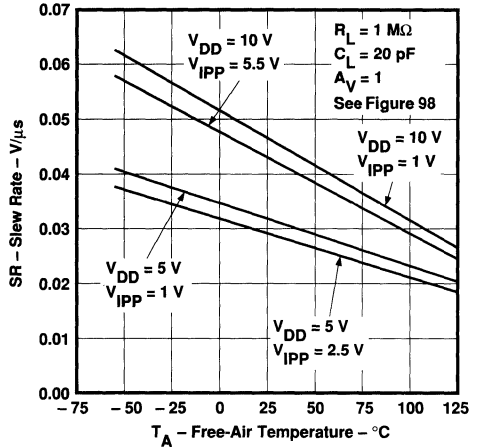


FIGURE 87

**BIAS SELECT CURRENT
 vs
 SUPPLY VOLTAGE**

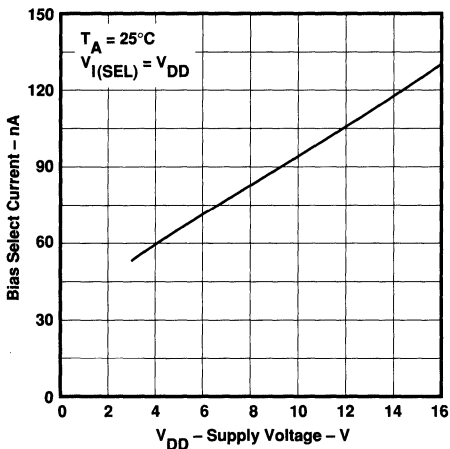


FIGURE 88

**HIGH-LEVEL OUTPUT VOLTAGE
 vs
 FREQUENCY**

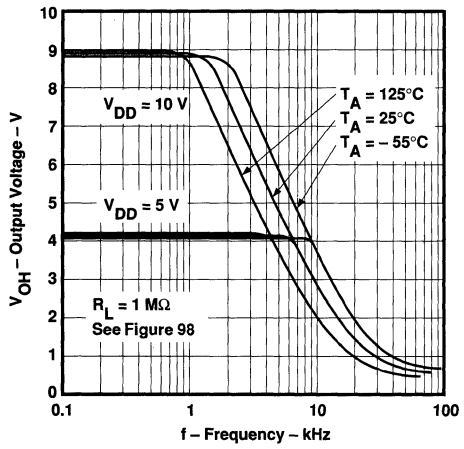


FIGURE 89

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

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TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

UNITY-GAIN BANDWIDTH
vs
FREE-AIR TEMPERATURE

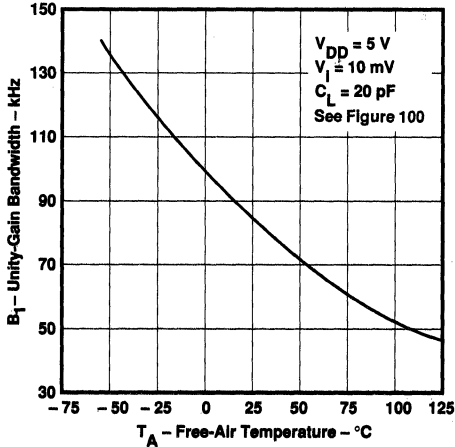


FIGURE 90

UNITY-GAIN BANDWIDTH
vs
SUPPLY VOLTAGE

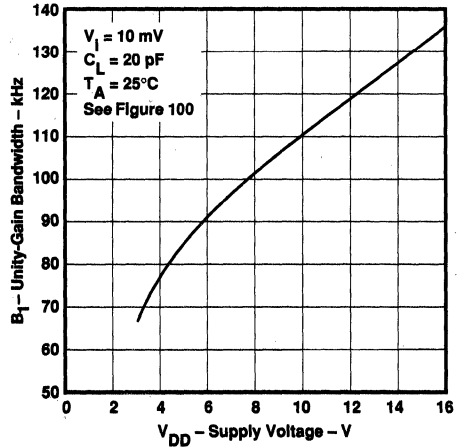


FIGURE 91

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
vs
FREQUENCY

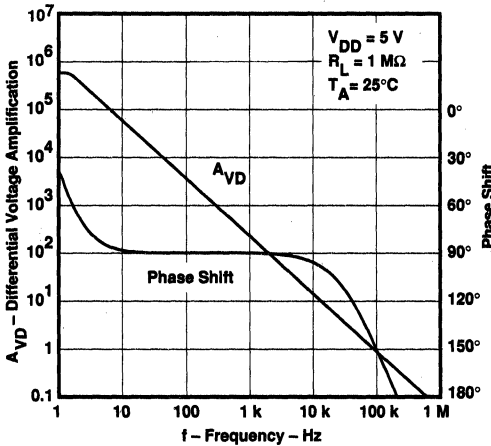


FIGURE 92

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
vs
FREQUENCY

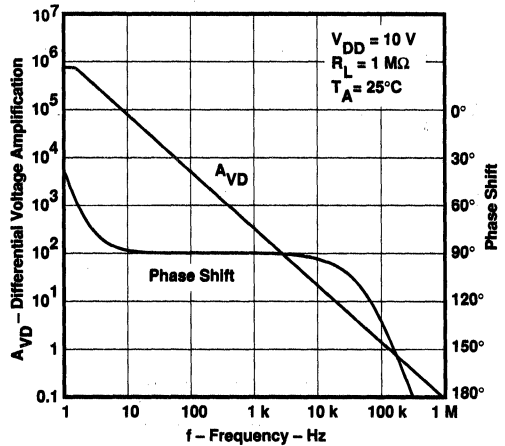


FIGURE 93

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

PHASE MARGIN
 vs
 SUPPLY VOLTAGE

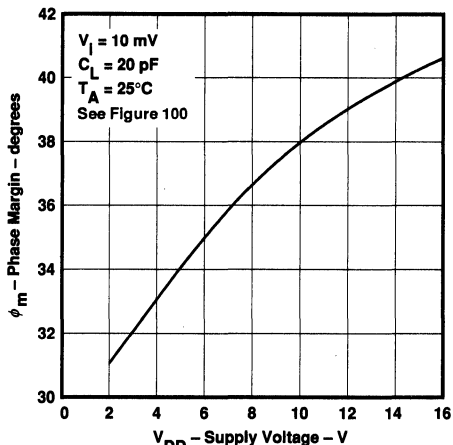


FIGURE 94

PHASE MARGIN
 vs
 FREE-AIR TEMPERATURE

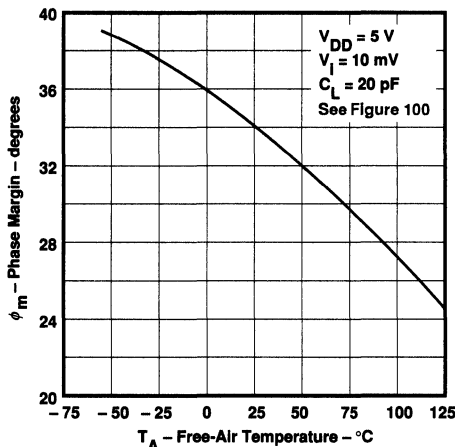


FIGURE 95

PHASE MARGIN
 vs
 CAPACITIVE LOAD

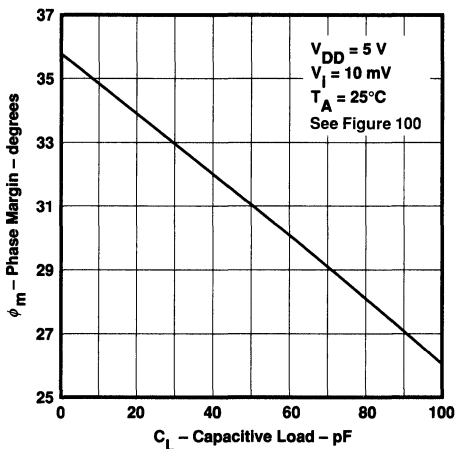


FIGURE 96

EQUIVALENT INPUT NOISE VOLTAGE
 vs
 FREQUENCY

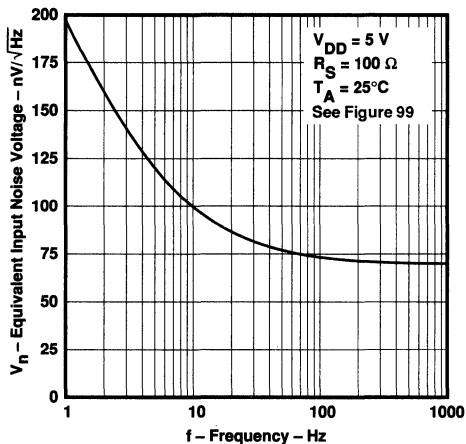


FIGURE 97

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of various devices.

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PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC271 is optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit will give the same result.



FIGURE 98. UNITY-GAIN AMPLIFIER

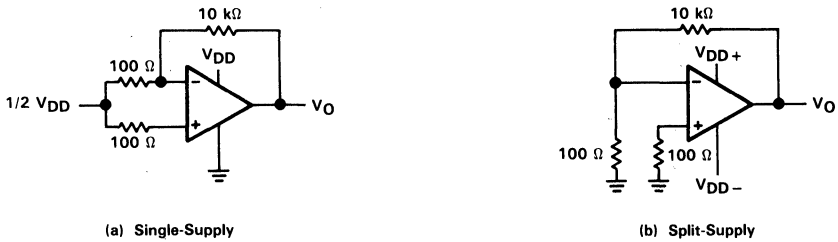


FIGURE 99. NOISE TEST CIRCUIT

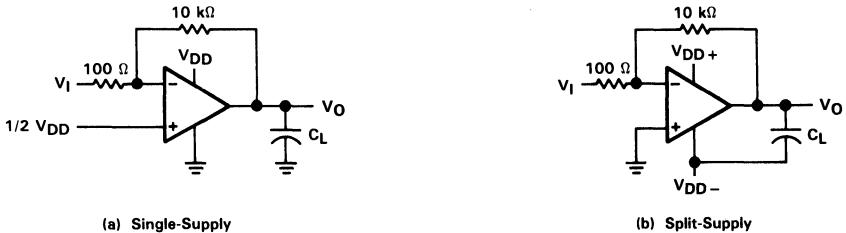


FIGURE 100. GAIN-OF-100 INVERTING AMPLIFIER

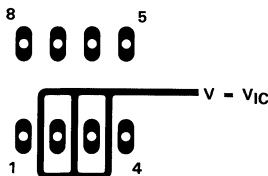
PARAMETER MEASUREMENT INFORMATION

input bias current

Because of the high input impedance of the TLC271 op amp, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 101). Leakages that would otherwise flow to the inputs will be shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the "open-socket" leakage readings from the readings obtained with a device in the test socket.

One word of caution . . . many automatic testers as well as some bench-top op amp testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an "open-socket" reading is not feasible using this method.



**FIGURE 101. ISOLATION METAL AROUND DEVICE INPUTS
(JG AND P DUAL-IN-LINE PACKAGE)**

low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to the Typical Characteristics section of this data sheet.

input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture will result in leakage and contact resistance which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

full-power response

Full-power response, the frequency above which the op amp slew rate limits the output voltage swing, is often specified two ways . . . full-linear response and full-peak response. The full-linear response is

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PARAMETER MEASUREMENT INFORMATION

generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for "significant" distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 98. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 102). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

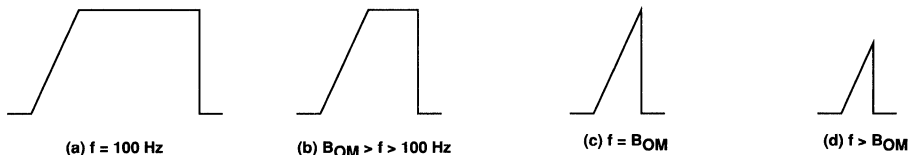


FIGURE 102. FULL-POWER-RESPONSE OUTPUT SIGNAL

test time

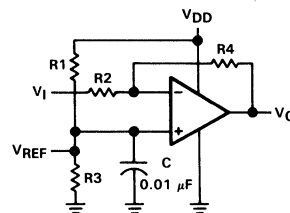
Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices, and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

TYPICAL APPLICATION DATA

single-supply operation

While the TLC271 performs well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 103). The low input bias current consumption of the TLC271 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.



$$V_{REF} = V_{DD} \frac{R_2}{R_1 + R_2}$$

$$V_O = \left(V_{REF} - V_i \right) \frac{R_4}{R_2} + V_{REF}$$

FIGURE 103. INVERTING AMPLIFIER WITH VOLTAGE REFERENCE

TYPICAL APPLICATION DATA

The TLC271 works well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 104); otherwise the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

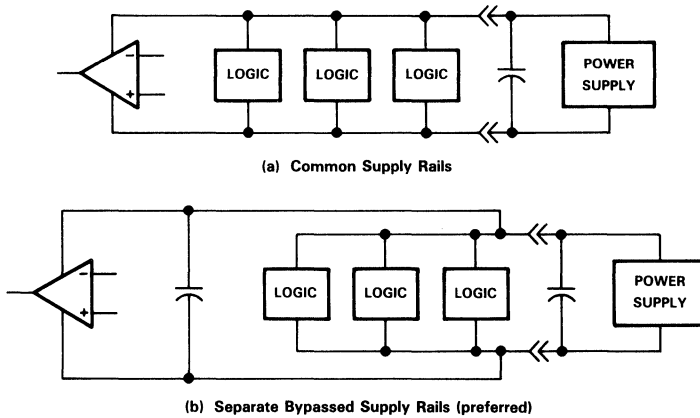


FIGURE 104. COMMON VERSUS SEPARATE SUPPLY RAILS

input offset voltage nulling

The TLC271 offers external input offset null control. Nulling of the input offset voltage may be achieved by adjusting a 25-k Ω potentiometer connected between the offset null terminals with the wiper connected as shown in Figure 105. The amount of nulling range varies with the bias selection. In the high-bias mode, the nulling range will allow the maximum offset voltage specified to be trimmed to zero. In low-bias and medium-bias modes, total nulling may not be possible.

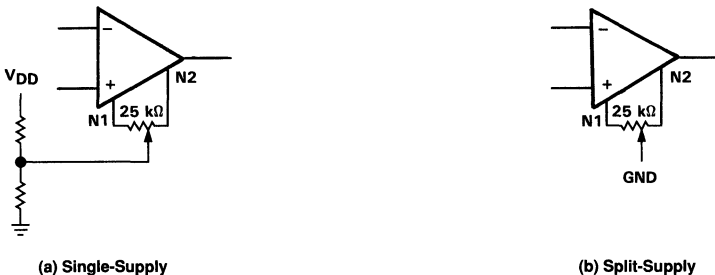
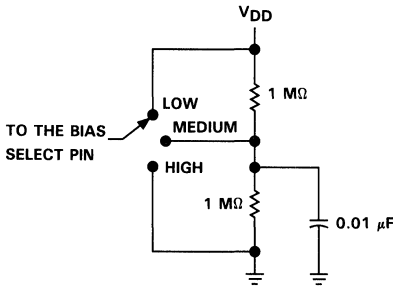


FIGURE 105. INPUT OFFSET VOLTAGE NULL CIRCUIT

TYPICAL APPLICATION DATA

bias selection

Bias selection is achieved by connecting the bias select pin to one of the three voltage levels (see Figure 106). For medium-bias applications, it is recommended that the bias select pin be connected to the mid-point between the supply rails. This is a simple procedure in split-supply applications, since this point is ground. In single-supply applications, the medium-bias mode will necessitate using a voltage divider as indicated. The use of large-value resistors in the voltage divider will reduce the current drain of the divider from the supply line. However, large-value resistors used in conjunction with a large-value capacitor will require significant time to charge up to the supply midpoint after the supply is switched on. A voltage other than the mid-point may be used if it is within the voltages specified in the following table.



BIAS MODE	BIAS SELECT VOLTAGE (Single Supply)
LOW	V_{DD}
MEDIUM	1 V to $V_{DD} - 1$ V
HIGH	GND

FIGURE 106. BIAS SELECTION FOR SINGLE-SUPPLY APPLICATIONS

input characteristics

The TLC271 is specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at $V_{DD} - 1$ V at $T_A = 25^\circ\text{C}$ and at $V_{DD} - 1.5$ V at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC271 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically 0.1 $\mu\text{V}/\text{month}$, including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC271 is well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 101 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 107).

The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.

TYPICAL APPLICATION DATA

noise performance

The noise specifications in op amp circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC271 results in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than 50 kΩ, since bipolar devices exhibit greater noise currents.

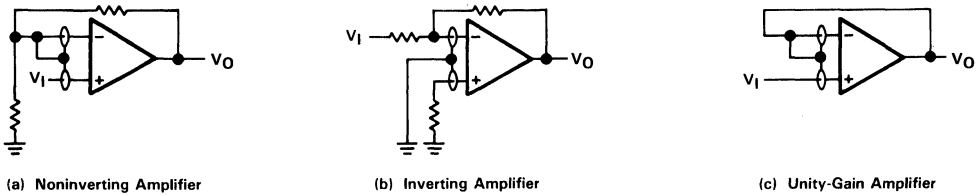


FIGURE 107. GUARD RING SCHEMES

feedback

Op amp circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 108). The value of this capacitor is optimized empirically.

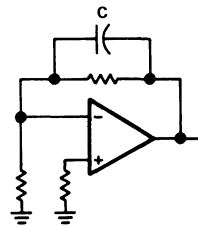


FIGURE 108. COMPENSATION FOR INPUT CAPACITANCE

electrostatic discharge protection

The TLC271 incorporates an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices, as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse biased diode.

latchup

Because CMOS devices are susceptible to latchup due to their inherent parasitic thyristors, the TLC271 inputs and output were designed to withstand –110-mA surge currents without sustaining latchup; however, techniques should be used to reduce the chance of latchup whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μF typical) located across the supply rails as close to the device as possible.

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The current path established if latchup occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latchup occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latchup occurring increases with increasing temperature and supply voltages.

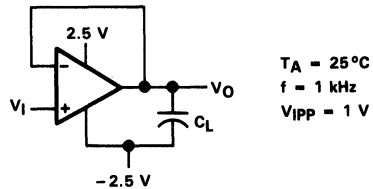
output characteristics

2

Operational Amplifiers

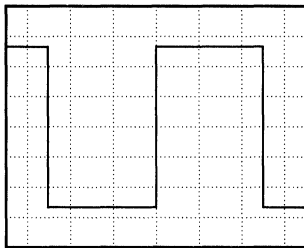
The output stage of the TLC271 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

All operating characteristics of the TLC27 were measured using a 20-pF load. The devices will drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figures 110, 111, and 112). In many cases, adding some compensation in the form of a series resistor in the feedback loop will alleviate the problem.

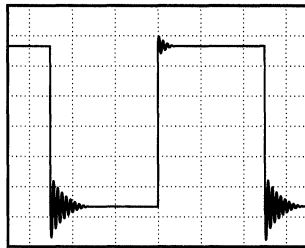


(d) Test Circuit

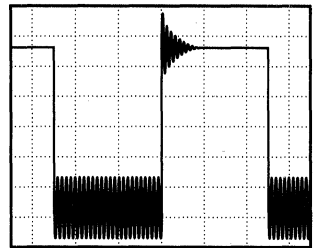
FIGURE 109. TEST CIRCUIT FOR OUTPUT CHARACTERISTICS



(a) $C_L = 20\text{ pF}$, $R_L = \text{NO LOAD}$



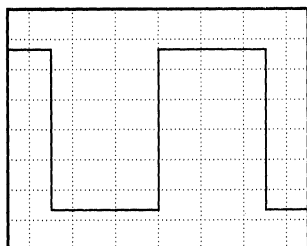
(b) $C_L = 130\text{ pF}$, $R_L = \text{NO LOAD}$



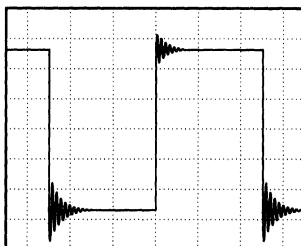
(c) $C_L = 150\text{ pF}$, $R_L = \text{NO LOAD}$

FIGURE 110. EFFECT OF CAPACITIVE LOADS IN HIGH-BIAS MODE

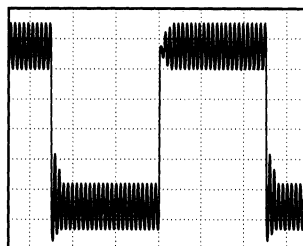
TYPICAL APPLICATION DATA



(a) $C_L = 20 \text{ pF}$, $R_L = \text{NO LOAD}$

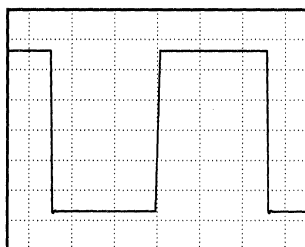


(b) $C_L = 170 \text{ pF}$, $R_L = \text{NO LOAD}$

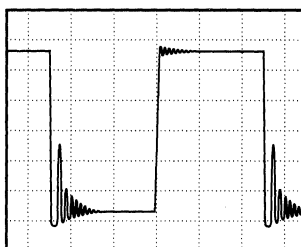


(c) $C_L = 190 \text{ pF}$, $R_L = \text{NO LOAD}$

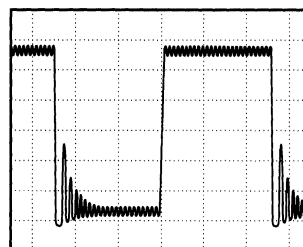
FIGURE 111. EFFECT OF CAPACITIVE LOADS IN MEDIUM-BIAS MODE



(a) $C_L = 20 \text{ pF}$, $R_L = \text{NO LOAD}$



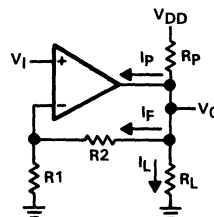
(b) $C_L = 260 \text{ pF}$, $R_L = \text{NO LOAD}$



(c) $C_L = 310 \text{ pF}$, $R_L = \text{NO LOAD}$

FIGURE 112. EFFECT OF CAPACITIVE LOADS IN LOW-BIAS MODE

Although the TLC271 possesses excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor (R_p) connected from the output to the positive supply rail (see Figure 113). There are two disadvantages to the use of this circuit. First, the NMOS pull-down transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately 60Ω and 180Ω , depending on how hard the op amp input is driven. With very low values of R_p , a voltage offset from 0 V at the output will occur. Second, pullup resistor R_p acts as a drain load to N4 and the gain of the op amp is reduced at output voltage levels where N5 is not supplying the output current.



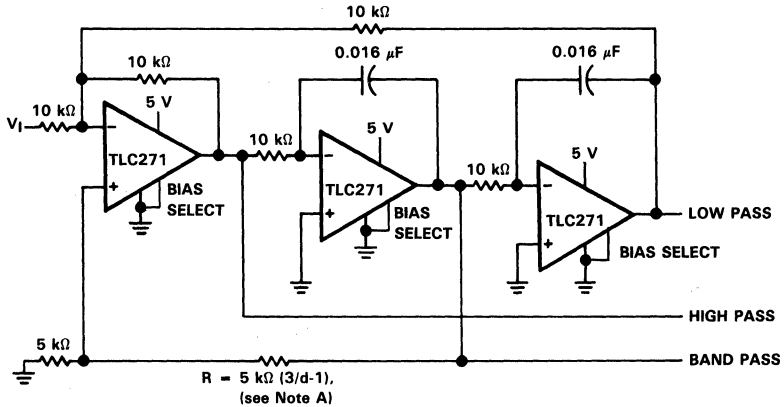
$$R_p = \frac{V_{DD} - V_O}{I_f + I_L + I_p}$$

I_p = Pullup current required by the op amp (typically $500 \mu\text{A}$)

FIGURE 113. RESISTIVE PULLUP TO INCREASE V_{OH}

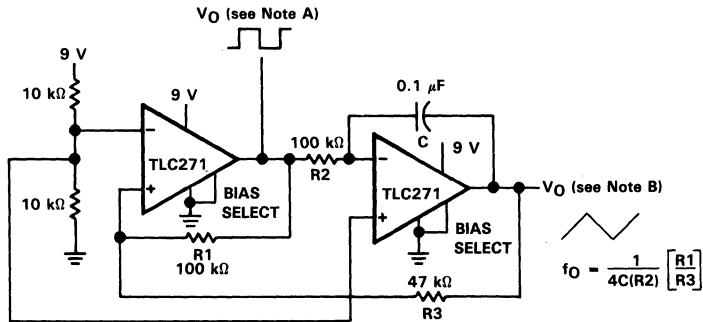
TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA (HIGH-BIAS MODE)



NOTES: A. $d = \text{damping factor}, 1/Q$
 B. Normalized to $10 \text{ k}\Omega$ and $f_c = 1 \text{ kHz}$

FIGURE 114. STATE VARIABLE FILTER



$$f_0 = \frac{1}{4C(R2)} \left[\frac{R1}{R3} \right]$$

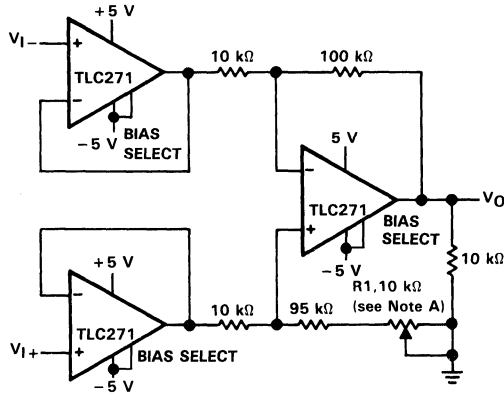
NOTES: A. $V_{OPP} = 8 \text{ V}$
 B. $V_{OPP} = 4 \text{ V}$

FIGURE 115. SINGLE-SUPPLY FUNCTION GENERATOR

2

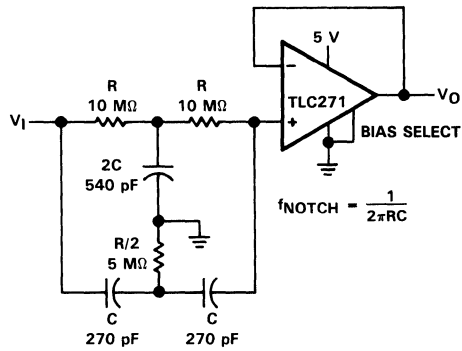
Operational Amplifiers

TYPICAL APPLICATION DATA (HIGH-BIAS MODE)



NOTE A: CMRR adjustment (must be noninductive).

FIGURE 116. LOW-POWER INSTRUMENTATION AMPLIFIER

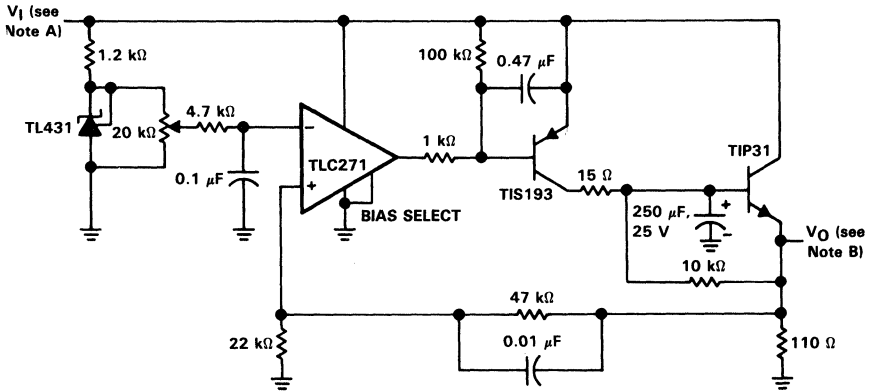


$$f_{\text{NOTCH}} = \frac{1}{2\pi RC}$$

FIGURE 117. SINGLE-SUPPLY TWIN-T NOTCH FILTER

TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA (HIGH-BIAS MODE)



NOTES: A. $V_I = 3.5$ to 15 V
 B. $V_O = 2.0$ V, 0 to 1 A

FIGURE 118. LOGIC-ARRAY POWER SUPPLY

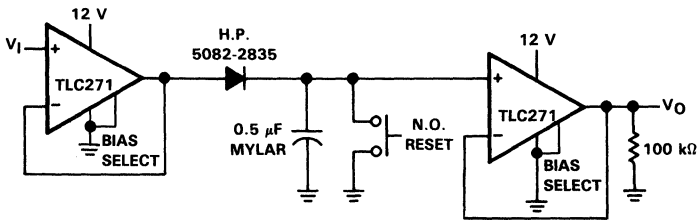
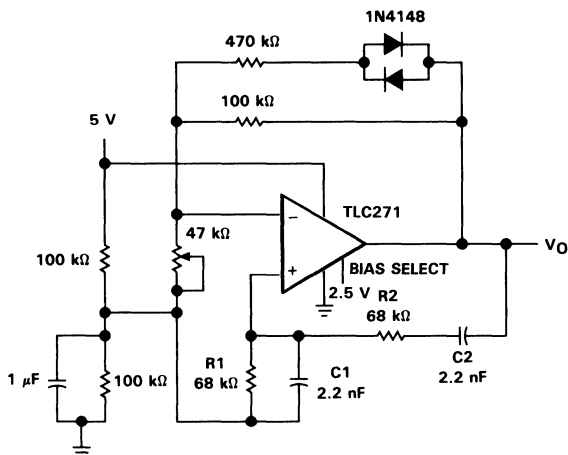


FIGURE 119. POSITIVE-PEAK DETECTOR

TYPICAL APPLICATION DATA (MEDIUM-BIAS MODE)



NOTES: $V_{OPP} \approx 2\text{ V}$

$$f_0 = \frac{1}{2\pi \sqrt{R1R2C1C2}}$$

FIGURE 120. WIEN OSCILLATOR

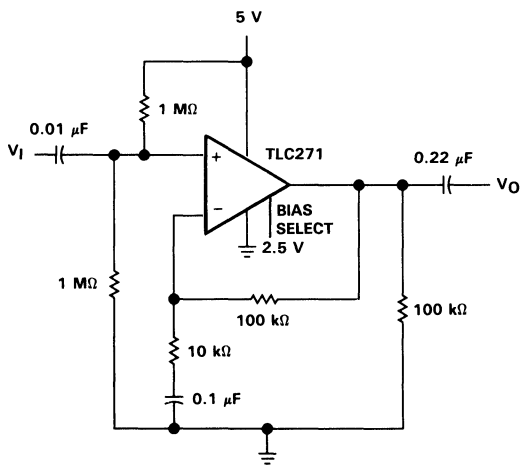
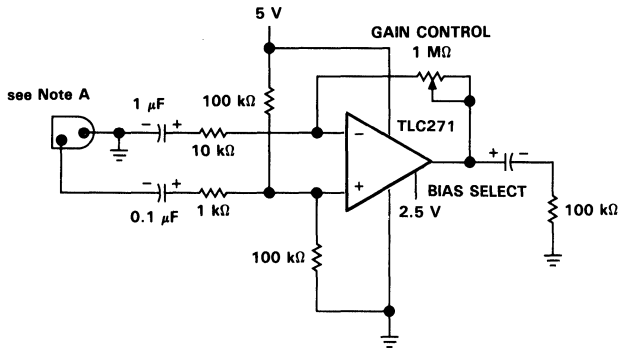


FIGURE 121. SINGLE-SUPPLY AC AMPLIFIER

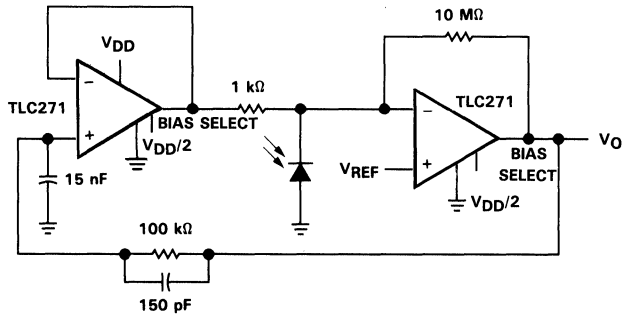
TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA (MEDIUM-BIAS MODE)



NOTE A.: Low to medium impedance dynamic mike

FIGURE 122. MICROPHONE PREAMPLIFIER

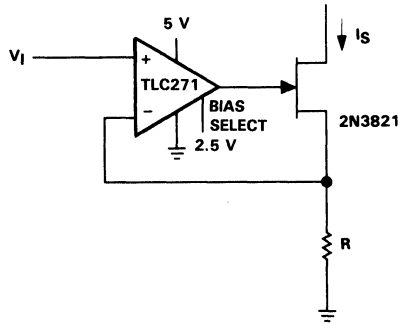


NOTES: $V_{DD} = 4 \text{ V to } 15 \text{ V}$
 $V_{REF} = 0 \text{ V to } V_{DD} - 2 \text{ V}$

FIGURE 123. PHOTO DIODE AMPLIFIER WITH AMBIENT LIGHT REJECTION

TLC271, TLC271A, TLC271B
**LinCMOS™ PROGRAMMABLE LOW-POWER
 OPERATIONAL AMPLIFIERS**

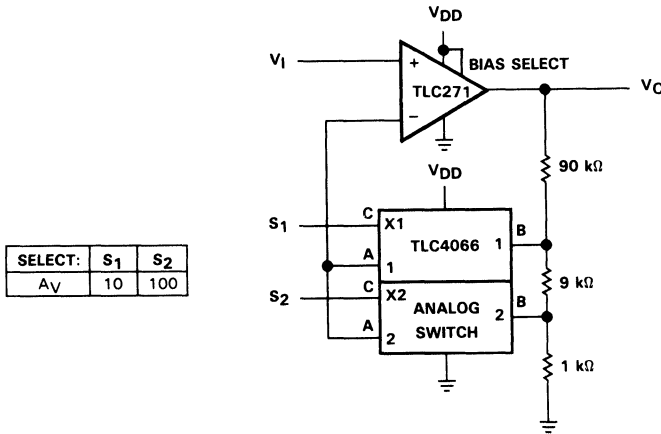
TYPICAL APPLICATION DATA (MEDIUM-BIAS MODE)



NOTES: $V_I = 0 \text{ V to } 3 \text{ V}$
 $I_S = \frac{V_I}{R}$

FIGURE 124. PRECISION LOW-CURRENT SINK

TYPICAL APPLICATION DATA (LOW-BIAS MODE)



NOTE: $V_{DD} = 5 \text{ V to } 12 \text{ V}$

FIGURE 125. AMPLIFIER WITH DIGITAL GAIN SELECTION

TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA (LOW-BIAS MODE)

2

Operational Amplifiers

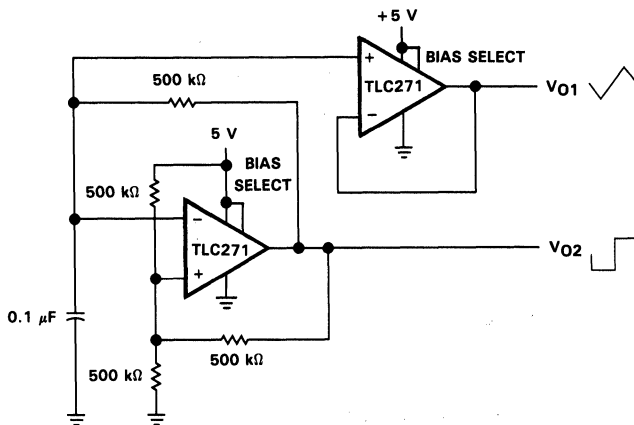
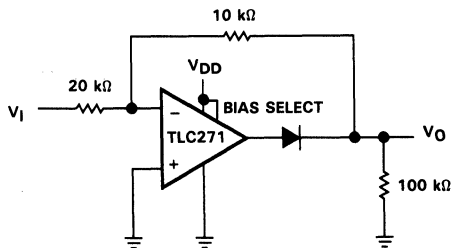


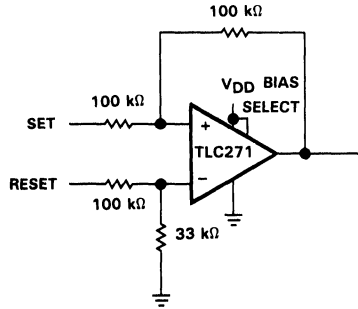
FIGURE 126. MULTIVIBRATOR



NOTE: $V_{DD} = 5 \text{ V to } 16 \text{ V}$

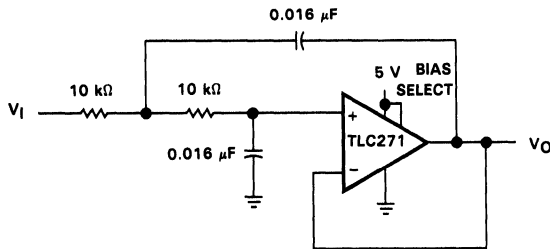
FIGURE 127. FULL-WAVE RECTIFIER

TYPICAL APPLICATION DATA (LOW-BIAS MODE)



NOTE: $V_{DD} = 5 \text{ V to } 16 \text{ V}$

FIGURE 128. SET/RESET FLIP-FLOP



NOTE: Normalized to $F_C = 1 \text{ kHz}$ and $R_L = 10 \text{ k}\Omega$

FIGURE 129. TWO-POLE LOW-PASS BUTTERWORTH FILTER

2

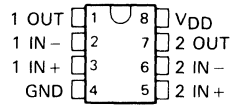
Operational Amplifiers

TLC272, TLC272A, TLC272B, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

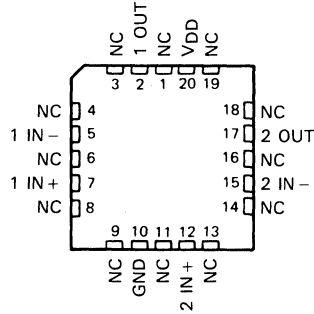
D3138, OCTOBER 1987—REVISED MARCH 1989

- **Trimmed Offset Voltage:**
TLC277 . . . 500 μV Max at 25 °C,
V_{DD} = 5 V
- **Input Offset Voltage Drift . . . Typically**
0.1 $\mu\text{V}/\text{Month}$, Including the First 30 Days
- **Wide Range of Supply Voltages over Specified Temperature Range:**
–55°C to 125°C . . . 4 V to 16 V
–40°C to 85°C . . . 4 V to 16 V
0°C to 70°C . . . 3 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix types)**
- **Low Noise . . . Typically 25 nV/ $\sqrt{\text{Hz}}$**
at f = 1 kHz
- **Output Voltage Range Includes Negative Rail**
- **High Input Impedance . . . 10¹² Ω Typical**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available in Tape and Reel**
- **Designed-In Latchup Immunity**

D, JG, OR P PACKAGE
(TOP VIEW)



FK PACKAGE
(TOP VIEW)

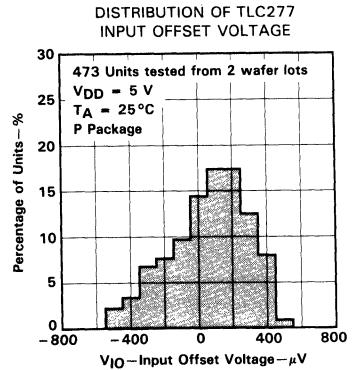


NC—No internal connection

AVAILABLE OPTIONS

T _A	V _{IO} max at 25 °C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	500 μV	TLC277CD	—	TLC277CJG	TLC277CP
	2 mV	TLC272BCD	—	TLC272BCJG	TLC272BCP
	5 mV	TLC272ACD	—	TLC272ACJG	TLC272ACP
–40°C to 85°C	10 mV	TLC272CD	—	TLC272CJG	TLC272CP
	500 μV	TLC277ID	—	TLC277IJG	TLC277IP
	2 mV	TLC272BID	—	TLC272BIJG	TLC272BIP
–55°C to 125°C	5 mV	TLC272AID	—	TLC272AIJG	TLC272AIP
	10 mV	TLC272ID	—	TLC272IJG	TLC272IP
–55°C to 125°C	500 μV	—	TLC277MFK	TLC277MJG	—
	10 mV	—	TLC272MFK	TLC272MJG	—

The D package is available in tape and reel. Add R suffix to the device type (e.g., TLC277CDR).



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Operational Amplifiers

TLC272, TLC272A, TLC272B, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

description

The TLC272 and TLC277 dual operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, low noise, and speeds approaching that of general-purpose BIFET devices.

These devices use Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, and high slew rates make these cost-effective devices ideal for applications which have previously been reserved for BIFET and NFET products. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC272 (10 mV) to the high-precision TLC277 (500 μ V). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC272 and TLC277. The devices also exhibit low voltage single-supply operation, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip carrier versions for high-density system applications.

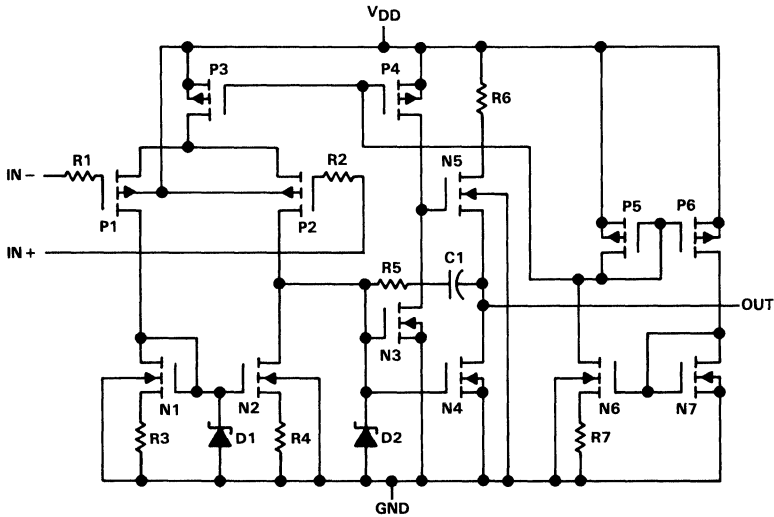
The device inputs and outputs are designed to withstand – 100-mA surge currents without sustaining latchup.

The TLC272 and TLC277 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

M-suffix devices are characterized for operation over the full military temperature range of –55°C to 125°C. I-suffix devices are characterized for operation from –40°C to 85°C, and C-suffix devices are characterized for operation from 0°C to 70°C.

TLC272, TLC272A, TLC272B, TLC277
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

equivalent schematic (each amplifier)



TLC272M, TLC277M
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC272M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
	TLC277M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	200	500	μV	
				Full range		3750		
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 125°C	2.1		$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				125°C	1.4	15	nA	
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				125°C	9	35	nA	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	0 to 4	-0.3 to 4.2	V	
				Full range	0 to 3.5		V	
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 10\text{ k}\Omega$	25°C	3.2	3.8	V	
				-55°C	3	3.8		
				125°C	3	3.8		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				-55°C	0	50		
				125°C	0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$	$R_L = 10\text{ k}\Omega$	25°C	5	23	V/mV	
				-55°C	3.5	35		
				125°C	3.5	16		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	80	dB	
				-55°C	60	81		
				125°C	60	84		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	65	95	dB	
				-55°C	60	90		
				125°C	60	97		
I_{DD}	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$, No load	$V_{IC} = 2.5\text{ V}$	25°C	1.4	3.2	mA	
				-55°C	2	5		
				125°C	1	2.2		

† Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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Operational Amplifiers

TLC272M, TLC277M
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC272M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
	TLC277M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	250	800	μV	
				Full range		4300		
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 125°C		2.2	$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				125°C	1.8	15	nA	
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				125°C	10	35	nA	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	0 to 9	-0.3 to 9.2	V	
				Full range	0 to 8.5		V	
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 10\text{ k}\Omega$	25°C	8	8.5	V	
				-55°C	7.8	8.5		
				125°C	7.8	8.4		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C		0 to 50	mV	
				-55°C		0 to 50		
				125°C		0 to 50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 10\text{ k}\Omega$	25°C	10	36	V/mV	
				-55°C	7	50		
				125°C	7	27		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	85	dB	
				-55°C	60	87		
				125°C	60	86		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	65	95	dB	
				-55°C	60	90		
				125°C	60	97		
I_{DD}	Supply current (two amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$	25°C	1.9	4	mA	
				-55°C		3 to 6		
				125°C	1.3	2.8		

† Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
 5. This range also applies to each input individually.

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Operational Amplifiers

TLC2721, TLC272AI, TLC272BI, TLC2771
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, V_{DD} = 5 V (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T _A [†]	MIN	TYP	MAX	UNIT
V _{IO}	Input offset voltage	TLC272I	V _O = 1.4 V, R _S = 50 Ω,	V _{IC} = 0 V, R _L = 10 kΩ	25°C	1.1	10	mV
				Full range		13		
	TLC272AI	V _O = 1.4 V, R _S = 50 Ω,	V _{IC} = 0 V, R _L = 10 kΩ	25°C	0.9	5	μV	
				Full range		7		
	TLC272BI	V _O = 1.4 V, R _S = 50 Ω,	V _{IC} = 0 V, R _L = 10 kΩ	25°C	230	2000	μV	
				Full range		3500		
	TLC277I	V _O = 1.4 V, R _S = 50 Ω,	V _{IC} = 0 V, R _L = 10 kΩ	25°C	200	500	μV	
				Full range		2000		
α _{VIO}	Average temperature coefficient of input offset voltage			25°C to 85°C	1.8		μV/°C	
I _{IO}	Input offset current (see Note 4)	V _O = 2.5 V,	V _{IC} = 2.5 V	25°C	0.1		pA	
I _{IB}	Input bias current (see Note 4)	V _O = 2.5 V,	V _{IC} = 2.5 V	85°C	24	1000	pA	
				25°C	0.6			
V _{ICR}	Common-mode input voltage range (see Note 5)			85°C	200	2000	pA	
				25°C	-0.2 to 4	-0.3 to 4.2	V	
V _{OH}	High-level output voltage	V _{ID} = 100 mV,	R _L = 10 kΩ	25°C	3.2	3.8	V	
				-40°C	3	3.8		
V _{OL}	Low-level output voltage	V _{ID} = -100 mV,	I _{OL} = 0	85°C	3	3.8	V	
				25°C	0	50	mV	
V _{OL}	Low-level output voltage	V _{ID} = -100 mV,	I _{OL} = 0	-40°C	0	50	mV	
				85°C	0	50		
A _{VD}	Large-signal differential voltage amplification	V _O = 0.25 V to 2 V,	R _L = 10 kΩ	25°C	5	23	V/mV	
				-40°C	3.5	32		
				85°C	3.5	19		
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICR} min		25°C	65	80	dB	
				-40°C	60	81		
				85°C	60	86		
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V,	V _O = 1.4 V	25°C	65	95	dB	
				-40°C	60	92		
				85°C	60	96		
I _{DD}	Supply current (two amplifiers)	V _O = 2.5 V, No load	V _{IC} = 2.5 V,	25°C	1.4	3.2	mA	
				-40°C	1.9	4.4		
				85°C	1.1	2.4		

† Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

TLC272I, TLC272AI, TLC272BI, TLC277I

LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC272I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
				Full range		13		
		TLC272AI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	0.9	5	mV
				Full range		7		
		TLC272BI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	290	2000	μV
				Full range		3500		
		TLC277I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	250	800	μV
				Full range		2900		
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 85°C		2		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C	0.1			pA
				85°C	26	1000		
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C	0.7			pA
				85°C	220	2000		
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2		V
				Full range	-0.2 to 8.5			
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$,	$R_L = 10\text{ k}\Omega$	25°C	8	8.5		V
				-40°C	7.8	8.5		
				85°C	7.8	8.5		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$,	$I_{OL} = 0$	25°C		0	50	mV
				-40°C		0	50	
				85°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$,	$R_L = 10\text{ k}\Omega$	25°C	10	36		V/mV
				-40°C	7	46		
				85°C	7	31		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	85		dB
				-40°C	60	87		
				85°C	60	88		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$,	$V_O = 1.4\text{ V}$	25°C	65	95		dB
				-40°C	60	92		
				85°C	60	96		
I_{DD}	Supply current (two amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$	25°C	1.9	4		mA
				-40°C	2.8	5		
				85°C	1.5	3.2		

† Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

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Operational Amplifiers

TLC272C, TLC272AC, TLC272BC, TLC277C LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC272C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
				Full range			12	
		TLC272AC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	0.9	5	
				Full range			6.5	
	TLC272BC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	230	2000	μV	
				Full range				3000
		TLC277C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	200		500
					Full range			
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C		1.8		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$, $V_{IC} = 2.5\text{ V}$		25°C		0.1		pA
				70°C		7	300	
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$, $V_{IC} = 2.5\text{ V}$		25°C		0.6		pA
				70°C		40	600	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2		V
				Full range			-0.2 to 3.5	
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$, $R_L = 10\text{ k}\Omega$		25°C		3.2	3.8	V
				0°C		3	3.8	
				70°C		3	3.8	
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$, $I_{OL} = 0$		25°C		0	50	mV
				0°C		0	50	
				70°C		0	50	
AVD	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$, $R_L = 10\text{ k}\Omega$		25°C		5	23	V/mV
				0°C		4	27	
				70°C		4	20	
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\ min}$		25°C		65	80	dB
				0°C		60	84	
				70°C		60	85	
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$, $V_O = 1.4\text{ V}$		25°C		65	95	dB
				0°C		60	94	
				70°C		60	96	
I_{DD}	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$, No load	$V_{IC} = 2.5\text{ V}$	25°C		1.4	3.2	mA
				0°C		1.6	3.6	
				70°C		1.2	2.6	

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

TLC272C, TLC272AC, TLC272BC, TLC277C
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A †	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC272C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
				Full range		12		
		TLC272AC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	0.9	5	μV
				Full range		6.5		
TLC272BC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	290	2000	μV		
		Full range		3000				
TLC277C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	250	800	μV		
		Full range		1900				
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C		2		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.1			pA
				70°C	7	300		
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.7			pA
				70°C	50	600		
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2		V
				Full range	-0.2 to 8.5			
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 10\text{ k}\Omega$	25°C	8	8.5		V
				0°C	7.8	8.5		
				70°C	7.8	8.4		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50		mV
				0°C	0	50		
				70°C	0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 10\text{ k}\Omega$	25°C	10	36		V/mV
				0°C	7.5	42		
				70°C	7.5	32		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	85		dB
				0°C	60	88		
				70°C	60	88		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	65	95		dB
				0°C	60	94		
				70°C	60	96		
I_{DD}	Supply current (two amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$	25°C	1.9	4		mA
				0°C	2.3	4.4		
				70°C	1.6	3.4		

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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TLC272M, TLC277M
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operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25 °C		3.6		V/ μ s
				-55 °C		4.7		
				125 °C		2.3		
			$V_{Ipp} = 2.5\text{ V}$	25 °C		2.9		
				-55 °C		3.7		
				125 °C		2		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25 °C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25 °C		320		kHz
				-55 °C		400		
				125 °C		230		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25 °C		1.7		MHz
				-55 °C		2.9		
				125 °C		1.1		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25 °C		46°		
				-55 °C		49°		
				125 °C		41°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25 °C		5.3		V/ μ s
				-55 °C		7.1		
				125 °C		3.1		
			$V_{Ipp} = 5.5\text{ V}$	25 °C		4.6		
				-55 °C		6.1		
				125 °C		2.7		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25 °C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25 °C		200		kHz
				-55 °C		280		
				125 °C		110		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25 °C		2.2		MHz
				-55 °C		3.4		
				125 °C		1.6		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25 °C		49°		
				-55 °C		52°		
				125 °C		44°		

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Operational Amplifiers

TLC272I, TLC272AI, TLC272BI, TLC277I
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operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C			3.6	$\text{V}/\mu\text{s}$
				-40°C			4.5	
				85°C			2.8	
			$V_{IPP} = 2.5\text{ V}$	25°C			2.9	
				-40°C			3.5	
				85°C			2.3	
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		25	$\text{nV}/\sqrt{\text{Hz}}$	
BOM	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25°C		320	kHz	
				-40°C		380		
				85°C		250		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		1.7	MHz	
				-40°C		2.6		
				85°C		1.2		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		46°		
				-40°C		49°		
				85°C		43°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C			5.3	$\text{V}/\mu\text{s}$
				-40°C			6.8	
				85°C			4	
			$V_{IPP} = 5.5\text{ V}$	25°C			4.6	
				-40°C			5.8	
				85°C			3.5	
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		25	$\text{nV}/\sqrt{\text{Hz}}$	
BOM	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25°C		200	kHz	
				-40°C		260		
				85°C		130		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		2.2	MHz	
				-40°C		3.1		
				85°C		1.7		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		49°		
				-40°C		52°		
				85°C		46°		

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Operational Amplifiers

TLC272C, TLC272AC, TLC272BC, TLC277C
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operating characteristics, VDD = 5 V

PARAMETER		TEST CONDITIONS		T _A [†]	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	R _L = 10 kΩ, C _L = 20 pF, See Figure 1	V _{IPP} = 1 V	25 °C		3.6		V/μs
				0 °C		4		
				70 °C		3		
			V _{IPP} = 2.5 V	25 °C		2.9		
				0 °C		3.1		
				70 °C		2.5		
V _n	Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 100 Ω,	25 °C		25		nV/√Hz
B _{OM}	Maximum output swing bandwidth	V _O = V _{OH} , R _L = 10 kΩ,	C _L = 20 pF, See Figure 1	25 °C		320		kHz
				0 °C		340		
				70 °C		260		
B ₁	Unity-gain bandwidth	V _i = 10 mV, See Figure 3	C _L = 20 pF,	25 °C		1.7		MHz
				0 °C		2		
				70 °C		1.3		
φ _m	Phase margin	V _i = 10 mV, C _L = 20 pF,	f = B ₁ , See Figure 3	25 °C		46°		
				0 °C		47°		
				70 °C		43°		

operating characteristics, VDD = 10 V

PARAMETER		TEST CONDITIONS		T _A [†]	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	R _L = 10 kΩ, C _L = 20 pF, See Figure 1	V _{IPP} = 1 V	25 °C		5.3		V/μs
				0 °C		5.9		
				70 °C		4.3		
			V _{IPP} = 5.5 V	25 °C		4.6		
				0 °C		5.1		
				70 °C		3.8		
V _n	Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 100 Ω,	25 °C		25		nV/√Hz
B _{OM}	Maximum output swing bandwidth	V _O = V _{OH} , R _L = 10 kΩ,	C _L = 20 pF, See Figure 1	25 °C		200		kHz
				0 °C		220		
				70 °C		140		
B ₁	Unity-gain bandwidth	V _i = 10 mV, See Figure 3	C _L = 20 pF,	25 °C		2.2		MHz
				0 °C		2.5		
				70 °C		1.8		
φ _m	Phase margin	V _i = 10 mV, C _L = 20 pF,	f = B ₁ , See Figure 3	25 °C		49°		
				0 °C		50°		
				70 °C		46°		

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Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC272 and TLC277 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit will give the same result.

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Operational Amplifiers



FIGURE 1. UNITY-GAIN AMPLIFIER

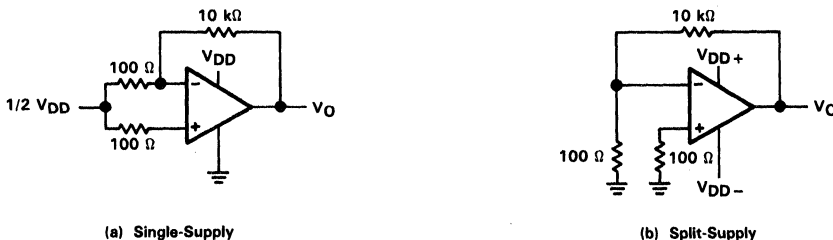


FIGURE 2. NOISE TEST CIRCUIT

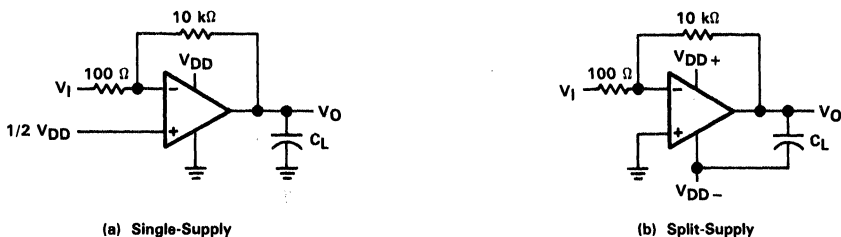


FIGURE 3. GAIN-OF-100 INVERTING AMPLIFIER

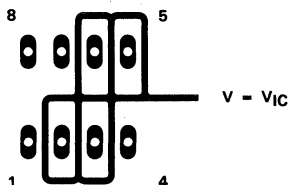
PARAMETER MEASUREMENT INFORMATION

input bias current

Because of the high input impedance of the TLC272 and TLC277 op amps, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs will be shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the "open-socket" leakage readings from the readings obtained with a device in the test socket.

One word of caution . . . many automatic testers as well as some bench-top op amp testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an "open-socket" reading is not feasible using this method.



**FIGURE 4. ISOLATION METAL AROUND DEVICE INPUTS
 (JG AND P DUAL-IN-LINE-PACKAGE)**

low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture will result in leakage and contact resistance which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

PARAMETER MEASUREMENT INFORMATION

full-power response

Full-power response, the frequency above which the op amp slew rate limits the output voltage swing, is often specified two ways . . . full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for "significant" distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

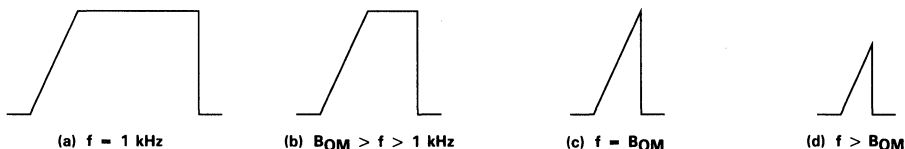


FIGURE 5. FULL-POWER-RESPONSE OUTPUT SIGNAL

test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

TLC272, TLC272A, TLC272B, TLC277
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC272
 INPUT OFFSET VOLTAGE

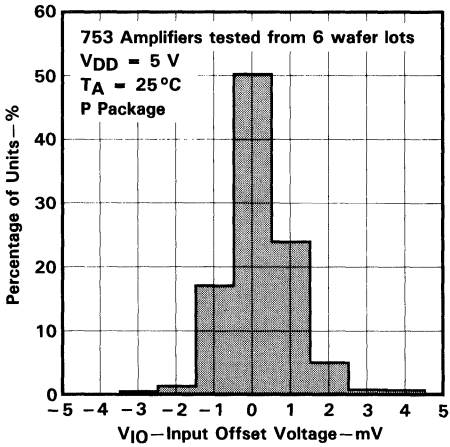


FIGURE 6

DISTRIBUTION OF TLC272
 INPUT OFFSET VOLTAGE

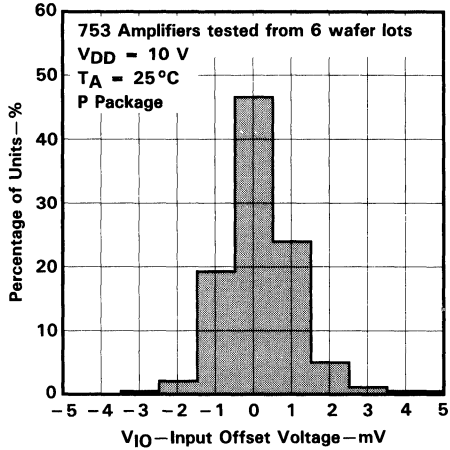


FIGURE 7

DISTRIBUTION OF TLC272 AND TLC277
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT

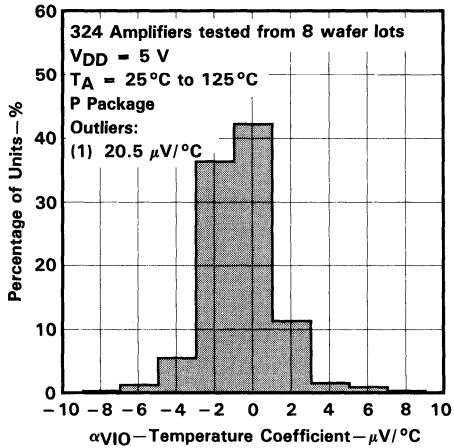


FIGURE 8

DISTRIBUTION OF TLC272 AND TLC277
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT

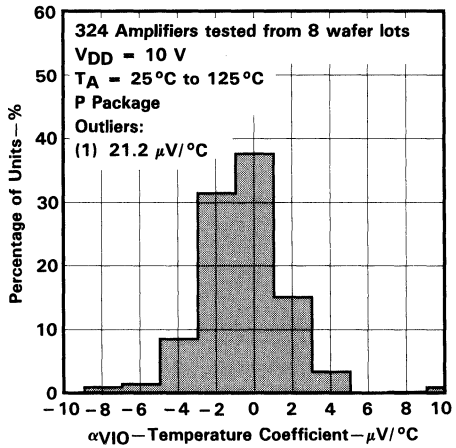


FIGURE 9

TLC272, TLC272A, TLC272B, TLC277
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

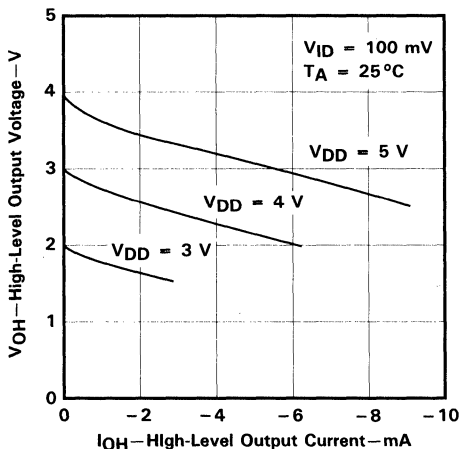


FIGURE 10

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

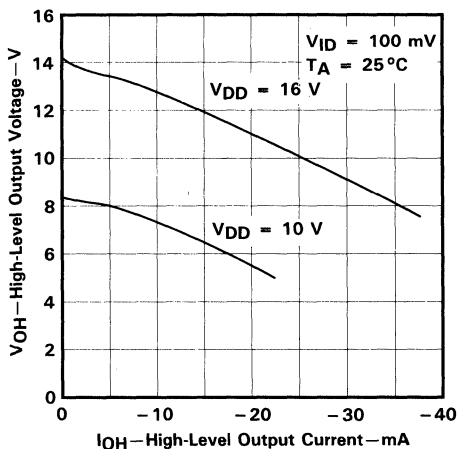


FIGURE 11

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 SUPPLY VOLTAGE

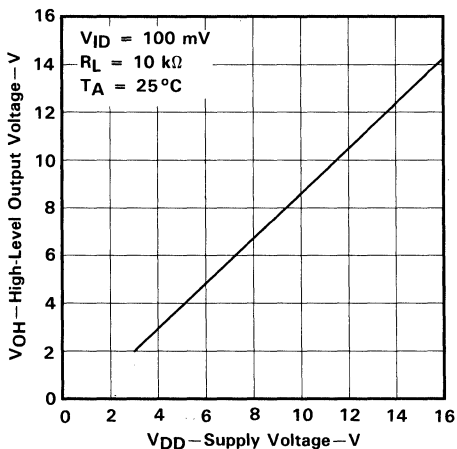


FIGURE 12

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

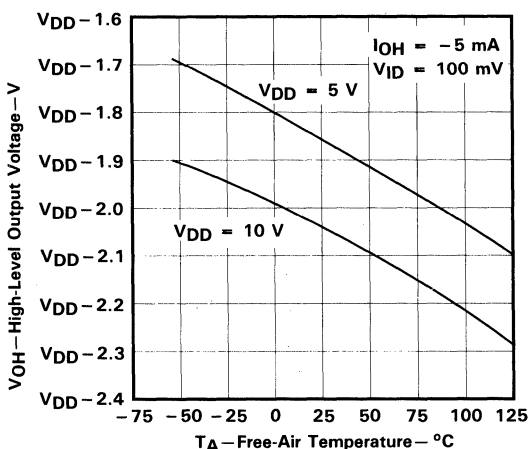


FIGURE 13

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

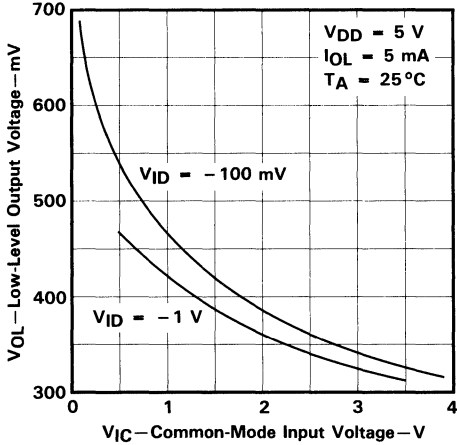


FIGURE 14

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

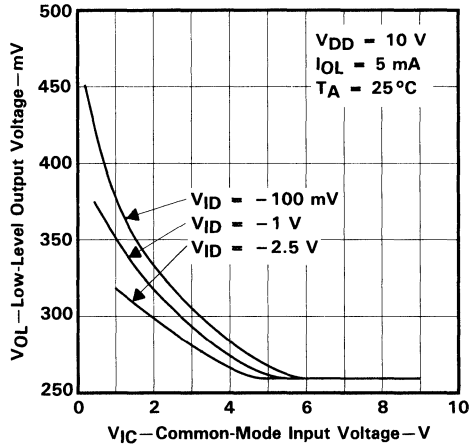


FIGURE 15

LOW-LEVEL OUTPUT VOLTAGE
 vs
 DIFFERENTIAL INPUT VOLTAGE

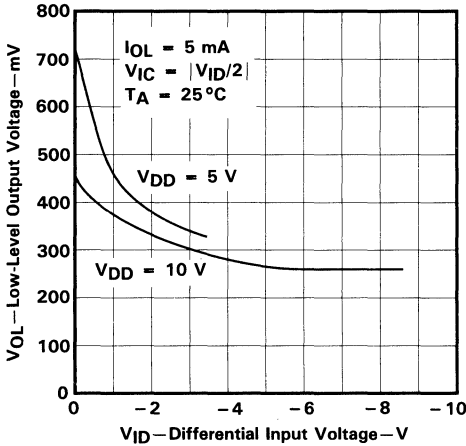


FIGURE 16

LOW-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

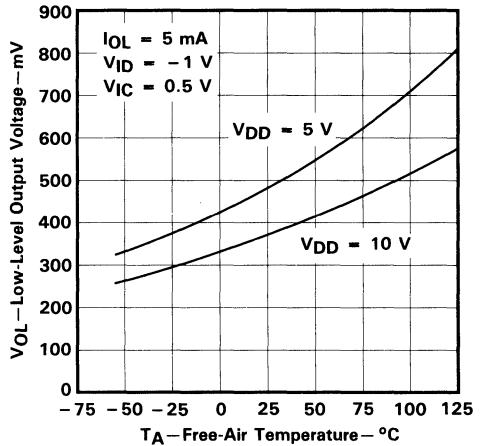


FIGURE 17

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC272, TLC272A, TLC272B, TLC277
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

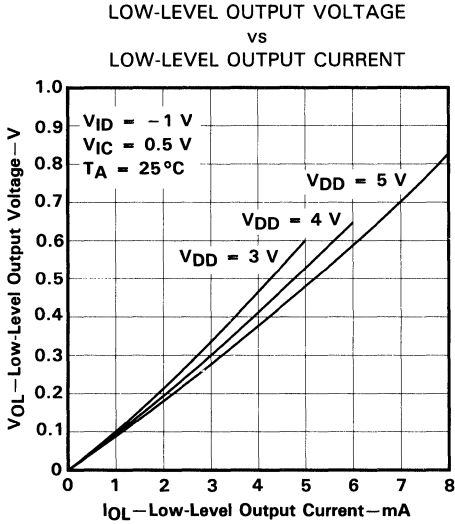


FIGURE 18

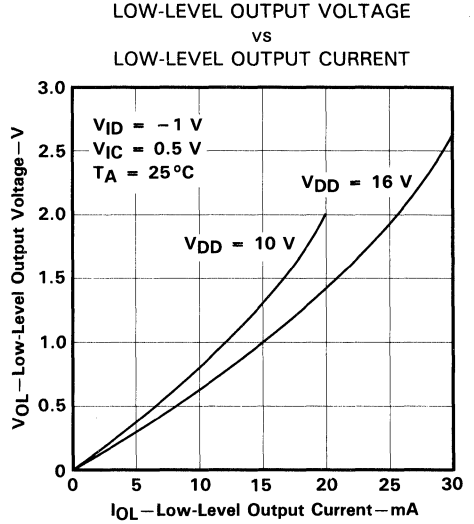


FIGURE 19

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 SUPPLY VOLTAGE

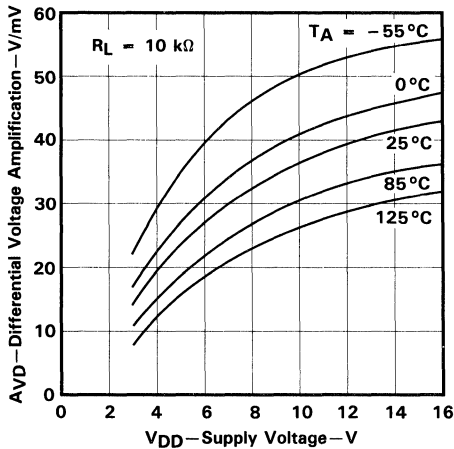


FIGURE 20

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 FREE-AIR TEMPERATURE

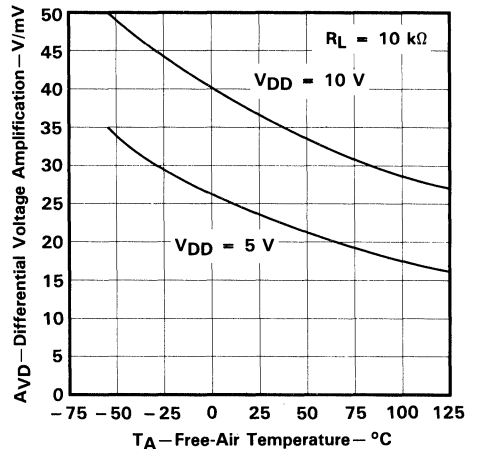


FIGURE 21

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC272, TLC272A, TLC272B, TLC277
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

INPUT BIAS CURRENT AND INPUT OFFSET CURRENT
 vs
 FREE-AIR TEMPERATURE

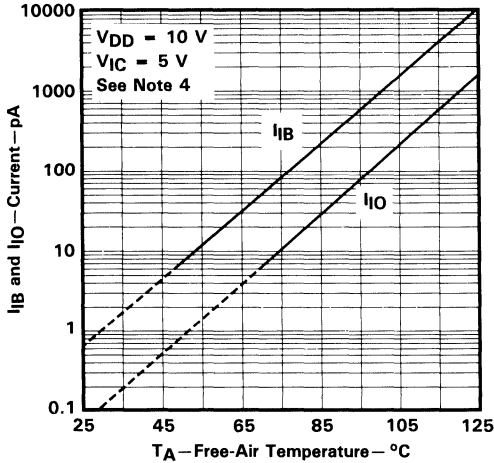


FIGURE 22

COMMON-MODE
 INPUT VOLTAGE POSITIVE LIMIT
 vs
 SUPPLY VOLTAGE

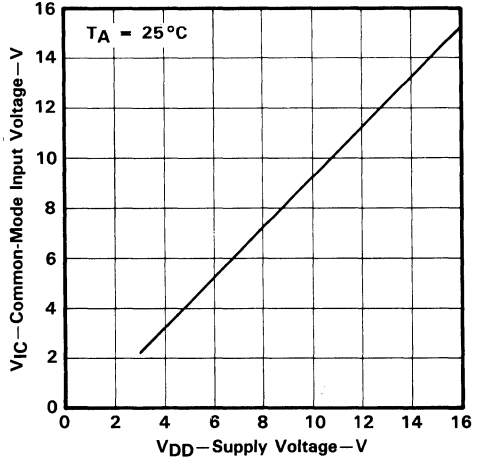


FIGURE 23

SUPPLY CURRENT
 vs
 SUPPLY VOLTAGE

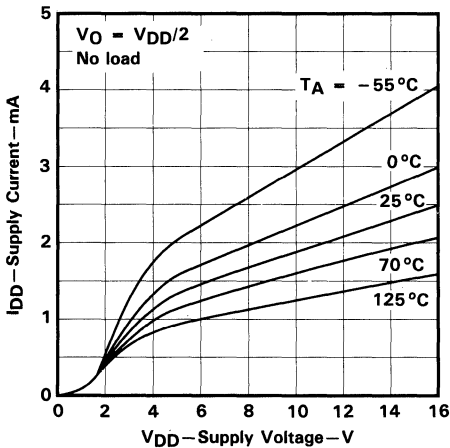


FIGURE 24

SUPPLY CURRENT
 vs
 FREE-AIR TEMPERATURE

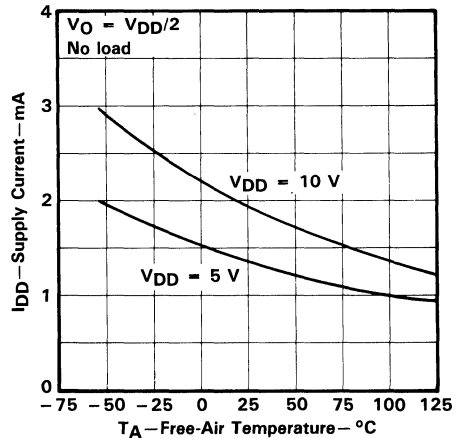


FIGURE 25

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.
 NOTE 4: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

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 Operational Amplifiers

TLC272, TLC272A, TLC272B, TLC277
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

SLEW RATE
 vs
SUPPLY VOLTAGE

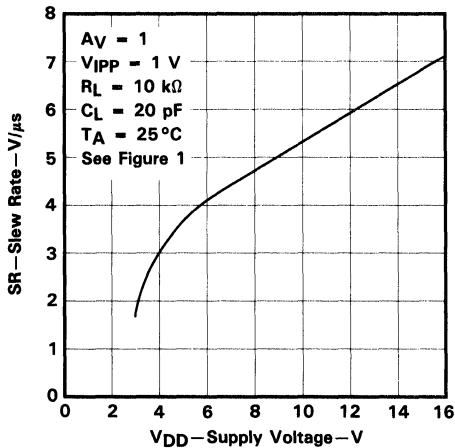


FIGURE 26

SLEW RATE
 vs
FREE-AIR TEMPERATURE

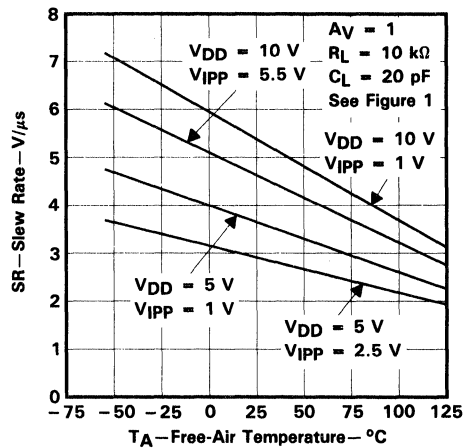


FIGURE 27

NORMALIZED SLEW RATE
 vs
FREE-AIR TEMPERATURE

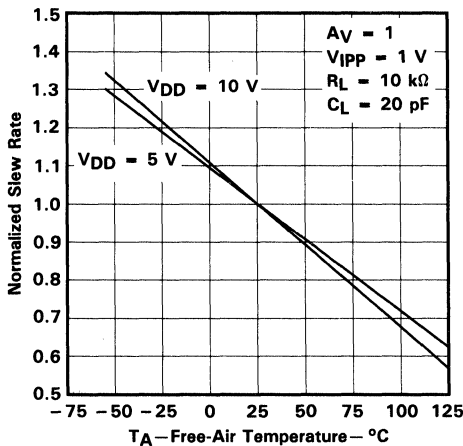


FIGURE 28

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 vs
FREQUENCY

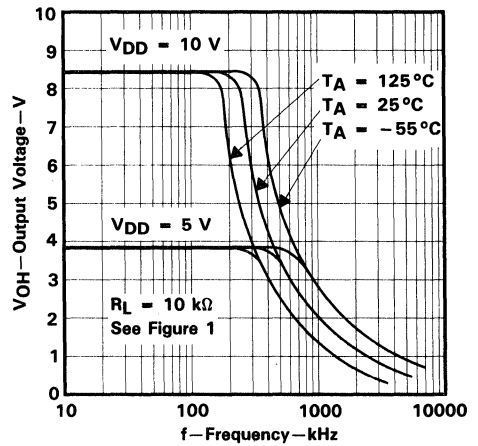


FIGURE 29

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

UNITY-GAIN BANDWIDTH
 vs
 FREE-AIR TEMPERATURE

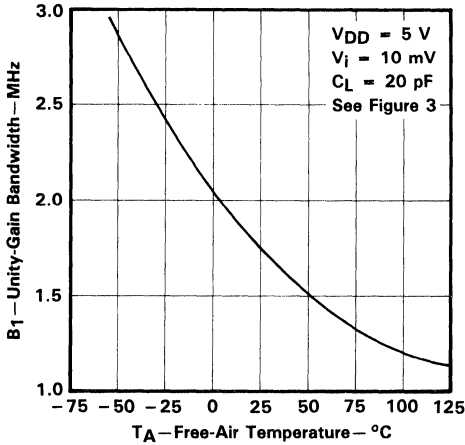


FIGURE 30

UNITY-GAIN BANDWIDTH
 vs
 SUPPLY VOLTAGE

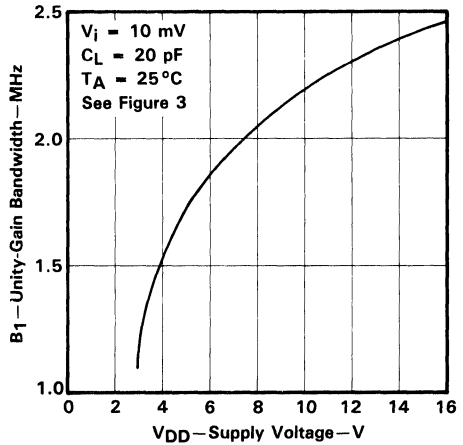


FIGURE 31

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

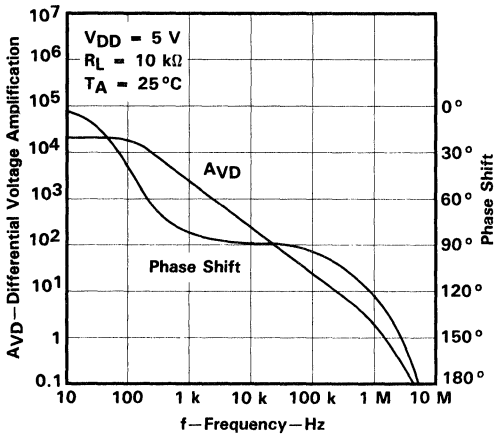


FIGURE 32

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

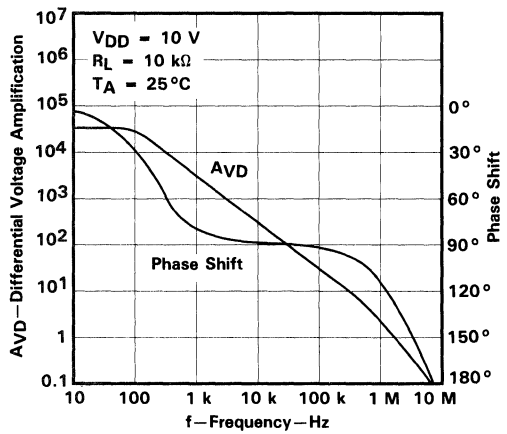


FIGURE 33

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

PHASE MARGIN
 vs
 SUPPLY VOLTAGE

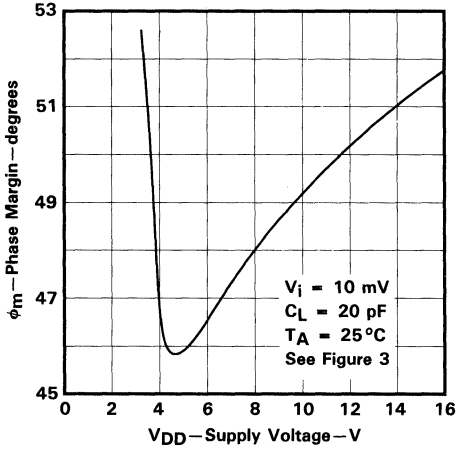


FIGURE 34

PHASE MARGIN
 vs
 FREE-AIR TEMPERATURE

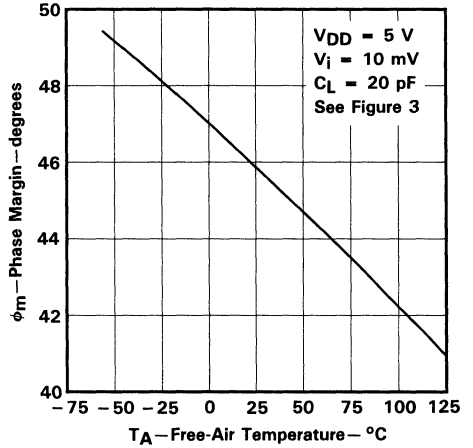


FIGURE 35

PHASE MARGIN
 vs
 CAPACITIVE LOAD

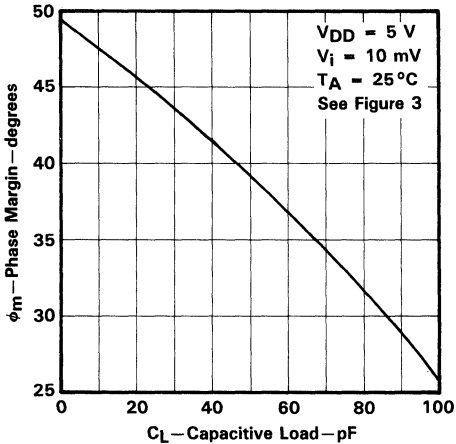


FIGURE 36

EQUIVALENT INPUT NOISE VOLTAGE
 vs
 FREQUENCY

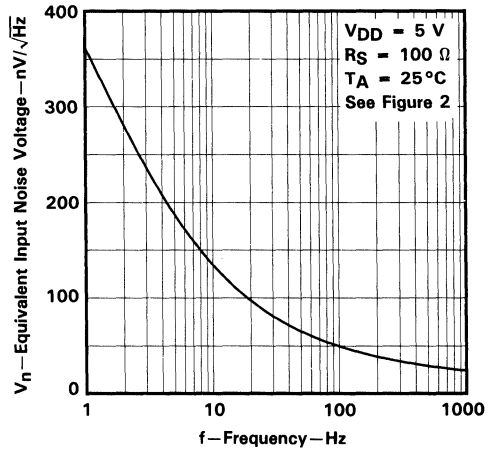


FIGURE 37

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL APPLICATION DATA

single-supply operation

While the TLC272 and TLC277 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC272 and TLC277 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC272 and TLC277 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

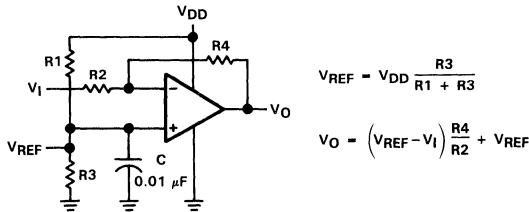


FIGURE 38. INVERTING AMPLIFIER WITH VOLTAGE REFERENCE

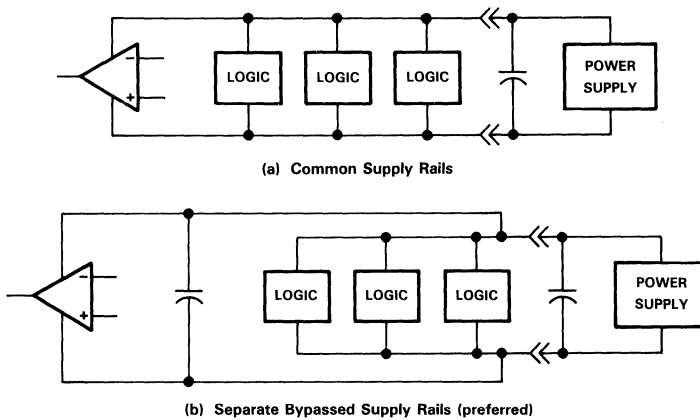


FIGURE 39. COMMON VS SEPARATE SUPPLY RAILS

TYPICAL APPLICATION DATA

input characteristics

The TLC272 and TLC277 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at $V_{DD} - 1\text{ V}$ at $T_A = 25^\circ\text{C}$ and at $V_{DD} - 1.5\text{ V}$ at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC272 and TLC277 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically $0.1\ \mu\text{V}/\text{month}$, including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC272 and TLC277 are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.

noise performance

The noise specifications in op amp circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC272 and TLC277 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than $50\ \text{k}\Omega$, since bipolar devices exhibit greater noise currents.

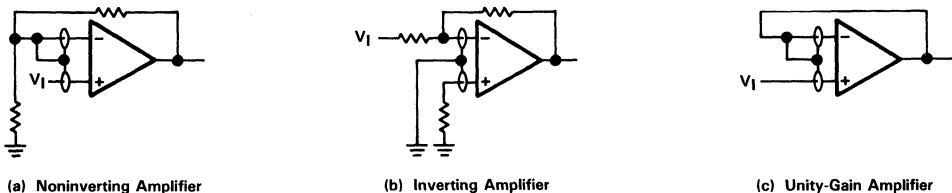


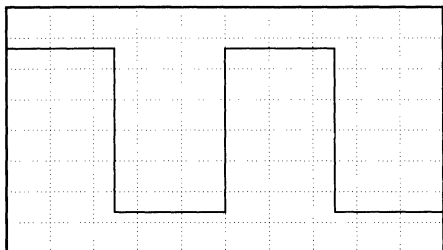
FIGURE 40. GUARD-RING SCHEMES

output characteristics

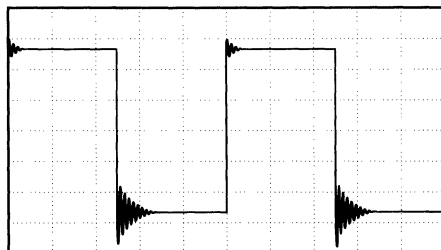
The output stage of the TLC272 and TLC277 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

All operating characteristics of the TLC272 and TLC277 were measured using a 20-pF load. The devices will drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance will alleviate the problem.

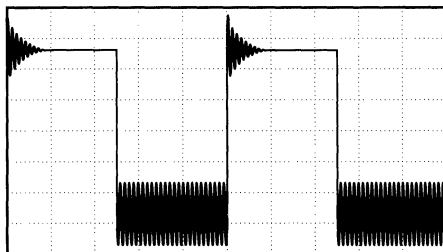
TYPICAL APPLICATION DATA



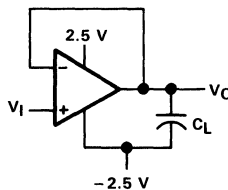
(a) $C_L = 20 \text{ pF}$, $R_L = \text{No load}$



(b) $C_L = 130 \text{ pF}$, $R_L = \text{No load}$



(c) $C_L = 150 \text{ pF}$, $R_L = \text{No load}$



$T_A = 25^\circ\text{C}$
 $f = 1 \text{ kHz}$
 $V_{IPP} = 1 \text{ V}$

(d) Test Circuit

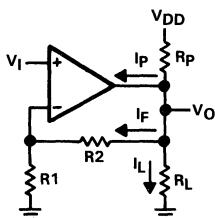
FIGURE 41. EFFECT OF CAPACITIVE LOADS AND TEST CIRCUIT

Although the TLC272 and TLC277 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor (R_p) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pull-down transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately 60Ω and 180Ω , depending on how hard the op amp input is driven. With very low values of R_p , a voltage offset from 0 V at the output will occur. Second, pullup resistor R_p acts as a drain load to N4 and the gain of the op amp is reduced at output voltage levels where N5 is not supplying the output current.

feedback

Op amp circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

TYPICAL APPLICATION DATA



$$R_p = \frac{V_{DD} - V_O}{I_F + I_L + I_p}$$

I_p = Pullup current required by the op amp (typically 500 μ A)

FIGURE 42. RESISTIVE PULLUP TO INCREASE V_{OH}

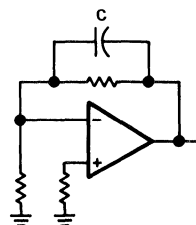


FIGURE 43. COMPENSATION FOR INPUT CAPACITANCE

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Operational Amplifiers

electrostatic discharge protection

The TLC272 and TLC277 incorporate an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

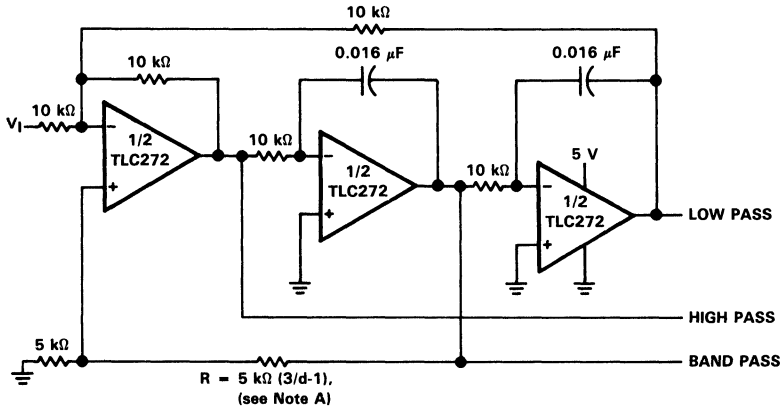
latchup

Because CMOS devices are susceptible to latchup due to their inherent parasitic thyristors, the TLC272 and TLC277 inputs and outputs were designed to withstand – 100-mA surge currents without sustaining latchup; however, techniques should be used to reduce the chance of latchup whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μ F typical) located across the supply rails as close to the device as possible.

The current path established if latchup occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latchup occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latchup occurring increases with increasing temperature and supply voltages.

TLC272, TLC272A, TLC272B, TLC277
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA



- NOTES: A. d = damping factor, $1/Q$
 B. Normalized to $10 \text{ k}\Omega$ and $f_c = 1 \text{ kHz}$

FIGURE 44. STATE VARIABLE FILTER

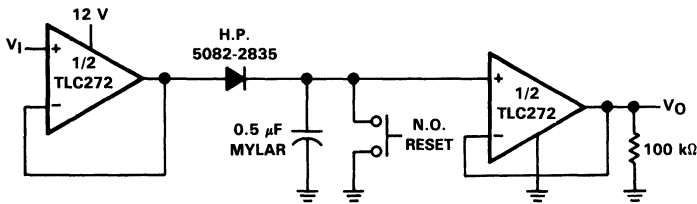
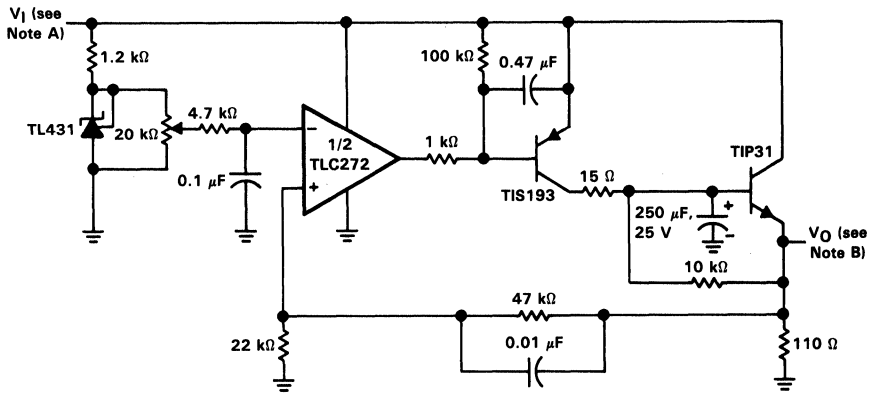


FIGURE 45. POSITIVE-PEAK DETECTOR

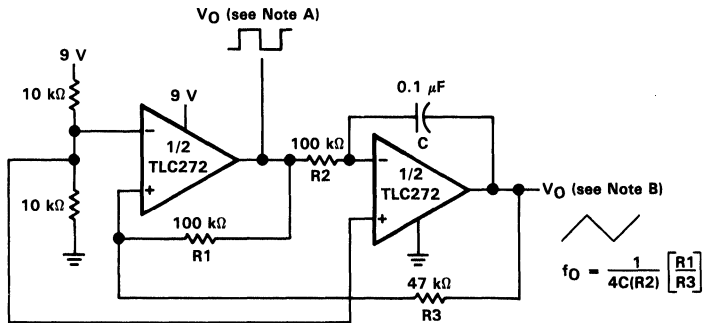
TLC272, TLC272A, TLC272B, TLC277
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA



NOTES: A. $V_I = 3.5$ to 15 V
 B. $V_O = 2.0$ V, 0 to 1 A

FIGURE 46. LOGIC ARRAY POWER SUPPLY



NOTES: A. $V_{OPP} = 8$ V
 B. $V_{OPP} = 4$ V

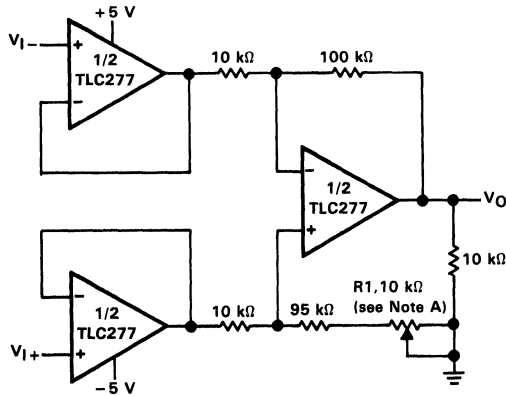
FIGURE 47. SINGLE-SUPPLY FUNCTION GENERATOR

$$f_0 = \frac{1}{4C(R_2)} \left[\frac{R_1}{R_3} \right]$$

2

Operational Amplifiers

TYPICAL APPLICATION DATA



NOTE A: CMRR adjustment (must be noninductive).

FIGURE 48. LOW-POWER INSTRUMENTATION AMPLIFIER

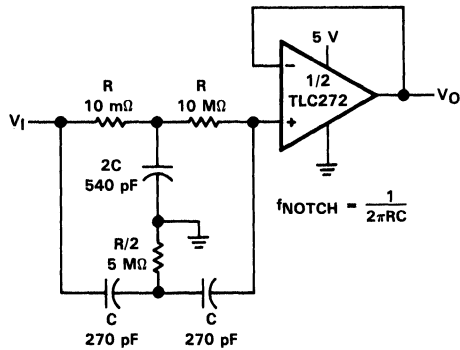


FIGURE 49. SINGLE-SUPPLY TWIN-T NOTCH FILTER

2

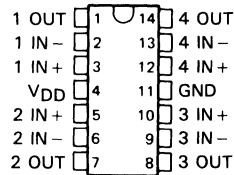
Operational Amplifiers

TLC274, TLC274A, TLC274B, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

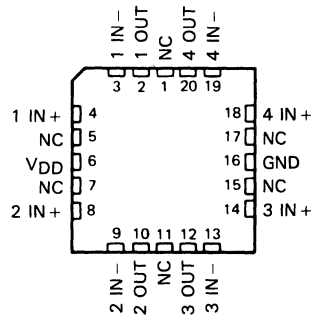
D3141, SEPTEMBER 1987—REVISED AUGUST 1988

- **Trimmed Offset Voltage:**
TLC279 . . . 900 μV Max at 25 °C,
VDD = 5 V
- **Input Offset Voltage Drift . . . Typically**
0.1 $\mu\text{V}/\text{Month}$, Including the First 30 Days
- **Wide Range of Supply Voltages over Specified Temperature Range:**
–55°C to 125°C . . . 4 V to 16 V
–40°C to 85°C . . . 4 V to 16 V
0°C to 70°C . . . 3 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix types)**
- **Low Noise . . . Typically 25 nV/ $\sqrt{\text{Hz}}$ at f = 1 kHz**
- **Output Voltage Range Includes Negative Rail**
- **High Input Impedance . . . 10¹² Ω Typical**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available in Tape and Reel**
- **Designed-In Latchup Immunity**

D, J, OR N PACKAGE
(TOP VIEW)



FK PACKAGE
(TOP VIEW)



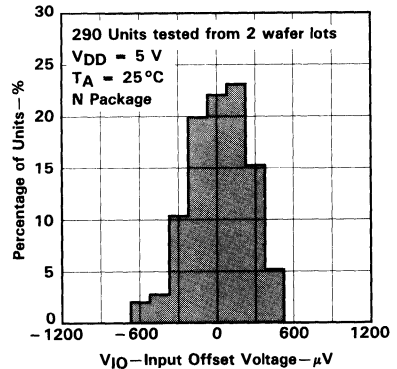
NC—No internal connection

AVAILABLE OPTIONS

TA	V _{IO} max at 25°C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	900 μV	TLC279CD	—	TLC279CJ	TLC279CN
	2 mV	TLC274BCD	—	TLC274BCJ	TLC274BCN
	5 mV	TLC274ACD	—	TLC274ACJ	TLC274ACN
	10 mV	TLC274CD	—	TLC274CJ	TLC274CN
–40°C to 85°C	900 μV	TLC279ID	—	TLC279IJ	TLC279IN
	2 mV	TLC274BID	—	TLC274BIJ	TLC274BIN
	5 mV	TLC274AID	—	TLC274AIJ	TLC274AIN
	10 mV	TLC274ID	—	TLC274IJ	TLC274IN
–55°C to 125°C	900 μV	—	TLC279MFK	TLC279MJ	—
	10 mV	—	TLC274MFK	TLC274MJ	—

The D package is available in tape and reel. Add R suffix to the device type (e.g., TLC279CDR).

DISTRIBUTION OF TLC279
INPUT OFFSET VOLTAGE



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TLC274, TLC274A, TLC274B, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

description

The TLC274 and TLC279 quad operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, low noise, and speeds approaching that of general-purpose BiFET devices.

These devices use Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, and high slew rates make these cost-effective devices ideal for applications which have previously been reserved for BiFET and NFET products. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC274 (10 mV) to the high-precision TLC279 (900 μ V). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC274 and TLC279. The devices also exhibit low voltage single-supply operation, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip carrier versions for high-density system applications.

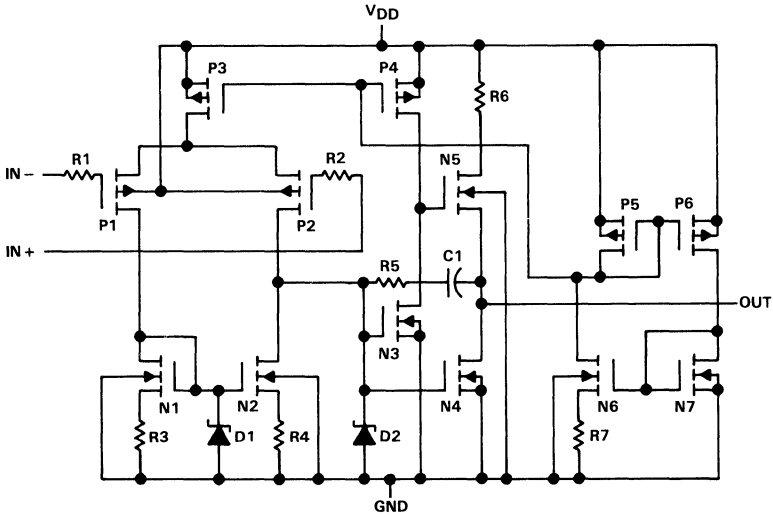
The device inputs and outputs are designed to withstand -100 -mA surge currents without sustaining latchup.

The TLC274 and TLC279 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . I-suffix devices are characterized for operation from -40°C to 85°C , and C-suffix devices are characterized for operation from 0°C to 70°C .

TLC274, TLC274A, TLC274B, TLC279
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

equivalent schematic (each amplifier)



TLC274, TLC274A, TLC274B, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

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Operational Amplifiers

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{DD} (see Note 1)	18 V
Differential input voltage (see Note 2)	$\pm V_{DD}$
Input voltage range, V_I (any input)	$-0.3 \text{ V to } V_{DD}$
Input current, I_I	$\pm 5 \text{ mA}$
Output current, I_O (each output)	$\pm 30 \text{ mA}$
Total current into V_{DD} terminal	45 mA
Total current out of ground terminal	45 mA
Duration of short-circuit current at (or below) 25 °C (see Note 3)	Unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, T_A : M-suffix	$-55 \text{ }^\circ\text{C to } 125 \text{ }^\circ\text{C}$
I-suffix	$-40 \text{ }^\circ\text{C to } 85 \text{ }^\circ\text{C}$
C-suffix	$0 \text{ }^\circ\text{C to } 70 \text{ }^\circ\text{C}$
Storage temperature range	$-65 \text{ }^\circ\text{C to } 150 \text{ }^\circ\text{C}$
Case temperature for 60 seconds: FK package	260 °C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D and N package	260 °C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300 °C

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.
 2. Differential voltages are at the noninverting input with respect to the inverting input.
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25 \text{ }^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25 \text{ }^\circ\text{C}$	$T_A = 70 \text{ }^\circ\text{C}$	$T_A = 85 \text{ }^\circ\text{C}$	$T_A = 125 \text{ }^\circ\text{C}$
	POWER RATING		POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	
FK	1375 mW	11 mW/°C	880 mW	715 mW	275 mW
J (M-suffix)	1375 mW	11 mW/°C	880 mW	715 mW	275 mW
J (C-, I-suffix)	1025 mW	8.2 mW/°C	656 mW	533 mW	
N	1575 mW	12.6 mW/°C	1008 mW	819 mW	

recommended operating conditions

	M-SUFFIX			I-SUFFIX			C-SUFFIX			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, V_{DD}	4		16	4		16	3		16	V
Common-mode input voltage, V_{IC}	$V_{DD} = 5 \text{ V}$			0		3.5	-0.2		3.5	V
	$V_{DD} = 10 \text{ V}$			0		8.5	-0.2		8.5	
Operating free-air temperature, T_A	-55		125	-40		85	0		70	°C

TLC274M, TLC279M
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC274M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $R_L = 10\text{ k}\Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25 °C	1.1	10	mV
				Full range			12	
		TLC279M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $R_L = 10\text{ k}\Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25 °C	320	900	μV
				Full range			3750	
α_{VIO}	Average temperature coefficient of input offset voltage			25 °C to 125 °C		2.1		$\mu\text{V}/^\circ\text{C}$
I_{JO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$, $V_{IC} = 2.5\text{ V}$		25 °C		0.1		pA
				125 °C		1.4	15	nA
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$, $V_{IC} = 2.5\text{ V}$		25 °C		0.6		pA
				125 °C		9	35	nA
V_{ICR}	Common-mode input voltage range (see Note 5)			25 °C	0 to 4	-0.3 to 4.2		V
				Full range	0 to 3.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$, $R_L = 10\text{ k}\Omega$		25 °C	3.2	3.8		V
				-55 °C	3	3.8		
				125 °C	3	3.8		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$, $I_{OL} = 0$		25 °C		0	50	mV
				-55 °C		0	50	
				125 °C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$, $R_L = 10\text{ k}\Omega$		25 °C	5	23		V/mV
				-55 °C	3.5	35		
				125 °C	3.5	16		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25 °C	65	80		dB
				-55 °C	60	81		
				125 °C	60	84		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$, $V_O = 1.4\text{ V}$	No load	25 °C	65	95		dB
				-55 °C	60	90		
				125 °C	60	97		
I_{DD}	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$, No load	$V_{IC} = 2.5\text{ V}$, No load	25 °C		2.7	6.4	mA
				-55 °C		4	10	
				125 °C		1.9	4.4	

† Full range is -55 °C to 125 °C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

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Operational Amplifiers

TLC274M, TLC279M

LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC274M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$,	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC279M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$,	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	370	1200	μV
					Full range		4300	
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 125°C	2.2			$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C	0.1			pA
				125°C	1.8	15	nA	
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C	0.7			pA
				125°C	10	35	nA	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	0 to 9	-0.3 to 9.2		V
				Full range	0 to 8.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$,	$R_L = 10\text{ k}\Omega$	25°C	8	8.5		V
				-55°C	7.8	8.5		
				125°C	7.8	8.4		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$,	$I_{OL} = 0$	25°C	0	50		mV
				-55°C	0	50		
				125°C	0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$,	$R_L = 10\text{ k}\Omega$	25°C	10	36		V/mV
				-55°C	7	50		
				125°C	7	27		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min}$		25°C	65	85		dB
				-55°C	60	87		
				125°C	60	86		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$,	$V_O = 1.4\text{ V}$	25°C	65	95		dB
				-55°C	60	90		
				125°C	60	97		
I_{DD}	Supply current (four amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$,	25°C	3.8	8		mA
				-55°C	6.0	12		
				125°C	2.5	5.6		

† Full range is -55°C to 125°C .

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

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Operational Amplifiers

TLC274I, TLC274AI, TLC274BI, TLC279I LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, V_{DD} = 5 V (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T _A [†]	MIN	TYP	MAX	UNIT
V _{IO}	Input offset voltage	TLC274I	V _O = 1.4 V, R _S = 50 Ω,	V _{IC} = 0 V,	25 °C	1.1	10	mV
				R _L = 10 kΩ	Full range		13	
		TLC274AI	V _O = 1.4 V, R _S = 50 Ω,	V _{IC} = 0 V,	25 °C	0.9	5	
				R _L = 10 kΩ	Full range		7	
	TLC274BI	V _O = 1.4 V, R _S = 50 Ω,	V _{IC} = 0 V,	R _L = 10 kΩ	25 °C	340	2000	μV
					Full range		3500	
	TLC279I	V _O = 1.4 V, R _S = 50 Ω,	V _{IC} = 0 V,	R _L = 10 kΩ	25 °C	320	900	
					Full range		2000	
α _{VIO}	Average temperature coefficient of input offset voltage			25 °C to 85 °C		1.8		μV/°C
I _{IO}	Input offset current (see Note 4)	V _O = 2.5 V,	V _{IC} = 2.5 V	25 °C		0.1		pA
				85 °C		24	1000	
I _{IB}	Input bias current (see Note 4)	V _O = 2.5 V,	V _{IC} = 2.5 V	25 °C		0.6		pA
				85 °C		200	2000	
V _{ICR}	Common-mode input voltage range (see Note 5)			25 °C	-0.2 to 4	-0.3 to 4.2		V
				Full range	-0.2 to 3.5			V
V _{OH}	High-level output voltage	V _{ID} = 100 mV,	R _L = 10 kΩ	25 °C	3.2	3.8		V
				-40 °C	3	3.8		
				85 °C	3	3.8		
V _{OL}	Low-level output voltage	V _{ID} = -100 mV,	I _{OL} = 0	25 °C		0	50	mV
				-40 °C		0	50	
				85 °C		0	50	
A _{VD}	Large-signal differential voltage amplification	V _O = 0.25 V to 2 V,	R _L = 10 kΩ	25 °C	5	23		V/mV
				-40 °C	3.5	32		
				85 °C	3.5	19		
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICR} min		25 °C	65	80		dB
				-40 °C	60	81		
				85 °C	60	86		
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V		25 °C	65	95		dB
				-40 °C	60	92		
				85 °C	60	96		
I _{DD}	Supply current (four amplifiers)	V _O = 2.5 V, No load	V _{IC} = 2.5 V,	25 °C		2.7	6.4	mA
				-40 °C		3.8	8.8	
				85 °C		2.1	4.8	

[†] Full range is -40 °C to 85 °C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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Operational Amplifiers

TLC274I, TLC274AI, TLC274BI, TLC279I

LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC274I	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0\text{ V},$ $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
				Full range		13		
		TLC274AI	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0\text{ V},$ $R_L = 10\text{ k}\Omega$	25°C	0.9	5	
				Full range		7		
		TLC274BI	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0\text{ V},$ $R_L = 10\text{ k}\Omega$	25°C	390	2000	μV
Full range				3500				
TLC279I	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0\text{ V},$ $R_L = 10\text{ k}\Omega$	25°C	370	1200			
	Full range			2900				
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 85°C		2		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V},$	$V_{IC} = 5\text{ V}$	25°C	0.1			pA
				85°C	26	1000		
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V},$	$V_{IC} = 5\text{ V}$	25°C	0.7			pA
				85°C	220	2000		
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2	-0.3		V
					to	to		
				9	9.2			
				Full range	-0.2			V
					to			
					8.5			
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV},$	$R_L = 10\text{ k}\Omega$	25°C	8	8.5		V
				-40°C	7.8	8.5		
				85°C	7.8	8.5		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV},$	$I_{OL} = 0$	25°C		0	50	mV
				-40°C		0	50	
				85°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V},$	$R_L = 10\text{ k}\Omega$	25°C	10	36		V/mV
				-40°C	7	47		
				85°C	7	31		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min}$		25°C	65	85		dB
				-40°C	60	87		
				85°C	60	88		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V},$	$V_O = 1.4\text{ V}$	25°C	65	95		dB
				-40°C	60	92		
				85°C	60	96		
I_{DD}	Supply current (four amplifiers)	$V_O = 5\text{ V},$ No load	$V_{IC} = 5\text{ V},$	25°C	3.8	8		mA
				-40°C	5.5	10		
				85°C	2.9	6.4		

† Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

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Operational Amplifiers

TLC274C, TLC274AC, TLC274BC, TLC279C
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A †	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC274C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC274AC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	0.9	5	mV
					Full range		6.5	
		TLC274BC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	340	2000	μV
					Full range		3000	
		TLC279C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25°C	320	900	μV
					Full range		1500	
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C	1.8			$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1			pA
				70°C	7	300		
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6			pA
				70°C	40	600		
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2		V
				Full range	-0.2 to 3.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 10\text{ k}\Omega$	25°C	3.2	3.8		V
				0°C	3	3.8		
				70°C	3	3.8		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50		mV
				0°C	0	50		
				70°C	0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$	$R_L = 10\text{ k}\Omega$	25°C	5	23		V/mV
				0°C	4	27		
				70°C	4	20		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	80		dB
				0°C	60	84		
				70°C	60	85		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to } 10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	65	95		dB
				0°C	60	94		
				70°C	60	96		
I_{DD}	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$, No load	$V_{IC} = 2.5\text{ V}$	25°C	2.7	6.4		mA
				0°C	3.1	7.2		
				70°C	2.3	5.2		

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
 5. This range also applies to each input individually.

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Operational Amplifiers

TLC274C, TLC274AC, TLC274BC, TLC279C
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC274C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $R_L = 10\text{ k}\Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25 °C	1.1	10	mV
					Full range		12	
		TLC274AC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $R_L = 10\text{ k}\Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25 °C	0.9	5	
					Full range		6.5	
		TLC274BC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $R_L = 10\text{ k}\Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25 °C	390	2000	μV
					Full range		3000	
		TLC279C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $R_L = 10\text{ k}\Omega$	$V_{IC} = 0\text{ V}$, $R_L = 10\text{ k}\Omega$	25 °C	370	1200	
					Full range		1900	
α_{VIO}	Average temperature coefficient of input offset voltage			25 °C to 70 °C		2		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$, $V_{IC} = 5\text{ V}$		25 °C		0.1		pA
				70 °C		7	300	
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$, $V_{IC} = 5\text{ V}$		25 °C		0.7		pA
				70 °C		50	600	
V_{ICR}	Common-mode input voltage range (see Note 5)			25 °C	-0.2 to 9	-0.3 to 9.2		V
				Full range	-0.2 to 8.5			
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$, $R_L = 10\text{ k}\Omega$		25 °C	8	8.5		V
				0 °C	7.8	8.5		
				70 °C	7.8	8.4		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$, $I_{OL} = 0$		25 °C	0	50		mV
				0 °C	0	50		
				70 °C	0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$, $R_L = 10\text{ k}\Omega$		25 °C	10	36		V/mV
				0 °C	7.5	42		
				70 °C	7.5	32		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25 °C	65	85		dB
				0 °C	60	88		
				70 °C	60	88		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$, $V_O = 1.4\text{ V}$		25 °C	65	95		dB
				0 °C	60	94		
				70 °C	60	96		
I_{DD}	Supply current (four amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$	25 °C		3.8	8	mA
				0 °C		4.5	8.8	
				70 °C		3.2	6.8	

† Full range is 0 °C to 70 °C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
 5. This range also applies to each input individually.

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Operational Amplifiers

TLC274M, TLC279M
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C		3.6		V/ μ s
				-55°C		4.7		
				125°C		2.3		
			$V_{Ipp} = 2.5\text{ V}$	25°C		2.9		
				-55°C		3.7		
				125°C		2		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		25		nV/ $\sqrt{\text{Hz}}$
BOM	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25°C		320		kHz
				-55°C		400		
				125°C		230		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		1.7		MHz
				-55°C		2.9		
				125°C		1.1		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		46°		
				-55°C		49°		
				125°C		41°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C		5.3		V/ μ s
				-55°C		7.1		
				125°C		3.1		
			$V_{Ipp} = 5.5\text{ V}$	25°C		4.6		
				-55°C		6.1		
				125°C		2.7		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		25		nV/ $\sqrt{\text{Hz}}$
BOM	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25°C		200		kHz
				-55°C		280		
				125°C		110		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		2.2		MHz
				-55°C		3.4		
				125°C		1.6		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		49°		
				-55°C		52°		
				125°C		44°		

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Operational Amplifiers

TLC274I, TLC274AI, TLC274BI, TLC279I
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25 °C		3.6		V/ μ s
			-40 °C		4.5		
			85 °C		2.8		
		$V_{IPP} = 2.5\text{ V}$	25 °C		2.9		
			-40 °C		3.5		
			85 °C		2.3		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25 °C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25 °C		320		kHz
			-40 °C		380		
			85 °C		250		
B_1 Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25 °C		1.7		MHz
			-40 °C		2.6		
			85 °C		1.2		
ϕ_m Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25 °C		46°		
			-40 °C		49°		
			85 °C		43°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25 °C		5.3		V/ μ s
			-40 °C		6.7		
			85 °C		4		
		$V_{IPP} = 5.5\text{ V}$	25 °C		4.6		
			-40 °C		5.8		
			85 °C		3.5		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25 °C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25 °C		200		kHz
			-40 °C		260		
			85 °C		130		
B_1 Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25 °C		2.2		MHz
			-40 °C		3.1		
			85 °C		1.7		
ϕ_m Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25 °C		49°		
			-40 °C		52°		
			85 °C		46°		

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Operational Amplifiers

TLC274C, TLC274AC, TLC274BC, TLC279C
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

operating characteristics, V_{DD} = 5 V

PARAMETER	TEST CONDITIONS		T _A [†]	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	R _L = 10 kΩ, C _L = 20 pF, See Figure 1	V _I PP = 1 V	25°C		3.6		V/μs
			0°C		4		
			70°C		3		
		V _I PP = 2.5 V	25°C		2.9		
			0°C		3.1		
			70°C		2.5		
V _n Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 100 Ω,	25°C		25		nV/√Hz
B _{OM} Maximum output swing bandwidth	V _O = V _{OH} , R _L = 10 kΩ,	C _L = 20 pF, See Figure 1	25°C		320		kHz
			0°C		340		
			70°C		260		
B ₁ Unity-gain bandwidth	V _i = 10 mV, See Figure 3	C _L = 20 pF,	25°C		1.7		MHz
			0°C		2		
			70°C		1.3		
φ _m Phase margin	V _i = 10 mV, C _L = 20 pF,	f = B ₁ , See Figure 3	25°C		46°		
			0°C		47°		
			70°C		44°		

operating characteristics, V_{DD} = 10 V

PARAMETER	TEST CONDITIONS		T _A [†]	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	R _L = 10 kΩ, C _L = 20 pF, See Figure 1	V _I PP = 1 V	25°C		5.3		V/μs
			0°C		5.9		
			70°C		4.3		
		V _I PP = 5.5 V	25°C		4.6		
			0°C		5.1		
			70°C		3.8		
V _n Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 100 Ω,	25°C		25		nV/√Hz
B _{OM} Maximum output swing bandwidth	V _O = V _{OH} , R _L = 10 kΩ,	C _L = 20 pF, See Figure 1	25°C		200		kHz
			0°C		220		
			70°C		140		
B ₁ Unity-gain bandwidth	V _i = 10 mV, See Figure 3	C _L = 20 pF,	25°C		2.2		MHz
			0°C		2.5		
			70°C		1.8		
φ _m Phase margin	V _i = 10 mV, C _L = 20 pF,	f = B ₁ , See Figure 3	25°C		49°		
			0°C		50°		
			70°C		46°		

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Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC274 and TLC279 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit will give the same result.



FIGURE 1. UNITY-GAIN AMPLIFIER

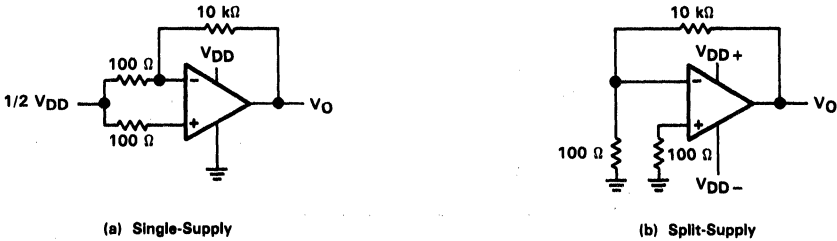


FIGURE 2. NOISE TEST CIRCUIT

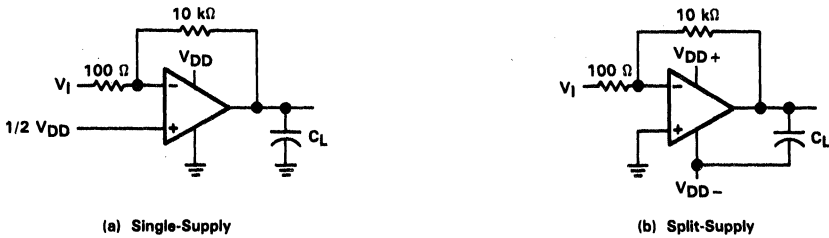


FIGURE 3. GAIN-OF-100 INVERTING AMPLIFIER

PARAMETER MEASUREMENT INFORMATION

input bias current

Because of the high input impedance of the TLC274 and TLC279 op amps, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs will be shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the "open-socket" leakage readings from the readings obtained with a device in the test socket.

One word of caution . . . many automatic testers as well as some bench-top op amp testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an "open-socket" reading is not feasible using this method.

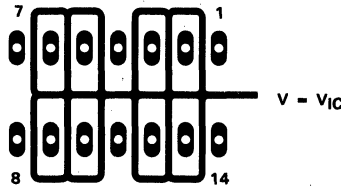


FIGURE 4. ISOLATION METAL AROUND DEVICE INPUTS
 (J AND N DUAL-IN-LINE-PACKAGE)

low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture will result in leakage and contact resistance which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

TLC274, TLC274A, TLC274B, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

full-power response

Full-power response, the frequency above which the op amp slew rate limits the output voltage swing, is often specified two ways . . . full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for "significant" distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

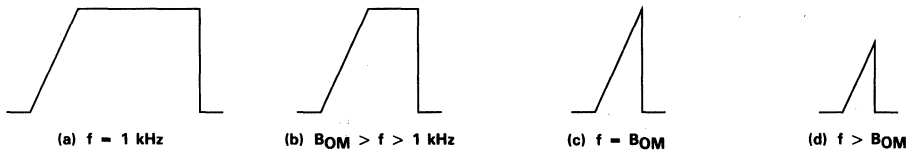


FIGURE 5. FULL-POWER-RESPONSE OUTPUT SIGNAL

test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC274
 INPUT OFFSET VOLTAGE

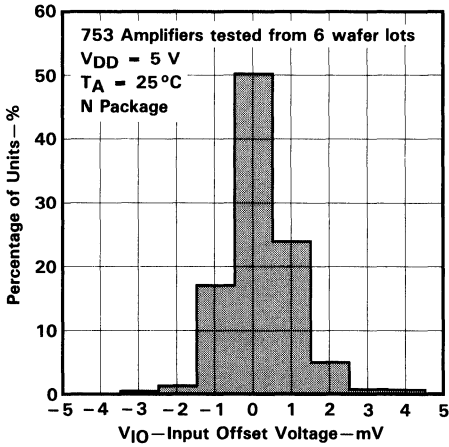


FIGURE 6

DISTRIBUTION OF TLC274
 INPUT OFFSET VOLTAGE

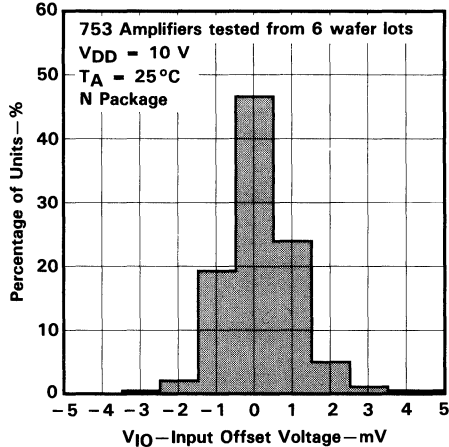


FIGURE 7

DISTRIBUTION OF TLC274 AND TLC279
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT

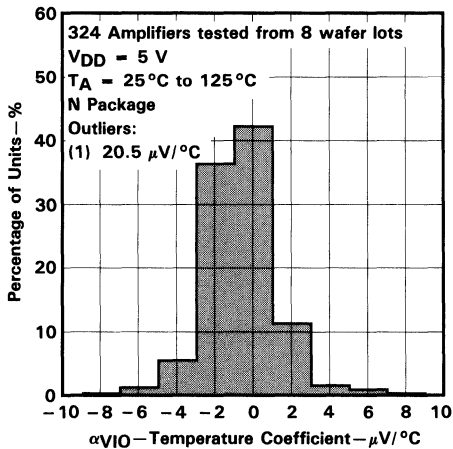


FIGURE 8

DISTRIBUTION OF TLC274 AND TLC279
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT

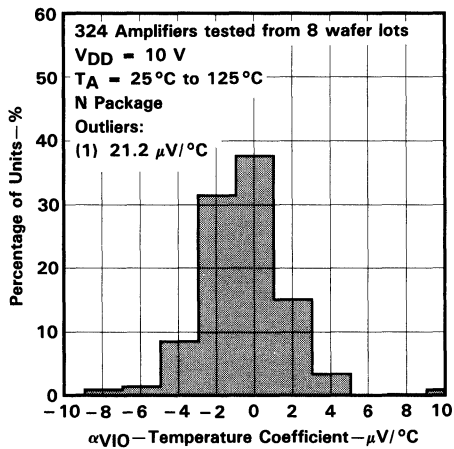


FIGURE 9

TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

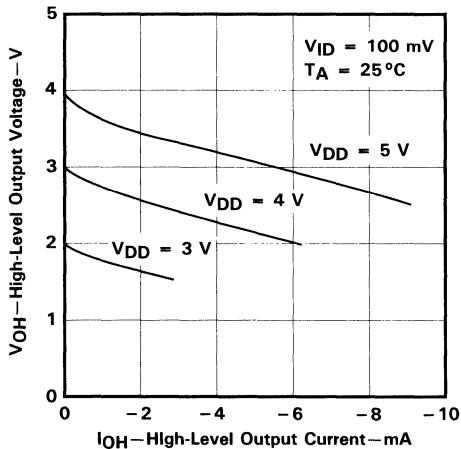


FIGURE 10

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

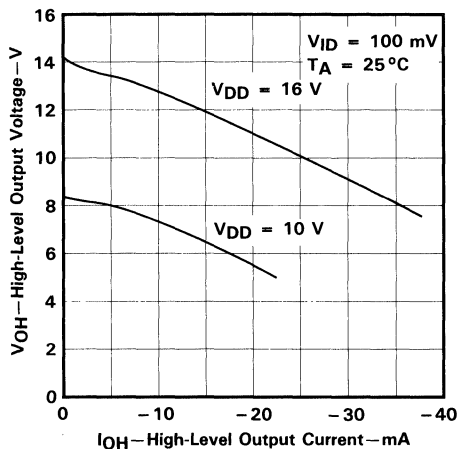


FIGURE 11

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 SUPPLY VOLTAGE

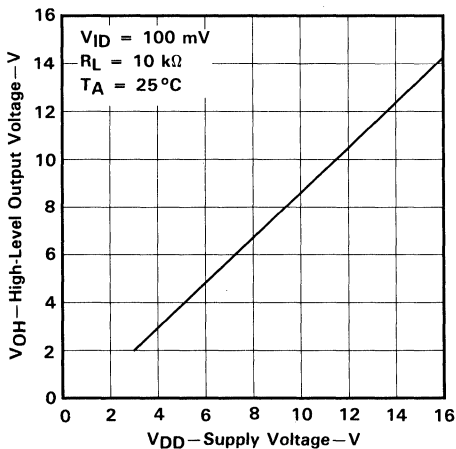


FIGURE 12

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

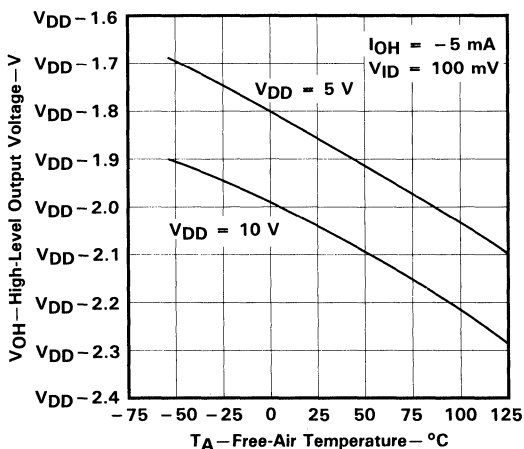


FIGURE 13

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

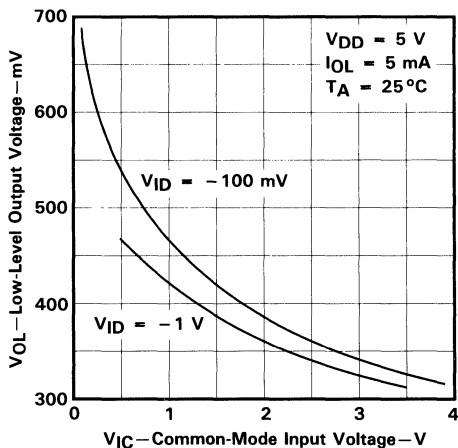


FIGURE 14

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

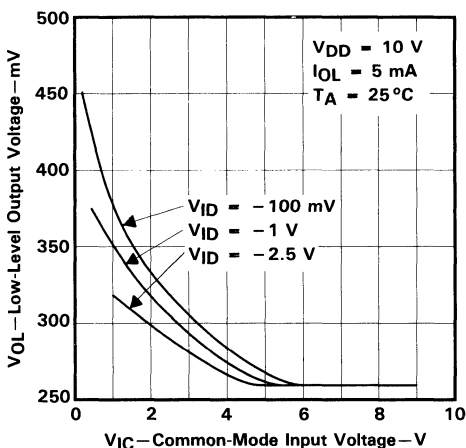


FIGURE 15

LOW-LEVEL OUTPUT VOLTAGE
 vs
 DIFFERENTIAL INPUT VOLTAGE

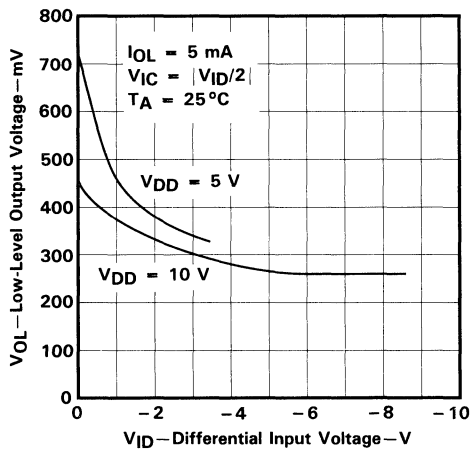


FIGURE 16

LOW-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

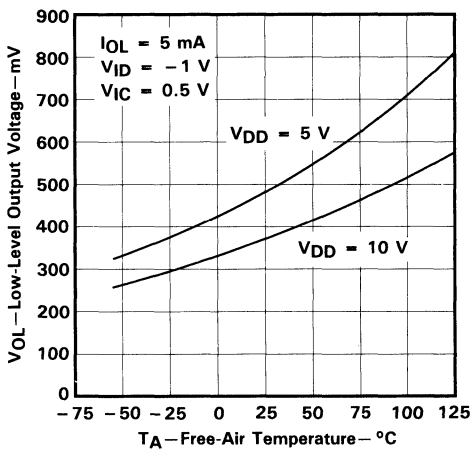


FIGURE 17

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

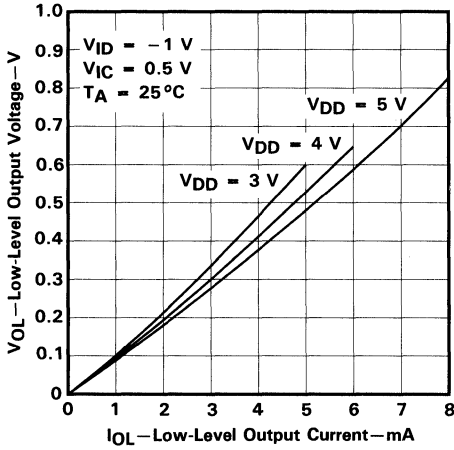


FIGURE 18

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

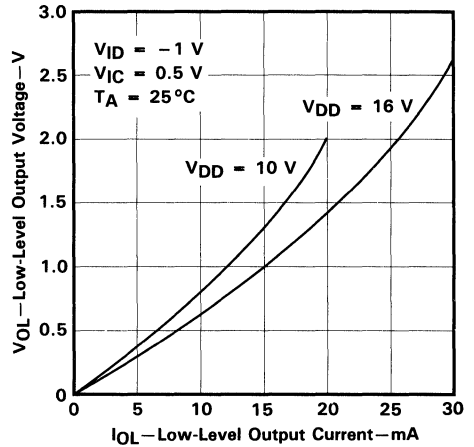


FIGURE 19

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 SUPPLY VOLTAGE

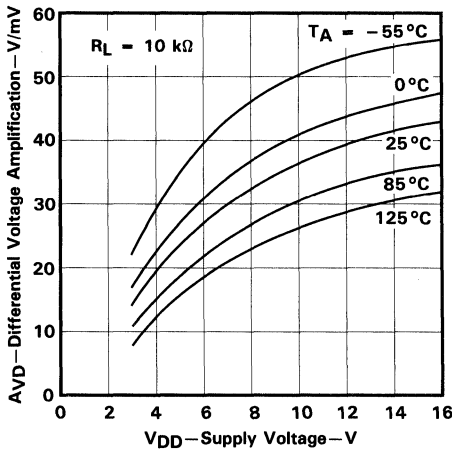


FIGURE 20

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 FREE-AIR TEMPERATURE

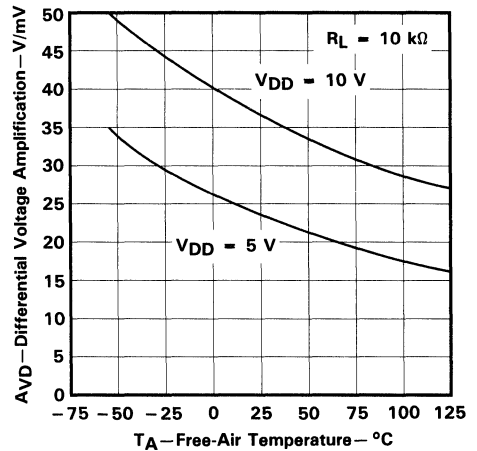


FIGURE 21

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC274, TLC274A, TLC274B, TLC279
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TYPICAL CHARACTERISTICS†

INPUT BIAS CURRENT AND INPUT OFFSET CURRENT
 vs
 FREE-AIR TEMPERATURE

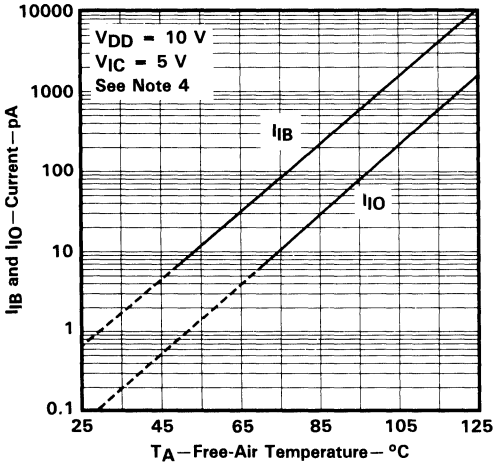


FIGURE 22

COMMON-MODE
 INPUT VOLTAGE POSITIVE LIMIT
 vs
 SUPPLY VOLTAGE

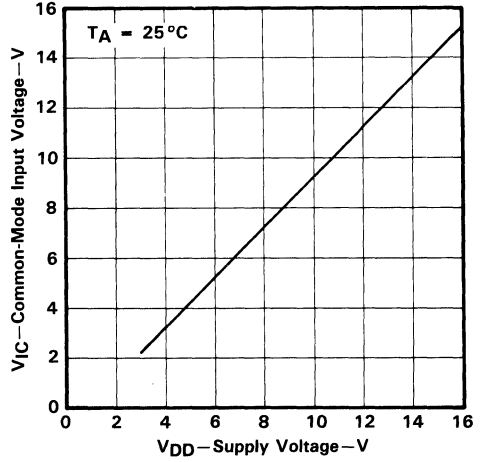


FIGURE 23

SUPPLY CURRENT
 vs
 SUPPLY VOLTAGE

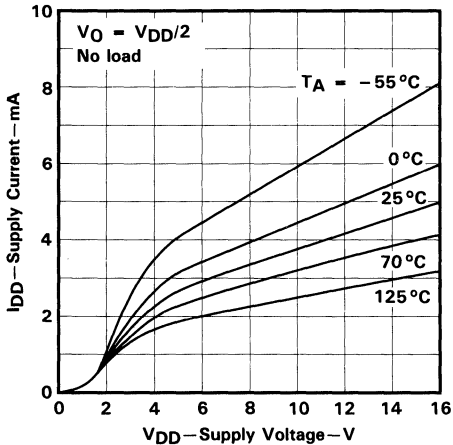


FIGURE 24

SUPPLY CURRENT
 vs
 FREE-AIR TEMPERATURE

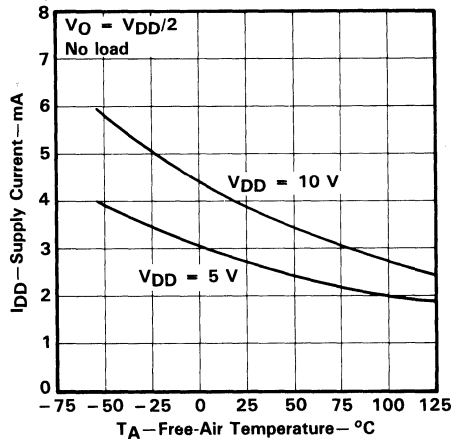


FIGURE 25

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.
 NOTE 4: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

TYPICAL CHARACTERISTICS†

**SLEW RATE
 vs
 SUPPLY VOLTAGE**

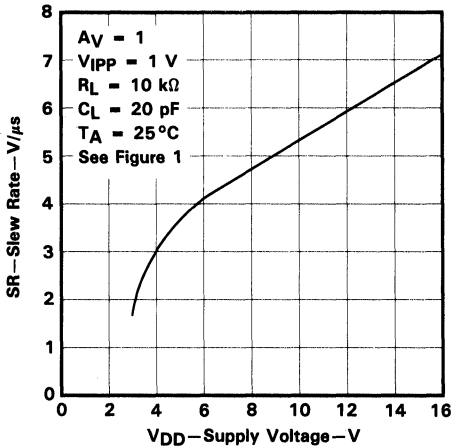


FIGURE 26

**SLEW RATE
 vs
 FREE-AIR TEMPERATURE**

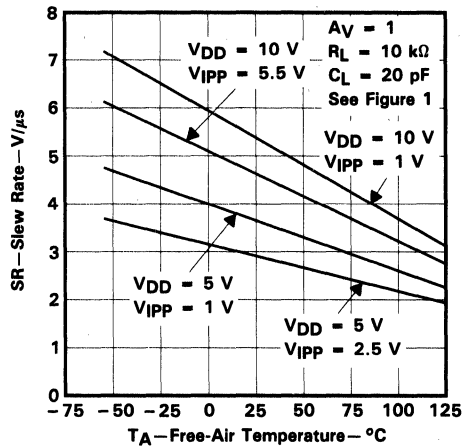


FIGURE 27

**NORMALIZED SLEW RATE
 vs
 FREE-AIR TEMPERATURE**

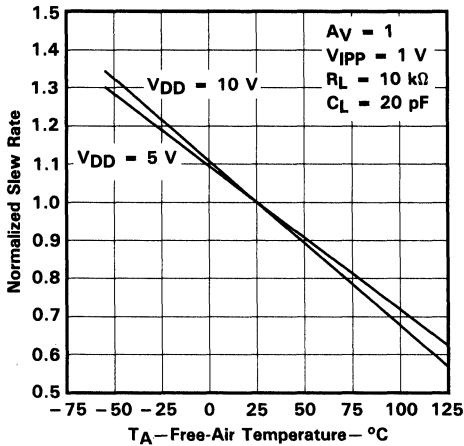


FIGURE 28

**MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 vs
 FREQUENCY**

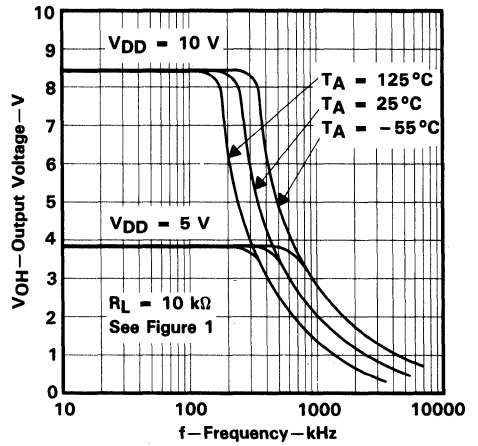


FIGURE 29

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC274, TLC274A, TLC274B, TLC279
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

UNITY-GAIN BANDWIDTH
 vs
 FREE-AIR TEMPERATURE

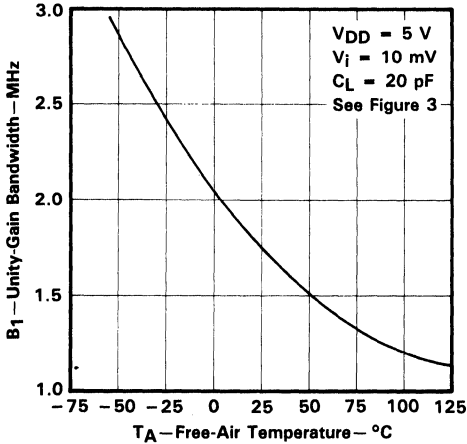


FIGURE 30

UNITY-GAIN BANDWIDTH
 vs
 SUPPLY VOLTAGE

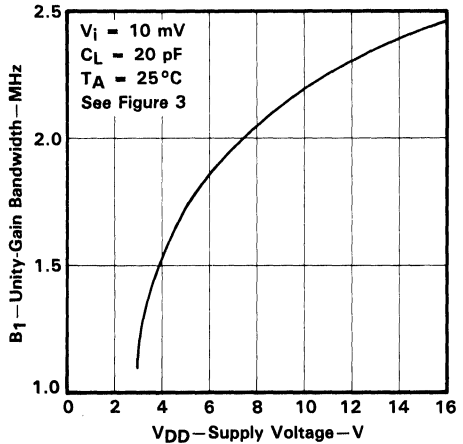


FIGURE 31

2
Operational Amplifiers

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

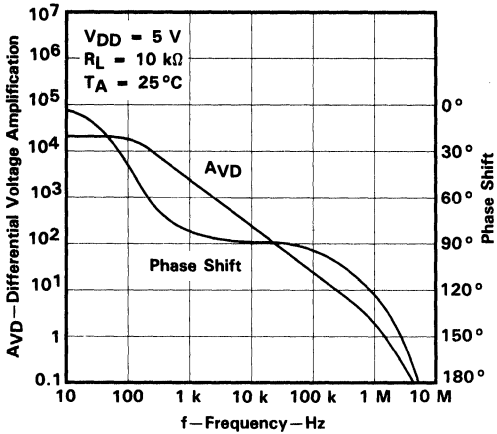


FIGURE 32

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

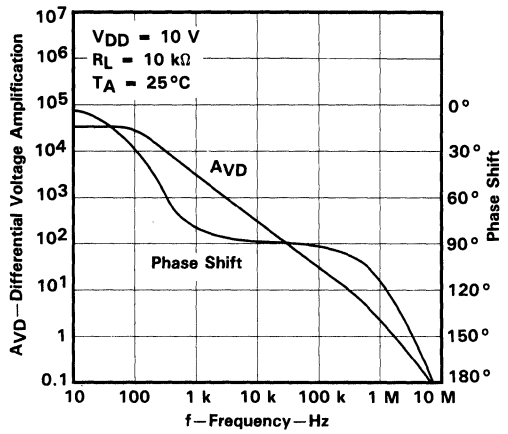


FIGURE 33

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

PHASE MARGIN
 vs
 SUPPLY VOLTAGE

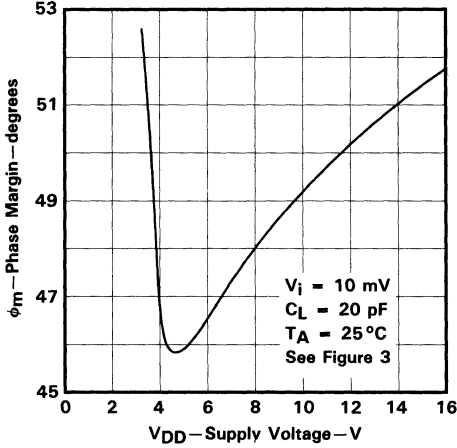


FIGURE 34

PHASE MARGIN
 vs
 FREE-AIR TEMPERATURE

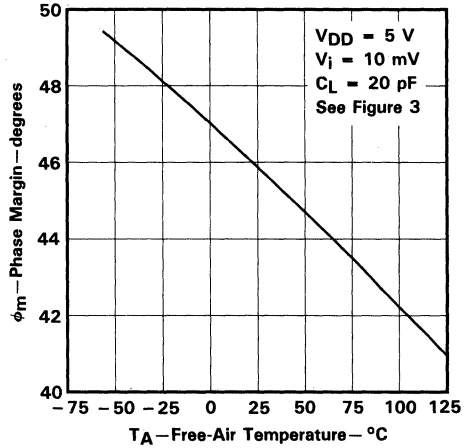


FIGURE 35

PHASE MARGIN
 vs
 CAPACITIVE LOAD

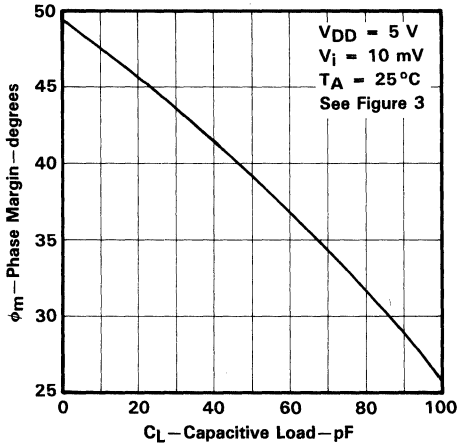


FIGURE 36

EQUIVALENT INPUT NOISE VOLTAGE
 vs
 FREQUENCY

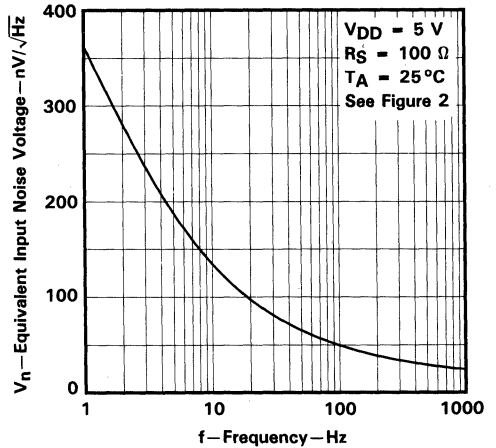


FIGURE 37

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL APPLICATION DATA

single-supply operation

While the TLC274 and TLC279 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC274 and TLC279 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC274 and TLC279 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

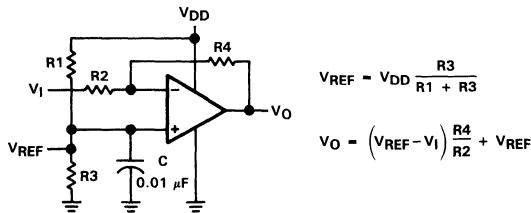


FIGURE 38. INVERTING AMPLIFIER WITH VOLTAGE REFERENCE

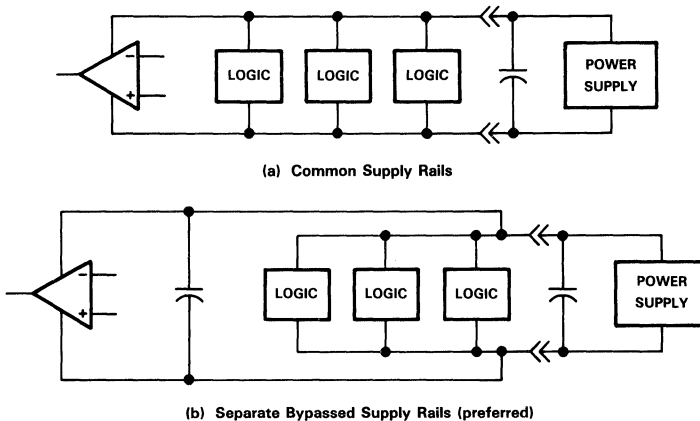


FIGURE 39. COMMON VS SEPARATE SUPPLY RAILS

TYPICAL APPLICATION DATA

input characteristics

The TLC274 and TLC279 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at $V_{DD} - 1\text{ V}$ at $T_A = 25^\circ\text{C}$ and at $V_{DD} - 1.5\text{ V}$ at all other temperatures.

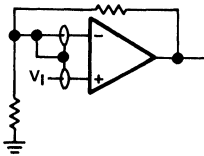
The use of the polysilicon-gate process and the careful input circuit design gives the TLC274 and TLC279 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically $0.1\ \mu\text{V}/\text{month}$, including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC274 and TLC279 are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

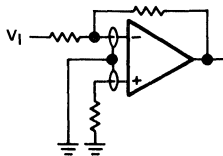
The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.

noise performance

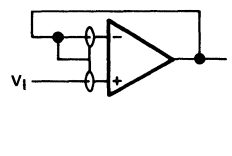
The noise specifications in op amp circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC274 and TLC279 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than $50\text{ k}\Omega$, since bipolar devices exhibit greater noise currents.



(a) Noninverting Amplifier



(b) Inverting Amplifier



(c) Unity-Gain Amplifier

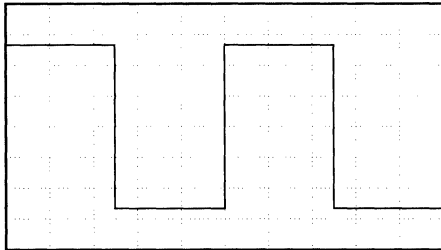
FIGURE 40. GUARD-RING SCHEMES

output characteristics

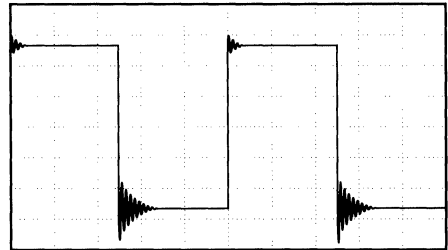
The output stage of the TLC274 and TLC279 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

All operating characteristics of the TLC274 and TLC279 were measured using a 20-pF load. The devices will drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance will alleviate the problem.

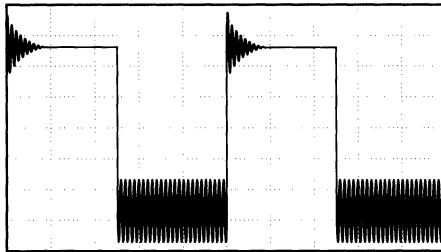
TYPICAL APPLICATION DATA



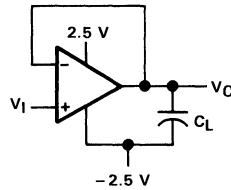
(a) $C_L = 20 \text{ pF}$, $R_L = \text{No load}$



(b) $C_L = 130 \text{ pF}$, $R_L = \text{No load}$



(c) $C_L = 150 \text{ pF}$, $R_L = \text{No load}$



(d) Test Circuit

$T_A = 25^\circ\text{C}$
 $f = 1 \text{ kHz}$
 $V_{\text{IPP}} = 1 \text{ V}$

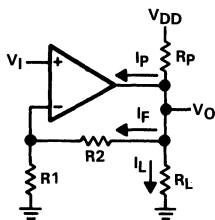
FIGURE 41. EFFECT OF CAPACITIVE LOADS AND TEST CIRCUIT

Although the TLC274 and TLC279 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor (R_p) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pull-down transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately 60Ω and 180Ω , depending on how hard the op amp input is driven. With very low values of R_p , a voltage offset from 0 V at the output will occur. Second, pullup resistor R_p acts as a drain load to N4 and the gain of the op amp is reduced at output voltage levels where N5 is not supplying the output current.

feedback

Op amp circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

TYPICAL APPLICATION DATA



$$R_P = \frac{V_{DD} - V_O}{I_F + I_L + I_P}$$

I_P = Pullup current required by the op amp (typically 500 μ A)

FIGURE 42. RESISTIVE PULLUP TO INCREASE V_{OH}

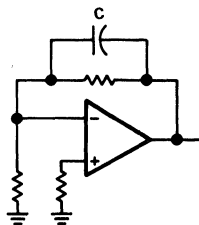


FIGURE 43. COMPENSATION FOR INPUT CAPACITANCE

2

Operational Amplifiers

electrostatic discharge protection

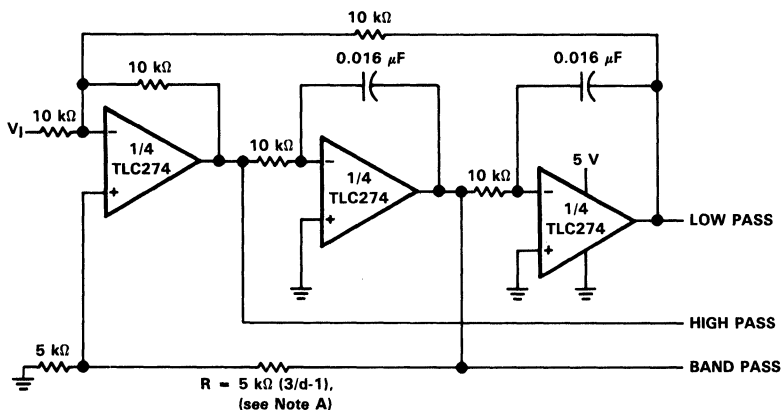
The TLC274 and TLC279 incorporate an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

latchup

Because CMOS devices are susceptible to latchup due to their inherent parasitic thyristors, the TLC274 and TLC279 inputs and outputs were designed to withstand – 100-mA surge currents without sustaining latchup; however, techniques should be used to reduce the chance of latchup whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μ F typical) located across the supply rails as close to the device as possible.

The current path established if latchup occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latchup occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latchup occurring increases with increasing temperature and supply voltages.

TYPICAL APPLICATION DATA



NOTES: A. d = damping factor, $1/Q$
 B. Normalized to $10\text{ k}\Omega$ and $f_c = 1\text{ kHz}$

FIGURE 44. STATE VARIABLE FILTER

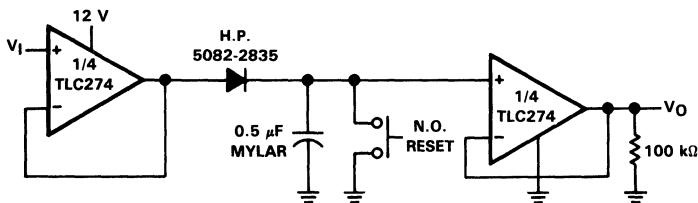


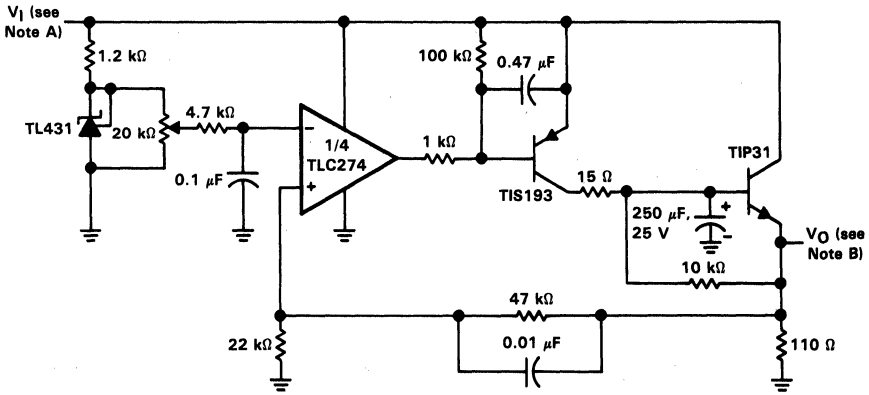
FIGURE 45. POSITIVE-PEAK DETECTOR

TLC274, TLC274A, TLC274B, TLC279
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

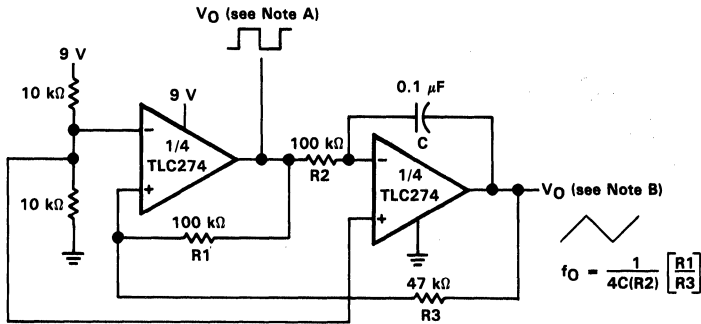
2

Operational Amplifiers



NOTES: A. $V_I = 3.5$ to 15 V
 B. $V_O = 2.0$ V, 0 to 1 A

FIGURE 46. LOGIC ARRAY POWER SUPPLY



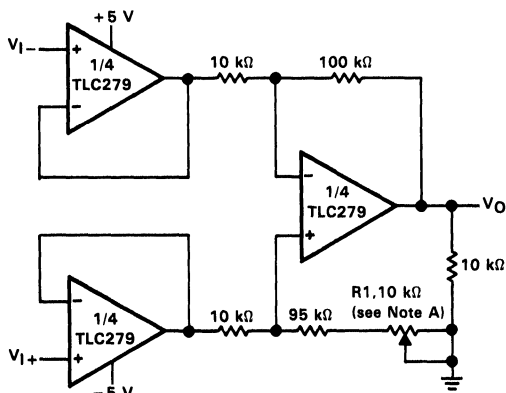
NOTES: A. $V_{OPP} = 8$ V
 B. $V_{OPP} = 4$ V

$$f_0 = \frac{1}{4C(R2)} \left[\frac{R1}{R3} \right]$$

FIGURE 47. SINGLE-SUPPLY FUNCTION GENERATOR

TLC274, TLC274A, TLC274B, TLC279
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA



NOTE A: CMRR adjustment (must be noninductive).

FIGURE 48. LOW-POWER INSTRUMENTATION AMPLIFIER

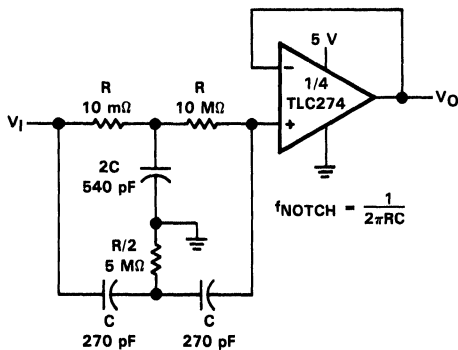


FIGURE 49. SINGLE-SUPPLY TWIN-T NOTCH FILTER

2

Operational Amplifiers

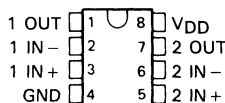
TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

D3139, OCTOBER 1987—REVISED AUGUST 1988

- **Trimmed Offset Voltage:**
TLC27L7 . . . 500 μV Max at 25°C,
 $V_{\text{DD}} = 5\text{ V}$
- **Input Offset Voltage Drift . . . Typically**
0.1 $\mu\text{V}/\text{Month}$, Including the First 30 Days
- **Wide Range of Supply Voltages over Specified Temperature Range:**
-55°C to 125°C . . . 4 V to 16 V
-40°C to 85°C . . . 4 V to 16 V
0°C to 70°C . . . 3 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix types)**
- **Ultralow Power . . . Typically 95 μW at 25°C, $V_{\text{DD}} = 5\text{ V}$**
- **Output Voltage Range Includes Negative Rail**
- **High Input Impedance . . . 10¹² Ω Typical**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available in Tape and Reel**
- **Designed-In Latchup Immunity**

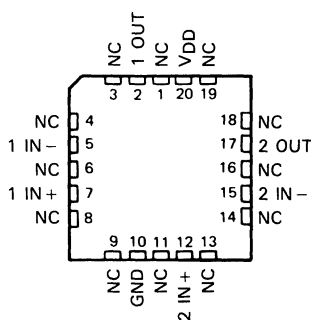
D, JG, OR P PACKAGE

(TOP VIEW)



FK PACKAGE

(TOP VIEW)



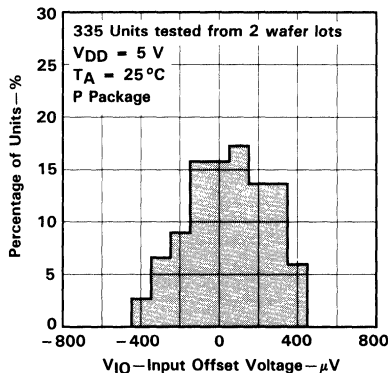
NC—No internal connection

AVAILABLE OPTIONS

T _A	V _{IO} max at 25°C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	500 μV	TLC27L7CD	—	TLC27L7CJG	TLC27L7CP
	2 mV	TLC27L2BCD	—	TLC27L2BCJG	TLC27L2BCP
	5 mV	TLC27L2ACD	—	TLC27L2ACJG	TLC27L2ACP
-40°C to 85°C	10 mV	TLC27L2CD	—	TLC27L2CJG	TLC27L2CP
	500 μV	TLC27L7ID	—	TLC27L7IJG	TLC27L7IP
	2 mV	TLC27L2BID	—	TLC27L2BIJG	TLC27L2BIP
-55°C to 125°C	5 mV	TLC27L2AID	—	TLC27L2AIJG	TLC27L2AIP
	10 mV	TLC27L2ID	—	TLC27L2IJG	TLC27L2IP
-55°C to 125°C	500 μV	—	TLC27L7MFK	TLC27L7MJG	—
	10 mV	—	TLC27L2MFK	TLC27L2MJG	—

The D package is available taped and reeled. Add R suffix to the device type (e.g., TLC27L7CDR).

DISTRIBUTION OF TLC27L7
INPUT OFFSET VOLTAGE



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2-607

TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

description

The TLC27L2 and TLC27L7 dual operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, extremely low power, and high gain.

These devices use Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, and low power consumption make these cost-effective devices ideal for high gain, low frequency, low power applications. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC27L2 (10 mV) to the high-precision TLC27L7 (500 μ V). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC27L2 and TLC27L7. The devices also exhibit low voltage single-supply operation and ultra-low power consumption, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip carrier versions for high-density system applications.

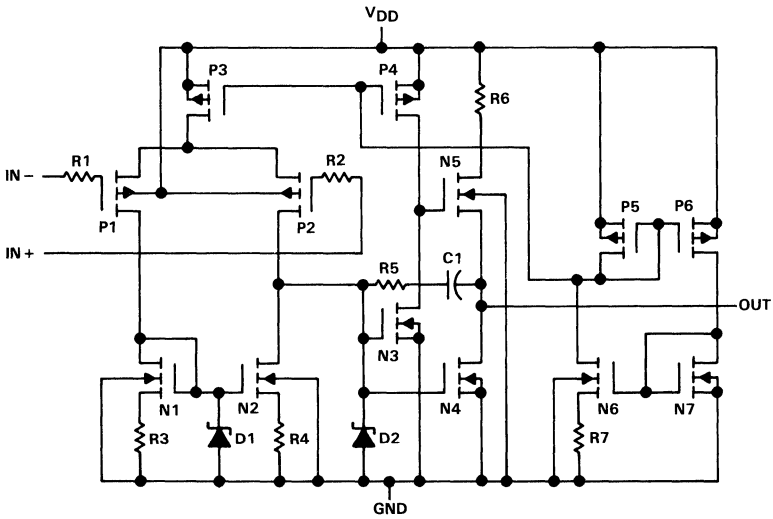
The device inputs and outputs are designed to withstand -100 -mA surge currents without sustaining latchup.

The TLC27L2 and TLC27L7 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . I-suffix devices are characterized for operation from -40°C to 85°C , and C-suffix devices are characterized for operation from 0°C to 70°C .

TLC27L2, TLC27L2A, TLC27L2B, TLC27L7
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

equivalent schematic (each amplifier)



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Operational Amplifiers

TLC27L2M, TLC27L7M LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT	
V_{IO}	Input offset voltage	TLC27L2M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV	
				Full range			12		
		TLC27L7M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C		170	500	μV
					Full range			3750	
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 125°C		1.4		$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$,	$V_{IC} = 2.5\text{ V}$	25°C		0.1		pA	
				125°C		1.4	15	nA	
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$,	$V_{IC} = 2.5\text{ V}$	25°C		0.6		pA	
				125°C		9	35	nA	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	0 to 4	-0.3 to 4.2		V	
				Full range	0 to 3.5			V	
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$,	$R_L = 1\text{ M}\Omega$	25°C	3.2	4.1		V	
				-55°C	3	4.1			
				125°C	3	4.2			
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$,	$I_{OL} = 0$	25°C		0	50	mV	
				-55°C		0	50		
				125°C		0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$,	$R_L = 1\text{ M}\Omega$	25°C	50	500		V/mV	
				-55°C	25	1000			
				125°C	25	200			
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min}$		25°C	65	94		dB	
				-55°C	60	95			
				125°C	60	85			
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$,	$V_O = 1.4\text{ V}$	25°C	70	97		dB	
				-55°C	60	97			
				125°C	60	98			
I_{DD}	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$, No load	$V_{IC} = 2.5\text{ V}$,	25°C		20	34	μA	
				-55°C		35	60		
				125°C		14	24		

† Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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Operational Amplifiers

TLC27L2M, TLC27L7M LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A †	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27L2M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	1.1	10		mV
				Full range		12		
	TLC27L7M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	190	800		μV	
			Full range		4300			
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 125°C	1.4			$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$, $V_{IC} = 5\text{ V}$		25°C	0.1			pA
				125°C	1.8	15		nA
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$, $V_{IC} = 5\text{ V}$		25°C	0.7			pA
				125°C	10	35		nA
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	0 to 9	-0.3 to 9.2		V
				Full range	0 to 8.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$, $R_L = 1\text{ M}\Omega$		25°C	8	8.9		V
				-55°C	7.8	8.8		
				125°C	7.8	9		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$, $I_{OL} = 0$		25°C	0	50		mV
				-55°C	0	50		
				125°C	0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$, $R_L = 1\text{ M}\Omega$		25°C	50	860		V/mV
				-55°C	25	1750		
				125°C	25	380		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	97		dB
				-55°C	60	97		
				125°C	60	91		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$, $V_O = 1.4\text{ V}$		25°C	70	97		dB
				-55°C	60	97		
				125°C	60	98		
I_{DD}	Supply current (two amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$	25°C	29	46		μA
				-55°C	56	96		
				125°C	18	30		

† Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

TLC27L2C, TLC27L2AC, TLC27L2BC, TLC27L7C
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A †	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27L2I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		13	
		TLC27L2AI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	0.9	5	
					Full range		7	
TLC27L2BI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	235	2000	μV		
			Full range		3500			
TLC27L7I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	190	800	μV		
			Full range		2900			
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 85°C		2.1		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C		0.1		pA
				85°C		26	1000	
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C		0.7		pA
				85°C		220	2000	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2		V
				Full range	-0.2 to 8.5			
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$,	$R_L = 1\text{ M}\Omega$	25°C	8	8.9		V
				-40°C	7.8	8.9		
				85°C	7.8	8.9		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$,	$I_{OL} = 0$	25°C		0	50	mV
				-40°C		0	50	
				85°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$,	$R_L = 1\text{ M}\Omega$	25°C	50	860		V/mV
				-40°C	50	1550		
				85°C	50	585		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	97		dB
				-40°C	60	97		
				85°C	60	98		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$,	$V_O = 1.4\text{ V}$	25°C	70	97		dB
				-40°C	60	97		
				85°C	60	98		
I_{DD}	Supply current (two amplifiers)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$,	25°C		29	46	μA
				-40°C		49	86	
		No load			20	36		

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
 5. This range also applies to each input individually.

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Operational Amplifiers

TLC27L2I, TLC27L2AI, TLC27L2BI, TLC27L7I
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27L2I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		13	
		TLC27L2AI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	0.9	5	mV
					Full range		7	
TLC27L2BI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	235	2000	μV		
			Full range		3500			
TLC27L7I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	190	800	μV		
			Full range		2900			
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 85°C		1		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C		0.1		pA
				85°C		26	1000	
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C		0.7		pA
				85°C		220	2000	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2		V
				Full range	-0.2 to 8.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	25°C	8	8.9		V
				-40°C	7.8	8.9		
				85°C	7.8	8.9		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C		0	50	mV
				-40°C		0	50	
				85°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 1\text{ M}\Omega$	25°C	50	860		V/mV
				-40°C	50	1550		
				85°C	50	585		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	97		dB
				-40°C	60	97		
				85°C	60	98		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	97		dB
				-40°C	60	97		
				85°C	60	98		
I_{DD}	Supply current (two amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$	25°C		29	46	μA
				-40°C		49	86	
				85°C		20	36	

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
 5. This range also applies to each input individually.

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Operational Amplifiers

TLC27L2I, TLC27L2AI, TLC27L2BI, TLC27L7I LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A †	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27L2C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27L2AC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	0.9	5	
					Full range		6.5	
	TLC27L2BC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	204	2000	μV	
				Full range		3000		
		TLC27L7C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	170		500
					Full range			1500
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C	1.1			$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$,	$V_{IC} = 2.5\text{ V}$	25°C	0.1			pA
				70°C	7	300		
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$,	$V_{IC} = 2.5\text{ V}$	25°C	0.6			pA
				70°C	50	600		
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2		V
				Full range	-0.2 to 3.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$,	$R_L = 1\text{ M}\Omega$	25°C	3.2	4.1		V
				0°C	3	4.1		
				70°C	3	4.2		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$,	$I_{OL} = 0$	25°C		0	50	mV
				0°C		0	50	
				70°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$, $R_L = 1\text{ M}\Omega$		25°C	50	500		V/mV
				0°C	50	700		
				70°C	50	380		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	94		dB
				0°C	60	95		
				70°C	60	95		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to } 10\text{ V}$, $V_O = 1.4\text{ V}$		25°C	70	97		dB
				0°C	60	97		
				70°C	60	98		
I_{DD}	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$,	No load	$V_{IC} = 2.5\text{ V}$,	25°C	20	34	μA
					0°C	24	42	
					70°C	16	28	

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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Operational Amplifiers

TLC27L2C, TLC27L2AC, TLC27L2BC, TLC27L7C
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A †	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27L2C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27L2AC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	0.9	5	
					Full range		6.5	
TLC27L2BC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	235	2000			
			Full range		3000			
TLC27L7C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	190	800		μV	
Full range				1900				
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C		1		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C		0.1		pA
				70°C		8	300	
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C		0.7		pA
				70°C		50	600	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2		V
				Full range	-0.2 to 8.5		V	
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	25°C	8	8.9		V
				0°C	7.8	8.9		
				70°C	7.8	8.9		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C		0	50	mV
				0°C		0	50	
				70°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 1\text{ M}\Omega$	25°C	50	860		V/mV
				0°C	50	1025		
				70°C	50	660		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min}$		25°C	65	97		dB
				0°C	60	97		
				70°C	60	97		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	97		dB
				0°C	60	97		
				70°C	60	98		
I_{DD}	Supply current (two amplifiers)	$V_O = 5\text{ V}$	No load	$V_{IC} = 5\text{ V}$	25°C	29	46	μA
					0°C	36	66	
					70°C	22	40	

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
 5. This range also applies to each input individually.

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Operational Amplifiers

TLC27L2M, TLC27L7M
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operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25 °C		0.03		V/ μ s
				-55 °C		0.04		
				125 °C		0.02		
			$V_{Ipp} = 2.5\text{ V}$	25 °C		0.03		
				-55 °C		0.04		
				125 °C		0.02		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$	25 °C		68		nV/ $\sqrt{\text{Hz}}$
B _{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$	$C_L = 20\text{ pF}$, See Figure 1	25 °C		5		kHz
				-55 °C		8		
				125 °C		3		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$	25 °C		85		kHz
				-55 °C		140		
				125 °C		45		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$	$f = B_1$, See Figure 3	25 °C		34 °		
				-55 °C		39 °		
				125 °C		25 °		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25 °C		0.05		V/ μ s
				-55 °C		0.06		
				125 °C		0.03		
			$V_{Ipp} = 5.5\text{ V}$	25 °C		0.04		
				-55 °C		0.06		
				125 °C		0.03		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$	25 °C		68		nV/ $\sqrt{\text{Hz}}$
B _{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$	$C_L = 20\text{ pF}$, See Figure 1	25 °C		1		kHz
				-55 °C		1.5		
				125 °C		0.7		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$	25 °C		110		kHz
				-55 °C		165		
				125 °C		70		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$	$f = B_1$, See Figure 3	25 °C		38 °		
				-55 °C		43 °		
				125 °C		29 °		

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Operational Amplifiers

TLC27L2I, TLC27L2AI, TLC27L2BI, TLC27L7I
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operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C		0.03		V/ μ s
				-40°C		0.04		
				85°C		0.03		
			$V_{Ipp} = 2.5\text{ V}$	25°C		0.03		
				-40°C		0.04		
				85°C		0.02		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		68		nV/ $\sqrt{\text{Hz}}$
B _{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25°C		5		kHz
				-40°C		7		
				85°C		4		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		85		kHz
				-40°C		130		
				85°C		55		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		34°		
				-40°C		38°		
				85°C		29°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C		0.05		V/ μ s
				-40°C		0.06		
				85°C		0.03		
			$V_{Ipp} = 5.5\text{ V}$	25°C		0.04		
				-40°C		0.05		
				85°C		0.03		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		68		nV/ $\sqrt{\text{Hz}}$
B _{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25°C		1		kHz
				-40°C		1.4		
				85°C		0.8		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		110		kHz
				-40°C		155		
				85°C		80		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		38°		
				-40°C		42°		
				85°C		32°		

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Operational Amplifiers

TLC27L2C, TLC27L2AC, TLC27L2BC, TLC27L7C LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25 °C		0.03		$V/\mu\text{s}$
				0 °C		0.04		
				70 °C		0.03		
			$V_{IPP} = 2.5\text{ V}$	25 °C		0.03		
				0 °C		0.03		
				70 °C		0.02		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$	25 °C		68		$\text{nV}/\sqrt{\text{Hz}}$
BOM	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25 °C		5		kHz
				0 °C		6		
				70 °C		4.5		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25 °C		85		kHz
				0 °C		100		
				70 °C		65		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25 °C		34 °		
				0 °C		36 °		
				70 °C		30 °		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25 °C		0.05		$V/\mu\text{s}$
				0 °C		0.05		
				70 °C		0.04		
			$V_{IPP} = 5.5\text{ V}$	25 °C		0.04		
				0 °C		0.05		
				70 °C		0.04		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$	25 °C		68		$\text{nV}/\sqrt{\text{Hz}}$
BOM	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25 °C		1		kHz
				0 °C		1.3		
				70 °C		0.9		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25 °C		110		kHz
				0 °C		125		
				70 °C		90		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25 °C		38 °		
				0 °C		40 °		
				70 °C		34 °		

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Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC27L2 and TLC27L7 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit will give the same result.



FIGURE 1. UNITY-GAIN AMPLIFIER

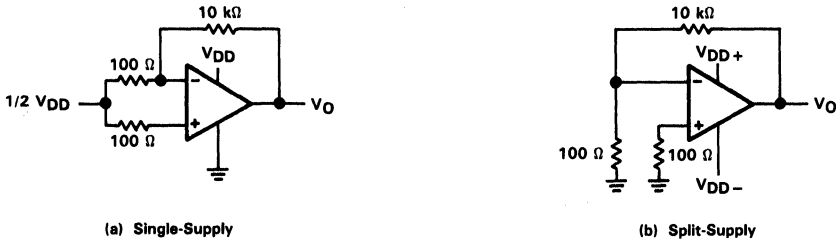


FIGURE 2. NOISE TEST CIRCUIT

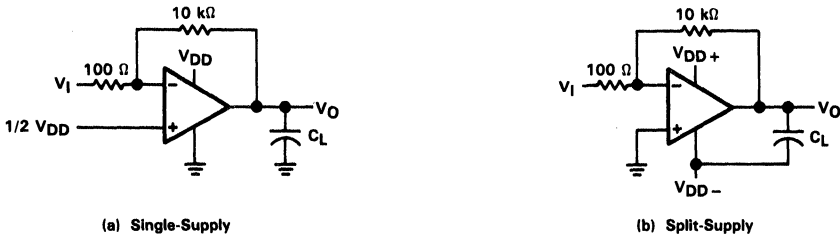


FIGURE 3. GAIN-OF-100 INVERTING AMPLIFIER

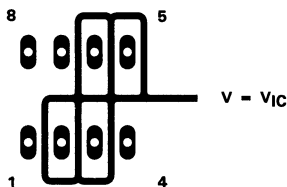
PARAMETER MEASUREMENT INFORMATION

input bias current

Because of the high input impedance of the TLC27L2 and TLC27L7 op amps, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs will be shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the "open-socket" leakage readings from the readings obtained with a device in the test socket.

One word of caution . . . many automatic testers as well as some bench-top op amp testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an "open-socket" reading is not feasible using this method.



**FIGURE 4. ISOLATION METAL AROUND DEVICE INPUTS
 (JG AND P DUAL-IN-LINE-PACKAGE)**

low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture will result in leakage and contact resistance which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

PARAMETER MEASUREMENT INFORMATION

full-power response

Full-power response, the frequency above which the op amp slew rate limits the output voltage swing, is often specified two ways . . . full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for "significant" distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

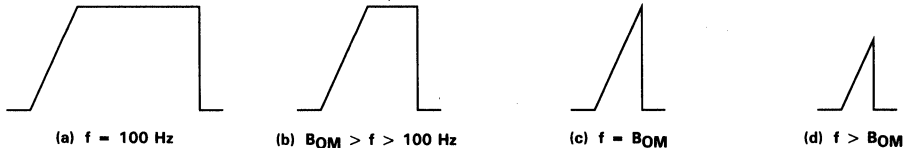


FIGURE 5. FULL-POWER-RESPONSE OUTPUT SIGNAL

test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

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Operational Amplifiers

TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC27L2
 INPUT OFFSET VOLTAGE

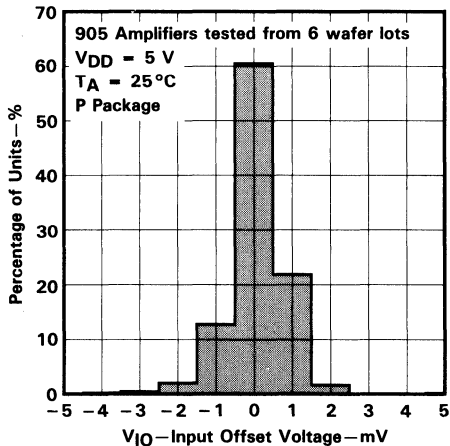


FIGURE 6

DISTRIBUTION OF TLC27L2
 INPUT OFFSET VOLTAGE

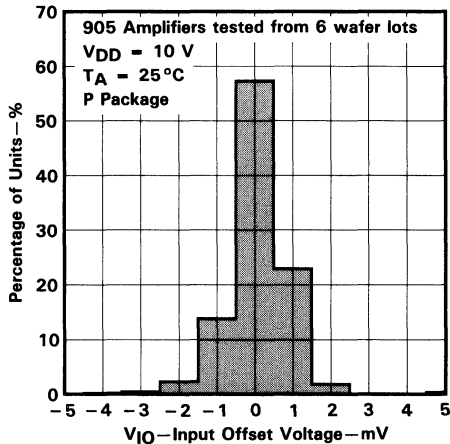


FIGURE 7

DISTRIBUTION OF TLC27L2 AND TLC27L7
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT

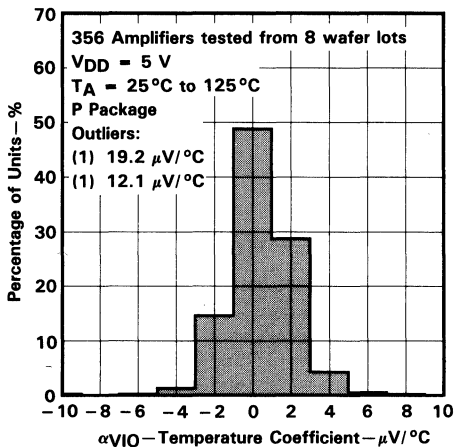


FIGURE 8

DISTRIBUTION OF TLC27L2 AND TLC27L7
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT

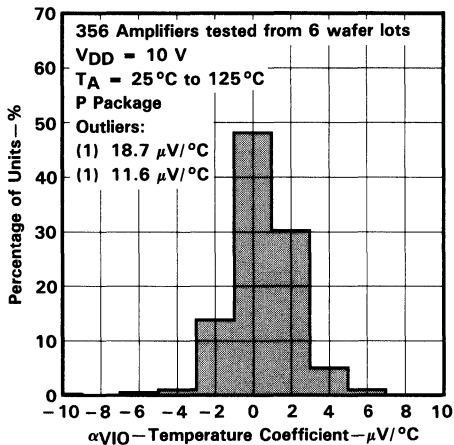


FIGURE 9

TLC27L2, TLC27L2A, TLC27L2B, TLC27L7
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

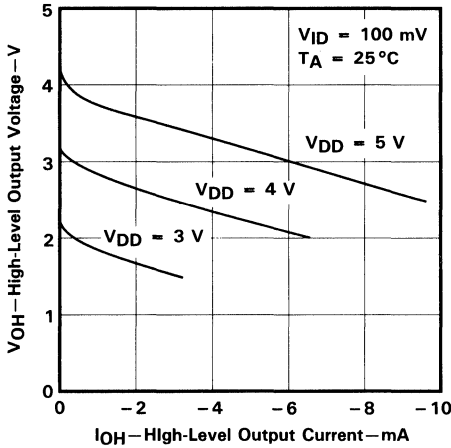


FIGURE 10

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

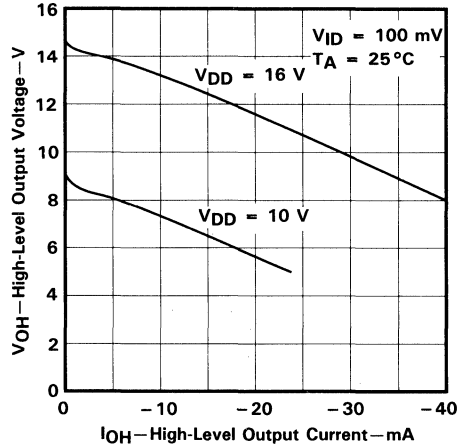


FIGURE 11

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 SUPPLY VOLTAGE

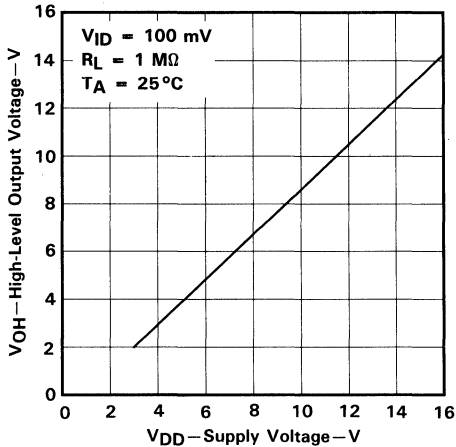


FIGURE 12

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

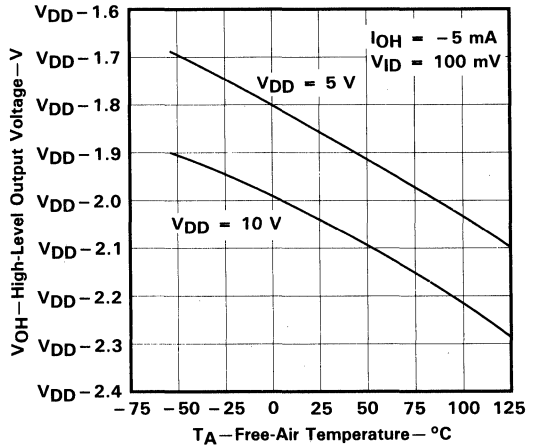


FIGURE 13

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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Operational Amplifiers

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

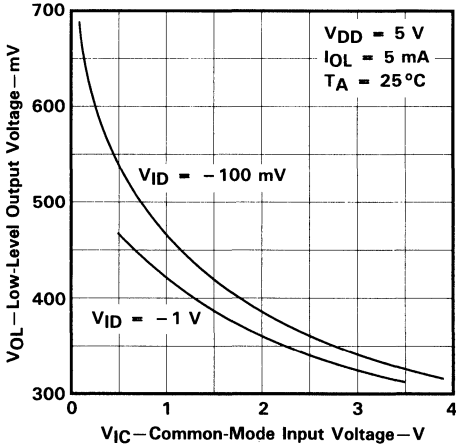


FIGURE 14

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

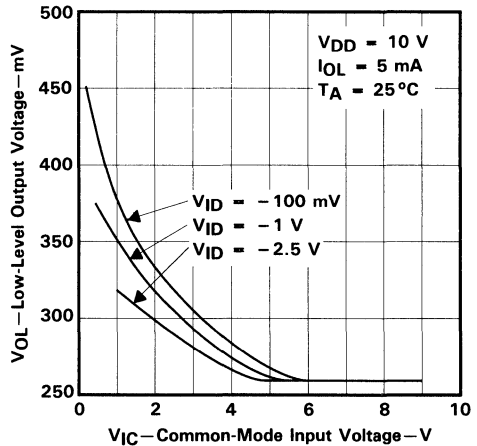


FIGURE 15

LOW-LEVEL OUTPUT VOLTAGE
 vs
 DIFFERENTIAL INPUT VOLTAGE

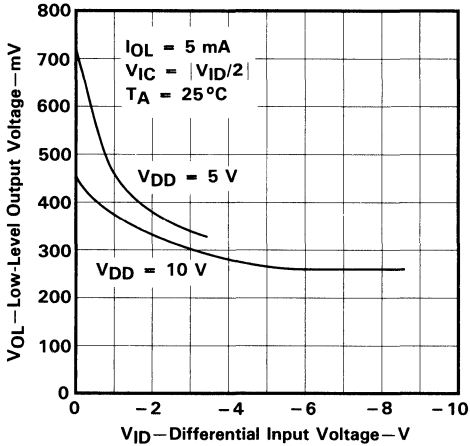


FIGURE 16

LOW-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

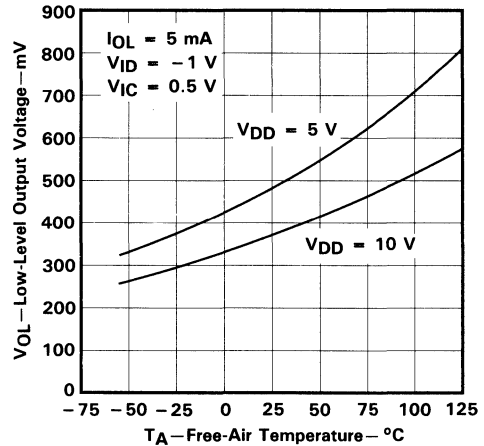


FIGURE 17

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC27L2, TLC27L2A, TLC27L2B, TLC27L7
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

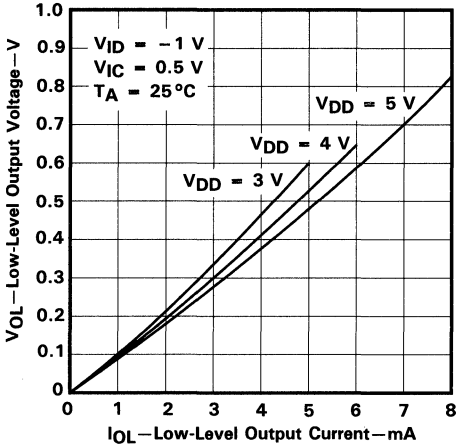


FIGURE 18

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

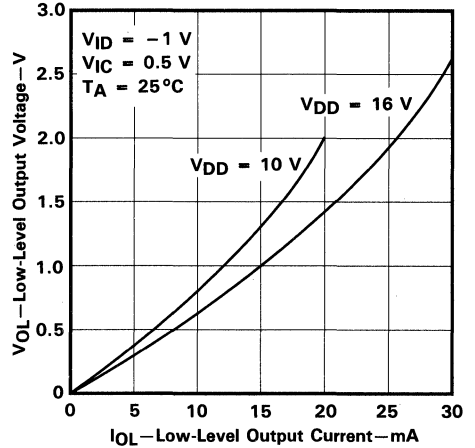


FIGURE 19

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 SUPPLY VOLTAGE

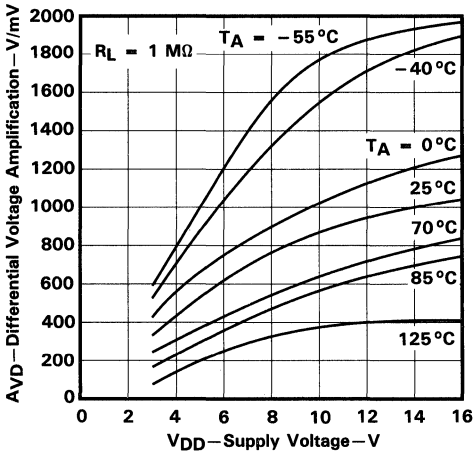


FIGURE 20

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 FREE-AIR TEMPERATURE

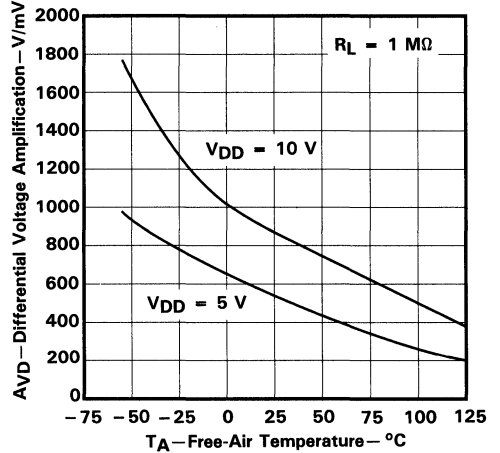


FIGURE 21

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

INPUT BIAS CURRENT AND INPUT OFFSET CURRENT
 vs
 FREE-AIR TEMPERATURE

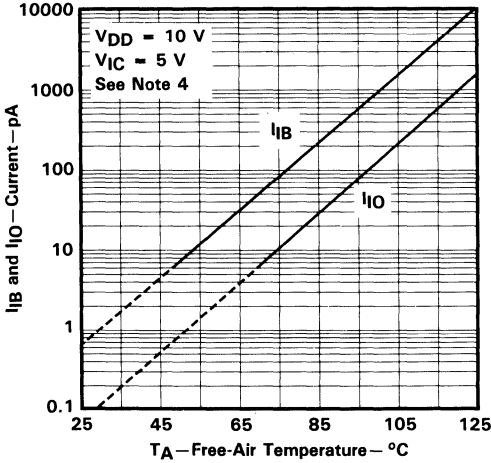


FIGURE 22

COMMON-MODE
 INPUT VOLTAGE POSITIVE LIMIT
 vs
 SUPPLY VOLTAGE

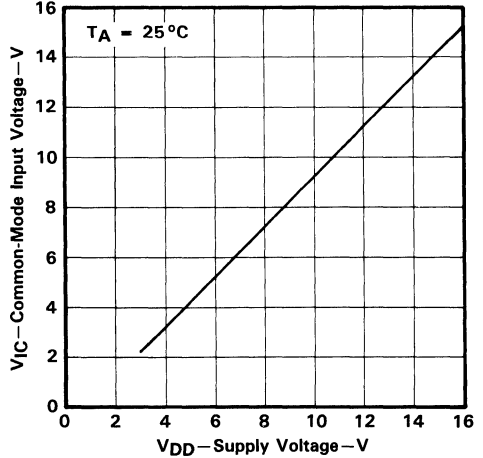


FIGURE 23

SUPPLY CURRENT
 vs
 SUPPLY VOLTAGE

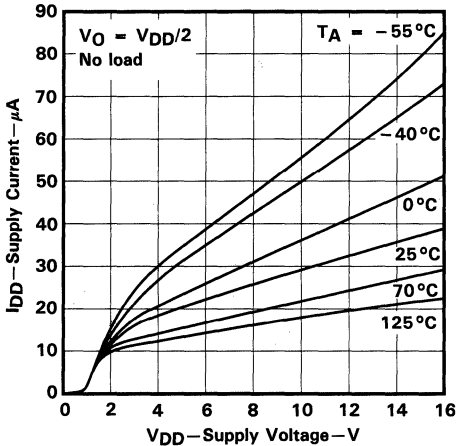


FIGURE 24

SUPPLY CURRENT
 vs
 FREE-AIR TEMPERATURE

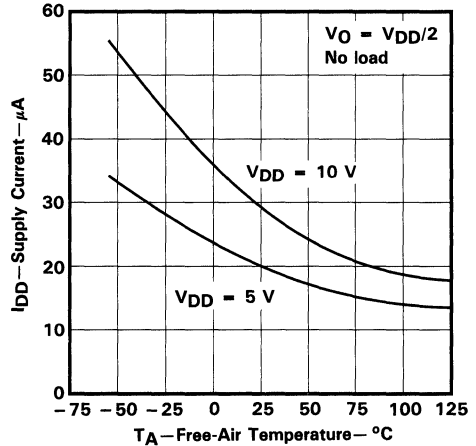


FIGURE 25

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.
 NOTE 4: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

TYPICAL CHARACTERISTICS†

SLEW RATE
 vs
 SUPPLY VOLTAGE

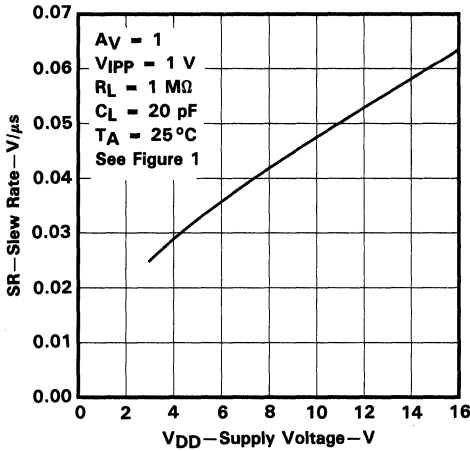


FIGURE 26

SLEW RATE
 vs
 FREE-AIR TEMPERATURE

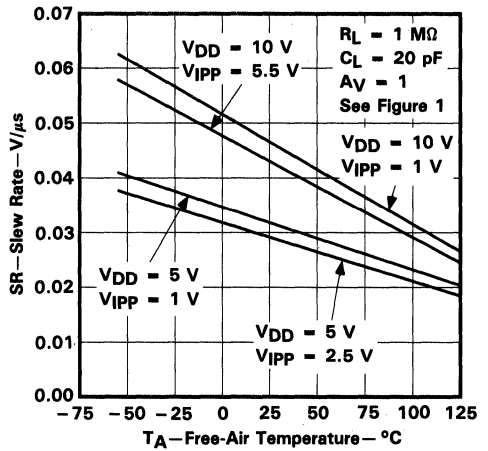


FIGURE 27

NORMALIZED SLEW RATE
 vs
 FREE-AIR TEMPERATURE

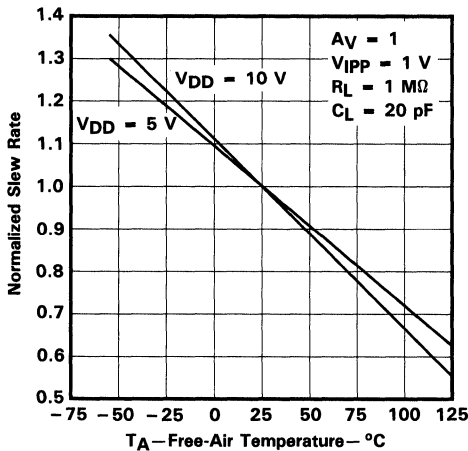


FIGURE 28

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 vs
 FREQUENCY

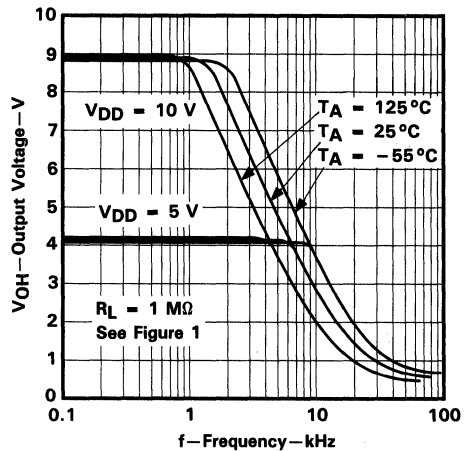


FIGURE 29

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC27L2, TLC27L2A, TLC27L2B, TLC27L7
 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

UNITY-GAIN BANDWIDTH
 vs
 FREE-AIR TEMPERATURE

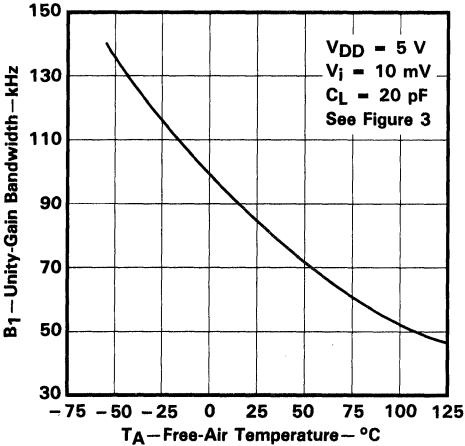


FIGURE 30

UNITY-GAIN BANDWIDTH
 vs
 SUPPLY VOLTAGE

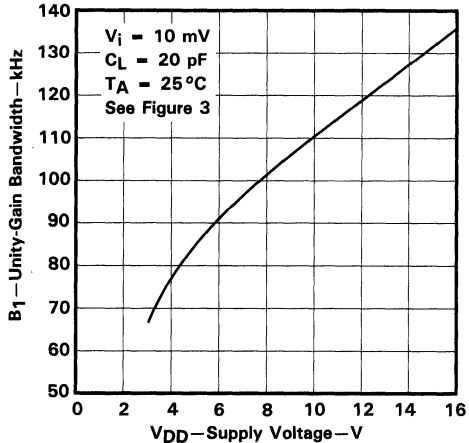


FIGURE 31

2
 Operational Amplifiers

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

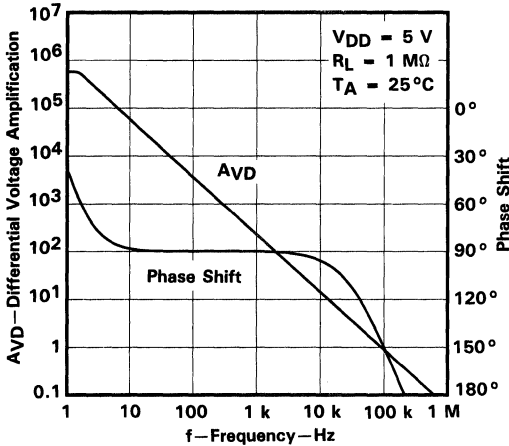


FIGURE 32

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

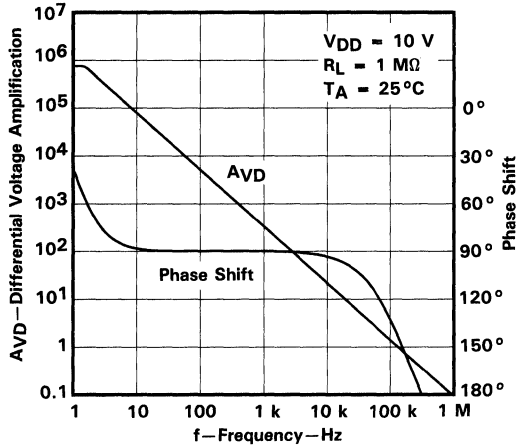


FIGURE 33

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

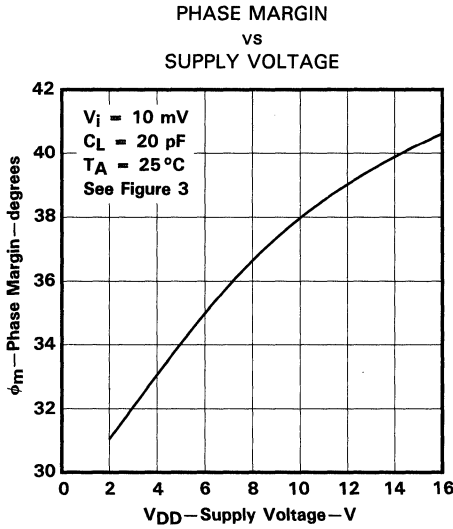


FIGURE 34

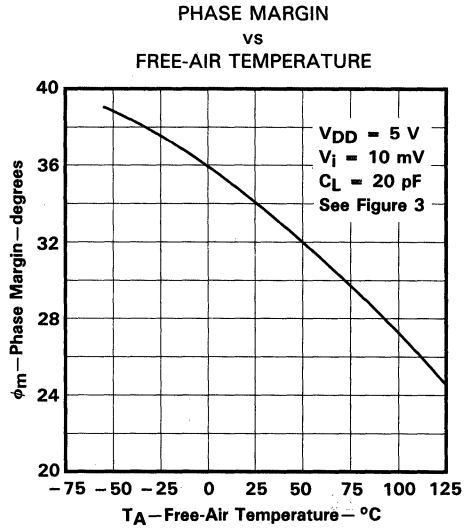


FIGURE 35

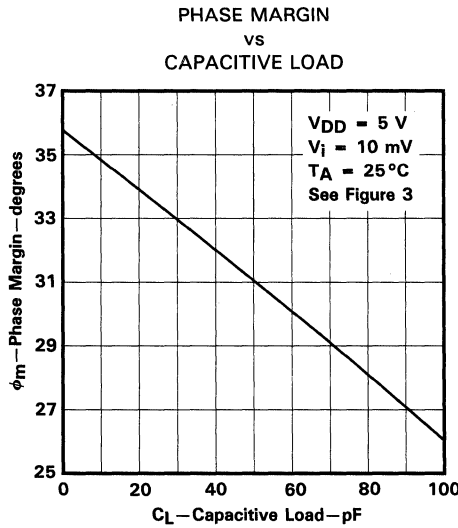


FIGURE 36

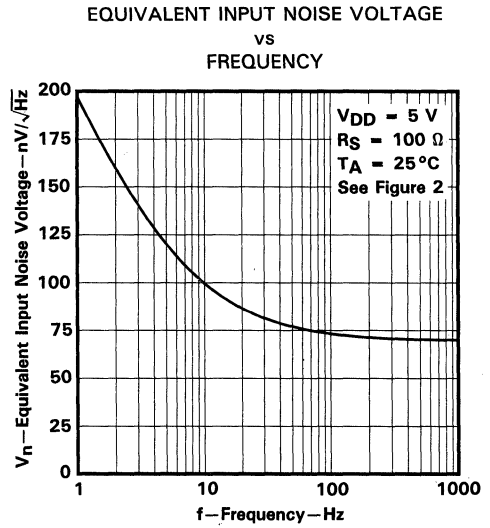


FIGURE 37

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL APPLICATION DATA

single-supply operation

While the TLC27L2 and TLC27L7 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC27L2 and TLC27L7 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC27L2 and TLC27L7 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

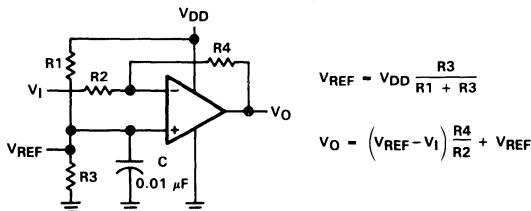


FIGURE 38. INVERTING AMPLIFIER WITH VOLTAGE REFERENCE

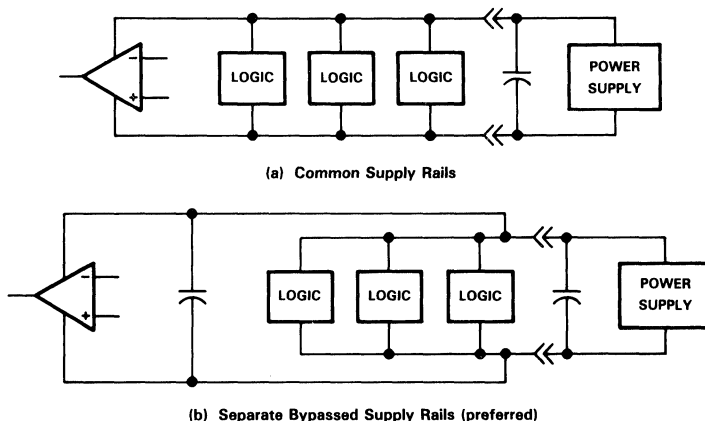


FIGURE 39. COMMON VS SEPARATE SUPPLY RAILS

TYPICAL APPLICATION DATA

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Operational Amplifiers

input characteristics

The TLC27L2 and TLC27L7 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at $V_{DD} - 1\text{ V}$ at $T_A = 25^\circ\text{C}$ and at $V_{DD} - 1.5\text{ V}$ at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC27L2 and TLC27L7 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically $0.1\ \mu\text{V}/\text{month}$, including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC27L2 and TLC27L7 are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.

noise performance

The noise specifications in op amp circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC27L2 and TLC27L7 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than $50\text{ k}\Omega$, since bipolar devices exhibit greater noise currents.

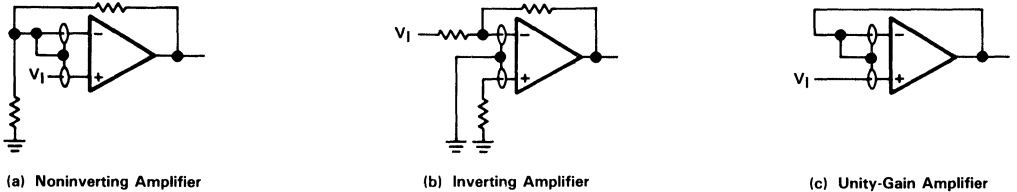


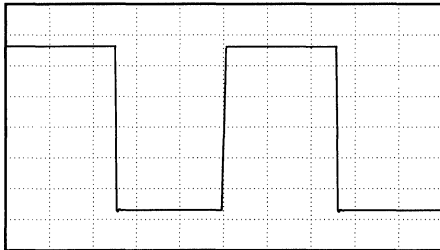
FIGURE 40. GUARD-RING SCHEMES

output characteristics

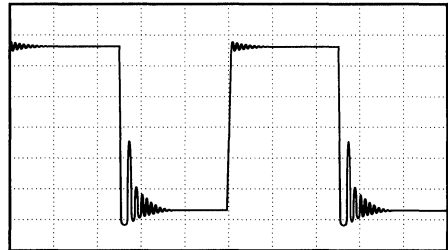
The output stage of the TLC27L2 and TLC27L7 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

All operating characteristics of the TLC27L2 and TLC27L7 were measured using a 20-pF load. The devices will drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance will alleviate the problem.

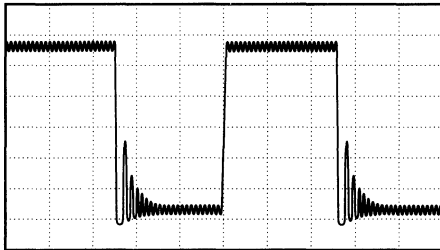
TYPICAL APPLICATION DATA



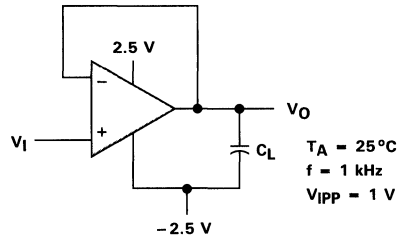
(a) $C_L = 20 \text{ pF}$, $R_L = \text{No load}$



(b) $C_L = 260 \text{ pF}$, $R_L = \text{No load}$



(c) $C_L = 310 \text{ pF}$, $R_L = \text{No load}$



(d) Test Circuit

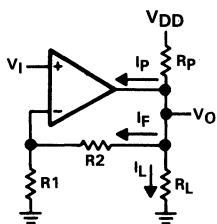
FIGURE 41. EFFECT OF CAPACITIVE LOADS AND TEST CIRCUIT

Although the TLC27L2 and TLC27L7 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor (R_p) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pull-down transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately 60Ω and 180Ω , depending on how hard the op amp input is driven. With very low values of R_p , a voltage offset from 0 V at the output will occur. Second, pullup resistor R_p acts as a drain load to N4 and the gain of the op amp is reduced at output voltage levels where N5 is not supplying the output current.

feedback

Op amp circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

TYPICAL APPLICATION DATA



$$R_p = \frac{V_{DD} - V_O}{I_F + I_L + I_p}$$

I_p = Pullup current required by the op amp (typically 500 μ A)

FIGURE 42. RESISTIVE PULLUP TO INCREASE VOH

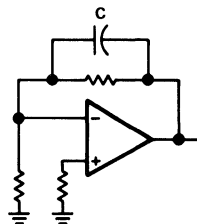


FIGURE 43. COMPENSATION FOR INPUT CAPACITANCE

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Operational Amplifiers

electrostatic discharge protection

The TLC27L2 and TLC27L7 incorporate an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices, as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

latchup

Because CMOS devices are susceptible to latchup due to their inherent parasitic thyristors, the TLC27L2 and TLC27L7 inputs and outputs were designed to withstand – 100-mA surge currents without sustaining latchup; however, techniques should be used to reduce the chance of latchup whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μ F typical) located across the supply rails as close to the device as possible.

The current path established if latchup occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latchup occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latchup occurring increases with increasing temperature and supply voltages.

TLC27L2, TLC27L2A, TLC27L2B, TLC27L7
 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

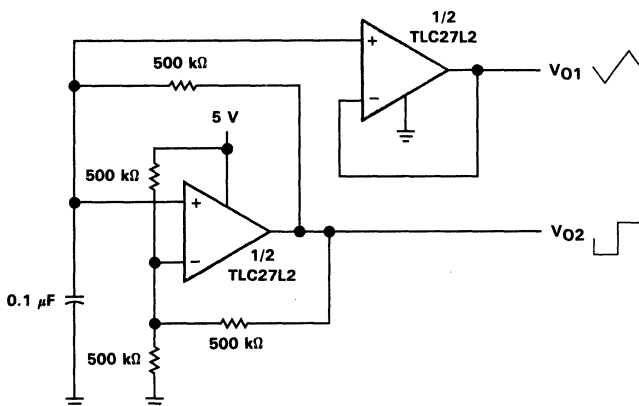
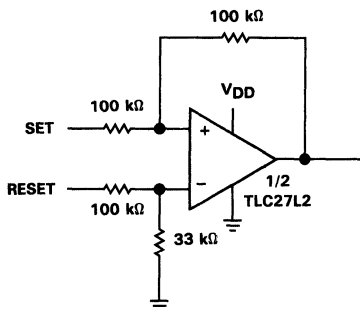


FIGURE 44. MULTIVIBRATOR



NOTE: $V_{DD} = 5 \text{ V to } 16 \text{ V}$

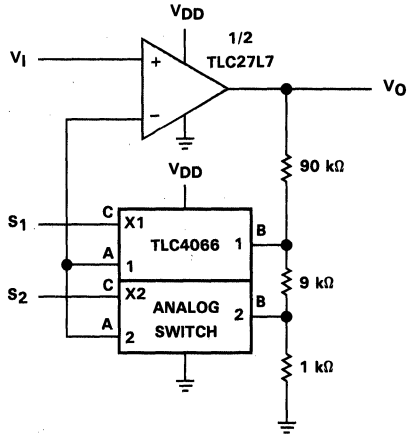
FIGURE 45. SET/RESET FLIP-FLOP

TYPICAL APPLICATION DATA

2

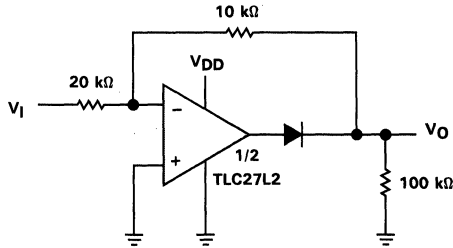
Operational Amplifiers

SELECT:	S ₁	S ₂
A _V	10	100



NOTE: V_{DD} = 5 V to 12 V

FIGURE 46. AMPLIFIER WITH DIGITAL GAIN SELECTION

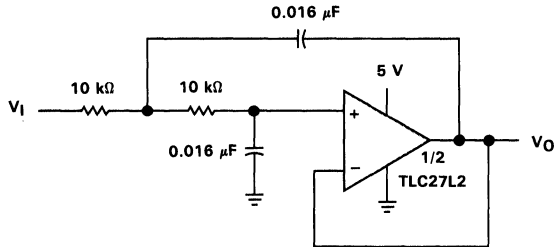


NOTE: V_{DD} = 5 V to 16 V

FIGURE 47. FULL-WAVE RECTIFIER

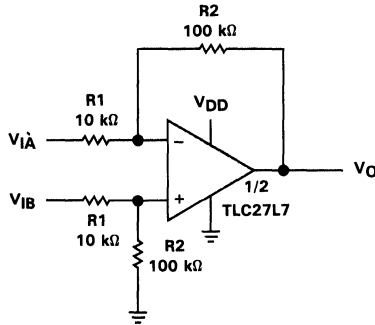
TLC27L2, TLC27L2A, TLC27L2B, TLC27L7
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TYPICAL APPLICATION DATA



NOTE: Normalized to $F_C = 1$ kHz and $R_L = 10$ kΩ

FIGURE 48. TWO-POLE LOW-PASS BUTTERWORTH FILTER



NOTES: $V_{DD} = 5$ V to 16 V

$$V_O = \frac{R_2}{R_1} (V_{IB} - V_{IA})$$

FIGURE 49. DIFFERENCE AMPLIFIER

2

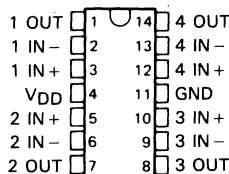
Operational Amplifiers

TLC27L4, TLC27L4A, TLC27L4B, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

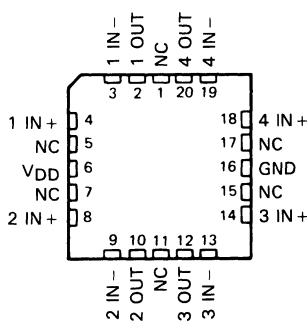
D3142, OCTOBER 1987—REVISED AUGUST 1988

- **Trimmed Offset Voltage:**
TLC27L9 . . . 900 μV Max at 25°C,
V_{DD} = 5 V
- **Input Offset Voltage Drift . . . Typically**
0.1 $\mu\text{V}/\text{Month}$, Including the First 30 Days
- **Wide Range of Supply Voltages over Specified Temperature Range:**
–55°C to 125°C . . . 4 V to 16 V
–40°C to 85°C . . . 4 V to 16 V
0°C to 70°C . . . 3 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range**
Extends Below the Negative Rail (C-Suffix,
I-Suffix types)
- **Ultra-Low Power . . . Typically 195 μW**
at 25°C, V_{DD} = 5 V
- **Output Voltage Range Includes Negative Rail**
- **High Input Impedance . . . 10¹² Ω Typical**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available**
in Tape and Reel
- **Designed-In Latchup Immunity**

D, J, OR N PACKAGE
(TOP VIEW)



FK PACKAGE
(TOP VIEW)



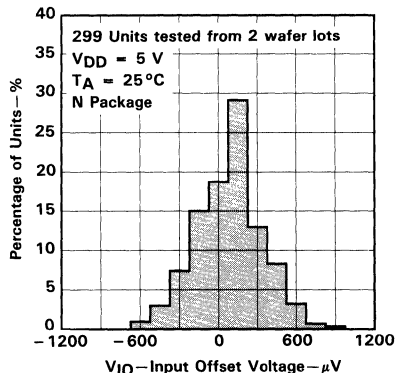
NC—No internal connection

AVAILABLE OPTIONS

T _A	V _{IO} max at 25°C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	900 μV	TLC27L9CD	—	TLC27L9CJ	TLC27L9CN
	2 mV	TLC27L4BCD	—	TLC27L4BCJ	TLC27L4BCN
	5 mV	TLC27L4ACD	—	TLC27L4ACJ	TLC27L4ACN
	10 mV	TLC27L4CD	—	TLC27L4CJ	TLC27L4CN
–40°C to 85°C	900 μV	TLC27L9ID	—	TLC27L9IJ	TLC27L9IN
	2 mV	TLC27L4BID	—	TLC27L4BIJ	TLC27L4BIN
	5 mV	TLC27L4AID	—	TLC27L4AIJ	TLC27L4AIN
	10 mV	TLC27L4ID	—	TLC27L4IJ	TLC27L4IN
–55°C to 125°C	900 μV	—	TLC27L9MFK	TLC27L9MJ	—
	10 mV	—	TLC27L4MFK	TLC27L4MJ	—

The D package is available in tape and reel. Add R suffix to the device type (e.g., TLC27L9CDR).

DISTRIBUTION OF TLC27L9
INPUT OFFSET VOLTAGE



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TLC27L4, TLC27L4A, TLC27L4B, TLC27L9

LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

description

The TLC27L4 and TLC27L9 quad operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, extremely low power, and high gain.

These devices use Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, and low power consumption make these cost-effective devices ideal for high gain, low frequency, low power applications. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC27L4 (10 mV) to the high-precision TLC27L9 (900 μ V). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC27L4 and TLC27L9. The devices also exhibit low voltage single-supply operation and ultra-low power consumption, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip carrier versions for high-density system applications.

The device inputs and outputs are designed to withstand -100 -mA surge currents without sustaining latchup.

The TLC27L4 and TLC27L9 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . I-suffix devices are characterized for operation from -40°C to 85°C , and C-suffix devices are characterized for operation from 0°C to 70°C .

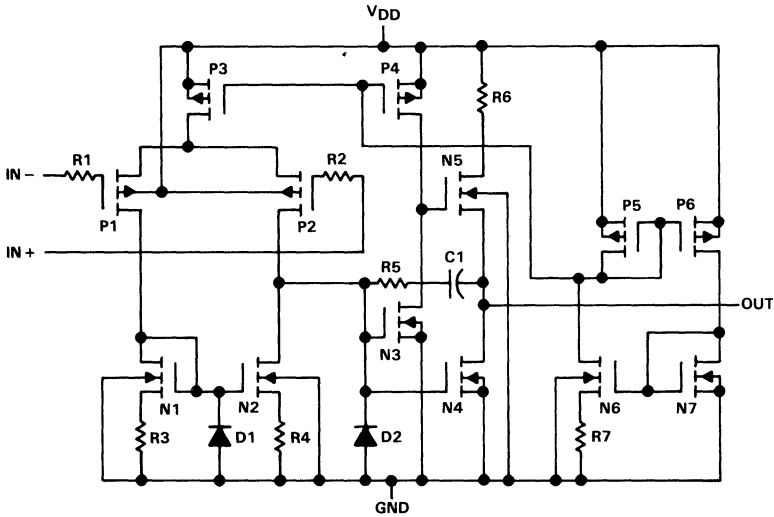
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Operational Amplifiers



TLC27L4, TLC27L4A, TLC27L4B, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

equivalent schematic (each amplifier)



TLC27L4, TLC27L4A, TLC27L4B, TLC27L9
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{DD} (see Note 1)	18 V
Differential input voltage (see Note 2)	$\pm V_{DD}$
Input voltage range, V_I (any input)	-0.3 V to V_{DD}
Input current, I_I	± 5 mA
Output current, I_O (each output)	± 30 mA
Total current into V_{DD} terminal	45 mA
Total current out of ground terminal	45 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	Unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, T_A : M-suffix	-55°C to 125°C
I-suffix	-40°C to 85°C
C-suffix	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D and N package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.
 2. Differential voltages are at the noninverting input with respect to the inverting input.
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING		POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	
FK	1375 mW	11 mW/°C	880 mW	715 mW	275 mW
J (M-suffix)	1375 mW	11 mW/°C	880 mW	715 mW	275 mW
J (C-, I-suffix)	1025 mW	8.2 mW/°C	656 mW	533 mW	
N	1575 mW	12.6 mW/°C	1008 mW	819 mW	

recommended operating conditions

		M-SUFFIX		I-SUFFIX		C-SUFFIX		UNIT			
		MIN	NOM	MAX	MIN	NOM	MAX				
Supply voltage, V_{DD}		4		16	4		16	3		16	V
Common-mode input voltage, V_{IC}	$V_{DD} = 5$ V	0		3.5	-0.2		3.5	-0.2		3.5	V
	$V_{DD} = 10$ V	0		8.5	-0.2		8.5	-0.2		8.5	V
Operating free-air temperature, T_A		-55		125	-40		85	0		70	°C

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Operational Amplifiers

TLC27L4M, TLC27L9M LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A †	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27L4M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $R_L = 1\text{ M}\Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
				Full range			12	
		TLC27L9M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $R_L = 1\text{ M}\Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	200	900	μV
				Full range			3750	
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 125°C		1.4		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$, $V_{IC} = 2.5\text{ V}$		25°C		0.1		pA
				125°C		1.4	15	nA
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$, $V_{IC} = 2.5\text{ V}$		25°C		0.6		pA
				125°C		9	35	nA
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	0 to 4	-0.3 to 4.2		V
				Full range	0 to 3.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$, $R_L = 1\text{ M}\Omega$		25°C	3.2	4.1		V
				-55°C	3	4.1		
				125°C	3	4.2		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$, $I_{OL} = 0$		25°C		0	50	mV
				-55°C		0	50	
				125°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$, $R_L = 1\text{ M}\Omega$		25°C	50	480		V/mV
				-55°C	25	950		
				125°C	25	200		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min}$		25°C	65	94		dB
				-55°C	60	95		
				125°C	60	85		
kSVR	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$, $V_O = 1.4\text{ V}$		25°C	70	97		dB
				-55°C	60	97		
				125°C	60	98		
I_{DD}	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$, No load	$V_{IC} = 2.5\text{ V}$	25°C		39	68	μA
				-55°C		69	120	
				125°C		27	48	

† Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

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Operational Amplifiers

TLC27L4M, TLC27L9M
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27L4M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$,	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
	TLC27L9M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$,	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	210	1200	μV	
				Full range		4300		
αV_{IO}	Average temperature coefficient of input offset voltage			25°C to 125°C	1.4		$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				125°C	1.8	15	nA	
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				125°C	10	35	nA	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	0 to 9	-0.3 to 9.2	V	
				Full range	0 to 8.5		V	
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$,	$R_L = 1\text{ M}\Omega$	25°C	8	8.9	V	
				-55°C	7.8	8.8		
				125°C	7.8	9		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$,	$I_{OL} = 0$	25°C	0	50	mV	
				-55°C	0	50		
				125°C	0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$,	$R_L = 1\text{ M}\Omega$	25°C	50	800	V/mV	
				-55°C	25	1750		
				125°C	25	380		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min}$		25°C	65	97	dB	
				-55°C	60	97		
				125°C	60	91		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$,	$V_O = 1.4\text{ V}$	25°C	70	97	dB	
				-55°C	60	97		
				125°C	60	98		
I_{DD}	Supply current (four amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$,	25°C	57	92	μA	
				-55°C	111	192		
				125°C	35	60		

† Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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Operational Amplifiers

TLC27L4I, TLC27L4AI, TLC27L4BI, TLC27L9I

LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27L4I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		13	
	TLC27L4AI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	0.9	5	mV	
				Full range		7		
	TLC27L4BI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	240	2000	μV	
				Full range		3500		
	TLC27L9I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	200	900	μV	
				Full range		2000		
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 85°C		1.1		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C		0.1		pA
				85°C		24	1000	
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C		0.6		pA
				85°C		200	2000	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2		V
				Full range	-0.2 to 3.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	25°C	3.2	4.1		V
				-40°C	3	4.1		
				85°C	3	4.2		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C		0	50	mV
				-40°C		0	50	
				85°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$	$R_L = 1\text{ M}\Omega$	25°C	50	480		V/mV
				-40°C	50	900		
				85°C	50	330		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	94		dB
				-40°C	60	95		
				85°C	60	95		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to } 10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	97		dB
				-40°C	60	97		
				85°C	60	98		
I_{DD}	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$, No load	$V_{IC} = 2.5\text{ V}$	25°C		39	68	μA
				-40°C		62	108	
				85°C		29	52	

† Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

2

Operational Amplifiers

TLC27L4I, TLC27L4AI, TLC27L4BI, TLC27L9I

LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A †	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27L4I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
				Full range		13		
		TLC27L4AI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	0.9	5	
				Full range		7		
	TLC27L4BI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	260	2000	μV	
				Full range		3500		
		TLC27L9I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	210		1200
					Full range			2900
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 85°C		1		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C		0.1		pA
				85°C		26	1000	
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C		0.7		pA
				85°C		220	2000	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2		V
				Full range	-0.2 to 8.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$,	$R_L = 1\text{ M}\Omega$	25°C	8	8.9		V
				-40°C	7.8	8.9		
				85°C	7.8	8.9		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$,	$I_{OL} = 0$	25°C		0	50	mV
				-40°C		0	50	
				85°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$,	$R_L = 1\text{ M}\Omega$	25°C	50	800		V/mV
				-40°C	50	1550		
				85°C	50	585		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	97		dB
				-40°C	60	97		
				85°C	60	98		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$,	$V_O = 1.4\text{ V}$	25°C	70	97		dB
				-40°C	60	97		
				85°C	60	98		
I_{DD}	Supply current (four amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$	25°C		57	92	μA
				-40°C		98	172	
				85°C		40	72	

† Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

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Operational Amplifiers

TLC27L4C, TLC27L4AC, TLC27L4BC, TLC27L9C LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27L4C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27L4AC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	0.9	5	
					Full range		6.5	
	TLC27L4BC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	240	2000	μV	
				Full range		3000		
	TLC27L9C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	200	900		
				Full range		1500		
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C		1.1		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C		0.1		pA
				70°C		7	300	
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C		0.6		pA
				70°C		40	600	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2		V
				Full range		-0.2 to 3.5		
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	25°C	3.2	4.1		V
				0°C	3	4.1		
				70°C	3	4.2		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C		0	50	mV
				0°C		0	50	
				70°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$	$R_L = 1\text{ M}\Omega$	25°C	50	520		V/mV
				0°C	50	680		
				70°C	50	380		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	94		dB
				0°C	60	95		
				70°C	60	95		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to } 10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	97		dB
				0°C	60	97		
				70°C	60	98		
I_{DD}	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$	No load	$V_{IC} = 2.5\text{ V}$	25°C	40	68	μA
					0°C	48	84	
					70°C	31	56	

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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Operational Amplifiers

TLC27L4C, TLC27L4AC, TLC27L4BC, TLC27L9C
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27L4C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27L4AC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	0.9	5	mV
					Full range		6.5	
TLC27L4BC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	260	2000	μV		
			Full range		3000			
TLC27L9C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 1\text{ M}\Omega$	25°C	210	1200	μV		
	Full range				1900			
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C		1		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C		0.1		pA
				70°C		7	300	
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C		0.7		pA
				70°C		50	600	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2		V
				Full range	-0.2 to 8.5		V	
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	25°C	8	8.9		V
				0°C	7.8	8.9		
				70°C	7.8	8.9		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C		0	50	mV
				0°C		0	50	
				70°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 1\text{ M}\Omega$	25°C	50	870		V/mV
				0°C	50	1020		
				70°C	50	660		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	97		dB
				0°C	60	97		
				70°C	60	97		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	97		dB
				0°C	60	97		
				70°C	60	98		
I_{DD}	Supply current (four amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$	25°C		57	92	μA
				0°C		72	132	
				70°C		44	80	

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
 5. This range also applies to each input individually.

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Operational Amplifiers

TLC27L4M, TLC27L9M
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

operating characteristics, V_{DD} = 5 V

PARAMETER	TEST CONDITIONS		T _A	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	R _L = 1 MΩ, C _L = 20 pF, See Figure 1	V _{IPP} = 1 V	25 °C		0.03		V/μs
			-55 °C		0.04		
			125 °C		0.02		
		V _{IPP} = 2.5 V	25 °C		0.03		
			-55 °C		0.04		
			125 °C		0.02		
V _n Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 100 Ω,	25 °C		70		nV/√Hz
B _{OM} Maximum output swing bandwidth	V _O = V _{OH} , R _L = 1 MΩ, See Figure 1	C _L = 20 pF, See Figure 1	25 °C		5		kHz
			-55 °C		8		
			125 °C		3		
B ₁ Unity-gain bandwidth	V _i = 10 mV, See Figure 3	C _L = 20 pF,	25 °C		85		kHz
			-55 °C		140		
			125 °C		45		
φ _m Phase margin	V _i = 10 mV, C _L = 20 pF,	f = B ₁ , See Figure 3	25 °C		34 °		
			-55 °C		39 °		
			125 °C		25 °		

operating characteristics, V_{DD} = 10 V

PARAMETER	TEST CONDITIONS		T _A	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	R _L = 1 MΩ, C _L = 20 pF, See Figure 1	V _{IPP} = 1 V	25 °C		0.05		V/μs
			-55 °C		0.06		
			125 °C		0.03		
		V _{IPP} = 5.5 V	25 °C		0.04		
			-55 °C		0.06		
			125 °C		0.03		
V _n Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 100 Ω,	25 °C		70		nV/√Hz
B _{OM} Maximum output swing bandwidth	V _O = V _{OH} , R _L = 1 MΩ, See Figure 1	C _L = 20 pF, See Figure 1	25 °C		1		kHz
			-55 °C		1.5		
			125 °C		0.7		
B ₁ Unity-gain bandwidth	V _i = 10 mV, See Figure 3	C _L = 20 pF,	25 °C		110		kHz
			-55 °C		165		
			125 °C		70		
φ _m Phase margin	V _i = 10 mV, C _L = 20 pF,	f = B ₁ , See Figure 3	25 °C		38 °		
			-55 °C		43 °		
			125 °C		29 °		

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Operational Amplifiers

TLC27L4I, TLC27L4AI, TLC27L4BI, TLC27L9I
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C		0.03		V/ μ s
				-40°C		0.04		
				85°C		0.03		
			$V_{IPP} = 2.5\text{ V}$	25°C		0.03		
				-40°C		0.04		
				85°C		0.02		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		70		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C		5		kHz
				-40°C		7		
				85°C		4		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		85		kHz
				-40°C		130		
				85°C		55		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		34°		
				-40°C		38°		
				85°C		28°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C		0.05		V/ μ s
				-40°C		0.06		
				85°C		0.03		
			$V_{IPP} = 5.5\text{ V}$	25°C		0.04		
				-40°C		0.05		
				85°C		0.03		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		70		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C		1		kHz
				-40°C		1.4		
				85°C		0.8		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		110		kHz
				-40°C		155		
				85°C		80		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		38°		
				-40°C		42°		
				85°C		32°		

TLC27L4C, TLC27L4AC, TLC27L4BC, TLC27L9C LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C		0.03		V/ μ s
			0°C		0.04		
			70°C		0.03		
		$V_{Ipp} = 2.5\text{ V}$	25°C		0.03		
			0°C		0.03		
			70°C		0.02		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		70		nV/ $\sqrt{\text{Hz}}$
BOM Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25°C		5		kHz
			0°C		6		
			70°C		4.5		
B ₁ Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		85		kHz
			0°C		100		
			70°C		65		
ϕ_m Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		34°		
			0°C		36°		
			70°C		30°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C		0.05		V/ μ s
			0°C		0.05		
			70°C		0.04		
		$V_{Ipp} = 5.5\text{ V}$	25°C		0.04		
			0°C		0.05		
			70°C		0.04		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		70		nV/ $\sqrt{\text{Hz}}$
BOM Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25°C		1		kHz
			0°C		1.3		
			70°C		0.9		
B ₁ Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		110		kHz
			0°C		125		
			70°C		90		
ϕ_m Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		38°		
			0°C		40°		
			70°C		34°		

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Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC27L4 and TLC27L9 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit will give the same result.

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Operational Amplifiers



FIGURE 1. UNITY-GAIN AMPLIFIER

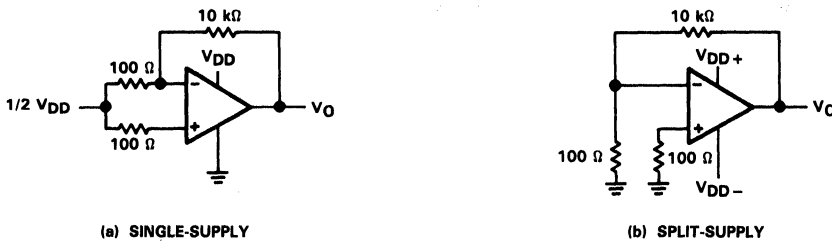


FIGURE 2. NOISE TEST CIRCUIT

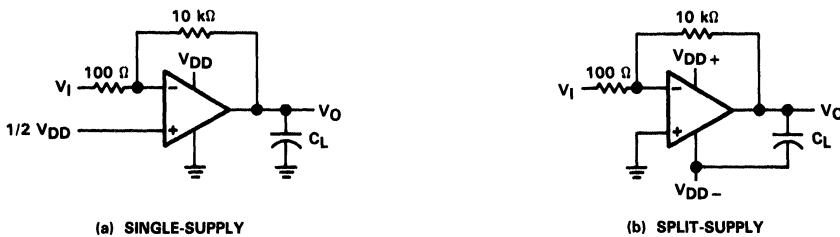


FIGURE 3. GAIN-OF-100 INVERTING AMPLIFIER

PARAMETER MEASUREMENT INFORMATION

input bias current

Because of the high input impedance of the TLC27L4 and TLC27L9 op amps, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs will be shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the "open-socket" leakage readings from the readings obtained with a device in the test socket.

One word of caution . . . many automatic testers as well as some bench-top op amp testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an "open-socket" reading is not feasible using this method.

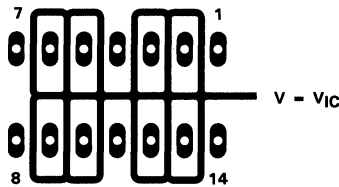


FIGURE 4. ISOLATION METAL AROUND DEVICE INPUTS
(J AND N DUAL-IN-LINE-PACKAGE)

low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture will result in leakage and contact resistance which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

PARAMETER MEASUREMENT INFORMATION

full-power response

Full-power response, the frequency above which the op amp slew rate limits the output voltage swing, is often specified two ways . . . full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for "significant" distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

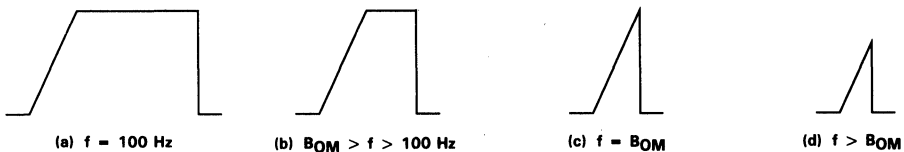


FIGURE 5. FULL-POWER-RESPONSE OUTPUT SIGNAL

test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC27L4
 INPUT OFFSET VOLTAGE

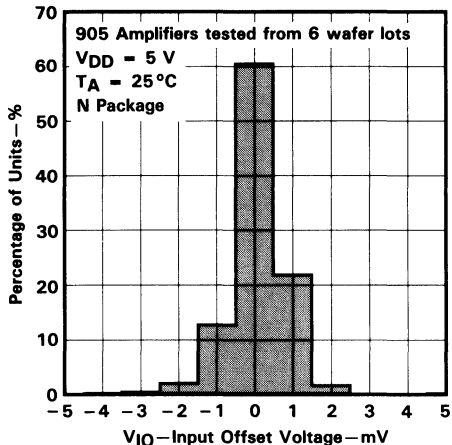


FIGURE 6

DISTRIBUTION OF TLC27L4
 INPUT OFFSET VOLTAGE

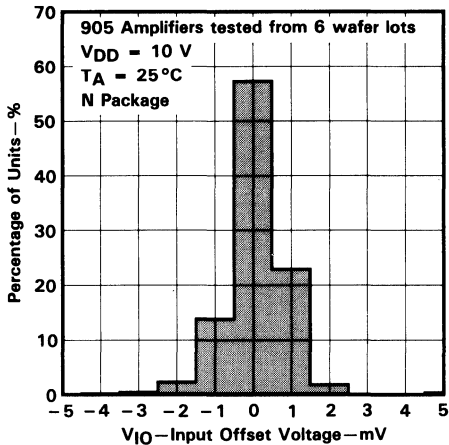


FIGURE 7

DISTRIBUTION OF TLC27L4 AND TLC27L9
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT

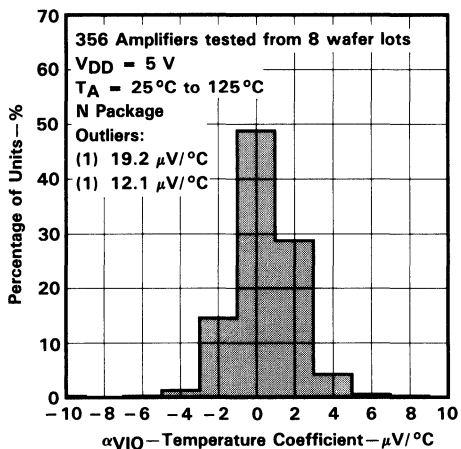


FIGURE 8

DISTRIBUTION OF TLC27L4 AND TLC27L9
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT

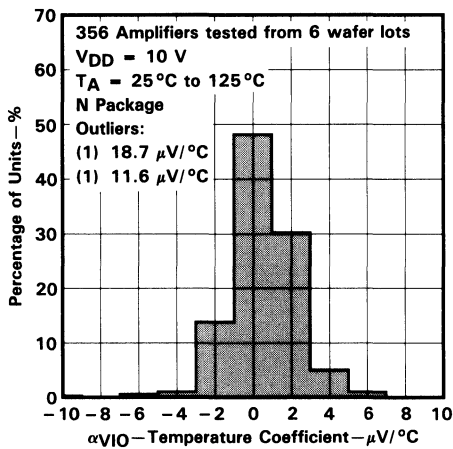


FIGURE 9

TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

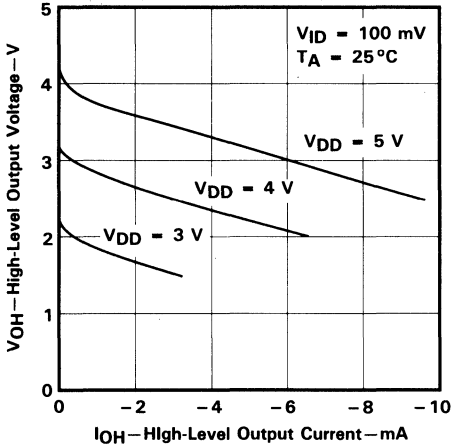


FIGURE 10

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

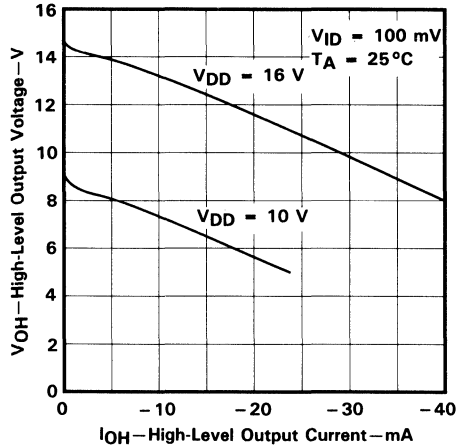


FIGURE 11

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 SUPPLY VOLTAGE

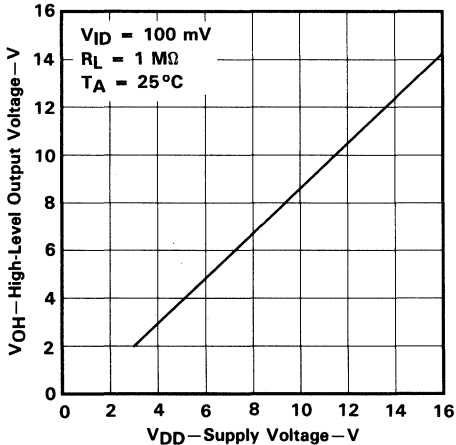


FIGURE 12

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

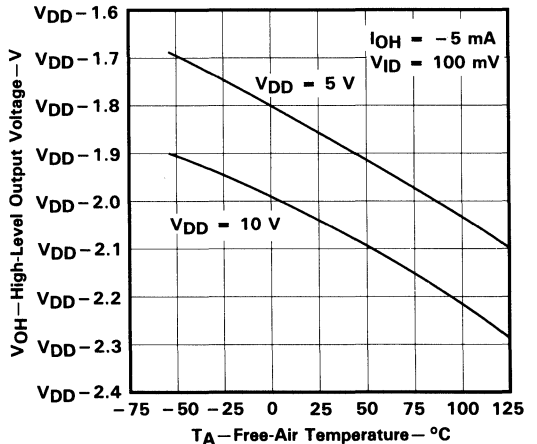


FIGURE 13

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

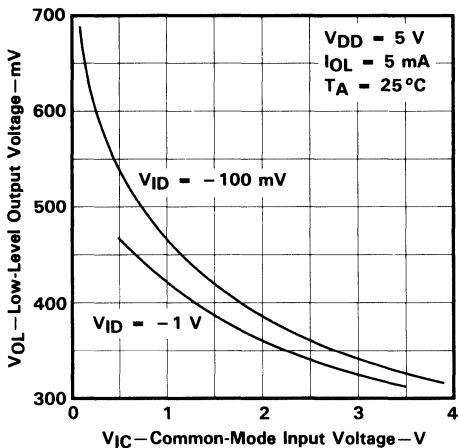


FIGURE 14

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

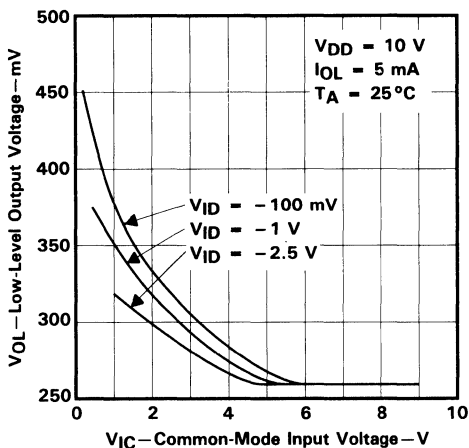


FIGURE 15

LOW-LEVEL OUTPUT VOLTAGE
 vs
 DIFFERENTIAL INPUT VOLTAGE

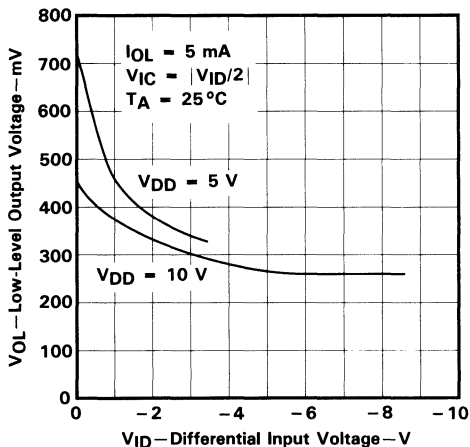


FIGURE 16

LOW-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

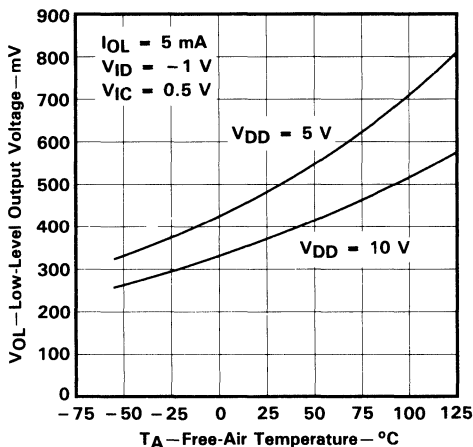


FIGURE 17

2
 Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC27L4, TLC27L4A, TLC27L4B, TLC27L9
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

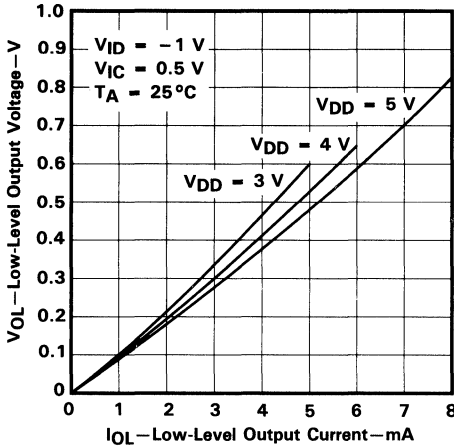


FIGURE 18

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

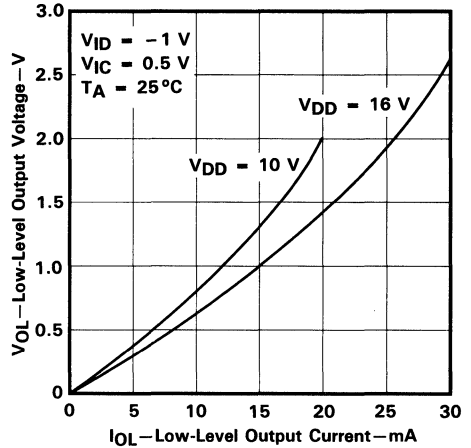


FIGURE 19

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 SUPPLY VOLTAGE

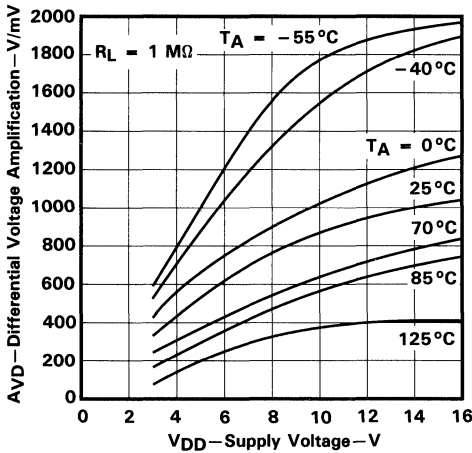


FIGURE 20

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 FREE-AIR TEMPERATURE

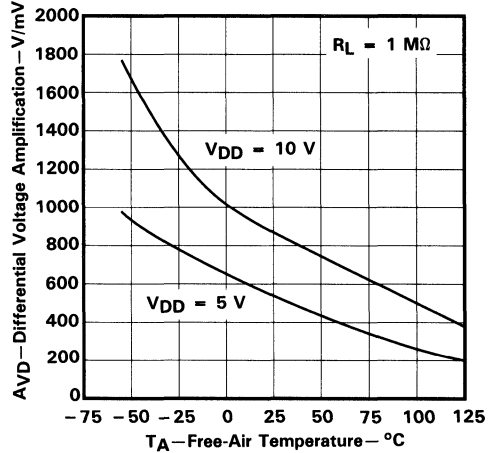


FIGURE 21

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC27L4, TLC27L4A, TLC27L4B, TLC27L9
 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

INPUT BIAS CURRENT AND INPUT OFFSET CURRENT
 vs
 FREE-AIR TEMPERATURE

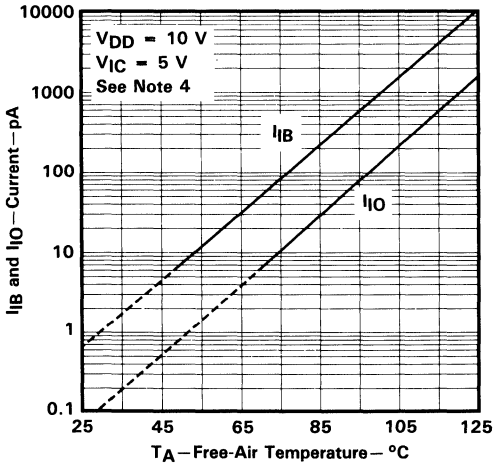


FIGURE 22

COMMON-MODE
 INPUT VOLTAGE POSITIVE LIMIT
 vs
 SUPPLY VOLTAGE

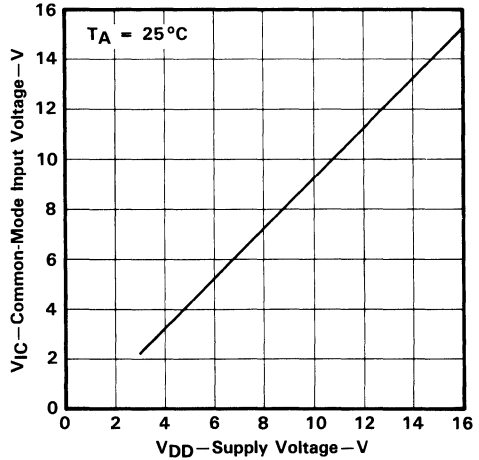


FIGURE 23

SUPPLY CURRENT
 vs
 SUPPLY VOLTAGE

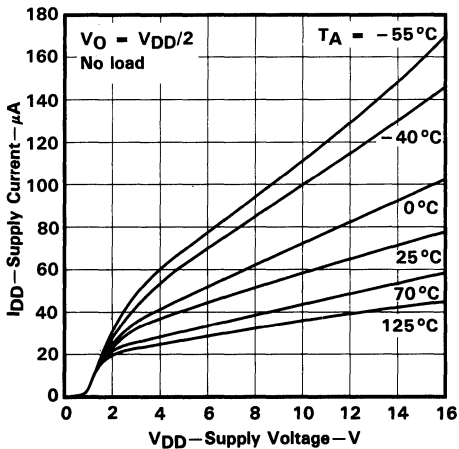


FIGURE 24

SUPPLY CURRENT
 vs
 FREE-AIR TEMPERATURE

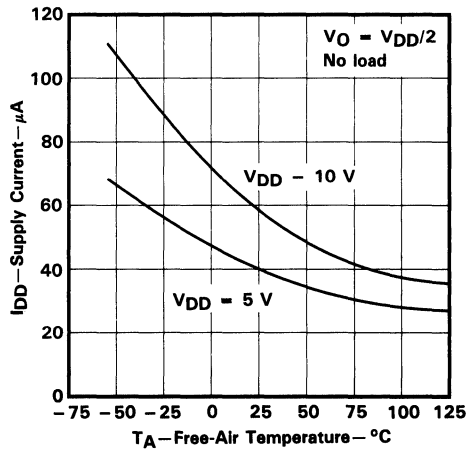
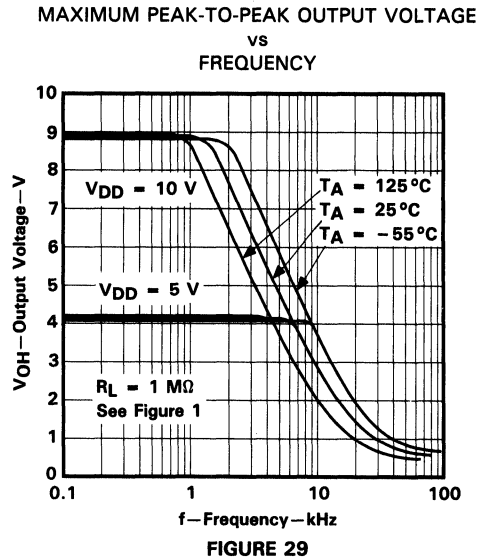
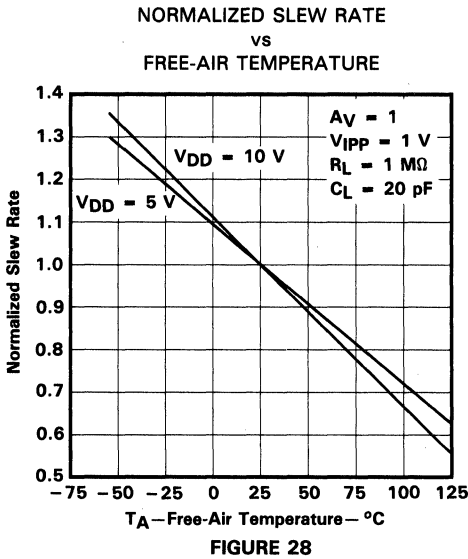
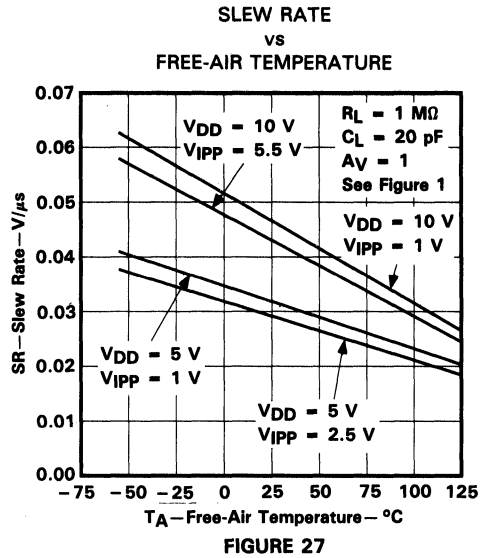
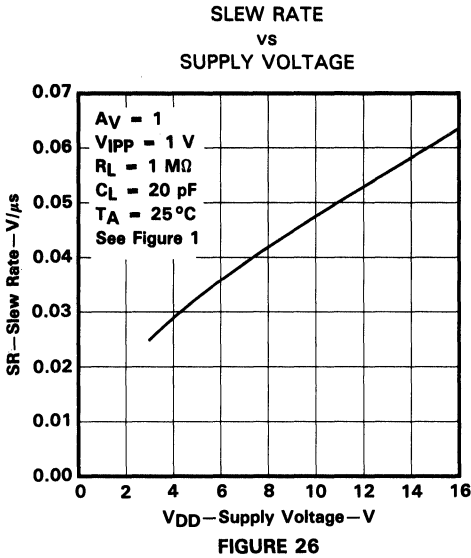


FIGURE 25

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.
 NOTE 4: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

TYPICAL CHARACTERISTICS†

2
Operational Amplifiers



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC27L4, TLC27M2A, TLC27M2B, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

UNITY-GAIN BANDWIDTH
vs
FREE-AIR TEMPERATURE

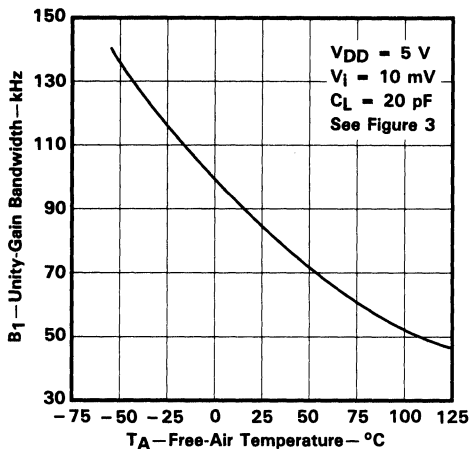


FIGURE 30

UNITY-GAIN BANDWIDTH
vs
SUPPLY VOLTAGE

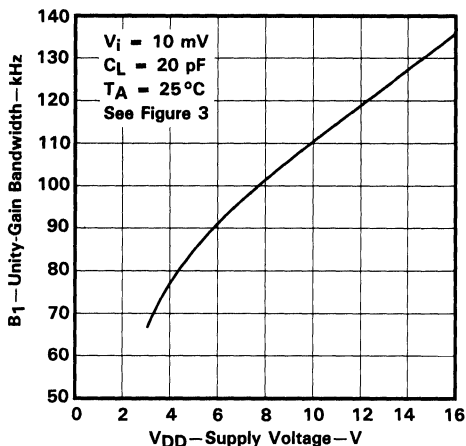


FIGURE 31

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
vs
FREQUENCY

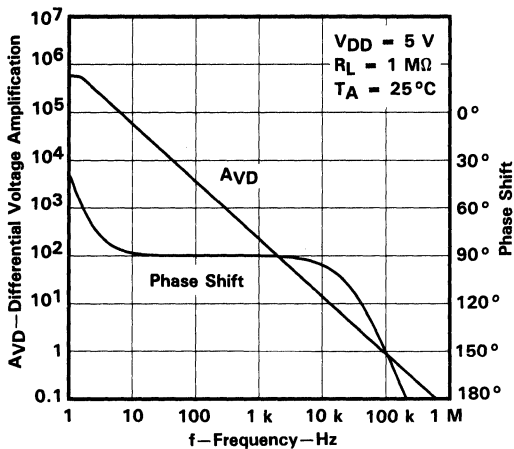


FIGURE 32

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
vs
FREQUENCY

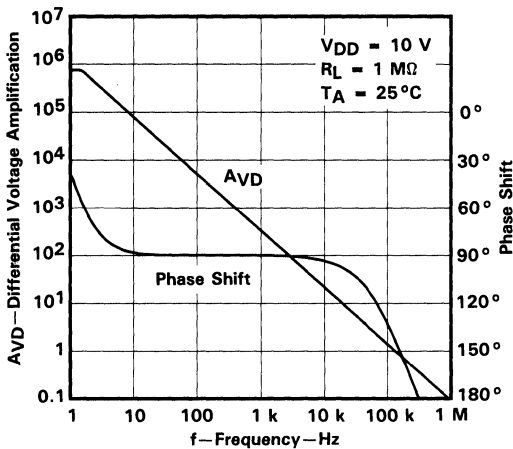


FIGURE 33

2

Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

PHASE MARGIN
 vs
 SUPPLY VOLTAGE

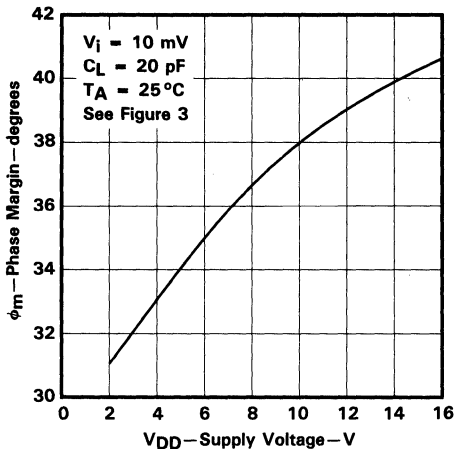


FIGURE 34

PHASE MARGIN
 vs
 FREE-AIR TEMPERATURE

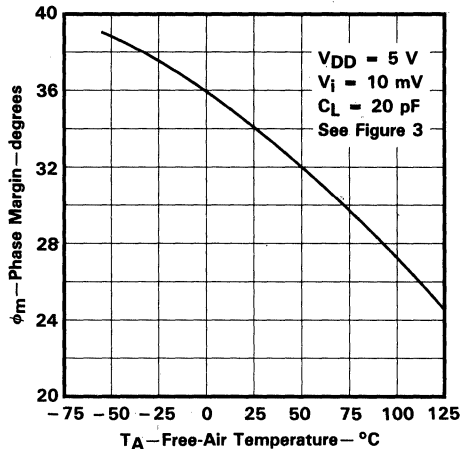


FIGURE 35

PHASE MARGIN
 vs
 CAPACITIVE LOAD

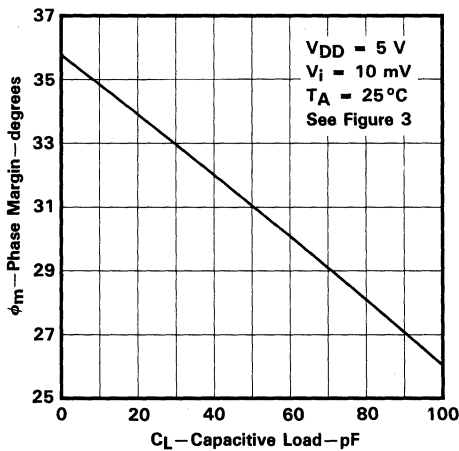


FIGURE 36

EQUIVALENT INPUT NOISE VOLTAGE
 vs
 FREQUENCY

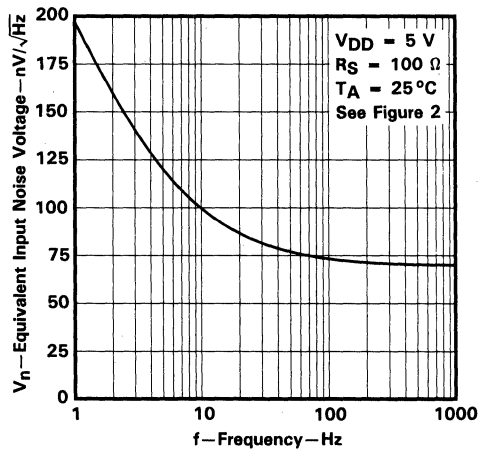


FIGURE 37

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

single-supply operation

While the TLC27L4 and TLC27L9 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC27L4 and TLC27L9 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC27L4 and TLC27L9 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

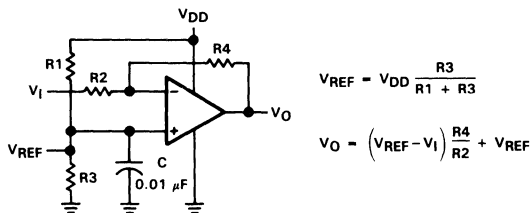


FIGURE 38. INVERTING AMPLIFIER WITH VOLTAGE REFERENCE

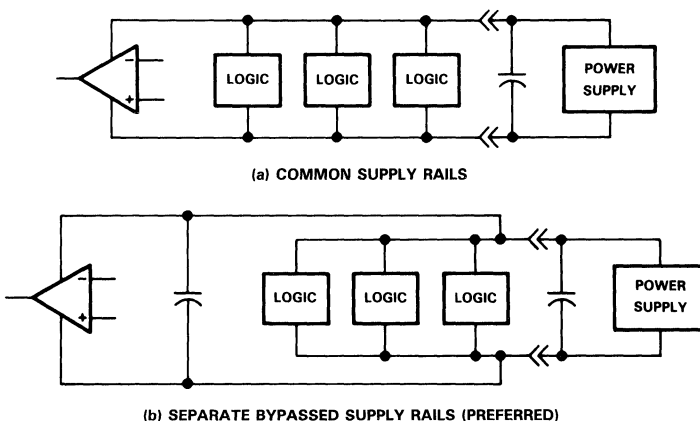


FIGURE 39. COMMON VS SEPARATE SUPPLY RAILS

TYPICAL APPLICATION DATA

input characteristics

The TLC27L4 and TLC27L9 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at $V_{DD} - 1\text{ V}$ at $T_A = 25^\circ\text{C}$ and at $V_{DD} - 1.5\text{ V}$ at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC27L4 and TLC27L9 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically $0.1\ \mu\text{V}/\text{month}$, including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC27L4 and TLC27L9 are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.

noise performance

The noise specifications in op amp circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC27L4 and TLC27L9 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than $50\ \text{k}\Omega$, since bipolar devices exhibit greater noise currents.

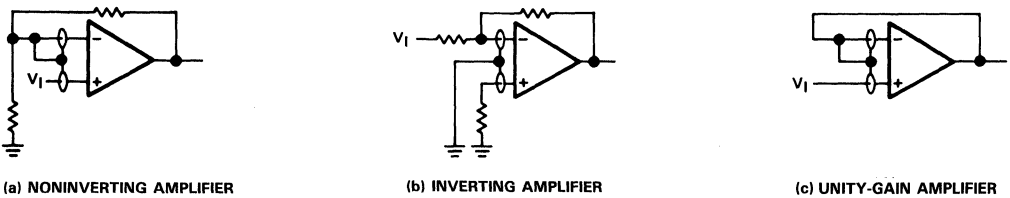


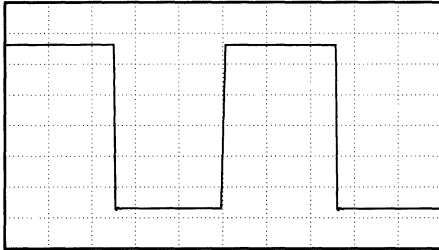
FIGURE 40. GUARD-RING SCHEMES

output characteristics

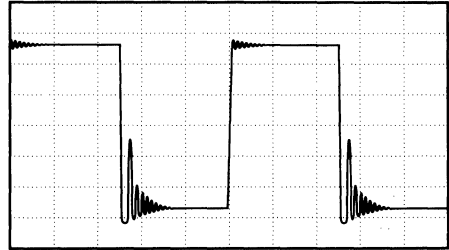
The output stage of the TLC27L4 and TLC27L9 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

All operating characteristics of the TLC27L4 and TLC27L9 were measured using a 20-pF load. The devices will drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance will alleviate the problem.

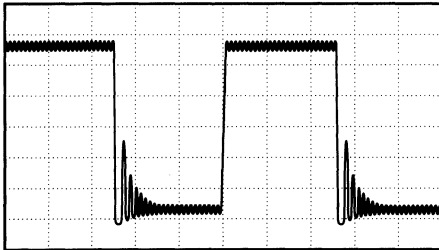
TYPICAL APPLICATION DATA



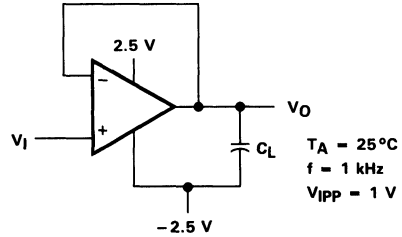
(a) $C_L = 20 \text{ pF}$, $R_L = \text{NO LOAD}$



(b) $C_L = 260 \text{ pF}$, $R_L = \text{NO LOAD}$



(c) $C_L = 310 \text{ pF}$, $R_L = \text{NO LOAD}$



(d) TEST CIRCUIT

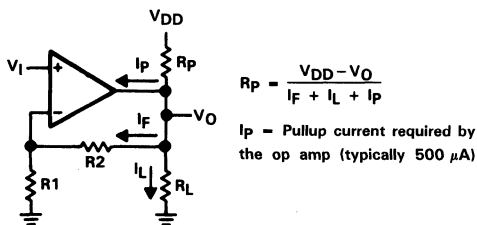
FIGURE 41. EFFECT OF CAPACITIVE LOADS AND TEST CIRCUIT

Although the TLC27L4 and TLC27L9 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor (R_p) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pull-down transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately 60Ω and 180Ω , depending on how hard the op amp input is driven. With very low values of R_p , a voltage offset from 0 V at the output will occur. Second, pullup resistor R_p acts as a drain load to N4 and the gain of the op amp is reduced at output voltage levels where N5 is not supplying the output current.

feedback

Op amp circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

TYPICAL APPLICATION DATA



$$R_p = \frac{V_{DD} - V_{O}}{I_f + I_L + I_p}$$

I_p = Pullup current required by the op amp (typically 500 μ A)

FIGURE 42. RESISTIVE PULLUP TO INCREASE V_{OH}

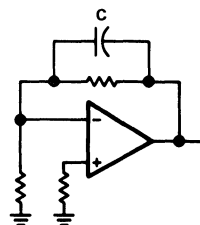


FIGURE 43. COMPENSATION FOR INPUT CAPACITANCE

2

Operational Amplifiers

electrostatic discharge protection

The TLC27L4 and TLC27L9 incorporate an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices, as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

latchup

Because CMOS devices are susceptible to latchup due to their inherent parasitic thyristors, the TLC27L4 and TLC27L9 inputs and outputs were designed to withstand – 100-mA surge currents without sustaining latchup; however, techniques should be used to reduce the chance of latchup whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μ F typical) located across the supply rails as close to the device as possible.

The current path established if latchup occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latchup occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latchup occurring increases with increasing temperature and supply voltages.

TLC27L4, TLC27L4A, TLC27L4B, TLC27L9
 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

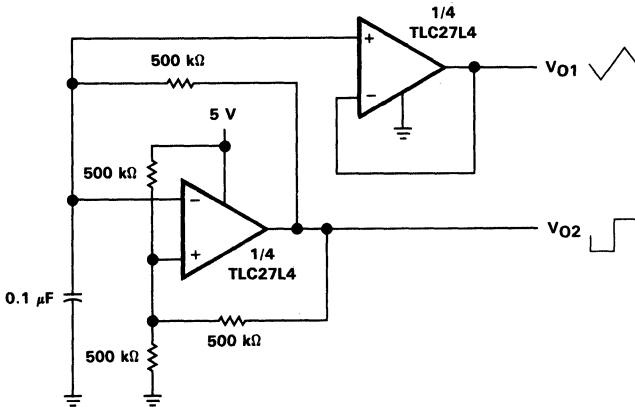
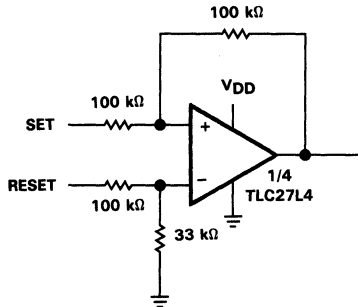


FIGURE 44. MULTIVIBRATOR



NOTE: $V_{DD} = 5\text{ V to }16\text{ V}$

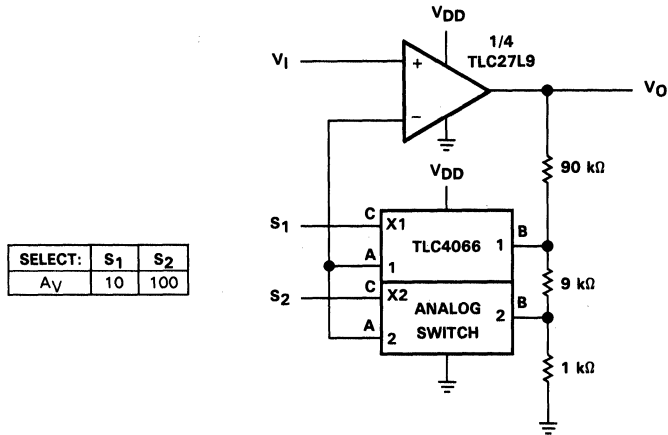
FIGURE 45. SET/RESET FLIP-FLOP

TLC27L4, TLC27L4A, TLC27L4B, TLC27L9
 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

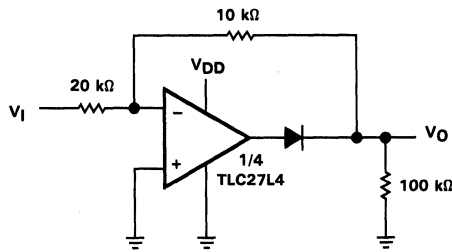
2

Operational Amplifiers



NOTE: V_{DD} = 5 V to 12 V

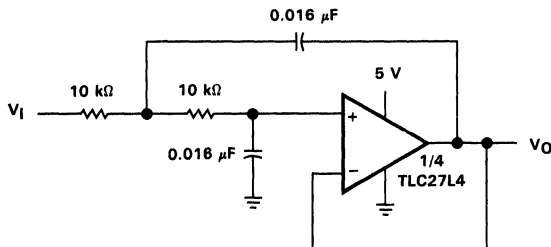
FIGURE 46. AMPLIFIER WITH DIGITAL GAIN SELECTION



NOTE: V_{DD} = 5 V to 16 V

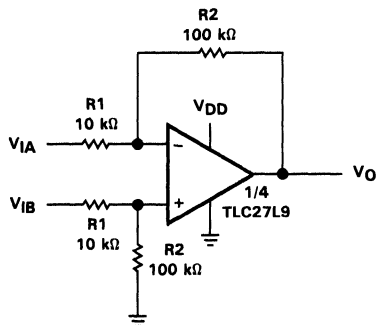
FIGURE 47. FULL-WAVE RECTIFIER

TYPICAL APPLICATION DATA



NOTE: Normalized to $F_C = 1 \text{ kHz}$ and $R_L = 10 \text{ k}\Omega$

FIGURE 48. TWO-POLE LOW-PASS BUTTERWORTH FILTER



NOTES: $V_{DD} = 5 \text{ V to } 16 \text{ V}$

$$V_O = \frac{R_2}{R_1} (V_{IB} - V_{IA})$$

FIGURE 49. DIFFERENCE AMPLIFIER

2

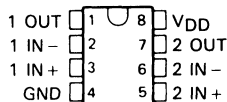
Operational Amplifiers

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

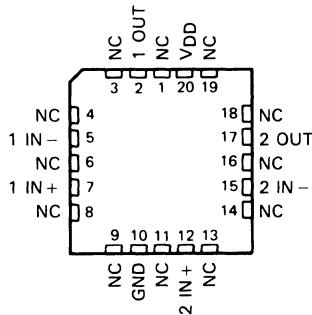
D3140, OCTOBER 1987—REVISED FEBRUARY 1989

- **Trimmed Offset Voltage:**
TLC27M7 . . . 500 μV Max at 25 °C,
VDD = 5 V
- **Input Offset Voltage Drift . . . Typically**
0.1 $\mu\text{V}/\text{Month}$, Including the First 30 Days
- **Wide Range of Supply Voltages over Specified Temperature Range:**
–55°C to 125°C . . . 4 V to 16 V
–40°C to 85°C . . . 4 V to 16 V
0°C to 70°C . . . 3 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix types)**
- **Low Noise . . . Typically 32 nV/ $\sqrt{\text{Hz}}$ at f = 1 kHz**
- **Low Power . . . Typically 2.1 mW at 25°C, VDD = 5 V**
- **Output Voltage Range Includes Negative Rail**
- **High Input Impedance . . . 10¹² Ω Typical**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available in Tape and Reel**
- **Designed-In Latchup Immunity**

D, JG, OR P PACKAGE
(TOP VIEW)



FK PACKAGE
(TOP VIEW)



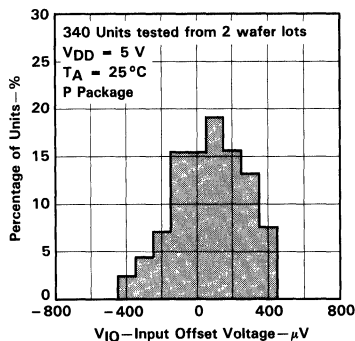
NC—No internal connection

AVAILABLE OPTIONS

T _A	V _{IO} max at 25°C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	500 μV	TLC27M7CD	—	TLC27M7CJG	TLC27M7CP
	2 mV	TLC27M2BCD	—	TLC27M2BCJG	TLC27M2BCP
	5 mV	TLC27M2ACD	—	TLC27M2ACJG	TLC27M2ACP
	10 mV	TLC27M2CD	—	TLC27M2CJG	TLC27M2CP
–40°C to 85°C	500 μV	TLC27M7ID	—	TLC27M7IJG	TLC27M7IP
	2 mV	TLC27M2BID	—	TLC27M2BIJG	TLC27M2BIP
	5 mV	TLC27M2AID	—	TLC27M2AIJG	TLC27M2AIP
	10 mV	TLC27M2ID	—	TLC27M2IJG	TLC27M2IP
–55°C to 125°C	500 μV	—	TLC27M7MFK	TLC27M7MJG	—
	10 mV	—	TLC27M2MFK	TLC27M2MJG	—

The D package is available in tape and reel. Add R suffix to the device type (e.g., TLC27M7CDR).

DISTRIBUTION OF TLC27M7
INPUT OFFSET VOLTAGE



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Operational Amplifiers

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

description

The TLC27M2 and TLC27M7 dual operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, low noise, and speeds comparable to that of general-purpose bipolar devices. These devices use Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance and low bias currents make these cost-effective devices ideal for applications which have previously been reserved for general-purpose bipolar products, but with only a fraction of the power consumption. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC27M2 (10 mV) to the high-precision TLC27M7 (500 μ V). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC27M2 and TLC27M7. The devices also exhibit low voltage single-supply operation and low power consumption, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip carrier versions for high-density system applications.

The device inputs and outputs are designed to withstand –100-mA surge currents without sustaining latchup.

The TLC27M2 and TLC27M7 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

M-suffix devices are characterized for operation over the full military temperature range of –55°C to 125°C. I-suffix devices are characterized for operation from –40°C to 85°C, and C-suffix devices are characterized for operation from 0°C to 70°C.

TLC27M2M, TLC27M7M
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A †	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27M2M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 100\text{ k}\Omega$	25 °C	1.1	10	mV
					Full range		12	
		TLC27M7M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 100\text{ k}\Omega$	25 °C	185	500	μV
					Full range		3750	
α_{VIO}	Average temperature coefficient of input offset voltage			25 °C to 125 °C		1.7		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25 °C	0.1			pA
				125 °C	1.4	15		nA
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25 °C	0.6			pA
				125 °C	9	35		nA
V_{ICR}	Common-mode input voltage range (see Note 5)			25 °C	0 to 4	-0.3 to 4.2		V
				Full range	0 to 3.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 100\text{ k}\Omega$	25 °C	3.2	3.9		V
				-55 °C	3	3.9		
				125 °C	3	4		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25 °C	0	50		mV
				-55 °C	0	50		
				125 °C	0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$	$R_L = 100\text{ k}\Omega$	25 °C	25	170		V/mV
				-55 °C	15	290		
				125 °C	15	120		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25 °C	65	91		dB
				-55 °C	60	89		
				125 °C	60	91		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25 °C	70	93		dB
				-55 °C	60	91		
				125 °C	60	94		
I_{DD}	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$, No load	$V_{IC} = 2.5\text{ V}$	25 °C	210	560		μA
				-55 °C	340	880		
				125 °C	140	360		

† Full range is -55 °C to 125 °C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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Operational Amplifiers

TLC27M2M, TLC27M7M
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT	
V_{IO}	Input offset voltage	TLC27M2M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV	
				Full range			12		
	TLC27M7M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 100\text{ k}\Omega$	25°C	190	800	μV		
			Full range			4300			
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 125°C		2.1		$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current (see Note 4)		$V_O = 5\text{ V}$, $V_{IC} = 5\text{ V}$		25°C	0.1		pA	
					125°C	1.8	15	nA	
I_{IB}	Input bias current (see Note 4)		$V_O = 5\text{ V}$, $V_{IC} = 5\text{ V}$		25°C	0.7		pA	
					125°C	10	35	nA	
V_{ICR}	Common-mode input voltage range (see Note 5)				25°C	0 to 9	-0.3 to 9.2	V	
					Full range	0 to 8.5		V	
V_{OH}	High-level output voltage		$V_{ID} = 100\text{ mV}$, $R_L = 100\text{ k}\Omega$		25°C	8	8.7	V	
					-55°C	7.8	8.6		
					125°C	7.8	8.8		
V_{OL}	Low-level output voltage		$V_{ID} = -100\text{ mV}$, $I_{OL} = 0$		25°C	0	50	mV	
					-55°C	0	50		
					125°C	0	50		
A_{VD}	Large-signal differential voltage amplification		$V_O = 1\text{ V to }6\text{ V}$, $R_L = 100\text{ k}\Omega$		25°C	25	275	V/mV	
					-55°C	15	420		
					125°C	15	190		
CMRR	Common-mode rejection ratio		$V_{IC} = V_{ICR\text{ min}}$		25°C	65	94	dB	
					-55°C	60	93		
					125°C	60	93		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)		$V_{DD} = 5\text{ V to }10\text{ V}$, $V_O = 1.4\text{ V}$		25°C	70	93	dB	
					-55°C	60	91		
					125°C	60	94		
I_{DD}	Supply current (two amplifiers)		$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$		25°C	285	600	μA
						-55°C	490	1000	
						125°C	180	480	

† Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
 5. This range also applies to each input individually.

TLC27M2I, TLC27M2AI, TLC27M2BI, TLC27M7I
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, V_{DD} = 5 V (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T _A [†]	MIN	TYP	MAX	UNIT
V _{IO}	Input offset voltage	TLC27M2I	V _O = 1.4 V, R _S = 50 Ω,	V _{IC} = 0 V, R _L = 100 kΩ	25°C	1.1	10	mV
				Full range			13	
		TLC27M2AI	V _O = 1.4 V, R _S = 50 Ω,	V _{IC} = 0 V, R _L = 100 kΩ	25°C	0.9	5	
				Full range			7	
	TLC27M2BI	V _O = 1.4 V, R _S = 50 Ω,	V _{IC} = 0 V, R _L = 100 kΩ	25°C	220	2000	μV	
				Full range				3500
		TLC27M7I	V _O = 1.4 V, R _S = 50 Ω,	V _{IC} = 0 V, R _L = 100 kΩ	25°C	185		500
					Full range			
α _{VIO}	Average temperature coefficient of input offset voltage			25°C to 85°C		1.7		μV/°C
I _{IO}	Input offset current (see Note 4)	V _O = 2.5 V,	V _{IC} = 2.5 V		25°C	0.1		pA
				85°C	24	1000		
I _{IB}	Input bias current (see Note 4)	V _O = 2.5 V,	V _{IC} = 2.5 V		25°C	0.6		pA
				85°C	200	2000		
V _{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2		V
				Full range	-0.2 to 3.5			V
V _{OH}	High-level output voltage	V _{ID} = 100 mV,	R _L = 100 kΩ	25°C	3.2	3.9		V
				-40°C	3	3.9		
				85°C	3	4		
V _{OL}	Low-level output voltage	V _{ID} = -100 mV,	I _{OL} = 0	25°C		0	50	mV
				-40°C		0	50	
				85°C		0	50	
A _{VD}	Large-signal differential voltage amplification	V _O = 0.25 V to 2 V,	R _L = 100 kΩ	25°C	25	170		V/mV
				-40°C	15	270		
				85°C	15	130		
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICR} min		25°C	65	91		dB
				-40°C	60	90		
				85°C	60	90		
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V		25°C	70	93		dB
				-40°C	60	91		
				85°C	60	94		
I _{DD}	Supply current (two amplifiers)	V _O = 2.5 V, No load	V _{IC} = 2.5 V,	25°C		210	560	μA
				-40°C		315	800	
				85°C		160	400	

† Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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Operational Amplifiers

TLC27M2I, TLC27M2AI, TLC27M2BI, TLC27M7I

LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27M2I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		13	
		TLC27M2AI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 100\text{ k}\Omega$	25°C	0.9	5	mV
					Full range		7	
TLC27M2BI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 100\text{ k}\Omega$	25°C	224	2000	μV		
			Full range		3500			
TLC27M7I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 100\text{ k}\Omega$	25°C	190	800	μV		
	Full range					2900		
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 85°C		2.1		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C		0.1		pA
				85°C		26	1000	
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C		0.7		pA
				85°C		220	2000	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2	-0.3		V
					to	to		
				Full range	-0.2			V
					to			
					8.5			
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$, $R_L = 100\text{ k}\Omega$		25°C	8	8.7		V
				-40°C	7.8	8.7		
				85°C	7.8	8.7		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$, $I_{OL} = 0$		25°C		0	50	mV
				-40°C		0	50	
				85°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$, $R_L = 100\text{ k}\Omega$		25°C	25	275		V/mV
				-40°C	15	390		
				85°C	15	220		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	94		dB
				-40°C	60	93		
				85°C	60	94		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$, $V_O = 1.4\text{ V}$		25°C	70	93		dB
				-40°C	60	91		
				85°C	60	94		
I_{DD}	Supply current (two amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$	25°C		285	600	μA
				-40°C		450	900	
				85°C		205	520	

[†] Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

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Operational Amplifiers

TLC27M2C, TLC27M2AC, TLC27M2BC, TLC27M7C LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A †	MIN	TYP	MAX	UNIT	
V_{IO}	Input offset voltage	TLC27M2C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 100\text{ k}\Omega$	25 °C	1.1	10	mV	
					Full range		12		
		TLC27M2AC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 100\text{ k}\Omega$	25 °C	0.9	5		
					Full range		6.5		
	TLC27M2BC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 100\text{ k}\Omega$	25 °C	220	2000	μV		
				Full range		3000			
		TLC27M7C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0\text{ V}$, $R_L = 100\text{ k}\Omega$	25 °C	185		500	
					Full range			1500	
αV_{IO}	Average temperature coefficient of input offset voltage			25 °C to 70 °C		1.7		$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25 °C	0.1			pA	
				70 °C	7	300			
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25 °C	0.6			pA	
				70 °C	40	600			
V_{ICR}	Common-mode input voltage range (see Note 5)			25 °C	-0.2 to 4	-0.3 to 4.2		V	
				Full range	-0.2 to 3.5			V	
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 100\text{ k}\Omega$	25 °C	3.2	3.9		V	
				0 °C	3	3.9			
				70 °C	3	4			
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25 °C		0	50	mV	
				0 °C		0	50		
				70 °C		0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$	$R_L = 100\text{ k}\Omega$	25 °C	25	170		V/mV	
				0 °C	15	200			
				70 °C	15	140			
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$			25 °C	65	91	dB	
					0 °C	60	91		
					70 °C	60	92		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$			25 °C	70	93	dB
						0 °C	60	92	
						70 °C	60	94	
I_{DD}	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$, No load	$V_{IC} = 2.5\text{ V}$			25 °C	210	560	μA
						0 °C	250	640	
						70 °C	170	440	

† Full range is 0 °C to 70 °C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

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Operational Amplifiers

TLC27M2C, TLC27M2AC, TLC27M2BC, TLC27M7C LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27M2C	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0\text{ V},$ $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27M2AC	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0\text{ V},$ $R_L = 100\text{ k}\Omega$	25°C	0.9	5	
					Full range		6.5	
	TLC27M2BC	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0\text{ V},$ $R_L = 100\text{ k}\Omega$	25°C	224	2000	μV	
				Full range		3000		
		TLC27M7C	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0\text{ V},$ $R_L = 100\text{ k}\Omega$	25°C	190		800
					Full range			1900
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C		2.1		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V},$	$V_{IC} = 5\text{ V}$	25°C		0.1		pA
				70°C		7	300	
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V},$	$V_{IC} = 5\text{ V}$	25°C		0.7		pA
				70°C		50	600	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2		V
				Full range	-0.2 to 8.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV},$	$R_L = 100\text{ k}\Omega$	25°C	8	8.7		V
				0°C	7.8	8.7		
				70°C	7.8	8.7		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV},$	$I_{OL} = 0$	25°C		0	50	mV
				0°C		0	50	
				70°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V},$	$R_L = 100\text{ k}\Omega$	25°C	25	275		V/mV
				0°C	15	320		
				70°C	15	230		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	94		dB
				0°C	60	94		
				70°C	60	94		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V},$	$V_O = 1.4\text{ V}$	25°C	70	93		dB
				0°C	60	92		
				70°C	60	94		
I_{DD}	Supply current (two amplifiers)	$V_O = 5\text{ V},$ No load	$V_{IC} = 5\text{ V},$	25°C		285	600	μA
				0°C		345	800	
				70°C		220	560	

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

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Operational Amplifiers

TLC27M2M, TLC27M7M
linCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25 °C		0.43		V/ μ s
				-55 °C		0.54		
				125 °C		0.29		
			$V_{Ipp} = 2.5\text{ V}$	25 °C		0.40		
				-55 °C		0.49		
				125 °C		0.28		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25 °C		32		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25 °C		55		kHz
				-55 °C		80		
				125 °C		40		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25 °C		525		kHz
				-55 °C		850		
				125 °C		330		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25 °C		40°		
				-55 °C		44°		
				125 °C		36°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25 °C		0.62		V/ μ s
				-55 °C		0.81		
				125 °C		0.38		
			$V_{Ipp} = 5.5\text{ V}$	25 °C		0.56		
				-55 °C		0.73		
				125 °C		0.35		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25 °C		32		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25 °C		35		kHz
				-55 °C		50		
				125 °C		20		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25 °C		635		kHz
				-55 °C		960		
				125 °C		440		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25 °C		43°		
				-55 °C		47°		
				125 °C		39°		

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Operational Amplifiers

TLC27M2I, TLC27M2AI, TLC27M2BI, TLC27M7I
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C		0.43		$\text{V}/\mu\text{s}$
				-40°C		0.51		
				85°C		0.35		
			$V_{Ipp} = 2.5\text{ V}$	25°C		0.40		
				-40°C		0.48		
				85°C		0.32		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		32		$\text{nV}/\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25°C		55		kHz
				-40°C		75		
				85°C		45		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		525		kHz
				-40°C		770		
				85°C		370		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		40°		
				-40°C		43°		
				85°C		38°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C		0.62		$\text{V}/\mu\text{s}$
				-40°C		0.77		
				85°C		0.47		
			$V_{Ipp} = 5.5\text{ V}$	25°C		0.56		
				-40°C		0.70		
				85°C		0.44		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		32		$\text{nV}/\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25°C		35		kHz
				-40°C		45		
				85°C		25		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		635		kHz
				-40°C		880		
				85°C		480		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		43°		
				-40°C		46°		
				85°C		41°		

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Operational Amplifiers

TLC27M2C, TLC27M2AC, TLC27M2BC, TLC27M7C
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C		0.43		V/ μs
				0°C		0.46		
				70°C		0.36		
			$V_{IPP} = 2.5\text{ V}$	25°C		0.40		
				0°C		0.43		
				70°C		0.34		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		32		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C		55		kHz
				0°C		60		
				70°C		50		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		525		kHz
				0°C		600		
				70°C		400		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		40°		
				0°C		41°		
				70°C		39°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C		0.62		V/ μs
				0°C		0.67		
				70°C		0.51		
			$V_{IPP} = 5.5\text{ V}$	25°C		0.56		
				0°C		0.61		
				70°C		0.46		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		32		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C		35		kHz
				0°C		40		
				70°C		30		
B_1	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		635		kHz
				0°C		710		
				70°C		510		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		43°		
				0°C		44°		
				70°C		42°		

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Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC27M2 and TLC27M7 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit will give the same result.



FIGURE 1. UNITY-GAIN AMPLIFIER

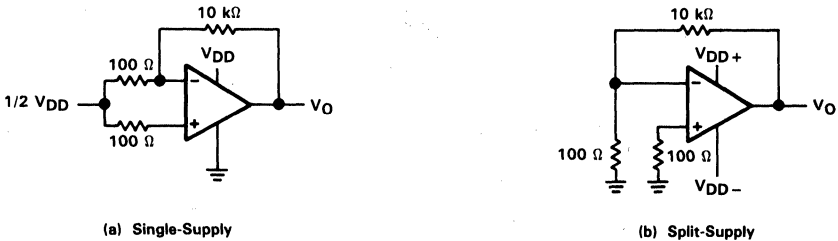


FIGURE 2. NOISE TEST CIRCUIT

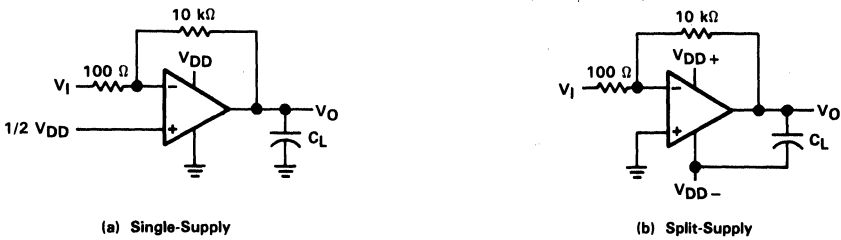


FIGURE 3. GAIN-OF-100 INVERTING AMPLIFIER

PARAMETER MEASUREMENT INFORMATION

input bias current

Because of the high input impedance of the TLC27M2 and TLC27M7 op amps, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs will be shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the "open-socket" leakage readings from the readings obtained with a device in the test socket.

One word of caution . . . many automatic testers as well as some bench-top op amp testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an "open-socket" reading is not feasible using this method.

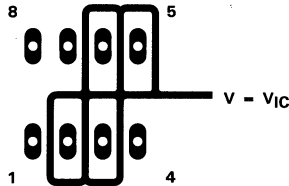


FIGURE 4. ISOLATION METAL AROUND DEVICE INPUTS
(JG AND P DUAL-IN-LINE-PACKAGE)

low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

PARAMETER MEASUREMENT INFORMATION

input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture will result in leakage and contact resistance which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

2 full-power response

Full-power response, the frequency above which the op amp slew rate limits the output voltage swing, is often specified two ways . . . full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for "significant" distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

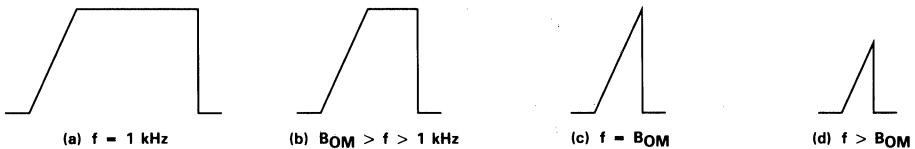


FIGURE 5. FULL-POWER-RESPONSE OUTPUT SIGNAL

test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7
 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC27M2
 INPUT OFFSET VOLTAGE

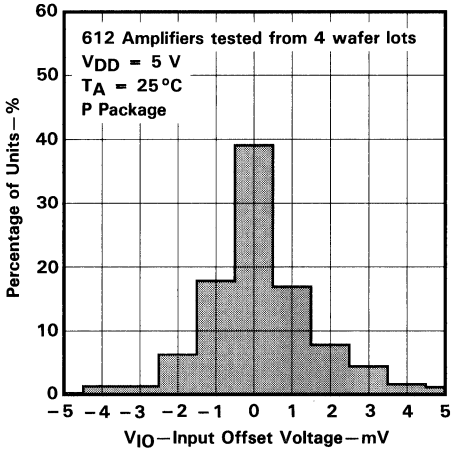


FIGURE 6

DISTRIBUTION OF TLC27M2
 INPUT OFFSET VOLTAGE

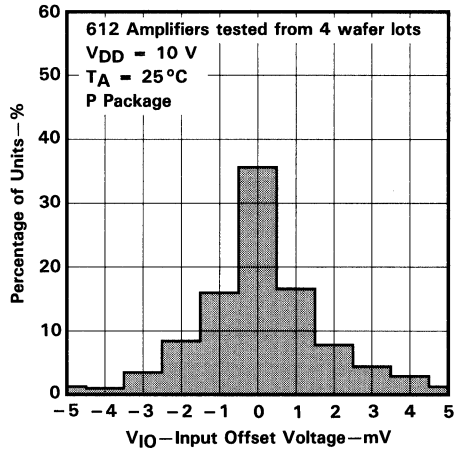


FIGURE 7

DISTRIBUTION OF TLC27M2 AND TLC27M7
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT

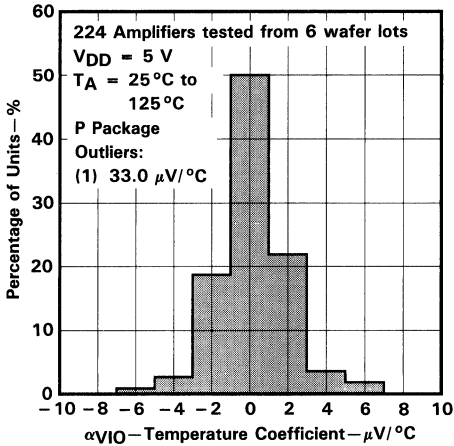


FIGURE 8

DISTRIBUTION OF TLC27M2 AND TLC27M7
 INPUT OFFSET VOLTAGE
 TEMPERATURE COEFFICIENT

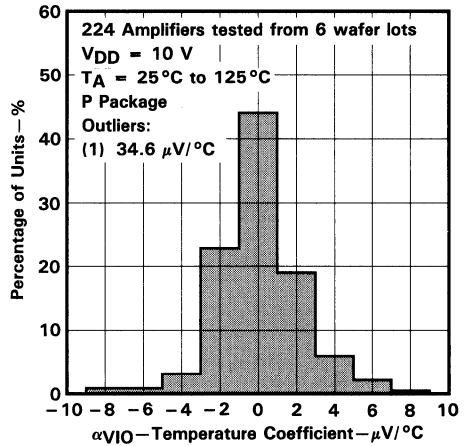


FIGURE 9

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

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Operational Amplifiers

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

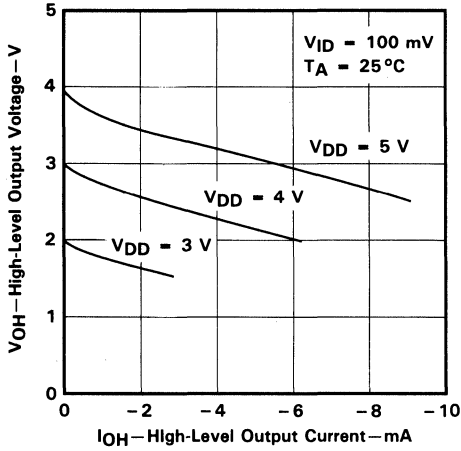


FIGURE 10

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

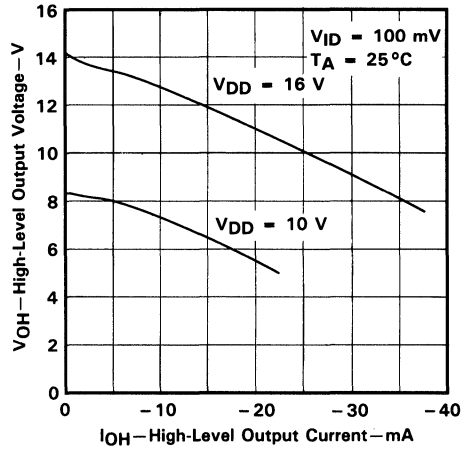


FIGURE 11

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 SUPPLY VOLTAGE

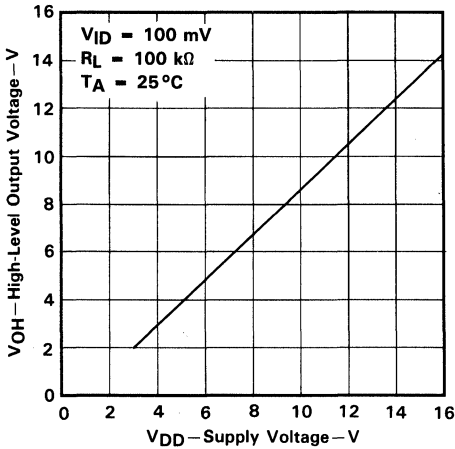


FIGURE 12

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

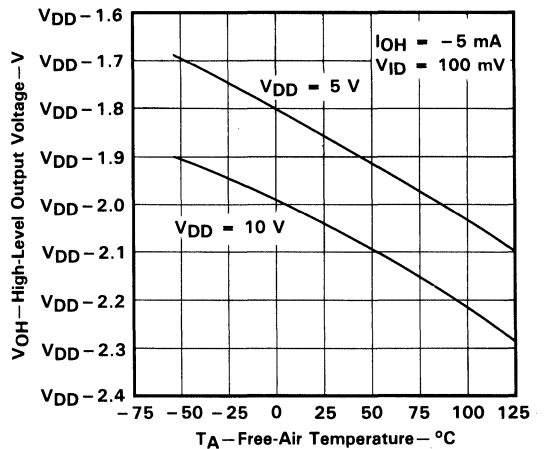


FIGURE 13

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

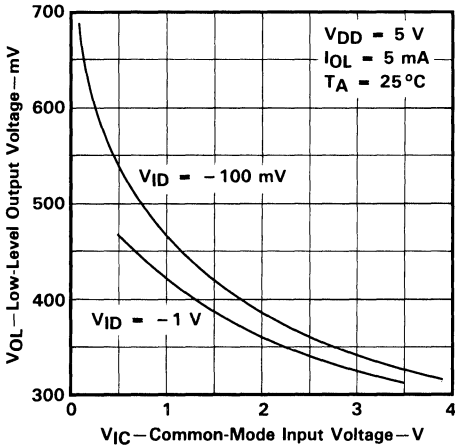


FIGURE 14

LOW-LEVEL OUTPUT VOLTAGE
 vs
 COMMON-MODE INPUT VOLTAGE

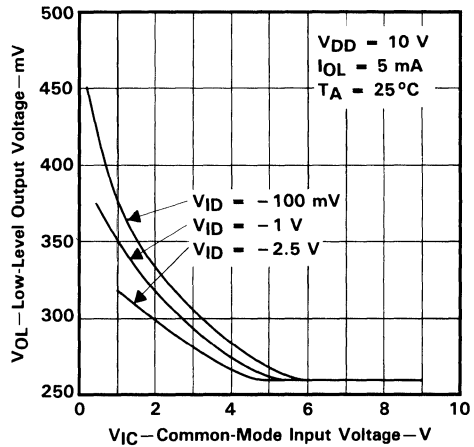


FIGURE 15

LOW-LEVEL OUTPUT VOLTAGE
 vs
 DIFFERENTIAL INPUT VOLTAGE

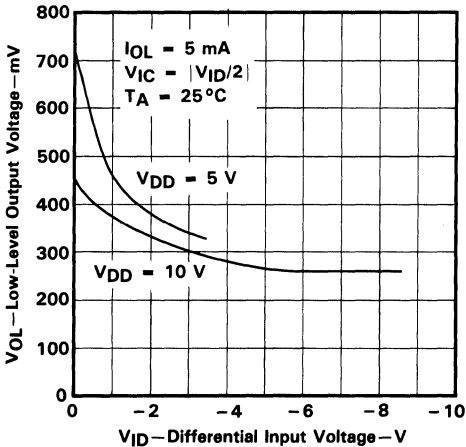


FIGURE 16

LOW-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

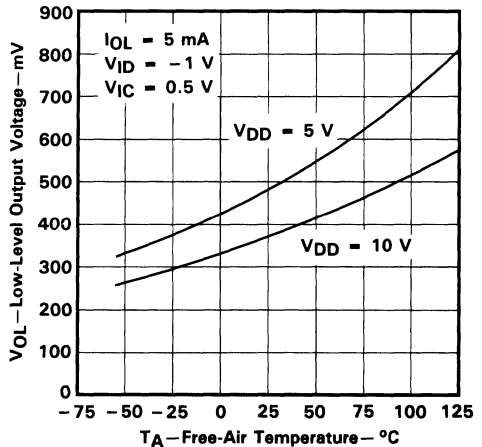


FIGURE 17

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

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Operational Amplifiers

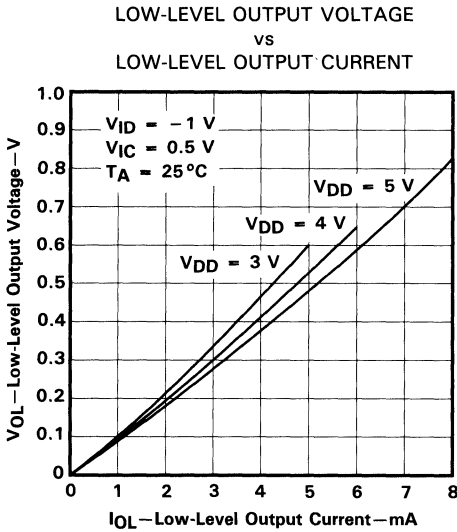


FIGURE 18

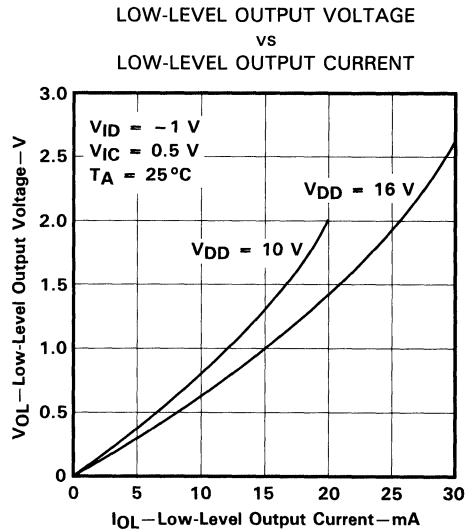


FIGURE 19

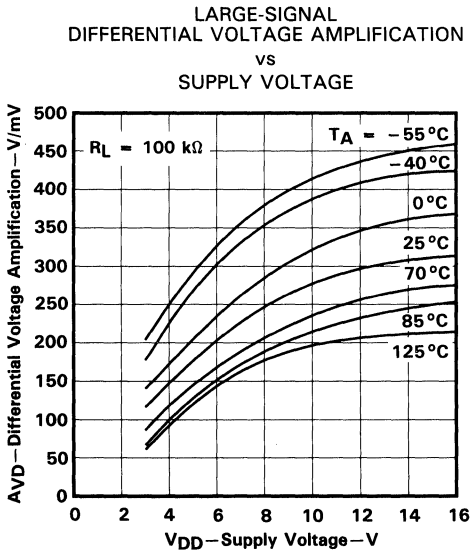


FIGURE 20

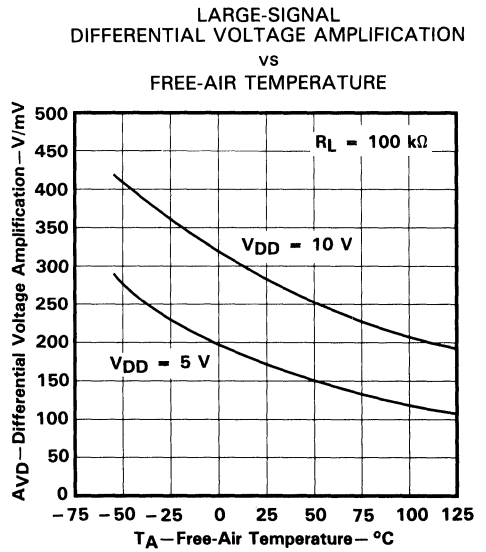


FIGURE 21

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

INPUT BIAS CURRENT AND INPUT OFFSET CURRENT
 vs
 FREE-AIR TEMPERATURE

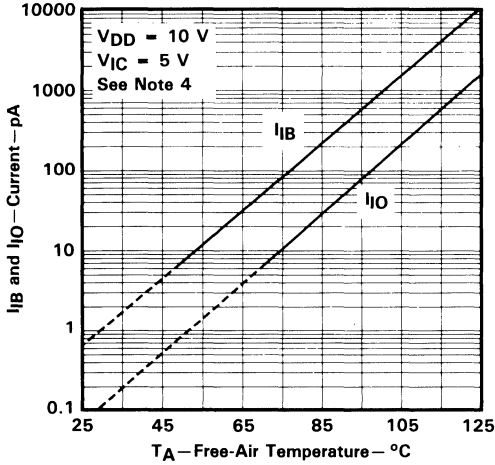


FIGURE 22

COMMON-MODE
 INPUT VOLTAGE POSITIVE LIMIT
 vs
 SUPPLY VOLTAGE

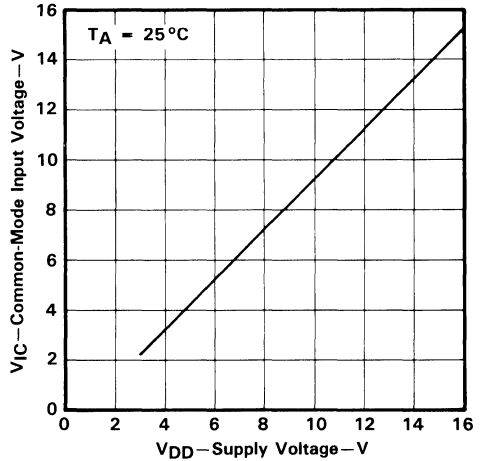


FIGURE 23

SUPPLY CURRENT
 vs
 SUPPLY VOLTAGE

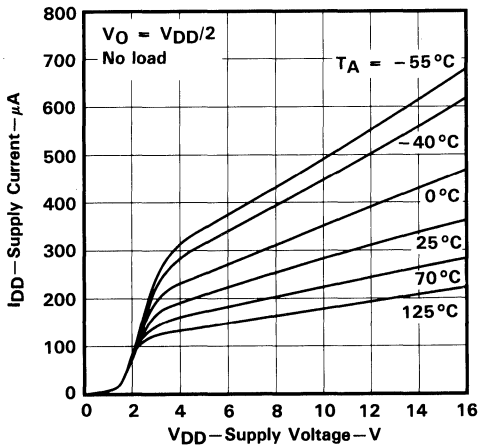


FIGURE 24

SUPPLY CURRENT
 vs
 FREE-AIR TEMPERATURE

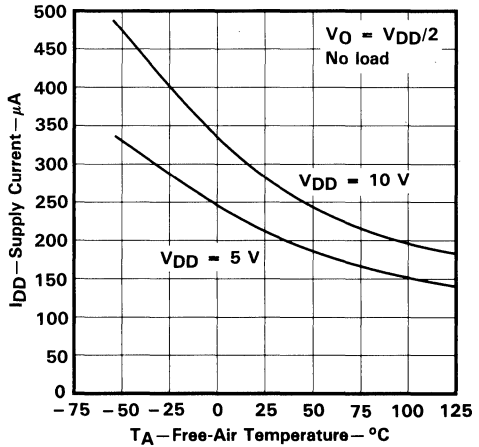


FIGURE 25

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.
 NOTE 4: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

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 Operational Amplifiers

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

**SLEW RATE
vs
SUPPLY VOLTAGE**

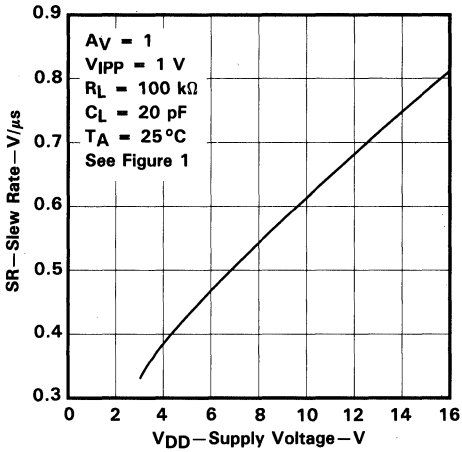


FIGURE 26

**SLEW RATE
vs
FREE-AIR TEMPERATURE**

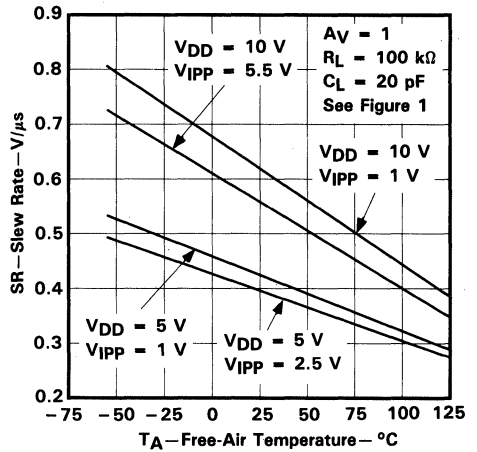


FIGURE 27

**NORMALIZED SLEW RATE
vs
FREE-AIR TEMPERATURE**

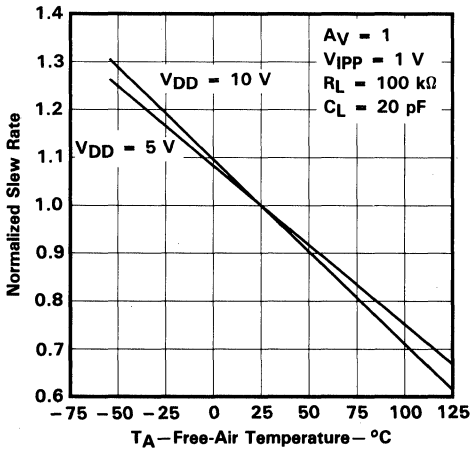


FIGURE 28

**MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
vs
FREQUENCY**

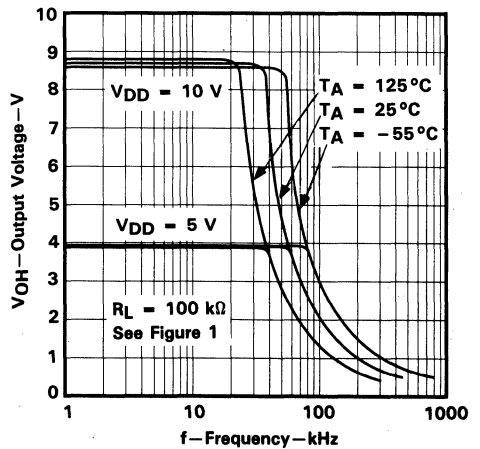


FIGURE 29

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

UNITY-GAIN BANDWIDTH
 vs
 FREE-AIR TEMPERATURE

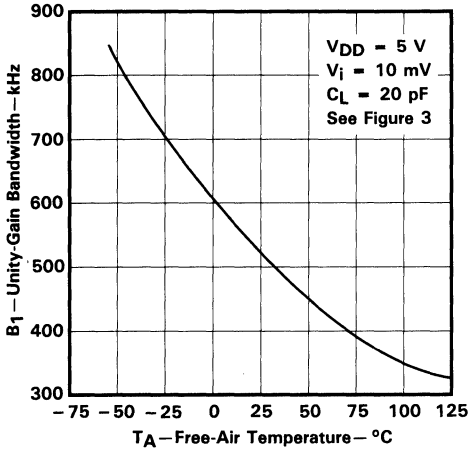


FIGURE 30

UNITY-GAIN BANDWIDTH
 vs
 SUPPLY VOLTAGE

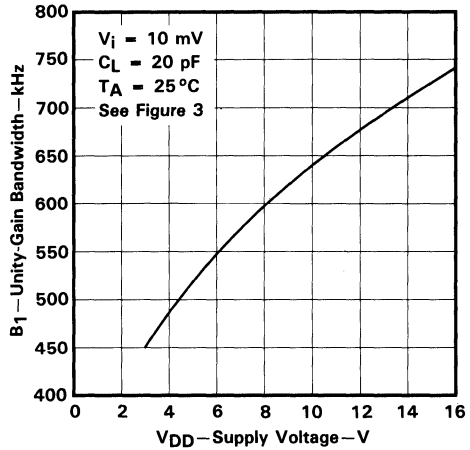


FIGURE 31

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

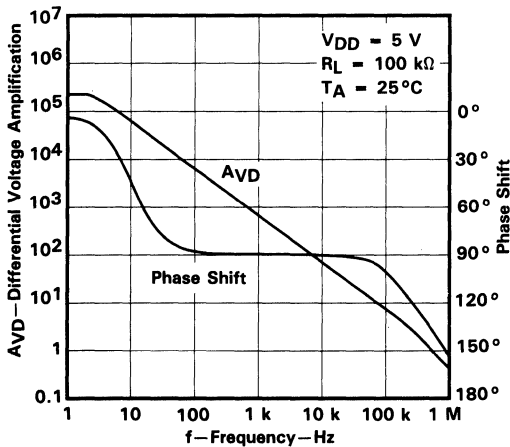


FIGURE 32

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

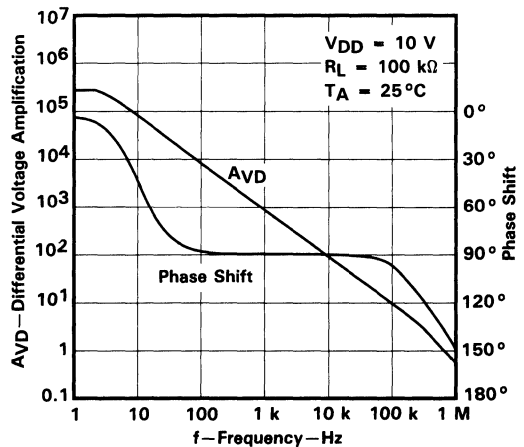


FIGURE 33

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

PHASE MARGIN
 vs
 SUPPLY VOLTAGE

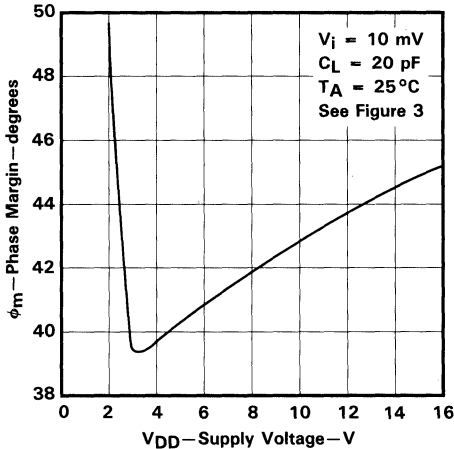


FIGURE 34

PHASE MARGIN
 vs
 FREE-AIR TEMPERATURE

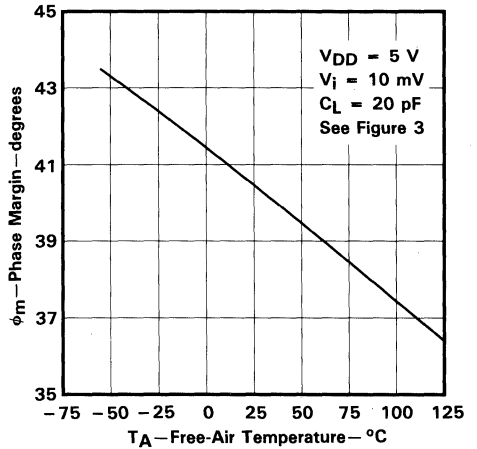


FIGURE 35

PHASE MARGIN
 vs
 CAPACITIVE LOAD

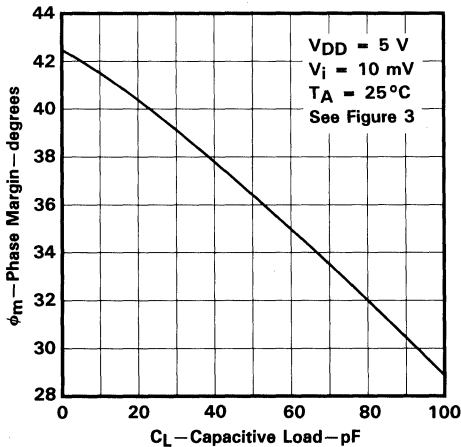


FIGURE 36

EQUIVALENT INPUT NOISE VOLTAGE
 vs
 FREQUENCY

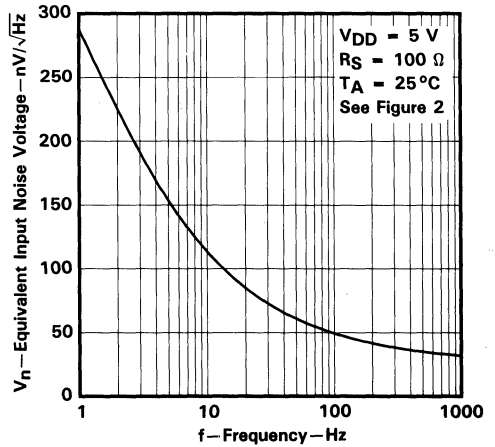


FIGURE 37

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL APPLICATION DATA

single-supply operation

While the TLC27M2 and TLC27M7 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC27M2 and TLC27M7 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC27M2 and TLC27M7 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

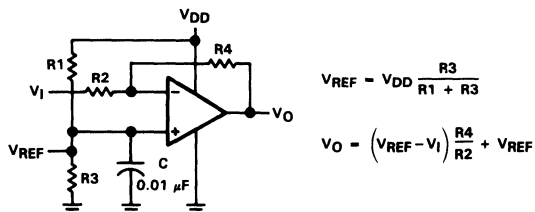


FIGURE 38. INVERTING AMPLIFIER WITH VOLTAGE REFERENCE

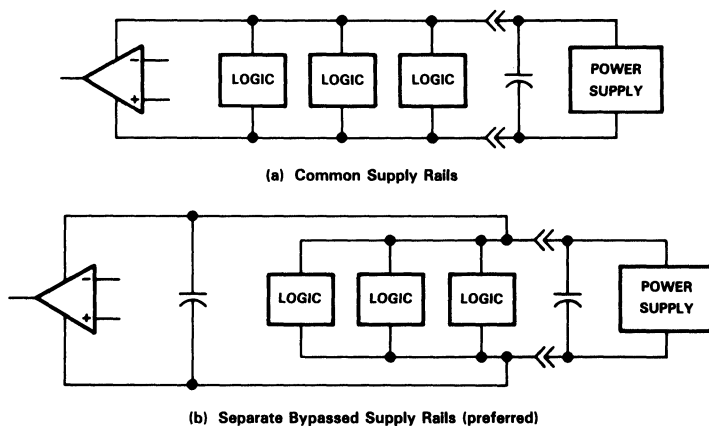


FIGURE 39. COMMON VS SEPARATE SUPPLY RAILS

TYPICAL APPLICATION DATA

input characteristics

The TLC27M2 and TLC27M7 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at $V_{DD} - 1\text{ V}$ at $T_A = 25^\circ\text{C}$ and at $V_{DD} - 1.5\text{ V}$ at all other temperatures.

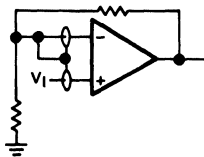
The use of the polysilicon-gate process and the careful input circuit design gives the TLC27M2 and TLC27M7 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically $0.1\ \mu\text{V}/\text{month}$, including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC27M2 and TLC27M7 are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

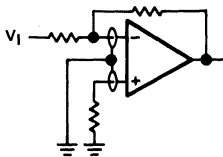
The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.

noise performance

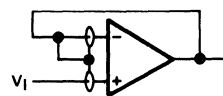
The noise specifications in op amp circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC27M2 and TLC27M7 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than $50\ \text{k}\Omega$, since bipolar devices exhibit greater noise currents.



(a) Noninverting Amplifier



(b) Inverting Amplifier



(c) Unity-Gain Amplifier

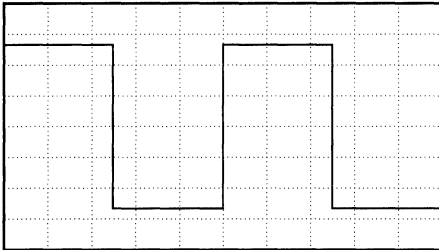
FIGURE 40. GUARD-RING SCHEMES

output characteristics

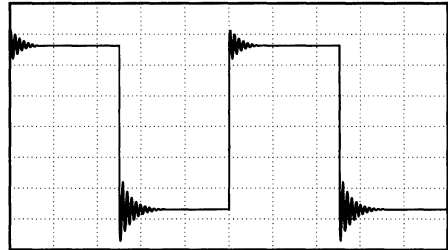
The output stage of the TLC27M2 and TLC27M7 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

All operating characteristics of the TLC27M2 and TLC27M7 were measured using a 20-pF load. The devices will drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance will alleviate the problem.

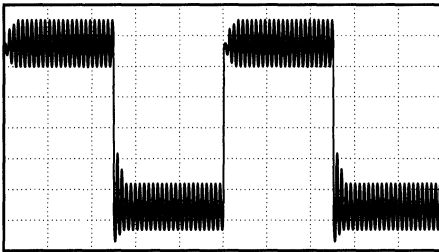
TYPICAL APPLICATION DATA



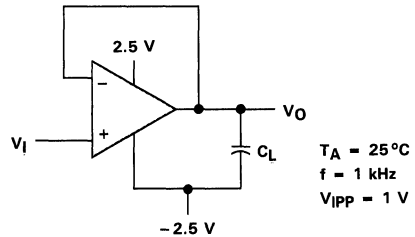
(a) $C_L = 20 \text{ pF}$, $R_L = \text{no load}$



(b) $C_L = 170 \text{ pF}$, $R_L = \text{no load}$



(c) $C_L = 190 \text{ pF}$, $R_L = \text{no load}$



(d) Test Circuit

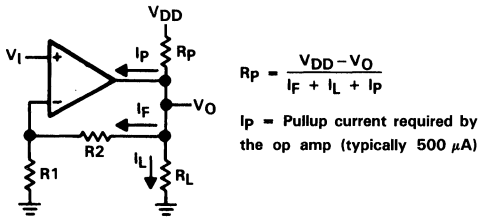
FIGURE 41. EFFECT OF CAPACITIVE LOADS AND TEST CIRCUIT

Although the TLC27M2 and TLC27M7 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor (R_p) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pull-down transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately 60Ω and 180Ω , depending on how hard the op amp input is driven. With very low values of R_p , a voltage offset from 0 V at the output will occur. Second, pullup resistor R_p acts as a drain load to N4 and the gain of the op amp is reduced at output voltage levels where N5 is not supplying the output current.

feedback

Op amp circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

TYPICAL APPLICATION DATA



$$R_p = \frac{V_{DD} - V_O}{I_F + I_L + I_p}$$

I_p = Pullup current required by the op amp (typically 500 μ A)

FIGURE 42. RESISTIVE PULLUP TO INCREASE V_{OH}

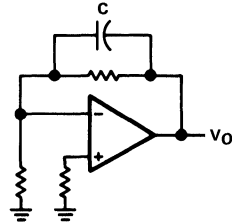


FIGURE 43. COMPENSATION FOR INPUT CAPACITANCE

2

Operational Amplifiers

electrostatic discharge protection

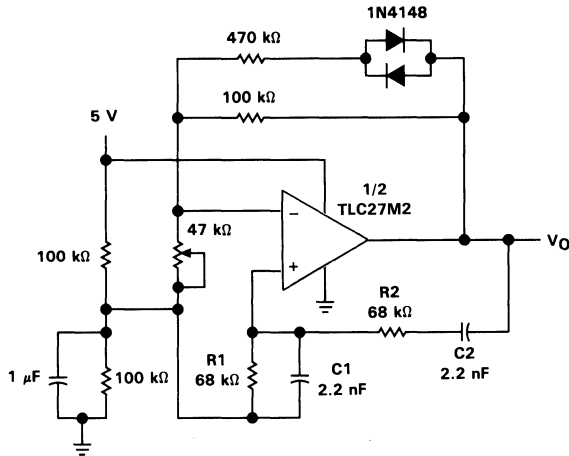
The TLC27M2 and TLC27M7 incorporate an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices, as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

latchup

Because CMOS devices are susceptible to latchup due to their inherent parasitic thyristors, the TLC27M2 and TLC27M7 inputs and outputs were designed to withstand – 100-mA surge currents without sustaining latchup; however, techniques should be used to reduce the chance of latchup whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μ F typical) located across the supply rails as close to the device as possible.

The current path established if latchup occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latchup occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latchup occurring increases with increasing temperature and supply voltages.

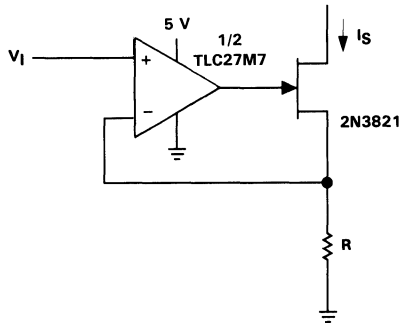
TYPICAL APPLICATION DATA



NOTES: $V_{OPP} \approx 2 \text{ V}$

$$f_O = \frac{1}{2\pi \sqrt{R1R2C1C2}}$$

FIGURE 44. WIEN OSCILLATOR

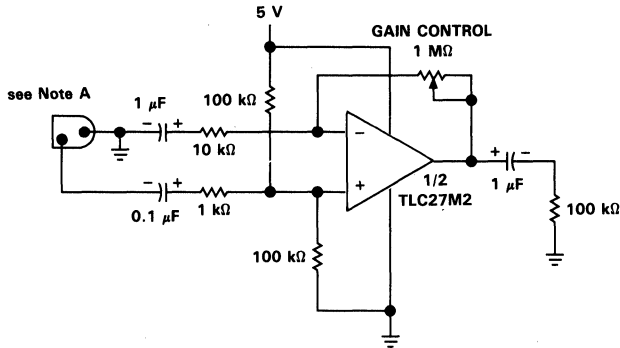


NOTES: $V_I = 0 \text{ V to } 3 \text{ V}$

$$I_S = \frac{V_I}{R}$$

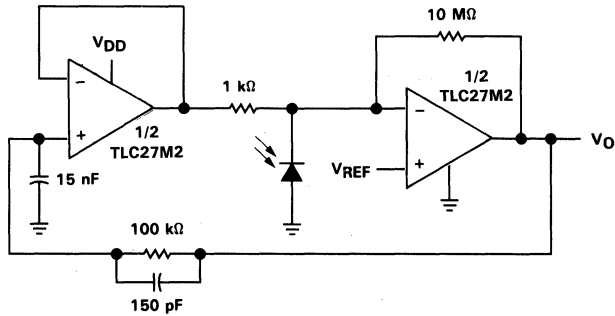
FIGURE 45. PRECISION LOW-CURRENT SINK

TYPICAL APPLICATION DATA



NOTE A.: Low to medium impedance dynamic mike

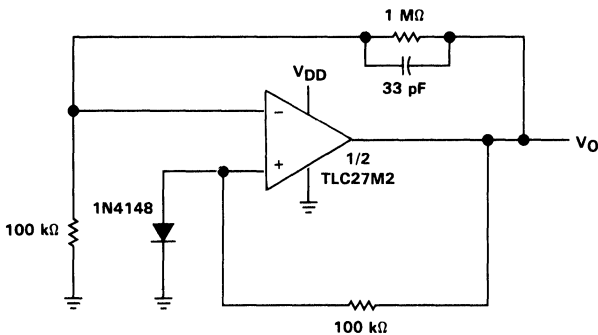
FIGURE 46. MICROPHONE PREAMPLIFIER



NOTES: $V_{DD} = 4 \text{ V to } 15 \text{ V}$
 $V_{REF} = 0 \text{ V to } V_{DD} - 2 \text{ V}$

FIGURE 47. PHOTO DIODE AMPLIFIER WITH AMBIENT LIGHT REJECTION

TYPICAL APPLICATION DATA



NOTES: $V_{DD} = 8 \text{ V to } 16 \text{ V}$
 $V_O = 5 \text{ V, } 10 \text{ mA}$

FIGURE 48. 5-V LOW-POWER VOLTAGE REGULATOR

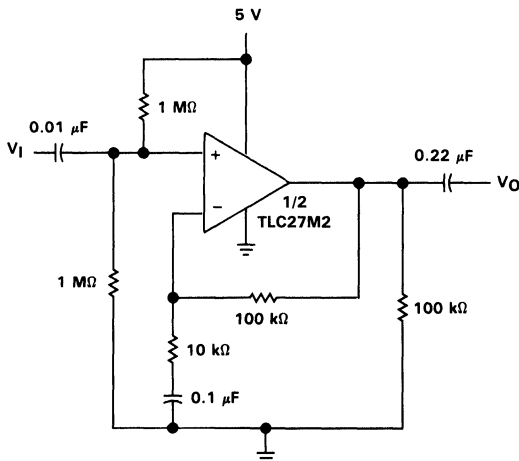


FIGURE 49. SINGLE-RAIL AC AMPLIFIER

2

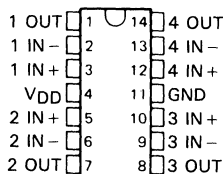
Operational Amplifiers

TLC27M4, TLC27M4A, TLC27M4B, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

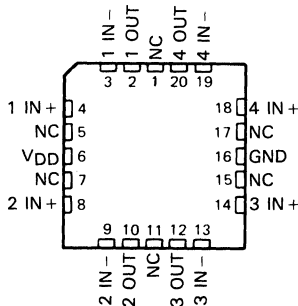
D3143, OCTOBER 1987—REVISED AUGUST 1988

- **Trimmed Offset Voltage:**
TLC27M9 . . . 900 μV Max at 25 °C,
 $V_{\text{DD}} = 5 \text{ V}$
- **Input Offset Voltage Drift . . . Typically**
0.1 $\mu\text{V}/\text{Month}$, Including the First 30 Days
- **Wide Range of Supply Voltages over Specified Temperature Range:**
–55 °C to 125 °C . . . 4 V to 16 V
–40 °C to 85 °C . . . 4 V to 16 V
0 °C to 70 °C . . . 3 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix Types)**
- **Low Noise . . . Typically 32 nV/ $\sqrt{\text{Hz}}$ at $f = 1 \text{ kHz}$**
- **Low Power . . . Typically 2.1 mW at 25 °C, $V_{\text{DD}} = 5 \text{ V}$**
- **Output Voltage Range Includes Negative Rail**
- **High Input Impedance . . . 10¹² Ω Typical**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available in Tape and Reel**
- **Designed-In Latchup Immunity**

D, J, OR N PACKAGE
(TOP VIEW)



FK PACKAGE
(TOP VIEW)



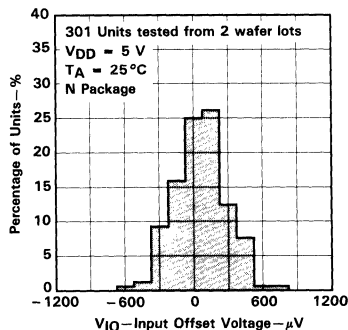
NC—No internal connection

AVAILABLE OPTIONS

T_A	V_{IOmax} at 25 °C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0 °C to 70 °C	900 μV	TLC27M9CD	—	TLC27M9CJ	TLC27M9CN
	2 mV	TLC27M4BCD	—	TLC27M4BCJ	TLC27M4BCN
	5 mV	TLC27M4ACD	—	TLC27M4ACJ	TLC27M4ACN
	10 mV	TLC27M4CD	—	TLC27M4CJ	TLC27M4CN
–40 °C to 85 °C	900 μV	TLC27M9ID	—	TLC27M9IJ	TLC27M9IN
	2 mV	TLC27M4BID	—	TLC27M4BIJ	TLC27M4BIN
	5 mV	TLC27M4AID	—	TLC27M4AIJ	TLC27M4AIN
	10 mV	TLC27M4ID	—	TLC27M4IJ	TLC27M4IN
–55 °C to 125 °C	900 μV	—	TLC27M9MFK	TLC27M9MJ	—
	10 mV	—	TLC27M4MFK	TLC27M4MJ	—

The D package is available in tape and reel. Add R suffix to the device type (e.g., TLC27M9CDR).

DISTRIBUTION OF TLC27M9
INPUT OFFSET VOLTAGE



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TLC27M4, TLC27M4A, TLC27M4B, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

description

The TLC27M4 and TLC27M9 quad operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, low noise, and speeds comparable to that of general-purpose bipolar devices. These devices use Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance and low bias currents make these cost-effective devices ideal for applications that have previously been reserved for general-purpose bipolar products, but with only a fraction of the power consumption. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC27M4 (10 mV) to the high-precision TLC27M9 (900 μ V). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC27M4 and TLC27M9. The devices also exhibit low voltage single-supply operation and low power consumption, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip carrier versions for high-density system applications.

The device inputs and outputs are designed to withstand -100 -mA surge currents without sustaining latchup.

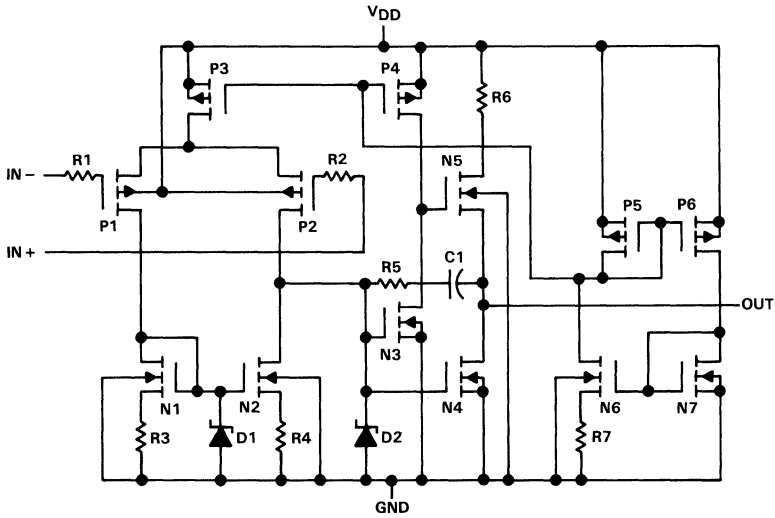
The TLC27M4 and TLC27M9 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I-suffix devices are characterized for operation from -40°C to 85°C , and the C-suffix devices are characterized for operation from 0°C to 70°C .



TLC27M4, TLC27M4A, TLC27M4B, TLC27M9
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

equivalent schematic (each amplifier)



TLC27M4M, TLC27M9M LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27M4M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C		1.1	10	mV
				Full range			12	
	TLC27M9M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C		210	900	μV	
			Full range			3750		
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 125°C		1.7		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$, $V_{IC} = 2.5\text{ V}$		25°C		0.1		pA
				125°C		1.4	15	nA
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$, $V_{IC} = 2.5\text{ V}$		25°C		0.6		pA
				125°C		9	35	nA
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	0 to 4	-0.3 to 4.2		V
				Full range	0 to 3.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$, $R_L = 100\text{ k}\Omega$		25°C		3.2	3.9	V
				-55°C		3	3.9	
				125°C		3	4	
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$, $I_{OL} = 0$		25°C		0	50	mV
				-55°C		0	50	
				125°C		0	50	
AVD	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$, $R_L = 100\text{ k}\Omega$		25°C		25	170	V/mV
				-55°C		15	290	
				125°C		15	120	
$CMRR$	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C		65	91	dB
				-55°C		60	89	
				125°C		60	91	
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$, $V_O = 1.4\text{ V}$		25°C		70	93	dB
				-55°C		60	91	
				125°C		60	94	
I_{DD}	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$, No load	$V_{IC} = 2.5\text{ V}$,	25°C		420	1120	μA
				-55°C		680	1760	
				125°C		280	720	

† Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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Operational Amplifiers

TLC27M4M, TLC27M9M

LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27M4M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$,	$V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27M9M	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$,	$V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C	220	1200	μV
					Full range		4300	
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 125°C		2.1		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C		0.1		pA
				125°C		1.8	15	nA
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C		0.7		pA
				125°C		10	35	nA
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	0 to 9	-0.3 to 9.2		V
				Full range	0 to 8.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$,	$R_L = 100\text{ k}\Omega$	25°C	8	8.7		V
				-55°C	7.8	8.6		
				125°C	7.8	8.8		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$,	$I_{OL} = 0$	25°C	0	0	50	mV
				-55°C	0	0	50	
				125°C	0	0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$,	$R_L = 100\text{ k}\Omega$	25°C	25	275		V/mV
				-55°C	15	420		
				125°C	15	190		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	94		dB
				-55°C	60	93		
				125°C	60	93		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$,	$V_O = 1.4\text{ V}$	25°C	70	93		dB
				-55°C	60	91		
				125°C	60	94		
I_{DD}	Supply current (four amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$,	25°C		570	1200	μA
				-55°C		980	2000	
				125°C		360	960	

† Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

TLC27M4I, TLC27M4AI, TLC27M4BI, TLC27M9I LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27M4I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		13	
		TLC27M4AI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C	0.9	5	
					Full range		7	
	TLC27M4BI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C	250	2000	μV	
				Full range		3500		
		TLC27M9I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C	210		900
					Full range			2000
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 85°C		1.7		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C		0.1		pA
				85°C		24	1000	
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C		0.6		pA
				85°C		200	2000	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2		V
				Full range	-0.2 to 3.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 100\text{ k}\Omega$	25°C	3.2	3.9		V
				-40°C	3	3.9		
				85°C	3	4		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C		0	50	mV
				-40°C		0	50	
				85°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$	$R_L = 100\text{ k}\Omega$	25°C	25	170		V/mV
				-40°C	15	270		
				85°C	15	130		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	91		dB
				-40°C	60	90		
				85°C	60	90		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	93		dB
				-40°C	60	91		
				85°C	60	94		
I_{DD}	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$, No load	$V_{IC} = 2.5\text{ V}$	25°C		420	1120	μA
				-40°C		630	1600	
				85°C		320	800	

† Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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Operational Amplifiers

TLC27M4C, TLC27M4AC, TLC27M4BC, TLC27M9C

LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A †	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27M4I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$,	$V_{IC} = 0$,	25°C	1.1	10	mV
				$R_L = 100\text{ k}\Omega$	Full range		13	
		TLC27M4AI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$,	$V_{IC} = 0$,	25°C	0.9	5	mV
				$R_L = 100\text{ k}\Omega$	Full range		7	
		TLC27M4BI	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$,	$V_{IC} = 0$,	25°C	260	2000	μV
				$R_L = 100\text{ k}\Omega$	Full range		3500	
		TLC27M9I	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$,	$V_{IC} = 0$,	25°C	220	1200	μV
				$R_L = 100\text{ k}\Omega$	Full range		2900	
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 85°C		2.1		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C		0.1		pA
				85°C		26	1000	
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$,	$V_{IC} = 5\text{ V}$	25°C		0.7		pA
				85°C		220	2000	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2		V
				Full range	-0.2 to 8.5			
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$,	$R_L = 100\text{ k}\Omega$	25°C	8	8.7		V
				-40°C	7.8	8.7		
				85°C	7.8	8.7		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$,	$I_{OL} = 0$	25°C		0	50	mV
				-40°C		0	50	
				85°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$,	$R_L = 100\text{ k}\Omega$	25°C	25	275		V/mV
				-40°C	15	390		
				85°C	15	220		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min}$		25°C	65	94		dB
				-40°C	60	93		
				85°C	60	94		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$,	$V_O = 1.4\text{ V}$	25°C	70	93		dB
				-40°C	60	91		
				85°C	60	94		
I_{DD}	Supply current (four amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$,	25°C		570	1200	μA
				-40°C		900	1800	
				85°C		410	1040	

† Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

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Operational Amplifiers

TLC27M4C, TLC27M4AC, TLC27M4BC, TLC27M9C LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27M4C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C		1.1	10	mV
				Full range			12	
		TLC27M4AC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C		0.9	5	
				Full range			6.5	
	TLC27M4BC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C		250	2000	μV	
			Full range			3000		
		TLC27M9C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$, $V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C		210		900
				Full range				1500
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C		1.7		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C		0.1		pA
				70°C		7	300	
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C		0.6		pA
				70°C		40	600	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2		V
				Full range	-0.2 to 3.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$, $R_L = 100\text{ k}\Omega$		25°C		3.2	3.9	V
				0°C		3	3.9	
				70°C		3	4	
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$, $I_{OL} = 0$		25°C		0	50	mV
				0°C		0	50	
				70°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$, $R_L = 100\text{ k}\Omega$		25°C		25	170	V/mV
				0°C		15	200	
				70°C		15	140	
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min}$		25°C		65	91	dB
				0°C		60	91	
				70°C		60	92	
				25°C		70	93	
kSVR	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to } 10\text{ V}$, $V_O = 1.4\text{ V}$		0°C		60	92	dB
				70°C		60	94	
				25°C		420	1120	
I_{DD}	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$, No load	$V_{IC} = 2.5\text{ V}$	25°C		500	1280	μA
				0°C		340	880	
				70°C				

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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Operational Amplifiers

TLC27M4I, TLC27M4AI, TLC27M4BI, TLC27M9I
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	TLC27M4C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27M4AC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C	0.9	5	μV
					Full range		6.5	
		TLC27M4BC	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C	260	2000	μV
Full range		3000						
TLC27M9C	$V_O = 1.4\text{ V}$, $R_S = 50\ \Omega$	$V_{IC} = 0$, $R_L = 100\text{ k}\Omega$	25°C	220	1200	μV		
Full range				1900				
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C		2.1		$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C		0.1		pA
				70°C		7	300	
I_{IB}	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C		0.7		pA
				70°C		50	600	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2		V
				Full range	-0.2 to 8.5			V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 100\text{ k}\Omega$	25°C	8	8.7		V
				0°C	7.8	8.7		
				70°C	7.8	8.7		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C		0	50	mV
				0°C		0	50	
				70°C		0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 100\text{ k}\Omega$	25°C	25	275		V/mV
				0°C	15	320		
				70°C	15	230		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	65	94		dB
				0°C	60	94		
				70°C	60	94		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	93		dB
				0°C	60	92		
				70°C	60	94		
I_{DD}	Supply current (four amplifiers)	$V_O = 5\text{ V}$, No load	$V_{IC} = 5\text{ V}$	25°C		570	1200	μA
				0°C		690	1600	
				70°C		440	1120	

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
 5. This range also applies to each input individually.

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Operational Amplifiers

TLC27M4M, TLC27M9M
linCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C		0.43		$V/\mu\text{s}$
				-55°C		0.54		
				125°C		0.29		
			$V_{IPP} = 2.5\text{ V}$	25°C		0.40		
				-55°C		0.50		
				125°C		0.28		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		32	$\text{nV}/\sqrt{\text{Hz}}$	
BOM	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25°C		55		kHz
				-55°C		80		
				125°C		40		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		525		kHz
				-55°C		850		
				125°C		330		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		40°		
				-55°C		44°		
				125°C		36°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C		0.62		$V/\mu\text{s}$
				-55°C		0.81		
				125°C		0.38		
			$V_{IPP} = 5.5\text{ V}$	25°C		0.56		
				-55°C		0.73		
				125°C		0.35		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		32	$\text{nV}/\sqrt{\text{Hz}}$	
BOM	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$, See Figure 1	25°C		35		kHz
				-55°C		50		
				125°C		20		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		635		kHz
				-55°C		960		
				125°C		440		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		43°		
				-55°C		47°		
				125°C		39°		

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Operational Amplifiers

TLC27M4I, TLC27M4AI, TLC27M4BI, TLC27M9I
in CMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{I_{PP}} = 1\text{ V}$	25°C		0.43		$V/\mu\text{s}$
				-40°C		0.51		
				85°C		0.35		
			$V_{I_{PP}} = 2.5\text{ V}$	25°C		0.40		
				-40°C		0.48		
				85°C		0.32		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		32	$\text{nV}/\sqrt{\text{Hz}}$	
B _{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C		55		kHz
				-40°C		75		
				85°C		45		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		525		kHz
				-40°C		770		
				85°C		370		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		40°		
				-40°C		43°		
				85°C		38°		

operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{I_{PP}} = 1\text{ V}$	25°C		0.62		$V/\mu\text{s}$
				-40°C		0.77		
				85°C		0.47		
			$V_{I_{PP}} = 5.5\text{ V}$	25°C		0.56		
				-40°C		0.70		
				85°C		0.44		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 100\ \Omega$,	25°C		32	$\text{nV}/\sqrt{\text{Hz}}$	
B _{OM}	Maximum output swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C		35		kHz
				-40°C		45		
				85°C		25		
B ₁	Unity-gain bandwidth	$V_i = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	25°C		635		kHz
				-40°C		880		
				85°C		480		
ϕ_m	Phase margin	$V_i = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C		43°		
				-40°C		46°		
				85°C		41°		

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Operational Amplifiers

TLC27M4C, TLC27M4AC, TLC27M4BC, TLC27M9C
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

operating characteristics, V_{DD} = 5 V

PARAMETER		TEST CONDITIONS		T _A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	R _L = 100 kΩ, C _L = 20 pF, See Figure 1	V _{IPP} = 1 V	25 °C		0.43		V/μs
				0 °C		0.46		
				70 °C		0.36		
			V _{IPP} = 2.5 V	25 °C		0.40		
				0 °C		0.43		
				70 °C		0.34		
V _n	Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 100 Ω,	25 °C		32		nV/√Hz
B _{OM}	Maximum output swing bandwidth	V _O = V _{OH} , R _L = 100 kΩ,	C _L = 20 pF, See Figure 1	25 °C		55		kHz
				0 °C		60		
				70 °C		50		
B ₁	Unity-gain bandwidth	V _i = 10 mV, See Figure 3	C _L = 20 pF,	25 °C		525		kHz
				0 °C		610		
				70 °C		400		
φ _m	Phase margin	V _i = 10 mV, C _L = 20 pF,	f = B ₁ , See Figure 3	25 °C		40°		
				0 °C		41°		
				70 °C		39°		

operating characteristics, V_{DD} = 10 V

PARAMETER		TEST CONDITIONS		T _A	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	R _L = 100 kΩ, C _L = 20 pF, See Figure 1	V _{IPP} = 1 V	25 °C		0.62		V/μs
				0 °C		0.67		
				70 °C		0.51		
			V _{IPP} = 5.5 V	25 °C		0.56		
				0 °C		0.61		
				70 °C		0.46		
V _n	Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 100 Ω,	25 °C		32		nV/√Hz
B _{OM}	Maximum output swing bandwidth	V _O = V _{OH} , R _L = 100 kΩ,	C _L = 20 pF, See Figure 1	25 °C		35		kHz
				0 °C		40		
				70 °C		30		
B ₁	Unity-gain bandwidth	V _i = 10 mV, See Figure 3	C _L = 20 pF,	25 °C		635		kHz
				0 °C		710		
				70 °C		510		
φ _m	Phase margin	V _i = 10 mV, C _L = 20 pF,	f = B ₁ , See Figure 3	25 °C		43°		
				0 °C		44°		
				70 °C		42°		

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Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC27M4 and TLC27M9 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit will give the same result.

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Operational Amplifiers



FIGURE 1. UNITY-GAIN AMPLIFIER

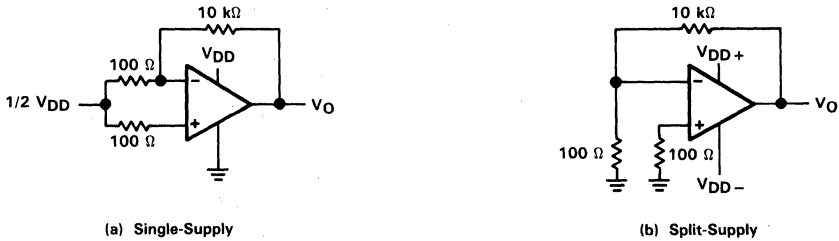


FIGURE 2. NOISE TEST CIRCUIT

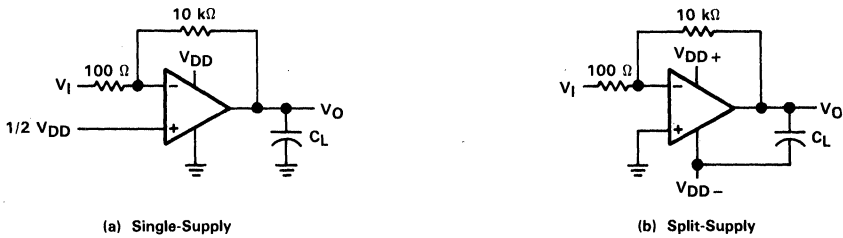


FIGURE 3. GAIN-OF-100 INVERTING AMPLIFIER

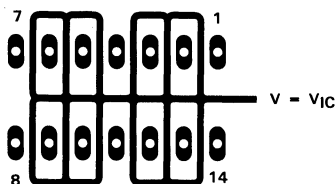
PARAMETER MEASUREMENT INFORMATION

input bias current

Because of the high input impedance of the TLC27M4 and TLC27M9 op amps, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs will be shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the "open-socket" leakage readings from the readings obtained with a device in the test socket.

One word of caution . . . many automatic testers as well as some bench-top op amp testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an "open-socket" reading is not feasible using this method.



**FIGURE 4. ISOLATION METAL AROUND DEVICE INPUTS
 (J AND N DUAL-IN-LINE-PACKAGE)**

low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

PARAMETER MEASUREMENT INFORMATION

input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture will result in leakage and contact resistance, which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

2

Operational Amplifiers

full-power response

Full-power response, the frequency above which the op amp slew rate limits the output voltage swing, is often specified two ways . . . full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for "significant" distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

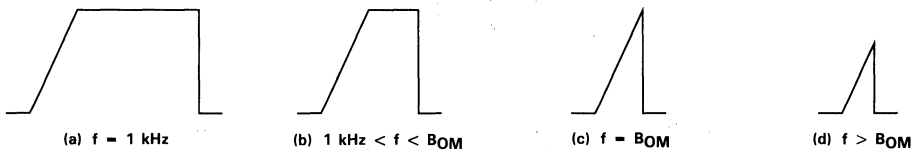


FIGURE 5. FULL-POWER-RESPONSE OUTPUT SIGNAL

test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

TLC27M4, TLC27M4A, TLC27M4B, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC27M4
INPUT OFFSET VOLTAGE

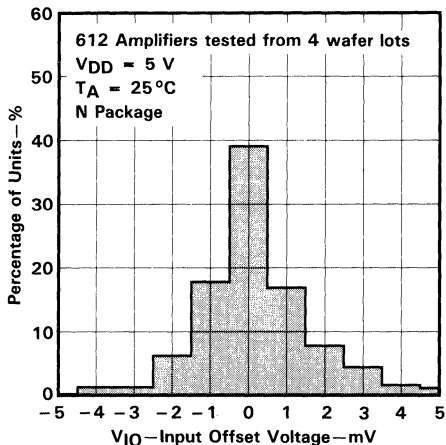


FIGURE 6

DISTRIBUTION OF TLC27M4
INPUT OFFSET VOLTAGE

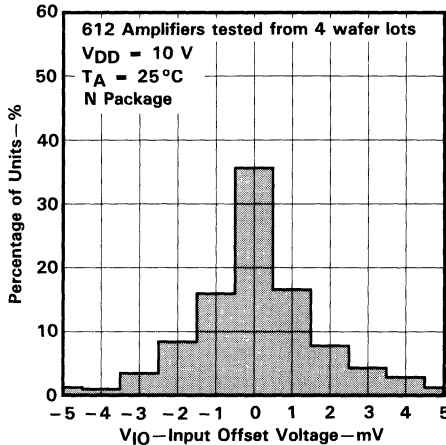


FIGURE 7

DISTRIBUTION OF TLC27M4 AND TLC27M9
INPUT OFFSET VOLTAGE
TEMPERATURE COEFFICIENT

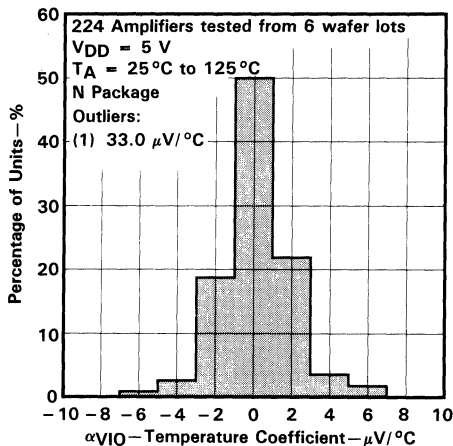


FIGURE 8

DISTRIBUTION OF TLC27M4 AND TLC27M9
INPUT OFFSET VOLTAGE
TEMPERATURE COEFFICIENT

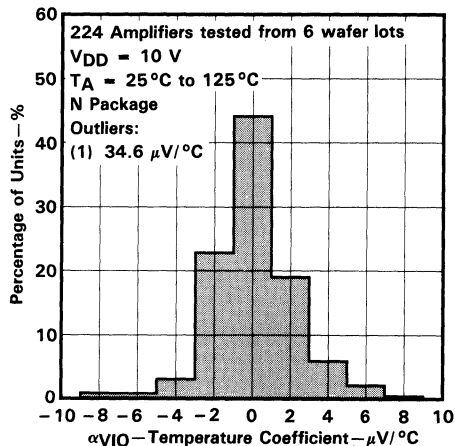


FIGURE 9

TLC27M4, TLC27M4A, TLC27M4B, TLC27M9
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

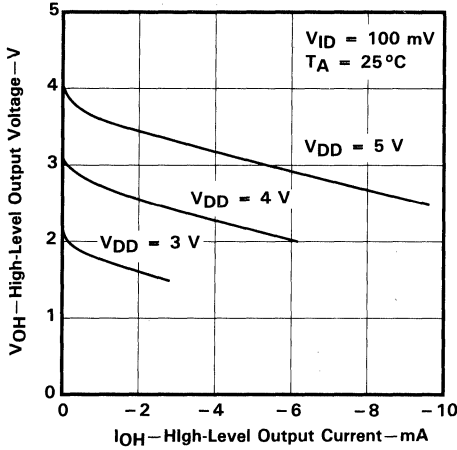


FIGURE 10

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 HIGH-LEVEL OUTPUT CURRENT

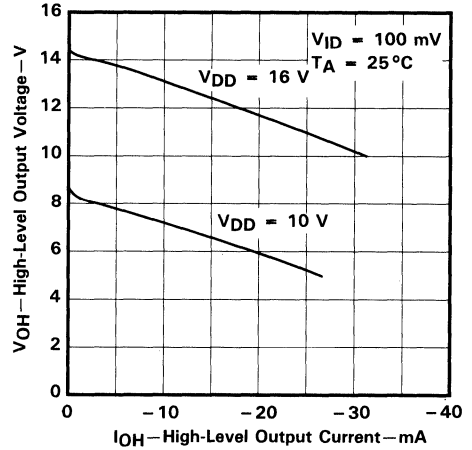


FIGURE 11

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 SUPPLY VOLTAGE

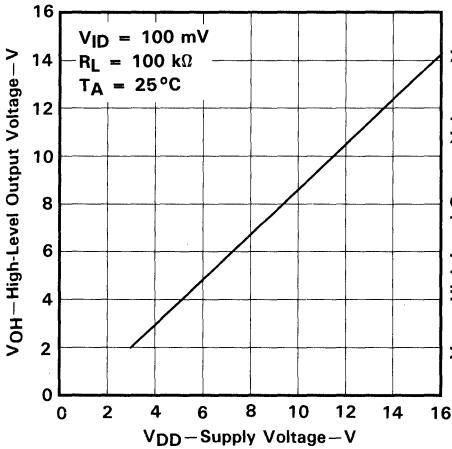


FIGURE 12

HIGH-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

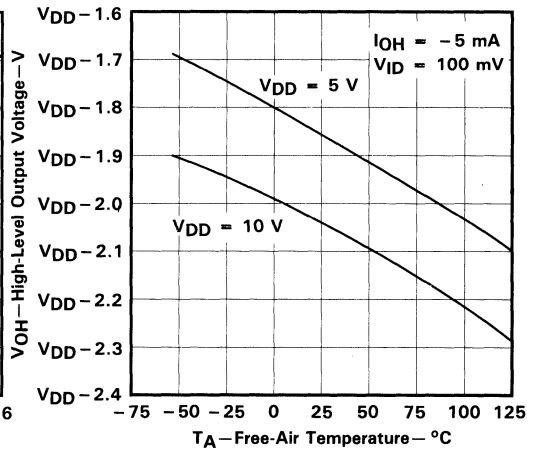


FIGURE 13

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC27M4, TLC27M4A, TLC27M4B, TLC27M9
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

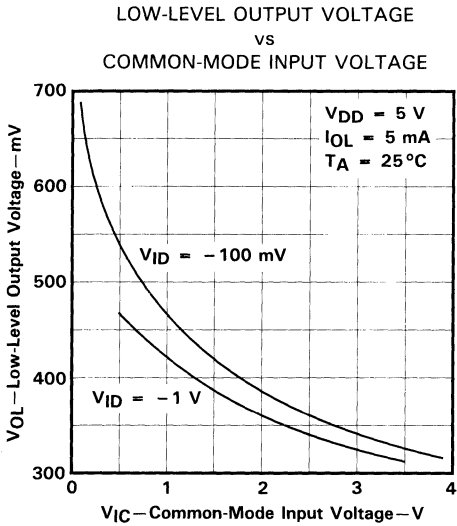


FIGURE 14

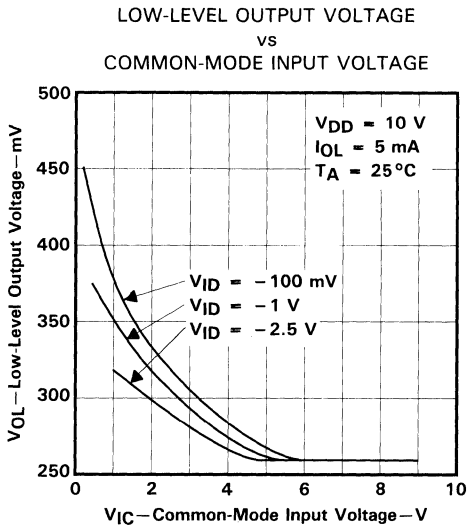


FIGURE 15

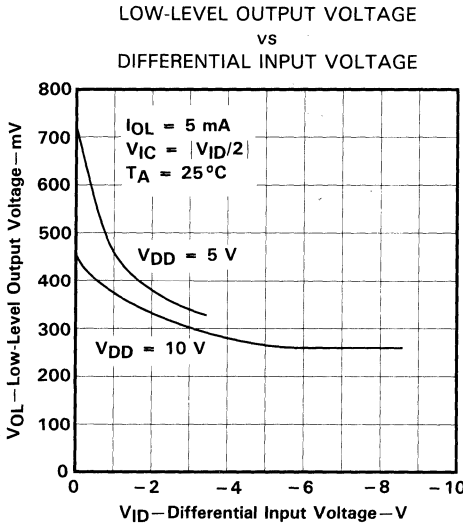


FIGURE 16

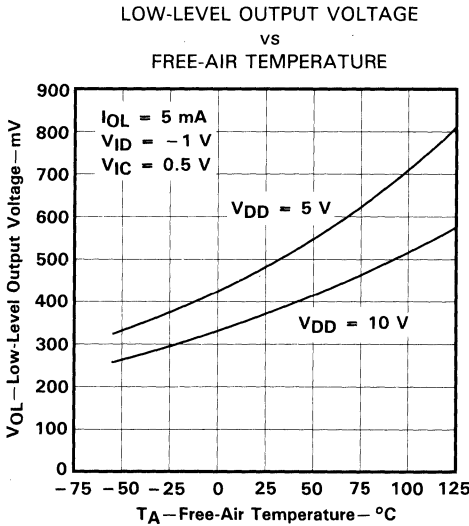


FIGURE 17

2
Operational Amplifiers

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

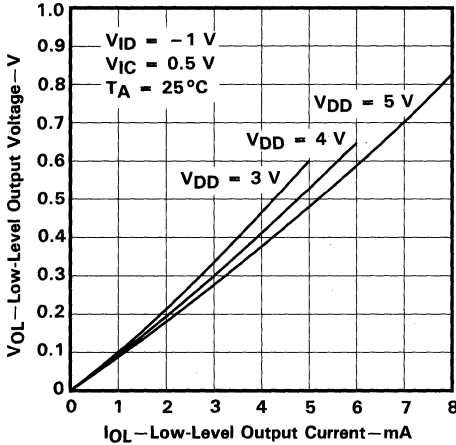


FIGURE 18

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

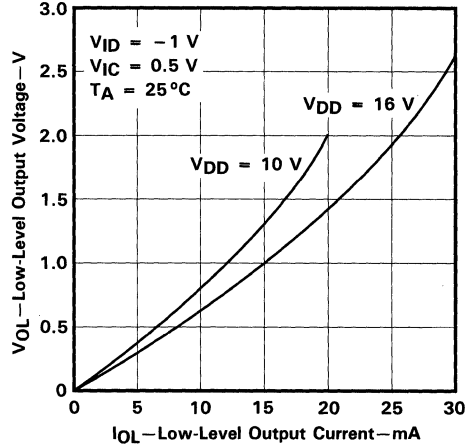


FIGURE 19

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 SUPPLY VOLTAGE

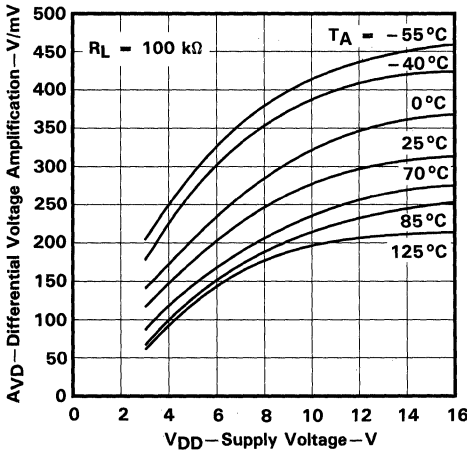


FIGURE 20

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 FREE-AIR TEMPERATURE

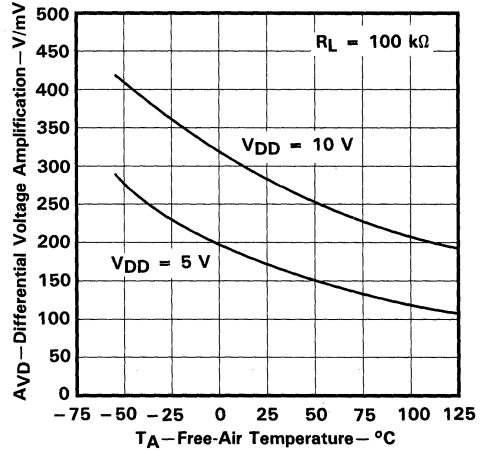


FIGURE 21

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

INPUT BIAS CURRENT AND INPUT OFFSET CURRENT
 vs
 FREE-AIR TEMPERATURE

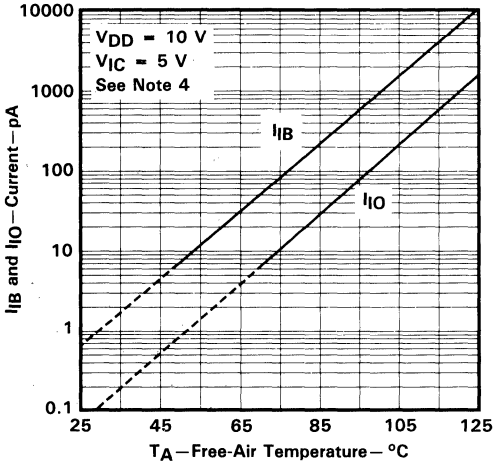


FIGURE 22

COMMON-MODE
 INPUT VOLTAGE POSITIVE LIMIT
 vs
 SUPPLY VOLTAGE

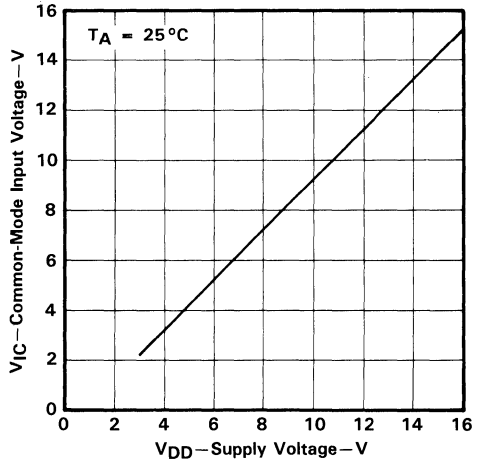


FIGURE 23

SUPPLY CURRENT
 vs
 SUPPLY VOLTAGE

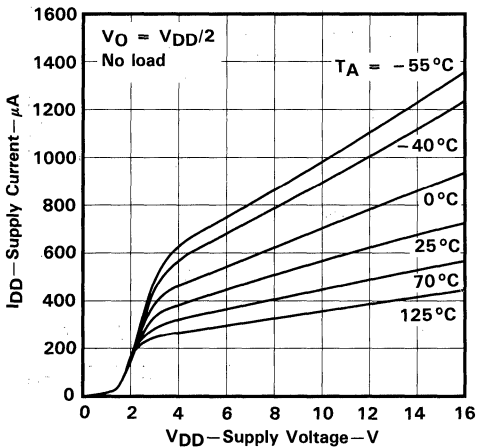


FIGURE 24

SUPPLY CURRENT
 vs
 FREE-AIR TEMPERATURE

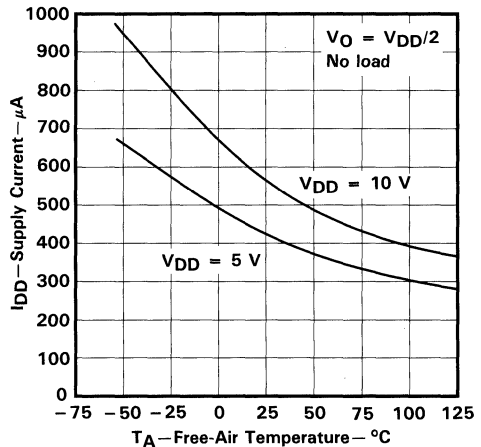


FIGURE 25

2
 Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.
 NOTE 4: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

TYPICAL CHARACTERISTICS†

SLEW RATE
 vs
 SUPPLY VOLTAGE

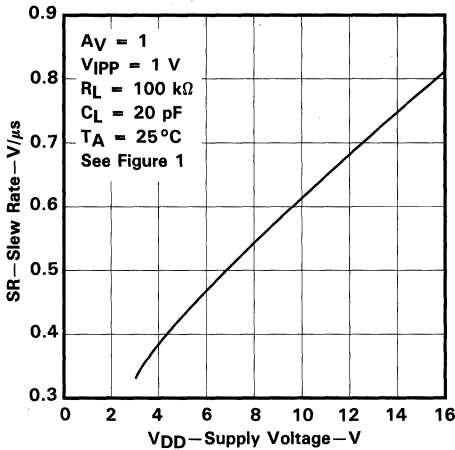


FIGURE 26

SLEW RATE
 vs
 FREE-AIR TEMPERATURE

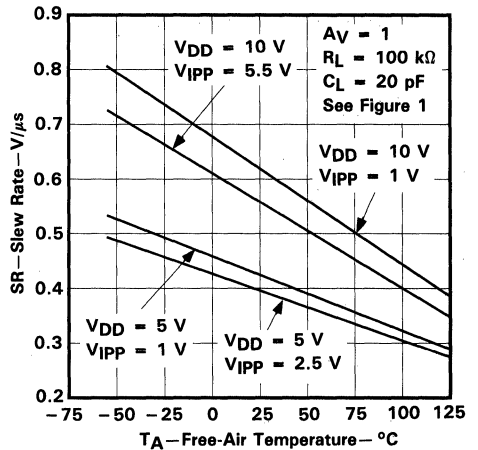


FIGURE 27

NORMALIZED SLEW RATE
 vs
 FREE-AIR TEMPERATURE

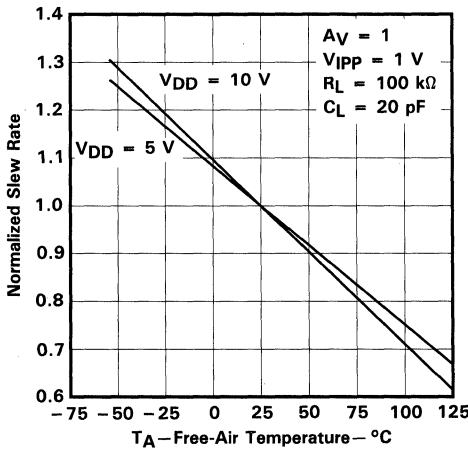


FIGURE 28

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 vs
 FREQUENCY

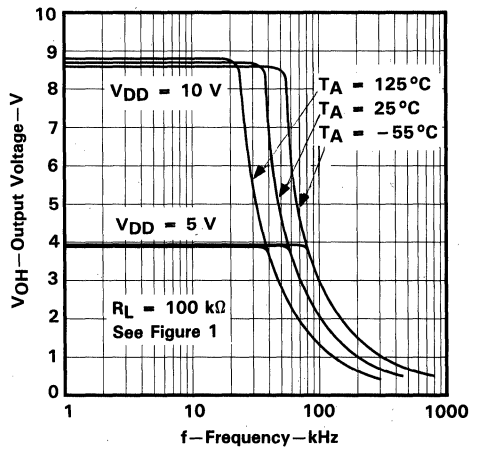


FIGURE 29

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

UNITY-GAIN BANDWIDTH
 vs
 FREE-AIR TEMPERATURE

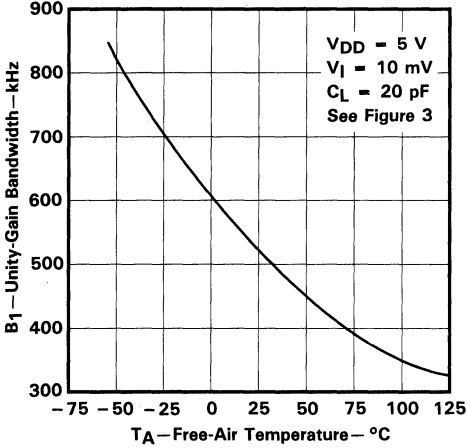


FIGURE 30

UNITY-GAIN BANDWIDTH
 vs
 SUPPLY VOLTAGE

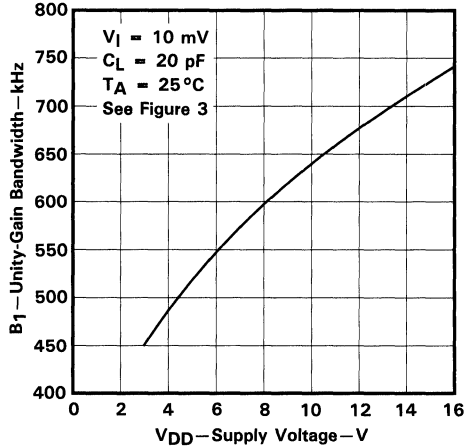


FIGURE 31

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

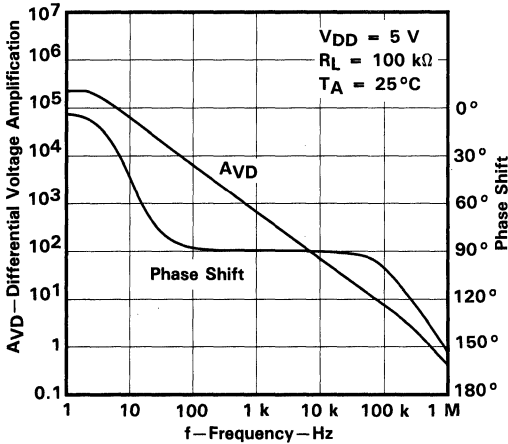


FIGURE 32

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 vs
 FREQUENCY

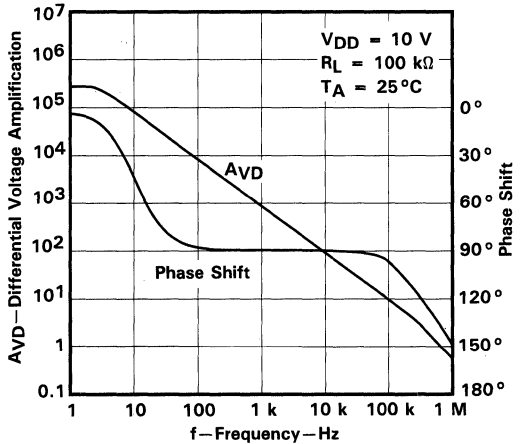


FIGURE 33

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

PHASE MARGIN
 vs
 SUPPLY VOLTAGE

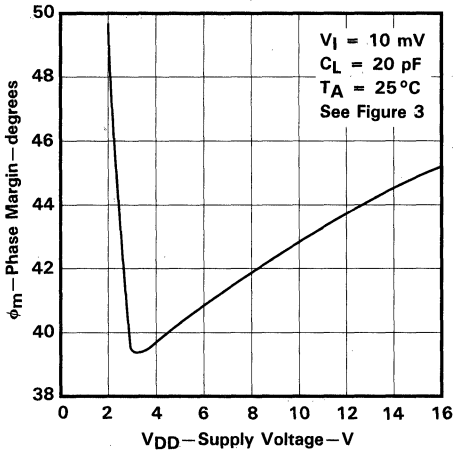


FIGURE 34

PHASE MARGIN
 vs
 FREE-AIR TEMPERATURE

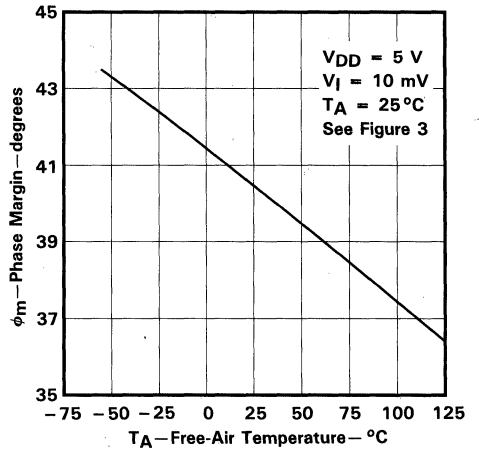


FIGURE 35

PHASE MARGIN
 vs
 CAPACITIVE LOAD

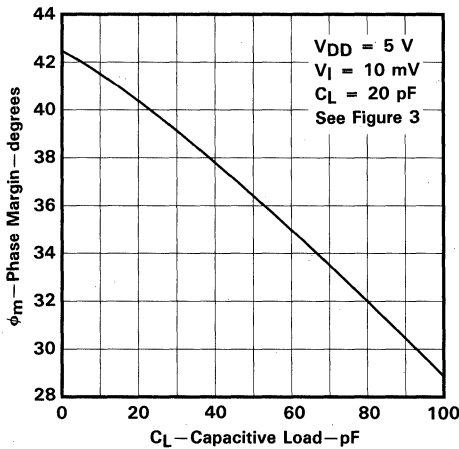


FIGURE 36

EQUIVALENT INPUT NOISE VOLTAGE
 vs
 FREQUENCY

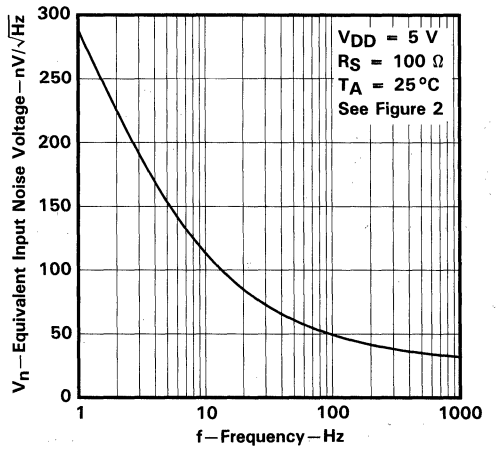


FIGURE 37

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL APPLICATION DATA

single-supply operation

While the TLC27M4 and TLC27M9 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common-mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC27M4 and TLC27M9 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC27M4 and TLC27M9 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

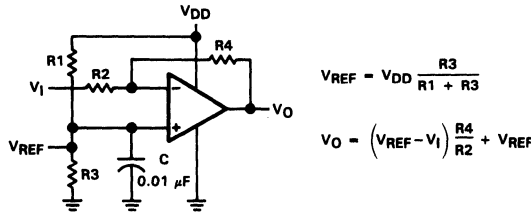


FIGURE 38. INVERTING AMPLIFIER WITH VOLTAGE REFERENCE

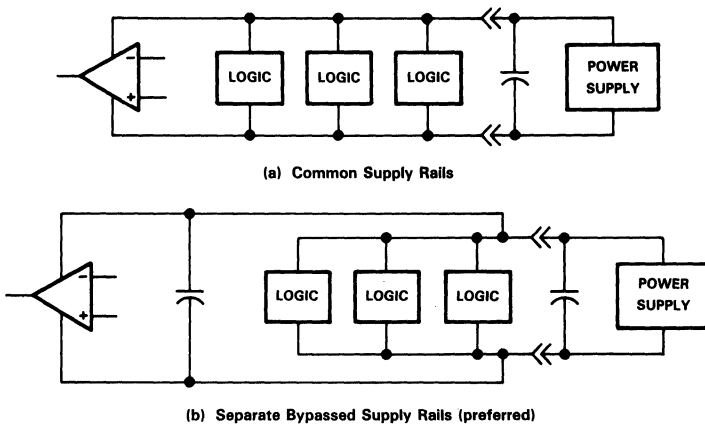


FIGURE 39. COMMON VS SEPARATE SUPPLY RAILS

TYPICAL APPLICATION DATA

input characteristics

The TLC27M4 and TLC27M9 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at $V_{DD} - 1\text{ V}$ at $T_A = 25^\circ\text{C}$ and at $V_{DD} - 1.5\text{ V}$ at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC27M4 and TLC27M9 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically $0.1\ \mu\text{V}/\text{month}$, including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC27M4 and TLC27M9 are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.

noise performance

The noise specifications in op amp circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC27M4 and TLC27M9 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than $50\ \text{k}\Omega$, since bipolar devices exhibit greater noise currents.

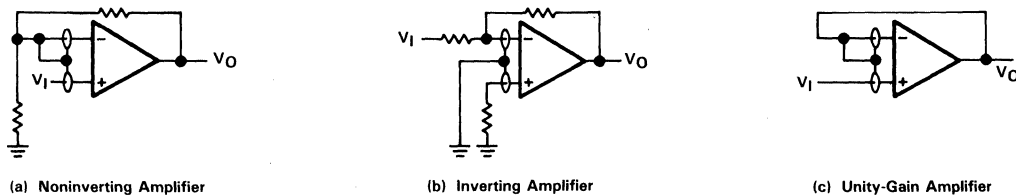


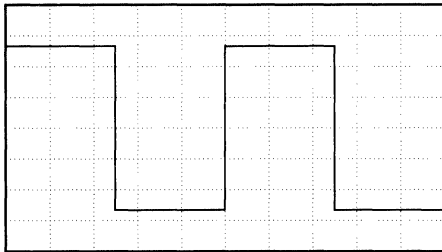
FIGURE 40. GUARD-RING SCHEMES

output characteristics

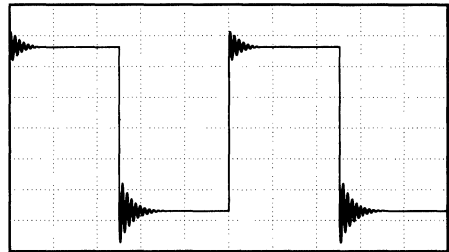
The output stage of the TLC27M4 and TLC27M9 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

All operating characteristics of the TLC27M4 and TLC27M9 were measured using a 20-pF load. The devices will drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance will alleviate the problem.

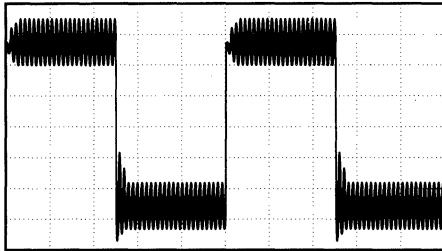
TYPICAL APPLICATION DATA



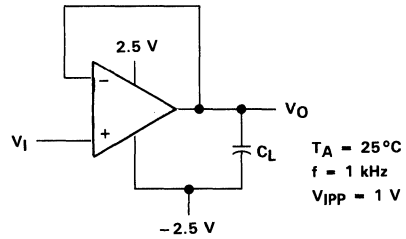
(a) $C_L = 20 \text{ pF}$, $R_L = \text{No load}$



(b) $C_L = 170 \text{ pF}$, $R_L = \text{No load}$



(c) $C_L = 190 \text{ pF}$, $R_L = \text{No load}$



(d) Test Circuit

FIGURE 41. EFFECT OF CAPACITIVE LOADS AND TEST CIRCUIT

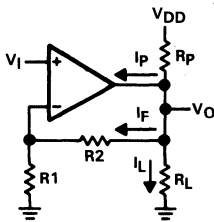
Although the TLC27M4 and TLC27M9 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor (R_p) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pull-down transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately 60Ω and 180Ω , depending on how hard the op amp input is driven. With very low values of R_p , a voltage offset from 0 V at the output will occur. Second, pullup resistor R_p acts as a drain load to N4 and the gain of the op amp is reduced at output voltage levels where N5 is not supplying the output current.

feedback

Op amp circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

TLC27M4, TLC27M4A, TLC27M4B, TLC27M9
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA



$$R_p = \frac{V_{DD} - V_O}{I_F + I_L + I_p}$$

I_p = Pullup current required by the op amp (typically 500 μ A)

FIGURE 42. RESISTIVE PULLUP TO INCREASE V_{OH}

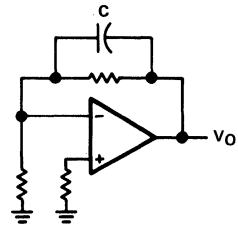


FIGURE 43. COMPENSATION FOR INPUT CAPACITANCE

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Operational Amplifiers

electrostatic discharge protection

The TLC27M4 and TLC27M9 incorporate an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

latchup

Because CMOS devices are susceptible to latchup due to their inherent parasitic thyristors, the TLC27M4 and TLC27M9 inputs and outputs were designed to withstand -100 -mA surge currents without sustaining latchup; however, techniques should be used to reduce the chance of latchup whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μ F typical) located across the supply rails as close to the device as possible.

The current path established if latchup occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latchup occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latchup occurring increases with increasing temperature and supply voltages.

TLC27M4, TLC27M4A, TLC27M4B, TLC27M9
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

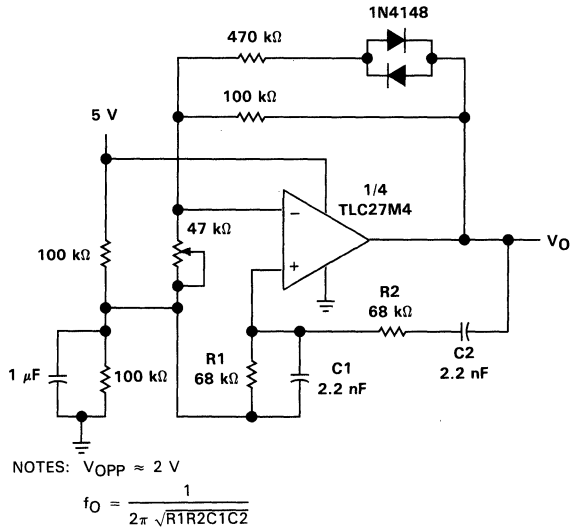


FIGURE 44. WIEN OSCILLATOR

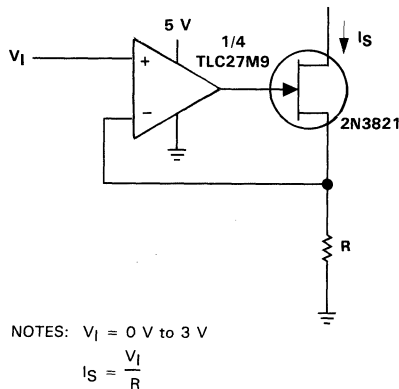
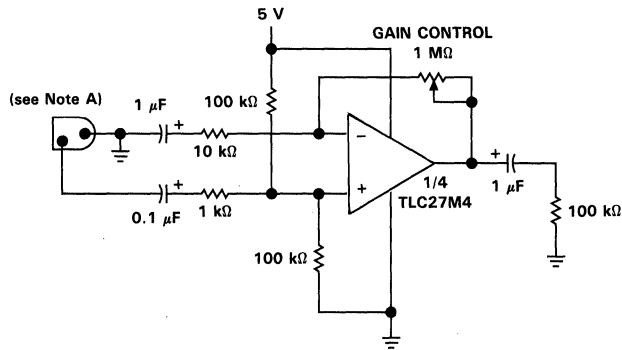


FIGURE 45. PRECISION LOW-CURRENT SINK

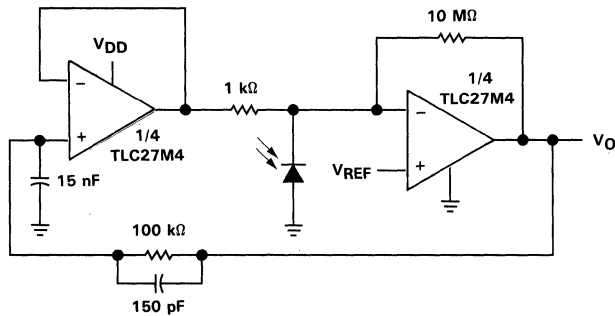
TLC27M4, TLC27M4A, TLC27M4B, TLC27M9
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA



NOTE A: Low to medium impedance dynamic mike.

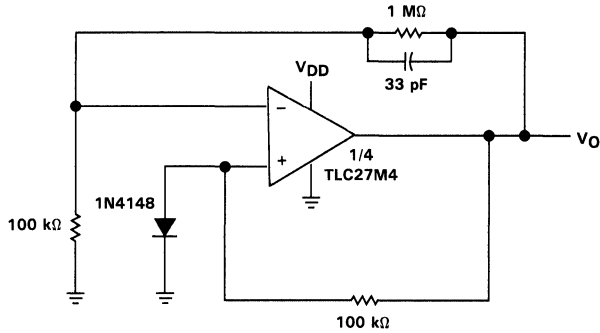
FIGURE 46. MICROPHONE PREAMPLIFIER



NOTES: $V_{DD} = 4\text{ V to }15\text{ V}$
 $V_{REF} = 0\text{ V to }V_{DD} - 2\text{ V}$

FIGURE 47. PHOTO DIODE AMPLIFIER WITH AMBIENT LIGHT REJECTION

TYPICAL APPLICATION DATA



NOTES: $V_{DD} = 8\text{ V to }16\text{ V}$
 $V_O = 5\text{ V, }10\text{ mA}$

FIGURE 48. 5-V LOW-POWER VOLTAGE REGULATOR

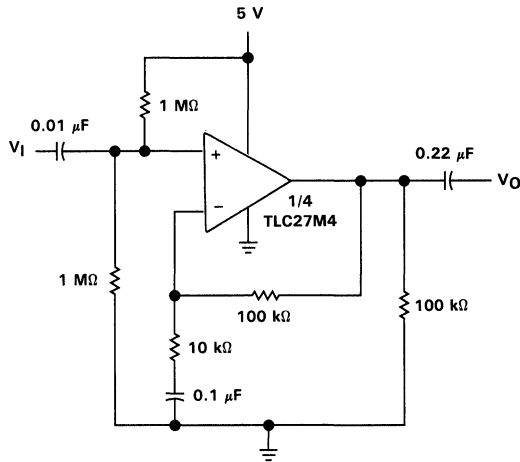


FIGURE 49. SINGLE-RAIL AC AMPLIFIER

1. The first step in the design of an operational amplifier is the selection of the active device. This is typically a bipolar junction transistor (BJT) or a metal-oxide-semiconductor field-effect transistor (MOSFET). The choice of device depends on the required performance characteristics, such as gain, bandwidth, and input/output impedance.

2

Operational Amplifiers

Operational amplifiers (op-amps) are a class of high-gain, DC-coupled voltage amplifiers with differential inputs. They are designed to perform a wide variety of mathematical operations, such as addition, subtraction, integration, and differentiation. The basic op-amp configuration consists of a differential input stage, a voltage divider network, and a feedback loop. The feedback loop is used to control the gain and bandwidth of the amplifier. The most common op-amp configuration is the inverting amplifier, which has a gain of $-R_f/R_{in}$, where R_f is the feedback resistor and R_{in} is the input resistor. Other common configurations include the non-inverting amplifier, the voltage follower, and the summing amplifier. The op-amp is a versatile and powerful tool for analog signal processing, and it is widely used in a wide range of applications, from audio amplification to signal conditioning and control systems.

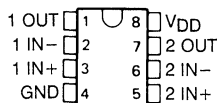
LinCMOS™ μ POWER PRECISION DUAL OPERATIONAL AMPLIFIERS

TLC1078

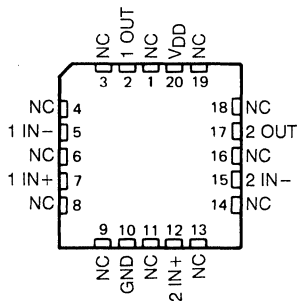
D3146, AUGUST 1988—REVISED OCTOBER 1988

- Power Dissipation as Low as 10 μ W Typ per Amplifier
- Operates on a Single Silver-Oxide Watch Battery, $V_{DD} = 1.4$ V Min
- $V_{IO} \dots 450$ μ V Max in DIP and Small-Outline Package
- Input Offset Voltage Drift $\dots 0.1$ μ V/Month Typ, Including the First 30 Days
- High-Impedance LinCMOS™ Inputs $I_{IB} = 0.6$ pA Typ
- High Open-Loop Gain $\dots 800,000$ Typ
- Output Drive Capability > 20 mA
- Slew Rate $\dots 47$ V/ms Typ
- Common-Mode Input Voltage Range Extends Below the Negative Rail
- Output Voltage Range Includes Negative Rail
- On-Chip ESD-Protection Circuitry
- Small-Outline Package Option Also Available in Tape and Reel

D, JG, OR P PACKAGE
(TOP VIEW)

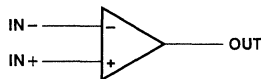


FK PACKAGE
(TOP VIEW)

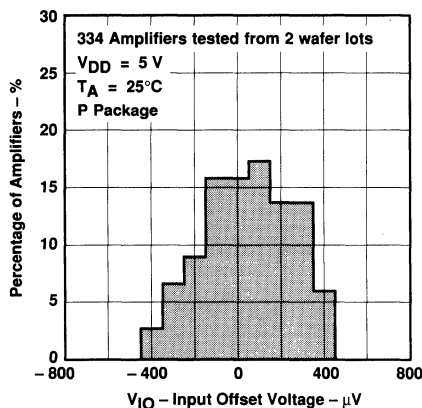


NC—No internal connection

symbol (each amplifier)



DISTRIBUTION OF TLC1078
INPUT OFFSET VOLTAGE



description

The TLC1078 operational amplifier offers ultra-low offset voltage, high gain, 110-kHz bandwidth, 47-V/ms slew rate, and just 150- μ W power dissipation per amplifier.

With a supply voltage of 1.4 V, common-mode input to the negative rail, and output swing to the negative rail, the TLC1078C is an ideal solution for low-voltage battery-operated systems. The 20-mA output drive capability means that the TLC1078 can easily drive small resistive and large capacitive loads when needed, while maintaining ultra-low standby power dissipation.

Since this device is functionally compatible as well as pin compatible with the TLC27L2 and TLC27L7, the TLC1078 easily upgrades existing designs that can benefit from its improved performance.

The TLC1078 incorporates internal ESD-protection circuits that will prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care

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2

Operational Amplifiers

TLC1078 LinCMOS™ μ POWER PRECISION DUAL OPERATIONAL AMPLIFIERS

description (continued)

should be exercised when handling these devices as exposure to ESD may result in degradation of the device parametric performance. The TLC1078 design also inhibits latchup of the device inputs and outputs even with surge currents as large as 100 mA.

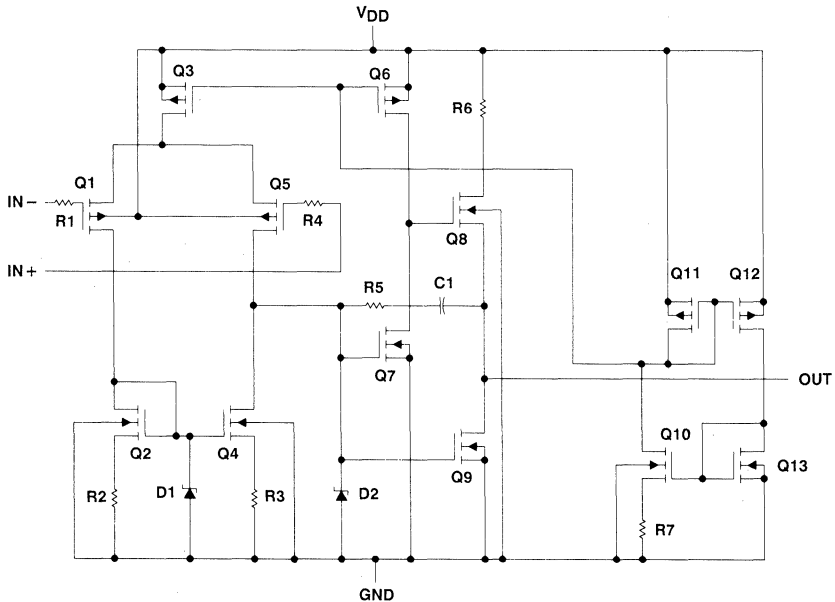
The M- suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I- suffix devices are characterized for operation from -40°C to 85°C . The C- suffix devices are characterized for operation from 0°C to 70°C . The wide range of packaging options includes small-outline and chip-carrier versions for high-density system applications.

AVAILABLE OPTIONS

T _A	PACKAGE			
	SMALL OUTLINE (D) SEE NOTE 1	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	TLC1078CD	—	TLC1078CJG	TLC1078CP
-40°C to 85°C	TLC1078ID	—	TLC1078IJG	TLC1078IP
-55°C to 125°C	—	TLC1078MFK	TLC1078MJG	—

NOTE 1: The D package is available taped and reeled. Add the suffix R to the device type (e.g., TLC1078CDR).

equivalent schematic (each amplifier)



TLC1078C

LinCMOS™ μ POWER PRECISION DUAL OPERATIONAL AMPLIFIERS

2 Operational Amplifiers

electrical characteristics over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$V_{DD} = 5\text{ V}$			$V_{DD} = 10\text{ V}$			UNIT		
		T_A	MIN	TYP	MAX	MIN	TYP		MAX	
V_{IO} Input offset voltage	$V_O = 1.4\text{ V}$, $V_{IC} = 0\text{ V}$, $R_S = 50\ \Omega$, $R_I = 1\text{ M}\Omega$	25°C		160	450		180	600	μV	
		Full range			800			950		
α_{VIO} Temperature coefficient of input offset voltage		25°C to 70°C		1.1			1		$\mu\text{V}/^\circ\text{C}$	
I_{IO} Input offset current (see Note 5)	$V_O = V_{DD}/2$, $V_{IC} = V_{DD}/2$	25°C		0.1			0.1		pA	
		70°C			7	300		7		300
I_{IB} Input bias current (see Note 5)		25°C		0.6			0.7		pA	
		70°C		40	600		50	600		
V_{ICR} Common-mode input voltage range (see Note 6)		25°C	-0.2 to 4	-0.3 to 4.2			-0.2 to 9	-0.3 to 9.2	V	
		Full range	-0.2 to 3.5				-0.2 to 8.5		V	
V_{OH} High-level output voltage	$V_{ID} = 100\text{ mV}$, $R_L = 1\text{ M}\Omega$	25°C	3.2	4.1			8.2	8.9	V	
		0°C	3.2	4.1			8.2	8.9		
		70°C	3.2	4.2			8.2	8.9		
V_{OL} Low-level output voltage	$V_{ID} = -100\text{ mV}$, $I_{OL} = 0$	25°C		0	25		0	25	mV	
		0°C		0	25		0	25		
		70°C		0	25		0	25		
A_{VD} Large-signal differential voltage amplification	$R_L = 1\text{ M}\Omega$ See Note 7	25°C	250	525			500	850	V/mV	
		0°C	250	680			500	1010		
		70°C	200	380			350	660		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$	25°C	70	95			75	97	dB	
		0°C	70	95			75	97		
		70°C	70	95			75	97		
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$, $V_O = 1.4\text{ V}$	25°C	75	98			75	98	dB	
		0°C	75	98			75	98		
		70°C	75	98			75	98		
I_{DD} Supply current (two amplifiers)	$V_O = V_{DD}/2$, $V_{IC} = V_{DD}/2$, No load	25°C		20	34			29	46	μA
		0°C		24	42			36	66	
		70°C		16	28			22	40	

NOTES: 5. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
 6. This range also applies to each input individually.
 7. At $V_{DD} = 5\text{ V}$, $V_O = 0.25\text{ V to }2\text{ V}$; at $V_{DD} = 10\text{ V}$, $V_O = 1\text{ V to }6\text{ V}$.

operating characteristics

PARAMETER	TEST CONDITIONS	$V_{DD} = 5\text{ V}$			$V_{DD} = 10\text{ V}$			UNIT
		T_A	MIN	TYP	MAX	MIN	TYP	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, $V_{IPP} = 1\text{ V}$, See Figure 1	25°C		32			47	V/ms
		0°C		35			51	
		70°C		27			38	
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, $R_S = 100\ \Omega$	25°C		68			68	$\text{nv}/\sqrt{\text{Hz}}$
B_1 Unity-gain bandwidth	$C_L = 20\text{ pF}$, See Figure 2	25°C		85			110	kHz
		0°C		100			125	
		70°C		65			90	
ϕ_m Phase margin at unity gain	$C_L = 20\text{ pF}$, See Figure 2	25°C		34°			38°	
		0°C		36°			40°	
		70°C		30°			34°	

TLC1078I

LinCMOS™ μ POWER PRECISION DUAL OPERATIONAL AMPLIFIERS

electrical characteristics over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T _A	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω , R _I = 1 M Ω	25°C		160	450		180	600	μ V
		Full range			950			1100	
α V _{IO} Temperature coefficient of input offset voltage	R _S = 50 Ω , R _I = 1 M Ω	25°C to 85°C		1.1			1		μ V/°C
I _{IO} Input offset current (see Note 5)		V _O = V _{DD} /2,	25°C		0.1			0.1	
	85°C			24	1000		26	1000	
I _{IB} Input bias current (see Note 5)	V _{IC} = V _{DD} /2	25°C		0.6			0.7		pA
		85°C		200	2000		220	2000	
V _{ICR} Common-mode input voltage range (see Note 6)		25°C	-0.2 to 4	-0.3 to 4.2		-0.2 to 9	-0.3 to 9.2		V
		Full range	-0.2 to 3.5			-0.2 to 8.5			V
V _{OH} High-level output voltage	V _{ID} = 100 mV, R _L = 1 M Ω	25°C	3.2	4.1		8.2	8.9		V
		-40°C	3.2	4.1		8.2	8.9		
		85°C	3.2	4.2		8.2	8.9		
V _{OL} Low-level output voltage	V _{ID} = -100 mV, I _{OL} = 0	25°C		0	25		0	25	mV
		-40°C		0	25		0	25	
		85°C		0	25		0	25	
A _{VD} Large-signal differential voltage amplification	R _L = 1 M Ω See Note 7	25°C	250	525		500	850		V/mV
		-40°C	250	900		500	1550		
		85°C	150	300		250	585		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min	25°C	70	95		75	97		dB
		-40°C	70	95		75	97		
		85°C	70	95		75	97		
k _{SVR} Supply-voltage rejection ratio (Δ V _{DD} / Δ V _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V	25°C	75	98		75	98		dB
		-40°C	75	98		75	98		
		85°C	75	98		75	98		
I _{DD} Supply current (two amplifiers)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C		20	34		29	46	μ A
		-40°C		31	54		50	86	
		85°C		15	26		20	36	

NOTES: 5. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
6. This range also applies to each input individually.
7. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

operating characteristics

PARAMETER	TEST CONDITIONS	T _A	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	R _L = 1 M Ω , C _L = 20 pF, V _I PP = 1 V, See Figure 1	25°C		32			47		V/ms
		-40°C		39			59		
		85°C		25			34		
V _n Equivalent input noise voltage	f = 1 kHz, R _S = 100 Ω	25°C		68			68		nv/ \sqrt Hz
B ₁ Unity-gain bandwidth	C _L = 20 pF, See Figure 2	25°C		85			110		kHz
		-40°C		130			155		
		85°C		55			80		
ϕ _m Phase margin at unity gain	C _L = 20 pF, See Figure 2	25°C		34°			38°		
		-40°C		38°			42°		
		85°C		28°			32°		

2
Operational Amplifiers



TLC1078M

LinCMOS™ μ POWER PRECISION DUAL OPERATIONAL AMPLIFIERS

2 Operational Amplifiers

electrical characteristics over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
		T _A	MIN	TYP	MAX	MIN	TYP	
V _{IO} Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω , R _I = 1 M Ω	25°C	160	450	180	600		μ V
α _{IO} Temperature coefficient of input offset voltage		Full range	1250		1400			μ V/°C
I _{IO} Input offset current (see Note 5)	V _O = V _{DD} / 2,	25°C	0.1		0.1			pA
I _{IB} Input bias current (see Note 5)	V _{IC} = V _{DD} / 2	125°C	1.4	15	1.8	15		nA
		25°C	0.6		0.7			pA
		125°C	9	35	10	35		nA
V _{ICR} Common-mode input voltage range (see Note 6)		25°C	0 to 4	-0.3 to 4.2	0 to 9	-0.3 to 9.2		V
		Full range	3.5 to		8.5 to			V
V _{OH} High-level output voltage	V _{ID} = 100 mV, R _L = 1 M Ω	25°C	3.2	4.1	8.2	8.9		V
		-55°C	3.2	4.1	8.2	8.8		
		125°C	3.2	4.2	8.2	9		
V _{OL} Low-level output voltage	V _{ID} = -100 mV, I _{OL} = 0	25°C	0 to 25		0 to 25			mV
		-55°C	0 to 25		0 to 25			
		125°C	0 to 25		0 to 25			
A _{VD} Large-signal differential voltage amplification	R _L = 1 M Ω See Note 7	25°C	250	525	500	850		V/mV
		-55°C	250	950	500	1750		
		125°C	75	200	150	380		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min	25°C	70	95	75	97		dB
		-55°C	70	95	75	97		
		125°C	70	85	75	91		
k _{SVR} Supply-voltage rejection ratio (Δ V _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V	25°C	75	98	75	98		dB
		-55°C	70	98	70	98		
		125°C	70	98	70	98		
I _{DD} Supply current (two amplifiers)	V _O = V _{DD} / 2, V _{IC} = V _{DD} / 2, No load	25°C	20 to 34		29 to 46			μ A
		-55°C	35 to 60		56 to 96			
		125°C	14 to 24		18 to 30			

NOTES: 5. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

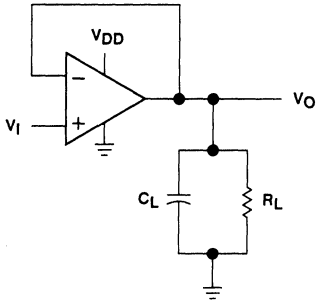
6. This range also applies to each input individually.

7. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

operating characteristics

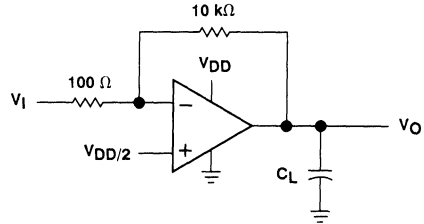
PARAMETER	TEST CONDITIONS	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
		T _A	MIN	TYP	MAX	MIN	TYP	
SR Slow rate at unity gain	R _L = 1 M Ω , C _L = 20 pF, V _{IPP} = 1 V, See Figure 1	25°C	32		47			V/ms
		-55°C	41		63			
		125°C	20		27			
V _n Equivalent input noise voltage	f = 1 kHz, R _S = 100 Ω	25°C	68		68			nv/ \sqrt Hz
		25°C	85		110			
B ₁ Unity-gain bandwidth	C _L = 20 pF, See Figure 2	-55°C	140		165			kHz
		125°C	45		70			
		25°C	34°		38°			
ϕ _m Phase margin at unity gain	C _L = 20 pF, See Figure 2	-55°C	39°		43°			
		125°C	25°		29°			

PARAMETER MEASUREMENT INFORMATION



C_L includes fixture capacitance.

FIGURE 1. SLEW RATE TEST CIRCUIT



C_L includes fixture capacitance.

FIGURE 2. UNITY-GAIN BANDWIDTH AND PHASE MARGIN TEST CIRCUIT

TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC1078 INPUT OFFSET VOLTAGE TEMPERATURE COEFFICIENT

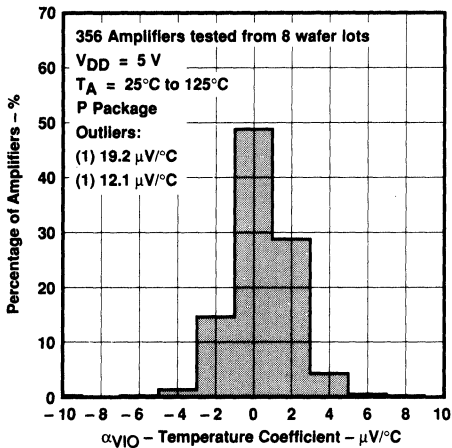


FIGURE 3

DISTRIBUTION OF TLC1078 INPUT OFFSET VOLTAGE TEMPERATURE COEFFICIENT

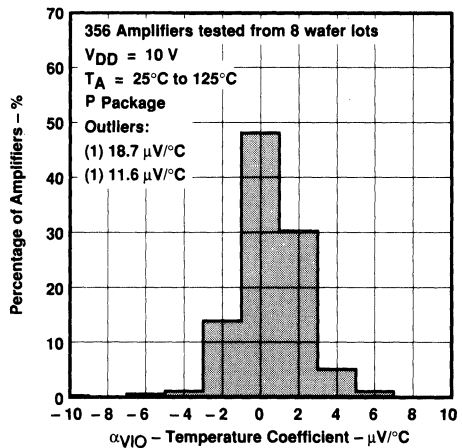


FIGURE 4

TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT VOLTAGE
 VS
 HIGH-LEVEL OUTPUT CURRENT

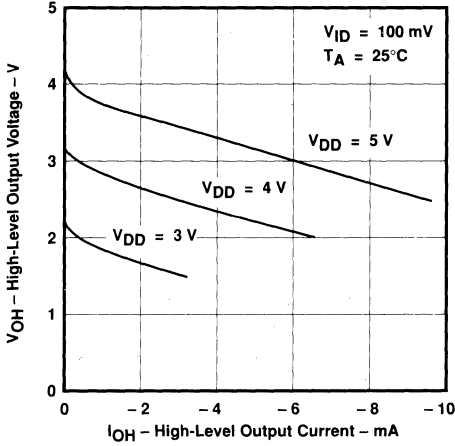


FIGURE 5

HIGH-LEVEL OUTPUT VOLTAGE
 VS
 HIGH-LEVEL OUTPUT CURRENT

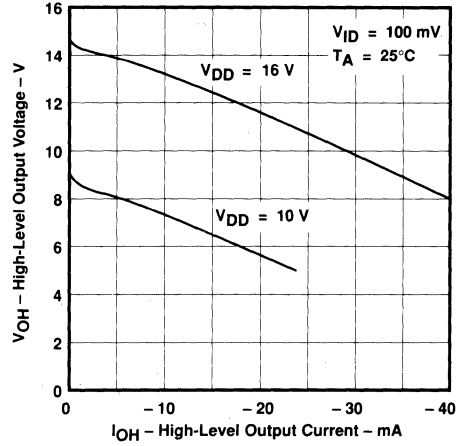


FIGURE 6

HIGH-LEVEL OUTPUT VOLTAGE
 VS
 SUPPLY VOLTAGE

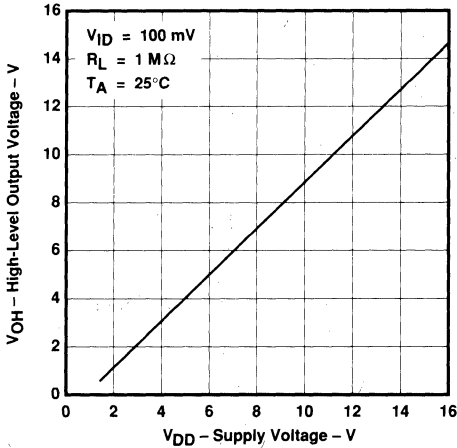


FIGURE 7

HIGH-LEVEL OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

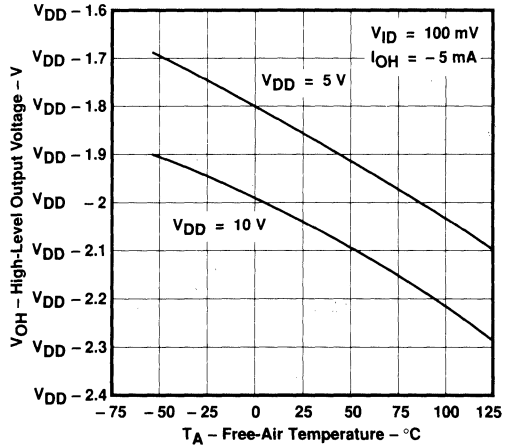


FIGURE 8

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
 VS
 COMMON-MODE INPUT VOLTAGE

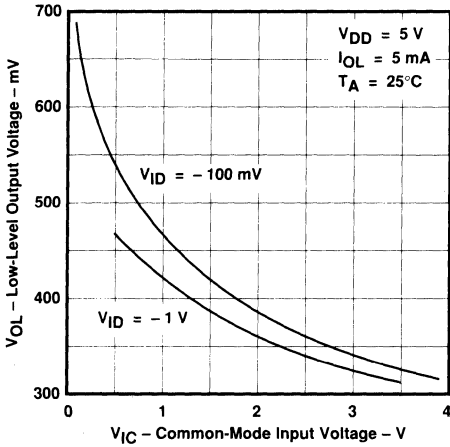


FIGURE 9

LOW-LEVEL OUTPUT VOLTAGE
 VS
 COMMON-MODE INPUT VOLTAGE

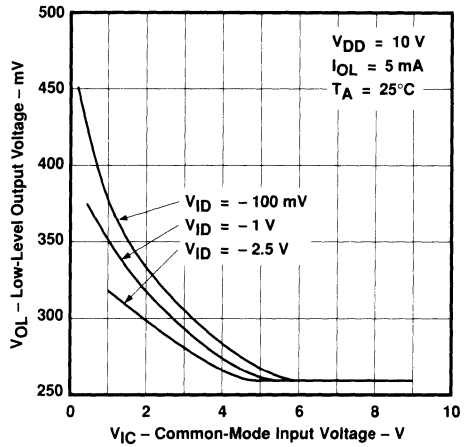


FIGURE 10

LOW-LEVEL OUTPUT VOLTAGE
 VS
 DIFFERENTIAL INPUT VOLTAGE

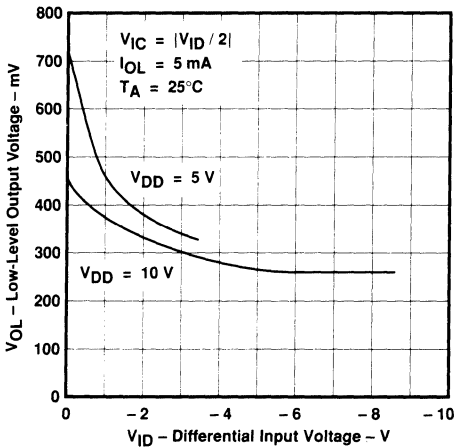


FIGURE 11

LOW-LEVEL OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

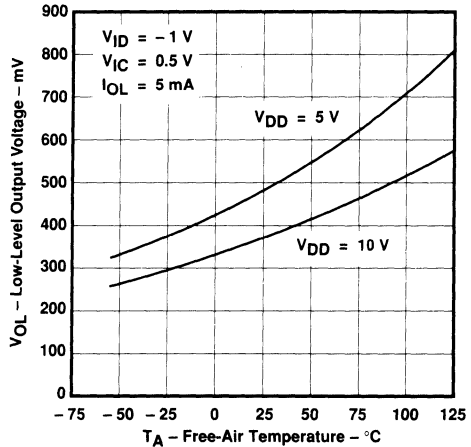


FIGURE 12

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
 VS
 LOW-LEVEL OUTPUT CURRENT

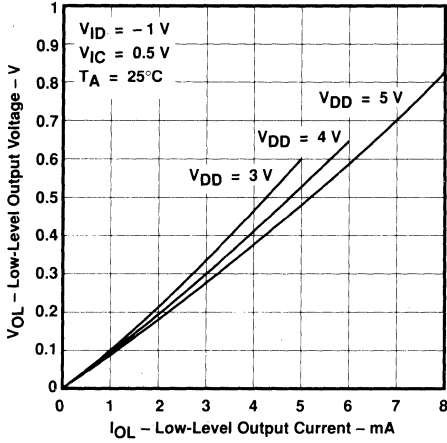


FIGURE 13

LOW-LEVEL OUTPUT VOLTAGE
 VS
 LOW-LEVEL OUTPUT CURRENT

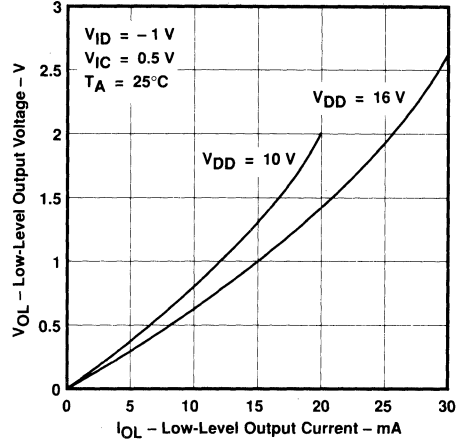


FIGURE 14

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 VS
 SUPPLY VOLTAGE

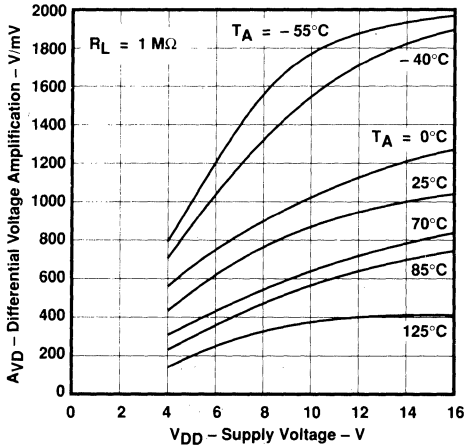


FIGURE 15

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 VS
 FREE-AIR TEMPERATURE

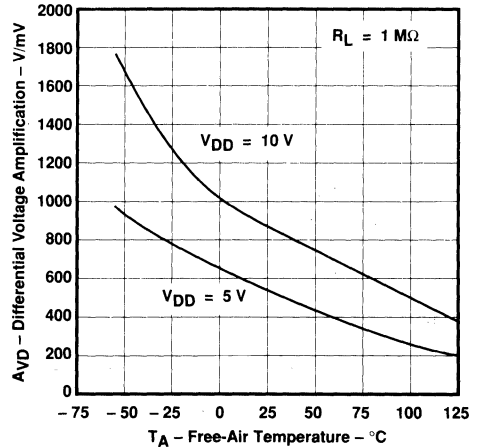


FIGURE 16

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

INPUT BIAS CURRENT AND INPUT OFFSET CURRENT

vs
FREE-AIR TEMPERATURE

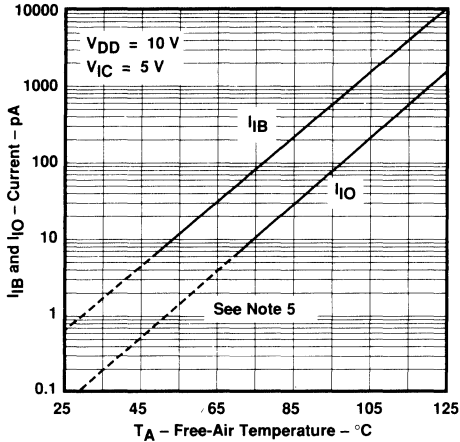


FIGURE 17

COMMON-MODE INPUT VOLTAGE POSITIVE LIMIT

vs
SUPPLY VOLTAGE

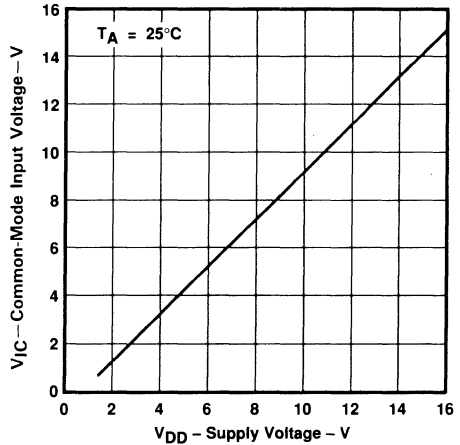


FIGURE 18

SUPPLY CURRENT
vs
SUPPLY VOLTAGE

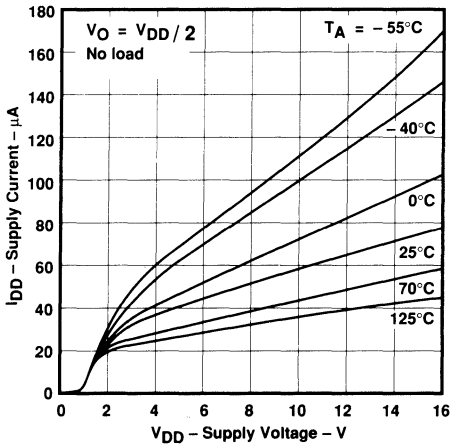


FIGURE 19

SUPPLY CURRENT
vs
FREE-AIR TEMPERATURE

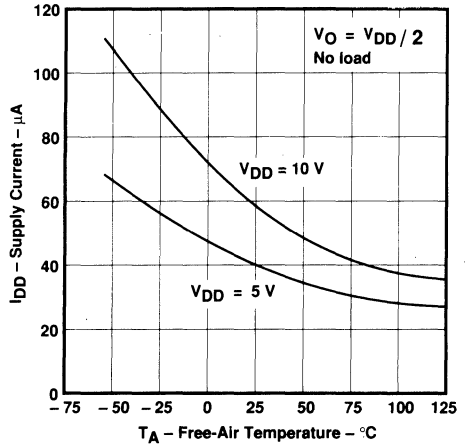


FIGURE 20

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.
NOTE 5: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

TYPICAL CHARACTERISTICS†

2

Operational Amplifiers

SLEW RATE
 VS
 SUPPLY VOLTAGE

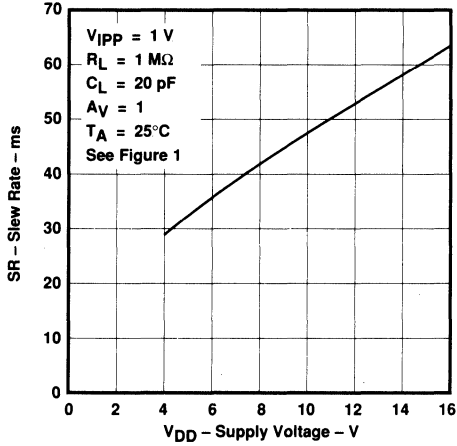


FIGURE 21

SLEW RATE
 VS
 FREE-AIR TEMPERATURE

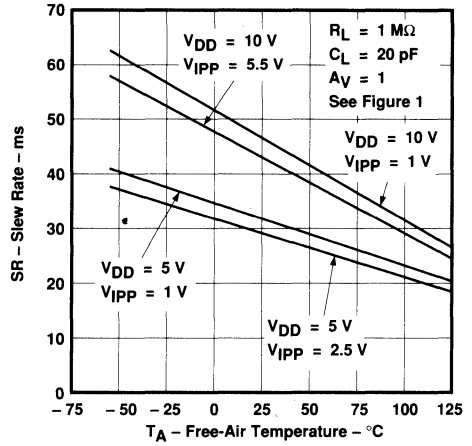


FIGURE 22

NORMALIZED SLEW RATE
 VS
 FREE-AIR TEMPERATURE

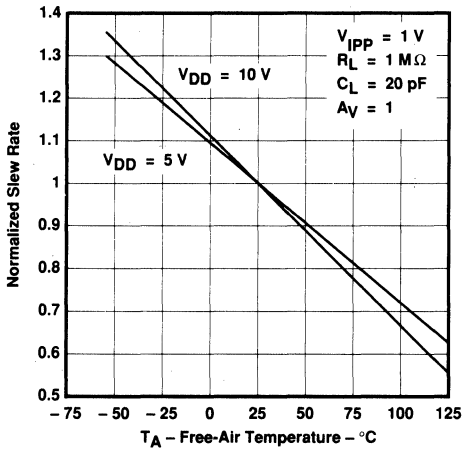


FIGURE 23

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 VS
 FREQUENCY

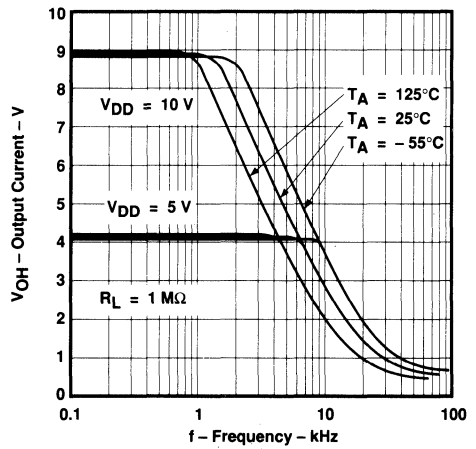


FIGURE 24

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

UNITY-GAIN BANDWIDTH
VS
FREE-AIR TEMPERATURE

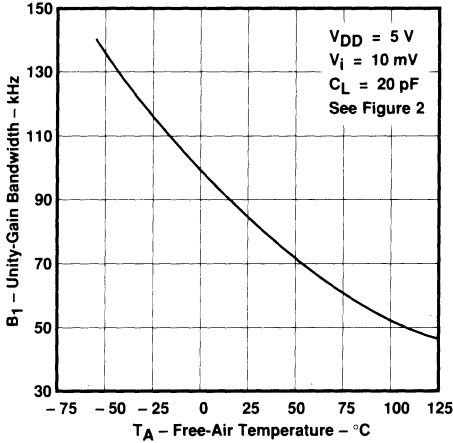


FIGURE 25

UNITY-GAIN BANDWIDTH
VS
SUPPLY VOLTAGE

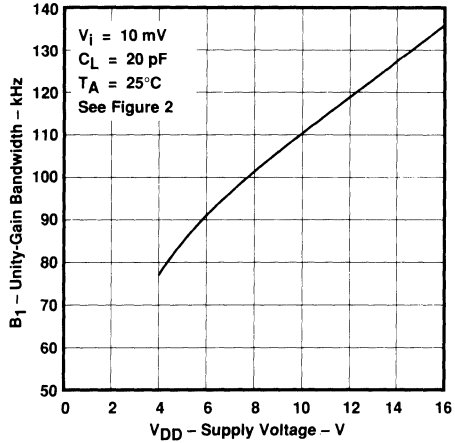


FIGURE 26

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
VS
FREQUENCY

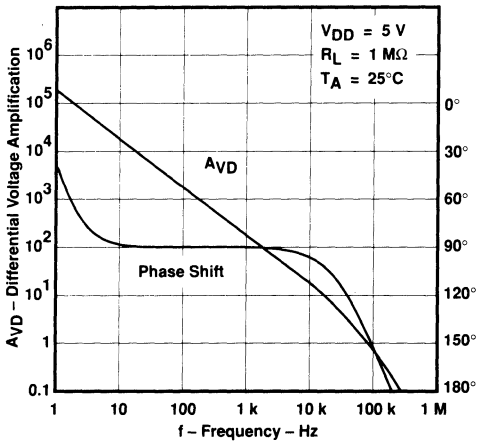


FIGURE 27

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
VS
FREQUENCY

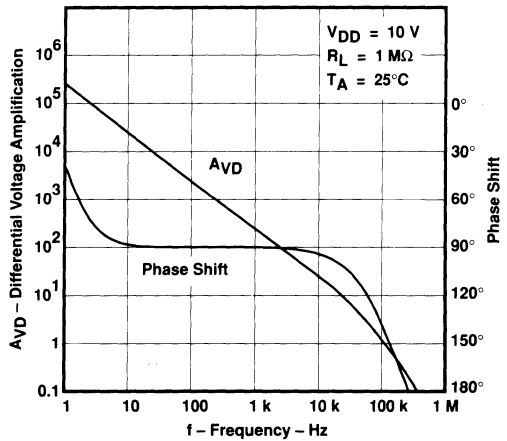


FIGURE 28

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

PHASE MARGIN
 VS
 SUPPLY VOLTAGE

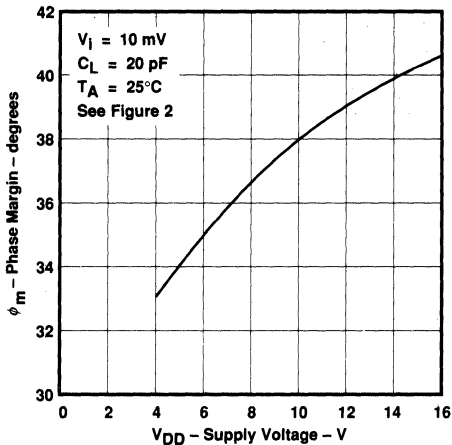


FIGURE 29

PHASE MARGIN
 VS
 FREE-AIR TEMPERATURE

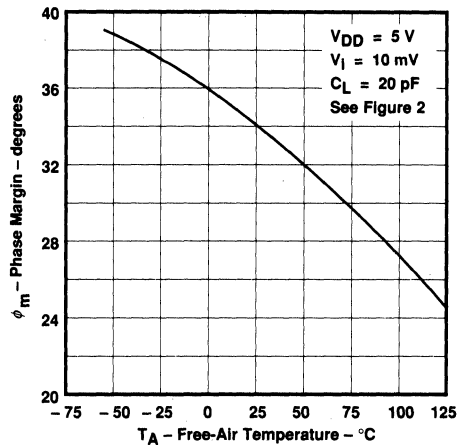


FIGURE 30

PHASE MARGIN
 VS
 CAPACITIVE LOAD

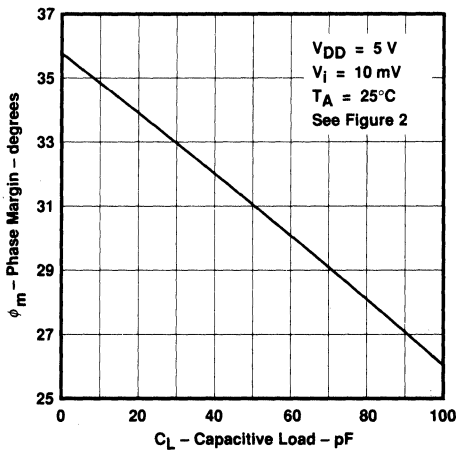


FIGURE 31

EQUIVALENT INPUT NOISE VOLTAGE
 VS
 FREQUENCY

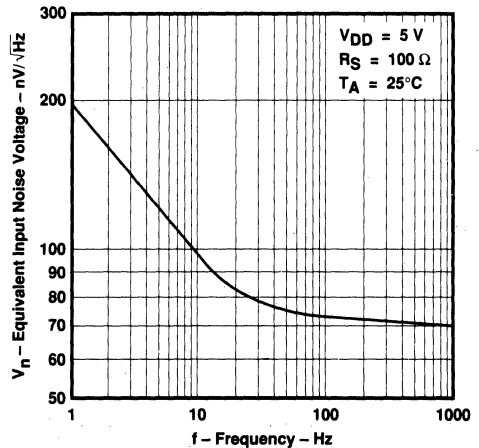


FIGURE 32

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

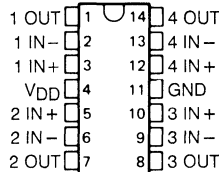
LinCMOS™ μ POWER PRECISION QUAD OPERATIONAL AMPLIFIERS

TLC1079

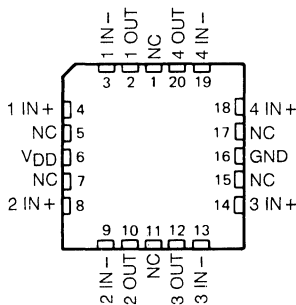
D3147, AUGUST 1988—REVISED OCTOBER 1988

- Power Dissipation as Low as 10 μ W Typ per Amplifier
- Operates on a Single Silver-Oxide Watch Battery, $V_{DD} = 1.4$ V Min
- $V_{IO} \dots 850$ μ V Max in DIP and Small-Outline Package
- Input Offset Voltage Drift $\dots 0.1$ μ V/Month Typ, Including the First 30 Days
- High-Impedance LinCMOS™ Inputs $I_{IB} = 0.6$ pA Typ
- High Open-Loop Gain $\dots 800,000$ Typ
- Output Drive Capability > 20 mA
- Slew Rate $\dots 47$ V/ms Typ
- Common-Mode Input Voltage Range Extends Below the Negative Rail
- Output Voltage Range Includes Negative Rail
- On-Chip ESD-Protection Circuitry
- 14-Pin Small-Outline Package Option Also Available in Tape and Reel

D, J, OR N PACKAGE
(TOP VIEW)



FK PACKAGE
(TOP VIEW)



NC—No internal connection

description

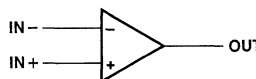
The TLC1079 operational amplifier offers ultra-low offset voltage, high gain, 110-kHz bandwidth, 47-V/ms slew rate, and just 150- μ W power dissipation per amplifier.

With a supply voltage of 1.4 V, common-mode input to the negative rail, and output swing to the negative rail, the TLC1079 is an ideal solution for low-voltage, battery-operated systems. The 20-mA output drive capability means that the TLC1079 can easily drive small resistive and large capacitive loads when needed, while maintaining ultra-low standby power dissipation.

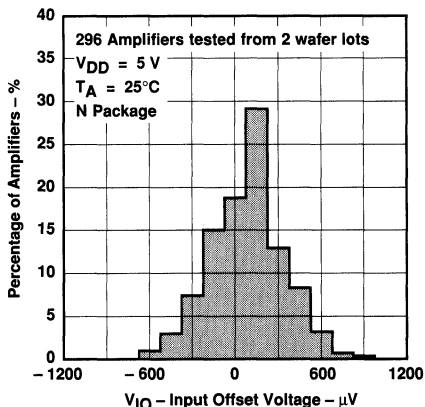
Since this device is functionally compatible as well as pin compatible with the TLC27L4 and TLC27L9, the TLC1079 easily upgrades existing designs that can benefit from its improved performance.

The TLC1079 incorporates internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care

symbol (each amplifier)



DISTRIBUTION OF TLC1079
INPUT OFFSET VOLTAGE



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TLC1079 LinCMOS™ μ POWER PRECISION QUAD OPERATIONAL AMPLIFIERS

description (continued)

should be exercised when handling these devices as exposure to ESD may result in degradation of the device parametric performance. The TLC1079 design also inhibits latchup of the device inputs and outputs even with surge currents as large as 100 mA.

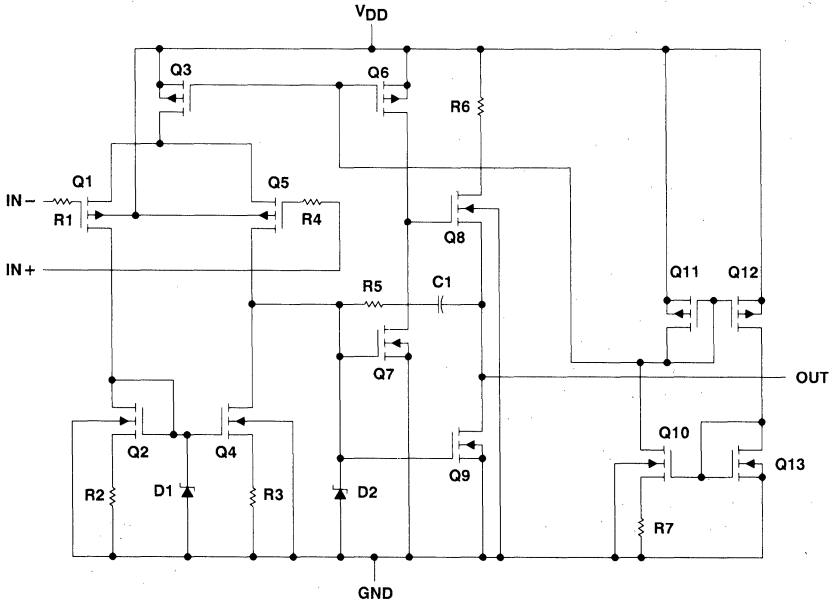
The M- suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I- suffix devices are characterized for operation from -40°C to 85°C . The C- suffix devices are characterized for operation from 0°C to 70°C . The wide range of packaging options includes small-outline and chip-carrier versions for high-density system applications.

AVAILABLE OPTIONS

T _A	PACKAGE			
	SMALL OUTLINE (D) SEE NOTE 1	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	TLC1079CD	—	TLC1079CJ	TLC1079CN
-40°C to 85°C	TLC1079ID	—	TLC1079IJ	TLC1079IN
-55°C to 125°C	—	TLC1079MFK	TLC1079MJ	—

NOTE 1: The D package is available taped and reeled. Add the suffix R to the device type (e.g., TLC1079CDR).

equivalent schematic (each amplifier)



2

Operational Amplifiers

TLC1079C LinCMOS™ μ POWER PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	V _{DD} = 5 V			V _{DD} = 10 V			UNIT	
		T _A	MIN	TYP	MAX	MIN	TYP		MAX
V _{IO} Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω , R _I = 1 M Ω	25°C		190	850			1150	μ V
α V _{IO} Temperature coefficient of input offset voltage		Full range		1200			1500		
I _{IO} Input offset current (see Note 5)	V _O = V _{DD} / 2, V _{IC} = V _{DD} / 2	25°C	0.1			0.1			pA
I _{IB} Input bias current (see Note 5)		70°C	7	300		7	300		
V _{ICR} Common-mode input voltage range (see Note 6)		25°C	-0.2	-0.3		-0.2	-0.3		V
		Full range	to	to		to	to		
V _{OH} High-level output voltage	V _{ID} = 100 mV, R _L = 1 M Ω	25°C	3.2	4.1		8.2	8.9		V
		0°C	3.2	4.1		8.2	8.9		
		70°C	3.2	4.2		8.2	8.9		
V _{OL} Low-level output voltage	V _{ID} = -100 mV, I _{OL} = 0	25°C		0	25		0	25	mV
		0°C		0	25		0	25	
		70°C		0	25		0	25	
A _{VD} Large-signal differential voltage amplification	R _L = 1 M Ω See Note 7	25°C	250	525		500	850		V/mV
		0°C	250	700		500	1010		
		70°C	200	380		350	660		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min	25°C	70	95		75	97		dB
		0°C	70	95		75	97		
		70°C	70	95		75	97		
k _{SVR} Supply-voltage rejection ratio (Δ V _{DD} / Δ V _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V	25°C	75	98		75	98		dB
		0°C	75	98		75	98		
		70°C	75	98		75	98		
I _{DD} Supply current (four amplifiers)	V _O = V _{DD} / 2, V _{IC} = V _{DD} / 2, No load	25°C		40	68		57	92	μ A
		0°C		48	84		72	132	
		70°C		31	56		44	80	

NOTES: 5. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

6. This range also applies to each input individually.

7. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

operating characteristics

PARAMETER	TEST CONDITIONS	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
		T _A	MIN	TYP	MAX	MIN	TYP	
SR Slew rate at unity gain	R _L = 1 M Ω , C _L = 20 pF, V _{IPP} = 1 V, See Figure 1	25°C		32			47	V/ms
		0°C		35			51	
		70°C		27			38	
V _n Equivalent input noise voltage	f = 1 kHz, R _S = 100 Ω	25°C		68			68	nv/ \sqrt Hz
B ₁ Unity-gain bandwidth	C _L = 20 pF, See Figure 2	25°C		85			110	kHz
		0°C		100			125	
		70°C		65			90	
ϕ_m Phase margin at unity gain	C _L = 20 pF, See Figure 2	25°C		34°			38°	
		0°C		36°			40°	
		70°C		30°			34°	

TLC1079I

LinCMOS™ μ POWER PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	V _{DD} = 5 V			V _{DD} = 10 V			UNIT	
		T _A	MIN	TYP	MAX	MIN	TYP		MAX
V _{IO} Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω , R _I = 1 M Ω	25°C	190		850		200 1150		μ V
		Full range			1350		1650		
α V _{IO} Temperature coefficient of input offset voltage		25°C to 85°C	1.1				1		μ V/°C
I _{IO} Input offset current (see Note 5)	V _O = V _{DD} /2,	25°C	0.1				0.1		pA
		85°C	24		1000		26 1000		
I _{IB} Input bias current (see Note 5)	V _{IC} = V _{DD} /2	25°C	0.6				0.7		pA
		85°C	200		2000		220 2000		
V _{ICR} Common-mode input voltage range (see Note 6)		25°C	-0.2	-0.3		-0.2	-0.3		V
			to	to		to	to		
		Full range	4	4.2		9	9.2		V
			-0.2			-0.2			V
			to			to			V
			3.5			8.5			V
V _{OH} High-level output voltage	V _{ID} = 100 mV, R _L = 1 M Ω	25°C	3.2	4.1		8.2	8.9		V
		-40°C	3.2	4.1		8.2	8.9		
		85°C	3.2	4.2		8.2	8.9		
V _{OL} Low-level output voltage	V _{ID} = -100 mV, I _{OL} = 0	25°C		0	25		0	25	mV
		-40°C		0	25		0	25	
		85°C		0	25		0	25	
A _{VD} Large-signal differential voltage amplification	R _L = 1 M Ω See Note 7	25°C	250	525		500	850		V/mV
		-40°C	250	900		500	1550		
		85°C	150	330		250	585		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} min	25°C	70	95		75	97		dB
		-40°C	70	95		75	97		
		85°C	70	95		75	97		
k _{SVR} Supply-voltage rejection ratio (Δ V _{DD} / Δ V _{IO})	V _{DD} = 5 V to 10 V, V _O = 1.4 V	25°C	75	98		75	98		dB
		-40°C	75	98		75	98		
		85°C	75	98		75	98		
I _{DD} Supply current (four amplifiers)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C		40	68		57	92	μ A
		-40°C		62	108		98	172	
		85°C		29	52		40	72	

NOTES: 5. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
 6. This range also applies to each input individually.
 7. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

operating characteristics

PARAMETER	TEST CONDITIONS	V _{DD} = 5 V			V _{DD} = 10 V			UNIT
		T _A	MIN	TYP	MAX	MIN	TYP	
SR Slew rate at unity gain	R _L = 1 M Ω , C _L = 20 pF, V _I PP = 1 V, See Figure 1	25°C		32			47	V/ms
		-40°C		39			59	
		85°C		25			34	
V _n Equivalent input noise voltage	f = 1 kHz, R _S = 100 Ω	25°C		68			68	nv/ \sqrt Hz
B ₁ Unity-gain bandwidth	C _L = 20 pF, See Figure 2	25°C		85			110	kHz
		-40°C		130			155	
		85°C		55			80	
ϕ _m Phase margin at unity gain	C _L = 20 pF, See Figure 2	25°C		34°			38°	
		-40°C		38°			42°	
		85°C		28°			32°	

2

Operational Amplifiers

TLC1079M LinCMOS™ μ POWER PRECISION QUAD OPERATIONAL AMPLIFIERS

electrical characteristics over operating free-air temperature range (unless otherwise noted)

2

Operational Amplifiers

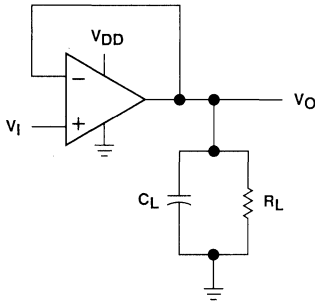
PARAMETER	TEST CONDITIONS	$V_{DD} = 5\text{ V}$			$V_{DD} = 10\text{ V}$			UNIT	
		T_A	MIN	TYP	MAX	MIN	TYP		MAX
V_{IO} Input offset voltage	$V_O = 1.4\text{ V}$, $V_{IC} = 0\text{ V}$, $R_S = 50\ \Omega$, $R_I = 1\text{ M}\Omega$	25°C		190	850		200	1150	μV
		Full range			1600			1900	
α_{VIO} Temperature coefficient of input offset voltage	$V_O = 1.4\text{ V}$, $V_{IC} = 0\text{ V}$, $R_S = 50\ \Omega$, $R_I = 1\text{ M}\Omega$	25°C to 125°C		1.4			1.4		$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current (see Note 5)		$V_O = V_{DD}/2$, $V_{IC} = V_{DD}/2$	25°C		0.1			0.1	
I_{IB} Input bias current (see Note 5)	$V_O = V_{DD}/2$, $V_{IC} = V_{DD}/2$	125°C		1.4	15		1.8	15	nA
		25°C		0.6			0.7		pA
V_{ICR} Common-mode input voltage range (see Note 6)		125°C		9	35		10	35	nA
		25°C	0	-0.3			0	-0.3	V
			to	to			to	to	
V_{OH} High-level output voltage	$V_{ID} = 100\text{ mV}$, $R_L = 1\text{ M}\Omega$	4	4.2			9	9.2		V
		Full range							V
V_{OL} Low-level output voltage	$V_{ID} = -100\text{ mV}$, $I_{OL} = 0$	0				0			
		to				to			
		3.5				8.5			
A_{VD} Large-signal differential voltage amplification	$R_L = 1\text{ M}\Omega$ See Note 7	25°C	3.2	4.1		8.2	8.9		V
		-55°C	3.2	4.1		8.2	8.8		
		125°C	3.2	4.2		8.2	9		
$CMRR$ Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min}$	25°C		0	25		0	25	mV
		-55°C		0	25		0	25	
		125°C		0	25		0	25	
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V}$, $V_O = 1.4\text{ V}$	25°C	250	525		500	850		V/mV
		-55°C	250	950		500	1750		
		125°C	75	200		150	380		
I_{DD} Supply current (four amplifiers)	$V_O = V_{DD}/2$, $V_{IC} = V_{DD}/2$, No load	25°C	70	95		75	97		dB
		-55°C	70	95		75	97		
		125°C	70	85		75	91		
ϕ_m Phase margin at unity gain	$C_L = 20\text{ pF}$, See Figure 2	25°C	75	98		75	98		dB
		-55°C	70	98		70	98		
		125°C	70	98		70	98		
B_1 Unity-gain bandwidth	$C_L = 20\text{ pF}$, See Figure 2	25°C		40	68		57	92	μA
		-55°C		69	120		111	192	
		125°C		27	48		35	60	

NOTES: 5. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
6. This range also applies to each input individually.
7. At $V_{DD} = 5\text{ V}$, $V_O = 0.25\text{ V to }2\text{ V}$; at $V_{DD} = 10\text{ V}$, $V_O = 1\text{ V to }6\text{ V}$.

operating characteristics

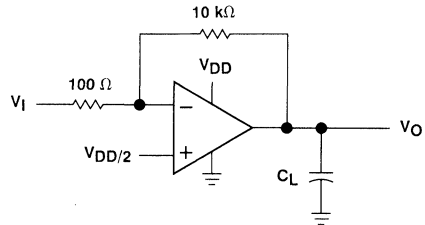
PARAMETER	TEST CONDITIONS	$V_{DD} = 5\text{ V}$			$V_{DD} = 10\text{ V}$			UNIT
		T_A	MIN	TYP	MAX	MIN	TYP	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, $V_{Ipp} = 1\text{ V}$, See Figure 1	25°C		32			47	V/ms
		-55°C		41			63	
		125°C		20			27	
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, $R_S = 100\ \Omega$	25°C		68			68	nv/ $\sqrt{\text{Hz}}$
B_1 Unity-gain bandwidth	$C_L = 20\text{ pF}$, See Figure 2	25°C		85			110	kHz
		-55°C		140			165	
		125°C		45			70	
ϕ_m Phase margin at unity gain	$C_L = 20\text{ pF}$, See Figure 2	25°C		34°			38°	
		-55°C		39°			43°	
		125°C		25°			29°	

PARAMETER MEASUREMENT INFORMATION



C_L includes fixture capacitance.

FIGURE 1. SLEW RATE TEST CIRCUIT



C_L includes fixture capacitance.

FIGURE 2. UNITY-GAIN BANDWIDTH AND PHASE MARGIN TEST CIRCUIT

TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC1079
INPUT OFFSET VOLTAGE
TEMPERATURE COEFFICIENT

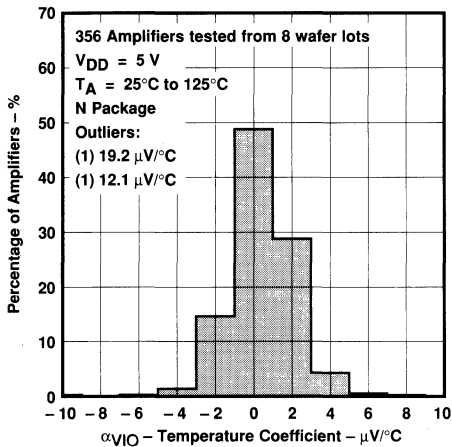


FIGURE 3

DISTRIBUTION OF TLC1079
INPUT OFFSET VOLTAGE
TEMPERATURE COEFFICIENT

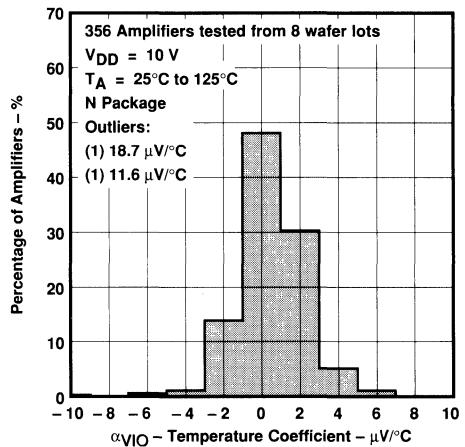


FIGURE 4

TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT VOLTAGE
 VS
 HIGH-LEVEL OUTPUT CURRENT

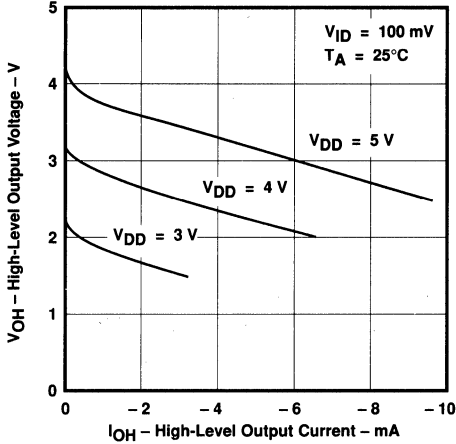


FIGURE 5

HIGH-LEVEL OUTPUT VOLTAGE
 VS
 HIGH-LEVEL OUTPUT CURRENT

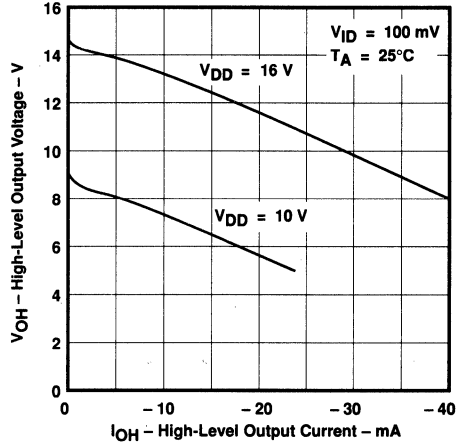


FIGURE 6

HIGH-LEVEL OUTPUT VOLTAGE
 VS
 SUPPLY VOLTAGE

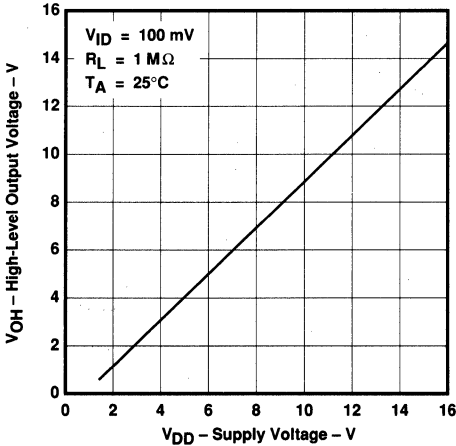


FIGURE 7

HIGH-LEVEL OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

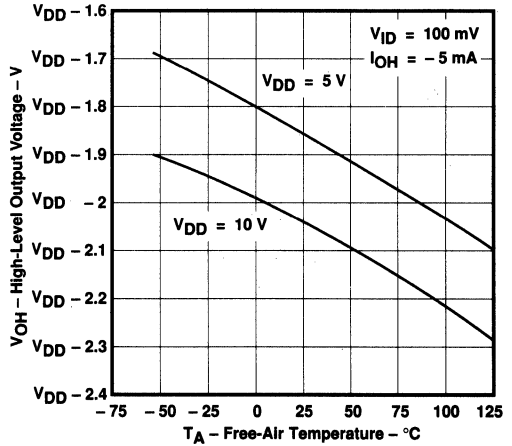
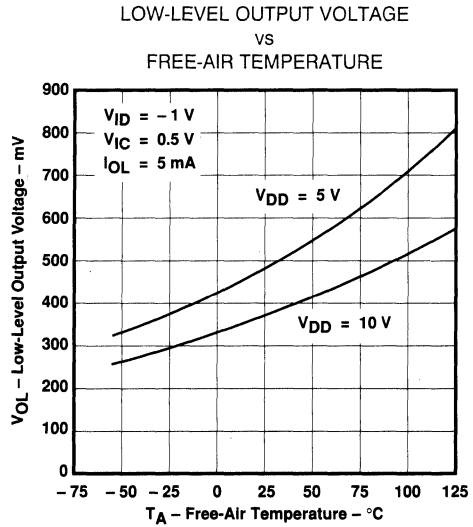
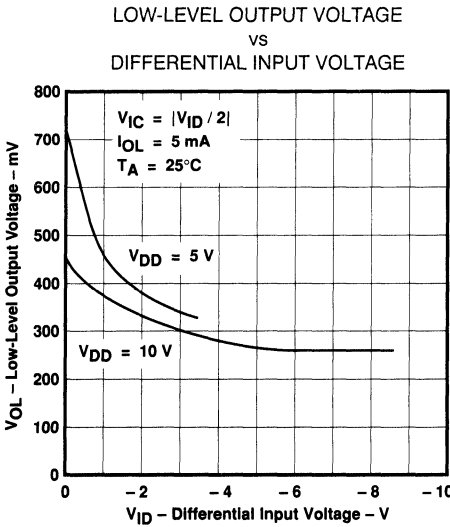
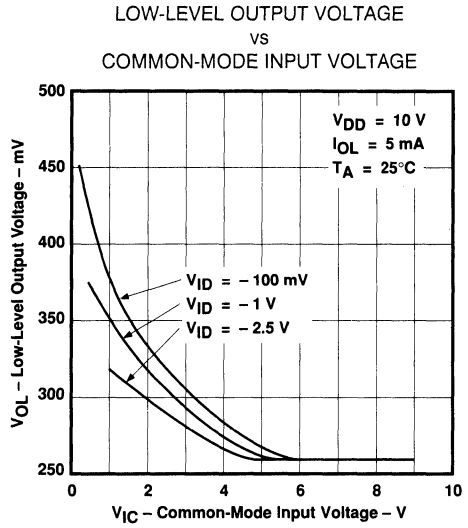
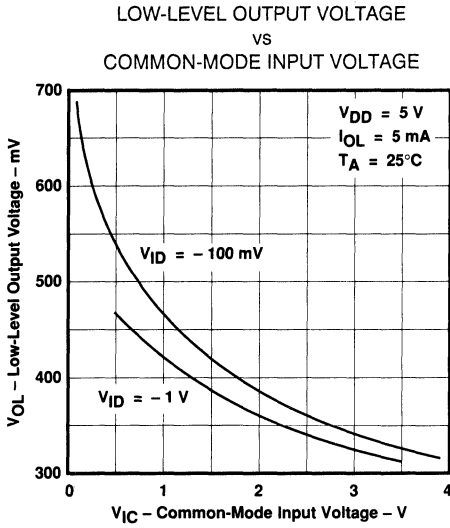


FIGURE 8

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
 VS
 LOW-LEVEL OUTPUT CURRENT

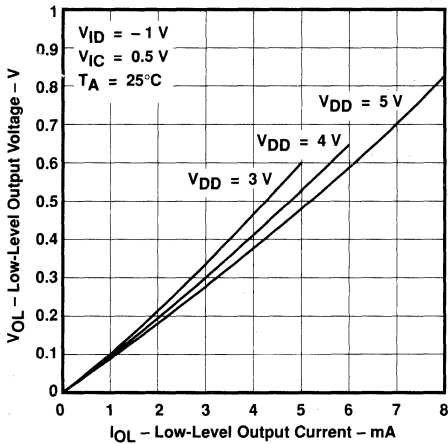


FIGURE 13

LOW-LEVEL OUTPUT VOLTAGE
 VS
 LOW-LEVEL OUTPUT CURRENT

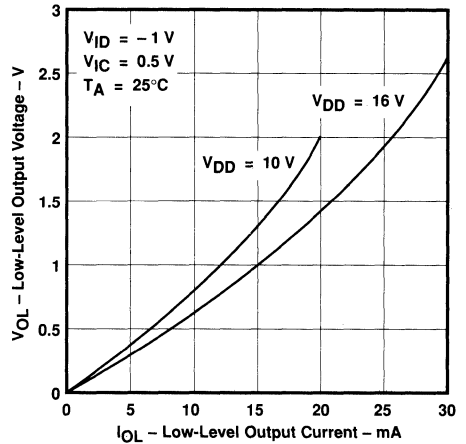


FIGURE 14

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 VS
 SUPPLY VOLTAGE

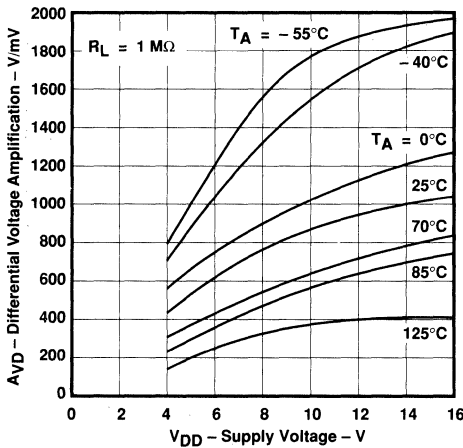


FIGURE 15

LARGE-SIGNAL
 DIFFERENTIAL VOLTAGE AMPLIFICATION
 VS
 FREE-AIR TEMPERATURE

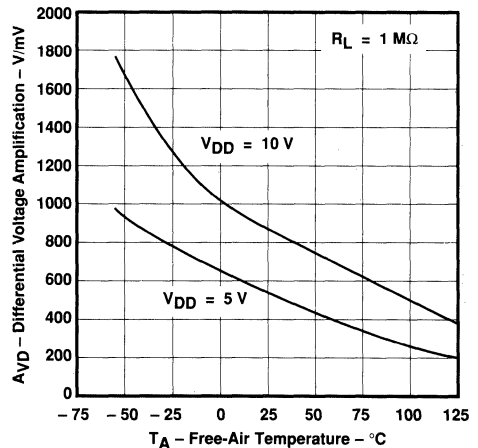


FIGURE 16

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

INPUT BIAS CURRENT AND INPUT OFFSET CURRENT
VS
FREE-AIR TEMPERATURE

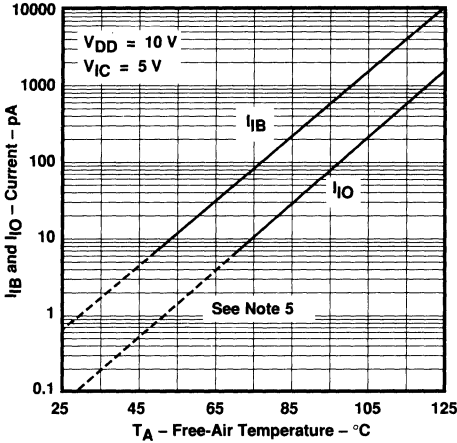


FIGURE 17

COMMON-MODE INPUT VOLTAGE POSITIVE LIMIT
VS
SUPPLY VOLTAGE

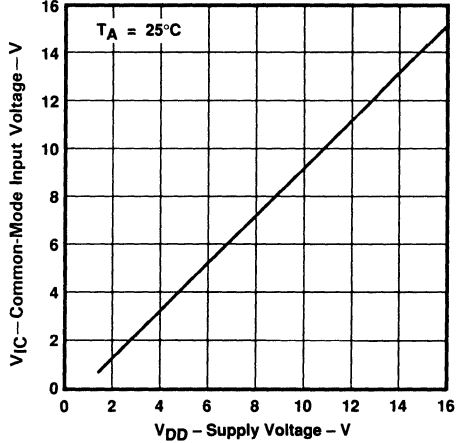


FIGURE 18

SUPPLY CURRENT
VS
SUPPLY VOLTAGE

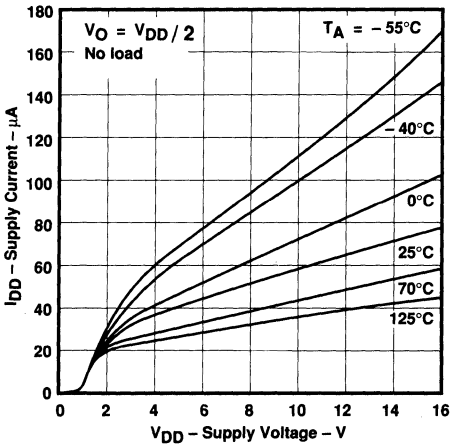


FIGURE 19

SUPPLY CURRENT
VS
FREE-AIR TEMPERATURE

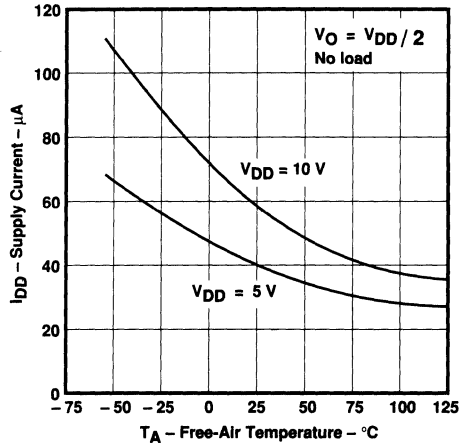


FIGURE 20

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.
NOTE 5: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

TYPICAL CHARACTERISTICS†

SLEW RATE
 VS
 SUPPLY VOLTAGE

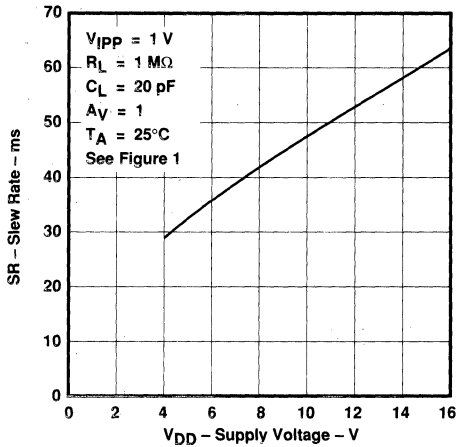


FIGURE 21

SLEW RATE
 VS
 FREE-AIR TEMPERATURE

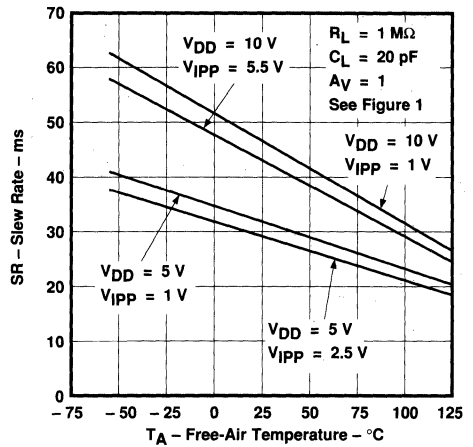


FIGURE 22

NORMALIZED SLEW RATE
 VS
 FREE-AIR TEMPERATURE

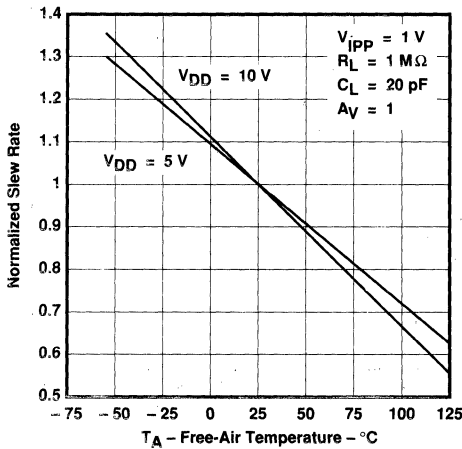


FIGURE 23

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 VS
 FREQUENCY

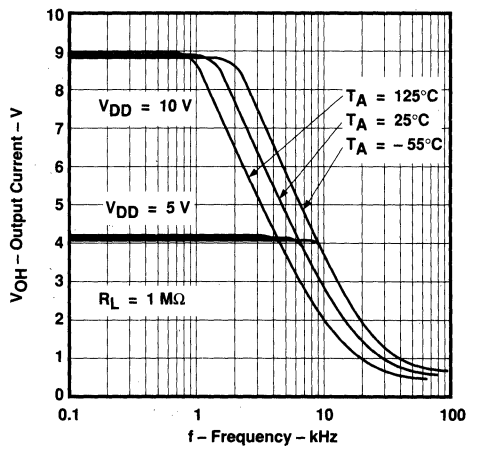


FIGURE 24

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

UNITY-GAIN BANDWIDTH
VS
FREE-AIR TEMPERATURE

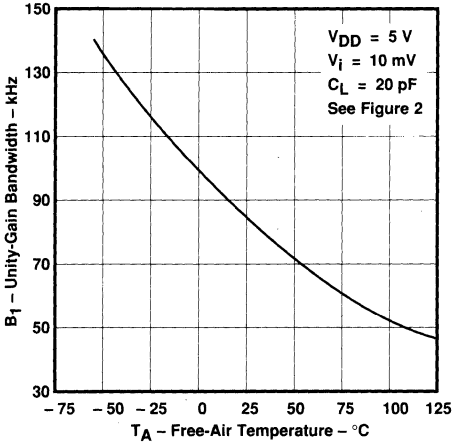


FIGURE 25

UNITY-GAIN BANDWIDTH
VS
SUPPLY VOLTAGE

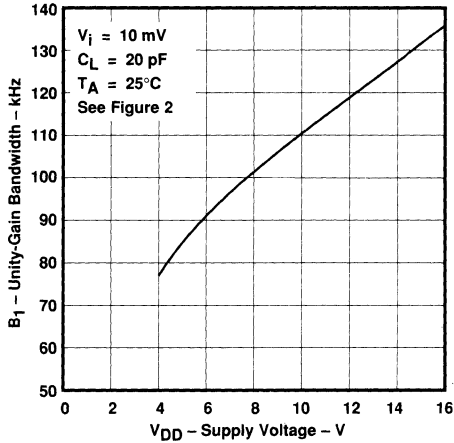


FIGURE 26

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
VS
FREQUENCY

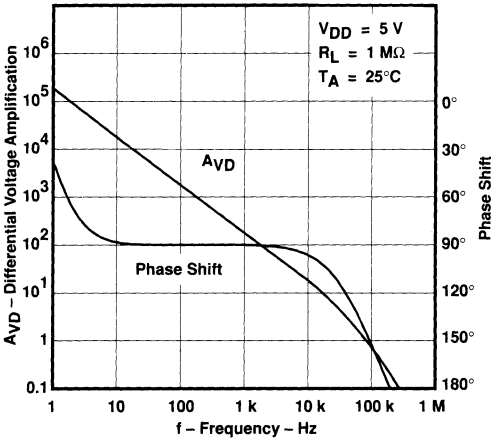


FIGURE 27

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
VS
FREQUENCY

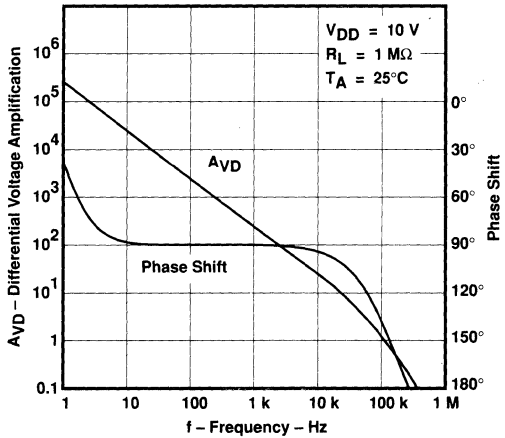


FIGURE 28

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

PHASE MARGIN
 VS
 SUPPLY VOLTAGE

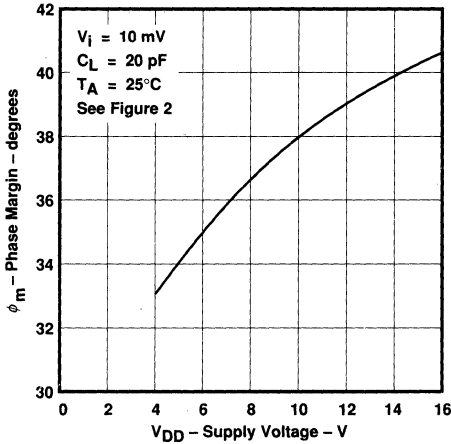


FIGURE 29

PHASE MARGIN
 VS
 FREE-AIR TEMPERATURE

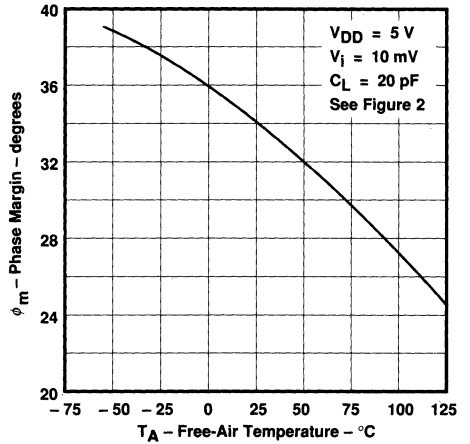


FIGURE 30

PHASE MARGIN
 VS
 CAPACITIVE LOAD

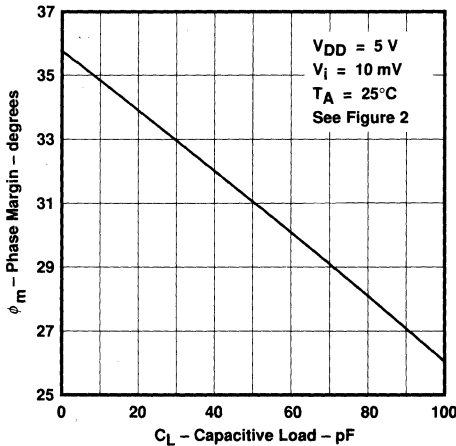


FIGURE 31

EQUIVALENT INPUT NOISE VOLTAGE
 VS
 FREQUENCY

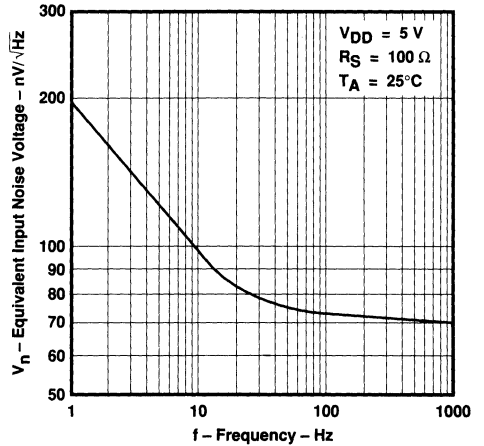


FIGURE 32

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC2201, TLC2201A, TLC2201B Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

D3173, NOVEMBER 1988

- **TLC2201B Is 100% Tested for Noise:**
25 nV/√Hz Max at f = 10 Hz
12 nV/√Hz Max at f = 1 kHz
- **Low Input Offset Voltage . . . 200 μV Max**
- **Excellent Offset Voltage Stability with Temperature . . . 0.5 μV/°C Typ**
- **Low Input Bias Current . . . 1 pA Typ at T_A = 25°C**
- **Fully Specified for Both Single-Supply and Split-Supply Operation**
- **Common-Mode Input Voltage Range Includes the Negative Rail**

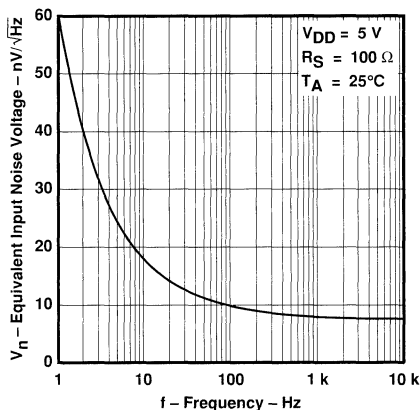
description

The TLC2201, TLC2201A, and TLC2201B are precision, low-noise operational amplifiers using Texas Instruments Advanced LinCMOS™ process. These devices combine the noise performance of the lowest-noise JFET amplifiers with the dc precision available previously only in bipolar amplifiers. The Advanced LinCMOS™ process uses silicon-gate technology to obtain input offset voltage stability with temperature and time that far exceeds that obtainable using metal-gate technology. In addition, this technology makes possible input impedance levels that meet or exceed levels offered by top-gate JFET and expensive dielectric-isolated devices.

The combination of excellent dc and noise performance with a common-mode input voltage range that includes the negative rail makes these devices an ideal choice for high-impedance, low-level signal conditioning applications in either single-supply or split-supply configurations.

The device inputs and outputs are designed to withstand -100-mA surge currents without sustaining latchup. In addition, internal ESD protection circuits prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.

TYPICAL EQUIVALENT
INPUT NOISE VOLTAGE
VS
FREQUENCY



2

Operational Amplifiers

AVAILABLE OPTIONS

T _A	V _{IO} max AT 25°C	V _n max f = 10 Hz AT 25°C	V _n max f = 1 kHz AT 25°C	PACKAGE				
				SMALL- OUTLINE (D)	PLASTIC DIP (P)	CERAMIC DIP (JG)	CHIP CARRIER (FK)	METAL CAN (L)
0°C	200 μV	25 nV/√Hz	12 nV/√Hz	TLC2201BCD	TLC2201BCP	TLC2201BCJG	—	TLC2201BCL
to	200 μV	35 nV/√Hz	15 nV/√Hz	TLC2201ACD	TLC2201ACP	TLC2201ACJG	—	TLC2201ACL
70°C	500 μV	—	—	TLC2201CD	TLC2201CP	TLC2201CJG	—	TLC2201CL
-40°C	200 μV	25 nV/√Hz	12 nV/√Hz	TLC2201BID	TLC2201BIP	TLC2201BIJG	—	TLC2201BIL
to	200 μV	35 nV/√Hz	15 nV/√Hz	TLC2201AID	TLC2201AIP	TLC2201AIJG	—	TLC2201AIL
85°C	500 μV	—	—	TLC2201ID	TLC2201IP	TLC2201IJG	—	TLC2201IL
-55°C	200 μV	25 nV/√Hz	12 nV/√Hz	TLC2201BMD	TLC2201BMP	TLC2201BMJG	TLC2201BMFK	TLC2201BML
to	200 μV	35 nV/√Hz	15 nV/√Hz	TLC2201AMD	TLC2201AMP	TLC2201AMJG	TLC2201AMFK	TLC2201AML
125°C	500 μV	—	—	TLC2201MD	TLC2201MP	TLC2201MJG	TLC2201MFK	TLC2201ML

D packages are available taped-and-reeled. Add "R" suffix to device type when ordering (e.g., TLC2201BCDR).

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PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS
INSTRUMENTS

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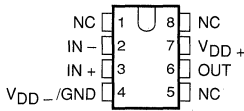
TLC2201, TLC2201A, TLC2201B

Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

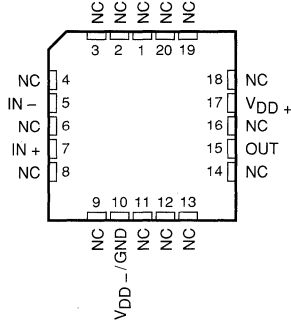
description (continued)

The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I-suffix devices are characterized for operation from -40°C to 85°C . The C-suffix devices are characterized for operation from 0°C to 70°C .

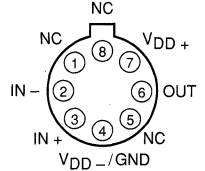
D, JG, or P PACKAGE
(TOP VIEW)



FK PACKAGE
(TOP VIEW)



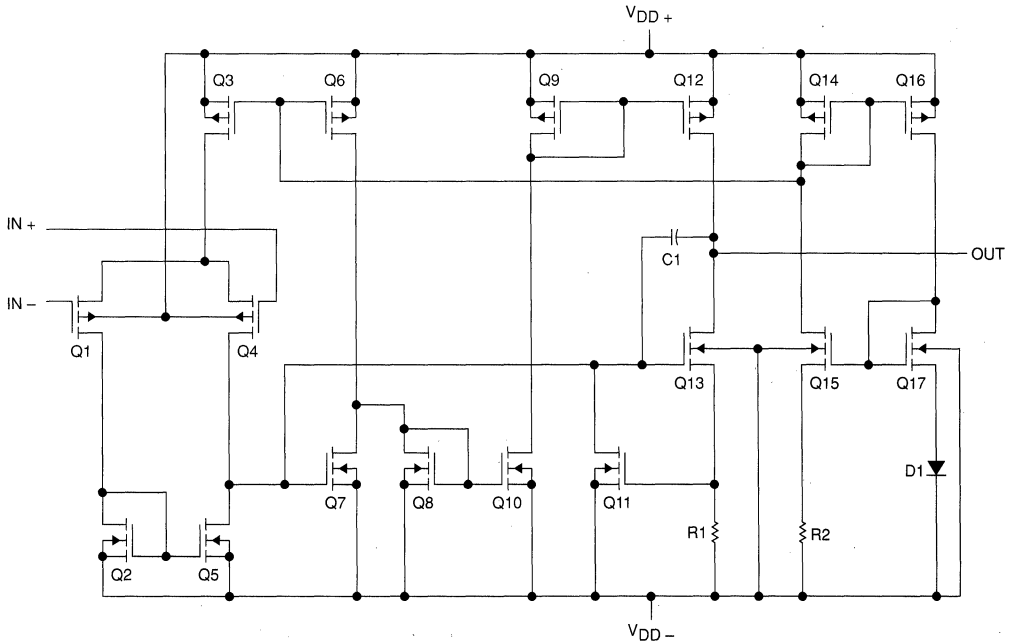
L PACKAGE
(TOP VIEW)



Pin 4 of the L package is in electrical contact with the case.

NC — No internal connection

equivalent schematic



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Operational Amplifiers

TLC2201, TLC2201A, TLC2201B
Advanced LinCMOS™ LOW-NOISE PRECISION
OPERATIONAL AMPLIFIERS

absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage, V_{DD+} (see Note 1)	8 V
Supply voltage, V_{DD-} (see Note 1)	-8 V
Differential input voltage (see Note 2)	± 16 V
Input voltage range, V_I (any input, see Note 1)	± 8 V
Input current, I_I (each input)	± 5 mA
Output current, I_O	± 50 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	Unlimited
Continuous total dissipation	see Dissipation Rating Table
Operating free-air temperature, T_A : M-suffix	-55°C to 125°C
I-suffix	-40°C to 85°C
C-suffix	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG or L package	300°C

2
Operational Amplifiers

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{DD+} and V_{DD-} .
 2. Differential voltages are at the noninverting input with respect to the inverting input.
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR		$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$		POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C		464 mW	377 mW	145 mW
FK	1375 mW	11.0 mW/°C		880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C		672 mW	546 mW	210 mW
L	650 mW	5.2 mW/°C		416 mW	338 mW	130 mW
P	1000 mW	8.0 mW/°C		640 mW	520 mW	200 mW

recommended operating conditions

	M-SUFFIX		I-SUFFIX		C-SUFFIX		UNIT
	MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, V_{DD}	± 2.3	± 8	± 2.3	± 8	± 2.3	± 8	V
Common-mode input voltage, V_{IC}	V_{DD-}	$V_{DD+} + 2.3$	V_{DD-}	$V_{DD+} + 2.3$	V_{DD-}	$V_{DD+} + 2.3$	V
Operating free-air temperature, T_A	-55	125	-40	85	0	70	°C

TLC2201M

Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} \pm = \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201M			UNIT
		MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	100	500	μV
		Full range		700	
α_{VIO} Temperature coefficient of input offset voltage		-55°C to 125°C	0.5		$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005	$\mu\text{V}/\text{mo}$
I_{IO} Input offset current		25°C	0.5		pA
		Full range		500	
I_{IB} Input bias current		25°C	1		pA
		Full range		500	
V_{ICR} Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7		V
V_{OM+} Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	V
Full range		4.7			
V_{OM-} Maximum negative peak output voltage swing		25°C	-4.7	-4.9	V
Full range	-4.7				
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	400	560	V/mV
		Full range	200		
	$V_O = \pm 4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	90	100	
		Full range	45		
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR}\ \text{min}, R_S = 50\ \Omega$	25°C	90	115	dB
		Full range	85		
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} \pm = \pm 2.3\text{ V to } \pm 8\text{ V}$	25°C	90	110	dB
		Full range	85		
I_{DD} Supply current	$V_O = 0, \text{ No load}$	25°C	1.1	1.5	mA
		Full range		1.5	

operating characteristics at specified free-air temperature, $V_{DD} \pm = \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201M			UNIT
		MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	2	2.7	$\text{V}/\mu\text{s}$
		Full range	1.3		
V_n Equivalent input noise voltage	$f = 10\ \text{Hz}$	25°C		18	$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C		8	
V_{NPP} Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C		0.5	μV
	$f = 0.1\ \text{to } 10\ \text{Hz}$	25°C		0.7	
I_n Equivalent input noise current		25°C		0.6	$\text{fA}/\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.9	MHz
ϕ_m Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		48°	

†Full range is -55°C to 125°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

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Operational Amplifiers

TLC2201AM, TLC2201BM

Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} \pm \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201AM			TLC2201BM			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	80	200	80	200	μV	
		Full range	400		400			
α_{VIO} Temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	-55°C to 125°C	0.5		0.5		$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.001	0.005	0.001	0.005	$\mu\text{V}/\text{mo}$	
I_{IO} Input offset current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.5		0.5		pA	
		Full range	500		500			
I_{IB} Input bias current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	1		1		pA	
		Full range	500		500			
V_{ICR} Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7	-5 to 2.7	-5 to 2.7	-5 to 2.7	V	
V_{OM+} Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	4.7	4.8	V	
V_{OM-} Maximum negative peak output voltage swing		Full range	4.7	4.7	4.7	4.7		
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	400	560	400	560	V/mV	
		Full range	200	200	200	200		
	$V_O = \pm 4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	90	100	90	100		
		Full range	45	45	45	45		
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR\ min}, R_S = 50\ \Omega$	25°C	90	115	90	115	dB	
		Full range	85	85	85	85		
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} \pm \pm 2.3\text{ V to } \pm 8\text{ V}$	25°C	90	110	90	110	dB	
		Full range	85	85	85	85		
I_{DD} Supply current	$V_O = 0, \text{ No load}$	25°C	1.1	1.5	1.1	1.5	mA	
		Full range	1.5		1.5			

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Operational Amplifiers

operating characteristics at specified free-air temperature, $V_{DD} \pm \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201AM			TLC2201BM			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	2	2.7	2	2.7	$\text{V}/\mu\text{s}$	
		Full range	1.3		1.3			
V_n Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C	18	35	18	25	$\text{nV}/\sqrt{\text{Hz}}$	
	$f = 1\ \text{kHz}$	25°C	8	15	8	12		
V_{NPP} Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C	0.5		0.5		μV	
	$f = 0.1\ \text{to } 10\ \text{Hz}$	25°C	0.7		0.7			
I_n Equivalent input noise current		25°C	0.6		0.6		$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.9		1.9		MHz	
ϕ_m Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	48°		48°			

†Full range is -55°C to 125°C.

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. This parameter is tested on a sample basis for the TLC2201A and on all devices for the TLC2201B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

TLC2201M
Advanced LinCMOS™ LOW-NOISE PRECISION
OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		TLC2201M			UNIT
				MIN	TYP	MAX	
V_{IO}	Input offset voltage		25°C	100	500	μV	
				Full range	700		
α_{VIO}	Temperature coefficient of input offset voltage		-55°C to 125°C	0.5		$\mu\text{V}/^\circ\text{C}$	
				Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, R_S = 50\ \Omega$		25°C
I_{IO}	Input offset current		25°C	0.5		pA	
				Full range	500		
I_{IB}	Input bias current		25°C	1		pA	
				Full range	500		
V_{ICR}	Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7		V	
V_{OH}	Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	V	
			Full range	4.7			
V_{OL}	Maximum low-level output voltage	$I_O = 0$	25°C	0	50	mV	
			Full range	50			
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V}, R_L = 500\ \text{k}\Omega,$ $V_O = 1\text{ V to }4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	150	315	V/mV	
			Full range	75			
			25°C	25	55		
			Full range	10			
CMRR	Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR\ \text{min}}, R_S = 50\ \Omega$	25°C	90	110	dB	
			Full range	85			
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} = 4.6\text{ V to }16\text{ V}$	25°C	90	110	dB	
			Full range	85			
I_{DD}	Supply current	$V_O = 2.5\text{ V},$ No load	25°C	1	1.5	mA	
			Full range	1.5			

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		TLC2201M			UNIT
				MIN	TYP	MAX	
SR	Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8	2.5	$\text{V}/\mu\text{s}$	
			Full range	1.1			
V_n	Equivalent input noise voltage	$f = 10\ \text{Hz}$ $f = 1\ \text{kHz}$	25°C	18		$\text{nV}/\sqrt{\text{Hz}}$	
			25°C	8			
V_{NPP}	Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\ \text{Hz}$ $f = 0.1\text{ to }10\ \text{Hz}$	25°C	0.5		μV	
			25°C	0.7			
I_n	Equivalent input noise current		25°C	0.6		$\text{fA}/\sqrt{\text{Hz}}$	
	Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8		MHz	
ϕ_m	Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	45°			

†Full range is -55°C to 125°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

TLC2201AM, TLC2201BM
Advanced LinCMOS™ LOW-NOISE PRECISION
OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201AM			TLC2201BM			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	80	200		80	200	μV
		Full range			400		400	
α_{VIO} Temperature coefficient of input offset voltage		-55°C to 125°C	0.5		0.5			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005		0.001	0.005	$\mu\text{V}/\text{mo}$
I_{IO} Input offset current		25°C	0.5		0.5			pA
		Full range	500		500			
I_{IB} Input bias current		25°C	1		1			pA
		Full range	500		500			
V_{ICR} Common-mode input voltage range		$R_S = 50\ \Omega$	Full range	0 to 2.7		0 to 2.7		V
V_{OH} Maximum high-level output voltage		$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	4.7	4.8	V
		Full range	4.7		4.7			
V_{OL} Maximum low-level output voltage	$I_O = 0$	25°C		0	50	0	50	mV
		Full range		50		50		
A_{VD} Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	150	315	150	315	V/mV	
		Full range	75		75			
	$V_O = 1\text{ V to }4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	25	55	25	55		
		Full range	10		10			
$CMRR$ Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR}\ \text{min}, R_S = 50\ \Omega$	25°C	90	110	90	110	dB	
		Full range	85		85			
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} = 4.6\text{ V to }16\text{ V}$	25°C	90	110	90	110	dB	
		Full range	85		85			
I_{DD} Supply current	$V_O = 2.5\text{ V}, \text{ No load}$	25°C		1	1.5	1	1.5	mA
		Full range		1.5		1.5		

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Operational Amplifiers

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201AM			TLC2201BM			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8	2.5	1.8	2.5	$\text{V}/\mu\text{s}$	
		Full range	1.1		1.1			
V_n Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C		18	35	18	25	$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C		8	15	8	12	
V_{NPP} Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\ \text{Hz}$	25°C	0.5		0.5		μV	
	$f = 0.1\text{ to }10\ \text{Hz}$	25°C	0.7		0.7			
I_n Equivalent input noise current		25°C	0.6		0.6		$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8		1.8		MHz	
ϕ_m Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	45°		45°			

†Full range is -55°C to 125°C.

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. This parameter is tested on a sample basis for the TLC2201A and on all devices for the TLC2201B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

TLC2201I
Advanced LinCMOS™ LOW-NOISE PRECISION
OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} \pm \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		TLC2201I			UNIT	
				MIN	TYP	MAX		
V_{IO}	Input offset voltage		25°C	100	500		μV	
				Full range	650			
α_{VIO}	Temperature coefficient of input offset voltage		-40°C to 85°C	0.5			$\mu\text{V}/^\circ\text{C}$	
	Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0,$	$R_S = 50\ \Omega$	25°C	0.001	0.005	$\mu\text{V}/\text{mo}$	
I_{IO}	Input offset current		25°C	0.5			pA	
				Full range	150			
I_{IB}	Input bias current		25°C	1			pA	
				Full range	150			
V_{ICR}	Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7			V	
V_{OM+}	Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8			V
V_{OM-}	Maximum negative peak output voltage swing		25°C	-4.7	-4.9			V
A_{VD}	Large-signal differential voltage amplification	$V_O = \pm 4\text{ V},$ $R_L = 500\ \text{k}\Omega,$	25°C	400	560			V/mV
			Full range	250				
		$V_O = \pm 4\text{ V},$ $R_L = 10\ \text{k}\Omega$	25°C	90	100			
			Full range	65				
CMRR	Common-mode rejection ratio	$V_O = 0,$ $V_{IC} = V_{ICR}\ \text{min},$ $R_S = 50\ \Omega$	25°C	90	115			dB
			Full range	85				
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} \pm = \pm 2.3\text{ V to } \pm 8\text{ V}$	25°C	90	110			dB
			Full range	85				
I_{DD}	Supply current	$V_O = 0,$ No load	25°C	1.1		1.5	mA	
			Full range			1.5		

operating characteristics at specified free-air temperature, $V_{DD} \pm \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		TLC2201I			UNIT	
				MIN	TYP	MAX		
SR	Slew rate at unity gain	$V_O = \pm 2.3\text{ V}, R_L = 10\ \text{k}\Omega,$ $C_L = 100\ \text{pF}$	25°C	2	2.7			$\text{V}/\mu\text{s}$
			Full range	1.4				
V_n	Equivalent input noise voltage	$f = 10\ \text{Hz}$	25°C	18				$\text{nV}/\sqrt{\text{Hz}}$
			25°C	8				
V_{NPP}	Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to}\ 1\ \text{Hz}$	25°C	0.5				μV
			25°C	0.7				
I_n	Equivalent input noise current		25°C	0.6				$\text{fA}/\sqrt{\text{Hz}}$
	Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega,$ $C_L = 100\ \text{pF}$	25°C	1.9				MHz
ϕ_m	Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	48°				

†Full range is -40°C to 85°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

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Operational Amplifiers

TLC2201AI, TLC2201BI

Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} \pm \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201AI			TLC2201BI			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage		25°C	80	200	80	200	μV	
		Full range	350					
α_{VIO} Temperature coefficient of input offset voltage		-40°C to 85°C	0.5			0.5	$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.001	0.005	0.001	0.005	$\mu\text{V}/\text{mo}$	
I_{IO} Input offset current		25°C	0.5			0.5		
		Full range	150			150		
I_{IB} Input bias current		25°C	1			1		
	Full range	150			150			
V_{ICR} Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7	-5 to 2.7	-5 to 2.7	V		
V_{OM+} Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	4.7	4.8	V	
Full range		4.7			4.7			
V_{OM-} Maximum negative peak output voltage swing		25°C	-4.7	-4.9	-4.7	-4.9	V	
Full range		-4.7			-4.7			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}, R_L = 500\ \text{k}\Omega$	25°C	400	560	400	560	V/mV	
		Full range	250			250		
	$V_O = \pm 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	90	100	90	100		
		Full range	65			65		
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR\ \text{min}}, R_S = 50\ \Omega$	25°C	90	115	90	115	dB	
		Full range	85			85		
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} \pm \pm 2.3\ \text{V to } \pm 8\ \text{V}$	25°C	90	110	90	110	dB	
		Full range	85			85		
I_{DD} Supply current	$V_O = 0, \text{ No load}$	25°C	1.1	1.5	1.1	1.5	mA	
		Full range	1.5			1.5		

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Operational Amplifiers

operating characteristics at specified free-air temperature, $V_{DD} \pm \pm 5\ \text{V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201AI			TLC2201BI			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3\ \text{V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	2	2.7	2	2.7	$\text{V}/\mu\text{s}$	
		Full range	1.4			1.4		
V_n Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C	18	35	18	25	$\text{nV}/\sqrt{\text{Hz}}$	
		25°C	8	15	8	12		
V_{NPP} Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C	0.5			0.5	μV	
	$f = 0.1\ \text{to } 10\ \text{Hz}$	25°C	0.7			0.7		
I_n Equivalent input noise current		25°C	0.6			0.6	$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.9			1.9	MHz	
ϕ_m Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	48°			48°		

†Full range is -40°C to 85°C.

- NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.
5. This parameter is tested on a sample basis for the TLC2201A and on all devices for the TLC2201B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

TLC22011
Advanced LinCMOS™ LOW-NOISE PRECISION
OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		TLC22011			UNIT	
				MIN	TYP	MAX		
V_{IO}	Input offset voltage			25°C	100	500	μV	
				Full range		650		
α_{VIO}	Temperature coefficient of input offset voltage			-40°C to 85°C	0.5		$\mu\text{V}/^\circ\text{C}$	
	Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0$,	$R_S = 50\ \Omega$	25°C	0.001	0.005	$\mu\text{V}/\text{mo}$	
I_{IO}	Input offset current			25°C	0.5		pA	
				Full range		150		
I_{IB}	Input bias current			25°C	1		pA	
				Full range		150		
V_{ICR}	Common-mode input voltage range	$R_S = 50\ \Omega$		Full range	0 to 2.7		V	
V_{OH}	Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$		25°C	4.7	4.8	V	
				Full range	4.7			
V_{OL}	Maximum low-level output voltage	$I_O = 0$		25°C		0 50	mV	
				Full range		50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V}$, $R_L = 500\ \text{k}\Omega$,		25°C	150	315	V/mV	
				Full range	100			
				$V_O = 1\text{ V to }4\text{ V}$, $R_L = 10\ \text{k}\Omega$	25°C	25		55
				Full range	15			
CMRR	Common-mode rejection ratio	$V_O = 0$, $V_{IC} = V_{ICR}\ \text{min}$, $R_S = 50\ \Omega$		25°C	90	110	dB	
				Full range	85			
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} = 4.6\text{ V to }16\text{ V}$		25°C	90	110	dB	
				Full range	85			
I_{DD}	Supply current	$V_O = 2.5\text{ V}$, No load		25°C	1	1.5	mA	
				Full range		1.5		

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		TLC22011			UNIT
				MIN	TYP	MAX	
SR	Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}$, $R_L = 10\ \text{k}\Omega$, $C_L = 100\ \text{pF}$		25°C	1.8	2.5	V/ μs
				Full range	1.2		
V_n	Equivalent input noise voltage	$f = 10\ \text{Hz}$		25°C		18	nV/ $\sqrt{\text{Hz}}$
				25°C		8	
V_{NPP}	Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\ \text{Hz}$		25°C		0.5	μV
				25°C		0.7	
I_n	Equivalent input noise current			25°C		0.6	fA/ $\sqrt{\text{Hz}}$
	Gain-bandwidth product	$f = 10\ \text{kHz}$, $R_L = 10\ \text{k}\Omega$, $C_L = 100\ \text{pF}$		25°C		1.8	MHz
ϕ_m	Phase margin at unity gain	$R_L = 10\ \text{k}\Omega$, $C_L = 100\ \text{pF}$		25°C		45°	

†Full range is -40°C to 85°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

TLC2201AI, TLC2201BI

Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201AI			TLC2201BI			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	80	200	80	200	μV	
		Full range	350					
α_{VIO} Temperature coefficient of input offset voltage		-40°C to 85°C	0.5			0.5	$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005	0.001	0.005	$\mu\text{V}/\text{mo}$	
I_{IO} Input offset current		25°C	0.5			0.5	pA	
		Full range	150			150		
I_{IB} Input bias current		25°C	1			1	pA	
		Full range	150			150		
V_{ICR} Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7		0 to 2.7		V	
V_{OH} Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	4.7	4.8	V	
		Full range	4.7					
V_{OL} Maximum low-level output voltage	$I_O = 0$	25°C		0	50	0	50	mV
		Full range		50			50	
A_{VD} Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	150	315	150	315	V/mV	
		Full range	100					
	$V_O = 1\text{ V to }4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	25	55	25	55		
		Full range	15					
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR\ \text{min}}, R_S = 50\ \Omega$	25°C	90	110	90	110	dB	
		Full range	85					
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} = 4.6\text{ V to }16\text{ V}$	25°C	90	110	90	110	dB	
		Full range	85					
I_{DD} Supply current	$V_O = 2.5\text{ V}, \text{ No load}$	25°C		1	1.5	1	1.5	mA
		Full range		1.5			1.5	

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Operational Amplifiers

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201AI			TLC2201BI			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8	2.5	1.8	2.5	$\text{V}/\mu\text{s}$	
		Full range	1.2					
V_n Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C	18	35	18	25	$\text{nV}/\sqrt{\text{Hz}}$	
		Full range	8					
V_{NPP} Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\ \text{Hz}$	25°C	0.5			0.5	μV	
		Full range	0.7					
I_n Equivalent input noise current	$f = 0.1\text{ to }10\ \text{Hz}$	25°C	0.6			0.6	$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8			1.8	MHz	
ϕ_m Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	45°			45°		

†Full range is -40°C to 85°C.

- NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.
 5. This parameter is tested on a sample basis for the TLC2201A and on all devices for the TLC2201B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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electrical characteristics at specified free-air temperature, $V_{DD\pm} = \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201C			UNIT
		MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	100	500	μV
		Full range		600	
α_{VIO} Temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	0°C to 70°C	0.5		$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005	$\mu\text{V}/\text{mo}$
I_{IO} Input offset current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.5		pA
		Full range		100	
I_{IB} Input bias current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	1		pA
		Full range		100	
V_{ICR} Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7		V
V_{OM+} Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	V
Full range		4.7			
V_{OM-} Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	-4.7	-4.9	V
Full range		-4.7			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 4\text{ V}, R_L = 500\ \text{k}\Omega,$	25°C	400	560	V/mV
		Full range	300		
	$V_O = \pm 4\text{ V}, R_L = 10\ \text{k}\Omega,$	25°C	90	100	
		Full range	70		
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR\ \text{min}}, R_S = 50\ \Omega$	25°C	90	115	dB
		Full range	85		
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD\pm} / \Delta V_{IO}$)	$V_{DD\pm} = \pm 2.3\text{ V to } \pm 8\text{ V}$	25°C	90	110	dB
		Full range	85		
I_{DD} Supply current	$V_O = 0, \text{ No load}$	25°C	1.1	1.5	mA
		Full range	1.5		

operating characteristics at specified free-air temperature, $V_{DD\pm} = \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201C			UNIT
		MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	2	2.7	$\text{V}/\mu\text{s}$
		Full range	1.5		
V_n Equivalent input noise voltage	$f = 10\ \text{Hz}$	25°C	18		$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C	8		
V_{NPP} Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C	0.5		μV
	$f = 0.1\ \text{to } 10\ \text{Hz}$	25°C	0.7		
I_n Equivalent input noise current		25°C	0.6		$\text{fA}/\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.9		MHz
ϕ_m Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	48°		

†Full range is 0°C to 70°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

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electrical characteristics at specified free-air temperature, $V_{DD} \pm \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201AC			TLC2201BC			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX		
V_{IO} Input offset voltage		25°C	80	200	80	200		μV	
		Full range	300			300			
α_{VIO} Temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	0°C to 70°C	0.5			0.5			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005	0.001	0.005		$\mu\text{V}/\text{mo}$	
I_{IO} Input offset current		25°C	0.5			0.5			pA
		Full range	100			100			
I_{IB} Input bias current	25°C	1			1			pA	
	Full range	100			100				
V_{ICR} Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7		-5 to 2.7		V		
V_{OM+} Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	4.7	4.8		V	
V_{OM-} Maximum negative peak output voltage swing		Full range	4.7			4.7			
		25°C	-4.7	-4.9	-4.7	-4.9			V
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	400	560	400	560		V/mV	
		Full range	300			300			
	$V_O = \pm 4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	90	100	90	100			
		Full range	70			70			
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR\ \text{min}}, R_S = 50\ \Omega$	25°C	90	115	90	115		dB	
		Full range	85			85			
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} \pm \pm 2.3\text{ V to } \pm 8\text{ V}$	25°C	90	110	90	110		dB	
		Full range	85			85			
I_{DD} Supply current	$V_O = 0, \text{ No load}$	25°C		1.1 1.5		1.1 1.5		mA	
		Full range	1.5			1.5			

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Operational Amplifiers

operating characteristics at specified free-air temperature, $V_{DD} \pm \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201AC			TLC2201BC			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX		
SR Slew rate at unity gain	$V_O = \pm 2.3\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	2	2.7	2	2.7		$\text{V}/\mu\text{s}$	
		Full range	1.5			1.5			
V_n Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C	18	35	18	25		$\text{nV}/\sqrt{\text{Hz}}$	
		$f = 1\ \text{kHz}$	8	15	8	12			
V_{NPP} Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C	0.5			0.5			μV
	$f = 0.1\ \text{to } 10\ \text{Hz}$	25°C	0.7			0.7			
I_n Equivalent input noise current	25°C	0.6			0.6			$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.9			1.9			MHz
ϕ_m Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	48°			48°			

†Full range is 0°C to 70°C.

- NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.
5. This parameter is tested on a sample basis for the TLC2201A and on all devices for the TLC2201B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

TLC2201C

Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS [†]		TLC2201C			UNIT
				MIN	TYP	MAX	
V_{IO}	Input offset voltage			25°C	100	500	μV
				Full range		600	
α_{VIO}	Temperature coefficient of input offset voltage			0°C to 70°C	0.5		$\mu\text{V}/^\circ\text{C}$
	Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0,$	$R_S = 50\ \Omega$	25°C	0.001	0.005	$\mu\text{V}/\text{mo}$
I_{IO}	Input offset current			25°C	0.5		pA
				Full range		100	
I_{IB}	Input bias current			25°C	1		pA
				Full range		100	
V_{ICR}	Common-mode input voltage range	$R_S = 50\ \Omega$		Full range	0 to 2.7		V
V_{OH}	Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$		25°C	4.7	4.8	V
				Full range	4.7		
V_{OL}	Maximum low-level output voltage	$I_O = 0$		25°C	0	50	mV
				Full range		50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V},$ $R_L = 500\ \text{k}\Omega,$		25°C	150	315	V/mV
				Full range	100		
				25°C	25	55	
				Full range	15		
CMRR	Common-mode rejection ratio	$V_O = 0,$ $V_{IC} = V_{ICR}\ \text{min},$ $R_S = 50\ \Omega$		25°C	90	110	dB
				Full range	85		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} = 4.6\text{ V to }16\text{ V}$		25°C	90	110	dB
				Full range	85		
I_{DD}	Supply current	$V_O = 2.5\text{ V},$ No load		25°C	1	1.5	mA
				Full range		1.5	

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS [†]		TLC2201C			UNIT
				MIN	TYP	MAX	
SR	Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V},$ $R_L = 10\ \text{k}\Omega,$ $C_L = 100\ \text{pF}$		25°C	1.8	2.5	$\text{V}/\mu\text{s}$
				Full range	1.3		
V_n	Equivalent input noise voltage	$f = 10\ \text{Hz}$		25°C		18	$\text{nV}/\sqrt{\text{Hz}}$
				25°C		8	
V_{NPP}	Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\ \text{Hz}$		25°C		0.5	μV
				25°C		0.7	
I_n	Equivalent input noise current			25°C		0.6	$\text{fA}/\sqrt{\text{Hz}}$
	Gain-bandwidth product	$f = 10\ \text{kHz},$ $R_L = 10\ \text{k}\Omega,$ $C_L = 100\ \text{pF}$		25°C		1.8	MHz
ϕ_m	Phase margin at unity gain	$R_L = 10\ \text{k}\Omega,$ $C_L = 100\ \text{pF}$		25°C		45°	

[†]Full range is 0°C to 70°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

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Operational Amplifiers

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electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201AC			TLC2201BC			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX		
V_{IO} Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	80	200	80	200		μV	
		Full range	300			300			
α_{VIO} Temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	0°C to 70°C	0.5			0.5			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005	0.001	0.005		$\mu\text{V}/\text{mo}$	
I_{IO} Input offset current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.5			0.5			pA
		Full range	100			100			
I_{IB} Input bias current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	1			1			pA
		Full range	100			100			
V_{ICR} Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7		0 to 2.7			V	
V_{OH} Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	4.7	4.8		V	
		Full range	4.7			4.7			
V_{OL} Maximum low-level output voltage	$I_O = 0$	25°C		0	50		0	50	mV
		Full range	50			50			
A_{VD} Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	150	315	150	315		V/mV	
		Full range	100			100			
	$V_O = 1\text{ V to }4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	25	55	25	55			
		Full range	15			15			
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR\ \text{min}}, R_S = 50\ \Omega$	25°C	90	110	90	110		dB	
		Full range	85			85			
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} = 4.6\text{ V to }16\text{ V}$	25°C	90	110	90	110		dB	
		Full range	85			85			
I_{DD} Supply current	$V_O = 2.5\text{ V}, \text{ No load}$	25°C		1	1.5		1	1.5	mA
		Full range	1.5			1.5			

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Operational Amplifiers

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2201AC			TLC2201BC			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX		
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8	2.5	1.8	2.5		V/ μs	
		Full range	1.3			1.3			
V_n Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C		18	35		18	25	nV/ $\sqrt{\text{Hz}}$
		25°C		8	15		8	12	
V_{NPP} Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\ \text{Hz}$	25°C	0.5			0.5			μV
		25°C	0.7			0.7			
I_n Equivalent input noise current	$f = 10\ \text{Hz}$	25°C	0.6			0.6			fA/ $\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8			1.8			MHz
ϕ_m Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	45°			45°			

†Full range is 0°C to 70°C.

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. This parameter is tested on a sample basis for the TLC2201A and on all devices for the TLC2201B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

PARAMETER MEASUREMENT INFORMATION

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Operational Amplifiers

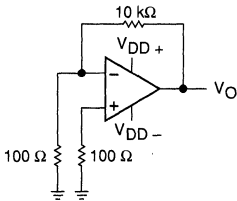
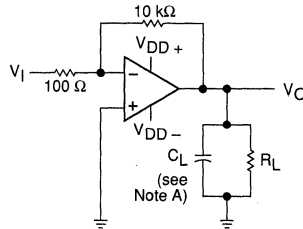
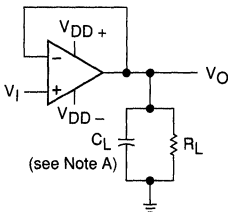


FIGURE 1. NOISE VOLTAGE TEST CIRCUIT



NOTE A: C_L includes fixture capacitance.

FIGURE 2. PHASE MARGIN TEST CIRCUIT



NOTE A: C_L includes fixture capacitance.

FIGURE 3. SLEW RATE TEST CIRCUIT

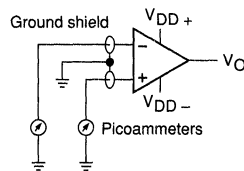


FIGURE 4. INPUT BIAS AND OFFSET CURRENT TEST CIRCUIT

typical values

Typical values as presented in this data sheet represent the median (50% point) of device parametric performance.

input bias and offset current

At the picoamp bias current level typical of the TLC2201, TLC2201A, and TLC2201B, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied but with no device in the socket. The device is then inserted in the socket and a second test measuring both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

noise

Texas Instruments offers automated production noise testing to meet individual applications requirements. Noise voltage at $f = 10$ Hz and $f = 1$ kHz is 100% tested on every TLC2201B device, while lot sample testing is performed on the TLC2201A. For other noise test requirements, please contact the factory.

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TYPICAL CHARACTERISTICS

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I_{IB}	Input bias current	vs Common-mode voltage	6
		vs Temperature	7
CMRR	Common-mode rejection ratio	vs Frequency	8
V_{OM}	Maximum peak output voltage	vs Output current	9
		vs Temperature	10
V_{OPP}	Maximum peak-to-peak output voltage	vs Frequency	11
		vs Frequency	12
V_{OH}	High-level output voltage	vs Current	13
		vs Temperature	14
		vs Output current	15
V_{OL}	Low-level output voltage	vs Temperature	16
		vs Frequency	17
A_{VD}	Differential voltage amplification	vs Temperature	18
		vs Supply voltage	19
I_{OS}	Short-circuit output current	vs Temperature	20
		vs Supply voltage	21
I_{DD}	Supply current	vs Temperature	22
		vs Supply voltage	23
SR	Slew rate	vs Temperature	24
		Small-signal	25, 26
V_{NPP}	Peak-to-peak equivalent input noise voltage	Large-signal	27, 28
		0.1 to 1 Hz	29
		0.1 to 10 Hz	30
	Gain-bandwidth product	vs Supply voltage	31
		vs Temperature	32
ϕ_m	Phase margin	vs Supply voltage	33
		vs Temperature	34
	Phase shift	vs Frequency	17

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TYPICAL CHARACTERISTICS†

DISTRIBUTION OF TLC2201
 INPUT OFFSET VOLTAGE

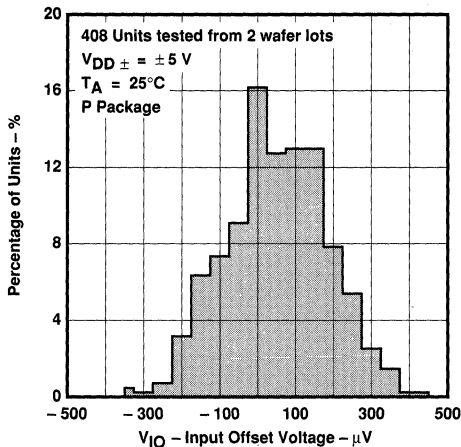


FIGURE 5

INPUT BIAS CURRENT
 vs
 COMMON-MODE INPUT VOLTAGE

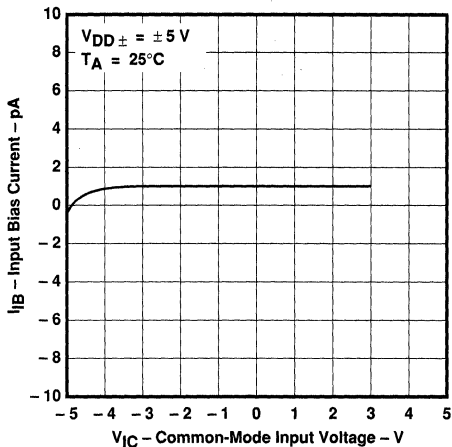


FIGURE 6

INPUT BIAS CURRENT
 vs
 FREE-AIR TEMPERATURE

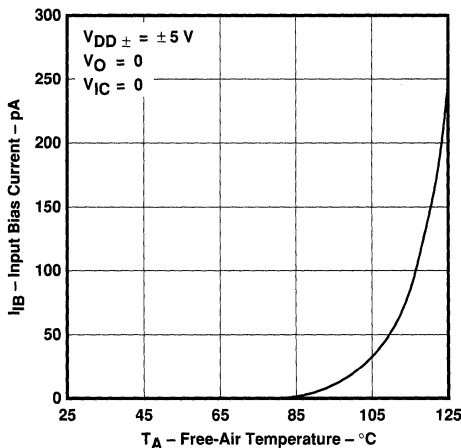


FIGURE 7

COMMON-MODE REJECTION RATIO
 vs
 FREQUENCY

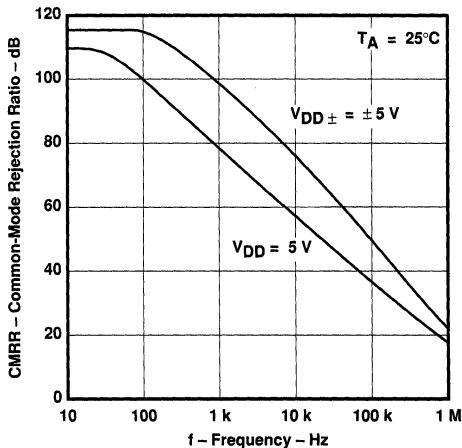


FIGURE 8

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 OUTPUT CURRENT

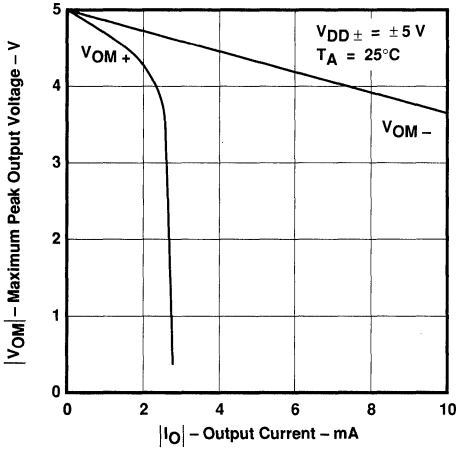


FIGURE 9

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

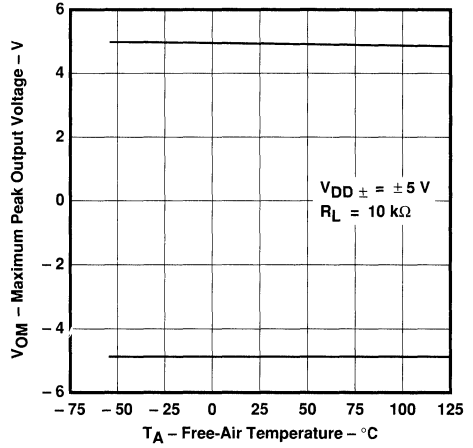


FIGURE 10

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 VS
 FREQUENCY

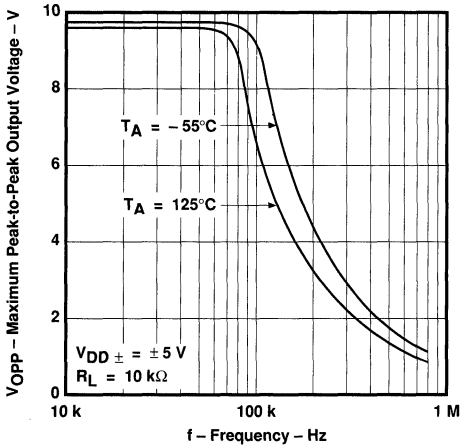


FIGURE 11

HIGH-LEVEL OUTPUT VOLTAGE
 VS
 FREQUENCY

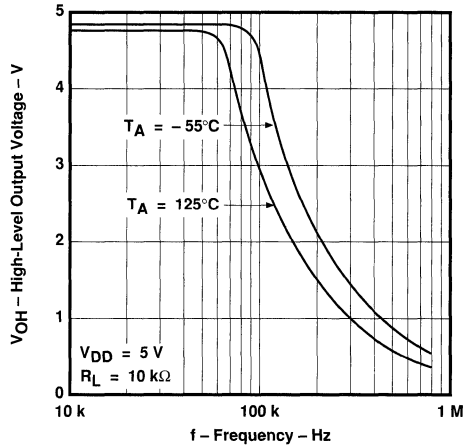


FIGURE 12

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC2201, TLC2201A, TLC2201B
Advanced LinCMOS™ LOW-NOISE PRECISION
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT VOLTAGE
 VS
 HIGH-LEVEL OUTPUT CURRENT

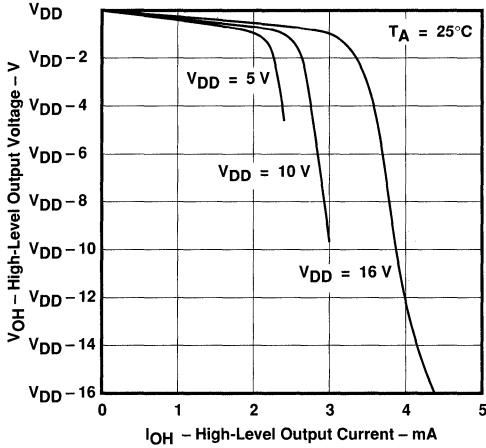


FIGURE 13

HIGH-LEVEL OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

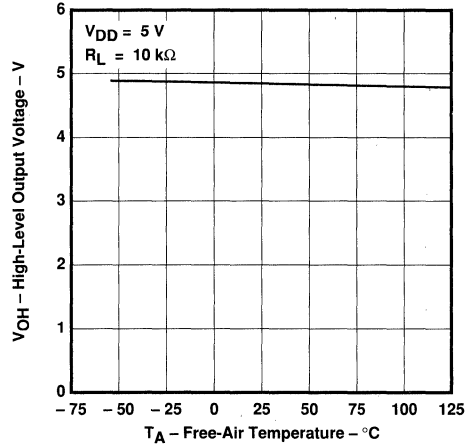


FIGURE 14

LOW-LEVEL OUTPUT VOLTAGE
 VS
 LOW-LEVEL OUTPUT CURRENT

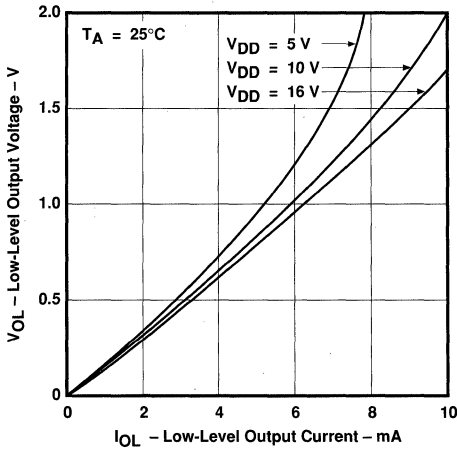


FIGURE 15

LOW-LEVEL OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

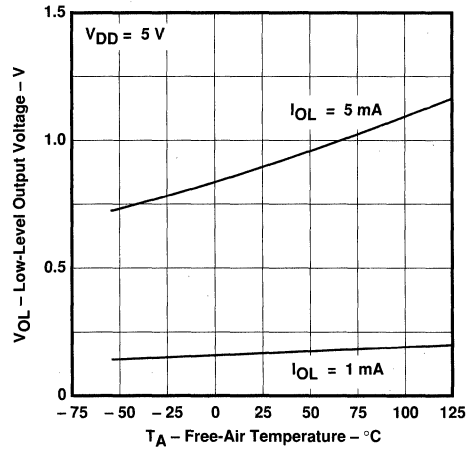


FIGURE 16

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC2201, TLC2201A, TLC2201B
 Advanced LinCMOS™ LOW-NOISE PRECISION
 OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE SHIFT
 VS
 FREQUENCY

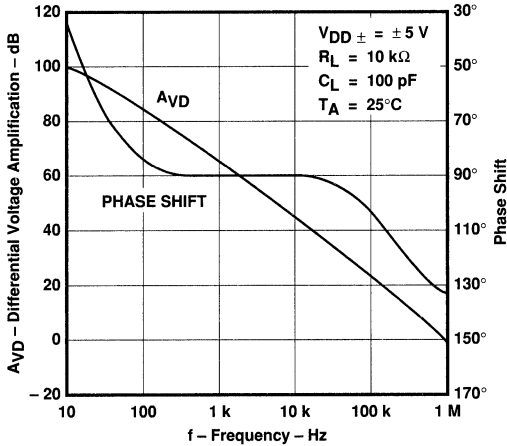


FIGURE 17

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION
 VS
 FREE-AIR TEMPERATURE

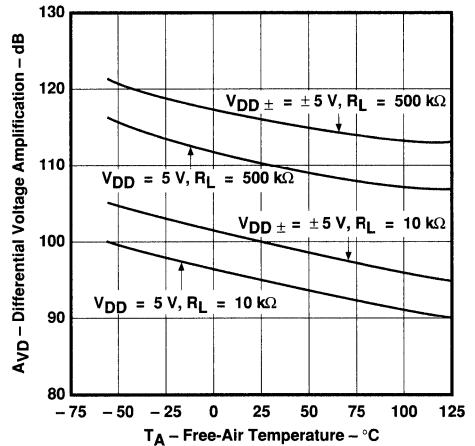


FIGURE 18

SHORT-CIRCUIT OUTPUT CURRENT
 VS
 SUPPLY VOLTAGE

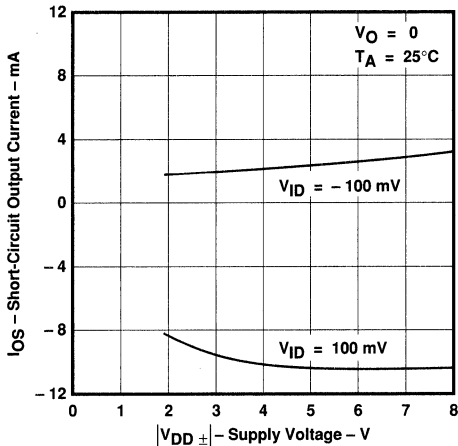


FIGURE 19

SHORT-CIRCUIT OUTPUT CURRENT
 VS
 FREE-AIR TEMPERATURE

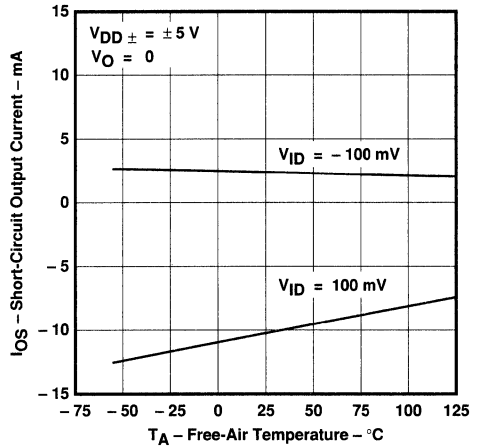


FIGURE 20

2
 Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC2201, TLC2201A, TLC2201B
Advanced LinCMOS™ LOW-NOISE PRECISION
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

2
Operational Amplifiers

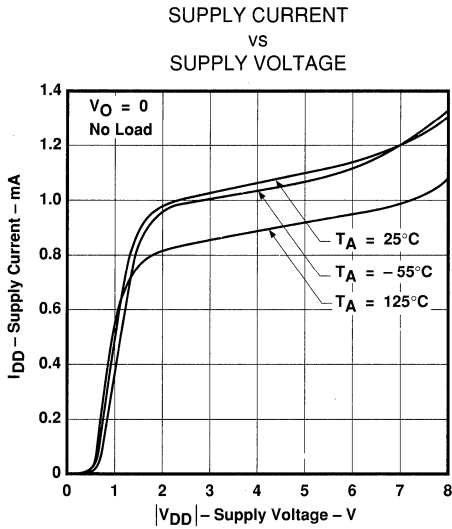


FIGURE 21

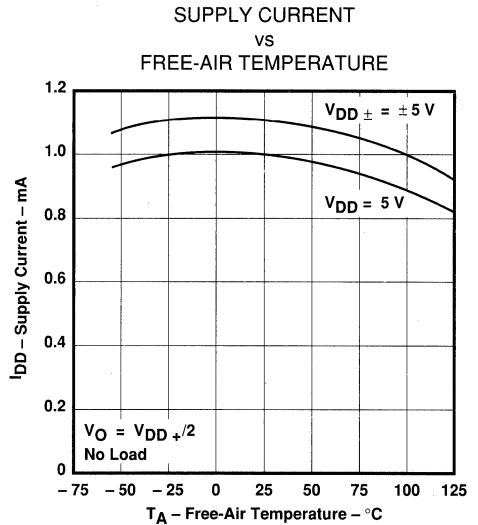


FIGURE 22

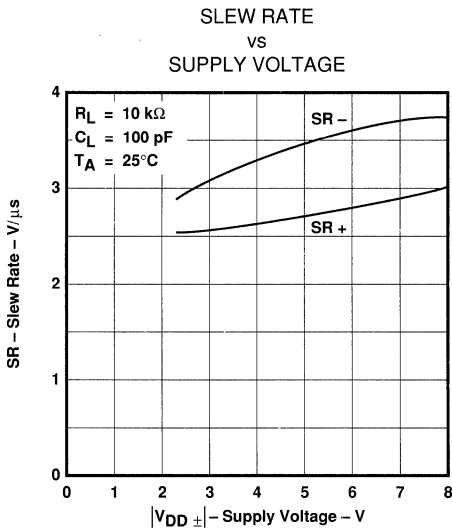


FIGURE 23

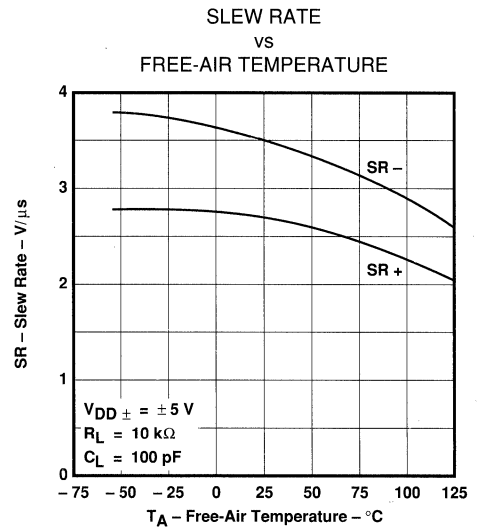


FIGURE 24

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

VOLTAGE-FOLLOWER
 SMALL-SIGNAL
 PULSE RESPONSE

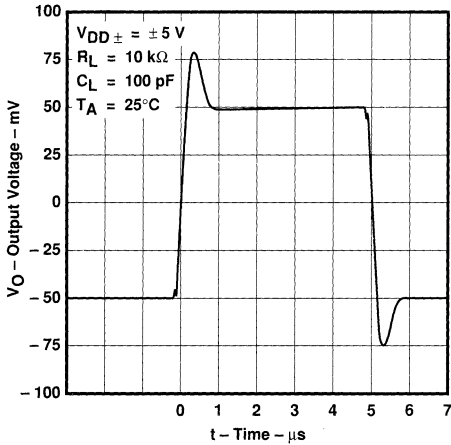


FIGURE 25

VOLTAGE-FOLLOWER
 SMALL-SIGNAL
 PULSE RESPONSE

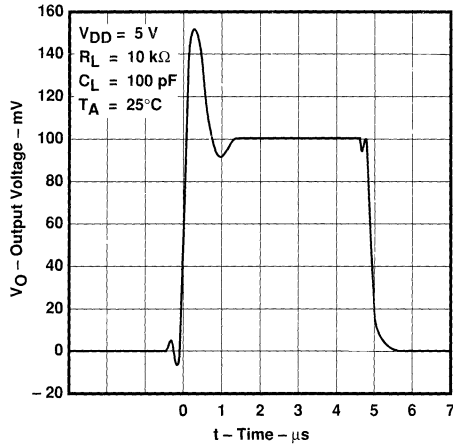


FIGURE 26

VOLTAGE-FOLLOWER
 LARGE-SIGNAL
 PULSE RESPONSE

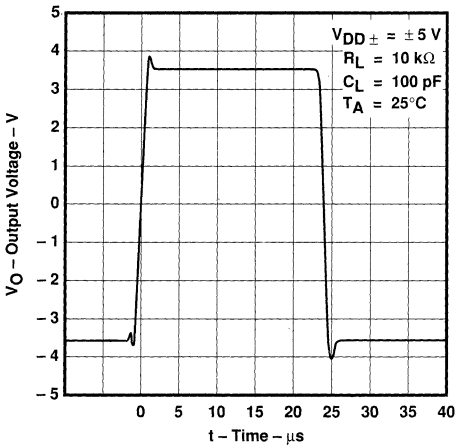


FIGURE 27

VOLTAGE-FOLLOWER
 LARGE-SIGNAL
 PULSE RESPONSE

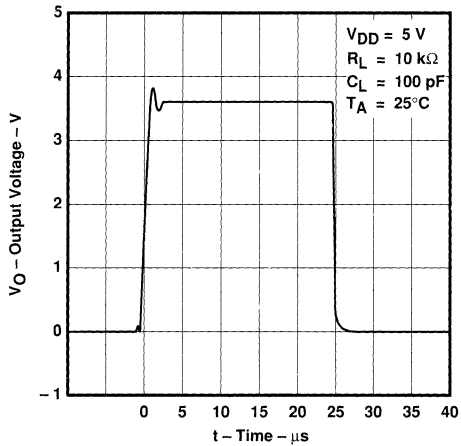


FIGURE 28

TLC2201, TLC2201A, TLC2201B
Advanced LinCMOS™ LOW-NOISE PRECISION
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

2
Operational Amplifiers

**PEAK-TO-PEAK EQUIVALENT
 INPUT NOISE VOLTAGE
 0.1 TO 1 Hz**

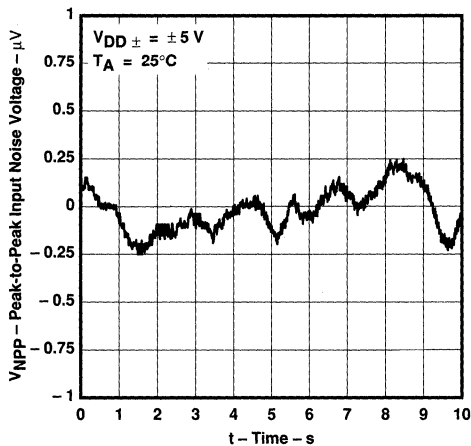


FIGURE 29

**PEAK-TO-PEAK EQUIVALENT
 INPUT NOISE VOLTAGE
 0.1 TO 10 Hz**

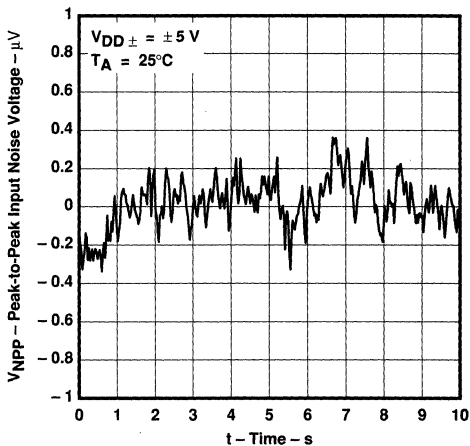


FIGURE 30

**GAIN-BANDWIDTH PRODUCT
 VS
 SUPPLY VOLTAGE**

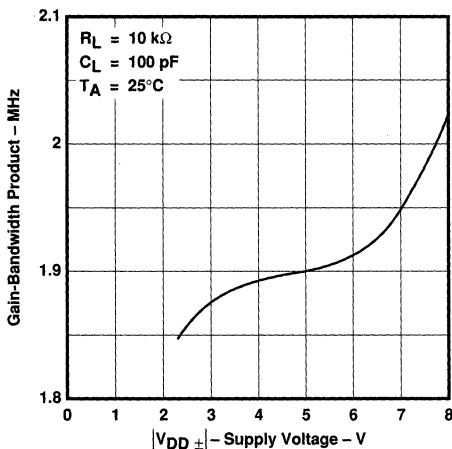


FIGURE 31

**GAIN-BANDWIDTH PRODUCT
 VS
 FREE-AIR TEMPERATURE**

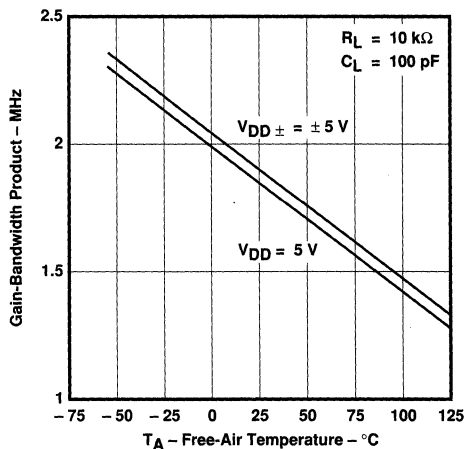


FIGURE 32

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

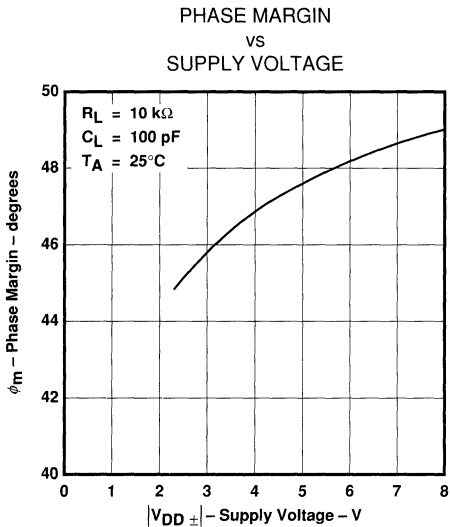


FIGURE 33

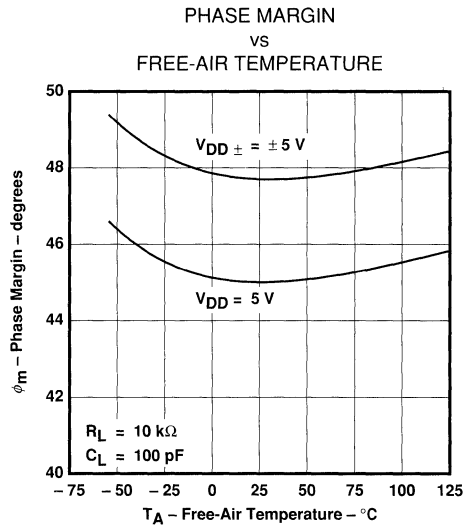


FIGURE 34

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL APPLICATION DATA

latchup avoidance

Because CMOS devices are susceptible to latchup due to their inherent parasitic thyristors, the TLC2201, TLC2201A, and TLC2201B inputs and outputs are designed to withstand -100-mA surge currents without sustaining latchup; however, techniques reducing the chance of latchup should be used whenever possible. Internal protection diodes should not be forward biased in normal operation. Applied input and output voltages should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μF typical) located across the supply rails as close to the device as possible.

electrostatic discharge protection

These devices use internal ESD protection circuits that prevent functional failures at voltages at or below 2000 V. Care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.

2

Operational Amplifiers

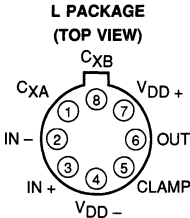
TLC2652, TLC2652A Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

D3157, SEPTEMBER 1988

- **Extremely Low Offset Voltage ... 1 μ V Max**
- **Extremely Low Change in Offset Voltage with Temperature ... 0.003 μ V/ $^{\circ}$ C Typ**
- **Low Input Offset Current ... 500 pA Max at $T_A = -55^{\circ}$ C to 125 $^{\circ}$ C**
- **AVD ... 135 dB Min**
- **CMRR and kSVR ... 120 dB Min**
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Includes the Negative Rail**
- **No Noise Degradation with External Capacitors Connected to V_{DD-}**

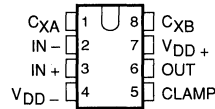
description

The TLC2652 and TLC2652A are high-precision chopper-stabilized operational amplifiers using Texas Instruments **Advanced LinCMOS™** process. This process in conjunction with unique chopper-stabilization circuitry produces operational amplifiers whose performance matches or exceeds that of similar devices available today.

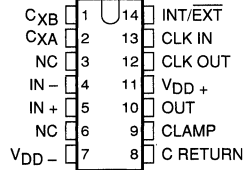


Pin 4 of the L package is in electrical contact with the case.

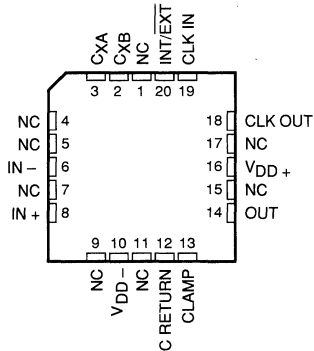
**D008, JG, or P PACKAGE
(TOP VIEW)**



**D014, J, or N PACKAGE
(TOP VIEW)**



**FK PACKAGE
(TOP VIEW)**



NC – No internal connection

AVAILABLE OPTIONS

T _A	V _{IO} max at 25 $^{\circ}$ C	PACKAGE								
		8-PIN				14-PIN				20-PIN
		SMALL-OUTLINE (D008)	PLASTIC DIP (P)	CERAMIC DIP (JG)	METAL CAN (L)	SMALL-OUTLINE (D014)	PLASTIC DIP (N)	CERAMIC DIP (J)	CHIP CARRIER (FK)	
0 $^{\circ}$ C to 70 $^{\circ}$ C	1 μ V	TLC2652AC-8D	TLC2652ACP	TLC2652ACJG	TLC2652ACL	TLC2652AC-14D	TLC2652ACN	TLC2652ACJ	—	
	3 μ V	TLC2652C-8D	TLC2652CP	TLC2652CJG	TLC2652CL	TLC2652C-14D	TLC2652CN	TLC2652CJ	—	
-40 $^{\circ}$ C to 85 $^{\circ}$ C	1 μ V	TLC2652AI-8D	TLC2652AIP	TLC2652AIJG	TLC2652AIL	TLC2652AI-14D	TLC2652AIN	TLC2652AIJ	—	
	3 μ V	TLC2652I-8D	TLC2652IP	TLC2652IJG	TLC2652IL	TLC2652I-14D	TLC2652IN	TLC2652IJ	—	
-55 $^{\circ}$ C to 125 $^{\circ}$ C	1 μ V	TLC2652AM-8D	TLC2652AMP	TLC2652AMJG	TLC2652AML	TLC2652AM-14D	TLC2652AMN	TLC2652AMJ	TLC2652AMFK	
	3 μ V	TLC2652M-8D	TLC2652MP	TLC2652MJG	TLC2652ML	TLC2652M-14D	TLC2652MN	TLC2652MJ	TLC2652MFK	

D008 and D014 packages are available taped and reeled. Add "R" suffix to device type when ordering (e.g., TLC2652AC-8DR).

Advanced LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TLC2652, TLC2652A

Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

description (continued)

Chopper stabilization techniques make possible extremely high dc precision by continuously nulling input offset voltage even during variations in temperature, time, common-mode voltage, and power supply voltage. In addition, low-frequency noise voltage is significantly reduced. This high precision, coupled with the extremely high input impedance of the CMOS input stage, makes the TLC2652 and TLC2652A an ideal choice for low-level signal processing applications such as strain gauges, thermocouples, and other transducer amplifiers. (For applications that require extremely low noise and higher usable bandwidth, use the TLC2654 or TLC2654A device, which has a chopping frequency of 10 kHz.)

The TLC2652 and TLC2652A input common-mode range includes the negative rail, thereby providing superior performance in either single-supply or split-supply applications, even at power supply voltage levels as low as ± 1.9 V.

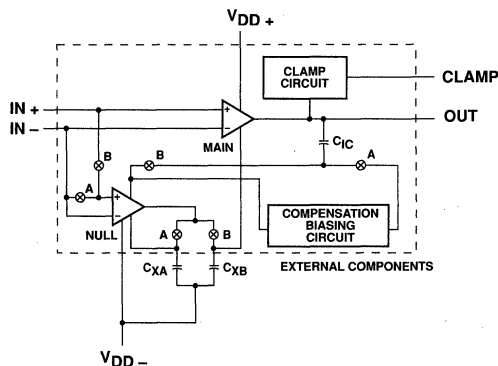
Two external capacitors are required for operation of the device; however, the on-chip chopper control circuitry is transparent to the user. On devices in the 14-pin and 20-pin packages, the control circuitry is made accessible to allow the user the option of controlling the clock frequency with an external frequency source. In addition, the clock threshold level of the TLC2652 and TLC2652A require no level shifting when used in the single-supply configuration with a normal CMOS or TTL clock input.

Innovative circuit techniques are used on the TLC2652 and TLC2652A to allow exceptionally fast overload recovery time. If desired, an output clamp pin is available to reduce the recovery time even further.

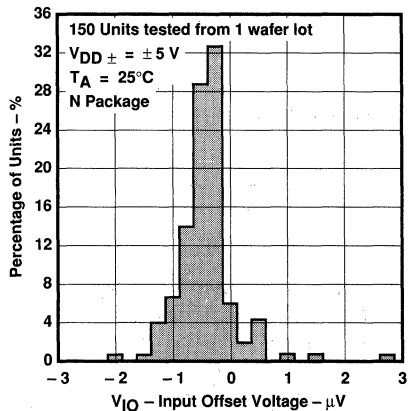
The device inputs and output are designed to withstand -100 -mA surge currents without sustaining latchup. Additionally, the TLC2652 and TLC2652A incorporate internal ESD protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.

The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I-suffix devices are characterized for operation from -40°C to 85°C . The C-suffix devices are characterized for operation from 0°C to 70°C .

functional block diagram



DISTRIBUTION OF TLC2652 INPUT OFFSET VOLTAGE



TLC2652M, TLC2652AM

Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} \pm = \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2652AM			TLC2652M			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.5	1	0.6	3	μV	
		Full range	4		6			
α_{VIO} Temperature coefficient of input offset voltage		-55°C to 125°C	0.003	0.03	0.003	0.03	$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)		25°C	0.003	0.02	0.003	0.06	$\mu\text{V}/\text{mo}$	
I_{IO} Input offset current		25°C	2		2		pA	
		Full range	500		500			
I_{IB} Input bias current		25°C	4		4		pA	
		Full range	500		500			
V_{ICR} Common-mode input voltage range		$R_S = 50\ \Omega$	Full range	-5 to 3.1	-5 to 3.1		V	
V_{OM+} Maximum positive peak output voltage swing		$R_L = 10\ \text{k}\Omega$, See Note 5	25°C	4.7	4.8	4.7	4.8	V
	Full range		4.7		4.7			
V_{OM-} Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$, See Note 5	25°C	-4.7	-4.9	-4.7	-4.9	V	
		Full range	-4.7		-4.7			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	135	150	120	150	dB	
		Full range	120		120			
f_{ch} Internal chopping frequency		25°C	450		450		Hz	
Clamp on-state current	$R_L = 100\ \text{k}\Omega$	25°C	25		25		μA	
		Full range	25		25			
Clamp off-state current	$V_O = -4\ \text{V to } 4\ \text{V}$	25°C	100		100		pA	
		Full range	100		100			
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR}\ \text{min}, R_S = 50\ \Omega$	25°C	120	140	120	140	dB	
		Full range	120		120			
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} \pm = \pm 1.9\ \text{V to } \pm 8\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C	120	135	120	135	dB	
		Full range	120		120			
I_{DD} Supply current	$V_O = 0, \text{No load}$	25°C	1.5	2.4	1.5	2.4	mA	
		Full range	2.5		2.5			

†Full range is -55°C to 125°C.

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. Output clamp is not connected.

2

Operational Amplifiers

TLC2652M, TLC2652AM
Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED
OPERATIONAL AMPLIFIERS

operating characteristics at specified free-air temperature, $V_{DD} \pm = \pm 5$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2652AM			TLC2652M			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR +	Positive slew rate at unity gain	$V_O = \pm 2.3$ V, $R_L = 10$ k Ω , $C_L = 100$ pF	25°C	2	2.8	2	2.8	V/ μ s
			Full range	1.3		1.3		
SR –	Negative slew rate at unity gain	$V_O = \pm 2.3$ V, $R_L = 10$ k Ω , $C_L = 100$ pF	25°C	2.3	3.1	2.3	3.1	V/ μ s
			Full range	1.6		1.6		
V_n	Equivalent input noise voltage	$f = 10$ Hz	25°C	94		94		nV/ $\sqrt{\text{Hz}}$
			$f = 1$ kHz	23		23		
V_{NPP}	Peak-to-peak equivalent input noise voltage	$f = 0$ to 1 Hz	25°C	0.8		0.8		μ V
			$f = 0$ to 10 Hz	2.8		2.8		
I_n	Equivalent input noise current	$f = 1$ kHz	25°C	0.004		0.004		pA/ $\sqrt{\text{Hz}}$
	Gain-bandwidth product	$f = 10$ kHz, $R_L = 10$ k Ω , $C_L = 100$ pF	25°C	1.9		1.9		MHz
ϕ_m	Phase margin at unity gain	$R_L = 10$ k Ω , $C_L = 100$ pF	25°C	48°		48°		

†Full range is – 55°C to 125°C.

2
Operational Amplifiers

TLC2652I, TLC2652AI

Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} \pm = \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†	TLC2652AI			TLC2652I			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO}	Input offset voltage		25°C	0.5	1	0.6	3	μV	
			Full range	2.95		4.95			
α_{VIO}	Temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	-40°C to 85°C	0.003	0.03	0.003	0.03	$\mu\text{V}/^\circ\text{C}$	
	Input offset voltage long-term drift (see Note 4)		25°C	0.003	0.02	0.003	0.06	$\mu\text{V}/\text{mo}$	
I_{IO}	Input offset current		25°C	2		2		μA	
I_{IB}	Input bias current		25°C	150		150		μA	
			Full range	4		4			
V_{ICR}	Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 3.1		-5 to 3.1		V	
			25°C	4.7	4.8	4.7	4.8		
V_{OM+}	Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$, See Note 5	25°C	4.7	4.8	4.7	4.8	V	
			Full range	4.7		4.7			
V_{OM-}	Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$, See Note 5	25°C	-4.7	-4.9	-4.7	-4.9	V	
			Full range	-4.7		-4.7			
A_{VD}	Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	135	150	120	150	dB	
			Full range	125		120			
f_{ch}	Internal chopping frequency		25°C	450		450		Hz	
			Clamp on-state current	$R_L = 100\ \text{k}\Omega$	25°C	25			25
	Clamp off-state current	$V_O = -4\ \text{V}$ to $4\ \text{V}$	25°C	100		100		μA	
			Full range	100		100			
CMRR	Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR}\ \text{min}, R_S = 50\ \Omega$	25°C	120	140	120	140	dB	
			Full range	120		120			
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} \pm = \pm 1.9\ \text{V}$ to $\pm 8\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C	120	135	120	135	dB	
			Full range	120		120			
I_{DD}	Supply current	$V_O = 0$, No load	25°C	1.5	2.4	1.5	2.4	mA	
			Full range	2.5		2.5			

†Full range is -40°C to 85°C.

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. Output clamp is not connected.

2

Operational Amplifiers

TLC2652I, TLC2652AI
Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED
OPERATIONAL AMPLIFIERS

operating characteristics at specified free-air temperature, $V_{DD} \pm = \pm 5$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2652AI			TLC2652I			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR +	Positive slew rate at unity gain $V_O = \pm 2.3$ V, $R_L = 10$ k Ω , $C_L = 100$ pF	25°C	2	2.8		2	2.8	V/ μ s
		Full range	1.4		1.4			
SR –	Negative slew rate at unity gain	25°C	2.3	3.1		2.3	3.1	V/ μ s
		Full range	1.7		1.7			
V_n	Equivalent input noise voltage (see Note 6)	f = 10 Hz	25°C	94	140		94	nV/ \sqrt Hz
		f = 1 kHz	25°C	23	35		23	
V_{NPP}	Peak-to-peak equivalent input noise voltage	f = 0 to 1 Hz	25°C	0.8		0.8		μ V
		f = 0 to 10 Hz	25°C	2.8		2.8		
I_n	Equivalent input noise current	f = 1 kHz	25°C	0.004		0.004	pA/ \sqrt Hz	
	Gain-bandwidth product	f = 10 kHz, $R_L = 10$ k Ω , $C_L = 100$ pF	25°C	1.9		1.9	MHz	
ϕ_m	Phase margin at unity gain	$R_L = 10$ k Ω , $C_L = 100$ pF	25°C	48°		48°		

†Full range is – 40°C to 85°C.

NOTE 6: This parameter is tested on a sample basis for the TLC2652A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

TLC2652C, TLC2652AC
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electrical characteristics at specified free-air temperature, $V_{DD} \pm \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2652AC			TLC2652C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.5	1	0.6	3	μV	
		Full range	2.35		4.35			
α_{VIO} Temperature coefficient of input offset voltage		0°C to 70°C	0.003	0.03	0.003	0.03	$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)		25°C	0.003	0.02	0.003	0.06	$\mu\text{V}/\text{mo}$	
I_{IO} Input offset current		25°C	2		2		pA	
		Full range	100		100			
I_{IB} Input bias current		25°C	4		4		pA	
		Full range	100		100			
V_{ICR} Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 3.1	-5 to 3.1		V		
V_{OM+} Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$, See Note 5	25°C	4.7	4.8	4.7	4.8	V	
		Full range	4.7		4.7			
V_{OM-} Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$, See Note 5	25°C	-4.7	-4.9	-4.7	-4.9	V	
		Full range	-4.7		-4.7			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	135	150	120	150	dB	
		Full range	130		120			
f_{ch} Internal chopping frequency		25°C	450		450		Hz	
Clamp on-state current	$R_L = 100\ \text{k}\Omega$	25°C	25		25		μA	
		Full range	25		25			
Clamp off-state current	$V_O = -4\ \text{V to } 4\ \text{V}$	25°C	100		100		pA	
		Full range	100		100			
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR}\ \text{min}, R_S = 50\ \Omega$	25°C	120	140	120	140	dB	
		Full range	120		120			
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} \pm = \pm 1.9\ \text{V to } \pm 8\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C	120	135	120	135	dB	
		Full range	120		120			
I_{DD} Supply current	$V_O = 0, \text{No load}$	25°C	1.5	2.4	1.5	2.4	mA	
		Full range	2.5		2.5			

†Full range is 0°C to 70°C.

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. Output clamp is not connected.

TLC2652C, TLC2652AC
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operating characteristics at specified free-air temperature, $V_{DD} \pm = \pm 5$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2652AC			TLC2652C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR +	Positive slew rate at unity gain $V_O = \pm 2.3$ V, $R_L = 10$ k Ω , $C_L = 100$ pF	25°C	2	2.8	2	2.8		V/ μ s
		Full range	1.5		1.5			
SR -	Negative slew rate at unity gain	25°C	2.3	3.1	2.3	3.1		V/ μ s
		Full range	1.8		1.8			
V_n	Equivalent input noise voltage (see Note 6)	f = 10 Hz	25°C	94	140	94		nV/ $\sqrt{\text{Hz}}$
		f = 1 kHz	25°C	23	35	23		
V_{NPP}	Peak-to-peak equivalent input noise voltage	f = 0 to 1 Hz	25°C	0.8		0.8		μ V
		f = 0 to 10 Hz	25°C	2.8		2.8		
I_n	Equivalent input noise current	f = 1 kHz	25°C	0.004		0.004		pA/ $\sqrt{\text{Hz}}$
	Gain-bandwidth product	f = 10 kHz, $R_L = 10$ k Ω , $C_L = 100$ pF	25°C	1.9		1.9		MHz
ϕ_m	Phase margin at unity gain	$R_L = 10$ k Ω , $C_L = 100$ pF	25°C	48°		48°		

†Full range is 0°C to 70°C.

NOTE 6: This parameter is tested on a sample basis for the TLC2652A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

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Operational Amplifiers

TLC2652, TLC2652A
Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED
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TYPICAL CHARACTERISTICS

table of graphs

			FIGURE
V_{IO}	Normalized input offset voltage	vs Chopping frequency	1
I_{IB}	Input bias current	vs Common-mode input voltage	2
		vs Chopping frequency	3
		vs Temperature	4
I_{IO}	Input offset current	vs Chopping frequency	5
		vs Temperature	6
	Clamp current	vs Output voltage	7
V_{OPP}	Maximum peak-to-peak output voltage swing	vs Frequency	8
V_{OM}	Maximum peak output voltage swing	vs Output current	9, 10
		vs Temperature	11, 12
A_{VD}	Differential voltage amplification	vs Frequency	13
		vs Temperature	14
f_{ch}	Chopping frequency	vs Supply voltage	15
		vs Temperature	16
I_{DD}	Supply current	vs Supply voltage	17
		vs Temperature	18
I_{OS}	Short-circuit output current	vs Supply voltage	19
		vs Temperature	20
SR	Slew rate	vs Supply voltage	21
		vs Temperature	22
	Pulse response	Small-signal	23
		Large-signal	24
V_{NPP}	Peak-to-peak equivalent input noise voltage	vs Chopping frequency	25, 26
V_n	Equivalent input noise voltage	vs Frequency	27
	Gain-bandwidth product	vs Supply voltage	28
		vs Temperature	29
ϕ_m	Phase margin	vs Supply voltage	30
		vs Temperature	31
		vs Load capacitance	32
	Phase shift	vs Frequency	13

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Operational Amplifiers

TYPICAL CHARACTERISTICS†

NORMALIZED INPUT OFFSET VOLTAGE
 VS
 CHOPPING FREQUENCY

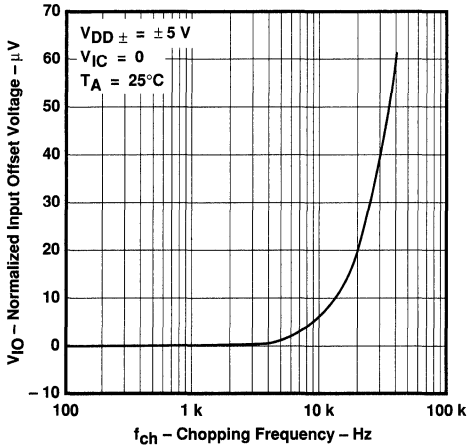


FIGURE 1

INPUT BIAS CURRENT
 VS
 COMMON-MODE INPUT VOLTAGE

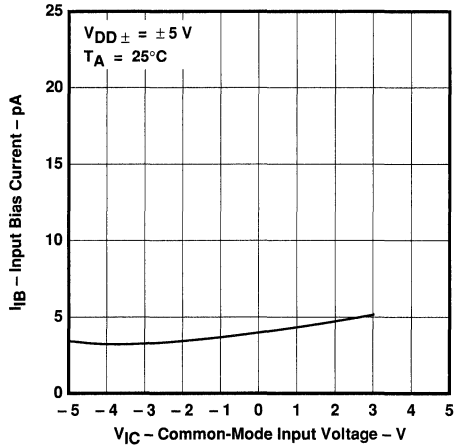


FIGURE 2

INPUT BIAS CURRENT
 VS
 CHOPPING FREQUENCY

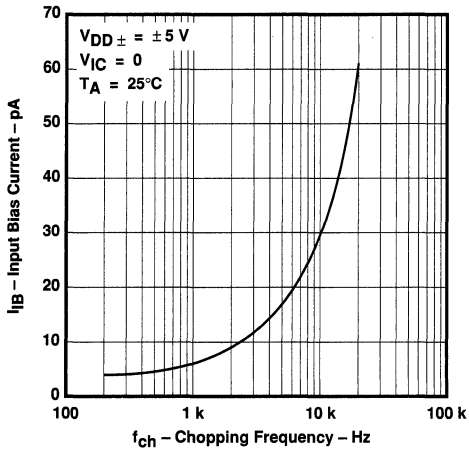


FIGURE 3

INPUT BIAS CURRENT
 VS
 FREE-AIR TEMPERATURE

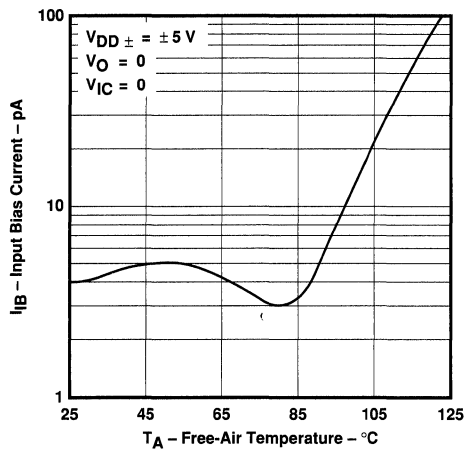


FIGURE 4

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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Operational Amplifiers

TYPICAL CHARACTERISTICS†

INPUT OFFSET CURRENT
vs
CHOPPING FREQUENCY

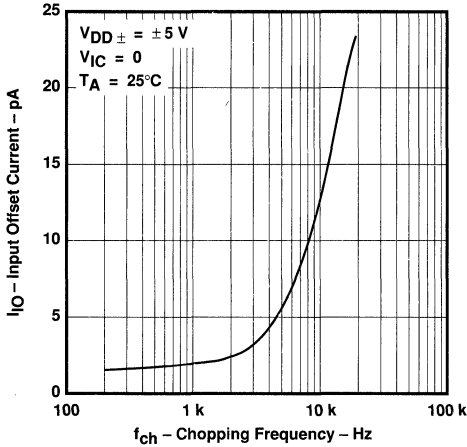


FIGURE 5

INPUT OFFSET CURRENT
vs
FREE-AIR TEMPERATURE

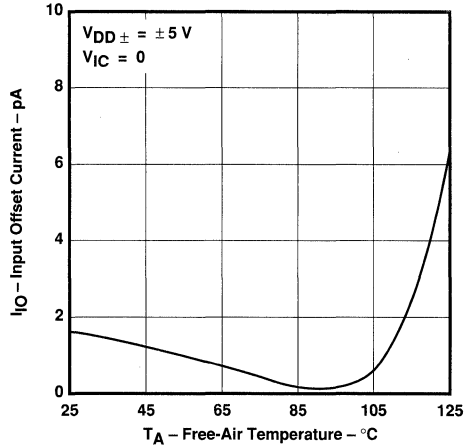


FIGURE 6

CLAMP CURRENT
vs
OUTPUT VOLTAGE

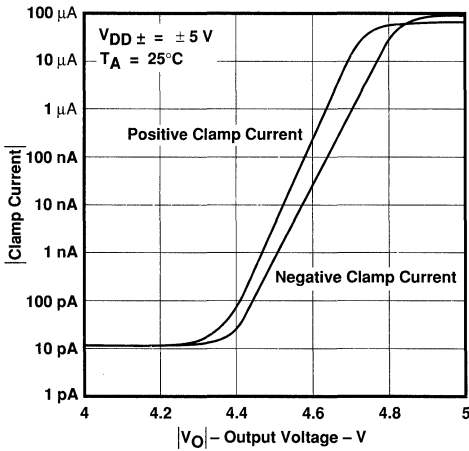


FIGURE 7

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
vs
FREQUENCY

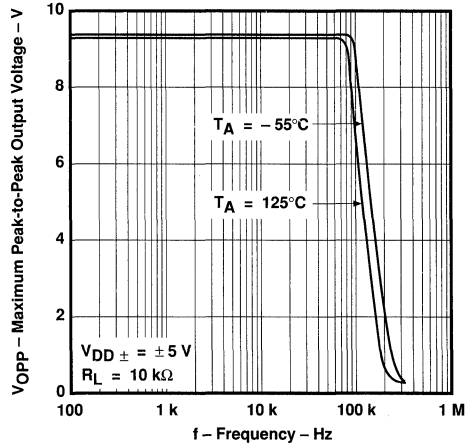


FIGURE 8

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE
VS
OUTPUT CURRENT

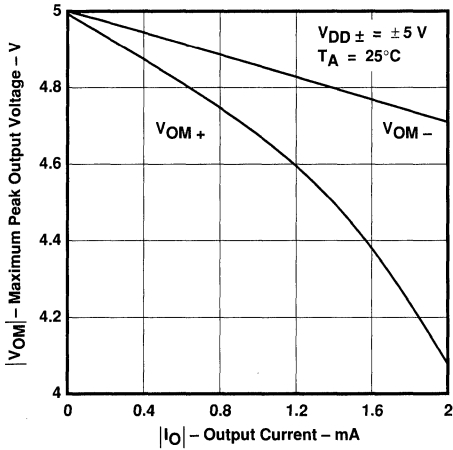


FIGURE 9

MAXIMUM PEAK OUTPUT VOLTAGE
VS
OUTPUT CURRENT

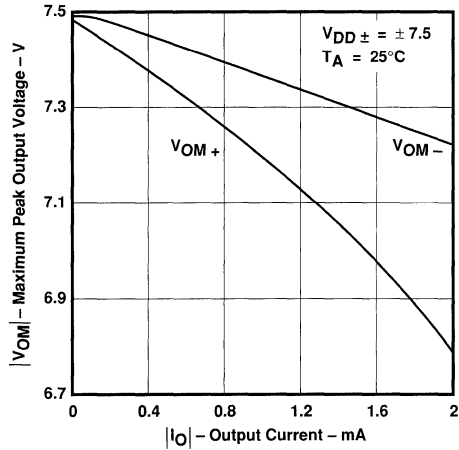


FIGURE 10

MAXIMUM PEAK OUTPUT VOLTAGE
VS
FREE-AIR TEMPERATURE

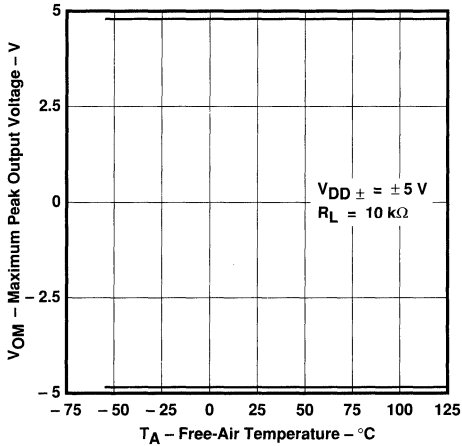


FIGURE 11

MAXIMUM PEAK OUTPUT VOLTAGE
VS
FREE-AIR TEMPERATURE

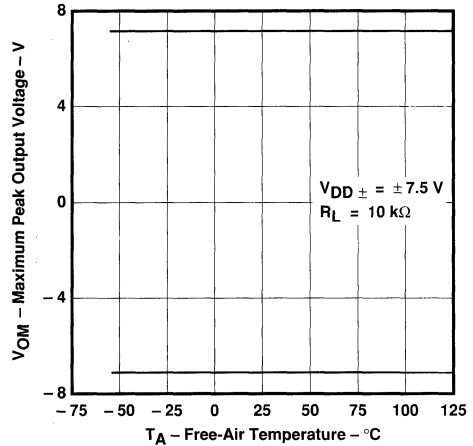


FIGURE 12

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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Operational Amplifiers

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OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT VS FREQUENCY

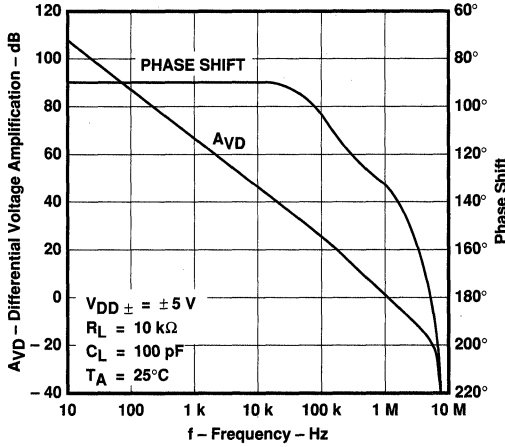


FIGURE 13

LARGE-SIGNAL VOLTAGE AMPLIFICATION VS FREE-AIR TEMPERATURE

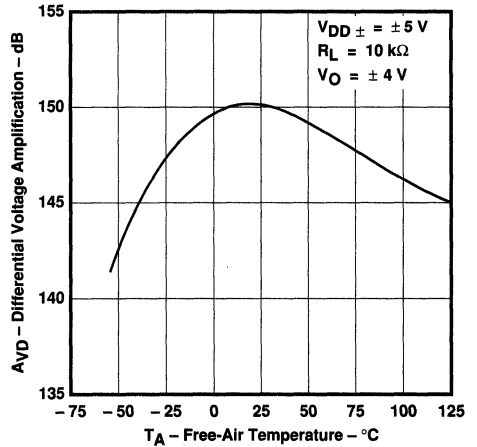


FIGURE 14

CHOPPING FREQUENCY VS SUPPLY VOLTAGE

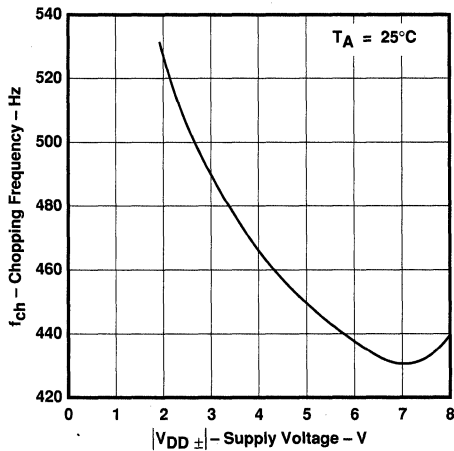


FIGURE 15

CHOPPING FREQUENCY VS FREE-AIR TEMPERATURE

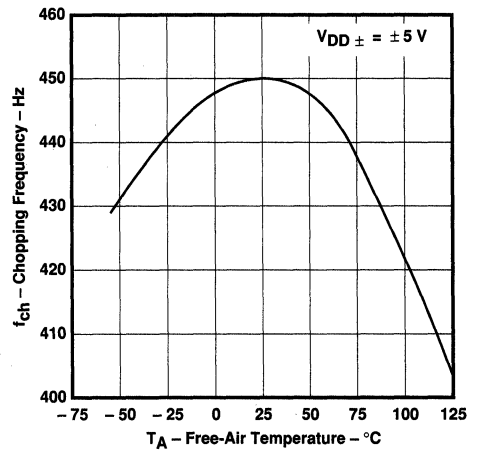


FIGURE 16

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

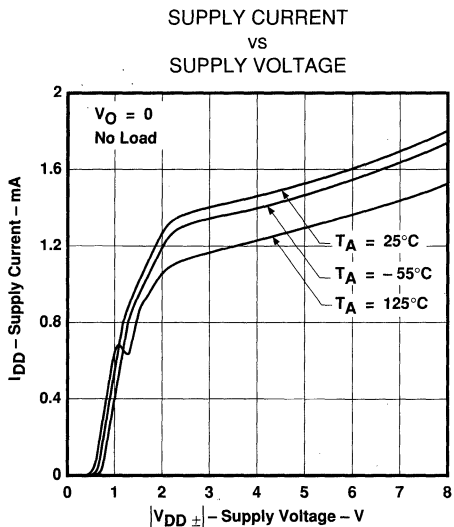


FIGURE 17

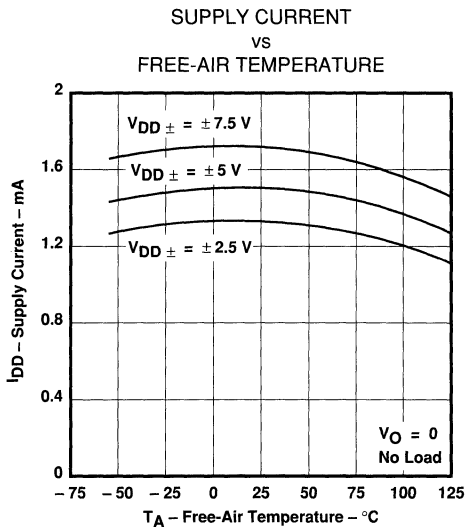


FIGURE 18

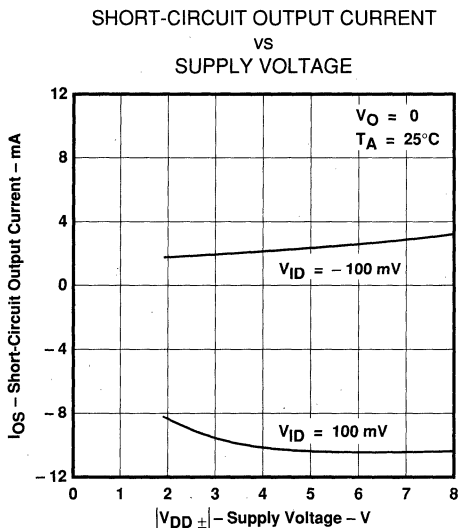


FIGURE 19

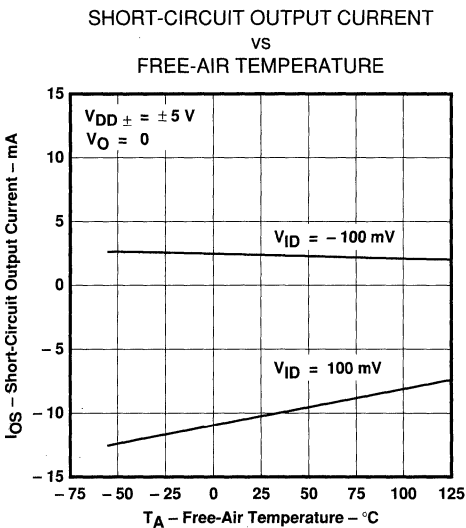


FIGURE 20

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

SLEW RATE
 VS
 SUPPLY VOLTAGE

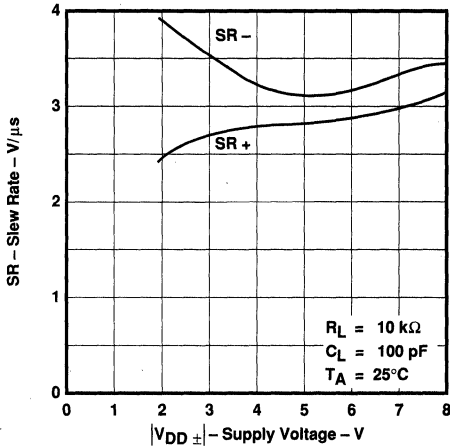


FIGURE 21

SLEW RATE
 VS
 FREE-AIR TEMPERATURE

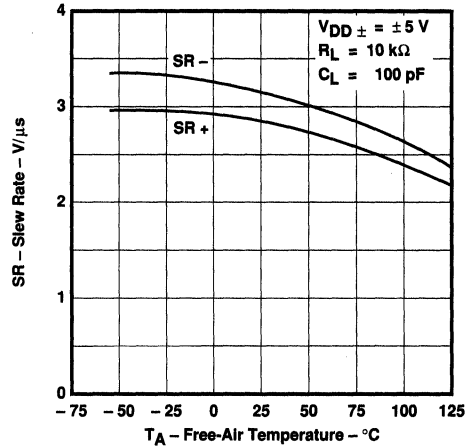


FIGURE 22

VOLTAGE-FOLLOWER
 SMALL-SIGNAL
 PULSE RESPONSE

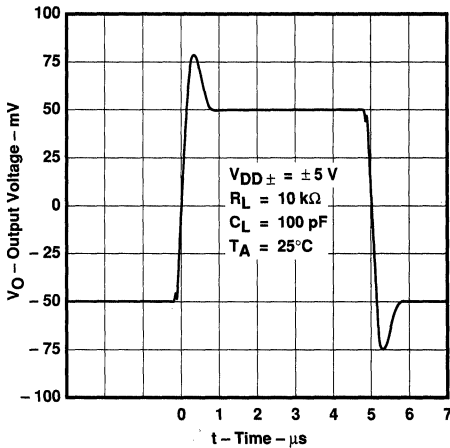


FIGURE 23

VOLTAGE-FOLLOWER
 LARGE-SIGNAL
 PULSE RESPONSE

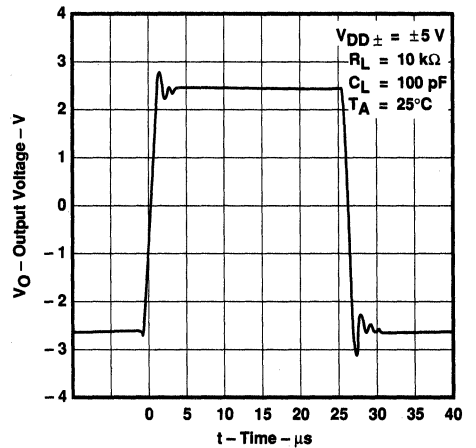


FIGURE 24

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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TYPICAL CHARACTERISTICS

PEAK-TO-PEAK INPUT NOISE VOLTAGE
 VS
 CHOPPING FREQUENCY

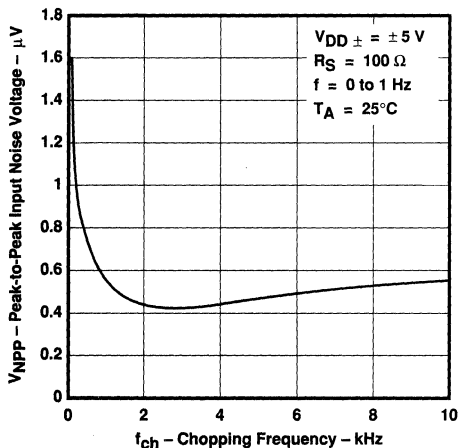


FIGURE 25

PEAK-TO-PEAK INPUT NOISE VOLTAGE
 VS
 CHOPPING FREQUENCY

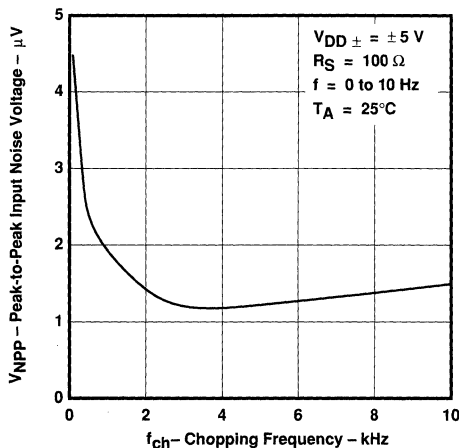


FIGURE 26

EQUIVALENT INPUT NOISE VOLTAGE
 VS
 FREQUENCY

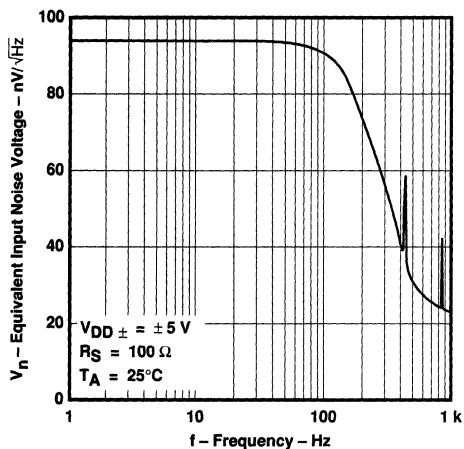


FIGURE 27

GAIN-BANDWIDTH PRODUCT
 VS
 SUPPLY VOLTAGE

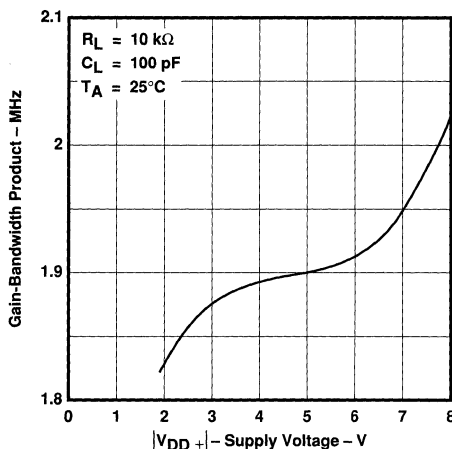


FIGURE 28

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Operational Amplifiers

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TYPICAL CHARACTERISTICS†

GAIN-BANDWIDTH PRODUCT
VS
FREE-AIR TEMPERATURE

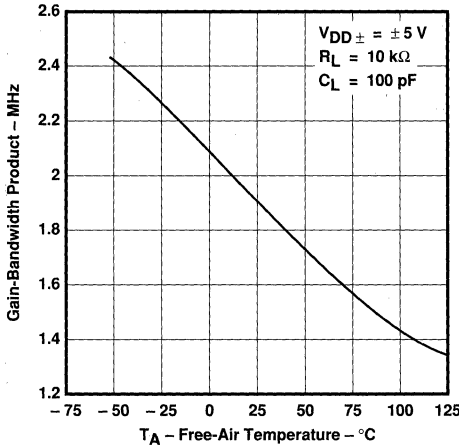


FIGURE 29

PHASE MARGIN
VS
SUPPLY VOLTAGE

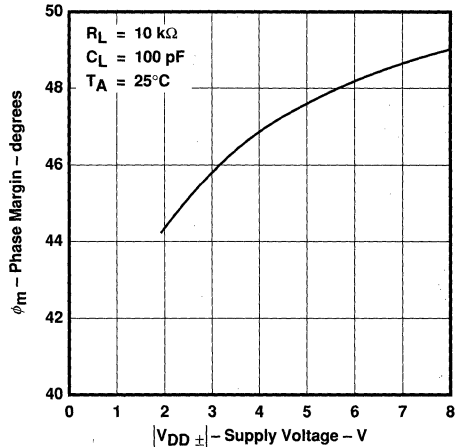


FIGURE 30

PHASE MARGIN
VS
FREE-AIR TEMPERATURE

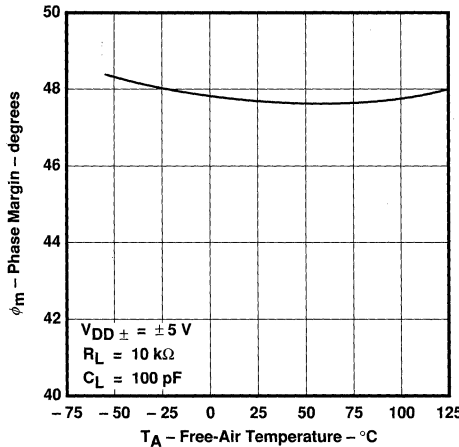


FIGURE 31

PHASE MARGIN
VS
LOAD CAPACITANCE

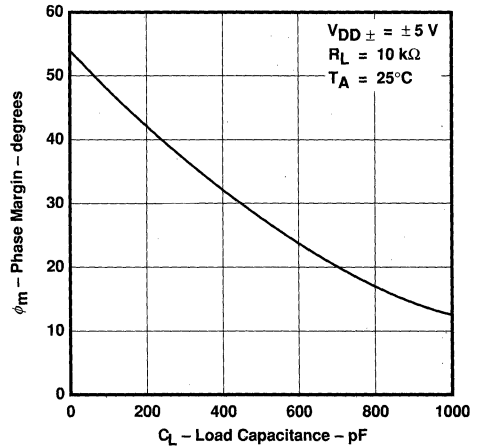


FIGURE 32

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL APPLICATION DATA

capacitor selection and placement

The two important factors to consider when selecting external capacitors C_{XA} and C_{XB} are leakage and dielectric absorption. Both factors can cause system degradation that can negate the performance advantages realized by using the TLC2652.

Degradation from capacitor leakage becomes more apparent with increasing temperatures. Low-leakage capacitors and standoffs are recommended for operation at $T_A = 125^\circ\text{C}$. In addition, guardbands are recommended around the capacitor connections on both sides of the printed circuit board to alleviate problems caused by surface leakage on circuit boards.

Capacitors with high dielectric absorption tend to take several seconds to settle upon application of power, which directly affects input offset voltage. In applications where fast settling of input offset voltage is needed, it is recommended that high-quality film capacitors, such as mylar, polystyrene, or polypropylene, be used. In other applications, however, a ceramic or other low-grade capacitor may suffice.

Unlike many choppers available today, the TLC2652 is designed to function with values of C_{XA} and C_{XB} in the range of $0.1\ \mu\text{F}$ to $1\ \mu\text{F}$ without degradation to input offset voltage or input noise voltage. These capacitors should be located as close as possible to the C_{XA} and C_{XB} pins and returned to either the V_{DD-} pin or the C RETURN pin. Note that on many choppers, connecting these capacitors to the V_{DD-} pin will cause degradation in the noise performance. This problem is eliminated on the TLC2652.

internal/external clock

The TLC2652 has an internal clock that sets the chopping frequency to a nominal value of 450 Hz. On 8-pin packages, the chopping frequency can only be controlled by the internal clock; however, on all 14-pin packages and the 20-pin FK package, the device chopping frequency may be set by the internal clock or controlled externally by use of the INT/EXT and CLK IN pins. To use the internal 450-Hz clock, no connection is necessary. If external clocking is desired, connect the INT/EXT pin to V_{DD-} and the external clock to CLK IN. The external clock trip point is 2.5 V above the negative rail; however, the CLK IN pin may be driven from the negative rail to 5 V above the negative rail. If this level is exceeded, damage could occur to the device unless the current into the CLK IN pin is limited to $\pm 5\ \text{mA}$. When operating in the single-supply configuration, this feature allows the TLC2652 to be driven directly by 5-V TTL and CMOS logic. A divide-by-two frequency divider interfaces with the CLK IN pin and sets the chopping frequency. The chopping frequency appears on the CLK OUT pin. The duty cycle of the external clock is not critical but should be kept between 30% and 60%.

overload recovery/output clamp

When large differential input voltage conditions are applied to the TLC2652, the nulling loop will attempt to prevent the output from saturating by driving C_{XA} and C_{XB} to internally-clamped voltage levels. Once the overdrive condition is removed, a period of time is required to allow the built-up charge to dissipate. This time period is defined as overload recovery time (see Figure 33). Typical overload recovery time for the TLC2652 is significantly faster than competitive products; however, if required, this time can be

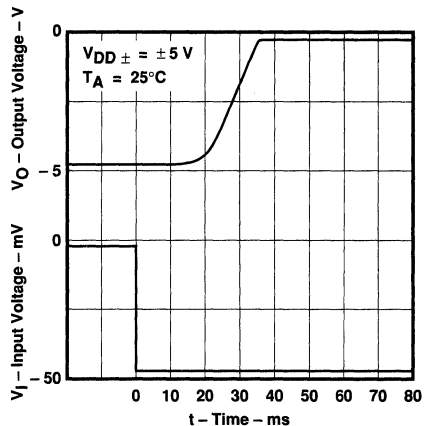


FIGURE 33. OVERLOAD RECOVERY

TYPICAL APPLICATION DATA

reduced further by using internal clamp circuitry accessible through the CLAMP pin.

The clamp is simply a switch that is automatically activated when the output is approximately 1 V from either supply rail. When connected to the inverting input (in parallel with the closed-loop feedback resistor), the closed-loop gain is reduced and the TLC2652 output is prevented from going into saturation. Since the output must source or sink current through the switch (see Figure 7), the maximum output voltage swing is slightly reduced.

thermoelectric effects

To take advantage of the extremely low offset voltage drift of the TLC2652, care must be taken to compensate for the thermoelectric effects present when two dissimilar metals are brought into contact with each other (such as device leads being soldered to a printed circuit board). Dissimilar metal junctions can produce thermoelectric voltages in the range of several microvolts per degree Celsius (orders of magnitude greater than the 0.003 $\mu\text{V}/^\circ\text{C}$ typical of the TLC2652).

To help minimize thermoelectric effects, careful attention should be paid to component selection and circuit board layout. Avoid the use of nonsoldered connections (such as sockets, relays, switches, etc.) in the input signal path. Cancel thermoelectric effects by duplicating the number of components and junctions in each device input. The use of low-thermoelectric-coefficient components, such as wire-wound resistors, is also beneficial.

latchup avoidance

Because CMOS devices are susceptible to latchup due to their inherent parasitic thyristors, the TLC2652 inputs and output were designed to withstand -100-mA surge currents without sustaining latchup; however, techniques to reduce the chance of latchup should be used whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltages should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μF typical) located across the supply rails as close to the device as is possible.

The current path established if latchup occurs is usually between the supply rails and is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor. The chance of latchup occurring increases with increasing temperature and supply voltage.

electrostatic discharge protection

The TLC2652 incorporates internal ESD protection circuits that prevent functional failures at voltages at or below 2000 V. Care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.

theory of operation

Chopper-stabilized operational amplifiers offer the best dc performance of any monolithic operational amplifier. This superior performance is the result of using two operational amplifiers – a main amplifier and a nulling amplifier – plus oscillator-controlled logic and two external capacitors to create a system that behaves as a single amplifier. With this approach, the TLC2652 achieves submicrovolt input offset voltage, submicrovolt noise voltage, and offset voltage variations with temperature in the $\text{nV}/^\circ\text{C}$ range.

The TLC2652 on-chip control logic produces two dominant clock phases; a nulling phase and an amplifying phase. The term *chopper-stabilized* derives from the process of switching between these two clock phases. Figure 34 shows a simplified block diagram of the TLC2652. Switches A and B are make-before-break types. During the nulling phase, switch A is closed, shorting the nulling amplifier inputs together and allowing the

TYPICAL APPLICATION DATA

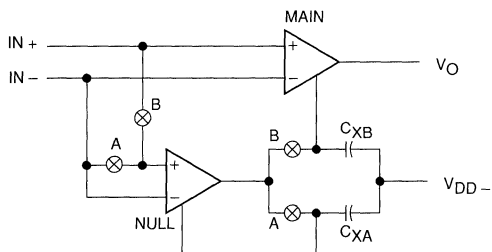


FIGURE 34. TLC2652 SIMPLIFIED BLOCK DIAGRAM

nulling amplifier to reduce its own input offset voltage by feeding its output signal back to an inverting input node. Simultaneously, external capacitor C_{XA} stores the nulling potential to allow the offset voltage of the amplifier to remain nulled during the amplifying phase.

During the amplifying phase, switch B is closed, connecting the output of the nulling amplifier to a noninverting input of the main amplifier. In this configuration, the input offset voltage of the main amplifier is nulled. Also, external capacitor C_{XB} stores the nulling potential to allow the offset voltage of the main amplifier to remain nulled during the next nulling phase.

This continuous chopping process allows offset voltage nulling during variations in time and temperature and over the common-mode input voltage range and power-supply range. In addition, because the low-frequency signal path is through both the null and main amplifiers, extremely high gain is achieved.

The low-frequency noise of a chopper amplifier depends on the magnitude of the component noise prior to chopping and the capability of the circuit to reduce this noise while chopping. The use of the Advanced LinCMOS process with its low-noise analog MOS transistors and patent-pending input stage design significantly reduces the input noise voltage.

The primary source of nonideal operation in chopper-stabilized amplifiers is error charge from the switches.

As charge imbalance accumulates on critical nodes, input offset voltage can increase, especially with increasing chopping frequency. This problem has been significantly reduced in the TLC2652 by use of a patent-pending compensation circuit and the Advanced LinCMOS process.

The TLC2652 incorporates a feed-forward design that ensures continuous frequency response. Essentially, the gain magnitude of the nulling amplifier and compensation network crosses unity at the break frequency of the main amplifier. As a result, the high-frequency response of the system is the same as the frequency response of the main amplifier. This approach also ensures that the slewing characteristics remain the same during both the nulling and amplifying phases.

2

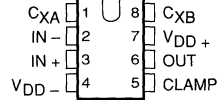
Operational Amplifiers

TLC2654, TLC2654A Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

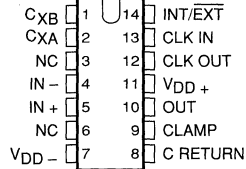
D3174, NOVEMBER 1988

- **Input Noise Voltage . . .**
 0.5 μV p-p Typ, $f = 0$ to 1 Hz
 1.5 μV p-p Typ, $f = 0$ to 10 Hz
 47 $\text{nV}/\sqrt{\text{Hz}}$ Typ, $f = 10$ Hz
 13 $\text{nV}/\sqrt{\text{Hz}}$ Typ, $f = 1$ kHz
- **High Chopping Frequency . . . 10 kHz Typ**
- **No Clock Noise Below 10 kHz**
- **No Intermodulation Error Below 5 kHz**
- **Low Input Offset Voltage . . . 10 μV Max**
- **Excellent Offset Voltage Stability with Temperature . . . 0.3 $\mu\text{V}/^\circ\text{C}$ Max**
- **A_{VD} . . . 135 dB Min**
- **CMRR . . . 110 dB Min**
- **k_{SVR} . . . 120 dB Min**
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Includes the Negative Rail**
- **No Noise Degradation with External Capacitors Connected to V_{DD} -**

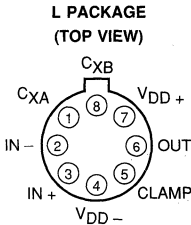
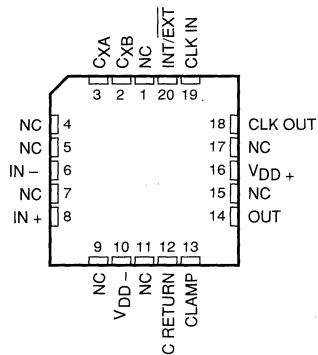
**D008, JG, or P PACKAGE
(TOP VIEW)**



**D014, J, or N PACKAGE
(TOP VIEW)**



**FK PACKAGE
(TOP VIEW)**



Pin 4 of the L package is in electrical contact with the case

NC - No internal connection

AVAILABLE OPTIONS

T _A	V _{IO} max AT 25°C	PACKAGE							
		8-PIN				14-PIN			20-PIN
		SMALL- OUTLINE (D008)	PLASTIC DIP (P)	CERAMIC DIP (JG)	METAL CAN (L)	SMALL- OUTLINE (D014)	PLASTIC DIP (N)	CERAMIC DIP (J)	CHIP CARRIER (FK)
0°C to 70°C	10 μV	TLC2654AC-8D	TLC2654ACP	TLC2654ACJG	TLC2654ACL	TLC2654AC-14D	TLC2654ACN	TLC2654ACJ	—
	20 μV	TLC2654C-8D	TLC2654CP	TLC2654CJG	TLC2654CL	TLC2654C-14D	TLC2654CN	TLC2654CJ	—
-40°C to 85°C	10 μV	TLC2654AI-8D	TLC2654AIP	TLC2654AIJG	TLC2654AIL	TLC2654AI-14D	TLC2654AIN	TLC2654AIJ	—
	20 μV	TLC2654I-8D	TLC2654IP	TLC2654IJG	TLC2654IL	TLC2654I-14D	TLC2654IN	TLC2654IJ	—
-55°C to 125°C	10 μV	TLC2654AM-8D	TLC2654AMP	TLC2654AMJG	TLC2654AML	TLC2654AM-14D	TLC2654AMN	TLC2654AMJ	TLC2654AMFK
	20 μV	TLC2654M-8D	TLC2654MP	TLC2654MJG	TLC2654ML	TLC2654M-14D	TLC2654MN	TLC2654MJ	TLC2654MFK

D008 and D014 packages are available taped-and-reeled. Add "R" suffix to device type when ordering (e.g., TLC2654AC-8DR).

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TLC2654, TLC2654A

Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

description

The TLC2654 and TLC2654A are low-noise chopper-stabilized operational amplifiers using the Advanced LinCMOS™ process. Combining this process with chopper stabilization circuitry makes possible excellent dc precision. In addition, circuit techniques have been added that give the TLC2654 and TLC2654A noise performance unsurpassed by similar devices.

Chopper stabilization techniques provide for extremely high dc precision by continuously nulling input offset voltage even during variations in temperature, time, common-mode voltage, and power supply voltage. The high chopping frequency of the TLC2654 and TLC2654A provides excellent noise performance in a frequency spectrum from near dc to 10 kHz. In addition, intermodulation or aliasing error is eliminated from frequencies up to 5 kHz.

This high dc precision and low noise, coupled with the extremely high input impedance of the CMOS input stage, make the TLC2654 and TLC2654A an ideal choice for a broad range of applications such as low-level low-frequency thermocouple amplifiers and strain gauges, as well as wide-bandwidth and subsonic circuits. (For applications requiring even greater dc precision, use the TLC2652 or TLC2652A device, which has a chopping frequency of 450 Hz.)

The TLC2654 and TLC2654A common-mode input voltage range includes the negative rail, thereby providing superior performance in either single-supply or split-supply applications, even at power supply voltage levels as low as ± 2.3 V.

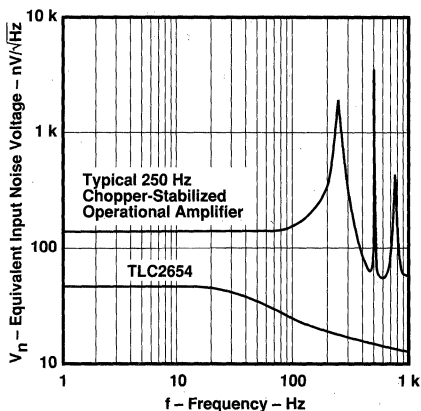
Two external capacitors are required to operate the device; however, the on-chip chopper control circuitry is transparent to the user. On devices in the 14-pin and 20-pin packages, the control circuitry is accessible, allowing the user the option of controlling the clock frequency with an external frequency source. In addition, the clock threshold of the TLC2654 and TLC2654A requires no level shifting when used in the single-supply configuration with a normal CMOS or TTL clock input.

Innovative circuit techniques used on the TLC2654 and TLC2654A allow exceptionally fast overload recovery time. An output clamp pin is available to reduce the recovery time further.

The device inputs and output are designed to withstand -100 mA surge currents without sustaining latchup. In addition, the TLC2654 and TLC2654A incorporate internal ESD protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015; however, care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.

The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I-suffix devices are characterized for operation from -40°C to 85°C . The C-suffix devices are characterized for operation from 0°C to 70°C .

EQUIVALENT INPUT NOISE VOLTAGE
VS
FREQUENCY



TLC2654M, TLC2654AM

Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} \pm = \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2654AM			TLC2654AM			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage		25°C	4	10	5	20	μV	
		Full range		40	50			
α_{VIO} Temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	-55°C to 125°C	0.004	0.3	0.004	0.3	$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)		25°C	0.003	0.02	0.003	0.06	$\mu\text{V}/\text{mo}$	
I_{IO} Input offset current		25°C	30		30		pA	
		Full range		500	500			
I_{IB} Input bias current		25°C	50		50		pA	
		Full range		500	500			
V_{ICR} Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7		-5 to 2.7		V	
V_{OM+} Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$, See Note 5	25°C	4.7	4.8	4.7	4.8	V	
		Full range	4.7		4.7			
V_{OM-} Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$, See Note 5	25°C	-4.7	-4.9	-4.7	-4.9	V	
		Full range	-4.7		-4.7			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	135	155	120	155	dB	
		Full range	120		120			
f_{ch} Internal chopping frequency		25°C	10		10		kHz	
Clamp on-state current	$R_L = 100\ \text{k}\Omega$	25°C	25		25		μA	
		Full range	25		25			
Clamp off-state current	$V_O = -4\ \text{V}$ to 4 V	25°C		100		100	pA	
		Full range		100		100		
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR\ \text{min}}, R_S = 50\ \Omega$	25°C	110	125	105	125	dB	
		Full range	110		105			
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} \pm = \pm 2.3\ \text{V}$ to $\pm 8\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C	120	125	110	125	dB	
		Full range	115		105			
I_{DD} Supply current	$V_O = 0, \text{ No load}$	25°C	1.5	2.1	1.5	2.1	mA	
		Full range		2.2		2.2		

†Full range is -55°C to 125°C.

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. Output clamp is not connected.

2

Operational Amplifiers

TLC2654M, TLC2654AM
Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED
OPERATIONAL AMPLIFIERS

operating characteristics at specified free-air temperature, $V_{DD} \pm = \pm 5 \text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2654AM			TLC2654M			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR + Positive slew rate at unity gain	$V_O = \pm 2.3 \text{ V}$, $R_L = 10 \text{ k}\Omega$, $C_L = 100 \text{ pF}$	25°C	1.5	2	1.5	2		V/ μs
		Full range	1.1		1.1			
SR – Negative slew rate at unity gain	$V_O = \pm 2.3 \text{ V}$, $R_L = 10 \text{ k}\Omega$, $C_L = 100 \text{ pF}$	25°C	2.3	3.7	2.3	3.7		V/ μs
		Full range	1.3		1.3			
V_n Equivalent input noise voltage	$f = 10 \text{ Hz}$ $f = 1 \text{ kHz}$	25°C		47		47		nV/ $\sqrt{\text{Hz}}$
		25°C		13		13		
V_{NPP} Peak-to-peak equivalent input noise voltage	$f = 0 \text{ to } 1 \text{ Hz}$ $f = 0 \text{ to } 10 \text{ Hz}$	25°C		0.5		0.5		μV
		25°C		1.5		1.5		
I_n Equivalent input noise current	$f = 1 \text{ kHz}$	25°C		0.004		0.004		pA/ $\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10 \text{ kHz}$, $R_L = 10 \text{ k}\Omega$, $C_L = 100 \text{ pF}$	25°C		1.9		1.9		MHz
ϕ_m Phase margin at unity gain	$R_L = 10 \text{ k}\Omega$, $C_L = 100 \text{ pF}$	25°C		48°		48°		

†Full range is – 55°C to 125°C.

2

Operational Amplifiers

TLC2654I, TLC2654AI

Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} \pm = \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2654AI			TLC2654I			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage		25°C	4	10	5	20	μV	
		Full range	30		40			
α_{VIO} Temperature coefficient of input offset voltage		-40°C to 85°C	0.004	0.3	0.004	0.3	$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.003	0.02	0.003	0.06	$\mu\text{V}/\text{mo}$	
I_{IO} Input offset current		25°C	30		30		pA	
	Full range	200		200				
I_{IB} Input bias current		25°C	50		50		pA	
		Full range	200		200			
V_{ICR} Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7		-5 to 2.7		V	
V_{OM+} Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$, See Note 5	25°C	4.7	4.8	4.7	4.8	V	
		Full range	4.7		4.7			
V_{OM-} Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$, See Note 5	25°C	-4.7	-4.9	-4.7	-4.9	V	
		Full range	-4.7		-4.7			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	135	155	120	155	dB	
		Full range	125		120			
f_{ch} Internal chopping frequency		25°C	10		10		kHz	
Clamp on-state current	$R_L = 100\ \text{k}\Omega$	25°C	25		25		μA	
		Full range	25		25			
Clamp off-state current	$V_O = -4\ \text{V to } 4\ \text{V}$	25°C	100		100		pA	
		Full range	100		100			
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR}\ \text{min}, R_S = 50\ \Omega$	25°C	110	125	105	125	dB	
		Full range	110		105			
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} \pm = \pm 2.3\ \text{V to } \pm 8\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C	120	125	110	125	dB	
		Full range	120		110			
I_{DD} Supply current	$V_O = 0, \text{ No load}$	25°C	1.5	2.1	1.5	2.1	mA	
		Full range	2.2		2.2			

†Full range is -40°C to 85°C.

- NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.
5. Output clamp is not connected.

2

Operational Amplifiers

TLC2654I, TLC2654AI
Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED
OPERATIONAL AMPLIFIERS

operating characteristics at specified free-air temperature, $V_{DD} \pm = \pm 5$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2654AI			TLC2654I			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR + Positive slew rate at unity gain	$V_O = \pm 2.3$ V, $R_L = 10$ k Ω , $C_L = 100$ pF	25°C	1.5	2	1.5	2		V/ μ s
		Full range	1.2		1.2			
SR – Negative slew rate at unity gain	$V_O = \pm 2.3$ V, $R_L = 10$ k Ω , $C_L = 100$ pF	25°C	2.3	3.7	2.3	3.7		V/ μ s
		Full range	1.5		1.5			
V_n Equivalent input noise voltage (see Note 6)	f = 10 Hz	25°C		47	75		47	nV/ $\sqrt{\text{Hz}}$
	f = 1 kHz	25°C		13	20		13	
V_{NPP} Peak-to-peak equivalent input noise voltage	f = 0 to 1 Hz	25°C		0.5			0.5	μ V
	f = 0 to 10 Hz	25°C		1.5			1.5	
I_n Equivalent input noise current	f = 1 kHz	25°C		0.004			0.004	pA/ $\sqrt{\text{Hz}}$
Gain-bandwidth product	f = 10 kHz, $R_L = 10$ k Ω , $C_L = 100$ pF	25°C		1.9			1.9	MHz
ϕ_m Phase margin at unity gain	$R_L = 10$ k Ω , $C_L = 100$ pF	25°C		48°			48°	

†Full range is – 40°C to 85°C.

NOTE 6: This parameter is tested on a sample basis for the TLC2654A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

2
Operational Amplifiers

TLC2654C, TLC2654AC

Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD} \pm = \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2654AC			TLC2654C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage		25°C	4	10	5	20	μV	
		Full range		24		34		
α_{VIO} Temperature coefficient of input offset voltage		0°C to 70°C	0.004	0.3	0.004	0.3	$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.003	0.02	0.003	0.06	$\mu\text{V}/\text{mo}$	
I_{IO} Input offset current		25°C	30		30		pA	
		Full range		150		150		
I_{IB} Input bias current		25°C	50		50		pA	
	Full range		150		150			
V_{ICR} Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7		-5 to 2.7		V	
V_{OM+} Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$, See Note 5	25°C	4.7	4.8	4.7	4.8	V	
		Full range	4.7		4.7			
V_{OM-} Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$, See Note 5	25°C	-4.7	-4.9	-4.7	-4.9	V	
		Full range	-4.7		-4.7			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	135	155	120	155	dB	
		Full range	130		120			
f_{ch} Internal chopping frequency		25°C	10		10		kHz	
Clamp on-state current	$R_L = 100\ \text{k}\Omega$	25°C	25		25		μA	
		Full range	25		25			
Clamp off-state current	$V_O = -4\ \text{V to } 4\ \text{V}$	25°C		100		100	pA	
		Full range		100		100		
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR}\ \text{min}, R_S = 50\ \Omega$	25°C	110	125	105	125	dB	
		Full range	110		105			
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} \pm = \pm 2.3\ \text{V to } \pm 8\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C	120	125	110	125	dB	
		Full range	120		110			
I_{DD} Supply current	$V_O = 0, \text{ No load}$	25°C	1.5	2.1	1.5	2.1	mA	
		Full range		2.2		2.2		

†Full range is 0°C to 70°C.

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$ extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. Output clamp is not connected.

2

Operational Amplifiers

TLC2654C, TLC2654AC
Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED
OPERATIONAL AMPLIFIERS

operating characteristics at specified free-air temperature, $V_{DD} \pm = \pm 5$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC2654AC			TLC2654C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR + Positive slew rate at unity gain	$V_O = \pm 2.3$ V, $R_L = 10$ k Ω , $C_L = 100$ pF	25°C	1.5	2	1.5	2	V/ μ s	
		Full range	1.3		1.3			
SR – Negative slew rate at unity gain	$V_O = \pm 2.3$ V, $R_L = 10$ k Ω , $C_L = 100$ pF	25°C	2.3	3.7	2.3	3.7	V/ μ s	
		Full range	1.7		1.7			
V_n Equivalent input noise voltage (see Note 6)	$f = 10$ Hz	25°C		47	75	47	nV/ $\sqrt{\text{Hz}}$	
	$f = 1$ kHz	25°C		13	20	13		
V_{NPP} Peak-to-peak equivalent input noise voltage	$f = 0$ to 1 Hz	25°C		0.5		0.5	μ V	
	$f = 0$ to 10 Hz	25°C		1.5		1.5		
I_n Equivalent input noise current	$f = 1$ kHz	25°C		0.004		0.004	pA/ $\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10$ kHz, $R_L = 10$ k Ω , $C_L = 100$ pF	25°C		1.9		1.9	MHz	
ϕ_m Phase margin at unity gain	$R_L = 10$ k Ω , $C_L = 100$ pF	25°C		48°		48°		

†Full range is 0°C to 70°C.

NOTE 6: This parameter is tested on a sample basis for the TLC2654A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

2
Operational Amplifiers

TLC2654, TLC2654A
Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

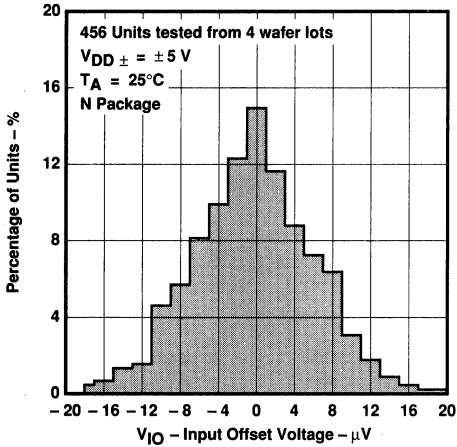
table of graphs

2
Operational Amplifiers

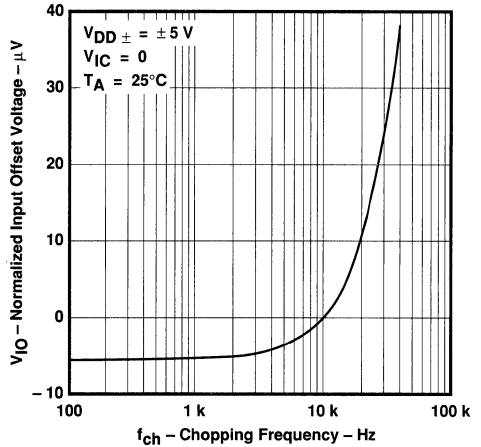
				FIGURE
V_{IO}	Input offset voltage	Distribution		1
	Normalized input offset voltage	vs Chopping frequency		2
I_{IO}	Input offset current	vs Chopping frequency		3
		vs Temperature		4
I_B	Input bias current	vs Common-mode voltage		5
		vs Chopping frequency		6
		vs Temperature		7
	Clamp current	vs Output voltage		8
V_{OM}	Maximum peak output voltage swing	vs Output current		9
		vs Temperature		10
V_{OPP}	Maximum peak-to-peak output voltage swing	vs Frequency		11
CMRR	Common-mode rejection ratio	vs Frequency		12
A_{VD}	Differential voltage amplification	vs Frequency		13
		vs Temperature		14
f_{ch}	Chopping frequency	vs Supply voltage		15
		vs Temperature		16
I_{DD}	Supply current	vs Supply voltage		17
		vs Temperature		18
I_{OS}	Short-circuit output current	vs Supply voltage		19
		vs Temperature		20
SR	Slew rate	vs Supply voltage		21
		vs Temperature		22
	Pulse response	Small-signal		23
		Large-signal		24
V_{NPP}	Peak-to-peak equivalent input noise voltage	vs Chopping frequency		25, 26
V_n	Equivalent input noise voltage	vs Frequency		27
k_{SVR}	Supply-voltage rejection ratio	vs Frequency		28
	Gain-bandwidth product	vs Supply voltage		29
		vs Temperature		30
	Phase margin	vs Supply voltage		31
		vs Temperature		32
ϕ_m	Phase shift	vs Frequency		13

TYPICAL CHARACTERISTICS†

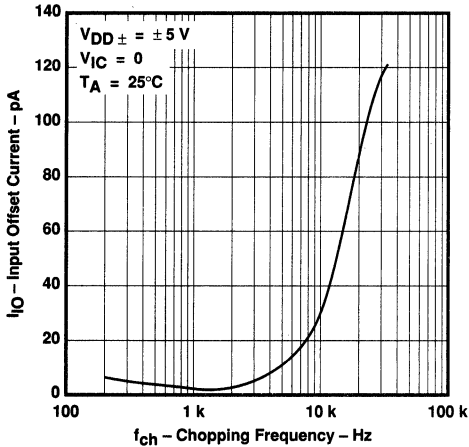
DISTRIBUTION OF TLC2654
INPUT OFFSET VOLTAGE



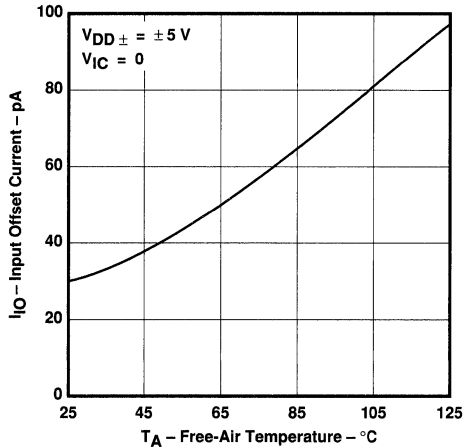
NORMALIZED INPUT OFFSET VOLTAGE
VS
CHOPPING FREQUENCY



INPUT OFFSET CURRENT
VS
CHOPPING FREQUENCY



INPUT OFFSET CURRENT
VS
FREE-AIR TEMPERATURE



†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC2654, TLC2654A

Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

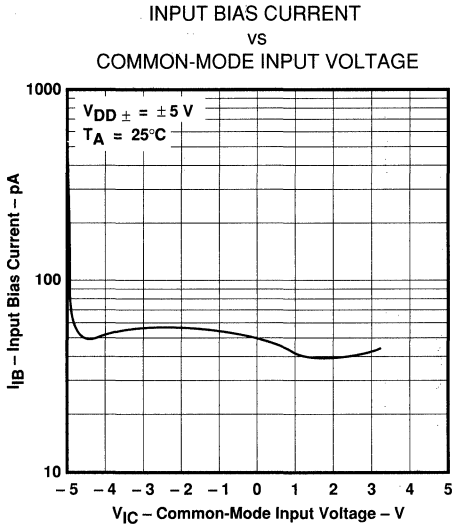


FIGURE 5

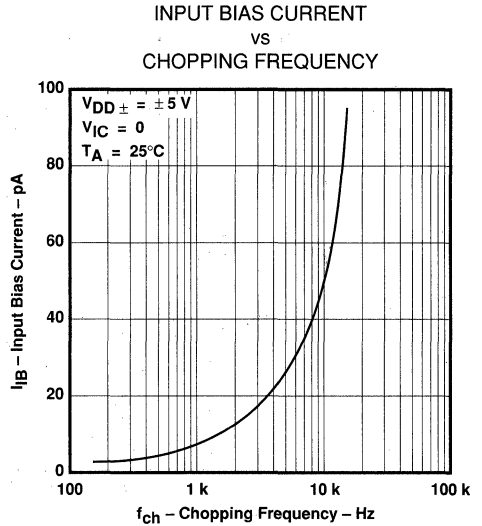


FIGURE 6

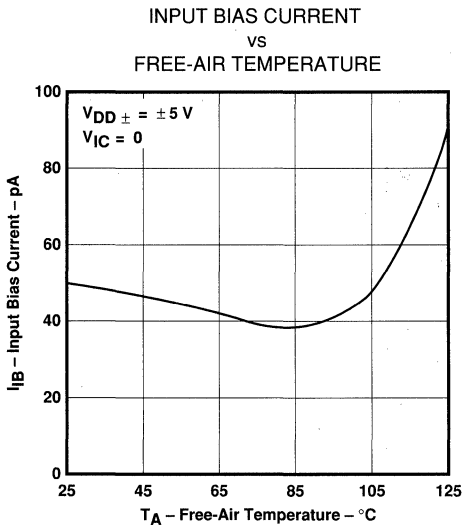


FIGURE 7

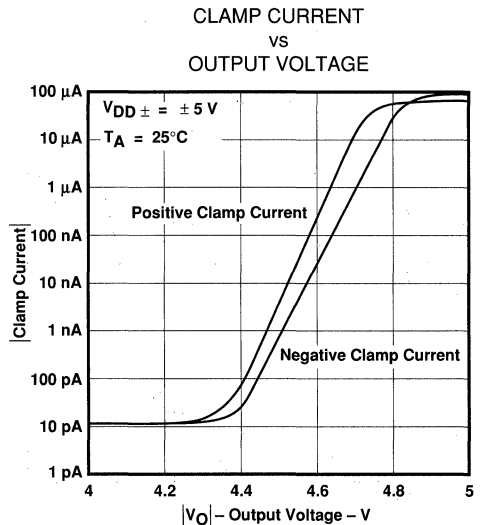


FIGURE 8

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC2654, TLC2654A
Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 OUTPUT CURRENT

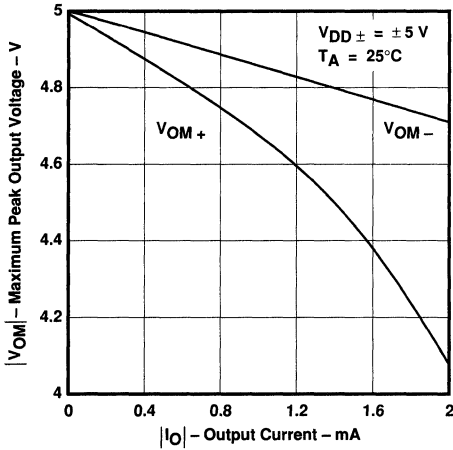


FIGURE 9

MAXIMUM PEAK OUTPUT VOLTAGE
 VS
 FREE-AIR TEMPERATURE

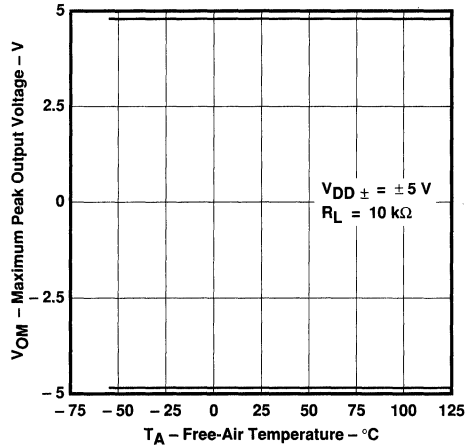


FIGURE 10

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 VS
 FREQUENCY

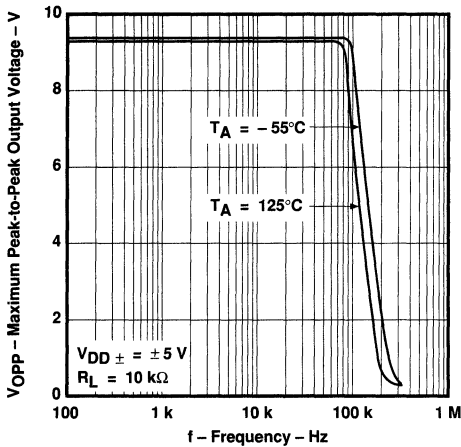


FIGURE 11

COMMON-MODE REJECTION RATIO
 VS
 FREQUENCY

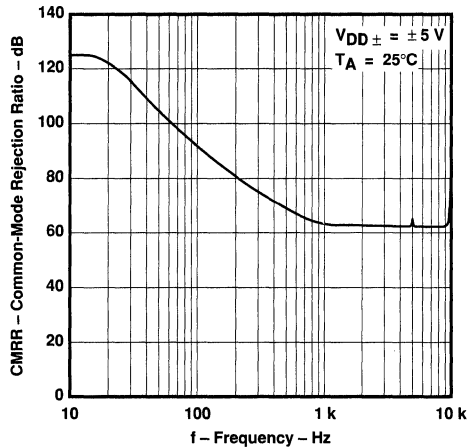


FIGURE 12

2
Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC2654, TLC2654A
Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT VS FREQUENCY

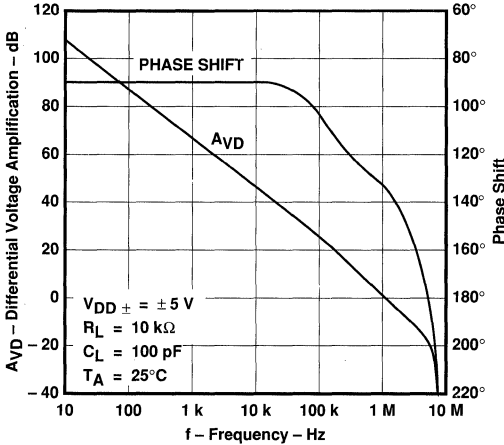


FIGURE 13

LARGE-SIGNAL VOLTAGE AMPLIFICATION VS FREE-AIR TEMPERATURE

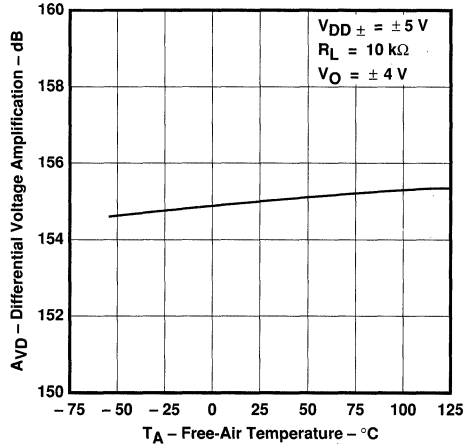


FIGURE 14

CHOPPING FREQUENCY VS SUPPLY VOLTAGE

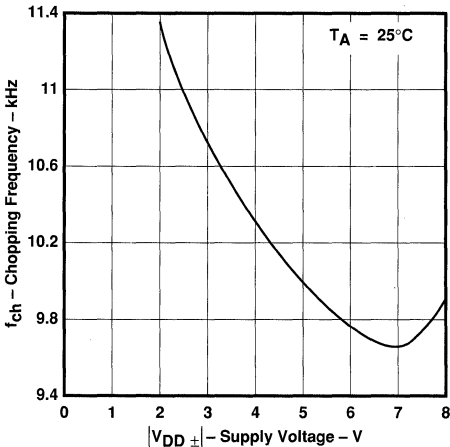


FIGURE 15

CHOPPING FREQUENCY VS FREE-AIR TEMPERATURE

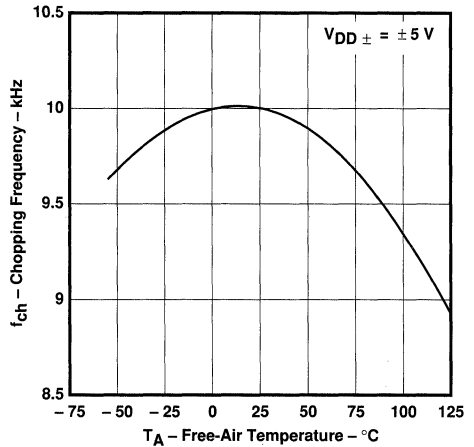


FIGURE 16

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

2
Operational Amplifiers

TLC2654, TLC2654A
Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

SUPPLY CURRENT
VS
SUPPLY VOLTAGE

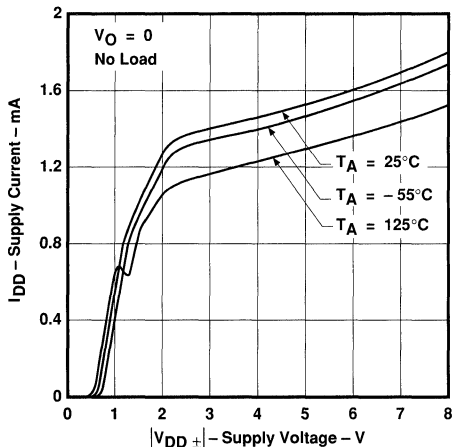


FIGURE 17

SUPPLY CURRENT
VS
FREE-AIR TEMPERATURE

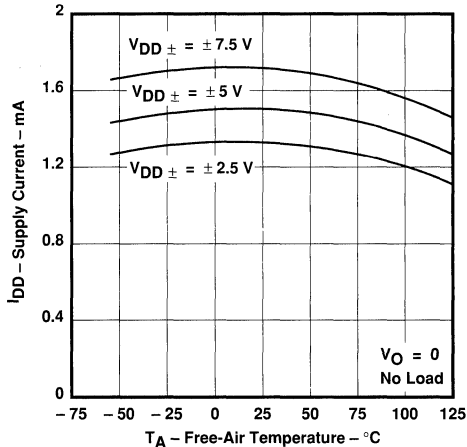


FIGURE 18

SHORT-CIRCUIT OUTPUT CURRENT
VS
SUPPLY VOLTAGE

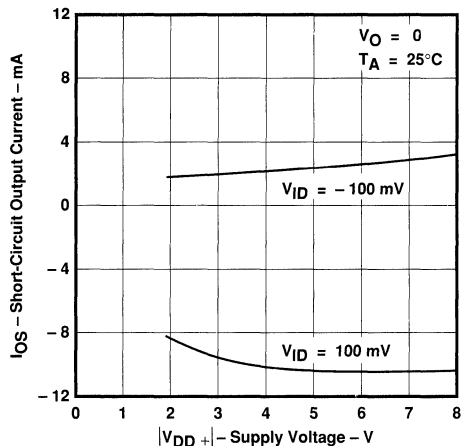


FIGURE 19

SHORT-CIRCUIT OUTPUT CURRENT
VS
FREE-AIR TEMPERATURE

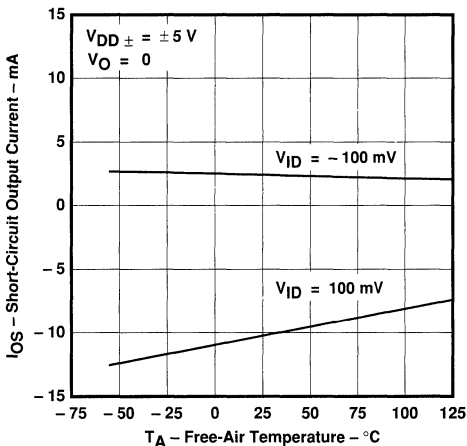


FIGURE 20

2
Operational Amplifiers

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC2654, TLC2654A
Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

SLEW RATE
VS
SUPPLY VOLTAGE

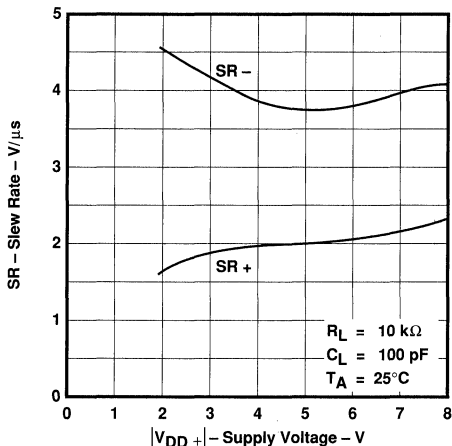


FIGURE 21

SLEW RATE
VS
FREE-AIR TEMPERATURE

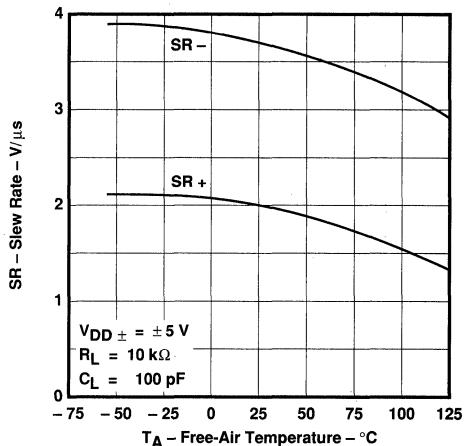


FIGURE 22

VOLTAGE-FOLLOWER
SMALL-SIGNAL
PULSE RESPONSE

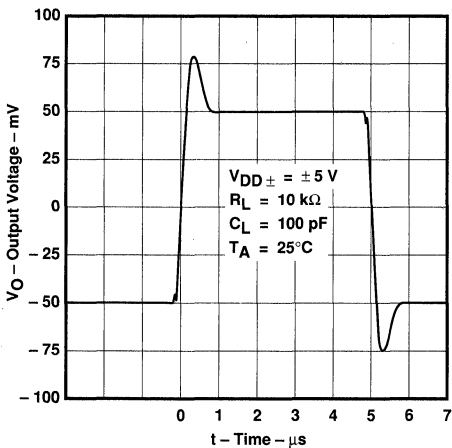


FIGURE 23

VOLTAGE-FOLLOWER
LARGE-SIGNAL
PULSE RESPONSE

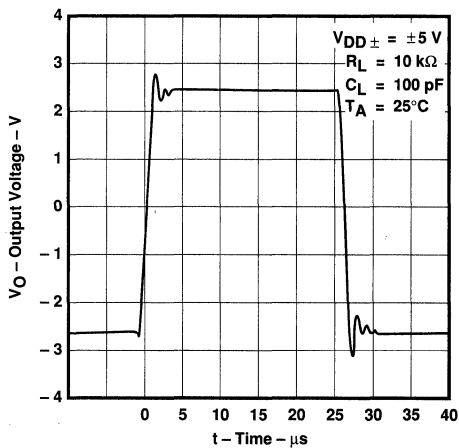


FIGURE 24

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

2
Operational Amplifiers

TLC2654, TLC2654A
Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

PEAK-TO-PEAK INPUT NOISE VOLTAGE
 VS
 CHOPPING FREQUENCY

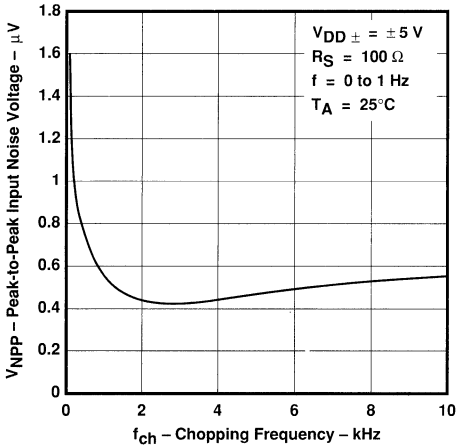


FIGURE 25

PEAK-TO-PEAK INPUT NOISE VOLTAGE
 VS
 CHOPPING FREQUENCY

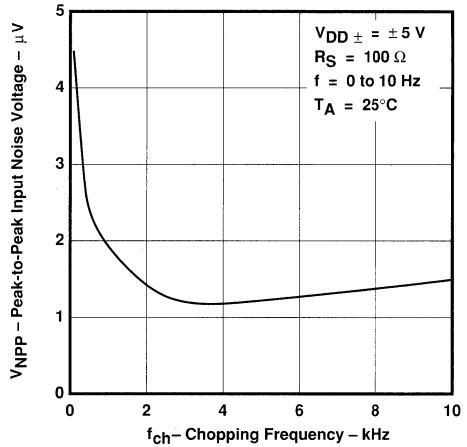


FIGURE 26

EQUIVALENT INPUT NOISE VOLTAGE
 VS
 FREQUENCY

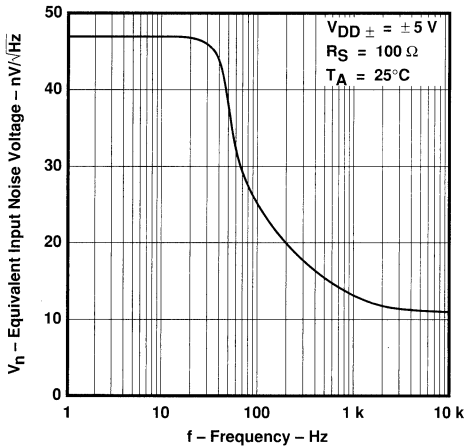


FIGURE 27

SUPPLY-VOLTAGE REJECTION RATIO
 VS
 FREQUENCY

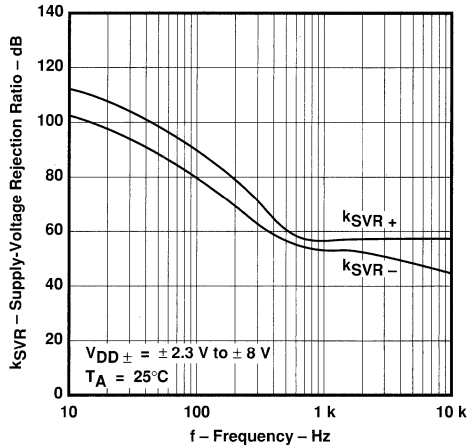


FIGURE 28

2
Operational Amplifiers

TLC2654, TLC2654A
Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED
OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS†

**GAIN-BANDWIDTH PRODUCT
 VS
 SUPPLY VOLTAGE**

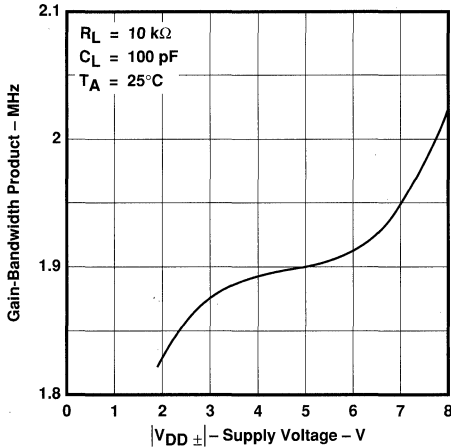


FIGURE 29

**GAIN-BANDWIDTH PRODUCT
 VS
 FREE-AIR TEMPERATURE**

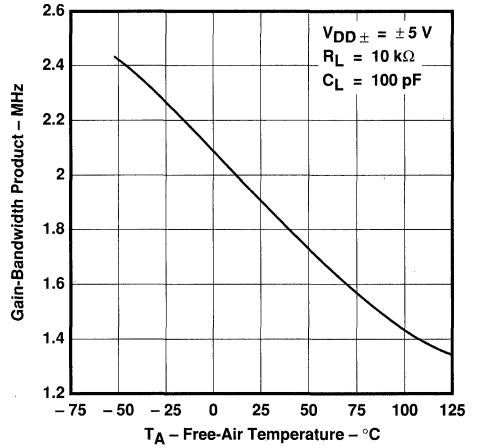


FIGURE 30

**PHASE MARGIN
 VS
 SUPPLY VOLTAGE**

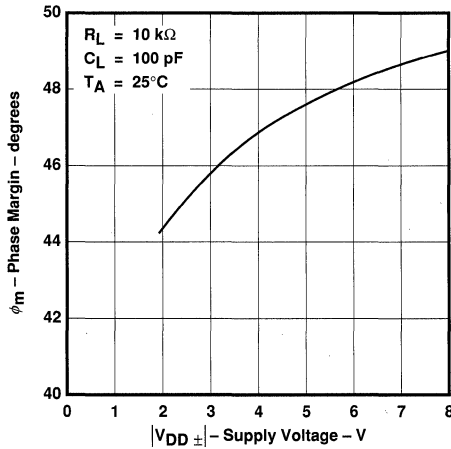


FIGURE 31

**PHASE MARGIN
 VS
 LOAD CAPACITANCE**

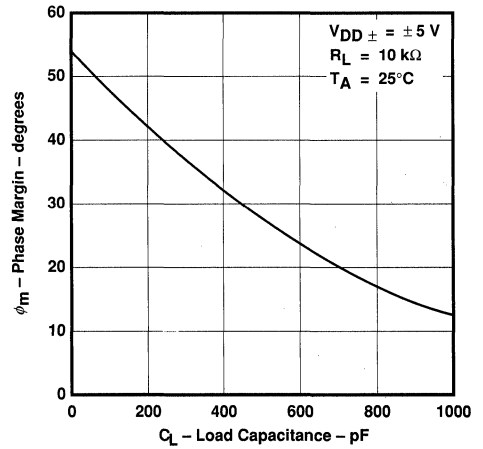


FIGURE 32

†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL APPLICATION DATA

capacitor selection and placement

Leakage and dielectric absorption are the two important factors to consider when selecting external capacitors C_{XA} and C_{XB} . Both factors can cause system degradation negating the performance advantages realized by using the TLC2654.

Degradation from capacitor leakage becomes more apparent with increasing temperatures. Low-leakage capacitors and standoffs are recommended for operation at $T_A = 125^\circ\text{C}$. In addition, guardbands around the capacitor connections on both sides of the printed circuit board are recommended to alleviate problems caused by surface leakage on circuit boards.

Capacitors with high dielectric absorption tend to take several seconds to settle upon application of power, which directly affects input offset voltage. In applications needing fast settling of input offset voltage, it is recommended that high-quality film capacitors, such as mylar, polystyrene, or polypropylene, be used. In other applications, however, a ceramic or other low-grade capacitor may suffice.

Unlike many choppers available today, the TLC2654 is designed to function with values of C_{XA} and C_{XB} in the range of $0.1\ \mu\text{F}$ to $1\ \mu\text{F}$ without degradation to input offset voltage or input noise voltage. These capacitors should be located as close as possible to the C_{XA} and C_{XB} pins and returned to either the V_{DD-} pin or the C RETURN pin. Note that on many choppers, connecting these capacitors to the V_{DD-} pin causes degradation in noise performance, a problem that is eliminated on the TLC2654.

internal/external clock

The TLC2654 has an internal clock that sets the chopping frequency to a nominal value of 10 kHz. On 8-pin packages, the chopping frequency can only be controlled by the internal clock; however, on all 14-pin packages and the 20-pin FK package, the device chopping frequency may be set by the internal clock or controlled externally by use of the INT/EXT and CLK IN pins. To use the internal 10-kHz clock, no connection is necessary. If external clocking is desired, connect the INT/EXT pin to V_{DD-} and the external clock to CLK IN. The external clock trip point is 2.5 V above the negative rail; however, the CLK IN pin may be driven from the negative rail to 5 V above the negative rail. This allows the TLC2654 to be driven directly by 5-V TTL and CMOS logic when operating in the single-supply configuration. If this 5-V level is exceeded, damage could occur to the device unless the current into the CLK IN pin is limited to $\pm 5\ \text{mA}$. A divide-by-two frequency divider interfaces with the CLK IN pin and sets the chopping frequency. The chopping frequency appears on the CLK OUT pin.

overload recovery/output clamp

When large differential input voltage conditions are applied to the TLC2654, the nulling loop attempts to prevent the output from saturating by driving C_{XA} and C_{XB} to internally-clamped voltage levels. Once the overdrive condition is removed, a period of time is required to allow the built-up charge to dissipate. This time period is defined as overload recovery time (see Figure 33). Typical overload recovery time for the TLC2654 is significantly faster than competitive products; however, if required, this time can be reduced further by use of internal clamp circuitry accessible through the CLAMP pin.

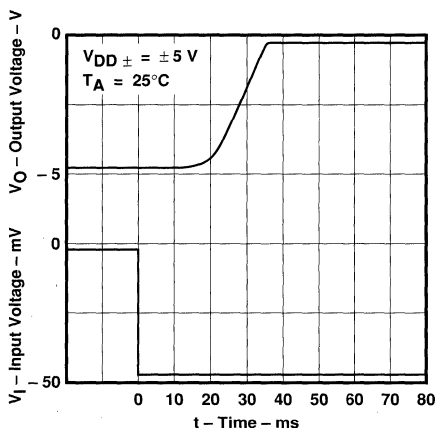


FIGURE 33. OVERLOAD RECOVERY

TLC2654, TLC2654A

Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

TYPICAL APPLICATION DATA

The clamp is simply a switch that is automatically activated when the output is approximately 1 V from either supply rail. When connected to the inverting input (in parallel with the closed-loop feedback resistor), the closed-loop gain is reduced and the TLC2654 output is prevented from going into saturation. Since the output must source or sink current through the switch (see Figure 8), the maximum output voltage swing is slightly reduced.

thermoelectric effects

To take advantage of the extremely low offset voltage temperature coefficient of the TLC2654, care must be taken to compensate for the thermoelectric effects present when two dissimilar metals are brought into contact with each other (such as device leads being soldered to a printed circuit board). It is not uncommon for dissimilar metal junctions to produce thermoelectric voltages in the range of several microvolts per degree Celsius (orders of magnitude greater than the 0.01- $\mu\text{V}/^\circ\text{C}$ typical of the TLC2654).

To help minimize thermoelectric effects, careful attention should be paid to component selection and circuit board layout. Avoid the use of nonsoldered connections (such as sockets, relays, switches, etc.) in the input signal path. Cancel thermoelectric effects by duplicating the number of components and junctions in each device input. The use of low-thermoelectric-coefficient components, such as wire-wound resistors, is also beneficial.

latchup avoidance

Because CMOS devices are susceptible to latchup due to their inherent parasitic thyristors, the TLC2654 inputs and output are designed to withstand -100-mA surge currents without sustaining latchup; however, techniques to reduce the chance of latchup should be used whenever possible. Internal protection diodes should not be forward biased in normal operation. Applied input and output voltages should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by using decoupling capacitors (0.1 μF typical) located across the supply rails as close to the device as possible.

The current path established if latchup occurs is usually between the supply rails and is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor. The chance of latchup occurring increases with increasing temperature and supply voltage.

electrostatic discharge protection

The TLC2654 incorporates internal ESD protection circuits that prevent functional failures at voltages at or below 2000 V. Care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.

theory of operation

Chopper-stabilized operational amplifiers offer the best dc performance of any monolithic operational amplifier. This superior performance is the result of using two operational amplifiers – a main amplifier and a nulling amplifier – plus oscillator-controlled logic and two external capacitors to create a system that behaves as a single amplifier. With this approach, the TLC2654 achieves submicrovolt input offset voltage, submicrovolt noise voltage, and offset voltage variations with temperature in the $\text{nV}/^\circ\text{C}$ range.

The TLC2654 on-chip control logic produces two dominant clock phases – a nulling phase and an amplifying phase. The term “chopper-stabilized” derives from the process of switching between these two clock phases. Figure 34 shows a simplified block diagram of the TLC2654. Switches A and B are make-before-break types. During the nulling phase, switch A is closed, shorting the nulling amplifier inputs together and allowing the nulling amplifier to reduce its own input offset voltage by feeding its output signal back to an inverting input

2

Operational Amplifiers

μA709AM, μA709M, μA709C GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

D942, FEBRUARY 1971 REVISED MAY 1988

- Common-Mode Input Range . . . ± 10 V Typical
- Designed to be Interchangeable with Fairchild μ A709A, μ A709, and μ A709C
- Maximum Peak-to-Peak Output Voltage Swing . . . 28-V Typical with 15-V Supplies

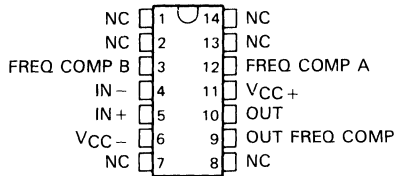
description

These circuits are general-purpose operational amplifiers, each having high-impedance differential inputs and a low-impedance output. Component matching, inherent with silicon monolithic circuit-fabrication techniques, produces an amplifier with low-drift and low-offset characteristics. Provisions are incorporated within the circuit whereby external components may be used to compensate the amplifier for stable operation under various feedback or load conditions. These amplifiers are particularly useful for applications requiring transfer or generation of linear or nonlinear functions.

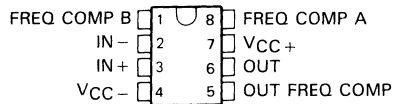
The μ A709A circuit features improved offset characteristics, reduced input-current requirements, and lower power dissipation when compared to the μ A709 circuit. In addition, maximum values of the average temperature coefficients of offset voltage and current are specified for the μ A709A.

The μ A709AM and μ A709M are characterized for operation over the full military temperature range of -55°C to 125°C . The μ A709C is characterized for operation from 0°C to 70°C .

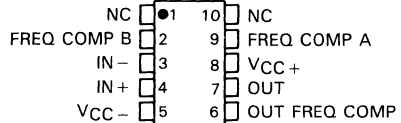
μA709AM, μA709M . . . J OR W PACKAGE
(TOP VIEW)



μA709AM, μA709M . . . JG PACKAGE
μA709C . . . D, JG, OR P PACKAGE
(TOP VIEW)

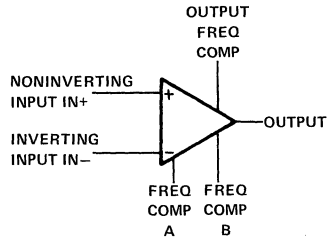


μA709AM, μA709M . . . U FLAT PACKAGE
(TOP VIEW)



NC—No internal connection

symbol



AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE					
		SMALL OUTLINE (D)	CERAMIC (J)	CERAMIC DIP (JG)	PLASTIC DIP (P)	FLAT PACK (U)	FLAT PACK (W)
0°C to 70°C	7.5 mV	μA709CD	—	μA709CJG	μA709CP	—	—
-55°C to 125°C	5 mV	—	μA709MJ	μA709MJG	—	μA709MU	μA709MW
	2 mV	—	μA709AMJ	μA709AMJG	—	μA709AMU	μA709AMW

The D package is available taped and reeled. Add the suffix R to the device type when ordering, (e.g., μ A709CDR)

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

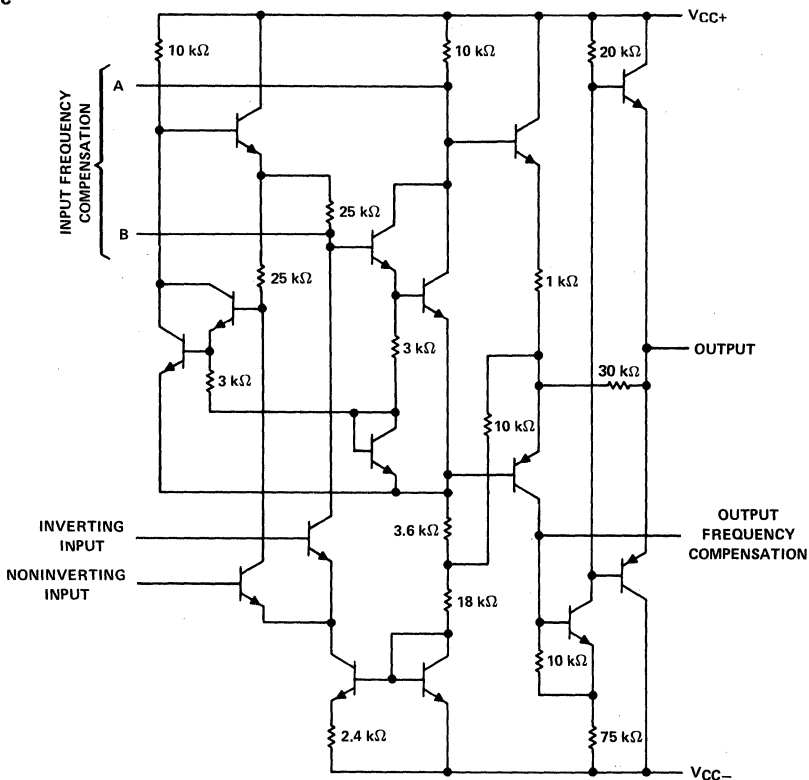
TEXAS INSTRUMENTS

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μA709AM, μA709M, μA709C GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

schematic



Component values shown are nominal.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	μA709AM μA709M	μA709C	UNIT
Supply voltage V_{CC+} (see Note 1)	18	18	V
Supply voltage V_{CC-} (see Note 1)	-18	-18	V
Differential input voltage (see Note 2)	±5	±5	V
Input voltage (either input, see Notes 1 and 3)	±10	±10	V
Duration of output short-circuit (see Note 4)	5	5	s
Continuous total dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J, JG, U, or W package	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package	260	°C

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 10 V, whichever is less.
 4. The output may be shorted to ground or either power supply.

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Operational Amplifiers

µA709AM, µA709M, µA709C GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING			POWER RATING	POWER RATING
D	300 mW	N/A	N/A	300 mW	N/A
J (µA709_M)	300 mW	11.0 mW/°C	123°C	300 mW	275 mW
JG (µA709_M)	300 mW	8.4 mW/°C	114°C	300 mW	210 mW
JG (µA709C)	300 mW	N/A	N/A	300 mW	N/A
P	300 mW	N/A	N/A	300 mW	N/A
U	300 mW	5.4 mW/°C	94°C	300 mW	135 mW
W	300 mW	8.0 mW/°C	113°C	300 mW	200 mW

electrical characteristics at specified free-air temperature, $V_{CC} \pm = \pm 9\text{ V to } \pm 15\text{ V}$ (unless otherwise noted)

2

Operational Amplifiers

PARAMETER	TEST CONDITIONS [†]		µA709AM			µA709M			UNIT
			MIN	TYP [‡]	MAX	MIN	TYP [‡]	MAX	
V_{IO} Input offset voltage	$V_O = 0$	$R_S \leq 10\text{ k}\Omega$	25°C	0.6	2	1	5	mV	
			Full range		3		6		
α_{VIO} Average temperature coefficient of input offset voltage	$V_O = 0$	$R_S = 50\ \Omega$	Full range	1.8	10	3		$\mu\text{V}/^\circ\text{C}$	
			$R_S = 10\text{ k}\Omega$	Full range	4.8	25	6		
I_{IO} Input offset current	$V_O = 0$		25°C	10	50	50	200	nA	
			-55°C	40	250	100	500		
			125°C	3.5	50	20	200		
α_{IIO} Average temperature coefficient of input offset current	$V_O = 0$		-55°C to 25°C	0.45	2.8			nA/°C	
			25°C to 125°C	0.08	0.5				
I_{IB} Input bias current	$V_O = 0$		25°C	0.1	0.2	0.2	0.5	μA	
			-55°C	0.3	0.6	0.5	1.5		
V_{ICR} Common-mode input voltage range	$V_{CC} \pm = \pm 15\text{ V}$		25°C	± 8	± 10	± 8	± 10	V	
			Full range	± 8		± 8			
V_{OPP} Maximum peak-to-peak output voltage swing	$V_{CC} \pm = \pm 15\text{ V}, R_L \geq 10\text{ k}\Omega$		25°C	24	28	24	28	V	
			Full range	24		24			
			$V_{CC} \pm = \pm 15\text{ V}, R_L = 2\text{ k}\Omega$	25°C	20	26	20		26
V_{DD} Large-signal differential voltage amplification	$V_{CC} \pm = \pm 15\text{ V}, R_L \geq 2\text{ k}\Omega, V_O = \pm 10\text{ V}$		25°C	45		45		V/mV	
			Full range	25	70	25	70		
r_i Input resistance			25°C	350	750	150	400	k Ω	
			-55°C	85	185	40	100		
r_o Output resistance	$V_O = 0$, See Note 5		25°C	150		150		Ω	
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$		25°C	80	110	70	90	dB	
			Full range	80		70			
k_{SVS} Power supply sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	$V_{CC} = \pm 9\text{ V to } \pm 15\text{ V}$		25°C	40	100	25	150	$\mu\text{V}/\text{V}$	
			Full range		100		150		
I_{CC} Supply current	$V_{CC} \pm = \pm 15\text{ V}, \text{ No load}, V_O = 0$		25°C	2.5	3.6	2.6	5.5	mA	
			-55°C	2.7	4.5				
			125°C	2.1	3				
P_D Total power dissipation	$V_{CC} \pm = \pm 15\text{ V}, \text{ No load}, V_O = 0$		25°C	75	108	78	165	mW	
			-55°C	81	135				
			125°C	63	90				

[†]All characteristics are specified under open-loop with zero common-mode input voltage unless otherwise specified. Full range for µA709AM and µA709M is -55°C to 125°C.

[‡]All typical values are at $V_{CC} \pm = \pm 15\text{ V}$.

NOTE 5: This typical value applies only at frequencies above a few hundred hertz because of the effects of drift and thermal feedback.

uA709AM, uA709M, uA709C
GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature (unless otherwise noted $V_{CC} \pm = \pm 15 \text{ V}$)

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Operational Amplifiers

PARAMETER	TEST CONDITIONS†	uA709C			UNIT	
		MIN	TYP	MAX		
V_{IO} Input offset voltage	$V_{CC} \pm = \pm 9 \text{ V to } \pm 15 \text{ V}, V_O = 0$	25°C	2	7.5	mV	
		Full range		10		
I_{IO} Input offset current	$V_{CC} \pm = \pm 9 \text{ V to } \pm 15 \text{ V}, V_O = 0$	25°C	100	500	nA	
		Full range		750		
I_{IB} Input bias current	$V_{CC} \pm = \pm 9 \text{ V to } \pm 15 \text{ V}, V_O = 0$	25°C	0.3	1.5	μA	
		Full range		2		
V_{ICR} Common-mode input voltage range		25°C	±8	±10	V	
V_{OPP} Maximum peak-to-peak output voltage swing	$R_L \geq 10 \text{ k}\Omega$	25°C	24	28	V	
		Full range	24			
		25°C	$R_L = 2 \text{ k}\Omega$	20		26
			$R_L \geq 2 \text{ k}\Omega$	20		
AVD Large-signal differential voltage amplification	$R_L \leq 2 \text{ k}\Omega, V_O = \pm 10 \text{ V}$	25°C	15	45	V/mV	
		Full range	12			
r_i Input resistance		25°C	50	250	kΩ	
		Full range	35			
r_o Output resistance	$V_O = 0,$ See Note 5	25°C		150	Ω	
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR} \text{ min}$	25°C	65	90	dB	
k_{SVS} Supply voltage sensitivity	$V_{CC} = \pm 9 \text{ V to } \pm 15 \text{ V}$	25°C	25	200	μV/V	
P_D Total power dissipation	$V_O = 0,$ No load	25°C	80	200	mW	

† All characteristics are specified under open-loop operation with zero volts common-mode voltage unless otherwise specified. Full range for uA709C is 0°C to 70°C.

NOTE 5: This typical value applies only at frequencies above a few hundred hertz because of the effects of drift and thermal feedback.

operating characteristics $V_{CC} \pm = \pm 9 \text{ V to } \pm 15 \text{ V}, T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	uA709AM uA709M uA709C			UNIT
		MIN	TYP	MAX	
t_r Rise time	$V_I = 20 \text{ mV}, R_L = 2 \text{ k}\Omega,$ See Figure 1	$C_L = 0$	0.3	1	μs
Overshoot factor		$C_L = 100 \text{ pF}$	6%	30%	

PARAMETER MEASUREMENT INFORMATION

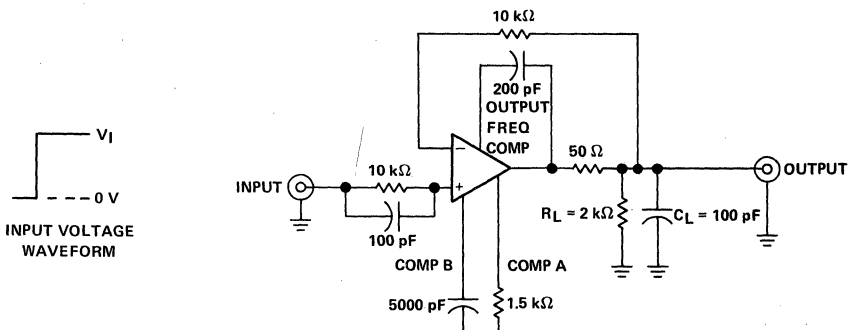
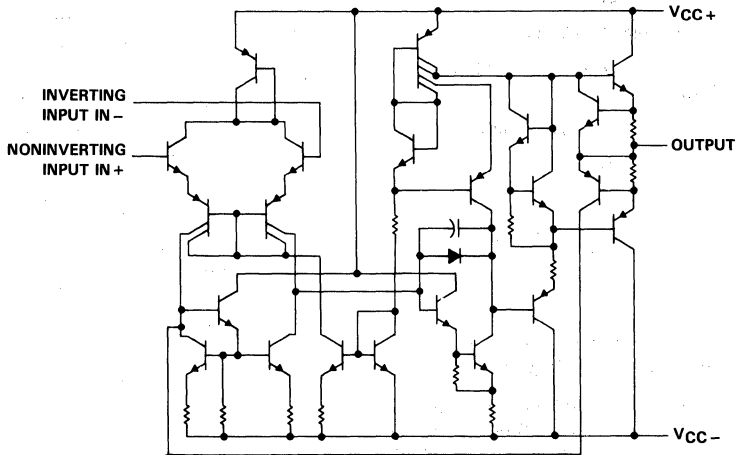


FIGURE 1. RISE TIME AND SLEW RATE

μA741M, μA741C GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

schematic



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	μA741M	μA741C	UNIT
Supply voltage V_{CC+} (see Note 1)	22	18	V
Supply voltage V_{CC-} (see Note 1)	-22	-18	V
Differential input voltage (see Note 2)	±30	±30	V
Input voltage any input (see Notes 1 and 3)	±15	±15	V
Voltage between either offset null terminal (N1/N2) and V_{CC-}	±0.5	±0.5	V
Duration of output short-circuit (see Note 4)	unlimited	unlimited	
Continuous total power dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Case temperature for 60 seconds	FK package		
	260		°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J, JG, or U package		
	300	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package		
		260	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	L package		
	300		°C

- NOTES:
1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V whichever is less.
 4. The output may be shorted to ground or either power supply. For the μA741M only, the unlimited duration of the short-circuit applies at (or below) 125°C case temperature or 75°C free-air temperature.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	DERATING ABOVE T_A	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING			POWER RATING	POWER RATING
D	500 mW	5.8 mW/°C	64°C	464 mW	N/A
FK	500 mW	11.0 mW/°C	105°C	500 mW	275 mW
J (μA741M)	500 mW	11.0 mW/°C	105°C	500 mW	275 mW
JG (μA741M)	500 mW	8.4 mW/°C	90°C	500 mW	210 mW
JG (all others)	500 mW	N/A	N/A	500 mW	N/A
L	500 mW	6.7 mW/°C	75°C	500 mW	167 mW
P	500 mW	N/A	N/A	500 mW	N/A
U	500 mW	5.4 mW/°C	57°C	432 mW	135 mW

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Operational Amplifiers

uA741M, uA741C
GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$

PARAMETER		TEST CONDITIONS†		uA741M			uA741C			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO}	Input offset voltage	$V_O = 0$	25°C	1		5	1		6	mV
			Full range			6	7.5			
$\Delta V_{IO(adj)}$	Offset voltage adjust range	$V_O = 0$	25°C	± 15			± 15			mV
I_{IO}	Input offset current	$V_O = 0$	25°C	20	200	20		200	nA	
			Full range	500		300				
I_{IB}	Input bias current	$V_O = 0$	25°C	80	500	80		500	nA	
			Full range	1500		800				
V_{ICR}	Common-mode input voltage range		25°C	± 12	± 13	± 12		± 13	V	
			Full range	± 12		± 12				
V_{OM}	Maximum peak output voltage swing	$R_L = 10\text{ k}\Omega$	25°C	± 12	± 14	± 12		± 14	V	
			Full range	± 12		± 12				
			25°C	± 10	± 13	± 10		± 13		
			Full range	± 10		± 10				
A_{VD}	Large-signal differential voltage amplification	$R_L \geq 2\text{ k}\Omega$	25°C	50	200	20		200	V/mV	
			Full range	25		15				
r_i	Input resistance	$V_O = 0$	25°C	0.3	2	0.3		2	MΩ	
r_o	Output resistance	$V_O = 0$, See Note 5	25°C	75		75			Ω	
C_i	Input capacitance		25°C	1.4		1.4			pF	
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$	25°C	70	90	70		90	dB	
			Full range	70		70				
k_{SVS}	Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	$V_{CC} = \pm 9\text{ V}$ to $\pm 15\text{ V}$	25°C	30	150	30		150	μV/V	
			Full range	150		150				
I_{OS}	Short-circuit output current		25°C	± 25	± 40	± 25		± 40	mA	
I_{CC}	Supply current	No load, $V_O = 0$	25°C	1.7	2.8	1.7		2.8	mA	
			Full range	3.3		3.3				
P_D	Total power dissipation	No load, $V_O = 0$	25°C	50	85	50		85	mW	
			Full range	100		100				

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for uA741M is -55°C to 125°C and for uA741C is 0°C to 70°C.

NOTE 5: This typical value applies only at frequencies above a few hundred hertz because of the effects of drift and thermal feedback.

operating characteristics, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS		uA741M			uA741C			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
t_r	Rise time	$V_I = 20\text{ mV}$, $C_L = 100\text{ pF}$, See Figure 1	$R_L = 2\text{ k}\Omega$, See Figure 1	0.3			0.3			μs
	Overshoot factor			5%			5%			
SR	Slew rate at unity gain	$V_I = 10\text{ V}$, $C_L = 100\text{ pF}$, See Figure 1	$R_L = 2\text{ k}\Omega$, See Figure 1	0.5			0.5			V/μs

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Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

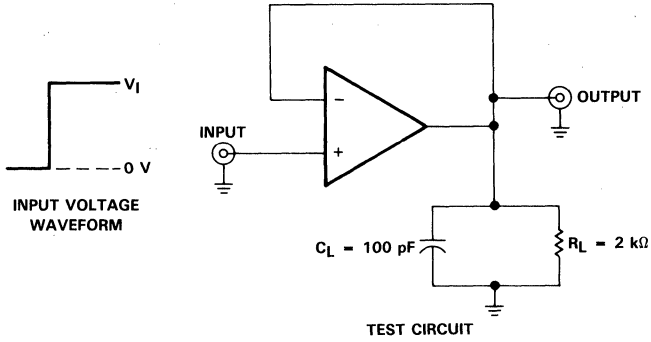


FIGURE 1. RISE TIME, OVERSHOOT, AND SLEW RATE

TYPICAL APPLICATION DATA

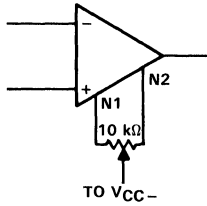


FIGURE 2. INPUT OFFSET VOLTAGE NULL CIRCUIT

TYPICAL CHARACTERISTICS

INPUT OFFSET CURRENT
 vs
FREE-AIR TEMPERATURE

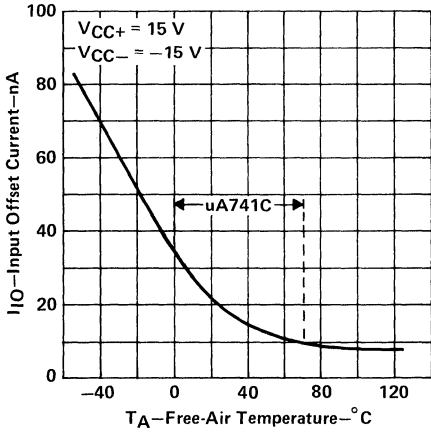


FIGURE 3

INPUT BIAS CURRENT
 vs
FREE-AIR TEMPERATURE

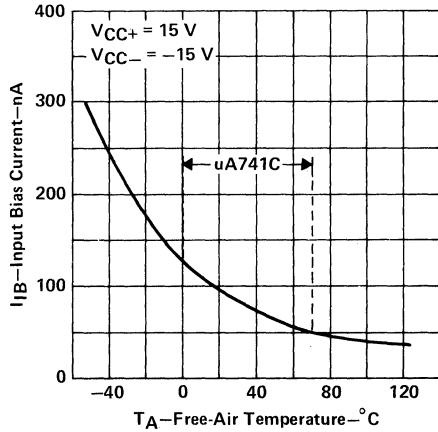


FIGURE 4

MAXIMUM PEAK OUTPUT VOLTAGE
 vs
LOAD RESISTANCE

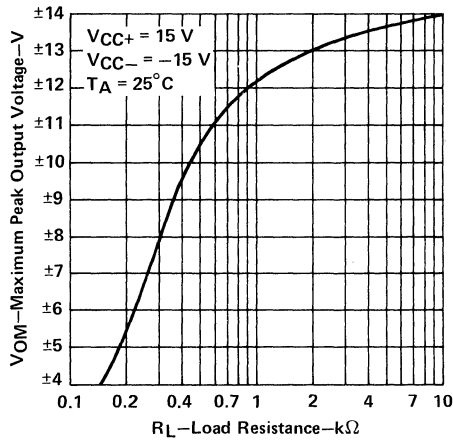


FIGURE 5

TYPICAL CHARACTERISTICS

**MAXIMUM PEAK OUTPUT VOLTAGE
vs
FREQUENCY**

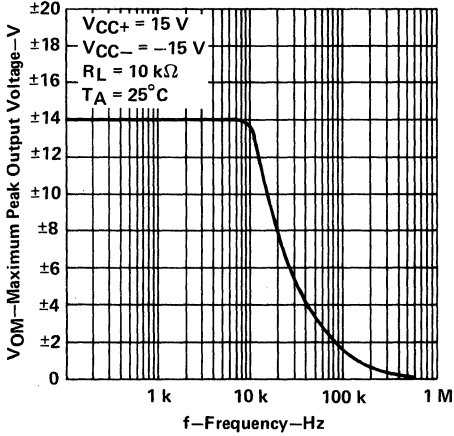


FIGURE 6

**OPEN-LOOP LARGE-SIGNAL DIFFERENTIAL
VOLTAGE AMPLIFICATION
vs
SUPPLY VOLTAGE**

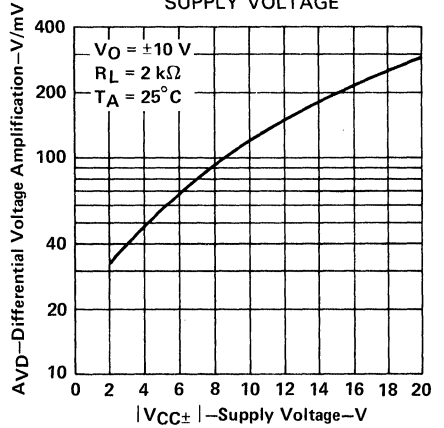


FIGURE 7

**OPEN-LOOP LARGE-SIGNAL DIFFERENTIAL
VOLTAGE AMPLIFICATION
vs
FREQUENCY**

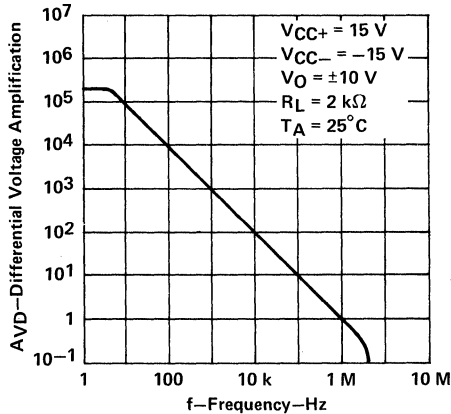


FIGURE 8

TYPICAL CHARACTERISTICS

COMMON-MODE REJECTION RATIO
vs
FREQUENCY

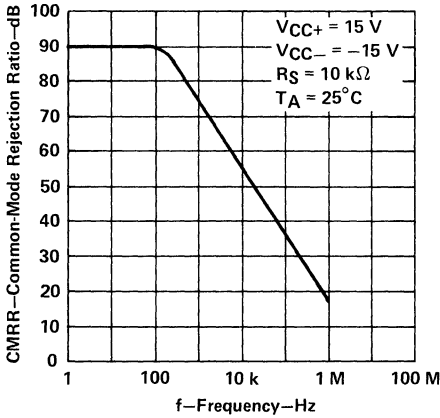


FIGURE 9

OUTPUT VOLTAGE
vs
ELAPSED TIME

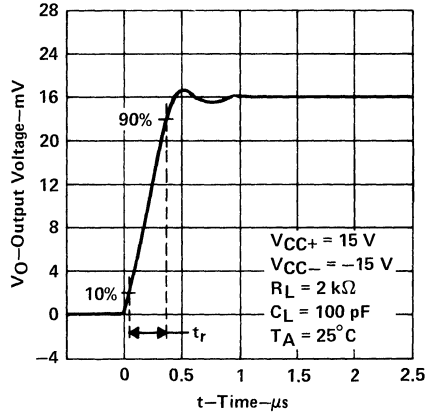


FIGURE 10

VOLTAGE-FOLLOWER
LARGE-SIGNAL PULSE RESPONSE

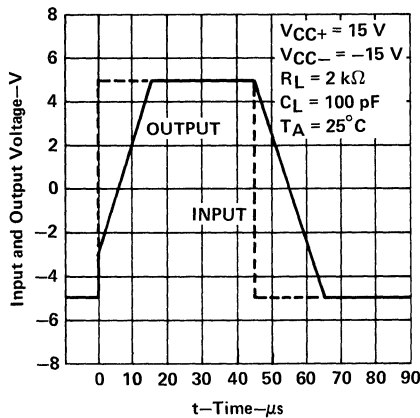


FIGURE 11

2

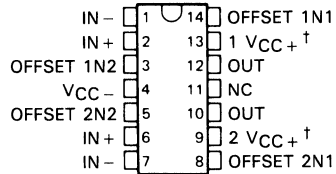
Operational Amplifiers

μA747M, μA747C DUAL GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

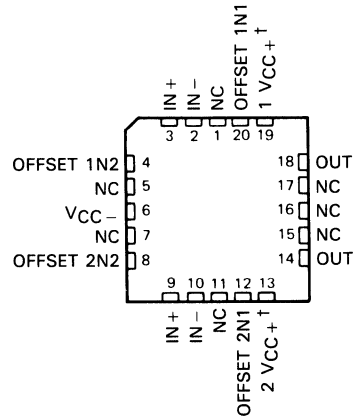
D971, FEBRUARY 1971—REVISED NOVEMBER 1988

- No Frequency Compensation Required
- Low Power Consumption
- Short-Circuit Protection
- Offset-Voltage Null Capability
- Wide Common-Mode and Differential Voltage Ranges
- No Latch-Up
- Designed to be Interchangeable with Fairchild μA747M and μA747C

D, J, OR N PACKAGE
OR W FLAT PACKAGE
(TOP VIEW)



μA747M . . . FK PACKAGE
(TOP VIEW)



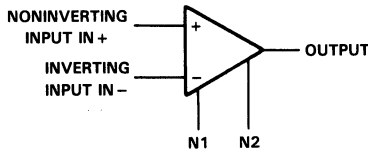
description

The μA747 is a dual general-purpose operational amplifier featuring offset-voltage null capability. Each half is electrically similar to μA741.

The high common-mode input voltage range and the absence of latch-up make this amplifier ideal for voltage-follower applications. The device is short-circuit protected and the internal frequency compensation ensures stability without external components. A low-value potentiometer may be connected between the offset null inputs to null out the offset voltage as shown in Figure 2.

The μA747M is characterized for operation over the full military temperature range of -55°C to 125°C; the μA747C is characterized for operation from 0°C to 70°C.

symbol (each amplifier)



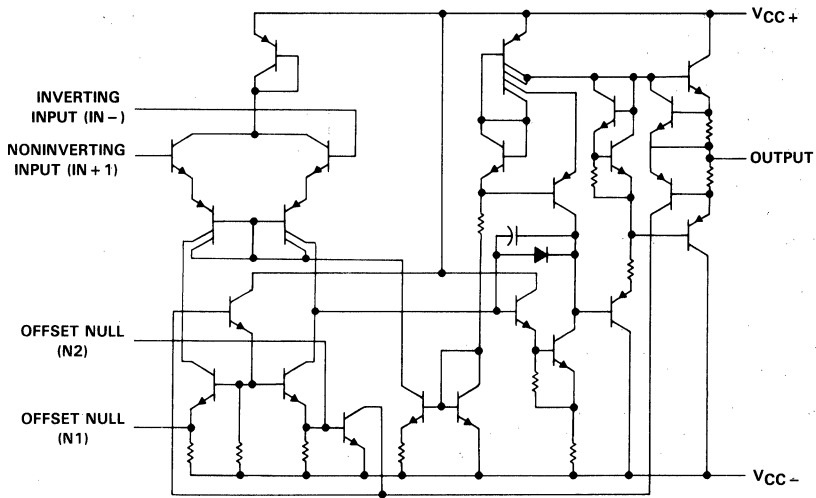
NC—No internal connection

†The two positive supply terminals (1 V_{CC} + and 2 V_{CC} +) are connected together internally.

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Operational Amplifiers

μA747M, μA747C DUAL GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

schematic (each amplifier)



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	μA747M	μA747C	UNIT
Supply voltage, V_{CC+} (see Note 1)	22	18	V
Supply voltage, V_{CC-} (see Note 1)	-22	-18	V
Differential input voltage (see Note 2)	±30	±30	V
Input voltage any input (see Notes 1 and 3)	±15	±15	V
Voltage between any offset null terminal (N1/N2) and V_{CC-}	±0.5	±0.5	V
Duration of output short-circuit (see Note 4)	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Case temperature for 60 seconds	FK package	260	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J or W package	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or N package	260	°C

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 volts, whichever is less.
 4. The output may be shorted to ground or either power supply. For the μA747M only, the unlimited duration of the short-circuit applies at (or below) 125°C case temperature or 75°C free-air temperature.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	800 mW	7.6 mW/°C	45°C	608 mW	—
FK	800 mW	11.0 mW/°C	77°C	800 mW	275 mW
J (μA747M)	800 mW	11.0 mW/°C	77°C	800 mW	275 mW
J (μA747C)	800 mW	8.2 mW/°C	52°C	656 mW	—
N	800 mW	9.2 mW/°C	63°C	736 mW	—
W	800 mW	8.0 mW/°C	50°C	640 mW	—

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Operational Amplifiers

µA747M, µA747C
DUAL GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$

PARAMETER		TEST CONDITIONS†		µA747M			µA747C			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	V _O = 0 V	25 °C	1	5		1	6	mV	
			Full range			6		7.5		
ΔV _{IO(adj)}	Offset voltage adjust range		25 °C	± 15		± 15		mV		
I _{IO}	Input offset current		25 °C	20	200		20	200	nA	
			Full range			500		300		
I _{IB}	Input bias current		25 °C	80	500		80	500	nA	
			Full range			1500		800		
V _{ICR}	Common-mode input voltage range		25 °C	± 12	± 13		± 12	± 13	V	
			Full range			± 12		± 12		
V _{OPP}	Maximum peak-to-peak output voltage swing	R _L = 10 kΩ	25 °C	24	28		24	28	V	
		R _L ≥ 10 kΩ	Full range			24		24		
		R _L = 2 kΩ	25 °C	20	26		20	26		
		R _L ≥ 2 kΩ	Full range			20		20		
A _{VD}	Large-signal differential voltage amplification	R _L ≥ 2 kΩ, V _O = ± 10 V	25 °C	50	200		25	200	V/mV	
			Full range			25		15		
r _i	Input resistance		25 °C	0.3	2		0.3	2	MΩ	
r _o	Output resistance	See Note 6	25 °C	75		75		Ω		
C _i	Input capacitance		25 °C	1.4		1.4		pF		
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICR}	25 °C	70	90		70	90	dB	
			Full range			70		70		
k _{SVS}	Supply voltage sensitivity (ΔV _{IO} /ΔV _{CC})	V _{CC} = ± 9 V to ± 15 V	25 °C	30	150		30	150	µV/V	
			Full range			150		150		
I _{OS}	Short-circuit output current		25 °C	± 25	± 40		± 25	± 40	mA	
I _{CC}	Supply current (each amplifier)	No load	25 °C	1.7	2.8		1.7	2.8	mA	
			Full range			3.3		3.3		
P _D	Power dissipation (each amplifier)	No load V _O = 0 V	25 °C	50	85		50	85	mW	
			Full range			100		100		
V _{O1} /V _{O2}	Channel separation		25 °C	120	0		120	0	dB	

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for µA747M is -55 °C to 125 °C and for µA747C is 0 °C to 70 °C.

NOTE 6: This typical value applies only at frequencies above a few hundred hertz because of the effects of drift and thermal feedback.

operating characteristics, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$, T_A = 25 °C

PARAMETER		TEST CONDITIONS		µA747M			µA747C			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
t _r	Rise time	V _I = 20 mV, C _L = 100 pF, See Figure 1	R _L = 2 kΩ	0.3			0.3			µs
	Overshoot factor			5%			5%			
SR	Slew rate at unity gain	V _I = 10 V, C _L = 100 pF, See Figure 1	R _L = 2 kΩ	0.5			0.5			V/µs

2

Operational Amplifiers

uA747M, uA747C
DUAL GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

PARAMETER MEASUREMENT INFORMATION

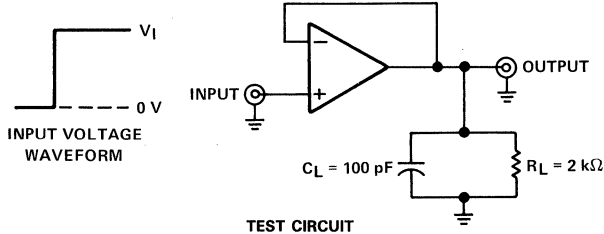


FIGURE 1. RISE TIME, OVERSHOOT, AND SLEW RATE

TYPICAL APPLICATION DATA

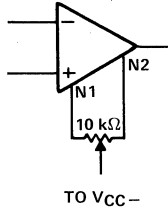


FIGURE 2. INPUT OFFSET VOLTAGE NULL CIRCUIT

TYPICAL CHARACTERISTICS

INPUT OFFSET CURRENT
vs
FREE-AIR TEMPERATURE

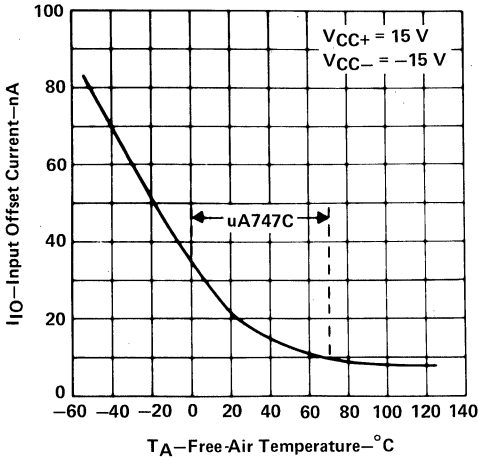


FIGURE 3

INPUT BIAS CURRENT
vs
FREE-AIR TEMPERATURE

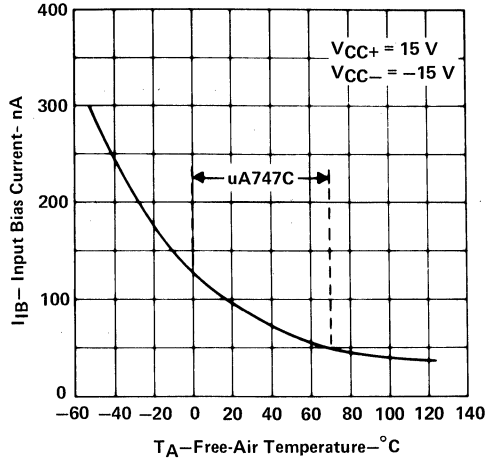


FIGURE 4

2

Operational Amplifiers

TYPICAL CHARACTERISTICS

MAXIMUM PEAK-TO-PEAK
 OUTPUT VOLTAGE
 vs
 LOAD RESISTANCE

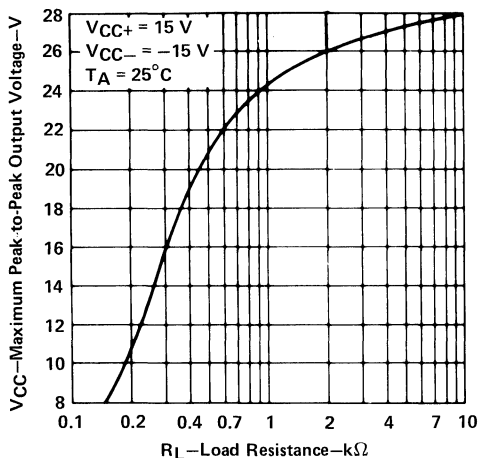


FIGURE 5

MAXIMUM PEAK-TO-PEAK
 OUTPUT VOLTAGE
 vs
 FREQUENCY

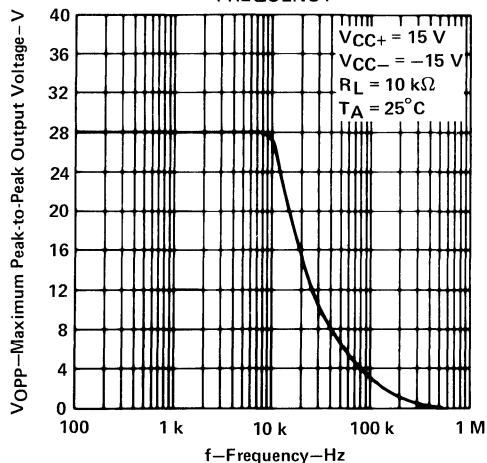


FIGURE 6

OPEN-LOOP LARGE-SIGNAL
 DIFFERENTIAL
 VOLTAGE AMPLIFICATION
 vs
 SUPPLY VOLTAGE

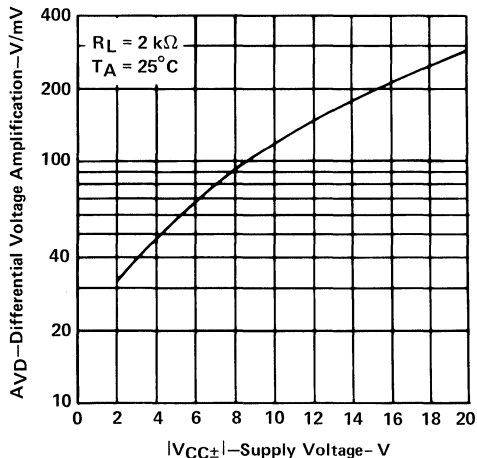


FIGURE 7

OPEN-LOOP LARGE-SIGNAL
 DIFFERENTIAL
 VOLTAGE AMPLIFICATION
 vs
 FREQUENCY

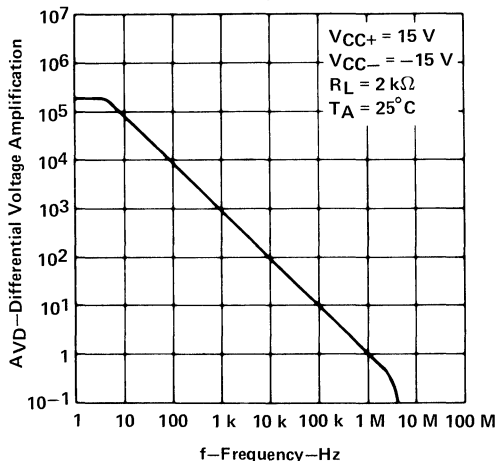


FIGURE 8

2
 Operational Amplifiers

TYPICAL CHARACTERISTICS

COMMON-MODE REJECTION RATIO
vs
FREQUENCY

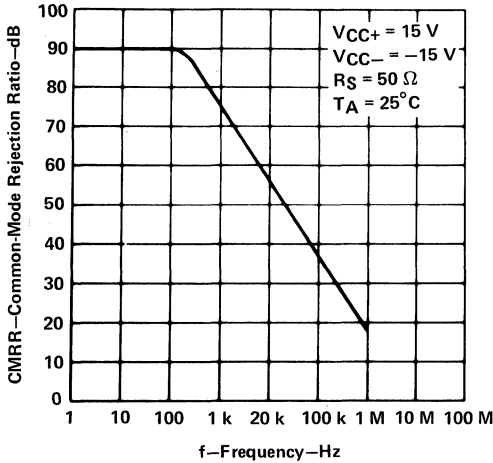


FIGURE 9

OUTPUT VOLTAGE
vs
ELAPSED TIME

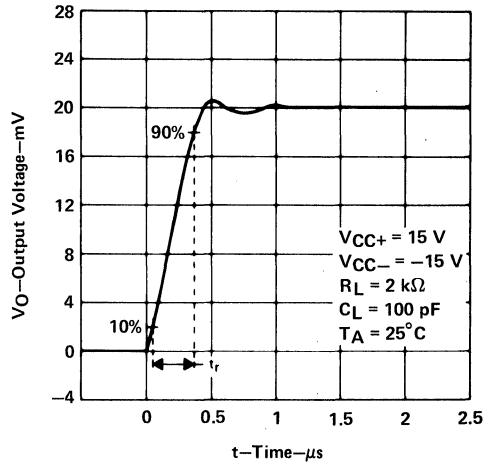


FIGURE 10

VOLTAGE-FOLLOWER
LARGE-SIGNAL PULSE RESPONSE

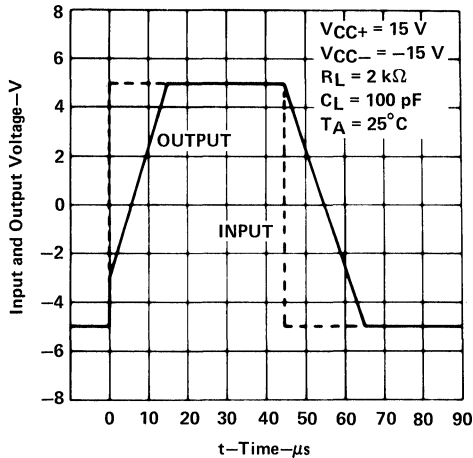


FIGURE 11

uA748M, uA748C GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

D921, DECEMBER 1970—REVISED NOVEMBER 1988

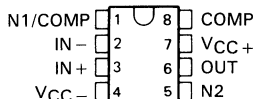
- Frequency and Transient Response Characteristics Adjustable
- Short-Circuit Protection
- Offset-Voltage Null Capability
- Wide Common-Mode and Differential Voltage Ranges
- Low Power Consumption
- No Latch-Up
- Same Pin Assignments as uA709

description

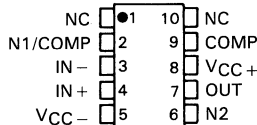
The uA748 is a general-purpose operational amplifier that offers the same advantages and attractive features as the uA741 except for internal compensation. External compensation can be as simple as a 30-pF capacitor for unity-gain conditions and, when the closed-loop gain is greater than one, can be changed to obtain wider bandwidth or higher slew rate. This circuit features high gain, large differential and common-mode input voltage range, and output short-circuit protection. Input offset voltage adjustment can be provided by connecting a variable resistor between the offset null pins as shown in Figure 12

The uA748M is characterized for operation over the full military temperature range of -55°C to 125°C ; the uA748C is characterized for operation from 0°C to 70°C .

uA748M . . . JG PACKAGE
uA748C . . . D, JG, OR P PACKAGE
(TOP VIEW)

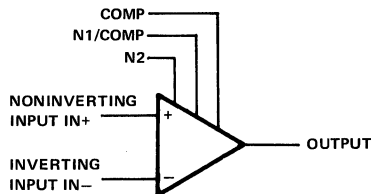


uA748M . . . U FLAT PACKAGE
(TOP VIEW)



NC—No internal connection

symbol

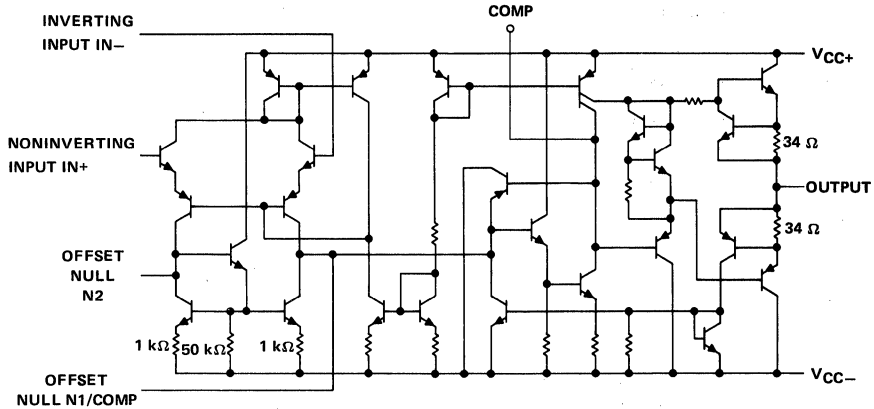


2

Operational Amplifiers

uA748M, uA748C GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

schematic



Resistor values shown are nominal.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	uA748M	uA748C	UNIT
Supply voltage V_{CC+} (see Note 1)	22	18	V
Supply voltage V_{CC-} (see Note 1)	-22	-18	V
Differential input voltage (see Note 2)	± 30	± 30	V
Input voltage (either input, see Notes 1 and 3)	± 15	± 15	V
Voltage between either offset null terminal (N1/N2) and V_{CC-}	-0.5 to 2	-0.5 to 2	V
Duration of output short-circuit (see Note 4)	unlimited	unlimited	
Continuous total power dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	JG or U package	300	300
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package		260

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
 4. The output may be shorted to ground or either power supply. For the uA748M only, the unlimited duration of the short-circuit applies at (or below) 125°C case temperature or 75°C free-air temperature.

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	500 mW	5.8 mW/°C	64°C	464 mW	N/A
JG (uA748M)	500 mW	8.4 mW/°C	90°C	500 mW	210 mW
JG	500 mW	N/A	N/A	500 mW	N/A
P	500 mW	N/A	N/A	500 mW	N/A
U	500 mW	5.4 mW/°C	57°C	432 mW	135 mW

μA748M, μA748C GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$, $C_C = 30\text{ pF}$

PARAMETER	TEST CONDITIONS [†]	μA748M			μA748C			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX		
V_{IO} Input offset voltage	$V_O = 0$	25°C	1	5		1	6	mV	
		Full range			6		7.5		
I_{IO} Input offset current	$V_O = 0$	25°C	20	200	20	200		nA	
		Full range		500		300			
I_{IB} Input bias current	$V_O = 0$	25°C	80	500	80	500		nA	
		Full range		1500		800			
V_{ICR} Common-mode input voltage range		25°C	±12	±13	±12	±13		V	
		Full range	±12		±12				
V_{OM} Maximum peak output voltage swing	$R_L = 10\text{ k}\Omega$	25°C	±12	±14	±12	±14		V	
		Full range	±12		±12				
		25°C	$R_L \geq 2\text{ k}\Omega$	±10	±13	±10	±13		
				Full range	±10		±10		
A_{VD} Large-signal differential voltage amplification	$R_L \geq 2\text{ k}\Omega$, $V_O = \pm 10\text{ V}$	25°C	50	200	20	200		V/mV	
		Full range	25		15				
r_i Input resistance		25°C	0.3	2	0.3	2		MΩ	
r_o Output resistance	$V_O = 0$, See Note 5	25°C		75		75		Ω	
C_i Input capacitance		25°C		1.4		1.4		pF	
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR}\text{ min.}$, $V_O = 0$	25°C	70	90	70	90		dB	
		Full range	70		70				
k_{SVS} Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	$V_{CC} = \pm 9\text{ V to } \pm 15\text{ V}$, $V_O = 0$	25°C	30	150	30	150		μV/V	
		Full range		150		150			
I_{OS} Short-circuit output current		25°C	±25	±40	±25	±40		mA	
I_{CC} Supply current	No load, $V_O = 0$	25°C	1.7	2.8	1.7	2.8		mA	
		Full range		3.3		3.3			
P_D Total power dissipation	No load, $V_O = 0$	25°C	50	85	50	85		mW	
		Full range		100		100			

[†] All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for μA748M is -55°C to 125°C and for μA748C is 0°C to 70°C.

NOTE 5: This typical value applies only at frequencies above a few hundred hertz because of the effects of drift and thermal feedback.

operating characteristics, $V_{CC+} = 15\text{ V}$, $V_{CC-} = -15\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	μA748M			μA748C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
t_r Rise time	$V_I = 20\text{ mV}$, $R_L = 2\text{ k}\Omega$, $C_L = 100\text{ pF}$, $C_C = 30\text{ pF}$, See Figure 1	0.3			0.3			μs
Overshoot factor		5%			5%			
SR Slew rate at unity gain	$V_I = 10\text{ V}$, $R_L = 2\text{ k}\Omega$, $C_L = 100\text{ pF}$, $C_C = 30\text{ pF}$, See Figure 1	0.5			0.5			V/μs

2

Operational Amplifiers

PARAMETER MEASUREMENT INFORMATION

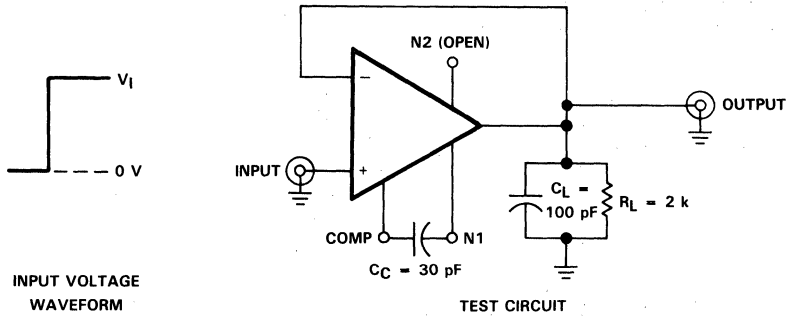


FIGURE 1. RISE TIME, OVERSHOOT, AND SLEW RATE

TYPICAL CHARACTERISTICS

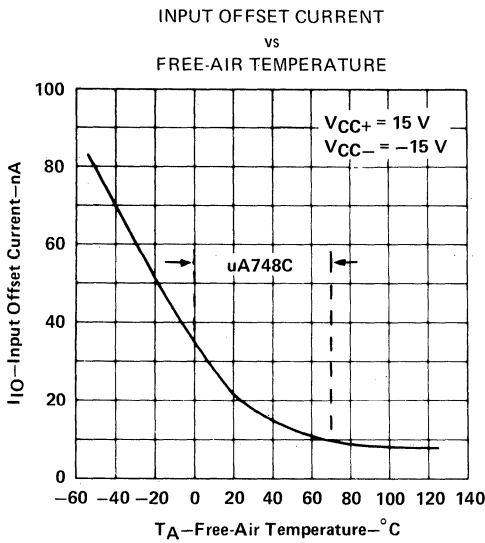


FIGURE 2

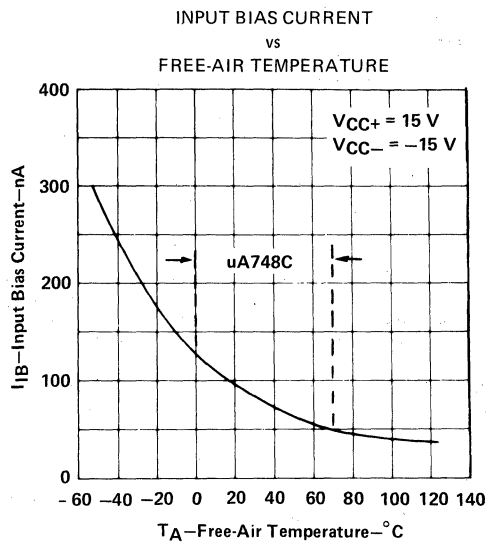


FIGURE 3

TYPICAL CHARACTERISTICS

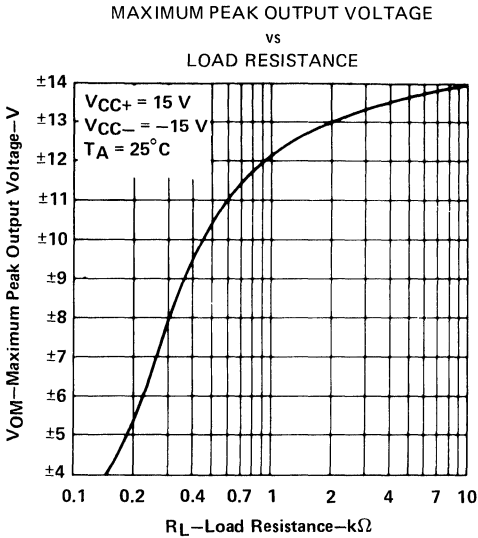


FIGURE 4

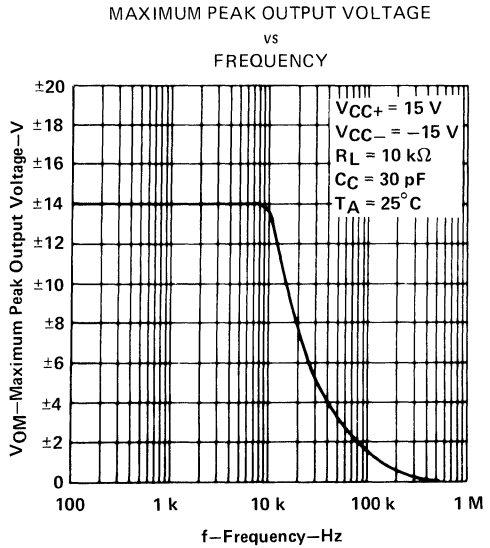


FIGURE 5

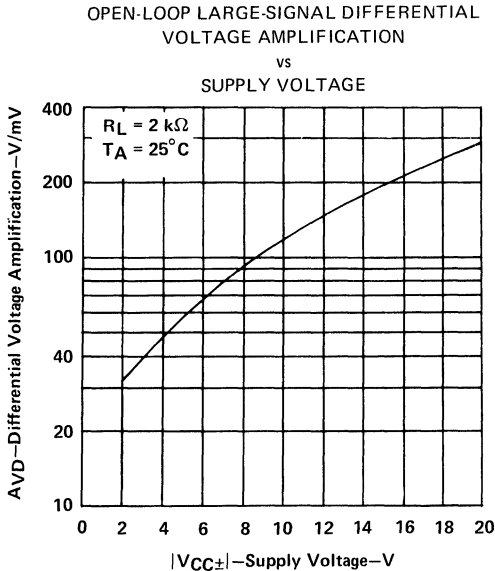


FIGURE 6

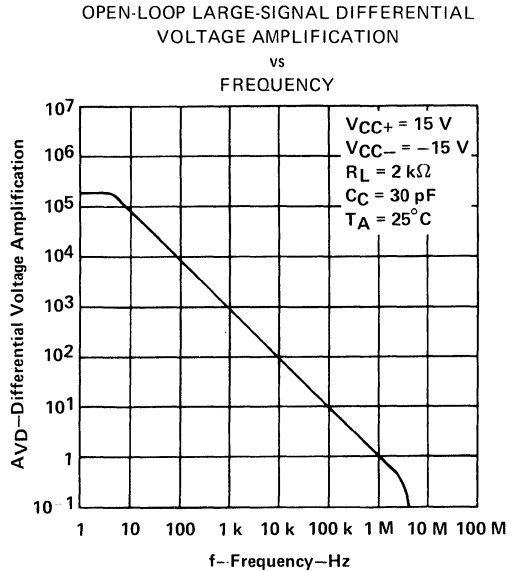


FIGURE 7

μA748M, μA748C
GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

TYPICAL CHARACTERISTICS

COMMON-MODE REJECTION RATIO
 vs
 FREQUENCY

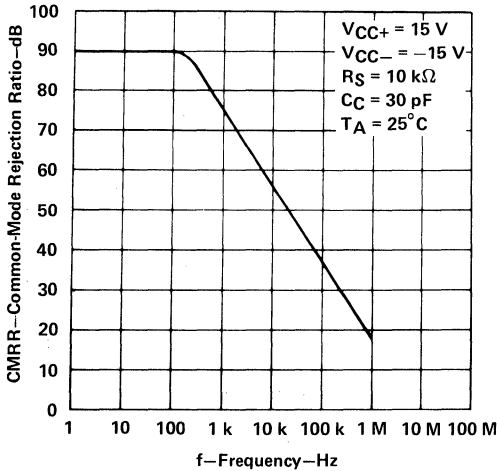


FIGURE 8

OUTPUT VOLTAGE
 vs
 ELAPSED TIME

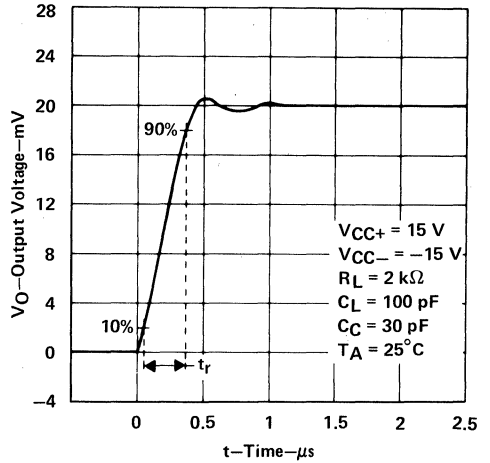


FIGURE 9

VOLTAGE-FOLLOWER
 LARGE-SIGNAL PULSE RESPONSE

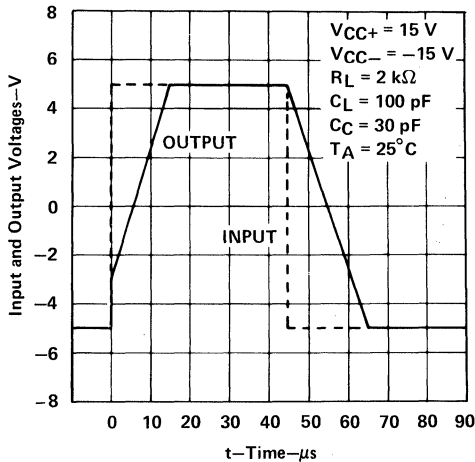


FIGURE 10

2
Operational Amplifiers

TYPICAL APPLICATION DATA

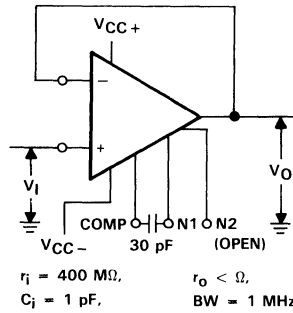
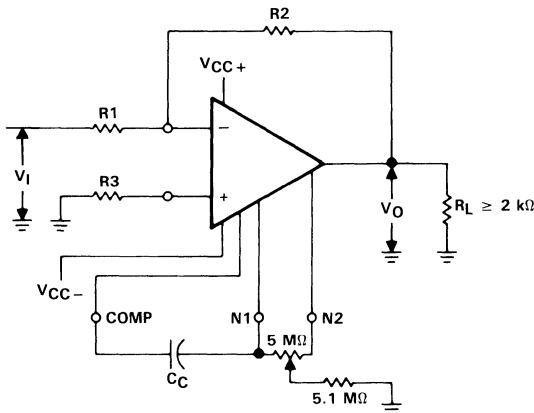


FIGURE 11. UNITY-GAIN VOLTAGE FOLLOWER



$$\frac{V_O}{V_I} = -\frac{R_2}{R_1}$$

$$C_C \geq \frac{R_1 \cdot 30 \text{ pF}}{R_1 + R_2}$$

$$R_3 = \frac{R_1 \cdot R_2}{R_1 + R_2}$$

FIGURE 12. INVERTING CIRCUIT WITH ADJUSTABLE GAIN
COMPENSATION, AND OFFSET ADJUSTMENT

2

Operational Amplifiers

General Information

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Operational Amplifiers

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Special Functions

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Product Previews

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Mechanical Data

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3

Voltage Comparators

LM111, LM211, LM311 DIFFERENTIAL COMPARATORS WITH STROBES

D1312, SEPTEMBER 1973—REVISED MARCH 1988

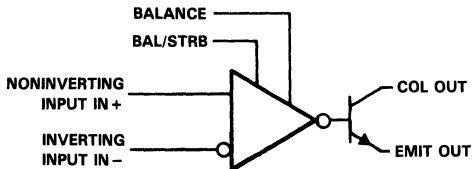
- **Fast Response Times**
- **Strobe Capability**
- **Designed to be Interchangeable with National Semiconductor LM111, LM211, and LM311**
- **Maximum Input Bias Current . . . 300 nA**
- **Maximum Input Offset Current . . . 70 nA**
- **Can Operate from Single 5-V Supply**

description

The LM111, LM211, and LM311 are single high-speed voltage comparators. These devices are designed to operate from a wide range of power supply voltage, including ± 15 -V supplies for operational amplifiers and 5-V supplies for logic systems. The output levels are compatible with most TTL and MOS circuits. These comparators are capable of driving lamps or relays and switching voltages up to 50 V at 50 mA. All inputs and outputs can be isolated from system ground. The outputs can drive loads referenced to ground, V_{CC+} or V_{CC-} . Offset balancing and strobe capability are available and the outputs can be wire-OR connected. If the strobe is low, the output will be in the off state regardless of the differential input.

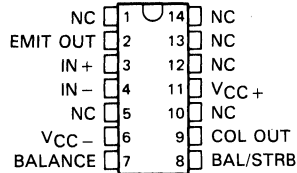
The LM111 is characterized for operation over the full military range of -55°C to 125°C . The LM211 is characterized for operation from -25°C to 85°C , and the LM311 is characterized for operation from 0°C to 70°C .

functional block diagram



LM111 . . . J PACKAGE

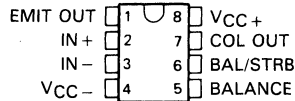
(TOP VIEW)



LM111 . . . JG PACKAGE

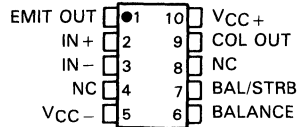
LM211, LM311 . . . D, JG, OR P PACKAGE

(TOP VIEW)



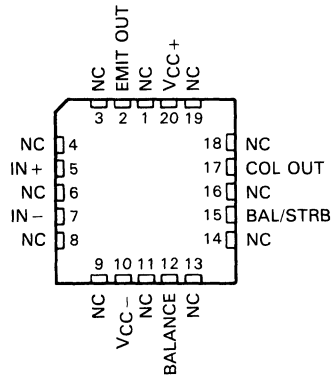
LM111 . . . U FLAT PACKAGE

(TOP VIEW)



LM111 . . . FK CHIP CARRIER PACKAGE

(TOP VIEW)



NC—No internal connection

3
Voltage Comparators

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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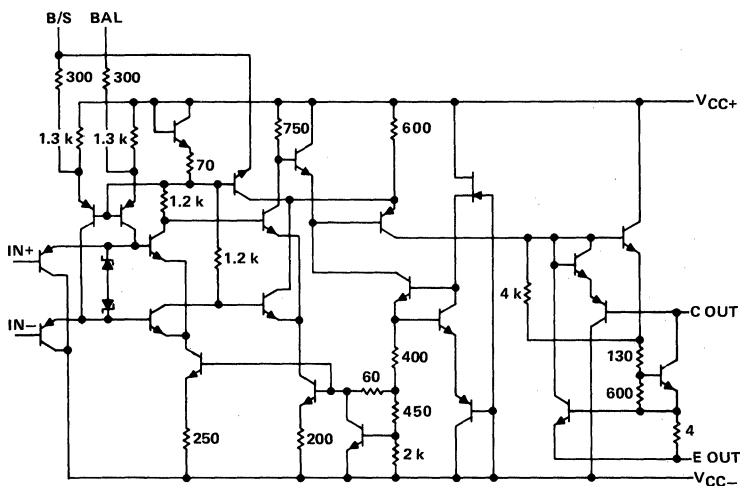
LM111, LM211, LM311 DIFFERENTIAL COMPARATORS WITH STROBES

AVAILABLE OPTIONS

OPERATING TEMPERATURE RANGE	V _{IO} MAX AT T _A = 25°C	PACKAGE					
		D SMALL OUTLINE	FK CERAMIC CHIP CARRIER	J CERAMIC DIP	JG CERAMIC DIP	P PLASTIC DIP	U FLATPACK
-55°C to 125°C	3 mV		LM111FK	LM111J			
-40°C to 85°C	3 mV	LM211D			LM211JG	LM211P	
0°C to 70°C	7.5 mV	LM311D			LM311JG	LM311P	LM111U

The D package is available in tape and reel. Add an R suffix when ordering, e.g., LM311DR.

schematic



- BAL Balance
- B/S Balance/Strobe
- C OUT Collector Output
- E OUT Emitter Output
- IN+ Noninverting Input
- IN- Inverting Input
- NC No Internal Connection
- VCC+ Positive Supply Voltage
- VCC- Negative Supply Voltage

Resistor values shown are nominal and in ohms.

3

Voltage Comparators

LM111, LM211, LM311 DIFFERENTIAL COMPARATORS WITH STROBES

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	LM111	LM211	LM311	UNIT
Supply voltage, V_{CC+} (see Note 1)	18	18	18	V
Supply voltage, V_{CC-} (see Note 1)	-18	-18	-18	V
Differential input voltage (see Note 2)	± 30	± 30	± 30	V
Input voltage (either input, see Notes 1 and 3)	± 15	± 15	± 15	V
Voltage from emitter output to V_{CC-}	30	30	30	V
Voltage from collector output to V_{CC-}	50	50	40	V
Duration of output short-circuit (see Note 4)	10	10	10	s
Continuous total dissipation	See Dissipation Rating Table			
Operating free-air temperature range	-55 to 125	-25 to 85	0 to 70	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Case temperature for 60 seconds: FK Package	260			$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: J, JG, or U package	300	300	300	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: D or P package		260	260	$^{\circ}\text{C}$

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^{\circ}\text{C}$ POWER RATING	$T_A = 85^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
D	500 mW	5.8 mW/ $^{\circ}\text{C}$	64 $^{\circ}\text{C}$	464 mW	377 mW	-
FK	500 mW	11.0 mW/ $^{\circ}\text{C}$	105 $^{\circ}\text{C}$	500 mW	500 mW	275 mW
J (LM111)	500 mW	11.0 mW/ $^{\circ}\text{C}$	105 $^{\circ}\text{C}$	500 mW	500 mW	275 mW
J	500 mW	8.2 mW/ $^{\circ}\text{C}$	89 $^{\circ}\text{C}$	500 mW	500 mW	-
JG (LM111)	500 mW	8.4 mW/ $^{\circ}\text{C}$	90 $^{\circ}\text{C}$	500 mW	500 mW	210 mW
JG	500 mW	6.6 mW/ $^{\circ}\text{C}$	74 $^{\circ}\text{C}$	500 mW	429 mW	-
P	500 mW	8.0 mW/ $^{\circ}\text{C}$	88 $^{\circ}\text{C}$	500 mW	500 mW	-
U	500 mW	5.4 mW/ $^{\circ}\text{C}$	57 $^{\circ}\text{C}$	432 mW	351 mW	135 mW

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or ± 15 volts, whichever is less.
 4. The output may be shorted to ground or either power supply.



Voltage Comparators

LM111, LM211, LM311

DIFFERENTIAL COMPARATORS WITH STROBES

electrical characteristics at specified free-air temperature, $V_{CC} \pm = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†		LM111, LM211			LM311			UNIT	
			MIN	TYP‡	MAX	MIN	TYP‡	MAX		
V_{IO} Input offset voltage	See Note 5	25°C	0.7			2			mV	
		Full range	4			10				
I_{IO} Input offset current	See Note 5	25°C	4			6			nA	
		Full range	20			70				
I_{IB} Input bias current	$V_O = 1\text{ V to }14\text{ V}$	25°C	75			100			nA	
		Full range	150			300				
$I_{IL(S)}$ Low-level strobe current (see Note 6)	$V_{(strobe)} = 0.3\text{ V}, V_{ID} \leq -10\text{ mV}$	25°C	-3			-3			mA	
V_{ICR} Common-mode input voltage range		Full range	13 to -14.5	13.8 to -14.7		13 to -14.5	13.8 to -14.7		V	
A_{VD} Large-signal differential voltage amplification	$V_O = 5\text{ V to }35\text{ V}, R_L = 1\text{ k}\Omega$	25°C	40	200		40	200		V/mV	
I_{OH} High-level (collector) output current	$I_{strobe} = -3\text{ mA}, V_{ID} = 5\text{ mV}, V_{OH} = 35\text{ V}$	25°C	0.2			10			nA	
		Full range	0.5						μA	
V_{OL} Low-level (collector-to-emitter) output voltage	$I_{OL} = 50\text{ mA}, V_{CC+} = 4.5\text{ V}, V_{CC-} = 0, I_{OL} = 8\text{ mA}$	$V_{ID} = 5\text{ mV}, V_{OH} = 35\text{ V}$	25°C	0.2			50			nA
		$V_{ID} = -5\text{ mV}$	25°C	0.75			1.5			
		$V_{ID} = -10\text{ mV}$	25°C				0.75			1.5
		Full range		0.23			0.4			
I_{CC+} Supply current from V_{CC+} , output low	$V_{ID} = -10\text{ mV},$ No load	25°C	5.1			6			mA	
I_{CC-} Supply current from V_{CC-} , output high	$V_{ID} = 10\text{ mV},$ No load	25°C	-4.1			-5			mA	

† Unless otherwise noted, all characteristics are measured with the balance and balance/strobe terminals open and the emitter output grounded. Full range for LM111 is -55°C to 125°C , for LM211 is -25°C to 85°C , and for LM311 is 0°C to 70°C .

‡ All typical values are at $T_A = 25^\circ\text{C}$.

NOTES: 5. The offset voltages and offset currents given are the maximum values required to drive the collector output up to 14 V or down to 1 V with a pull-up resistor of 7.5 k Ω to V_{CC+} . Thus these parameters actually define an error band and take into account the worst-case effects of voltage gain and input impedance.

6. The strobe should not be shorted to ground; it should be current driven at -3 to -5 mA, e.g., see Figures 13 and 27.

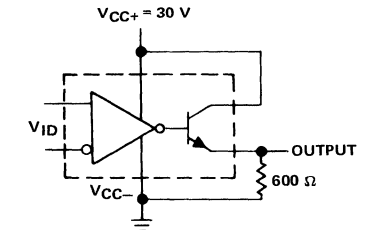
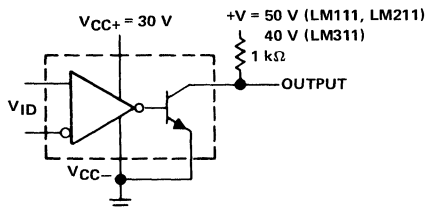
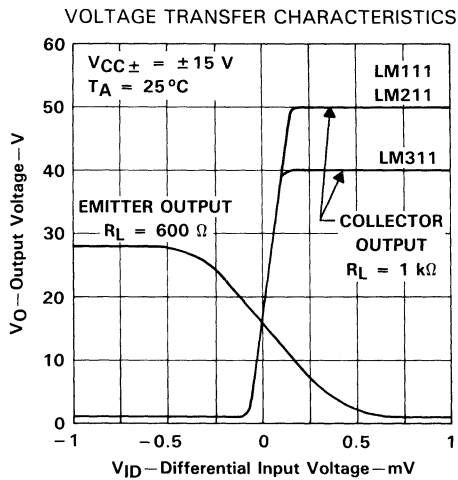
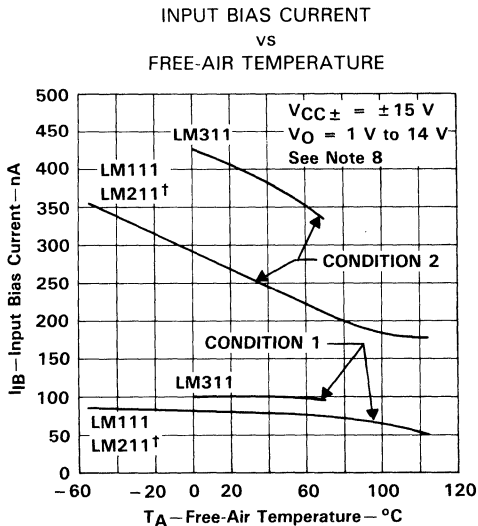
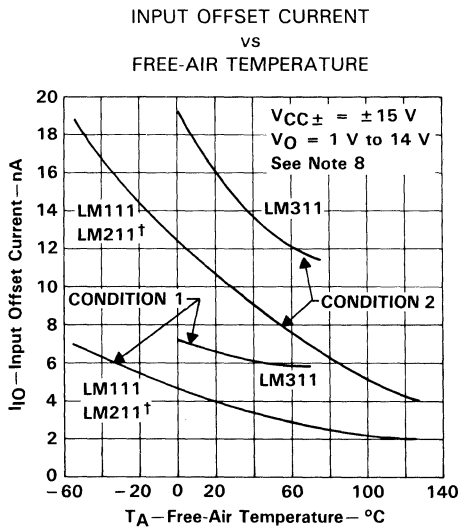
switching characteristics, $V_{CC+} = 15\text{ V}, V_{CC-} = -15\text{ V}, T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Response time, low-to-high-level output	$R_C = 500\ \Omega$ to 5 V, $C_L = 5\text{ pF}$, See Note 7	115			ns
Response time, high-to-low-level output		165			ns

NOTE 7: The response time specified is for a 100-mV input step with 5-mV overdrive and is the interval between the input step function and the instant when the output crosses 1.4 V.

LM111, LM211, LM311 DIFFERENTIAL COMPARATORS WITH STROBES

TYPICAL CHARACTERISTICS



¹Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.
NOTE 8: Condition 1 is with the balance and balance/strobe terminals open. Condition 2 is with the balance and balance/strobe terminals connected to V_{CC+} .

3
Voltage Comparators

TYPICAL CHARACTERISTICS

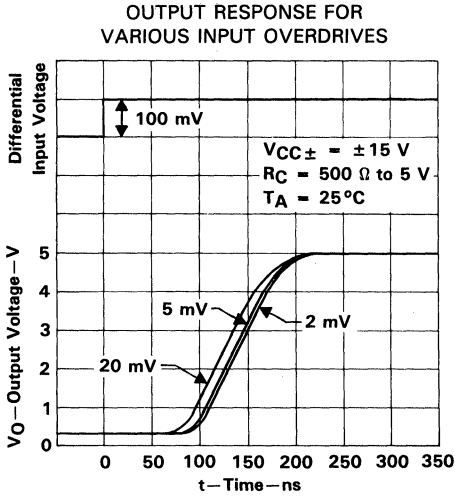


FIGURE 4

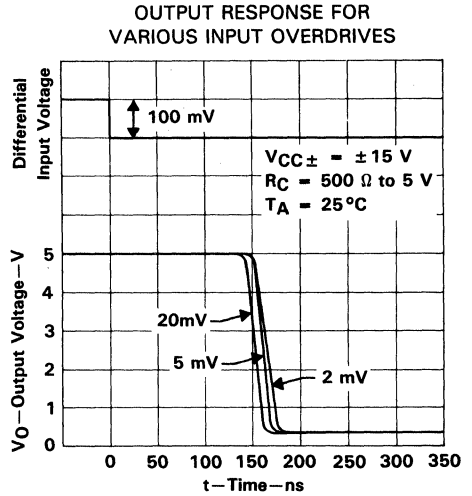
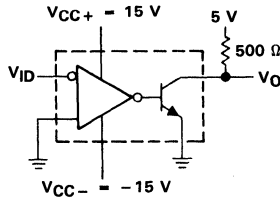


FIGURE 5



TEST CIRCUIT FOR FIGURES 4 AND 5

3
 Voltage Comparators

TYPICAL CHARACTERISTICS

OUTPUT RESPONSE FOR
VARIOUS INPUT OVERDRIVES

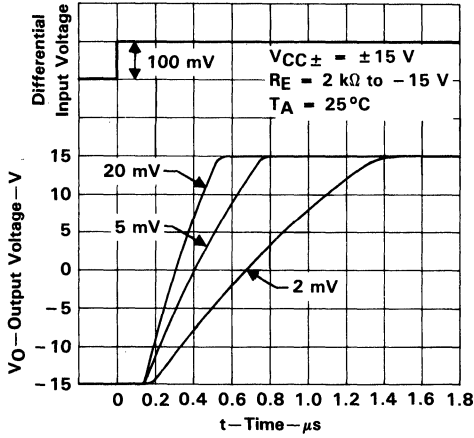


FIGURE 6

OUTPUT RESPONSE FOR
VARIOUS INPUT OVERDRIVES

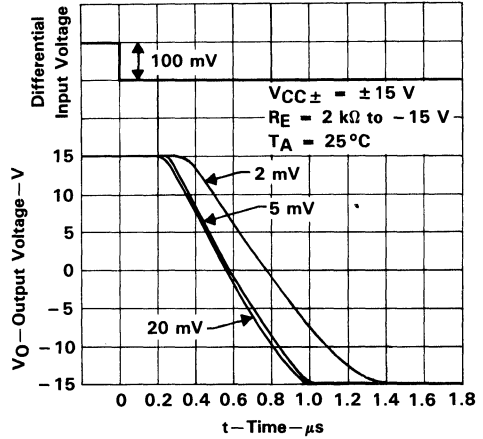
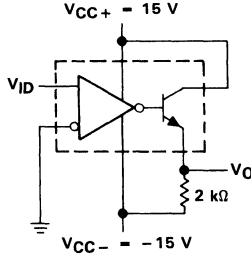


FIGURE 7



TEST CIRCUIT FOR FIGURES 6 AND 7

3

Voltage Comparators

TYPICAL CHARACTERISTICS

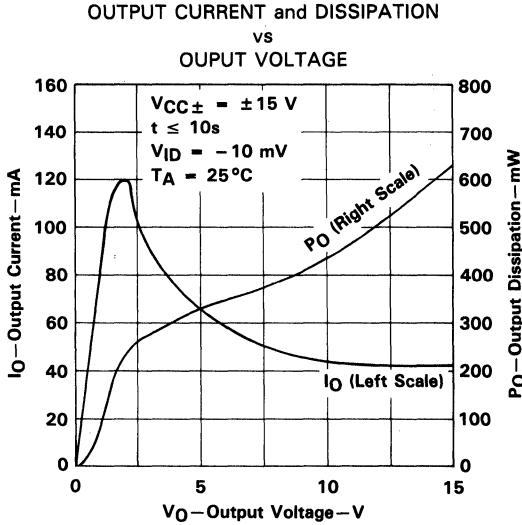


FIGURE 8

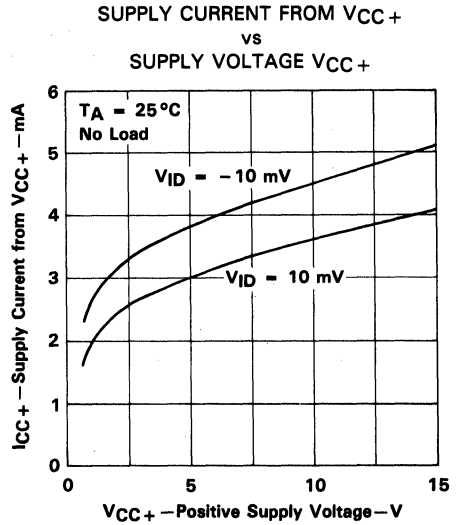


FIGURE 9

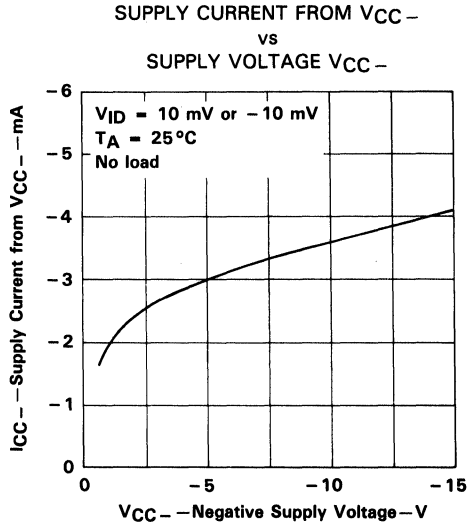


FIGURE 10

3

Voltage Comparators

TYPICAL APPLICATION DATA

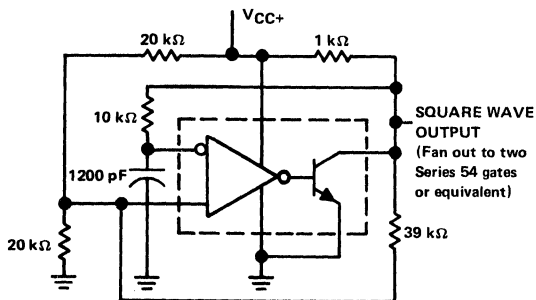


FIGURE 11. 100 kHz
FREE-RUNNING MULTIVIBRATOR

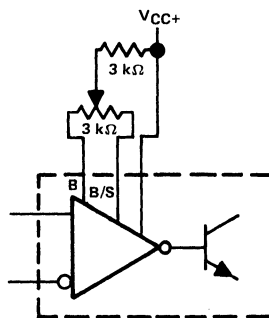


FIGURE 12. OFFSET BALANCING

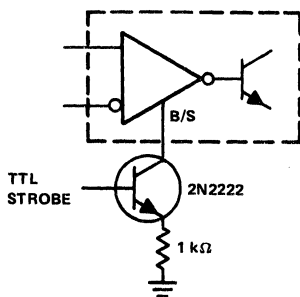


FIGURE 13. STROBING

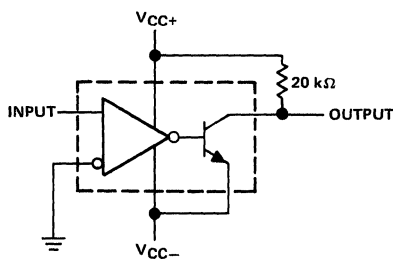
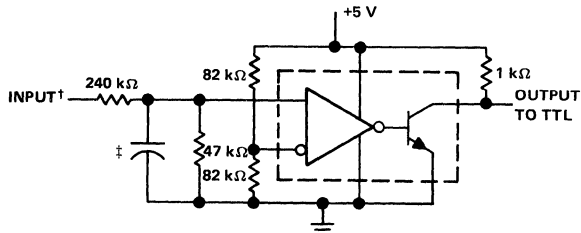


FIGURE 14. ZERO-CROSSING DETECTOR



† Resistor values shown are for a 0-to-30-V logic swing and a 15-V threshold.

‡ May be added to control speed and reduce susceptibility to noise spikes.

FIGURE 15. TTL INTERFACE WITH HIGH-LEVEL LOGIC

TYPICAL APPLICATION DATA

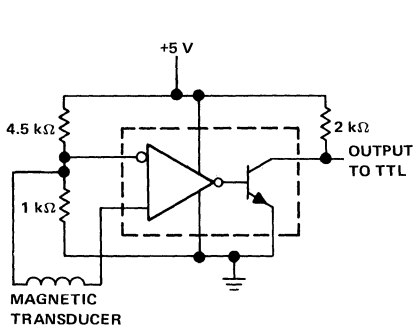


FIGURE 16. DETECTOR FOR MAGNETIC TRANSUDCER

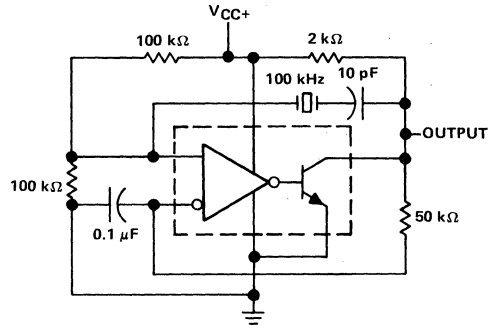


FIGURE 17. 100 kHz CRYSTAL OSCILLATOR

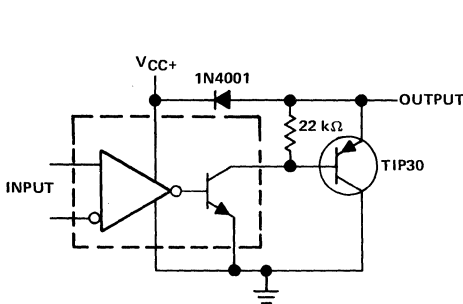
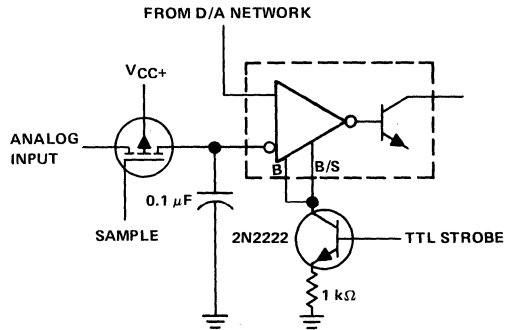


FIGURE 18. COMPARATOR AND SOLENOID DRIVER



Typical input current is 50 pA with inputs strobed off.
FIGURE 19. STROBING BOTH INPUT AND OUTPUT STAGES SIMULTANEOUSLY

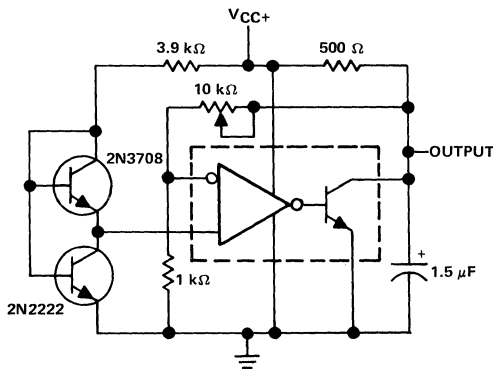


FIGURE 20. LOW-VOLTAGE ADJUSTABLE REFERENCE SUPPLY

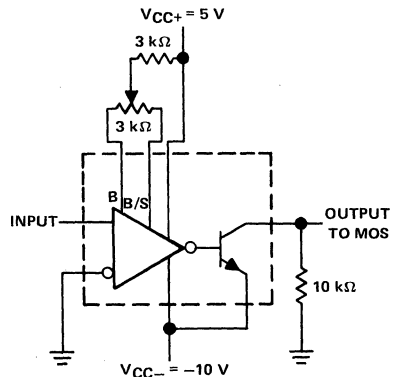
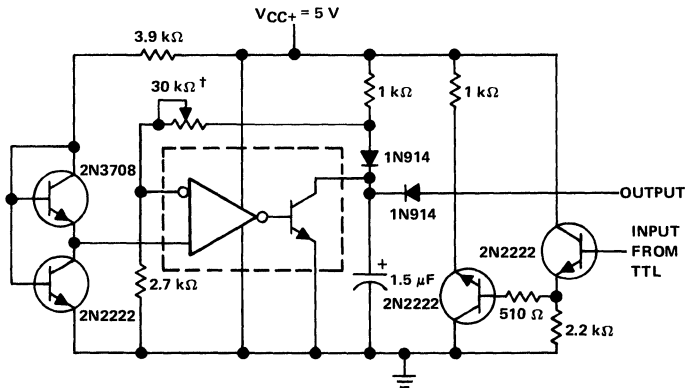


FIGURE 21. ZERO-CROSSING DETECTOR DRIVING MOS LOGIC

3 Voltage Comparators

TYPICAL APPLICATION DATA



† Adjust to set clamp level.

FIGURE 22. PRECISION SQUARER

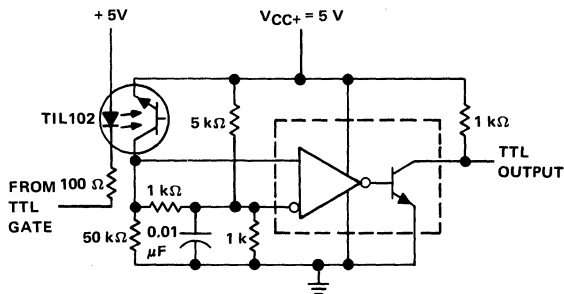


FIGURE 23. DIGITAL TRANSMISSION ISOLATOR

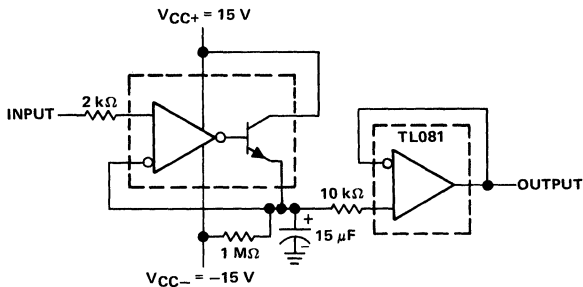


FIGURE 24. POSITIVE-PEAK DETECTOR

Voltage Comparators

TYPICAL APPLICATION DATA

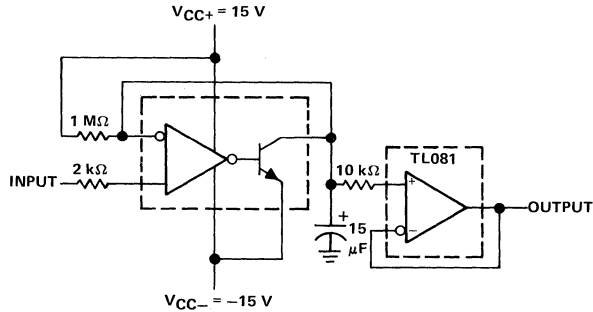
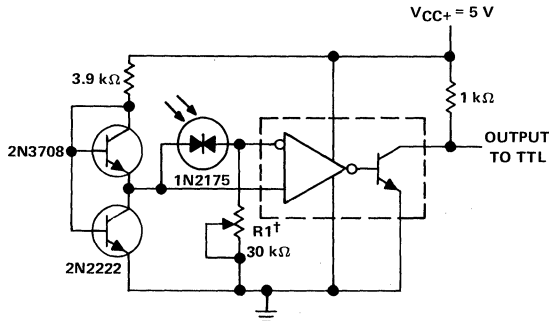
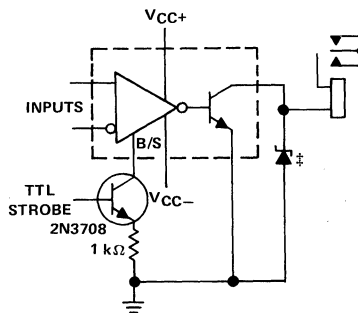


FIGURE 25. NEGATIVE-PEAK DETECTOR



† $R1$ sets the comparison level. At comparison, the photodiode has less than 5 mV across it decreasing dark current by an order of magnitude.

FIGURE 26. PRECISION PHOTODIODE COMPARATOR



‡Transient voltage and inductive kickback protection

FIGURE 27. RELAY DRIVER WITH STROBE

3
Voltage Comparators

LM111, LM211, LM311
DIFFERENTIAL COMPARATORS WITH STROBES

TYPICAL APPLICATION DATA

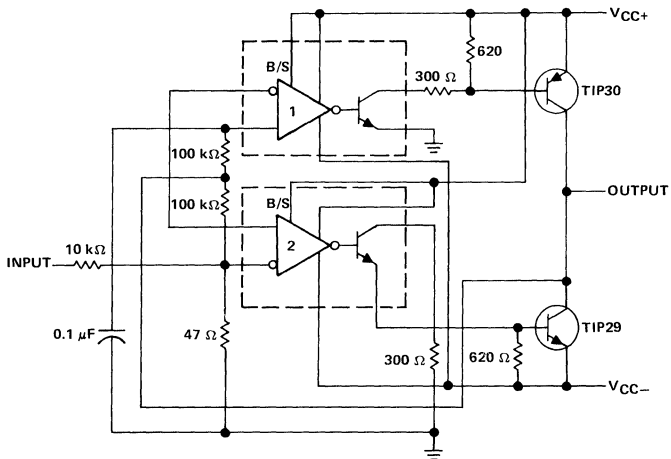


FIGURE 28. SWITCHING POWER AMPLIFIER

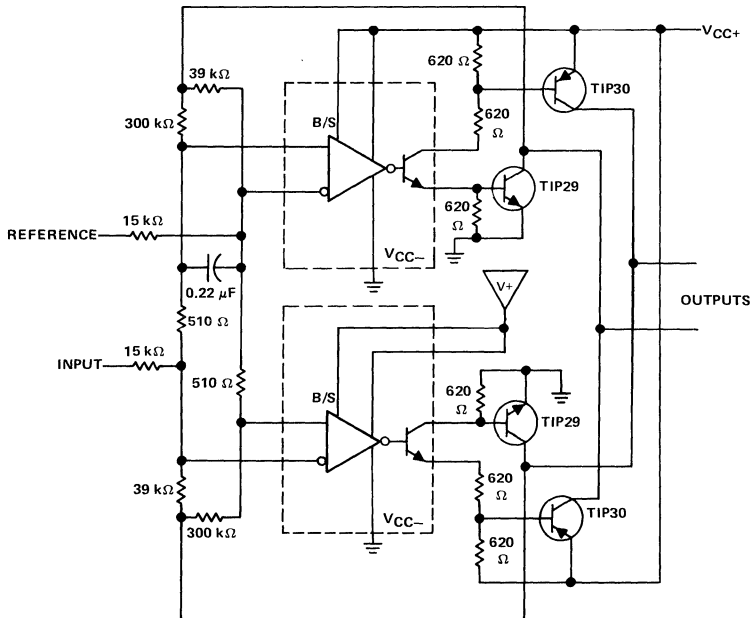


FIGURE 29. SWITCHING POWER AMPLIFIERS



Voltage Comparators

Voltage Comparators



LM139, LM239, LM339, LM139A LM239A, LM339A, LM2901 QUADRUPLE DIFFERENTIAL COMPARATORS

D1979, OCTOBER 1979 - REVISED APRIL 1988

- Single Supply or Dual Supplies
- Wide Range of Supply Voltage . . . 2 to 36 V
- Low Supply Current Drain Independent of Supply Voltage . . . 0.8 mA Typ
- Low Input Bias Current . . . 25 nA Typ
- Low Input Offset Current . . . 3 nA Typ (LM139)
- Low Input Offset Voltage . . . 2 mV Typ
- Common-Mode Input Voltage Range Includes Ground
- Differential Input Voltage Range Equal to Maximum-Rated Supply Voltage . . . ± 36 V
- Low Output Saturation Voltage
- Output Compatible with TTL, MOS, and CMOS

description

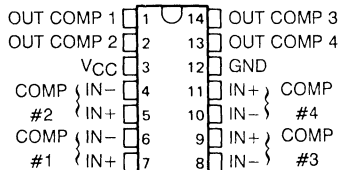
These devices consist of four independent voltage comparators that are designed to operate from a single power supply over a wide range of voltages. Operation from dual supplies is also possible as long as the difference between the two supplies is 2 V to 36 V and pin 3 is at least 1.5 V more positive than the input common-mode voltage. Current drain is independent of the supply voltage. The outputs can be connected to other open-collector outputs to achieve wired-AND relationships.

AVAILABLE OPTIONS

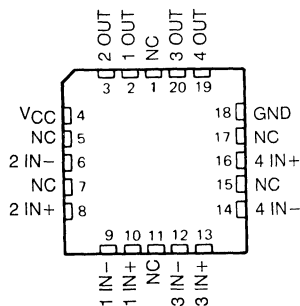
T _A	V _{IO} MAX at 25°C	PACKAGE			
		SMALL OUTLINE (D)	CERAMIC (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	5 mV	LM339D	—	LM339J	LM339N
—	2 mV	LM339AD	—	LM339AJ	LM339AN
-25°C to 85°C	5 mV	LM239D	—	LM239J	LM239N
—	2 mV	LM239AD	—	LM239AJ	LM239AN
-40°C to 125°C	7 mV	LM2901D	—	LM2901J	LM2901N
-55°C to 125°C	5 mV	—	LM139FK	LM139J	—
—	2 mV	—	LM139AFK	LM139AJ	—

The D package is available taped and reeled. Add the suffix R to the device type when ordering. (e.g., LM339DR)

LM139, LM139A . . . J PACKAGE
ALL OTHERS . . . D, J, OR N PACKAGE
(TOP VIEW)

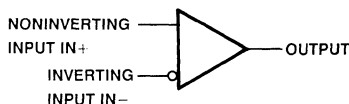


LM139, LM139A
FK CHIP CARRIER PACKAGE
(TOP VIEW)



NC—No internal connection

symbol (each comparator)



3

Voltage Comparators

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS
INSTRUMENTS

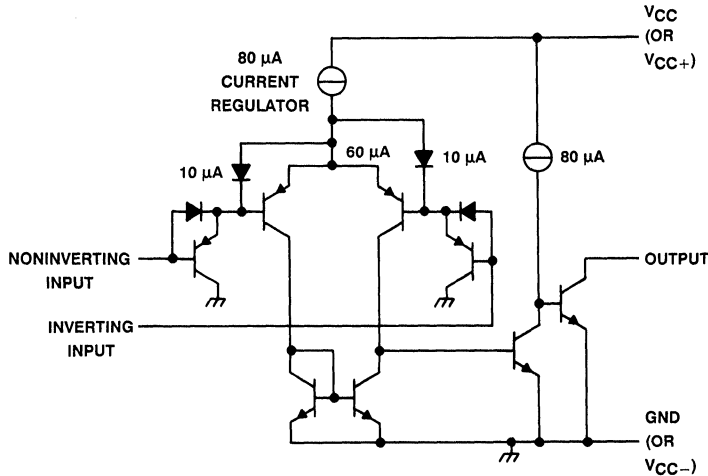
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3-17

LM139, LM239, LM339, LM139A LM239A, LM339A, LM2901 QUADRUPLE DIFFERENTIAL COMPARATORS

schematic (each comparator)



3

Voltage Comparators

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 2)	36 V
Differential input voltage (see Note 3)	± 36 V
Input voltage range (either input)	-0.3 V to 36 V
Output voltage	36 V
Output current	20 mA
Duration of output short-circuit to ground (see Note 4)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range:	
LM139	-55°C to 125°C
LM239, LM239A	-25°C to 85°C
LM339, LM339A	0°C to 70°C
LM2901	-40°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C

- NOTES: 2. All voltage values, except differential voltages, are with respect to the network ground terminal.
3. Differential voltages are at the noninverting input terminal with respect to the inverting input.
4. Short circuits from outputs to V_{CC} can cause excessive heating and eventual destruction.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	900 mW	7.6 mW/°C	31°C	608 mW	494 mW	—
FK	900 mW	11.0 mW/°C	68°C	880 mW	715 mW	275 mW
J (LM139, LM139A)	900 mW	11.0 mW/°C	68°C	880 mW	715 mW	275 mW
J (All others)	900 mW	8.2 mW/°C	40°C	656 mW	533 mW	—
N	900 mW	9.2 mW/°C	52°C	736 mW	598 mW	—

electrical characteristics at specified free-air temperature, $V_{CC} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	LM139			LM139A			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{CC} = 5\text{ V to }30\text{ V}$, $V_{IC} = V_{ICR\text{ min}}$, $V_O = 1.4\text{ V}$	25°C	2	5	1	2	mV	
		-55°C to 125°C	9			4		
I_{IO} Input offset current	$V_O = 1.4\text{ V}$	25°C	3	25	3	25	nA	
		-55°C to 125°C	100			100		
I_{IB} Input bias current	$V_O = 1.4\text{ V}$	25°C	-25	-100	-25	-100	nA	
		-55°C to 125°C	-300			-300		
V_{ICR} Common-mode input voltage range		25°C	0 to $V_{CC}-1.5$		0 to $V_{CC}-1.5$		V	
		-55°C to 125°C	0 to $V_{CC}-2$		0 to $V_{CC}-2$			
AV_D Large-signal differential voltage amplification	$V_{CC\pm} = \pm 7.5\text{ V}$, $V_O = -5\text{ V to }5\text{ V}$	25°C	200		50	200	V/mV	
I_{OH} High-level output current	$V_{ID} = 1\text{ V}$	$V_{OH} = 5\text{ V}$	25°C	0.1	0.1	nA		
		$V_{OH} = 30\text{ V}$	-55°C to 125°C	1			1	μA
V_{OL} Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 4\text{ mA}$	25°C	150	400	150	400	mV	
		-55°C to 125°C	700			700		
I_{OL} Low-level output current	$V_{ID} = -1\text{ V}$, $V_{OL} = 1.5\text{ V}$	25°C	6	16	6	16	mA	
I_{CC} Supply current (four comparators)	$V_O = 2.5\text{ V}$, No load	25°C	0.8	2	0.8	2	mA	

† All characteristics are measured with zero common-mode input voltage unless otherwise specified.

switching characteristics, $V_{CC} = 5\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
Response time	R_L connected to 5 V through 5.1 kΩ, $C_L = 15\text{ pF}$ ‡ See Note 4	100-mV input step with 5-mV overdrive		1.3		μs
		TTL-level input step		0.3		

‡ C_L includes probe and jig capacitance.

NOTE 4: The response time specified is the interval between the input step function and the instant when the output crosses 1.4 V.

LM139, LM139A
QUADRUPLE DIFFERENTIAL COMPARATORS

electrical characteristics at specified free-air temperature, $V_{CC} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	LM239, LM339			LM239A, LM339A			LM2901			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX		
V_{IO} Input offset voltage	$V_{CC} = 5\text{ V}$ to 30 V , $V_{IC} = V_{ICR\text{ min}}$, $V_O = 1.4\text{ V}$	25°C	2	5		1	2		2	7	mV	
		Full range		9			4			15		
I_{IO} Input offset current	$V_O = 1.4\text{ V}$	25°C	5	50		5	50		5	50	nA	
		Full range		150			150			200		
I_{IB} Input bias current	$V_O = 1.4\text{ V}$	25°C	-25	-250		-25	-250		-25	-250	nA	
		Full range		-400			-400			-500		
V_{ICR} Common-mode input voltage range		25°C	0 to $V_{CC}-1.5$			0 to $V_{CC}-1.5$			0 to $V_{CC}-1.5$		V	
		Full range	0 to $V_{CC}-2$			0 to $V_{CC}-2$			0 to $V_{CC}-2$			
A_{VD} Large-signal differential voltage amplification	$V_{CC} = 15\text{ V}$, $V_O = 1.4\text{ V}$ to 11.4 V , $R_L \geq 15\text{ k}\Omega$ to V_{CC}	25°C	50	200		50	200		25	100	V/mV	
I_{OH} High-level output current	$V_{ID} = 1\text{ V}$	$V_{OH} = 5\text{ V}$	25°C	0.1	50		0.1	50		0.1	50	nA
		$V_{OH} = 30\text{ V}$	Full range		1			1			1	μA
V_{OL} Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 4\text{ mA}$	25°C		150	400		150	400		150	500	mV
		Full range		700			700			700		
I_{OL} Low-level output current	$V_{ID} = -1\text{ V}$, $V_{OL} = 1.5\text{ V}$	25°C	6	16		6	16		6	16	mA	
I_{CC} Supply current (four comparators)	$V_O = 2.5\text{ V}$, No load	25°C		0.8	2		0.8	2		0.8	2	mA
	$V_{CC} = 30\text{ V}$, $V_O = 15\text{ V}$, No load								1	2.5		

† Full range (MIN to MAX) for LM239 and LM239A is -25°C to 85°C , for LM339 and LM339A is 0°C to 70°C , and for LM2901 is -40°C to 125°C . All characteristics are measured with zero common-mode input voltage unless otherwise specified.

switching characteristics, $V_{CC} = 5\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Response time	R_L connected to 5 V through $5.1\text{ k}\Omega$, $C_L = 15\text{ pF}$, [‡] See Note 5	100-mV input step with 5-mV overdrive		1.3	μs
		TTL-level input step		0.3	

[‡] C_L includes probe and jig capacitance.

NOTE 5: The response time specified is the interval between the input step function and the instant when the output crosses 1.4 V .

LM193, LM293, LM393, LM293A, LM393A, LM2903 DUAL DIFFERENTIAL COMPARATORS

D2232, JUNE 1976—REVISED NOVEMBER 1988

- Single Supply or Dual Supplies
- Wide Range of Supply Voltage . . . 2 to 36 V
- Low Supply Current Drain Independent of Supply Voltage . . . 0.5 mA Typ
- Low Input Bias Current . . . 25 nA Typ
- Low Input Offset Current . . . 3 nA Typ (LM139)
- Low Input Offset Voltage . . . 2 mV Typ
- Common-Mode Input Voltage Range Includes Ground
- Differential Input Voltage Range Equal to Maximum-Rated Supply Voltage . . . ± 36 V
- Low Output Saturation Voltage
- Output Compatible with TTL, MOS, and CMOS

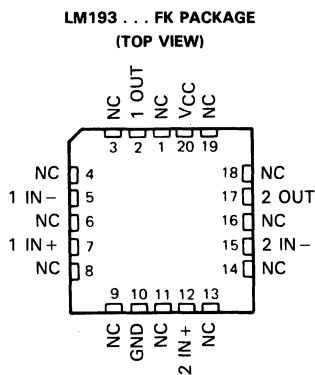
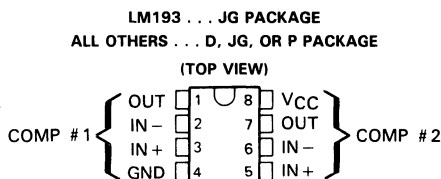
description

These devices consist of two independent voltage comparators that are designed to operate from a single power supply over a wide range of voltages. Operation from dual supplies is also possible as long as the difference between the two supplies is 2 V to 36 V and pin 8 is at least 1.5 V more positive than the input common-mode voltage. Current drain is independent of the supply voltage. The outputs can be connected to other open-collector outputs to achieve wired-AND relationships.

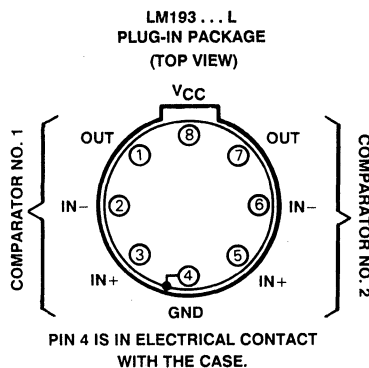
AVAILABLE OPTIONS

SYMBOLIZATION		OPERATING TEMPERATURE RANGE	V _{IO} MAX AT T _A = 25°C
DEVICE	PACKAGE SUFFIX		
LM193	L, FK, JG	-55°C to 125°C	5 mV
LM293	D, JG, P	-25°C to 85°C	5 mV
LM293A	D, JG, P	-25°C to 85°C	2 mV
LM393	D, JG, P	0°C to 70°C	5 mV
LM393A	D, JG, P	0°C to 70°C	2 mV
LM2903	D, JG, P	-40°C to 125°C	7 mV

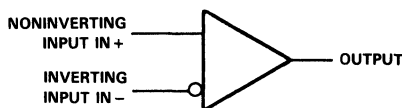
The D package is available in tape and reel. Add an R suffix when ordering. (e.g., LM393DR)



NC—No internal connection



symbol (each comparator)

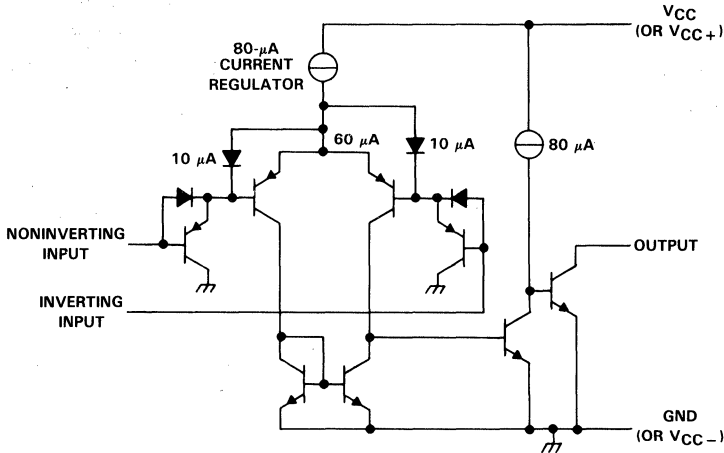


3

Voltage Comparators

LM193, LM293, LM393, LM293A, LM393A, LM2903 DUAL DIFFERENTIAL COMPARATORS

schematic (each comparator)



Current values shown are nominal.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	36 V
Differential input voltage (see Note 2)	± 36 V
Input voltage range (either input)	-0.3 V to 36 V
Output voltage	36 V
Output current	20 mA
Duration of output short-circuit to ground (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range:	
LM193	-55°C to 125°C
LM293, LM293A	-25°C to 85°C
LM393, LM393A	0°C to 70°C
LM2903	-40°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: L package	300°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the network ground terminal.
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input.
 3. Short circuits from outputs to V_{CC} can cause excessive heating and eventual destruction.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	25°C	464 mW	377 mW	—
FK	900 mW	11.0 mW/°C	68°C	880 mW	715 mW	275 mW
JG (LM193)	900 mW	8.4 mW/°C	43°C	672 mW	546 mW	210 mW
JG (All others)	825 mW	6.6 mW/°C	25°C	528 mW	429 mW	—
L	825 mW	6.6 mW/°C	25°C	528 mW	429 mW	165 mW
P	900 mW	8.0 mW/°C	37°C	640 mW	520 mW	—

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Voltage Comparators

electrical characteristics at specified free-air temperature, $V_{CC} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	LM193			LM293, LM393			LM293A, LM393A			LM2903			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{CC} = 5\text{ V}$ to 30 V , $V_{IC} = V_{ICR}$, $V_O = 1.4\text{ V}$	25°C	2	5	2	5	1	2	2	7	mV			
		Full range	9			9			4					
I_{IO} Input offset current	$V_O = 1.4\text{ V}$	25°C	3	25	5	50	5	50	5	50	nA			
		Full range	100			150			150					
I_{IB} Input bias current	$V_O = 1.4\text{ V}$	25°C	25	100	25	250	25	250	25	250	nA			
		Full range	300			400			400					
V_{ICR} Common-mode input voltage range‡		25°C	0 to $V_{CC}-1.5$			0 to $V_{CC}-1.5$			0 to $V_{CC}-1.5$			V		
		Full range	0 to $V_{CC}-2$			0 to $V_{CC}-2$			0 to $V_{CC}-2$					
A_{VD} Large-signal differential voltage amplification	$V_{CC} = 15\text{ V}$, $V_O = 1.4\text{ V}$ to 11.4 V , $R_L \geq 15\text{ k}\Omega$ to V_{CC}	25°C	50	200	50	200	50	200	25	100	V/mV			
I_{OH} High-level output current	$V_{OH} = 5\text{ V}$ $V_{ID} = 1\text{ V}$ $V_{OH} = 30\text{ V}$ $V_{ID} = 1\text{ V}$	25°C	0.1			0.1			0.1			nA		
		Full range	1			1			1					
V_{OL} Low-level output voltage	$I_{OL} = 4\text{ mA}$, $V_{ID} = -1\text{ V}$,	25°C	150	400	150	400	150	400	150	400	mV			
		Full range	700			700			700					
I_{OL} Low-level output current	$V_{OL} = 1.5\text{ V}$, $V_{ID} = -1\text{ V}$	25°C	6		6		6		6		mA			
I_{CC} Supply current	$R_L = \infty$	$V_{CC} = 5\text{ V}$	25°C	0.8	1	0.8	1	0.8	1	0.8	1	mA		
		$V_{CC} = 30\text{ V}$	Full range	2.5			2.5			2.5				

† Full range (MIN to MAX) for LM193 is -55°C to 125°C , for LM293 and LM293A is 25°C to 85°C , for the LM393 and LM393A is 0°C to 70°C , and for LM2903 is -40°C to 125°C . All characteristics are measured with zero common-mode input voltage unless otherwise specified.

‡ The voltage at either input or common-mode should not be allowed to go negative by more than 0.3 V . The upper end of the common-mode voltage range is $V_{CC+} - 1.5\text{ V}$, but either or both inputs can go to 30 V without damage.

switching characteristics, $V_{CC} = 5\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
Response time	R_L connected to 5 V through $5.1\text{ k}\Omega$, $C_L = 15\text{ pF}$, See Note 4	100-mV input step with 5-mV overdrive		1.3		μs
		TTL-level input step		0.3		

§ C_L includes probe and jig capacitance.

NOTE 4: The response time specified is the interval between the input step function and the instant when the output crosses 1.4 V .

LM193, LM293, LM393, LM293A, LM393A, LM2903
DUAL DIFFERENTIAL COMPARATORS

3

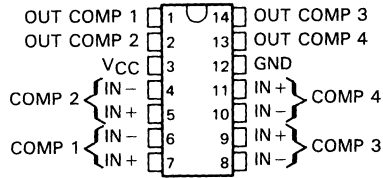
Voltage Comparators

LM3302 QUADRUPLE DIFFERENTIAL COMPARATOR

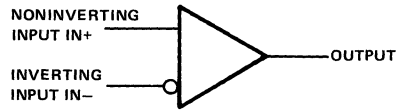
D2402, OCTOBER 1977—REVISED APRIL 1988

- Single Supply or Dual Supplies
- Wide Range of Supply Voltage . . . 2 to 28 Volts
- Low Supply Current Drain Independent of Supply Voltage . . . 0.8 mA Typ
- Low Input Bias Current . . . 25 nA Typ
- Low Input Offset Current . . . 3 nA Typ
- Low Input Offset Voltage . . . 3 mV Typ
- Common-Mode Input Voltage Range Includes Ground
- Differential Input Voltage Range Equal to Maximum-Rated Supply Voltage . . . ± 28 V
- Low Output Saturation Voltage
- Output Compatible with TTL, MOS, and CMOS

D, J, OR N PACKAGE
(TOP VIEW)



symbol (each comparator)



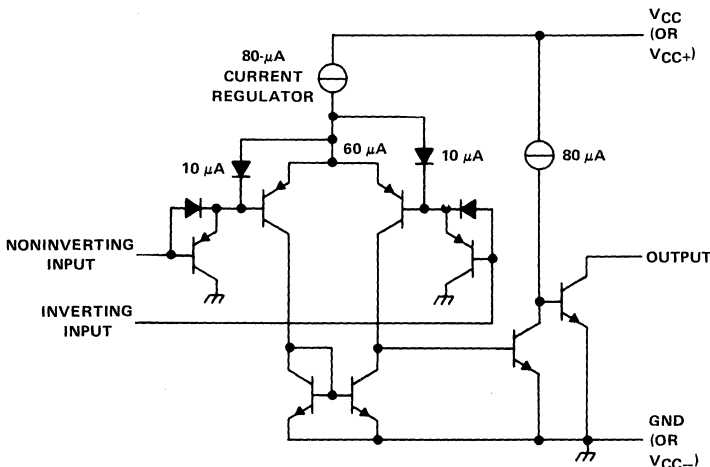
AVAILABLE OPTIONS

SYMBOLIZATION		OPERATING TEMPERATURE RANGE	V_{IO} MAX at 25°C
DEVICE	PACKAGE SUFFIX		
LM3302	D, J, N	-40°C to 85°C	20 mV

The D packages are available taped and reeled. Add the suffix R to the device type, when ordering. (i.e., LM3302DR)

description

This device consists of four independent voltage comparators that are designed to operate from a single power supply over a wide range of voltages. Operation from dual supplies is also possible so long as the difference between the two supplies is 2 V to 28 V and pin 3 is at least 1.5 V more positive than the input common-mode voltage. Current drain is independent of the supply voltage. The outputs can be connected to other open-collector outputs to achieve wired-AND relationships.



Current values shown are nominal.

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Voltage Comparators

LM3302 QUADRUPLE DIFFERENTIAL COMPARATOR

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	28 V
Differential input voltage (see Note 2)	± 28 V
Input voltage range (either input)	-0.3 V to 28 V
Output voltage	28 V
Output current	20 mA
Duration of output short-circuit to ground (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	-40°C to 85°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the network ground terminal.
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. Short circuits from the output to V_{CC} can cause excessive heating and eventual destruction.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW
J	1025 mW	8.2 mW/°C	656 mW	533 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW

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Voltage Comparators

electrical characteristics at specified free-air temperature, $V_{CC} = 5$ V (unless otherwise noted)

PARAMETER		TEST CONDITIONS†	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	$V_{CC} = 5$ V to 28 V, $V_{IC} = V_{ICR}$ min, $V_O = 1.4$ V	25°C	3	20	mV
			-40°C to 85°C		40	
I_{IO}	Input offset current	$V_O = 1.4$ V	25°C	3	100	nA
			-40°C to 85°C		300	
I_{IB}	Input bias current		25°C	-25	-500	nA
			-40°C to 85°C		-1000	
V_{ICR}	Common-mode input voltage range		25°C	0 to $V_{CC}-1.5$		V
			-40°C to 85°C		0 to $V_{CC}-2$	
A_{VD}	Large-signal differential voltage amplification	$V_{CC} = 15$ V, $V_O = 1.4$ V to 11.4 V, $R_L = 15$ k Ω to V_{CC}	25°C	2	30	V/mV
I_{OH}	High-level output current	$V_{ID} = 1$ V, $V_{OH} = 5$ V	25°C	0.1		nA
			-40°C to 85°C			1
V_{OL}	Low-level output voltage	$V_{ID} = 1$ V, $I_{OL} = 4$ mA	25°C	150	500	mV
			40°C to 85°C		700	
I_{OL}	Low-level output current	$V_{ID} = 1$ V, $V_{OL} = 1.5$ V	25°C	6	16	mA
I_{CC}	Supply current (four comparators)	$V_O = 2.5$ V, No load	25°C	0.8	2	mA

† All characteristics are measured with zero common-mode input voltage unless otherwise specified.

LM3302
QUADRUPLE DIFFERENTIAL COMPARATOR

switching characteristics, $V_{CC} = 5\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Response time	$R_L = 5.1\text{ k}\Omega$ to 5 V, $C_L = 15\text{ pF}^\ddagger$, See Note 4		1.3		μs
		100-mV input step with 5-mV overdrive TTL-level input step		0.3	

$^\ddagger C_L$ includes probe and jig capacitance.

NOTE 4: The response time specified is the interval between the input step function and the instant when the output crosses 1.4 V.



Voltage Comparators

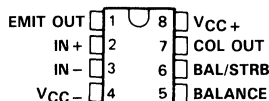
LP111, LP211, LP311

LOW-POWER DIFFERENTIAL COMPARATORS WITH STROBES

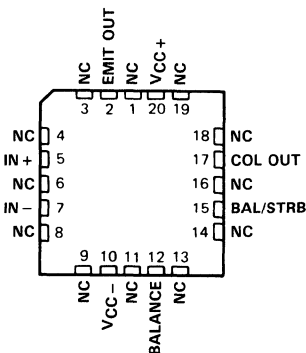
D3019, JUNE 1987—REVISED MAY 1988

- Low Power Drain — 900 μ W Typical with 5-V Supply
- Operates from ± 15 V or from a Single Supply as Low as 3 V
- Output Drive Capability of 25 mA
- Emitter Output Can Swing Below Negative Supply
- Response Time — 1.2 μ s Typ
- Low Input Currents:
Offset Current . . . 2 nA Typ
Bias Current . . . 15 nA Typ
- Wide Common-Mode Input Range:
— 14.5 V to 13.5 V with ± 15 -V Supply
- Same Pinout as LM111, LM211, LM311
- Designed to be Interchangeable with National Semiconductor LP311

LP111 . . . JG DUAL-IN-LINE PACKAGE
LP211, LP311 . . . D, JG, OR P PACKAGE
(TOP VIEW)



LP111 . . . FK CHIP CARRIER PACKAGE
(TOP VIEW)

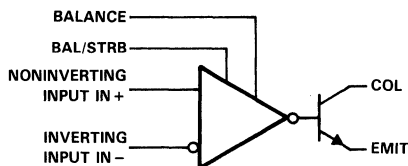


NC—No internal connection

description

The LP111, LP211, and LP311 are a low-power versions of the industry-standard LM111, LM211, and LM311. They take advantage of stable, high-value, ion-implanted resistors to perform the same function as the LM311 series, with a 30:1 reduction in power consumption but only a 6:1 slowdown in response time. Thus, they are well-suited for battery-powered applications and all other applications where fast response times are not needed. They operate over a wide range of supply voltages, from ± 18 V down to a single 3-V supply with less than 300 μ A current drain, but are still capable

functional block diagram



AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE			
		SMALL OUTLINE (D)	CERAMIC (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	7.5 mV	LP311D	—	LP311JG	LP311P
-25°C to 85°C	7.5 mV	LP211D	—	LP211JG	LP211P
-55°C to 125°C	7.5 mV	—	LP111FK	LP111JG	—

The D package is available taped and reeled. Add the suffix R to the device type when ordering, (e.g., LP311DR)

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Voltage Comparators

LP111, LP211, LP311

LOW-POWER DIFFERENTIAL COMPARATORS WITH STROBES

description (continued)

of driving a 25-mA load. The LP111, LP211, and LP311 are quite easy to apply free of oscillation if ordinary precautions are taken to minimize stray coupling from the output to either input or to the trim pins.

The LP111 is characterized for operation over the full military temperature range of -55°C to 125°C . The LP211 is characterized for operation from -25°C to 85°C , and the LP311 is characterized for operation from 0°C to 70°C .

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	18 V
Supply voltage, V_{CC-} (see Note 1)	-18 V
Differential input voltage (see Note 2)	± 30 V
Input voltage (either input, see Notes 1 and 3)	± 15 V
Voltage from emitter output to V_{CC-}	30 V
Voltage from collector output to V_{CC-}	40 V
Voltage from collector output to emitter output	40 V
Duration of output short-circuit (see Note 4)	40 V
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: LP111	-55°C to 125°C
LP211	-25°C to 85°C
LP311	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential input voltages are at the noninverting input terminal with respect to the inverting terminal.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage of ± 15 V, whichever is less.
 4. The output may be shorted to ground or to either power supply.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^{\circ}\text{C}$ POWER RATING	$T_A = 85^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
D	500 mW	5.8 mW/ $^{\circ}\text{C}$	64°C	464 mW	377 mW	—
FK	1375 mW	11.0 mW/ $^{\circ}\text{C}$	25°C	880 mW	715 mW	275 mW
JG (LP111)	1050 mW	8.4 mW/ $^{\circ}\text{C}$	25°C	672 mW	546 mW	210 mW
JG (LP_111)	825 mW	6.6 mW/ $^{\circ}\text{C}$	25°C	528 mW	429 mW	—
P	500 mW	8.0 mW/ $^{\circ}\text{C}$	88°C	500 mW	500 mW	—

recommended operating conditions

	MIN	NOM	MAX	UNITS
Input voltage ($ V_{CC\pm} \leq 15$ V)	$V_{CC-} + 0.5$		$V_{CC+} - 1.5$	V
Supply voltage, $V_{CC+} - V_{CC-}$	3.5		30	V

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Voltage Comparators

LP111, LP211, LP311
LOW-POWER DIFFERENTIAL COMPARATORS WITH STROBES

electrical characteristics at specified free-air temperature, $V_{CC\pm} = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP [†]	MAX	UNIT
V_{ID}	Input offset voltage	RS < 100 k Ω , See Note 5	25°C		2	7.5	mV
			Full Range			10	
I_{IO}	Input offset current	See Note 5	25°C		2	25	nA
			Full Range			35	
I_{IB}	Input bias current		25°C		15	100	nA
			Full Range			150	
V_{OL}	Low-level output voltage	$V_{ID} > 10\text{ mV}$, See Note 6 $I_{OL} = 25\text{ mA}$, $V_{CC} = 4.5\text{ V}$, $V_{CC-} = 0$, $V_{ID} < -10\text{ mV}$, $I_{OL} = 1.6\text{ mA}$, See Note 6	25°C		0.4	1.5	V
			Full Range	LP111	0.1	0.7	
				LP211 LP311	0.1	0.4	
	Low-level strobe current	$V_{(strobe)} = 0.3\text{ V}$, $V_{ID} < -10\text{ mV}$, See Note 7	25°C		100	300	μA
$I_{O(off)}$	Output off-state current	$V_{ID} > 10\text{ mV}$, $V_{CE} = 35\text{ V}$	25°C		0.2	100	nA
A_{VD}	Large signal differential voltage amplification	$R_L = 5\text{ k}\Omega$	25°C		40	100	V/mV
I_{CC+}	Supply current from V_{CC+}	$V_{ID} = -50\text{ mV}$, $R_L = \infty$	Full Range		150	300	μA
I_{CC-}	Supply current from V_{CC-}	$V_{ID} = 50\text{ mV}$, $R_L = \infty$	Full Range		-80	-180	μA

[†]All typical values are at $V_{CC\pm} = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$.

NOTES: 5. The offset voltages and offset currents given are the maximum values required to drive the output within 1 V of either supply with a 1-mA load. Thus, these parameters define an error band and take into account the worst-case effects of voltage gain and input impedance.

6. Voltages are with respect to EMIT OUT and V_{CC-} tied together.

7. The strobe should not be shorted to ground; it should be current driven at 100 μA to 300 μA .

switching characteristics at $V_{CC\pm} = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Response time	See Note 8		1.2		μs

NOTE 8: The response time is specified for a 100-mV input step with 5-mV overdrive.

3

Voltage Comparators

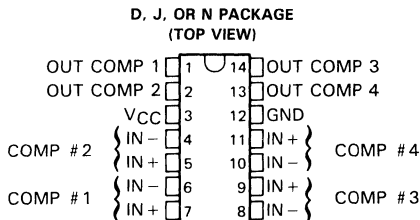


Voltage Comparators

LP239, LP339, LP2901 LOW-POWER QUAD DIFFERENTIAL COMPARATORS

D3044, OCTOBER 1987—REVISED MAY 1988

- **Ultralow Power Supply Current**
Drain . . . Typically 60 μ A
- **Low Input Biasing Current** . . . 3 nA
- **Low Input Offset Current** . . . ± 0.5 nA
- **Low Input Offset Voltage** . . . ± 2 mV
- **Common-Mode Input Voltage Includes Ground**
- **Output Voltage Compatible with MOS and CMOS Logic**
- **High Output Sink-Current Capability**
(30 mA at $V_O = 2$ V)
- **Power Supply Input Reverse-Voltage Protected**
- **Single-Power-Supply Operation**
- **Pin-for-Pin Compatible with LM239, LM339, LM2901**



description

The LP239, LP339, and LP2901 are low-power quadruple differential comparators. Each device consists of four independent voltage comparators designed specifically to operate from a single power supply and typically to draw 60- μ A drain current over a wide range of voltages. Operation from split power supplies is also possible and the ultralow power supply drain current is independent of the power supply voltage.

Applications include limit comparators, simple analog-to-digital converters, pulse generators, squarewave generators, time delay generators, voltage controlled oscillators, multivibrators, and high-voltage logic gates. The LP239, LP339, and LP2901 were specifically designed to interface with the CMOS logic family. The ultralow power supply current makes these products desirable in battery-powered applications.

The LP239 is characterized for operation from -25°C to 85°C . The LP339 is characterized for operation from 0°C to 70°C . The LP2901 is characterized for operation from -40°C to 85°C .

AVAILABLE OPTIONS

T_A	V_{IO} MAX AT 25°C	PACKAGE		
		SMALL-OUTLINE (D)	PLASTIC DIP (N)	CERAMIC DIP (J)
0°C to 70°C	± 5 mV	LP339D	LP339N	LP339J
-25°C to 85°C	± 5 mV	LP239D	LP239N	LP239J
-40°C to 85°C	± 5 mV	LP2901D	LP2901N	LP2901J

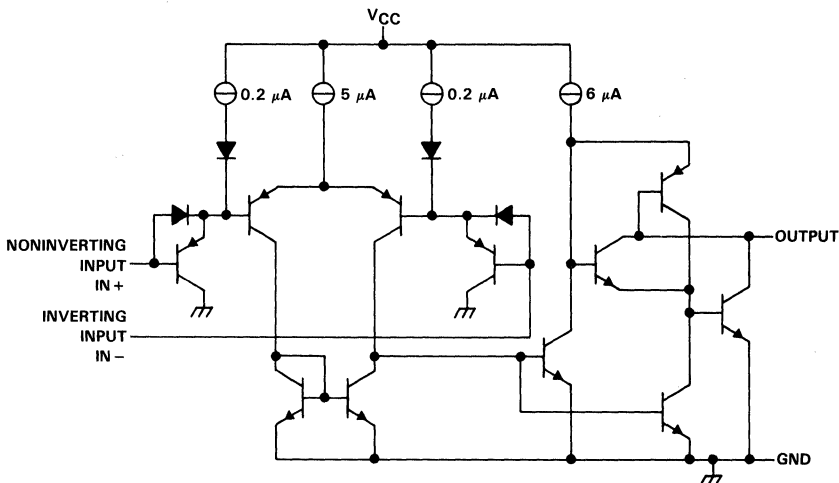
D packages are available taped-and-reeled. Add "R" suffix to device type when ordering (e.g., LP339DR).

3

Voltage Comparators

LP239, LP339, LP2901 LOW-POWER QUAD DIFFERENTIAL COMPARATORS

schematic diagram (each comparator)



3

Voltage Comparators

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	36 V
Differential input voltage, V_{ID} (see Note 2)	± 36 V
Input voltage range (either input)	-0.3 V to 36 V
Input current, $I_I \leq -0.3$ V (see Note 3)	-50 mA
Duration of output short-circuit to ground (see Note 4)	unlimited
Continuous total dissipation (see Note 5)	See Dissipation Rating Table
Operating free-air temperature range:	
LP239	-25°C to 85°C
LP339	0°C to 70°C
LP2901	-40°C to 85°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

- NOTES:
1. All voltage values, except differential voltages, are with respect to the network ground terminal.
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. This input current only exists when the voltage at any of the inputs is driven negative. The current flows through the collector-base junction of the input clamping device. In addition to the clamping device action, there is lateral n-p-n parasitic transistor action. This action is not destructive and normal output states are re-established when the input voltage returns to a value more positive than -0.3 V at $T_A = 25^\circ\text{C}$.
 4. Short circuits between outputs to V_{CC} can cause excessive heating and eventual destruction.
 5. If the output transistors are allowed to saturate, the low bias dissipation and the on-off characteristics of the outputs keep the dissipation very small (usually less than 100 mW).

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW
J	1025 mW	8.2 mW/°C	656 mW	533 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW

LP239, LP339, LP2901 LOW-POWER QUAD DIFFERENTIAL COMPARATORS

recommended operating conditions

	LP2901			LP239			LP339			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, V_{CC}	5		30	5		30	5		30	V
Common-mode input voltage, V_{IC}	$V_{CC} = 5\text{ V}$	0	3	0	3	0	3	0	3	V
	$V_{CC} = 30\text{ V}$	0	28	0	28	0	28	0	28	V
Input voltage, V_I	$V_{CC} = 5\text{ V}$	0	3	0	3	0	3	0	3	V
	$V_{CC} = 30\text{ V}$	0	28	0	28	0	28	0	28	V
Operating free-air temperature, T_A	-40		85	-25		85	0		70	°C

electrical characteristics, $V_{CC} = 5\text{ V}$, $T_A = 25\text{ °C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
V_{IO} Input offset voltage	$V_{CC} = 5\text{ V to }30\text{ V}$, $V_O = 2\text{ V}$, $R_S = 0$, See Note 6	25°C		± 2	± 5	mV
		Full range			± 9	
I_{IO} Input offset current		25°C		± 0.5	± 5	nA
		Full range		± 1	± 15	
I_{IB} Input bias current	See Note 7	25°C		- 2.5	- 25	nA
		Full range		- 4	- 40	
V_{ICR} Common-mode input voltage range	Single supply	25°C	0 to $V_{CC} - 1.5$			V
		Full range	0 to $V_{CC} - 2$			
A_{VD} Large-signal differential voltage amplification	$V_{CC} = 15\text{ V}$, $R_L = 15\text{ k}\Omega$			500		V/mV
Output sink current	$V_{I-} = 1\text{ V}$, $V_{I+} = 0$	$V_O = 2\text{ V}$ (see Note 8)	25°C	20	30	mA
		Full range		15		
		$V_O = 0.4\text{ V}$	25°C	0.2	0.7	
Output leakage current	$V_{I+} = 1\text{ V}$, $V_{I-} = 0$	$V_O = 5\text{ V}$	25°C	0.1		nA
		Full range				1
V_{ID} Differential input voltage	$V_I \leq 0$ (or V_{CC-} on split supplies)				36	V
I_{CC} Supply current	$R_L = \infty$ all comparators			60	100	µA

NOTES: 6. V_{IO} is measured over the full common-mode input voltage range.

7. Because of the p-n-p input stage, the direction of the current is out of the device. This current is essentially constant (i.e., independent of the output state). Therefore, no loading change exists on the reference or input lines as long as the common-mode input voltage range is not exceeded.

8. The output sink current is a function of the output voltage. These devices have a bimodal output section that allows them to sink (via a Darlington connection) large currents at output voltages greater than 1.5 V, and smaller currents at output voltages less than 1.5 V.

switching characteristics, $V_{CC} = 5\text{ V}$, $T_A = 25\text{ °C}$, R_L connected to 5 V through 5.1 kΩ

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Large-signal response time	TTL logic swing, $V_{ref} = 1.4\text{ V}$		1.3		µs
Response time			8		µs

3

Voltage Comparators

LP239, LP339, LP2901 LOW-POWER QUAD DIFFERENTIAL COMPARATORS

TYPICAL APPLICATION DATA

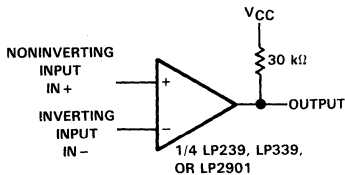


FIGURE 1. BASIC COMPARATOR

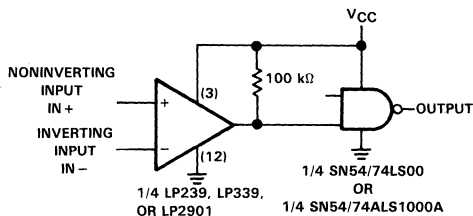


FIGURE 2. CMOS DRIVER

All pins of any unused comparators should be grounded. The bias network of the LP239, LP339, and LP2901 establishes a drain current that is independent of the magnitude of the power supply voltage over the range of 2 V to 30 V. It is usually necessary to use a bypass capacitor across the power supply line.

The differential input voltage may be larger than V_{CC} without damaging the device. Protection should be provided to prevent the input voltages from going negative by more than -0.3 V. The output section has two distinct modes of operation: a Darlington mode and a grounded-emitter mode. This unique drive circuit permits the device to sink 30 mA at $V_O = 2$ V in the Darlington mode and $700 \mu\text{A}$ at $V_O = 0.4$ V in the ground-emitter mode. Figure 3 is a simplified schematic diagram of the output section. The output section is configured in a Darlington connection (ignoring Q3). Therefore, if the output voltage is held high enough (above 1 V), Q1 is not saturated and the output current is limited only by the product of the h_{FE} of Q1, the h_{FE} of Q2, and I1 and by the $60\text{-}\Omega$ saturation resistance of Q2. The devices are capable of driving LEDs, relays, etc., in this mode while maintaining an ultralow power supply current of $60 \mu\text{A}$ typically.

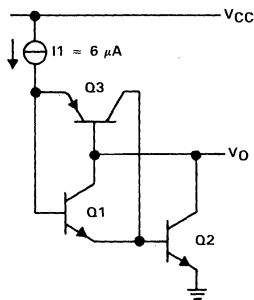


FIGURE 3. OUTPUT SECTION SCHEMATIC DIAGRAM

Without transistor Q3, if the output voltage were allowed to drop below 0.8 V, transistor Q1 would saturate and the output current would drop to zero. The circuit would be unable to pull low current loads down to ground or the negative supply, if used. Transistor Q3 has been included to bypass transistor Q1 under these conditions and apply the current I1 directly to the base of Q2. The output sink current is now approximately I1 times the h_{FE} of Q2 ($700 \mu\text{A}$ at $V_O = 0.4$ V). The output of the devices exhibit a bimodal characteristic with a smooth transition between modes.

In both cases, the output is an uncommitted collector. Therefore several outputs can be tied together to provide a dot logic function. An output pull-up resistor can be connected to any available power supply voltage within the permitted power supply voltage range, and there is no restriction on this voltage based on the magnitude of the voltage that is applied to the V_{CC} terminal of the package.

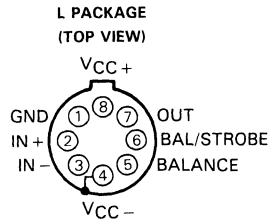
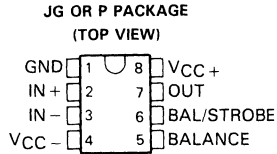
3

Voltage Comparators

LT1011, LT1011A VOLTAGE COMPARATORS

D3179, JANUARY 1989

- Low Input Offset Voltage . . . 1.5 or 0.5 mV Max
- Maximum Input Bias Current . . . 50 or 25 nA
- Low Input Offset Current . . . 4 or 3 nA Max
- Output Response Time . . . 250 ns Max
- Voltage Gain . . . 200 V/mV Min
- Output Current . . . 50 mA Source or Sink
- Differential Input Voltage . . . ± 30 V
- Can Operate from Single 5-V Supply
- Pin-Compatible with LM111 Series
- Designed to be Interchangeable with Linear Technology LT1011 and LT1011A



Pin 4 (L package) is in electrical contact with the case.

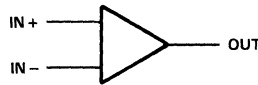
description

The LT1011 and LT1011A are general-purpose comparators that are pin-compatible with the LM111. The LT1011A offers significantly better input characteristics than the LM111: four times lower bias current, six times lower offset voltage, and five times higher voltage gain. Additionally, the supply current is considerably lower than that of the LM111 with no loss in speed. The offset voltage temperature coefficient of the LT1011A is $15 \mu\text{V}/^\circ\text{C}$. The LT1011 and LT1011A are fully specified for dc parameters and output response time when operating from a single 5-V supply.

The LT1011 and LT1011A can be used in high-accuracy (≥ 12 -bit) systems without trimming. The devices retain all the versatile features of the LM111 including single-supply operation (3 V to 36 V) or dual-supply operation (± 1.5 V to ± 18 V) and a floating transistor output with 50-mA source or sink capability. The devices can drive loads that are referenced to ground, the negative supply, or the positive supply, and are specified up to 50 V between V_{CC-} and the collector output. A differential input voltage up to the full supply voltage is allowed, even with ± 18 -V supplies, enabling the inputs to be clamped to the supplies with simple diode clamps.

M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . C-suffix devices are characterized for operation from 0°C to 70°C .

symbol



AVAILABLE OPTIONS

T _A	V _{IO} MAX at 25°C	PACKAGE		
		CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)
0°C to 70°C	1.5 mV 0.5 mV	LT1011CJG LT1011ACJG	LT1011CL LT1011ACL	LT1011CP LT1011ACP
-55°C to 125°C	1.5 mV 0.5 mV	LT1011MJG LT1011AMJG	LT1011ML LT1011AML	

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

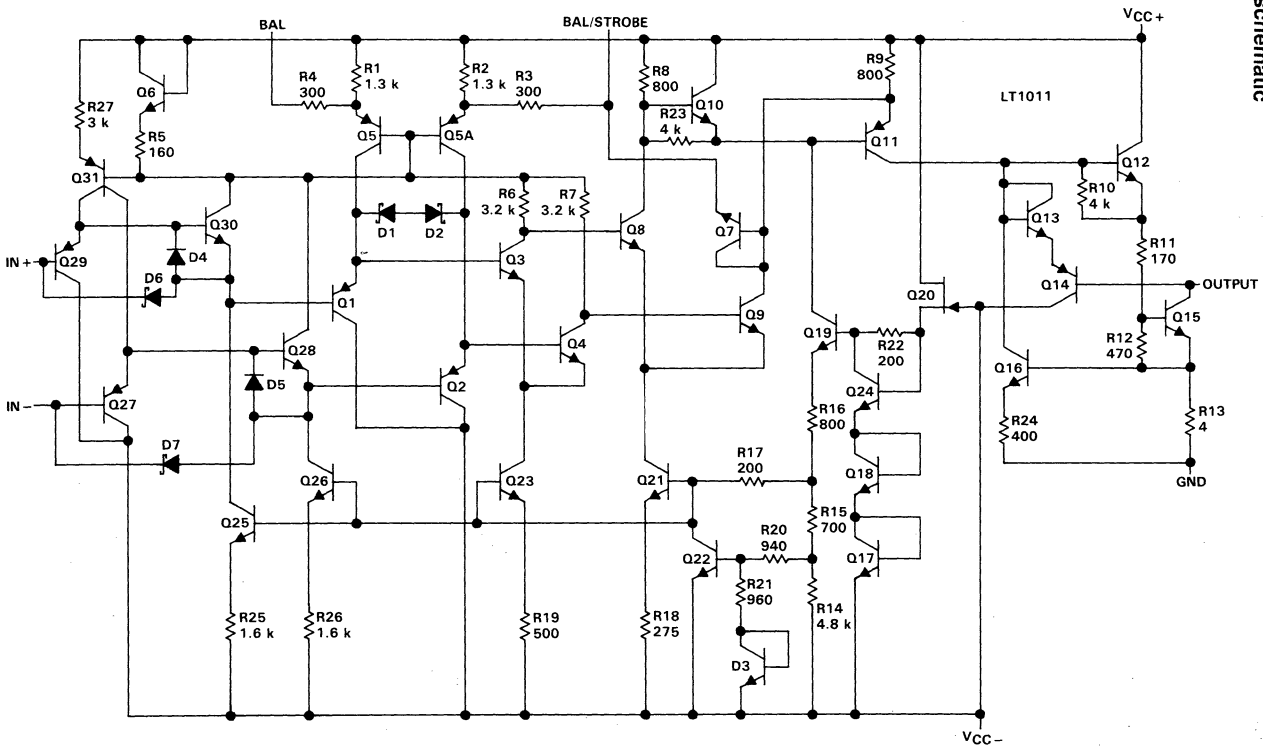
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LT1011, LT1011A VOLTAGE COMPARATORS

Schematic



Resistor values shown are nominal and in ohms.

Voltage Comparators



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LT1011, LT1011A VOLTAGE COMPARATORS

electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$, $V_{IC} = 0$, $R_S = 0$, pin 1 at V_{CC-} , output at pin 7 (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A^\dagger	LT1011			LT1011A			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$I_O = 1.5\text{ mA}$, $V_O = 0$	25°C	0.6 1.5			0.3 0.5			mV
		Full range	3			1			
	$R_S \leq 50\text{ k}\Omega$, See Note 5	25°C	2			0.75			
		Full range	3			1.5			
α_{VIO} Average temperature coefficient of input offset voltage	See Note 6	Full range	4 25			4 15			$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current	See Note 5	25°C	0.2 4			0.2 3			nA
		Full range	6			5			
I_{IB} Input bias current	$I_O = 1.5\text{ mA}$, $V_O = 0$	25°C	-20 \pm 50			-15 \pm 25			nA
		25°C	-25 \pm 65			-20 \pm 35			
	See Note 5	Full range	\pm 80			\pm 50			
$I_{IL(S)}$ Low-level strobe current (See Note 7)		25°C	-500			-500			μA
V_{ICR} Common-mode input voltage range		Full range	-14.5 to 13			-14.5 to 13			V
A_{VD} Large-signal differential voltage amplification	$R_L = 1\text{ k}\Omega$ to V_{CC+} , $V_O = -10\text{ V}$ to 14.5 V	25°C	200 500			200 500			V/mV
V_{OL} Low-level output voltage	$V_{ID} = -5\text{ mV}$, $I_{OL} = 8\text{ mA}$, Pin 1 at 0 V	Full range	0.4			0.4			V
	$V_{ID} = -5\text{ mV}$, $I_{OL} = 50\text{ mA}$, Pin 1 at 0 V	Full range	1.5			1.5			
$I_{O(Ik)}$ Output leakage current	$V_{ID} = 5\text{ mV}$, Pin 1 at -15 V, $V_O = 35\text{ V}$ (25 V for LT1011C)	25°C	0.2 10			0.2 10			nA
		Full range	500			500			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\text{ min}}$, $R_S \leq 50\text{ k}\Omega$	25°C	90 115			94 115			dB
I_{CC+} Supply current from V_{CC+}		25°C	3.2 4			3.2 4			mA
I_{CC-} Supply current from V_{CC-}		25°C	-1.7 -2.5			-1.7 -2.5			mA
C_i Input capacitance		25°C	6			6			pF

† Full range is -55°C to 125°C for the LT1011M and LT1011AM. Full range is 0°C to 70°C for the LT1011C and LT1011AC.

NOTES: 5. These specifications apply for single supply voltages from 5 V to 30 V and dual supply voltages from $\pm 2.5\text{ V}$ to $\pm 15\text{ V}$ for the entire input voltage range, and for both high and low output states. The high state is $I_{OH} \geq 100\text{ }\mu\text{A}$ and $V_O \geq (V_{CC+} - 1\text{ V})$. The low state is $I_{OL} \leq 8\text{ mA}$ and $V_O \leq 0.8\text{ V}$. Therefore, this specification defines a worst-case error band that includes effects due to common-mode signals, voltage gain, and output load.

6. Average temperature coefficient is calculated by dividing the offset voltage difference measured at minimum and maximum temperatures by the temperature difference.

7. This is the minimum current that must be drawn from the strobe to ensure that the output is off regardless of differential input voltage.

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Voltage Comparators

LT1011, LT1011A VOLTAGE COMPARATORS

electrical characteristics, $V_{CC+} = 5\text{ V}$, $V_{CC-} = 0$, $V_{IC} = 0$, $R_S = 0$, pin 1 at 0 V, output at pin 7 (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A^\dagger	LT1011			LT1011A			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$R_S \leq 50\text{ k}\Omega$, See Note 5	25°C			2		0.75	mV	
		Full range			3		1.5		
I_{IO} Input offset current	See Note 8	25°C		0.2	4	0.2	3	nA	
		Full range			6		5		
I_{IB} Input bias current	See Note 8	25°C		25	65	20	35	nA	
		Full range			80		50		
$I_{IL(S)}$ Low-level strobe current (See Note 7)		25°C			-500		-500	μA	
V_{ICR} Common-mode input voltage range		Full range	0.5 to 3			0.5 to 3		V	
A_{VD} Large-signal differential voltage amplification	$R_L = 0.5\text{ k}\Omega$ to V_{CC+} , $V_O = 0.5\text{ V}$ to 4.5 V	25°C	50	300		50	300	V/mV	
V_{OL} Low-level output voltage	$V_{ID} = -5\text{ mV}$, $I_{OL} = 8\text{ mA}$	Full range			0.4		0.4	V	
	$V_{ID} = -5\text{ mV}$, $I_{OL} = 50\text{ mA}$	Full range			1.5		1.5		
I_O Output leakage current	$V_{ID} = 5\text{ mV}$, $V_O = 50\text{ V}$ (40 V for LT1011C)	25°C		0.2	10	0.2	10	nA	
		Full range			500		500		
I_{CC+} Supply current from V_{CC+}		25°C		3.2	4	3.2	4	mA	
I_{CC-} Supply current from V_{CC-}		25°C		-1.7	-2.5	-1.7	-2.5	mA	

† Full range is -55°C to 125°C for the LT1011M and LT1011AM. Full range is 0°C to 70°C for the LT1011C and LT1011AC.

NOTES: 6. Average temperature coefficient is calculated by dividing the offset voltage difference measured at minimum and maximum temperatures by the temperature difference.

7. This is the minimum current that must be drawn from the strobe to ensure that the output is off regardless of differential input voltage.
 8. These specifications apply for all single-supply voltages from 5 V to 30 V for the entire input voltage range, and for both high and low output states. The high state is $I_{OH} \geq 100\text{ }\mu\text{A}$ and $V_O \geq (V_{CC+} - 1\text{ V})$. The low state is $I_{OL} \leq 8\text{ mA}$ and $V_O \leq 0.8\text{ V}$. Therefore, this specification defines a worst-case error band that includes effects due to common-mode signals, voltage gain, and output load.

switching characteristics, $V_{CC+} = 5\text{ V}$, $V_{CC-} = 0$, pin 1 at 0 V, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	LT1011			LT1011A			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
Output response time	$R_C = 500\text{ }\Omega$ to 5 V, $C_L = 5\text{ pF}$, See Note 9		150	250		150	250	ns

NOTE 9: The response time specified is for a 100-mV input step with 5-mV overdrive and is the interval between the input step function and the instant when the output crosses 1.4 V.

Voltage Comparators

TYPICAL CHARACTERISTICS†

INPUT OFFSET VOLTAGE
vs
COMMON-MODE INPUT VOLTAGE

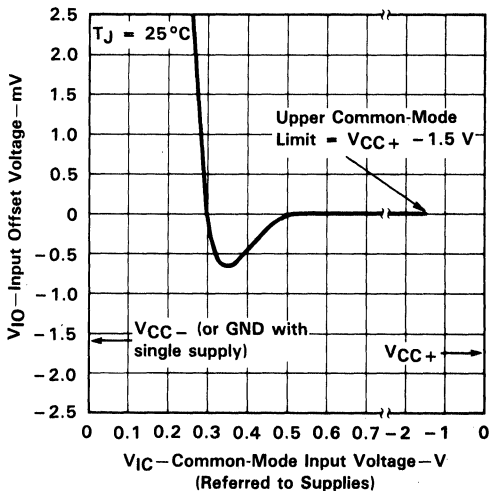


FIGURE 1

OFFSET ADJUSTMENT CHARACTERISTICS

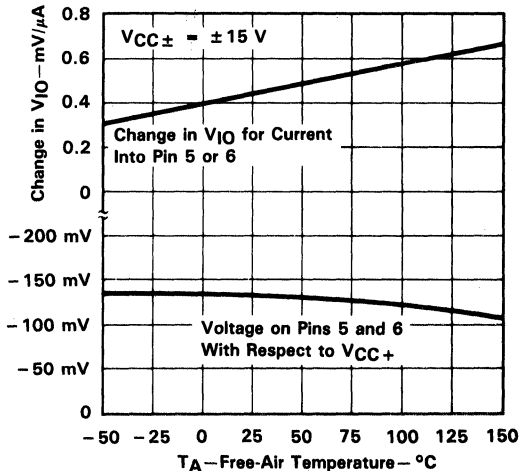


FIGURE 2

INPUT OFFSET CURRENT
vs
FREE-AIR TEMPERATURE

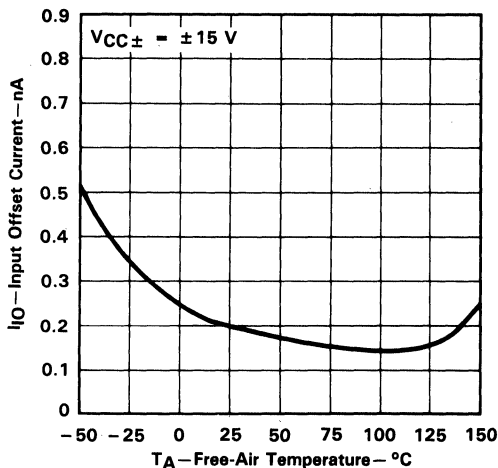


FIGURE 3

LT1011
INPUT BIAS CURRENT
vs
FREE-AIR TEMPERATURE

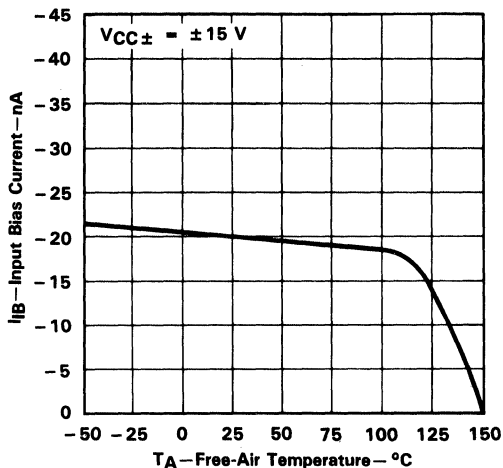


FIGURE 4

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

INPUT VOLTAGE LIMITS
vs
FREE-AIR TEMPERATURE

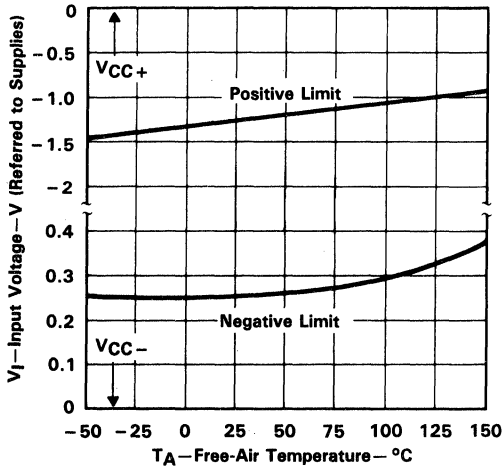


FIGURE 5

LT1011
INPUT CHARACTERISTICS
(EITHER INPUT WITH OTHER
INPUT GROUNDED)

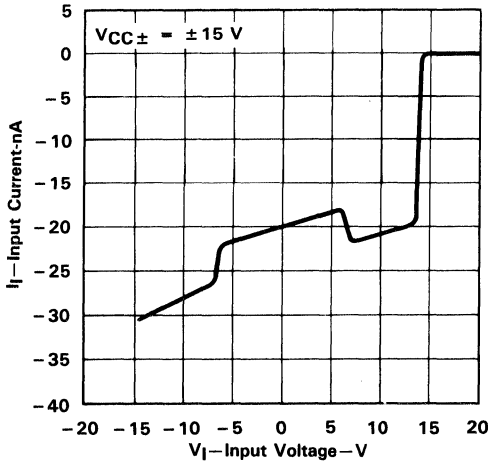


FIGURE 6

EQUIVALENT OFFSET VOLTAGE
vs
SOURCE RESISTANCE

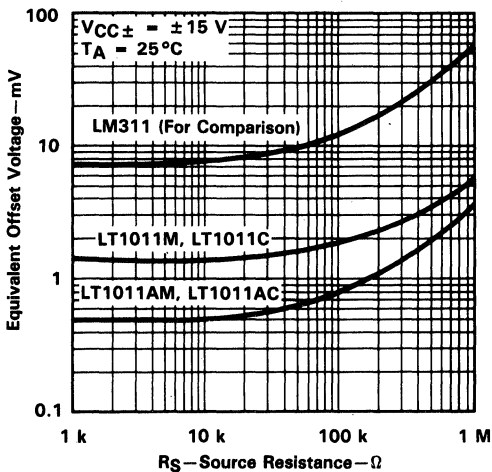


FIGURE 7

VOLTAGE TRANSFER CHARACTERISTICS

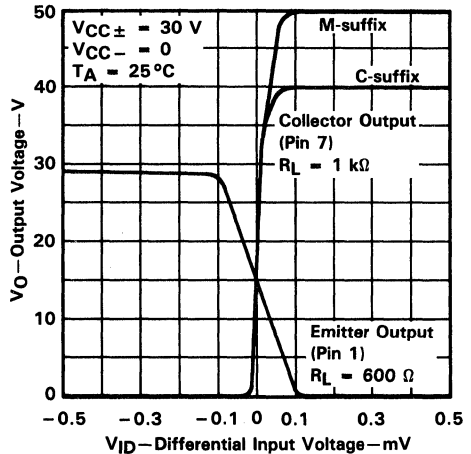


FIGURE 8

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

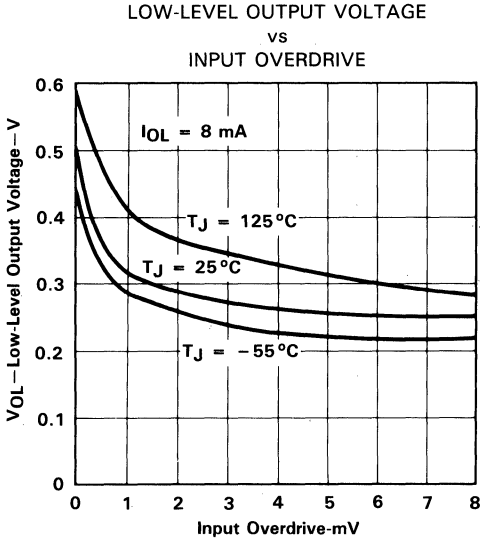


FIGURE 9

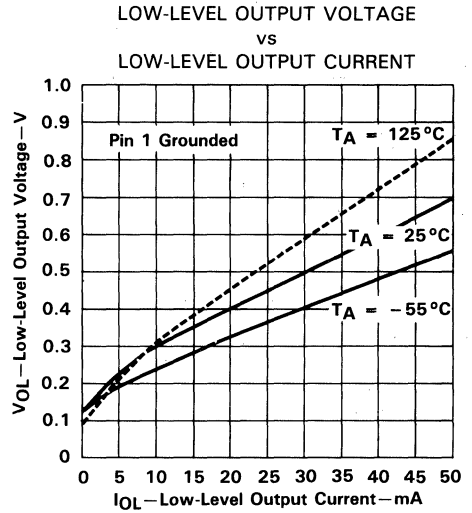


FIGURE 10

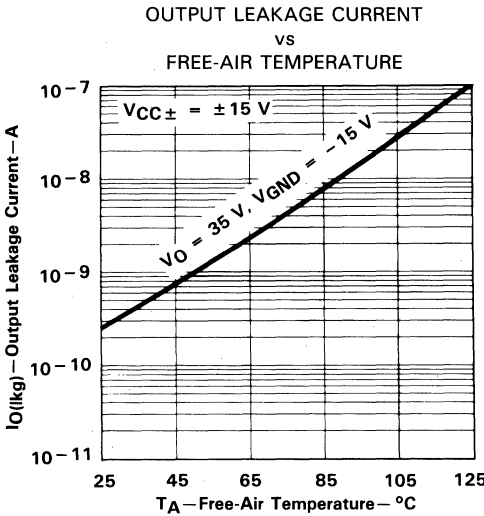


FIGURE 11

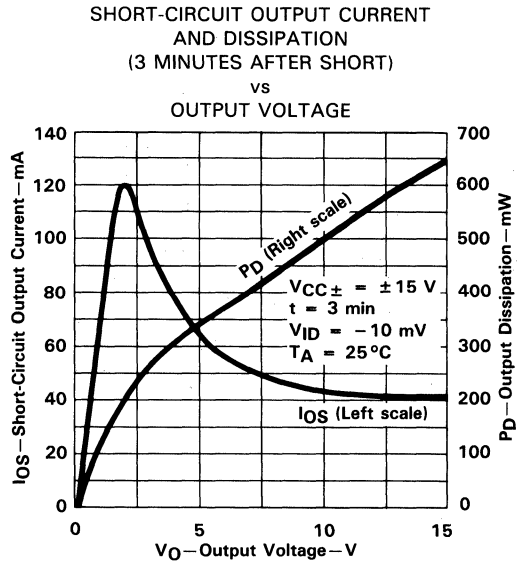


FIGURE 12

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

SUPPLY CURRENT
vs
SUPPLY VOLTAGE

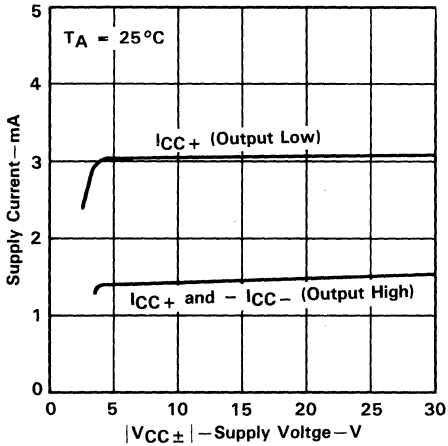


FIGURE 13

SUPPLY CURRENT
vs
FREE-AIR TEMPERATURE

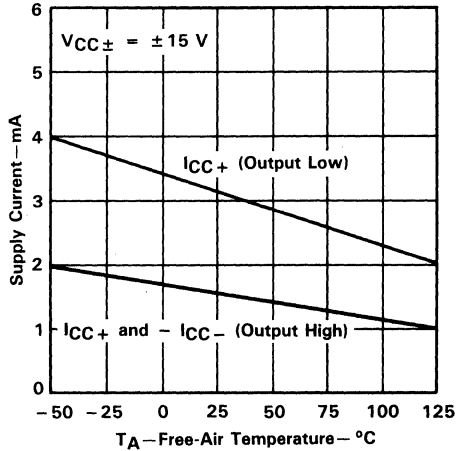


FIGURE 14

OUTPUT SATURATION VOLTAGE
(GROUND INPUT)
vs
OUTPUT CURRENT

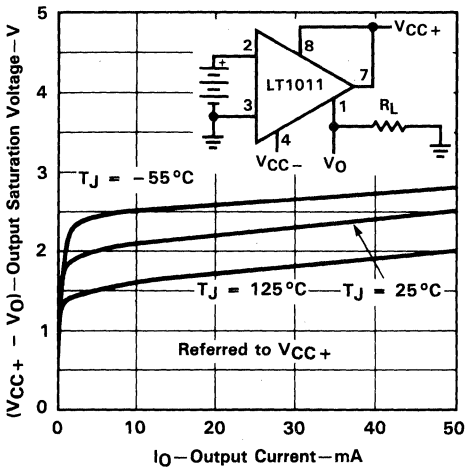


FIGURE 15

OUTPUT RESPONSE TIME
vs
INPUT STEP SIZE

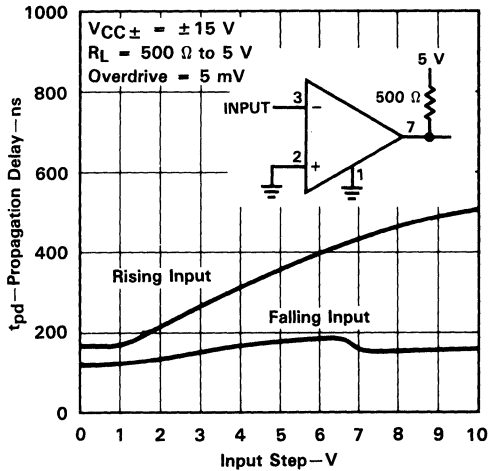


FIGURE 16

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

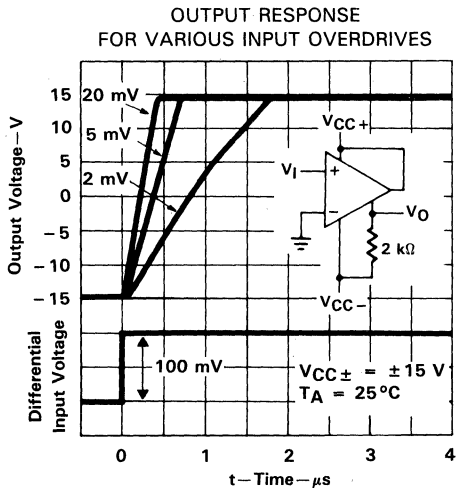


FIGURE 17

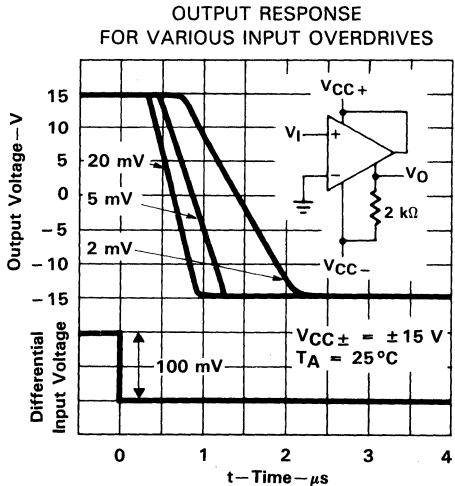


FIGURE 18

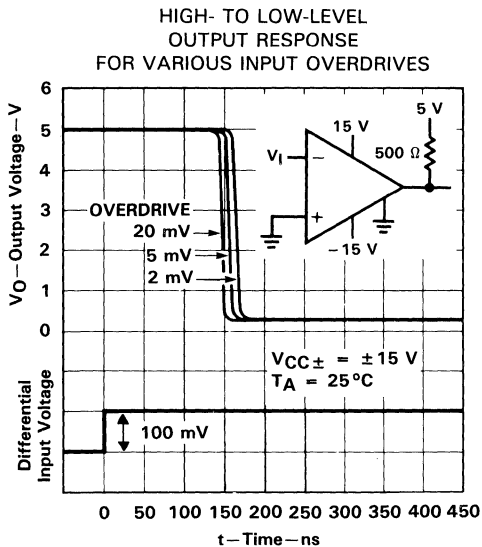


FIGURE 19

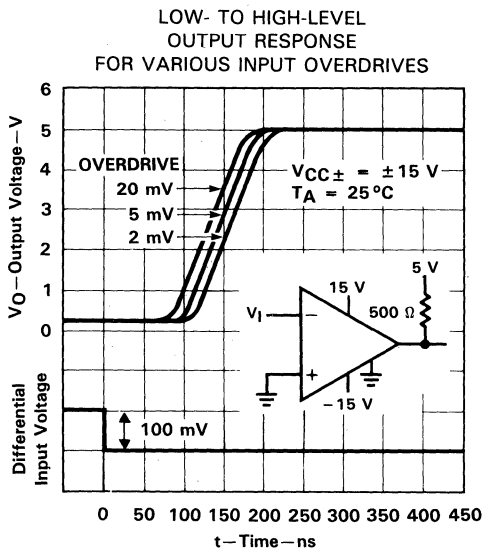


FIGURE 20

TYPICAL APPLICATION DATA

preventing oscillation problems

Oscillation problems in comparators are often caused by stray capacitance between the output and inputs or between the output and other sensitive pins on the comparator. This is especially true for comparators with high gain and wide bandwidth, like the LT1011 ($GBW \geq 10 \text{ GHz}$), that are designed for fast switching with millivolt input signal levels. Because oscillation problems tend to occur at frequencies around 5 MHz, where the LT1011 has a gain of approximately 2 V/mV, attenuation of output signals must be at least 2000:1 at 5 MHz as measured at the inputs. If the source impedance is 1 k Ω , the effective stray capacitance between output and input must have a reactance of more than $(2000)(1 \text{ k}\Omega) = 2 \text{ M}\Omega$, or less than 2 pF. The actual inter-lead capacitance between input and output pins on the LT1011 is less than 0.002 pF when cut to mounting length for printed circuit boards. Additional stray capacitance due to printed circuit traces must be minimized by routing the output trace directly away from input lines and, if possible, running ground traces next to input traces to provide shielding.

Additional steps to prevent oscillation problems are:

1. Bypass the strobe/balance pins with a 0.01- μF capacitor connected from pin 5 to pin 6 to eliminate stray capacitive feedback from the output to the balance pins. The balance pins are nearly as sensitive to stray capacitive feedback as the inputs.
2. Bypass the negative supply (pin 4) with a 0.1- μF ceramic capacitor close to the comparator. A 0.1- μF capacitor can also be used for the positive supply (pin 8) if the pull-up load is tied to a separate supply. When the pull-up load is tied directly to pin 8, use a 2- μF solid tantalum bypass capacitor.
3. Bypass any slow-moving or dc input with a capacitor ($\geq 0.01 \mu\text{F}$) close to the comparator to reduce high-frequency source impedance.
4. Keep resistive source impedance as low as possible. If a resistor is added in series with one input, bypass it with a capacitor to balance source impedances for dc accuracy. The low input bias current of the LT1011 usually eliminates any need for source resistance balancing. A 5-k Ω imbalance, for example, creates only 0.25-mV offset.
5. Use hysteresis, which consists of shifting the input offset voltage of the comparator when the output changes state. Hysteresis forces the comparator to move quickly through its linear region, eliminating oscillations by "overdriving" the comparator under all input conditions. Hysteresis may be either ac or dc. An ac hysteresis technique does not shift the apparent offset voltage of the comparator but requires a minimum input signal slew rate to be effective. A dc hysteresis technique works for all input slew rates but creates a shift in offset voltage dependent on the previous condition of the input signal.

The circuit shown in Figure 21 is an excellent compromise between ac and dc hysteresis. The 0.003- μF capacitor from pin 6 to pin 8 generates ac hysteresis by slightly shifting the voltage on the balance pins; both pins move about 4 mV depending on the state of the output. If pin 6 is bypassed, a level of ac hysteresis is created that is sufficient to switch the output at a speed near the comparator's maximum speed.

A small amount of dc hysteresis is also used to prevent problems due to low values of input slew rate. The sensitivity of the balance pins to current is about 0.5-mV input referred offset for each microampere of balance pin current. The 15-M Ω resistor tied from output to pin 5 generates 0.5-mV dc hysteresis.

The circuit is especially useful for general-purpose comparator applications because it does not force any signals directly back onto the input signal source. Instead, it takes advantage of the unique properties of the balance pins to provide extremely fast, clean output switching even with low-frequency input signals in the millivolt range. The combination of ac and dc hysteresis creates clean oscillation-free switching with very small input errors. The curve in Figure 22 plots input referred error versus switching frequency for the circuit shown in Figure 21. Note that at low frequencies, the error is simply the dc hysteresis, while at high frequencies, an additional error is created by the ac hysteresis. The high-frequency error can be reduced by reducing C_H , but lower values may not provide clean switching with very low slew-rate input signals.

TYPICAL APPLICATION DATA

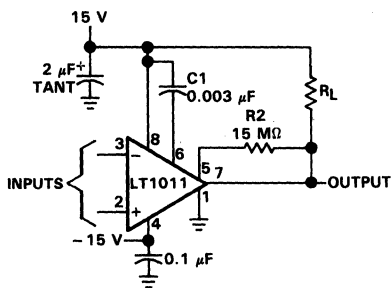


FIGURE 21. COMPARATOR WITH HYSTERESIS

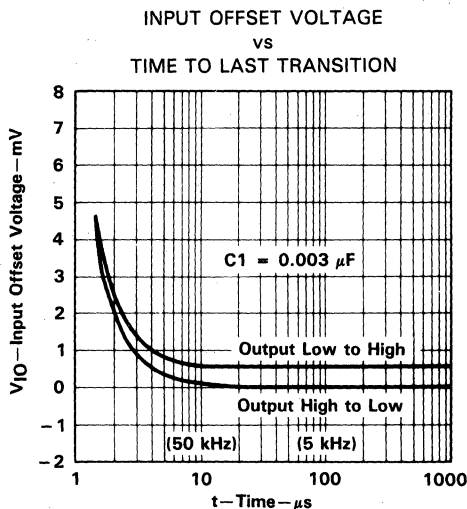


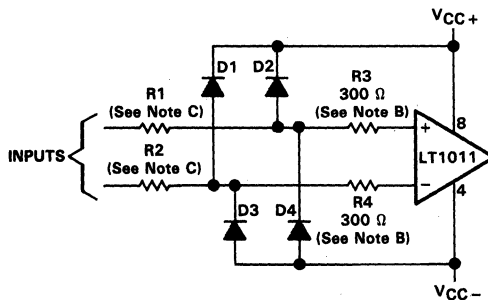
FIGURE 22

3

Voltage Comparators

input protection

The inputs to the LT1011 are particularly suited to general-purpose comparator applications because large differential and/or common-mode voltages can be tolerated without damage to the comparator. Either or both inputs can be raised 40 V above the negative supply, independent of the positive supply voltage. Internal forward biased diodes conduct when the inputs are taken below the negative supply. In this condition, input current must be limited to 1 mA. If very large (fault) input voltages must be accommodated, series resistors and clamp diodes should be used, as shown in Figure 23.



- NOTES: A. D1-D4 1N4148.
B. May be eliminated for fault current ≤ 1 mA.
C. Select according to allowable fault current and power dissipation.

FIGURE 23. LIMITING FAULT INPUT CURRENTS

TYPICAL APPLICATION DATA

The input resistors should limit fault current to a value between 0.1 mA and 20 mA. Power dissipation in the resistors must be considered for continuous faults, especially when the LT1011 supplies are off. Lightly loaded supplies may be forced to higher voltages by large fault currents flowing through D1-D4.

R3 and R4 limit input current to the LT1011 to less than 1 mA when the input signals are held below V_{CC-} . They may be eliminated if R1 and R2 are large enough to limit fault current to less than 1 mA.

input slew rate limitations

In the LT1011, step size is important because the slew rate of internal nodes increases response time for input step sizes larger than 1 V. For example, at 5-V step size, response time increases from 150 ns to 360 ns (see Figure 16). If response time is critical and large input signals are expected, clamp diodes across the inputs are recommended. The slew rate limitation can also affect performance when differential input voltage is low, but both inputs must slew quickly. The maximum suggested common-mode slew rate is 10 V/ μ s.

strobing

The LT1011 can be strobed by pulling current out of the strobe pin. The output transistor is forced to an off state, giving a high output at the collector (pin 7). Currents as low as -250μ A may cause strobing, but when the strobe current is low, strobe delay increases to between 200 ns and 300 ns. If strobe current is increased to -3 mA, strobe delay drops to about 60 ns. When the strobe current is 0, the voltage at the strobe pin is approximately 150 mV below V_{CC+} ; when the strobe current is increased to -3 mA, the strobe pin voltage is approximately 2 V below V_{CC+} . Do not ground the strobe pin; it must be current driven.

Figure 24 shows a typical strobe circuit. Note that there is no bypass capacitor between pins 5 and 6, which maximizes strobe speed but leaves the comparator more sensitive to oscillation problems for slow, low-level inputs. A 1-pF capacitor between the output and pin 5 greatly reduces oscillation problems without reducing strobe speed.

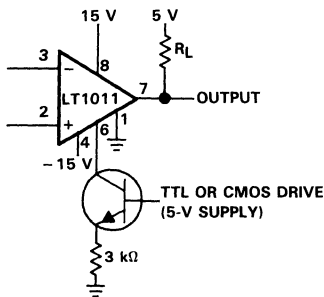


FIGURE 24. TYPICAL STROBE CIRCUIT

Placing a resistor from the output to pin 5 adds dc hysteresis. See step number 5 under "preventing oscillation problems."

The pin that is used for strobing (pin 6) is also one of the offset adjustment pins. Current into or out of pin 6 must be kept very low ($< 0.2 \mu$ A) when not strobing to prevent input offset voltage shifts.

output transistor

When the LT1011 output transistor is in the off state, negligible current flows into or out of the collector or emitter. The equivalent circuit is shown in Figure 25.

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Voltage Comparators

TYPICAL APPLICATION DATA

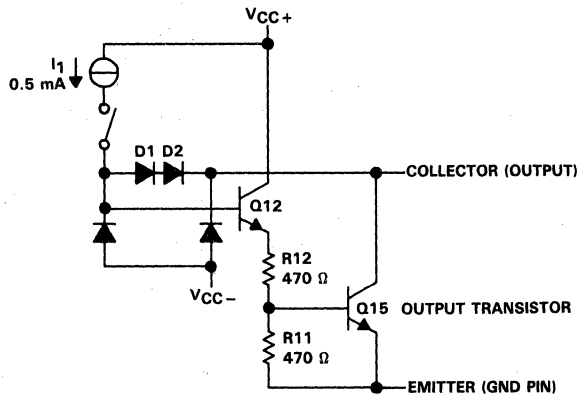


FIGURE 25. OUTPUT TRANSISTOR CIRCUITRY

3

Voltage Comparators

output transistor (continued)

In the off state, I_1 is switched off and both Q12 and Q15 turn off. The collector of Q15 can then be held above V_{CC-} without conducting current. The maximum voltage above V_{CC-} is 50 V for the LT1011 and 40 V for the LT1011C (these maximum voltages may exceed V_{CC+}). The emitter can be held at any voltage between V_{CC-} and V_{CC+} as long as the voltage is negative with respect to the collector.

In the on state, I_1 is connected, which turns on both Q12 and Q15. Diodes D1 and D2 prevent deep saturation of Q15 to improve speed and also limit the drive current of Q12. The R11/R12 divider sets the saturation voltage of Q15 and provides turn-off drive. Either the collector or emitter pin can be held at a voltage between V_{CC-} and V_{CC+} , which allows the remaining pin to drive the load. In typical applications, the emitter is connected to V_{CC-} or ground, and the collector drives a load tied to V_{CC+} or a separate positive supply.

When the emitter is used as the output, the collector is typically tied to V_{CC+} , and the load is connected to ground or V_{CC-} . Note that the emitter output is phase reversed with respect to the collector output so that the "+" and "-" input designations must be reversed. When the collector is tied to V_{CC+} , the voltage at the emitter in the one state is about 2 V below V_{CC+} .

input signal range

The input voltage range of the LT1011 is typically 300 mV above the negative supply and 1.5 V below the positive supply, independent of the actual supply voltages. This is the input voltage range over which the output will respond correctly when a voltage within the range is applied to one input and a higher or lower signal is applied to the other input. If one input is inside the range and one is outside, the output will be correct. If both inputs are outside the range, in opposite directions, the output will still be correct. If, however, both inputs are outside the range in the same direction, the output will not respond to the differential input; it will remain unconditionally off.

TYPICAL APPLICATIONS

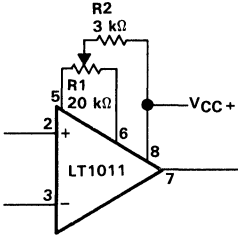
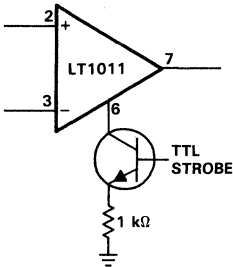
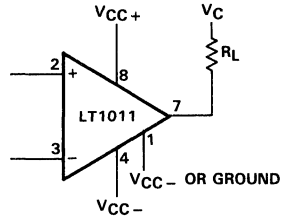


FIGURE 26. OFFSET BALANCING



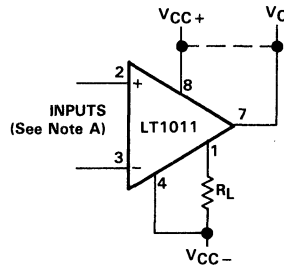
NOTE: Do not ground strobe pin.

FIGURE 27. STROBING



NOTE: V_C can be greater or less than V_{CC+} .

FIGURE 28. DRIVING LOAD REFERENCED TO POSITIVE SUPPLY



NOTE A: Input polarity is reversed when using Pin 1 for output.

FIGURE 29. DRIVING LOAD REFERENCED TO NEGATIVE SUPPLY

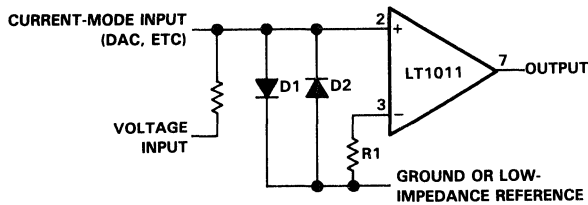
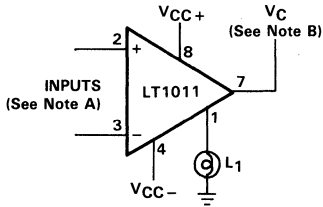


FIGURE 30. USING CLAMP DIODES TO IMPROVE FREQUENCY RESPONSE (See Figure 16)

LT1011, LT1011A VOLTAGE COMPARATORS

TYPICAL APPLICATIONS



- NOTES: A. Input polarity is reversed when using Pin 1 for output.
 B. V_C may be any voltage above V_{CC-} . Pin 1 swings to within approximately 2 V of V_{CC+} .

FIGURE 31. DRIVING LOAD REFERENCED TO GROUND

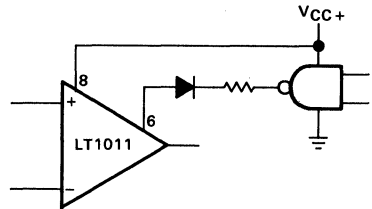
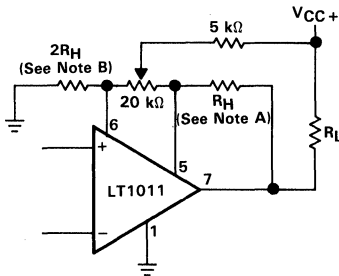


FIGURE 33. DIRECT STROBE DRIVE WHEN CMOS LOGIC USES SAME V_{CC+} SUPPLY AS LT1011 (Not applicable for TTL logic)

3

Voltage Comparators



- NOTES: A. Hysteresis is approximately 0.45 mV/ μ A change in current in R_H .
 B. This resistor causes hysteresis to be centered around V_{IO} .

FIGURE 32. COMBINING OFFSET ADJUSTMENT AND HYSTERESIS

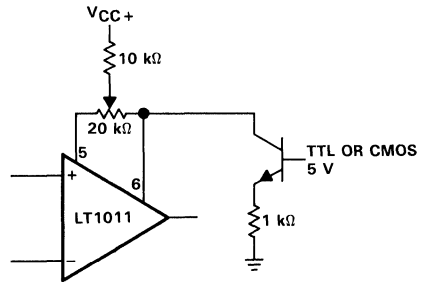
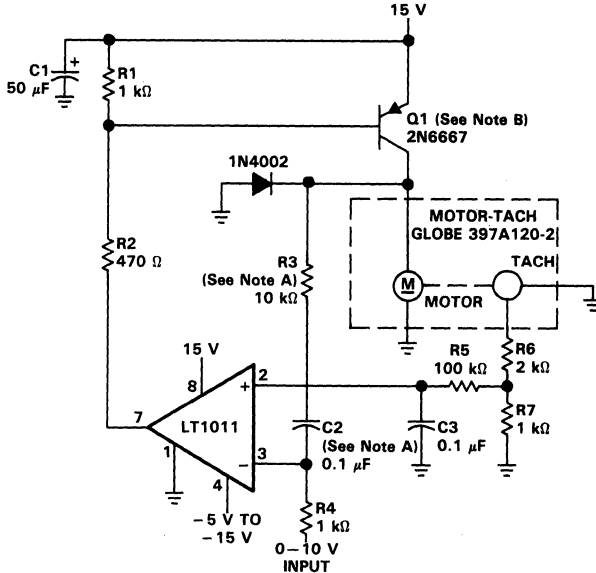


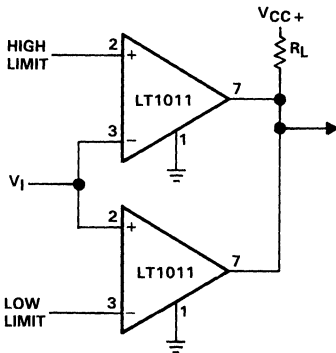
FIGURE 34. COMBINING OFFSET ADJUSTMENT AND STROBE

TYPICAL APPLICATIONS



NOTES: A. R3/C2 determines oscillation frequency of controller.
B. Q1 operates in switch mode.

FIGURE 35. HIGH-EFFICIENCY MOTOR SPEED CONTROLLER



NOTE: Output is high inside "window" and low above high limit or below low limit.

FIGURE 36. WINDOW DETECTOR

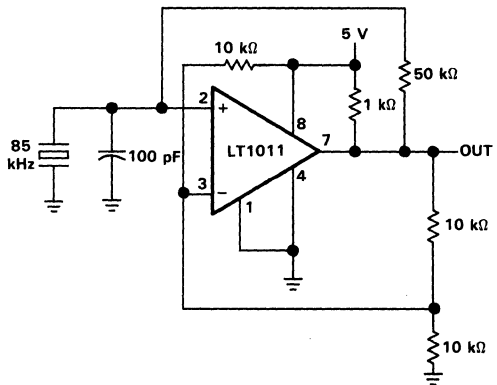
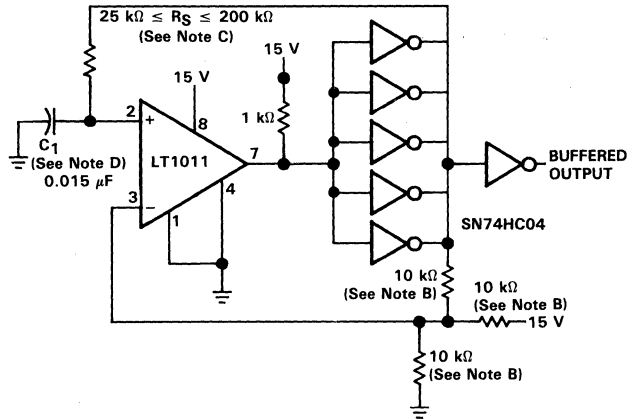


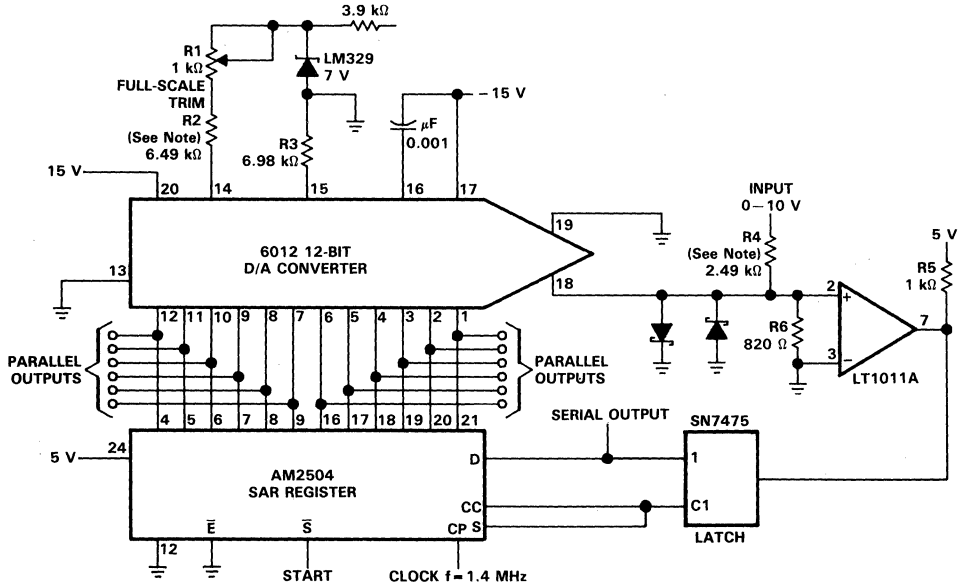
FIGURE 37. CRYSTAL OSCILLATOR

TYPICAL APPLICATIONS



- NOTES: A. Low drift and accurate frequency are obtained because this configuration rejects effects due to input offset voltage and input bias current of the comparator.
 B. 1% metal film.
 C. $R_S = \text{TRW type MTR-5/} + 120 \text{ ppm/}^\circ\text{C}$.
 D. $C_1 = 0.015 \mu\text{F} = \text{polystyrene : } 120 \text{ ppm/}^\circ\text{C} \pm 30 \text{ ppm WESCO type 32-P}$.
 E. Comparator contributes $\leq 10 \text{ ppm/}^\circ\text{C}$ drift for frequencies below 10 kHz.

FIGURE 38. LOW-DRIFT R/C OSCILLATOR



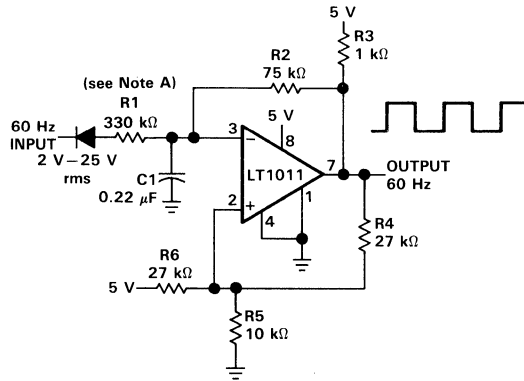
NOTE: R2 and R4 should TC track.

FIGURE 39. 10- μs 12-BIT A-D CONVERTER

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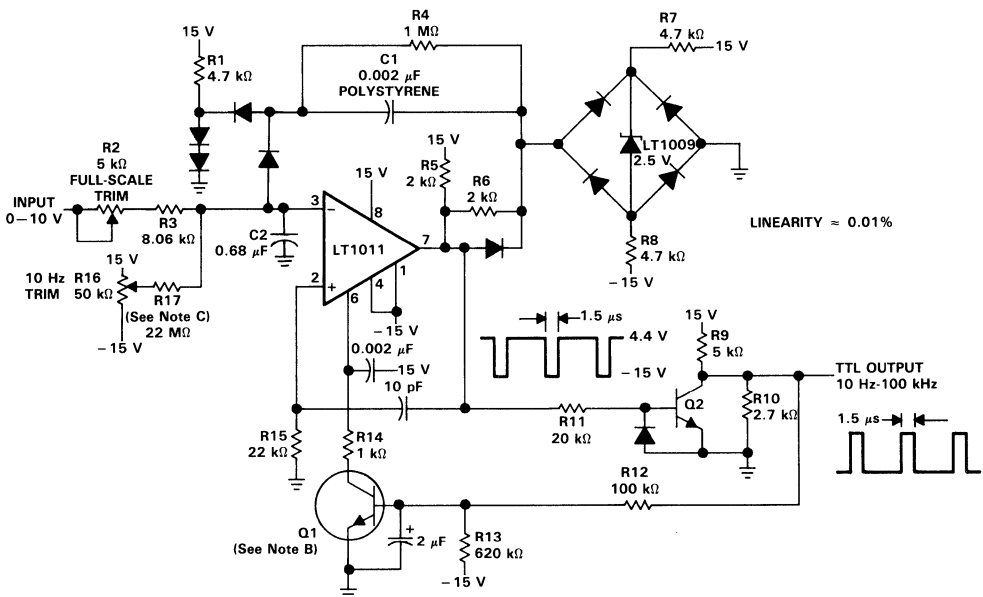
Voltage Comparators

TYPICAL APPLICATIONS



- NOTES: A. Increase R1 for larger input voltages.
B. LT1011 self-oscillates at approximately 60 Hz, thereby "locking" onto incoming line signal.

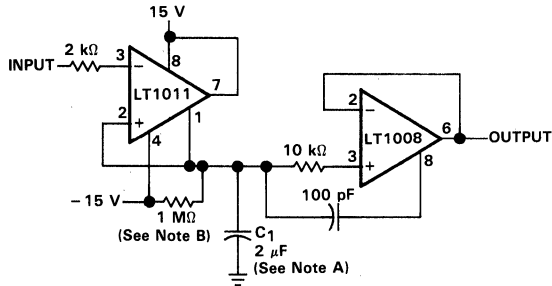
FIGURE 40. NOISE-IMMUNE 60-Hz LINE SYNCHRONIZATION



- NOTES: A. All diodes 1N4148, transistors 2N3904.
B. Used only to guarantee start-up.
C. R17 may be increased for better 10-Hz trim resolution.

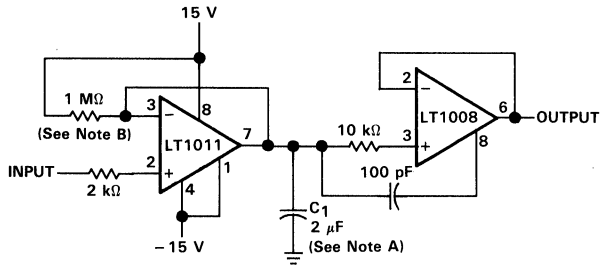
FIGURE 41. 10-Hz TO 100-kHz VOLTAGE-TO-FREQUENCY CONVERTER

TYPICAL APPLICATIONS



NOTES: A. Mylar
B. Set for required reset time constant.

FIGURE 42. POSITIVE PEAK DETECTOR



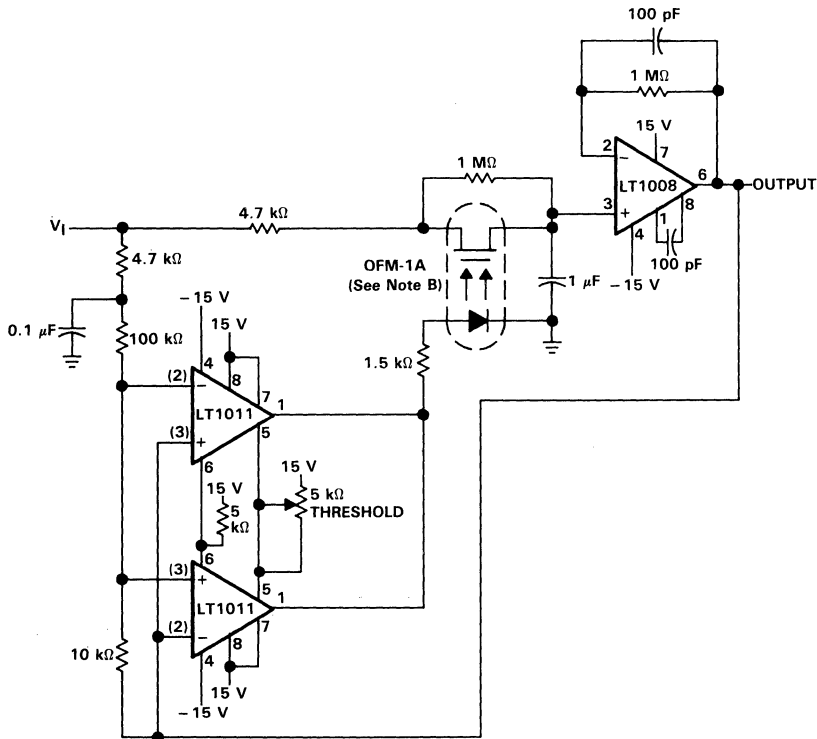
NOTES: A. Mylar
B. Select for required reset time constant.

FIGURE 43. NEGATIVE PEAK DETECTOR

3

Voltage Comparators

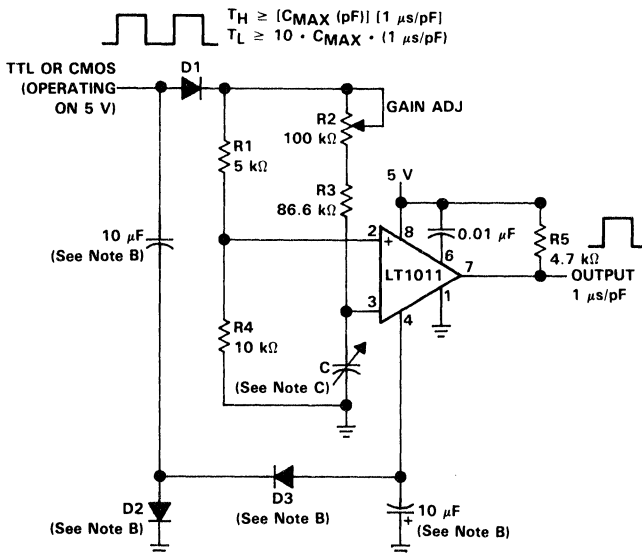
TYPICAL APPLICATIONS



- NOTES: A. The comparators drive the opto-coupled FET "on" when the difference between the output and the input exceeds threshold. When the output approaches the input, the FET turns "off" and low-pass filtering occurs.
B. From Theta-J Corporation, Woburn, Massachusetts.

FIGURE 44. FAST-SETTLING FILTER

TYPICAL APPLICATIONS



- NOTES: A. $PW = (R_2 + R_3) \cdot (C) \cdot [(R_1 + R_4)/R_1]$. The input capacitance of the LT1011 is approximately 6 pF. This is an offset term.
 B. These components may be eliminated if negative supply is available (-1 V to -15 V).
 C. Typical two sections of 365-pF variable capacitor when used as shaft-angle indication.

FIGURE 47. CAPACITANCE-TO-PULSE-WIDTH CONVERTER



3

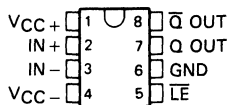
Voltage Comparators

LT1016 ULTRA-FAST PRECISION LATCHED COMPARATOR

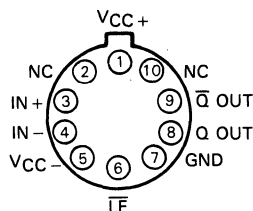
D3242, MAY 1988—REVISED MARCH 1989

- Ultra-Fast . . . 10 ns Typ t_{pd}
- Operates from Single 5-V or Dual ± 5 -V Supply
- Complementary TTL Outputs
- Low Input Offset Voltage . . . 0.8 mV or 1 mV Typ
- No Minimum Input Slew Rate Requirement
- No Supply Current Spiking
- Output Latch

D, JG, OR P PACKAGE
(TOP VIEW)



L PACKAGE
(TOP VIEW)



NC—No internal connection.

All leads of the L package are electrically insulated from the case.

description

The LT1016 is an ultra-fast comparator specifically designed to interface directly to TTL logic while operating from either a dual ± 5 -V supply or a single 5-V supply. The LT1016 offers tight offset voltage specifications and high gain for precision applications. Matched complementary outputs further extend the versatility of the LT1016.

The LT1016 features a unique output stage that provides active drive in both directions for maximum speed into TTL-logic or passive loads yet does not exhibit the large current spikes normally found in totem-pole output stages. This eliminates the need for a minimum input slew rate typical of other fast comparators. The LT1016's ability to remain stable with the outputs in the active region greatly reduces the problem of output "glitching" when the input signal is slow moving or is at a low level.

The LT1016 has a true latch for retaining input data at the outputs. The outputs remain latched as long as the latch enable input \overline{LE} is high. Quiescent negative supply current is only 3 mA, about ten times lower than competitive units. This feature reduces die temperature and allows the negative supply pin to be driven from virtually any supply voltage with a simple resistive divider. Device performance is not affected by variations in negative supply voltage.

The LT1016M is characterized for operation over the full military temperature range of -55°C to 125°C . The LT1016C is characterized for operation from 0°C to 70°C .

AVAILABLE OPTIONS

T _A	PACKAGE			
	SMALL OUTLINE (D)	CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)
0°C to 70°C	LT1016CD	LT1016CJG	LT1016CL	LT1016CP
-55°C to 125°C		LT1016MJG	LT1016ML	

The D package is available taped and reeled. Add the suffix R to the device type (e.g., LT1016CDR).

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Voltage Comparators

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS
INSTRUMENTS**

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LT1016

ULTRA-FAST PRECISION LATCHED COMPARATOR

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	7 V
Supply voltage, V_{CC-}	-7 V
Differential input voltage (see Note 2)	± 5 V
Input voltage (either input)	$V_{CC\pm}$
Latch enable input voltage	$V_{CC\pm}$
Output current, I_O	± 20 mA
Operating free-air temperature range: LT1016M	-55°C to 125°C
LT1016C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: L package	300°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. The output may be shorted to ground or to either power supply.

recommended operating conditions

		LT1016M			LT1016C			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, V_{CC+}				5		5	V	
Supply voltage, V_{CC-}				-5		-5	V	
Input voltage, V_I	$V_{CC\pm} = \pm 15$ V	-3.75	3.5	-3.75	3.5		V	
	$V_{CC+} = 5$ V, $V_{CC-} = 0$	1.25	3.5	1.25	1.25			
Operating free-air temperature, T_A		-55	125	0	70		°C	

3

Voltage Comparators

LT1016

ULTRA-FAST PRECISION LATCHED COMPARATOR

electrical characteristics, $V_{CC+} = 5\text{ V}$, $V_{CC-} = -5\text{ V}$, $V_{O(Q)} = 1.4\text{ V}$, \overline{LE} at 0 V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A †	LT1016M			LT1016C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage (see Note 4)	$R_S \leq 100\ \Omega$	25°C	0.8		2	1		3	mV
		Full range			3			3.5	
αV_{IO} Average temperature coefficient of input offset voltage		Full range	4			4			$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current (see Note 4)		25°C	0.3		1	0.3		1	μA
		Full range			1.3			1.3	
I_{IB} Input bias current (see Note 5)	$V_O = 1.4\text{ V}$	25°C	5		10	5		10	μA
		Full range			13			13	
V_{ICR} Common-mode input voltage range	Dual supply	Full range	-3.75 to 3.5			-3.75 to 3.5			V
	Single supply	Full range	1.25 to 3.5			1.25 to 3.5			
V_I Input voltage	\overline{LE} high	Full range	2			2			V
	\overline{LE} low	Full range			0.8			0.8	
V_{OH} High-level output voltage	$V_{CC+} \leq 4.6\text{ V}$, $I_O = 1\text{ mA}$	Full range	2.7			2.7			V
	$V_{CC+} \leq 4.6\text{ V}$, $I_O = 10\text{ mA}$		2.4			2.4			
V_{OL} Low-level output voltage	$I_O = 4\text{ mA}$	Full range			0.5			0.5	V
	$I_O = 10\text{ mA}$	25°C	0.4			0.4			
A_{VD} Small-signal differential voltage amplification	$V_O = 1\text{ V to } 2\text{ V}$	25°C	1400	3000		1400	3000	V/V	
$CMRR$ Common-mode rejection ratio	$V_{IC} = -3.75\text{ V to } 3.5\text{ V}$	Full range	80			80			dB
k_{SVR} Supply voltage rejection ratio	Positive supply, $V_{CC+} = 4.6\text{ V to } 5.4\text{ V}$	Full range	60			60			dB
	Negative supply, $V_{CC-} = 2\text{ V to } 7\text{ V}$		80			80			
I_{CC+} Supply current from V_{CC+}		Full range			35			35	mA
I_{CC-} Supply current from V_{CC-}		Full range			5			5	mA
I_l Latch pin input current		Full range			500			500	μA

†Full range is -55°C to 125°C for the LT1016M. Full range is 0°C to 70°C for the LT1016C.

NOTES: 4. Input offset voltage is defined as the average of the two voltages measured by forcing first one output and then the other to 1.4 V. Input offset current is defined in an analogous way.

5. Input bias current is defined as the average of the two input currents.

3

Voltage Comparators

LT1016
ULTRA-FAST PRECISION LATCHED COMPARATOR

switching characteristics, $V_{CC+} = 5\text{ V}$, $V_{CC-} = -5\text{ V}$, \overline{LE} at 0 V (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T_A †	MIN	TYP	MAX	UNIT
t_{pd}	Propagation delay time	$\Delta V_I = 100\text{ mV}$, 5-mV overdrive, See Note 6	25°C		10	14	ns
			Full range			16	
		$\Delta V_I = 100\text{ mV}$, 20-mV overdrive, See Note 6	25°C		10	14	
			Full range			16	
Δt_{pd}	Differential propagation delay	$\Delta V_I = 100\text{ mV}$, 5-mV overdrive, See Note 6	25°C			3	ns
	Latch minimum setup time		25°C		2		ns

† Full range is -55°C to 125°C for the LT1016M. Full range is 0°C to 70°C for the LT1016C.

NOTE 6: t_{pd} and Δt_{pd} cannot be measured in automatic-handling test equipment with low values of overdrive. The LT1016 is tested with a 1-V step and 500-mV overdrive. Correlation testing indicates that t_{pd} and Δt_{pd} limits shown can be met with this test. For low overdrive conditions, V_{IO} is added to the overdrive.

3

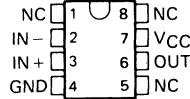
Voltage Comparators

TL331I, TL331C DIFFERENTIAL COMPARATORS

D2344, APRIL 1977—REVISED OCTOBER 1988

- Single Supply or Dual Supplies
- Wide Range of Supply Voltage
2 to 36 V
- Low Supply Current Drain Independent of Supply Voltage . . . 0.8 mA Typ
- Low Input Bias Current . . . 25 nA Typ
- Low Input Offset Current
3 to 5 nA Typ
- Low Input Offset Voltage . . . 2 mV Typ
- Common-Mode Input Voltage Range
Includes Ground
- Differential Input Voltage Range Equal to Maximum-Rated Supply Voltage . . . ± 36 V
- Low Output Saturation Voltage
- Output Compatible with TTL, MOS, and CMOS

D, JG, OR P PACKAGE
(TOP VIEW)



NC—No internal connection

AVAILABLE OPTIONS

T _A	V _{IO} MAX at 25°C	PACKAGE		
		SMALL OUTLINE (D)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	5 mV	TL331CD	TL331CJG	TL331CP
-25°C to 85°C	5 mV	TL331ID	TL331IJG	TL331IP

The D package is available taped and reeled. Add the suffix R to the device type (e.g., TL331CDR)

3

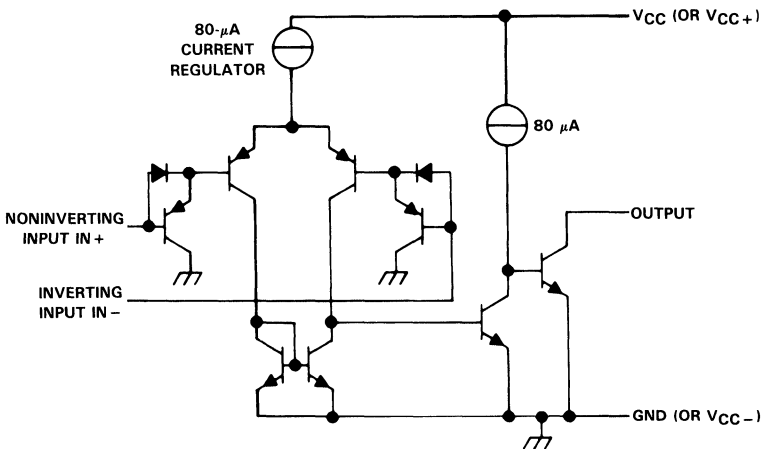
Voltage Comparators

description

The TL331 is a voltage comparator that is designed to operate from a single power supply over a wide range of voltages. Operation from dual supplies is also possible so long as the difference between the two supplies is 2 V to 36 V and pin 7 is at least 1.5 V more positive than the input common-mode voltage. Current drain is independent of the supply voltage.

The TL331I is characterized for operation from -25°C to 85°C. The TL331C is characterized for operation from 0°C to 70°C.

schematic



Current values shown are nominal.

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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TL331I, TL331C DIFFERENTIAL COMPARATORS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	36 V
Differential input voltage (see Note 2)	± 36 V
Input voltage range (either input)	-0.3 V to 36 V
Output voltage	36 V
Output current	20 mA
Duration of output short-circuit to ground (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: TL331I	-25°C to 85°C
TL331C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the network ground terminal.
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. Short circuits from the output to V_{CC} can cause excessive heating and eventual destruction.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING
D	680 mW	5.8 mW/°C	464 mW	377 mW
JG	680 mW	6.6 mW/°C	528 mW	429 mW
P	680 mW	8.0 mW/°C	640 mW	520 mW

recommended operating conditions

		TL331I		TL331C		UNIT
		MIN	MAX	MIN	MAX	
Supply voltage, V_{CC}		5	30	5	30	V
Common-mode input voltage, V_{IC}	$V_{CC} = 5$ V	0	3	0	3	V
	$V_{CC} = 30$ V	0	28	0	28	
Operating free-air temperature, T_A		-25	85	0	70	°C

3 Voltage Comparators

TL331I, TL331C DIFFERENTIAL COMPARATORS

electrical characteristics at specified free-air temperature, $V_{CC} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TL331I			TL331C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{CC} = 5\text{ V to }30\text{ V}$, $V_{IC} = V_{ICR}\text{ min}$, $V_O = 1.4\text{ V}$	25°C	2	5	2	5	mV	
		Full range	9			9		
I_{IO} Input offset current	$V_O = 1.4\text{ V}$	25°C	3	25	5	50	nA	
		Full range	100			150		
I_{IB} Input bias current		25°C	-25	-100	-25	-250	nA	
		Full range	-300			-400		
V_{ICR} Common-mode input voltage range	$V_{CC} = 5\text{ V to }30\text{ V}$	25°C	0 to $V_{CC}-1.5$	0 to $V_{CC}-1.5$	0 to $V_{CC}-1.5$	V		
		Full range	0 to $V_{CC}-2$	0 to $V_{CC}-2$	0 to $V_{CC}-2$			
A_{VD} Large-signal differential voltage amplification	$V_{CC} = 15\text{ V}$, $V_O = 1.4\text{ V to }11.4\text{ V}$, $R_L = 15\text{ k}\Omega\text{ to }V_{CC}$	25°C	200	200	200	V/mV		
I_{OH} High-level output current	$V_{ID} = 1\text{ V}$	$V_{OH} = 5\text{ V}$	25°C	0.1	0.1	nA		
		$V_{OH} = 30\text{ V}$	Full range	1			μA	
V_{OL} Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 4\text{ mA}$	25°C	150	400	150	400	mV	
		Full range	700			700		
I_{OL} Low-level output current	$V_{ID} = -1\text{ V}$, $V_{OL} = 1.5\text{ V}$	25°C	6	6	6	mA		
I_{CC} Supply current	$V_O = 2.5\text{ V}$, No load	25°C	0.5	0.8	0.5	0.8	mA	

† Full range (MIN to MAX) for the TL331I is -25°C to 85°C and for the TL331C is 0°C to 70°C . All characteristics are measured with zero common-mode input voltage unless otherwise specified.

switching characteristics, $V_{CC} = 5\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Response time	R_L connected to 5 V through 5.1 k Ω ,	100-mV input step with 5-mV overdrive			μs
	$C_L = 15\text{ pF}$, [‡] See Note 4	TTL-level input step			

[‡] C_L includes probe and jig capacitance.

NOTE 4: The response time specified is the interval between the input step function and the instant when the output crosses 1.4 V.

3
Voltage Comparators



Voltage Comparators

TL514M DUAL DIFFERENTIAL COMPARATOR WITH STROBE

D999, OCTOBER 1977 – REVISED MARCH 1988

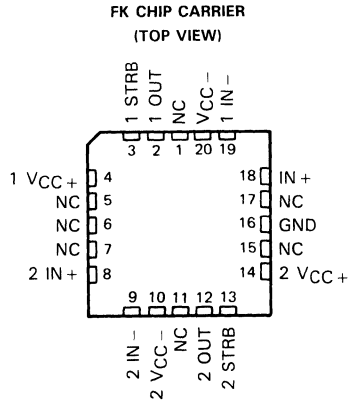
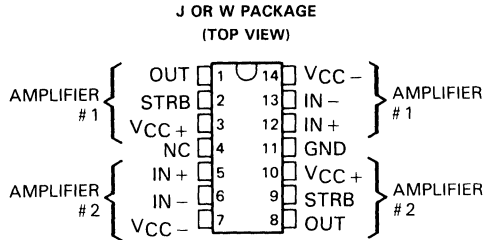
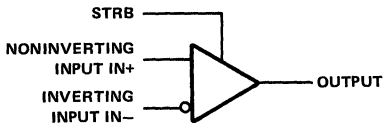
- Fast Response Times
- High Differential Voltage Amplification
- Low Offset Characteristics
- Outputs Compatible with Most TTL Circuits

description

The TL514 is an improved version of the TL720 dual high-speed voltage comparator. When compared with the TL720, these circuits feature higher amplification (typically 33,000) due to an extra amplification stage, increased accuracy because of lower offset characteristics, and greater flexibility with the addition of a strobe to each comparator. Since the output cannot be more positive than the strobe, a low-level input at the strobe will cause the output to go low regardless of the differential input.

These circuits are especially useful in applications requiring an amplitude discriminator, memory sense amplifier, or a high-speed limit detector. The TL514M is characterized for operation over the full military temperature range of -55°C to 125°C .

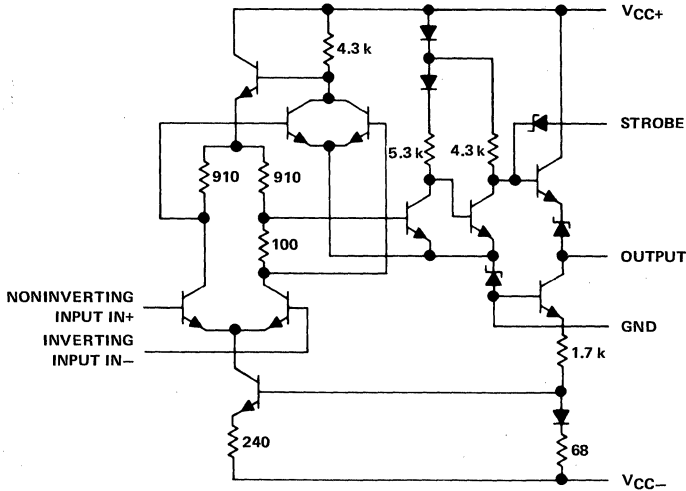
symbol (each comparator)



NC -- No internal connection

TL514M DUAL DIFFERENTIAL COMPARATOR WITH STROBE

schematic (each comparator)



Resistor values shown are nominal in ohms.

3

Voltage Comparators

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage V_{CC+} (see Note 1)	14 V
Supply voltage V_{CC-} (see Note 1)	-7 V
Differential input voltage (see Note 2)	± 5 V
Input voltage (any input, see Note 1)	± 7 V
Strobe voltage (see Note 1)	6 V
Peak output current ($t_W \leq 1$ s)	10 mA
Continuous total dissipation (either comparator or both together)	See Dissipation Rating Table
Operating free-air temperature range	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J or W package	300°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the network ground terminal.
2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 125^\circ\text{C}$
	POWER RATING			POWER RATING
FK	600 mW	11.0 mW/°C	95°C	275 mW
J	600 mW	11.0 mW/°C	95°C	275 mW
W	600 mW	8.0 mW/°C	75°C	200 mW

TL514M

DUAL DIFFERENTIAL COMPARATOR WITH STROBE

electrical characteristics at specified free-air temperature, $V_{CC+} = 12\text{ V}$, $V_{CC-} = -6\text{ V}$

PARAMETER		TEST CONDITIONS†		MIN	TYP	MAX	UNIT	
V_{IO}	Input offset voltage	$R_S \leq 20\ \Omega$, See Note 4	25 °C		0.6	2	mV	
			-55 °C to 125 °C			3		
α_{VIO}	Average temperature coefficient of input offset voltage	$R_S = 50\ \Omega$, See Note 4	-55 °C to 25 °C		3	10	$\mu\text{V}/^\circ\text{C}$	
			25 °C to 125 °C		3	10		
I_{IO}	Input offset current	See Note 4	25 °C		0.75	3	μA	
			-55 °C		1.8	7		
			125 °C		0.25	3		
α_{IIO}	Average temperature coefficient of input offset current	See Note 4	-55 °C to 25 °C		15	75	nA/°C	
			25 °C to 125 °C		5	25		
I_B	Input bias current	See Note 4	25 °C		7	15	μA	
			-55 °C		12	25		
$I_{H(S)}$	High-level strobe current	$V_{(\text{strobe})} = 5\text{ V}$, $V_{ID} = -5\text{ mV}$	25 °C		± 100		μA	
$I_{L(S)}$	Low-level strobe current	$V_{(\text{strobe})} = -100\text{ mV}$, $V_{ID} = 5\text{ mV}$	25 °C		-1	-2.5	mA	
V_{ICR}	Common-mode input voltage range	$V_{CC} = -7\text{ V}$	-55 °C to 125 °C		± 5		V	
V_{ID}	Differential input voltage range		-55 °C to 125 °C		± 5		V	
A_{VD}	Large-signal differential voltage amplification	No load, $V_O = 0$ to 2.5 V	25 °C		12.5	33	V/mV	
			-55 °C to 125 °C		10			
V_{OH}	High-level output voltage	$V_{ID} = 5\text{ mV}$, $I_{OH} = 0$	-55 °C to 125 °C		4 [§]	5	V	
			$V_{ID} = 5\text{ mV}$, $I_{OH} = -5\text{ mA}$	-55 °C to 125 °C		2.5		3.6 [§]
V_{OL}	Low-level output voltage	$V_{ID} = -5\text{ mV}$, $I_{OL} = 0$	-55 °C to 125 °C		-1	-0.5 [§]	0 [‡]	V
			$V_{(\text{strobe})} = 0.3\text{ V}$, $V_{ID} = 5\text{ mV}$, $I_{OL} = 0$	-55 °C to 125 °C		-1		0 [‡]
I_{OL}	Low-level output current	$V_{ID} = -5\text{ mV}$, $V_O = 0$	25 °C		2	2.4	mA	
			-55 °C		1	2.3		
			125 °C		0.5	2.3		
r_o	Output resistance	$V_O = 1.4\text{ V}$	25 °C		200		Ω	
$CMRR$	Common-mode rejection ratio	$R_S \leq 200\ \Omega$	-55 °C to 125 °C		80	100 [§]	dB	
I_{CC+}	Supply current from V_{CC+} †	$V_{ID} = -5\text{ mV}$, No load	-55 °C to 125 °C		11 [§]	18	mA	
I_{CC-}	Supply current from V_{CC-} †		-55 °C to 125 °C		-7 [§]	-14	mA	
P_D	Total power dissipation†		-55 °C to 125 °C		180 [§]	300	mW	

† Unless otherwise noted, all characteristics are measured with the strobe open.

‡ The algebraic convention, where the most-positive (least-negative) limit is designated as maximum, is used in this data sheet for logic levels only, e.g., when 0 V is the maximum, the minimum limit is a more-negative voltage.

§ These typical values are at $T_A = 25^\circ\text{C}$.

¶ Supply current and power dissipation limits apply for both comparators operating simultaneously.

NOTE 4: These characteristics are verified by measurements at the following temperatures and output voltage levels: $V_O = 1.8\text{ V}$ at $T_A = -55^\circ\text{C}$, $V_O = 1.4\text{ V}$ at $T_A = 25^\circ\text{C}$, and $V_O = 1\text{ V}$ at $T_A = 125^\circ\text{C}$. These output voltage levels were selected to approximate the logic threshold voltages of the types of digital logic circuits this comparator is intended to drive.

3

Voltage Comparators

TL514M

DUAL DIFFERENTIAL COMPARATOR WITH STROBE

switching characteristics, $V_{CC+} = 12\text{ V}$, $V_{CC-} = -6\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Response time	$R_L = \infty$ $C_L = 5\text{ pF}$, See Note 5		30	80	ns
Strobe release time	$R_L = \infty$ $C_L = 5\text{ pF}$, See Note 6		5	25	ns

NOTES: 5. The response time specified is for a 100-mV input step with 5-mV overdrive.

6. For testing purposes, the input bias conditions are selected to produce an output voltage of 1.4 V. A 5-mV overdrive is then added to the input bias voltage to produce an output voltage which rises above 1.4 V. The time interval is measured from the 50% point of the strobe voltage curve to the point where the overdriven output voltage crosses the 1.4-V level.

TYPICAL CHARACTERISTICS

LARGE-SIGNAL DIFFERENTIAL
VOLTAGE AMPLIFICATION
vs
FREE-AIR TEMPERATURE

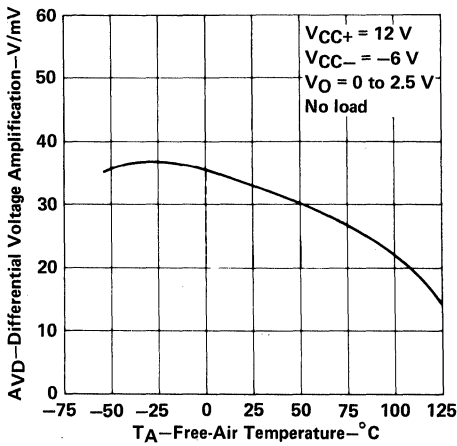


FIGURE 1

LARGE-SIGNAL DIFFERENTIAL
VOLTAGE AMPLIFICATION
vs
SUPPLY VOLTAGE

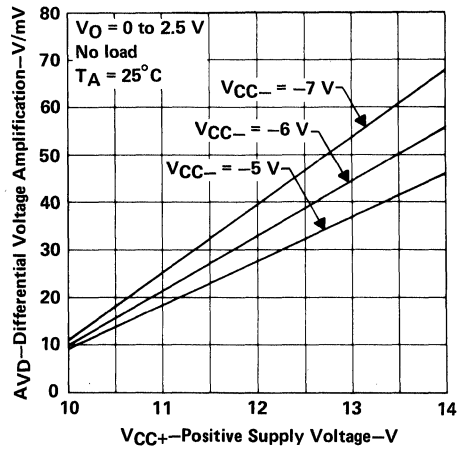


FIGURE 2

3

Voltage Comparators

TYPICAL CHARACTERISTICS

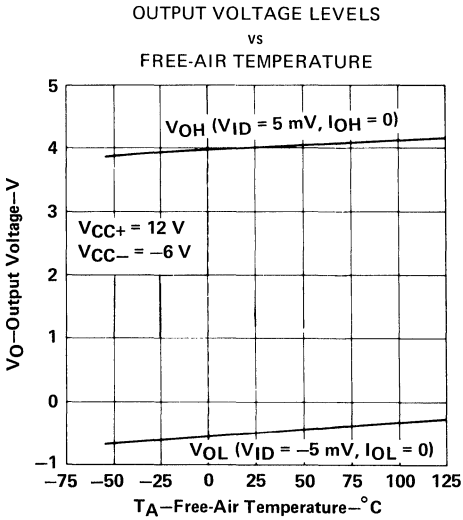


FIGURE 3

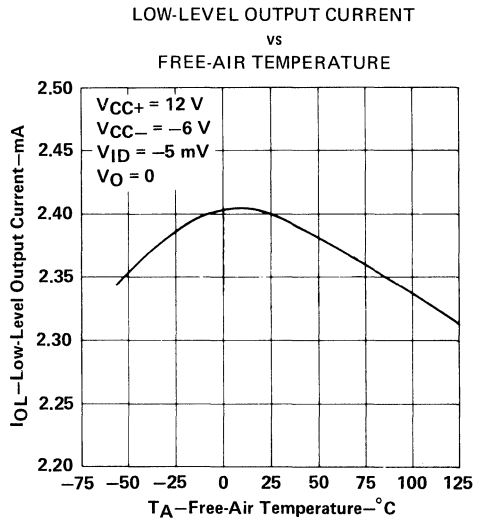


FIGURE 4

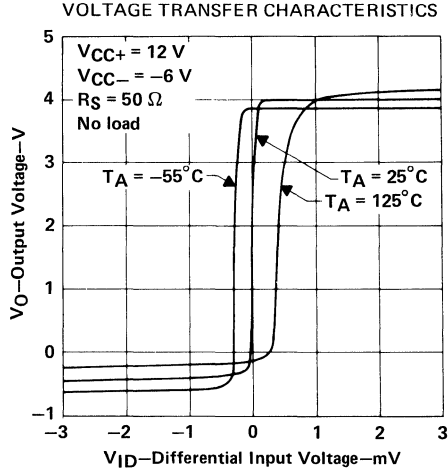
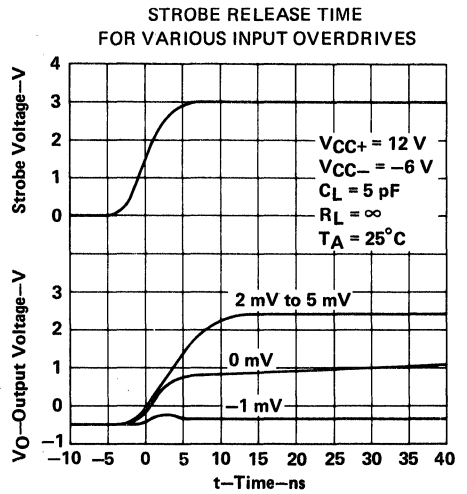
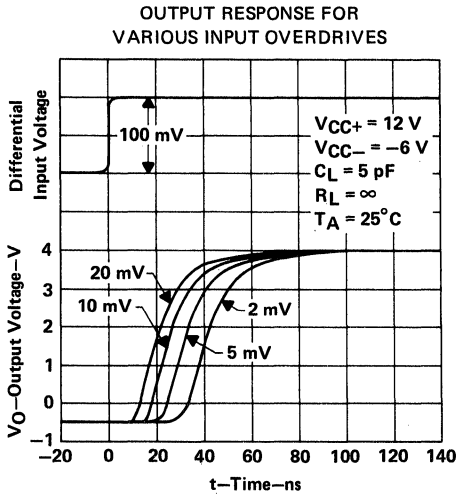
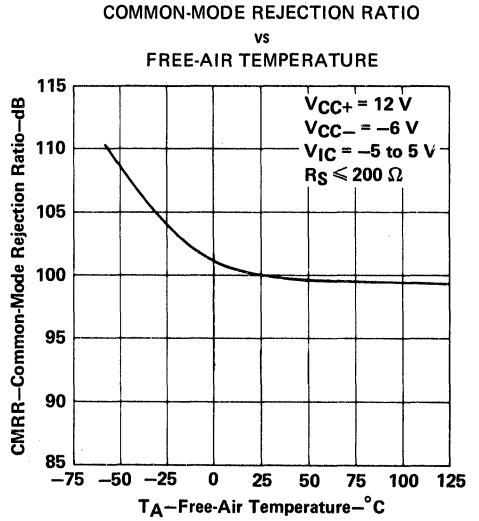
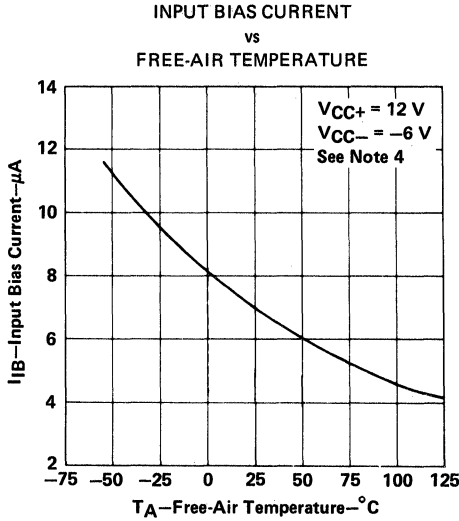


FIGURE 5

3
 Voltage Comparators

TL514M DUAL DIFFERENTIAL COMPARATOR WITH STROBE

TYPICAL CHARACTERISTICS



NOTE 4: These characteristics are verified by measurements at the following temperatures and output voltage levels: $V_O = 1.8 \text{ V}$ at $T_A = -55^\circ\text{C}$, $V_O = 1.4 \text{ V}$ at $T_A = 25^\circ\text{C}$, and $V_O = 1 \text{ V}$ at $T_A = 125^\circ\text{C}$. These output voltage levels were selected to approximate the logic threshold voltages of the types of digital logic circuits this comparator is intended to drive.

3

Voltage Comparators

TYPICAL CHARACTERISTICS

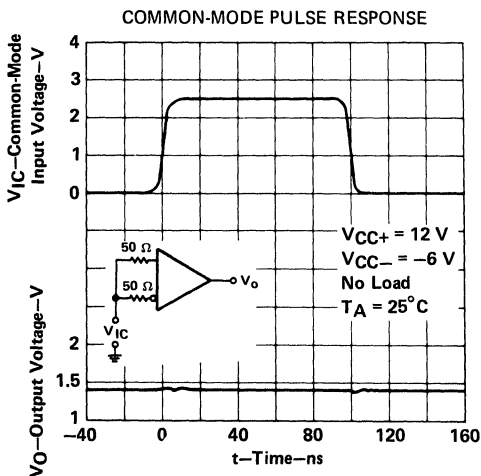


FIGURE 10

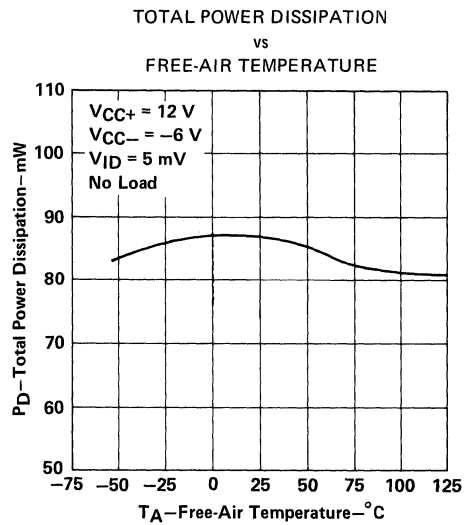


FIGURE 11

3

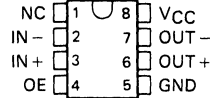
Voltage Comparators

TL712 DIFFERENTIAL COMPARATOR

D2741, JUNE 1983—REVISED MARCH 1988

- Operates from a 5-V Supply
- 0 to 5 V Common-Mode Input Voltage Range
- Self-Biased Inputs
- Complementary 3-State Outputs
- Enable Capability
- Hysteresis . . . 5 mV Typ
- Response Times . . . 25 ns Typ

D, JG, OR P PACKAGE
(TOP VIEW)



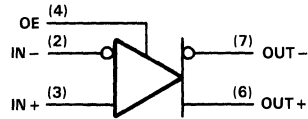
NC—No internal connection

description

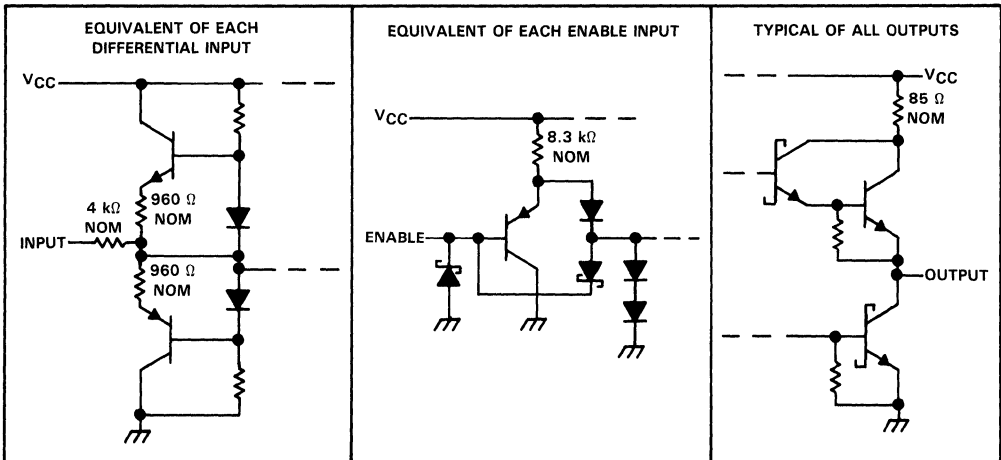
The TL712 is a high-speed comparator fabricated with bipolar Schottky† process technology. The circuit has differential analog inputs and complementary 3-state TTL-compatible logic outputs with symmetrical switching characteristics. When the output enable, OE, is low, both outputs are in the high-impedance state. This device operates from a single 5-V supply and is useful as a disk memory read-chain data comparator.

The TL712 is characterized for operation from 0°C to 70°C.

symbol (positive logic)



schematics of inputs and outputs



Voltage Comparators

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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TL712

DIFFERENTIAL COMPARATOR

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	7 V
Input voltage, any differential input	± 25 V
Differential input voltage (see Note 2)	± 25 V
Enable input voltage	7 V
Low-level output current	50 mA
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the network ground terminal.
2. Differential voltage values are at the noninverting terminal with respect to the inverting terminal.

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{CC}	4.75	5	5.25	V
Common-mode input voltage, V_{IC}			± 15	V
High-level output current, I_{OH}			-1	mA
Low-level output current, I_{OL}			16	mA
Operating free-air temperature, T_A	0		70	°C

electrical characteristics at $V_{CC} = 5$ V, $T_A = 25$ °C

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_T Threshold voltage (V_{T+} and V_{T-})	$V_{ICR} = 0$ to 5 V	-100 [†]		100	mV
V_{hys} Hysteresis ($V_{T+} - V_{T-}$)			5		mV
V_{OH} High-level output voltage	$V_{ID} = 100$ mV, $I_{OH} = -1$ mA	2.7	3.5		V
V_{OL} Low-level output voltage	$V_{ID} = -100$ mV, $I_{OL} = 16$ mA		0.4	0.5	V
I_{OZ} Off-state output current	$V_O = 2.4$ V			-20	μ A
I_I Enable current	$V_I = 5.5$ V			100	μ A
I_{IH} High-level enable current	$V_{IH} = 2.7$ V			20	μ A
I_{IL} Low-level enable current	$V_{IL} = 0.4$ V			-360	μ A
r_i Differential input resistance		4			k Ω
r_o Output resistance				100	Ω
I_{OS} Short-circuit output current		-15		-85	mA
I_{CC} Supply current	$V_{ID} = 0$, No load		17	20	mA

[†] The algebraic convention, where the more negative limit is designated as minimum, is used in this data sheet for input threshold voltage levels only.

switching characteristics, $V_{CC} = 5$ V, $T_A = 25$ °C

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{PLH} Propagation delay time, low-to-high-level output	TTL load (see Figure 1),		25		ns
t_{PHL} Propagation delay time, high-to-low-level output	See Note 3		25		ns

NOTE 3: The response time specified is for a 100-mV input step with 5-mV overdrive (105 mV total), and is the interval between the input step function and the instant when the output crosses 2.5 V.

PARAMETER MEASUREMENT INFORMATION

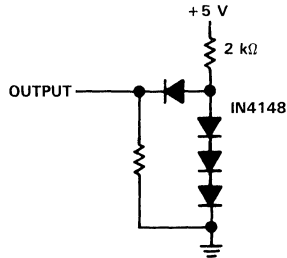


FIGURE 1. TTL OUTPUT LOAD CIRCUIT

TYPICAL CHARACTERISTICS

OUTPUT RESPONSE FOR VARIOUS
INPUT OVERDRIVES

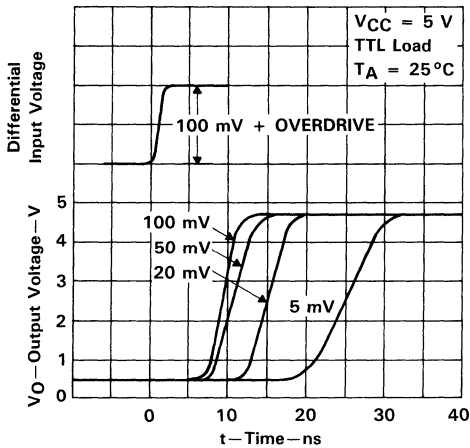


FIGURE 2

OUTPUT RESPONSE FOR VARIOUS
INPUT OVERDRIVES

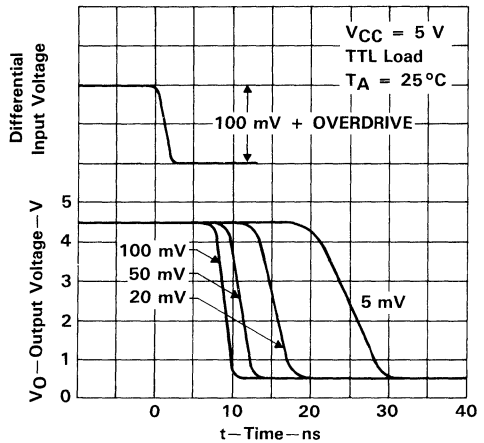


FIGURE 3

3
Voltage Comparators

TYPICAL CHARACTERISTICS

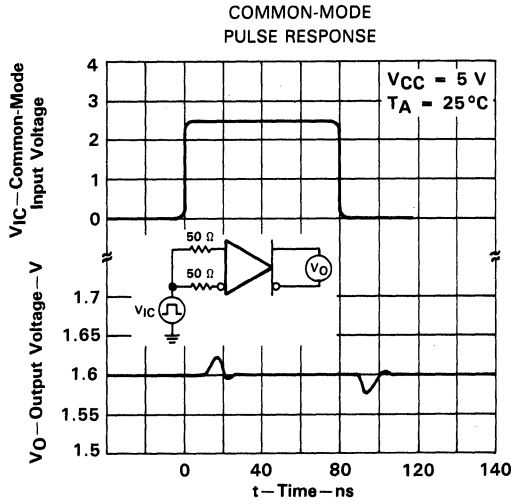


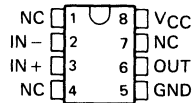
FIGURE 4

TL714C HIGH-SPEED DIFFERENTIAL COMPARATOR

D3131, DECEMBER 1988

- Operates from a 5-V Supply
- Self-Biasing Inputs
- Hysteresis . . . 10 mV Typical
- Response Time . . . 7 ns Typical
- Maximum Operating Frequency . . . 50 MHz Typical

D OR P PACKAGE
(TOP VIEW)



NC—No internal connection

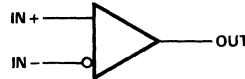
description

The TL714C is a high-speed differential comparator fabricated with bipolar Schottky process technology. The circuit has differential inputs and a TTL-compatible logic output with symmetrical switching characteristics.

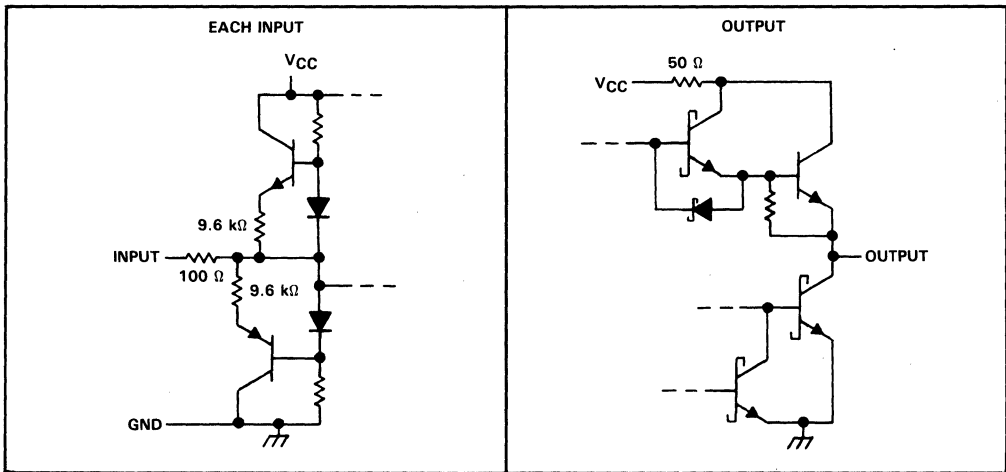
The device operates from a single 5-V supply and is useful as a disk-memory read-chain data comparator.

The TL714C is characterized for operation from 0°C to 70°C.

symbol



schematic of inputs and output



All resistor values shown are nominal.

3
Voltage Comparators

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TL714C HIGH-SPEED DIFFERENTIAL COMPARATOR

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	7 V
Differential input voltage, V_{ID} (see Note 2)	± 5 V
Input voltage range	V_{CC} to GND
Low-level output current, I_{OL}	40 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except for differential voltage, are with respect to the network ground.
2. Differential voltage values are at the noninverting terminal with respect to the inverting terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$ POWER RATING
D	500 mW	5.8 mW/°C	64°C	464 mW
P	500 mW	N/A	N/A	500 mW

recommended operating conditions

PARAMETER	MIN	MAX	UNIT
Supply voltage, V_{CC}	4.75	5.25	V
Common-mode input voltage, V_{IC}	1.4 to $V_{CC} - 1.4$		V
High-level output current, I_{OH}		-1	mA
Low-level output current, I_{OL}		16	mA
Operating free-air temperature, T_A	0	70	°C

electrical characteristics over free-air operating temperature range, $V_{CC} = 5$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP [†]	MAX	UNIT
V_T Threshold voltage	$V_{IC} = 1.4$ V to 3.6 V			± 20	mV
V_{hys} Hysteresis ($V_{T+} - V_{T-}$)		2	10	20	mV
V_{OH} High-level output voltage	$V_{ID} = 100$ mV, $I_{OH} = -1$ mA	2.7	3.4		V
V_{OL} Low-level output voltage	$V_{ID} = -100$ mV, $I_{OL} = 16$ mA		0.4	0.5	V
I_{OS} Short-circuit output current		-15		-85	mA
r_i Differential input resistance		2.9			k Ω
r_o Output resistance				100	Ω
I_{CC} Supply current	$V_{ID} = 0$, $I_O = 0$		7	20	mA

[†]All typical values are at $T_A = 25^\circ\text{C}$.

switching characteristics, $V_{CC} = 5$ V, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
f_{max} Maximum operating frequency	$V_{ID} = \pm 250$ mV, $t_r = t_f = 4$ ns, $C_L = 25$ pF, Input duty cycle = 50%		50		MHz
t_{PLH} Propagation delay time, low-to-high-level output	$V_{ID} = \pm 100$ mV, $C_L = 25$ pF, See Figures 1 and 2		7	25	ns
t_{PHL} Propagation delay time, high-to-low-level output			7	25	ns
t_r Rise time	$V_{ID} = \pm 100$ mV, $C_L = 25$ pF, See Figure 3		4	8	ns
t_f Fall time			4	8	ns

3

Voltage Comparators

PARAMETER MEASUREMENT INFORMATION

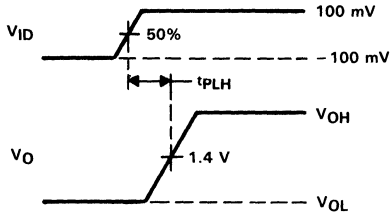


FIGURE 1. PROPAGATION DELAY TIME, LOW TO HIGH (t_{PLH})

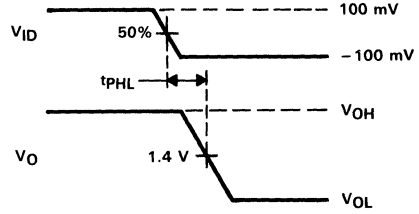


FIGURE 2. PROPAGATION DELAY TIME, HIGH TO LOW (t_{PHL})

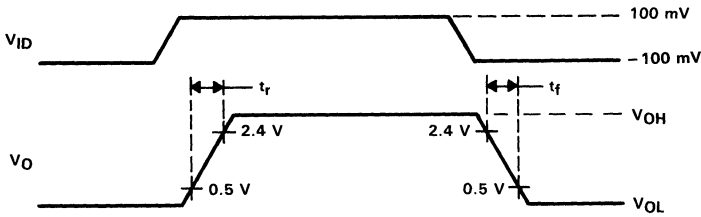


FIGURE 3. RISE AND FALL TIMES (t_r , t_f)

3

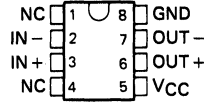
Voltage Comparators

TL721 DIFFERENTIAL COMPARATOR

D2781, FEBRUARY 1984—REVISED OCTOBER 1988

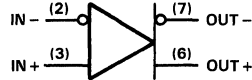
- Operates from a -5.2-V Power Supply
- Self-Biased Inputs
- Common-Mode Input Voltage Range
0 V to -5.2 V
- MECL III and MECL 10 000 Compatible
- Complementary ECL-Compatible Outputs
- Hysteresis . . . 5 mV Typ
- Response Times . . . 10 ns Typ

D, JG, OR P PACKAGE
(TOP VIEW)



NC—No internal connection

symbol



description

The TL721 is a high-speed voltage comparator fabricated with bipolar Schottky[†] process technology. The circuit has differential analog inputs and complementary ECL-compatible logic outputs with symmetrical switching characteristics. The device operates from a single -5.2-volt supply and is useful as a disk memory read-chain data comparator.

The TL721 is characterized for operation from 0°C to 70°C .

3

Voltage Comparators

[†]Integrated Schottky-Barrier diode-clamped transistor is patented by Texas Instruments. U.S. Patent Number 3,463,975.

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3-85

TL721

DIFFERENTIAL COMPARATOR

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	-7 V
Input voltage, any differential input	± 25 V
Differential input voltage (see Note 2)	± 25 V
Low-level output current	50 mA
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C

NOTES: 1. All voltage values, except differential voltages, are with respect to the network ground terminal.
 2. Differential voltage values are at the noninverting terminal with respect to the inverting terminal.

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{CC}		-5.2		V
Common-mode input voltage, V_{IC}			± 7	V
High-level output current, I_{OH}			-1	mA
Low-level output current, I_{OL}			16	mA
Operating free-air temperature, T_A	0		70	°C

electrical characteristics at $T_A = 25^\circ\text{C}$, $V_{CC} = -5.2$ V

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_T Threshold voltage (V_{T+} and V_{T-})	$V_{IC} = V_{ICR}$ min	-100 [†]		100	mV
V_{hys} Hysteresis ($V_{T+} - V_{T-}$)			5		mV
V_{OH} High-level output voltage	$V_{ID} = 100$ mV, $R_L = 50 \Omega$ to -2 V	-0.96 [†]		-0.81	V
V_{OL} Low-level output voltage	$V_{ID} = -100$ mV, $R_L = 50 \Omega$ to -2 V	-1.85 [†]		-1.65	V
V_{ICR} Common-mode input voltage range		0 to -5.2			V
r_{in} Input resistance		4			k Ω
I_{CC} Supply current	$V_{ID} = 0$, No load		-13	-17	mA

[†] The algebraic convention, in which the more negative limit is designated as minimum, is used in this data sheet for input threshold and output voltage levels only.

switching characteristics at $T_A = 25^\circ\text{C}$, $V_{CC} = -5.2$ V

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{PLH} Propagation delay time, low-to-high-level output	$\Delta V_{ID} = +200$ mV to -200 mV or -200 mV to +200 mV,		18		ns
t_{PHL} Propagation delay time, high-to-low-level output	$R_L = 50 \Omega$ to -2 V		18		ns

3 Voltage Comparators

TYPICAL CHARACTERISTICS

OUTPUT RESPONSES FOR VARIOUS
INPUT OVERDRIVES

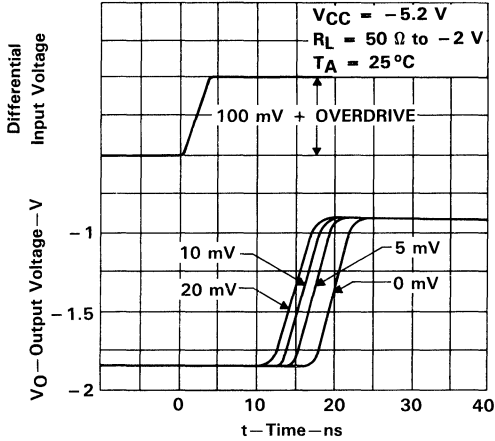


FIGURE 1

OUTPUT RESPONSES FOR VARIOUS
INPUT OVERDRIVES

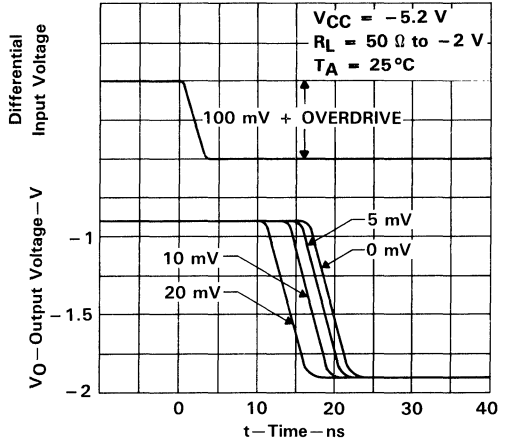


FIGURE 2

COMMON-MODE
PULSE RESPONSE

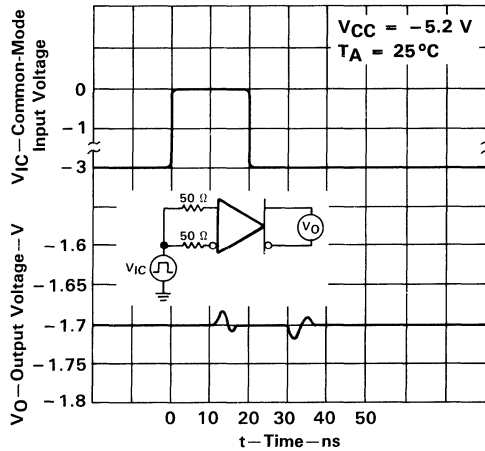


FIGURE 3

3
Voltage Comparators



Voltage Comparators

TLC339M, TLC339I, TLC339C QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

D3135, DECEMBER 1986—REVISED FEBRUARY 1989

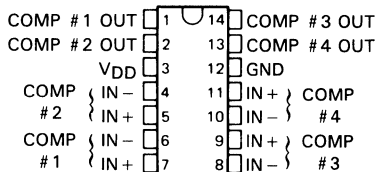
- Very Low Power . . . 200 μ W Typ at 5 V
- Fast Response Time . . . 2.5 μ s Typ with 5 mV Overdrive
- Single Supply Operation:
TLC339M . . . 4 V to 16 V
TLC339I . . . 3 V to 16 V
TLC339C . . . 3 V to 16 V
- High Input Impedance . . . $10^{12} \Omega$ Typ
- Input Offset Voltage Change at Worst Case Input Condition Typically 0.23 μ V/Month Including the First 30 Days
- On-Chip ESD Protection

description

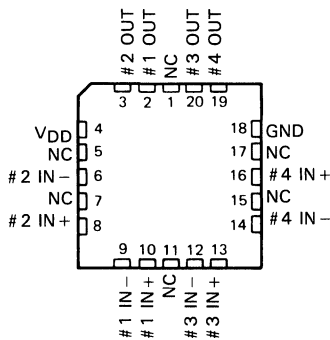
The TLC339 consists of four independent differential-voltage comparators designed to operate from a single supply. It is functionally similar to the LM339 but uses 1/20th the power for similar response times. The open-drain MOS output stage will interface to a variety of loads and supplies, as well as "wired" logic functions. For a similar device with a push-pull output configuration, see the TLC3704 data sheet.

Texas Instruments LinCMOS™ process offers superior analog performance to standard CMOS processes. Along with the standard CMOS advantages of low power without sacrificing speed, high input impedance, and low bias

TLC339M . . . J PACKAGE
TLC339I, TLC339C . . . D, J, OR N PACKAGE
(TOP VIEW)

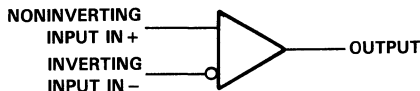


TLC339M . . . FK PACKAGE
(TOP VIEW)



NC—No internal connection

symbol (each comparator)



AVAILABLE OPTIONS

T _A	V _{IO} max at 25°C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	5 mV	TLC339CD	—	TLC339CJ	TLC339CN
-40°C to 85°C	5 mV	TLC339ID	—	TLC339IJ	TLC339IN
-55°C to 125°C	5 mV	—	TLC339MFK	TLC339MJ	—

The D package is available taped and reeled. Add the suffix R to the device type when ordering. (e.g., TLC339CDR)

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3

Voltage Comparators

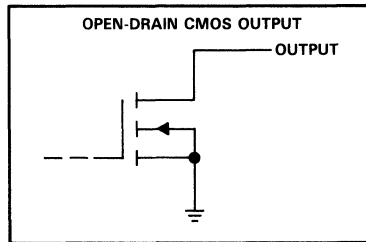
TLC339M, TLC339I, TLC339C QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

description (continued)

currents, the LinCMOS™ process offers extremely stable input offset voltages, even with differential input stresses of several volts. This characteristic makes it possible to build reliable CMOS comparators.

The TLC339M is characterized for operation over the full military temperature range of -55°C to 125°C . The TLC339I is characterized for operation over the extended industrial temperature range of -40°C to 85°C . The TLC339C is characterized for operation over the commercial temperature range of 0°C to 70°C .

schematic



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{DD} (see Note 1)	$-0.3\text{ V to }18\text{ V}$
Differential input voltage (see Note 2)	$\pm 18\text{ V}$
Input voltage, V_I	$-0.3\text{ V to }V_{DD}$
Output voltage, V_O	$-0.3\text{ V to }V_{DD}$
Input current, I_I	$\pm 5\text{ mA}$
Output current, I_O (each output)	20 mA
Total supply current into V_{DD} terminal	40 mA
Total current out of ground terminal	60 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: TLC339M	$-55^{\circ}\text{C to }125^{\circ}\text{C}$
TLC339I	$-40^{\circ}\text{C to }85^{\circ}\text{C}$
TLC339C	$0^{\circ}\text{C to }70^{\circ}\text{C}$
Storage temperature range	$-65^{\circ}\text{C to }150^{\circ}\text{C}$
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.
2. Differential voltages are at the noninverting input with respect to the inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$	DERATING FACTOR	$T_A = 70^{\circ}\text{C}$	$T_A = 85^{\circ}\text{C}$	$T_A = 125^{\circ}\text{C}$
	POWER RATING	ABOVE $T_A = 25^{\circ}\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/ $^{\circ}\text{C}$	608 mW	494 mW	—
FK	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	715 mW	275 mW
J (TLC339M)	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	715 mW	275 mW
J (TLC339I & C)	1025 mW	8.2 mW/ $^{\circ}\text{C}$	656 mW	533 mW	—
N	1150 mW	9.2 mW/ $^{\circ}\text{C}$	736 mW	598 mW	—

TLC339M

QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	4	5	16	V
Common-mode input voltage, V_{IC}	0		$V_{DD}-1.5$	V
Low-level output current, I_{OL}			20	mA
Operating free-air temperature, T_A	-55		125	°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5$ V (unless otherwise noted)

PARAMETER		TEST CONDITIONS†	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	$V_{IC} = V_{ICRmin}$, $V_{DD} = 5$ V to 10 V, See Note 3	25°C	1.4	5	mV
			-55°C to 125°C		10	
I_{IO}	Input offset current	$V_{IC} = 2.5$ V	25°C	1		pA
			125°C		15	nA
I_{IB}	Input bias current	$V_{IC} = 2.5$ V	25°C	5		pA
			125°C		30	nA
V_{ICR}	Common-mode input voltage range		25°C	0 to $V_{DD}-1$		V
			-55°C to 125°C	0 to $V_{DD}-1.5$		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84		dB
			125°C	84		
			-55°C	84		
kSVR	Supply voltage rejection ratio	$V_{DD} = 5$ V to 10 V	25°C	85		dB
			125°C	84		
			-55°C	84		
V_{OL}	Low-level output voltage	$V_{ID} = -1$ V, $I_{OL} = 6$ mA	25°C	300	400	mV
			125°C		800	
I_{OH}	High-level output current	$V_{ID} = 1$ V, $V_O = 5$ V	25°C	0.8	40	nA
			125°C		1	μA
I_{DD}	Supply current (four comparators)	No load, Outputs low	25°C	44	80	μA
			-55°C to 125°C		175	

† All characteristics are measured with zero common-mode voltage unless otherwise noted.

NOTE 3: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V with a 2.5-kΩ load to V_{DD} .

3

Voltage Comparators

TLC339I

QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	3	5	16	V
Common-mode input voltage, V_{IC}	-0.2	$V_{DD}-1.5$		V
Low-level output current, I_{OL}	8		20	mA
Operating free-air temperature, T_A	-40		85	°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5$ V (unless otherwise noted)

PARAMETER		TEST CONDITIONS†	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	$V_{IC} = V_{ICRmin}$, $V_{DD} = 5$ V to 10 V, See Note 4	25°C	1.4	5	mV
			-40°C to 85°C		7	
I_{IO}	Input offset current	$V_{IC} = 2.5$ V	25°C	1		pA
			85°C		1	nA
I_{IB}	Input bias current	$V_{IC} = 2.5$ V	25°C	5		pA
			85°C		2	nA
V_{ICR}	Common-mode input voltage range		25°C	0 to $V_{DD}-1$		V
			-40°C to 85°C	0 to $V_{DD}-1.5$		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84		dB
			85°C	84		
			-40°C	84		
k_{SVR}	Supply voltage rejection ratio	$V_{DD} = 5$ V to 10 V	25°C	85		dB
			85°C	85		
			-40°C	84		
V_{OL}	Low-level output voltage	$V_{ID} = -1$ V, $I_{OL} = 6$ mA	25°C	300	400	mV
			85°C		700	
I_{OH}	High-level output current	$V_{ID} = 1$ V, $V_O = 5$ V	25°C	0.8	40	nA
			85°C		1	
I_{DD}	Supply current (four comparators)	No load, Outputs low	25°C	44	80	μA
			-40°C to 85°C		125	

† All characteristics are measured with zero common-mode voltage unless otherwise noted.

NOTE 4: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.

3

Voltage Comparators

TLC339C

QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	3	5	16	V
Common-mode input voltage, V_{IC}	-0.2		$V_{DD}-1.5$	V
Low-level output current, I_{OL}		8	20	mA
Operating free-air temperature, T_A	0		70	°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	$V_{IC} = V_{ICRmin}$, $V_{DD} = 5\text{ V to }10\text{ V}$, See Note 4	25°C	1.4	5	mV
			0°C to 70°C		6.5	
I_{IO}	Input offset current	$V_{IC} = 2.5\text{ V}$	25°C	1		pA
			70°C		0.3	nA
I_{IB}	Input bias current	$V_{IC} = 2.5\text{ V}$	25°C	5		pA
			70°C		0.6	nA
V_{ICR}	Common-mode input voltage range		25°C	0 to $V_{DD}-1$		V
			0°C to 70°C	0 to $V_{DD}-1.5$		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84		dB
			70°C	84		
			0°C	84		
kSVR	Supply voltage rejection ratio	$V_{DD} = 5\text{ V to }10\text{ V}$	25°C	85		dB
			70°C	85		
			0°C	85		
V_{OL}	Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 6\text{ mA}$	25°C	300	400	mV
			70°C		650	
I_{OH}	High-level output current	$V_{ID} = 1\text{ V}$, $V_O = 5\text{ V}$	25°C	0.8	40	nA
			70°C		1	μA
I_{DD}	Supply current (four comparators)	No load, Outputs low	25°C	44	80	μA
			0°C to 70°C		100	

† All characteristics are measured with zero common-mode voltage unless otherwise noted.

NOTE 4: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.



Voltage Comparators

TLC339M, TLC339I, TLC339C QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

switching characteristics, $V_{DD} = 5\text{ V}$, $T_A = 25^\circ\text{C}$ (see Figure 3)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{PLH}	Propagation delay time, low-to-high level output	$f = 10\text{ kHz}$, $C_L = 15\text{ pF}$	Overdrive = 2 mV		4.5	μs
			Overdrive = 5 mV		2.5	
			Overdrive = 10 mV		1.7	
			Overdrive = 20 mV		1.2	
			Overdrive = 40 mV		1.0	
			$V_I = 1.4\text{ V}$ step at IN+ pin		1.1	
t_{PHL}	Propagation delay time, high-to-low level output	$f = 10\text{ kHz}$, $C_L = 15\text{ pF}$	Overdrive = 2 mV		3.6	μs
			Overdrive = 5 mV		2.1	
			Overdrive = 10 mV		1.3	
			Overdrive = 20 mV		0.85	
			Overdrive = 40 mV		0.55	
			$V_I = 1.4\text{ V}$ step at IN+ pin		0.10	
t_{THL}	Transition time, high-to-low level output	$f = 10\text{ kHz}$, $C_L = 15\text{ pF}$	Overdrive = 50 mV		20	ns

PARAMETER MEASUREMENT INFORMATION

3

Voltage Comparators

The TLC339 contains a digital output stage that, if held in the linear region of the transfer curve, can cause damage to the device. Conventional operational amplifier/comparator testing incorporates the use of a servo-loop that is designed to force the device output to a level within this linear region. Since the servo-loop method of testing cannot be used, the following alternatives for testing parameters such as input offset voltage, common-mode rejection, etc., are suggested.

To verify that the input offset voltage falls within the limits specified, the limit value is applied to the input as shown in Figure 1(a). With the noninverting input positive with respect to the inverting input, the output should be high. With the input polarity reversed, the output should be low.

A similar test can be made to verify the input offset voltage at the common-mode extremes. The supply voltages can be slewed as shown in Figure 1(b) for the V_{ICR} test, rather than changing the input voltages, to provide greater accuracy.

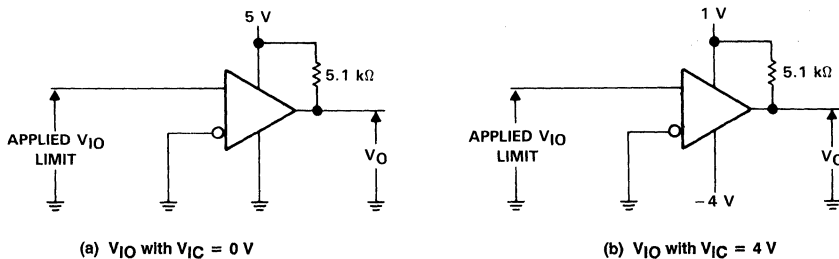


FIGURE 1. METHOD FOR VERIFYING THAT INPUT OFFSET VOLTAGE IS WITHIN SPECIFIED LIMITS

A close approximation of the input offset voltage can be obtained by using a binary search method to vary the differential input voltage while monitoring the output state. When the applied input voltage differential is equal but opposite in polarity to the input offset voltage, the output will change states.

PARAMETER MEASUREMENT INFORMATION

Figure 2 illustrates a practical circuit for direct dc measurement of input offset voltage that does not bias the comparator into the linear region. The circuit consists of a switching mode servo loop in which U1A generates a triangular waveform of approximately 20-mV amplitude. U1B acts as a buffer, with C2 and R4 removing any residual dc offset. The signal is then applied to the inverting input of the comparator under test, while the noninverting input is driven by the output of the integrator formed by U1C through the voltage divider formed by R9 and R10. The loop reaches a stable operating point when the output of the comparator under test has a duty cycle of exactly 50%, which can only occur when the incoming triangle wave is "sliced" symmetrically or when the voltage at the noninverting input exactly equals the input offset voltage.

Voltage divider R9 and R10 provides a step-up of the input offset voltage by a factor of 100 to make measurement easier. The values of R5, R8, R9, and R10 can significantly influence the accuracy of the reading; therefore, it is suggested that their tolerance level be 1% or lower.

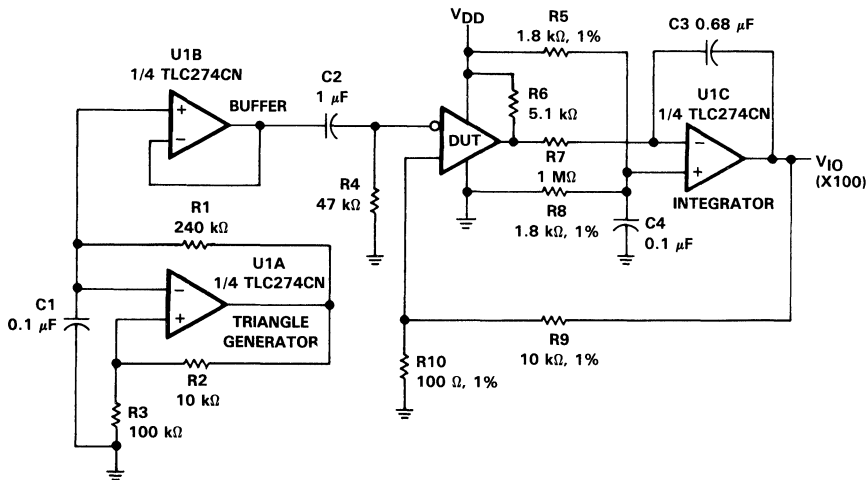


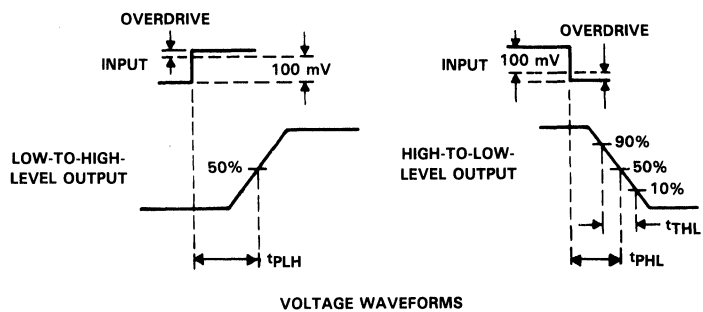
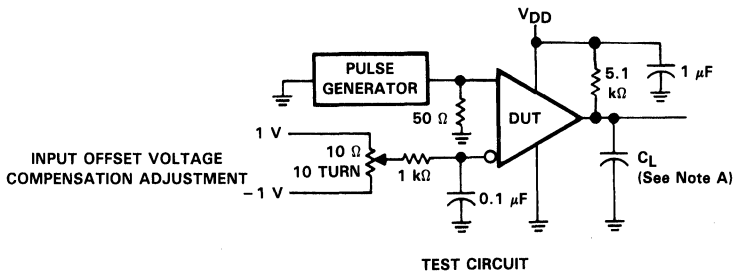
FIGURE 2. CIRCUIT FOR INPUT OFFSET VOLTAGE MEASUREMENT

Measuring the extremely low values of input current requires isolation from all other sources of leakage current and compensation for the leakage of the test socket and board. With a good picoammeter, the socket and board leakage can be measured with no device in the socket. Subsequently, this open socket leakage value can be subtracted from the measurement obtained with a device in the socket to obtain the actual input current of the device.

TLC339M, TLC339I, TLC339C
QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

PARAMETER MEASUREMENT INFORMATION

Propagation delay time is defined as the interval between the application of an input step function and the instant when the output reaches 50% of its maximum value. Propagation delay time, low-to-high-level output, is measured from the leading edge of the input pulse, while propagation delay time, high-to-low-level output, is measured from the trailing edge of the input pulse. Propagation delay time measurement at low input signal levels can be greatly affected by the input offset voltage. The offset voltage should be balanced by the adjustment at the inverting input (as shown in Figure 3) so that the circuit is just at the transition point. Then a low signal, for example 105-mV or 5-mV overdrive, will cause the output to change state.



NOTE A: C_L includes probe and jig capacitance.

FIGURE 3. PROPAGATION DELAY, RISE, AND FALL TIMES CIRCUIT AND VOLTAGE WAVEFORMS

3
Voltage Comparators

TLC339M, TLC339I, TLC339C
 QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

TYPICAL CHARACTERISTICS†

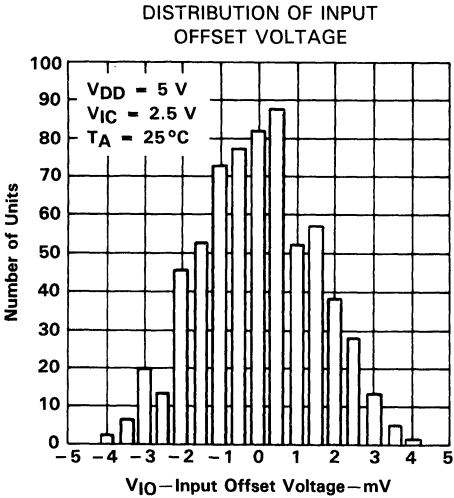


FIGURE 4

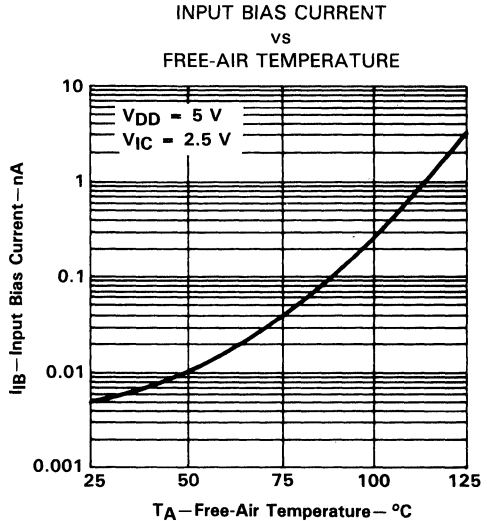


FIGURE 5

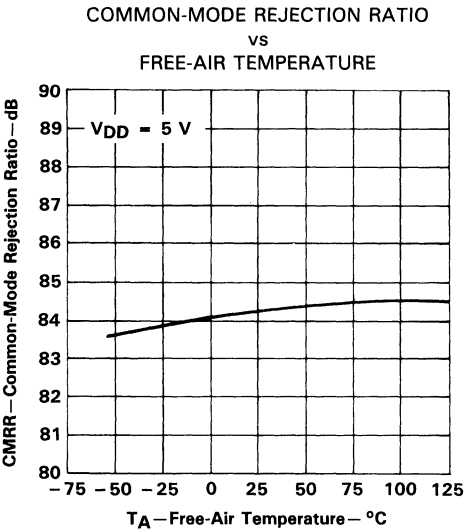


FIGURE 6

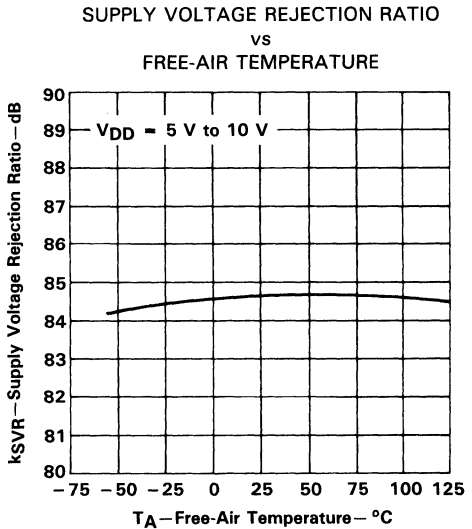


FIGURE 7

3

Voltage Comparators

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC339M, TLC339I, TLC339C
QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT CURRENT
 vs
 HIGH-LEVEL OUTPUT VOLTAGE

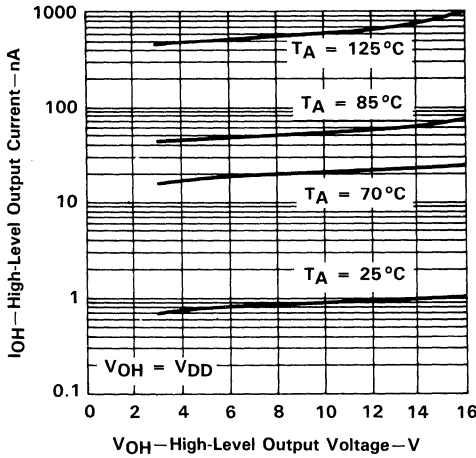


FIGURE 8

HIGH-LEVEL OUTPUT CURRENT
 vs
 FREE-AIR TEMPERATURE

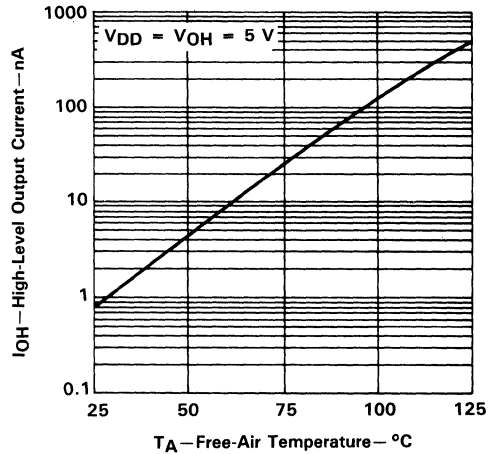


FIGURE 9

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

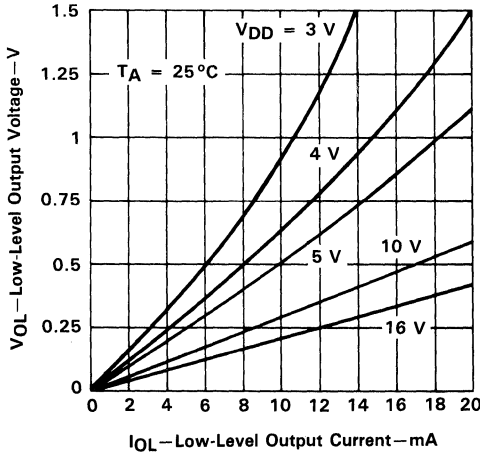


FIGURE 10

LOW-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

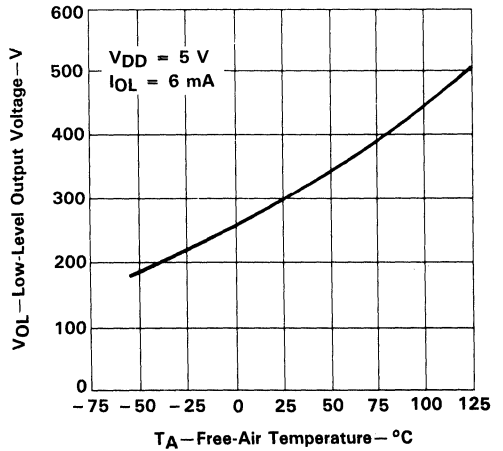


FIGURE 11

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TLC339M, TLC339I, TLC339C
QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

TYPICAL CHARACTERISTICS†

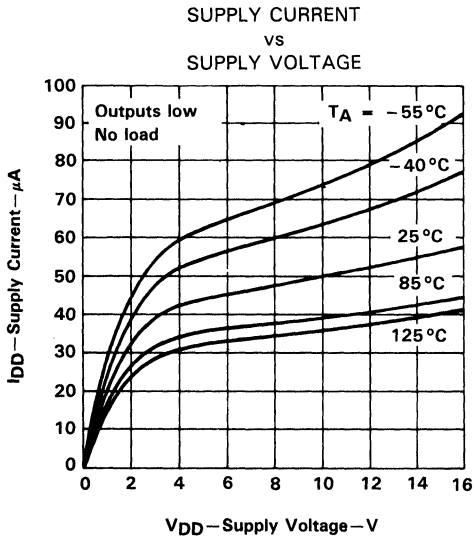


FIGURE 12

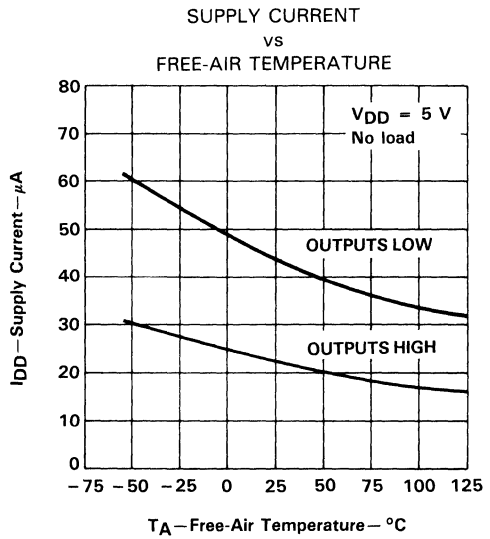


FIGURE 13

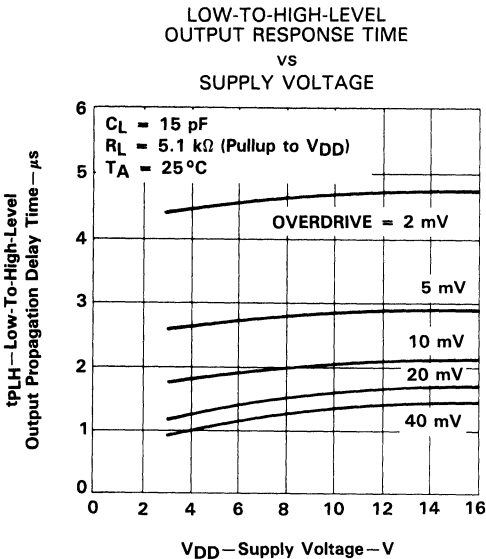


FIGURE 14

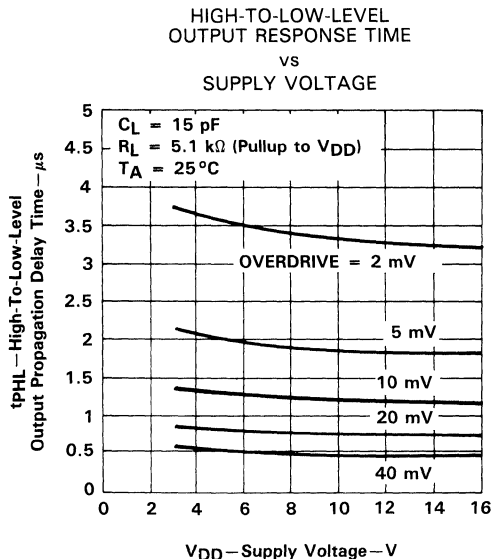


FIGURE 15

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC339M, TLC339I, TLC339C
QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

TYPICAL CHARACTERISTICS

**LOW-TO-HIGH-LEVEL OUTPUT
 PROPAGATION DELAY
 FOR VARIOUS OVERDRIVE VOLTAGES**

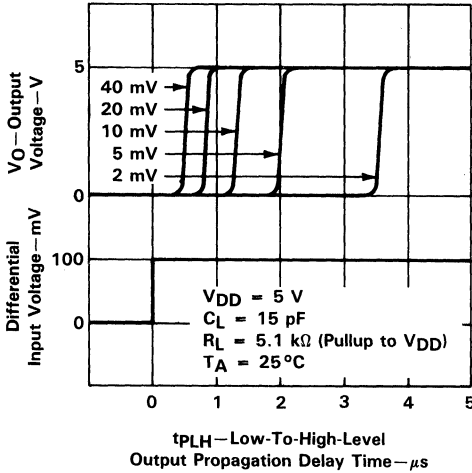


FIGURE 16

**OUTPUT FALL TIME
 vs
 SUPPLY VOLTAGE**

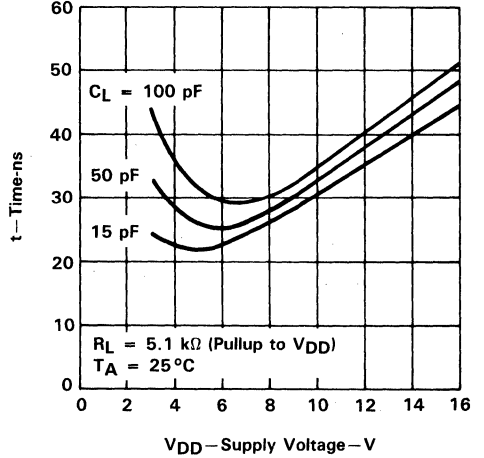


FIGURE 17

**HIGH-TO-LOW-LEVEL OUTPUT
 PROPAGATION DELAY
 FOR VARIOUS OVERDRIVE VOLTAGES**

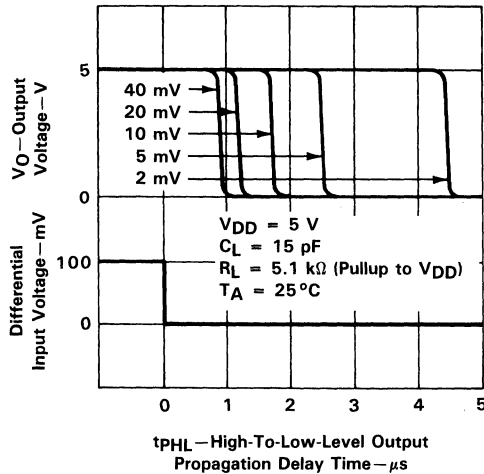


FIGURE 18

3

Voltage Comparators

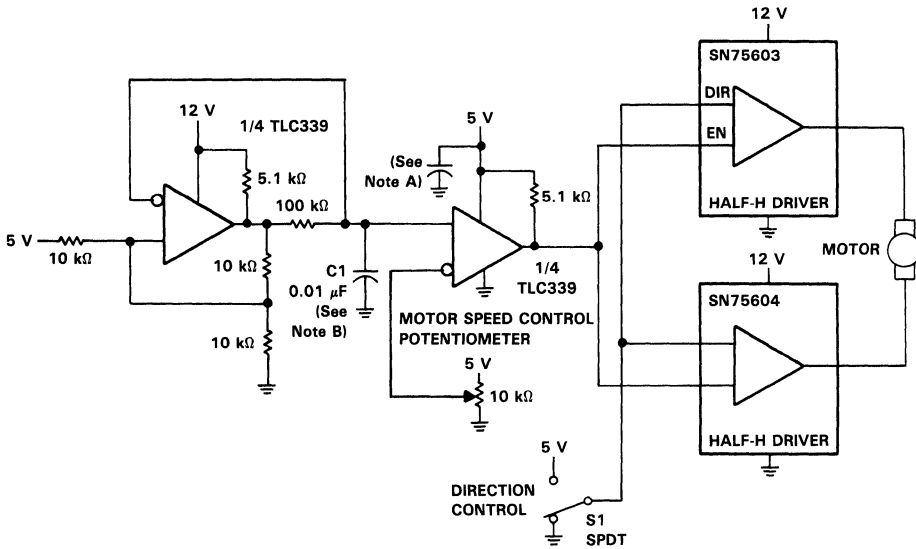
TYPICAL APPLICATION DATA

The inputs should always remain within the supply rails in order to avoid forward biasing the diodes in the electrostatic discharge (ESD) protection structure. If either input exceeds this range, the device will not be damaged as long as the input current is limited to less than 5 mA. To maintain the expected output state, the inputs must remain within the common-mode range. For example, at 25°C with $V_{DD} = 5\text{ V}$, both inputs must remain between -0.2 V and 4 V to assure proper device operation.

To assure reliable operation, the supply should be decoupled with a capacitor ($0.1\text{ }\mu\text{F}$) positioned as close to the device as possible.

Be careful to note the output and supply current limitations since the TLC339 does not provide current protection. For example, each output can source or sink a maximum of 20 mA; however, the total current to ground can only be an absolute maximum of 60 mA. This prohibits sinking 20 mA from each of the four outputs simultaneously since the total current to ground would be 80 mA.

The TLC339 has internal ESD protection circuits that will prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.



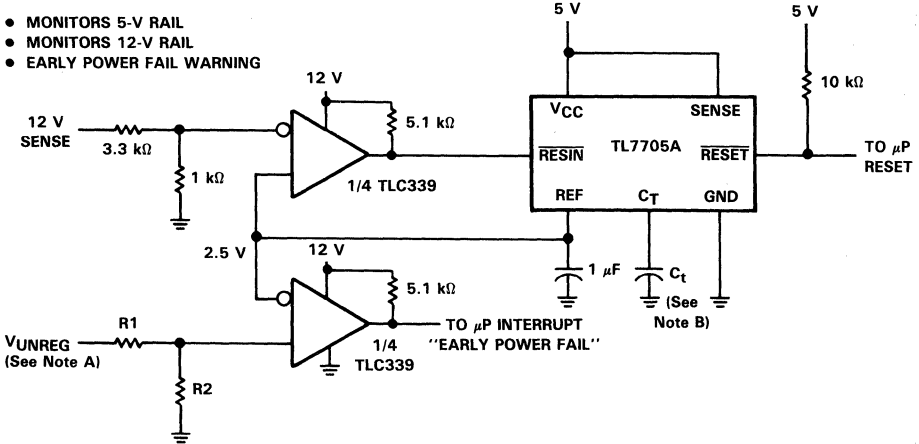
NOTES: A. The recommended minimum capacitance is $10\text{ }\mu\text{F}$ to eliminate common ground switching noise.
 B. Select C1 for change in oscillator frequency.

FIGURE 19. PULSE-WIDTH-MODULATED MOTOR SPEED CONTROLLER

TLC339M, TLC339I, TLC339C
QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

TYPICAL APPLICATION DATA

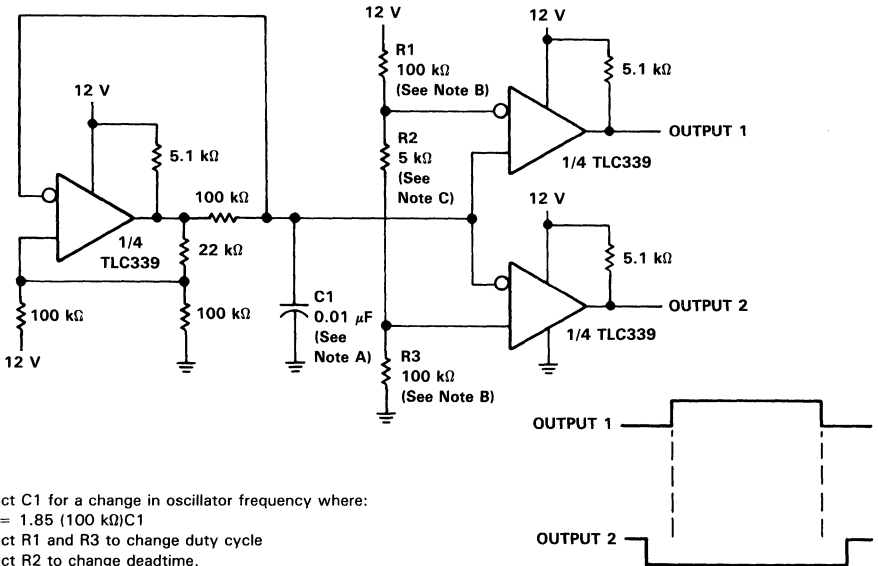
- MONITORS 5-V RAIL
- MONITORS 12-V RAIL
- EARLY POWER FAIL WARNING



NOTES: A. $V_{UNREG} = 2.5 \left(\frac{R1 + R2}{R2} \right)$

B. The value of C_T determines the time delay of reset.

FIGURE 20. ENHANCED SUPPLY SUPERVISOR



NOTES: A. Select $C1$ for a change in oscillator frequency where:

$1/f = 1.85 (100 \text{ k}\Omega) C1$

B. Select $R1$ and $R3$ to change duty cycle

C. Select $R2$ to change deadtime.

FIGURE 21. TWO-PHASE NONOVERLAPPING CLOCK GENERATOR

3

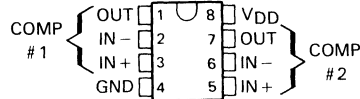
Voltage Comparators

TLC352M, TLC352I, TLC352C LinCMOS™ DUAL DIFFERENTIAL COMPARATORS

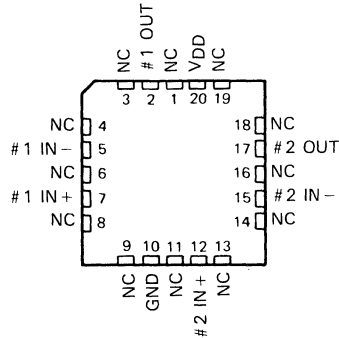
D2901, SEPTEMBER 1985—REVISED FEBRUARY 1989

- Single- or Dual-Supply Operation
- Wide Range of Supply Voltages . . . 1.4 V to 18 V
- Very Low Supply Current Drain
150 μA Typ at 5 V
65 μA Typ at 1.4 V
- Built-In ESD Protection
- High Input Impedance . . . $10^{12} \Omega$ Typ
- Extremely Low Input Bias Current 5 pA Typ
- Ultrastable Low Input Offset Voltage
- Input Offset Voltage Change at Worst-Case Input Conditions Typically 0.23 $\mu\text{V}/\text{Month}$, Including the First 30 Days
- Common-Mode Input Voltage Range Includes Ground
- Outputs Compatible with TTL, MOS, and CMOS
- Pin-Compatible with LM393

TLC352M . . . JG PACKAGE
TLC352I, TLC352C . . . D, JG, OR P PACKAGE
(TOP VIEW)

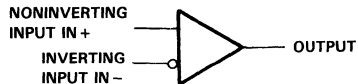


TLC352M . . . FK PACKAGE
(TOP VIEW)



NC—No internal connection

symbol (each comparator)



description

This device is fabricated using LinCMOS™ technology and consists of two independent voltage comparators, each designed to operate from a single power supply. Operation from dual supplies is also possible so long as the difference between the two supplies is 1.4 V to 18 V. Each device features extremely high input impedance (typically greater than $10^{12} \Omega$), which allows direct interface to high-impedance sources. The outputs are n-channel open-drain configurations and can be connected to achieve positive-logic wired-AND relationships. The capability of the TLC352 to operate from a 1.4-V supply makes this device ideal for low-voltage battery applications.

The TLC352 has internal electrostatic discharge (ESD) protection circuits and has been classified with a 2000-V ESD rating tested under MIL-STD-883C, Method 3015. However, care should be exercised in handling this device as exposure to ESD may result in degradation of the device parametric performance.

The TLC352M is characterized for operation over the full military temperature range of -55°C to 125°C . The TLC352I is characterized for operation over the industrial temperature range of -40°C to 85°C . The TLC352C is characterized for operation from 0°C to 70°C .

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PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS
INSTRUMENTS**

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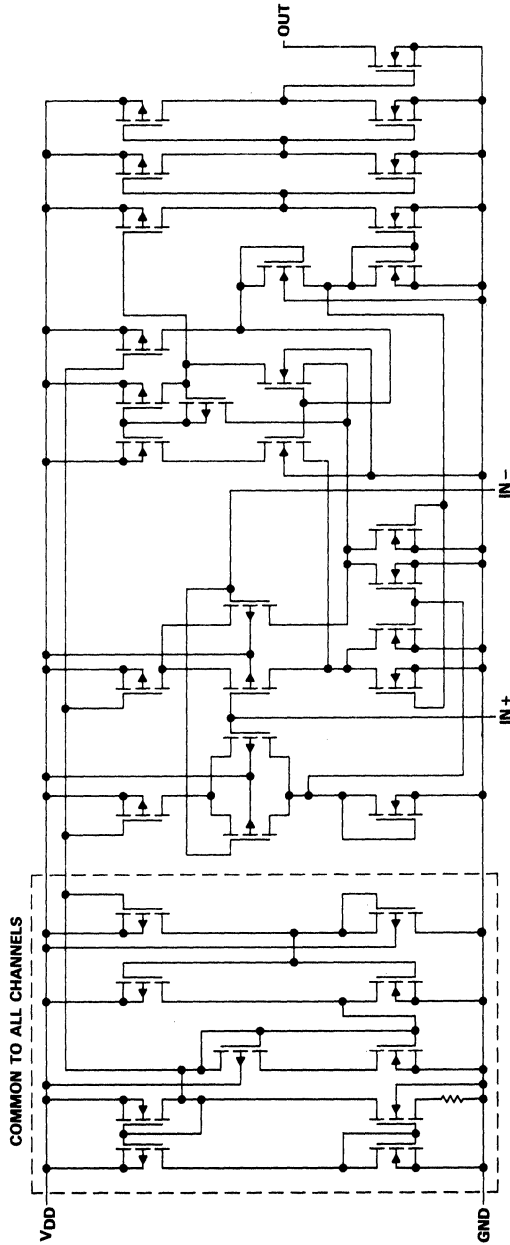
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3
Voltage Comparators

TLC352M, TLC352I, TLC352C
LinCMOS™ DUAL DIFFERENTIAL COMPARATORS

equivalent schematic (each comparator)



3

Voltage Comparators

TLC352M, TLC352I, TLC352C LinCMOS™ DUAL DIFFERENTIAL COMPARATORS

AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE			
		SMALL-OUTLINE (D)	CHIP-CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	5 mV	TLC352CD	—	TLC352CJG	TLC352CP
-40°C to 85°C	5 mV	TLC352ID	—	TLC352IJG	TLC352IP
-55°C to 125°C	5 mV	—	TLC352MFK	TLC352MJG	—

D packages are available taped and reeled. Add "R" suffix to device type when ordering (e.g., TLC352CDR).

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V _{DD} (see Note 1)	18 V
Differential input voltage, V _{ID} (see Note 2)	±18 V
Input voltage, V _I	V _{DD}
Input voltage range	-0.3 V to 18 V
Output voltage, V _O	18 V
Input current, I _I	±5 mA
Output current, I _O	20 mA
Duration of output short-circuit to ground (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: TLC352M	-55°C to 125°C
TLC352I	-40°C to 85°C
TLC352C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C

- NOTES: 1. All voltage values except differential voltages are with respect to network ground.
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. Short circuits from outputs to V_{DD} can cause excessive heating and eventual device destruction.

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C	DERATING FACTOR	DERATE ABOVE T _A	T _A = 70°C	T _A = 85°C	T _A = 125°C
	POWER RATING			POWER RATING	POWER RATING	POWER RATING
D	500 mW	5.8 mW/°C	64°C	464 mW	377 mW	N/A
FK	500 mW	11.0 mW/°C	104°C	500 mW	500 mW	275 mW
JG (TLC352M)	500 mW	8.4 mW/°C	90°C	500 mW	500 mW	210 mW
JG (TLC352I, TLC352C)	500 mW	6.6 mW/°C	74°C	500 mW	429 mW	N/A
P	500 mW	N/A	N/A	500 mW	500 mW	N/A

3

Voltage Comparators

Voltage Comparators



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recommended operating conditions

		M-SUFFIX		I-SUFFIX			C-SUFFIX			UNIT	
		MIN	NOM	MAX	MIN	NOM	MAX	MIN	NOM		MAX
Supply voltage, V_{DD}		4		16	3		16	3		16	V
Common-mode input voltage, V_{IC}	$V_{DD} = 5\text{ V}$	0		3.5	0		3.5	0		3.5	V
	$V_{DD} = 10\text{ V}$	0		8.5	0		8.5	0		8.5	
Operating free-air temperature, T_A		-55		125	-40		85	0		70	°C

electrical characteristics at specified free-air temperature, $V_{DD} = 1.4\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TLC352M			TLC352I			TLC352C			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX		
V_{IO} Input offset voltage	$V_{IC} = V_{ICR}\text{ min.}$ See Note 4	25°C	2		5	2		5	2		5	mV
		Full range			10			7			6.5	
I_{IO} Input offset current		25°C		1			1			1		pA
		MAX T_A			10			1			0.3	nA
I_{IB} Input bias current		25°C		5			5			5		pA
		MAX T_A			20			2			0.6	nA
V_{ICR} Common-mode input voltage range		Full range	0 to 0.2		0 to 0.2		0 to 0.2		0 to 0.2		V	
V_{OL} Low-level output voltage	$V_{ID} = -0.5\text{ V}, I_{OL} = 0.6\text{ mA}$	25°C		100		200		100		200		mV
		Full range			200					200		
I_{OL} Low-level output current	$V_{ID} = -0.5\text{ V}, V_{OL} = 0.3\text{ V}$	25°C	1	1.6		1	1.6		1	1.6	mA	
I_{DD} Supply current (two comparators)	$V_{DD} = 0.5\text{ V},$ No load	25°C		65		150		65		150		μA
		Full range			200					200		

†All characteristics are measured with zero common-mode input voltage unless otherwise noted. Full range is -55°C to 125°C for TLC352M, 0°C to 70°C for TLC352C, and -40°C to 85°C for TLC352I. IMPORTANT: See Parameter Measurement Information.

NOTE 4: The offset voltage limits given are the maximum values required to drive the output above 1.25 V or below 150 mV with a 10-kΩ resistor between the output and V_{DD} . They can be verified by applying the limit value to the input and checking for the appropriate output state.

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†		TLC352M			TLC352I			TLC352C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = V_{ICR\ min}$, See Note 5	25 °C	1		5	1		5	1		5	mV
			Full range			7			6.5			
I_{IO} Input offset current		25 °C	1			1			1			pA
			MAX T_A		10	1			0.3			nA
I_{IB} Input bias current		25 °C	5			5			5			pA
			MAX T_A		20	2			0.6			nA
V_{ICR} Common-mode input voltage range		25 °C	0 to $V_{DD}-1$			0 to $V_{DD}-1$			0 to $V_{DD}-1$			V
			Full range			0 to $V_{DD}-1.5$			0 to $V_{DD}-1.5$			
I_{OH} High-level output current	$V_{ID} = 1\text{ V}$	$V_{OH} = 5\text{ V}$	25 °C		0.1		0.1		0.1		nA	
		$V_{OH} = 15\text{ V}$	Full range			1		1		1		μA
V_{OL} Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 4\text{ mA}$	25 °C	150		400	150		400	150		400	mV
			Full range			700			700			
I_{OL} Low-level output current	$V_{ID} = -1\text{ V}$, $V_{OL} = 1.5\text{ V}$	25 °C	6	16		6	16		6	16	mA	
I_{DD} Supply current (two comparators)	$V_{DD} = 1\text{ V}$, No load	25 °C	0.15		0.3	0.15		0.3	0.15		0.3	μA
			Full range			0.4			0.4			

† All characteristics are measured with zero common-mode input voltage unless otherwise noted. Full range is -55 °C to 125 °C for TLC352M, 0 °C to 70 °C for TLC352C, and -40 °C to 85 °C for TLC352I. IMPORTANT: See Parameter Measurement Information.

NOTE 5: The offset voltage limits given are the maximum values required to drive the output above 4 V or below 400 mV with a 10-kΩ resistor between the output and V_{DD} . They can be verified by applying the limit value to the input and checking for the appropriate output state.

switching characteristics, $V_{DD} = 5\text{ V}$, $T_A = 25\text{ °C}$

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
Response time	R_L connected to 5 V through 5.1 kΩ, $C_L = 15\text{ pF}^\ddagger$, See Note 5	100-mV input step with 5-mV overdrive		650		ns
		TTL-level input step		200		

‡ C_L includes probe and jig capacitance.

NOTE 5: The response time specified is the interval between the input step function and the instant when the output crosses 1.4 V.

LINCMOS™ DUAL DIFFERENTIAL COMPARATORS
 TLC352M, TLC352I, TLC352C

PARAMETER MEASUREMENT INFORMATION

The digital output stage of the TLC352 can be damaged if it is held in the linear region of the transfer curve. Conventional operational amplifier/comparator testing incorporates the use of a servo-loop that is designed to force the device output to a level within this linear region. Since the servo-loop method of testing cannot be used, the following alternative for measuring parameters such as input offset voltage, common-mode rejection, etc., are offered.

To verify that the input offset voltage falls within the limits specified, the limit value is applied to the input as shown in Figure 1 (a). With the noninverting input positive with respect to the inverting input, the output should be high. With the input polarity reversed, the output should be low.

A similar test can be made to verify the input offset voltage at the common-mode extremes. The supply voltages can be slewed as shown in Figure 1 (b) for the V_{ICR} test, rather than changing the input voltages, to provide greater accuracy.

A close approximation of the input offset voltage can be obtained by using a binary search method to vary the differential input voltage while monitoring the output state. When the applied input voltage differential is equal, but opposite in polarity, to the input offset voltage, the output will change states.

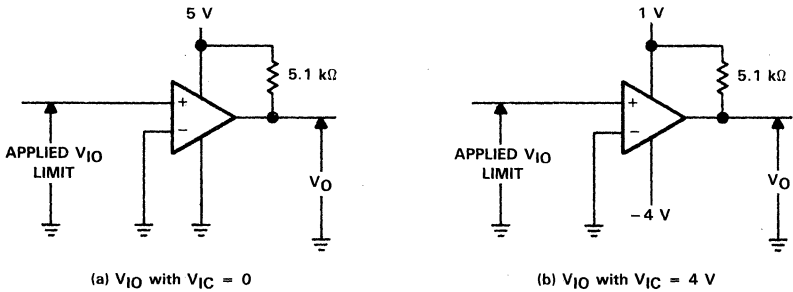


FIGURE 1. METHOD FOR VERIFYING THAT INPUT OFFSET VOLTAGE IS WITHIN SPECIFIED LIMITS

3
Voltage Comparators

PARAMETER MEASUREMENT INFORMATION

Figure 2 illustrates a practical circuit for direct dc measurement of input offset voltage that does not bias the comparator into the linear region. The circuit consists of a switching-mode servo loop in which U1a generates a triangular waveform of approximately 20-mV amplitude. U1b acts as a buffer, with C2 and R4 removing any residual dc offset. The signal is then applied to the inverting input of the comparator under test, while the noninverting input is driven by the output of the integrator formed by U1c through the voltage divider formed by R9 and R10. The loop reaches a stable operating point when the output of the comparator under test has a duty cycle of exactly 50%, which can only occur when the incoming triangle wave is "sliced" symmetrically or when the voltage at the noninverting input exactly equals the input offset voltage.

Voltage divider R9 and R10 provides a step-up of the input offset voltage by a factor of 100 to make measurement easier. The values of R5, R8, R9, and R10 can significantly influence the accuracy of the reading; therefore, it is suggested that their tolerance level be 1% or lower.

Measuring the extremely low values of input current requires isolation from all other sources of leakage current and compensation for the leakage of the test socket and board. With a good picoammeter, the socket and board leakage can be measured with no device in the socket. Subsequently, this open-socket leakage value can be subtracted from the measurement obtained with a device in the socket to obtain the actual input current of the device.

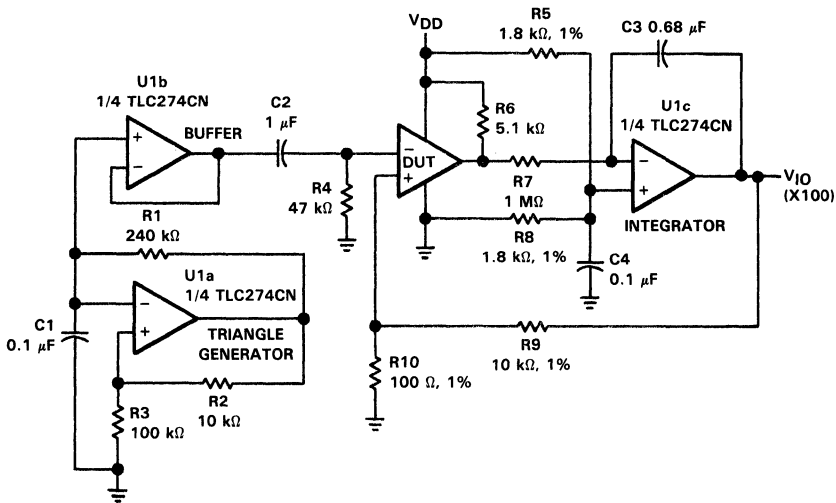
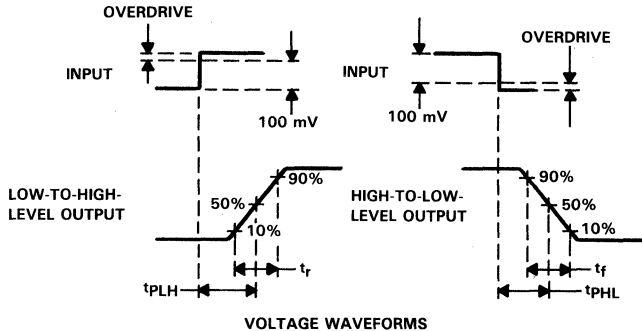
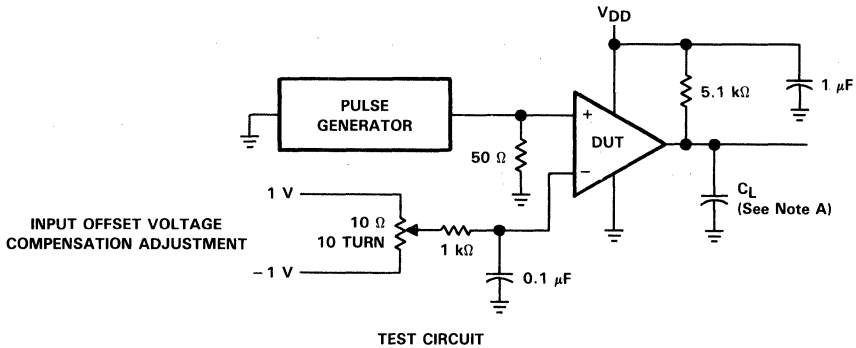


FIGURE 2. CIRCUIT FOR INPUT OFFSET VOLTAGE MEASUREMENT

3
 Voltage Comparators

PARAMETER MEASUREMENT INFORMATION

Response time is defined as the interval between the application of an input step function and the instant when the output reaches 50% of its maximum value. Response time, low-to-high-level output, is measured from the leading edge of the input pulse, while response time, high-to-low-level output, is measured from the trailing edge of the input pulse. Response-time measurement at low input signal levels can be greatly affected by the input offset voltage. The offset voltage should be balanced by the adjustment at the inverting input (as shown in Figure 3) so that the circuit is just at the transition point. Then a low signal, for example 105-mV or 5-mV overdrive, will cause the output to change state.



NOTE A: C_L includes probe and jig capacitance.

FIGURE 3. RESPONSE, RISE, AND FALL TIMES CIRCUIT AND VOLTAGE WAVEFORMS

LinCMOS™ QUADRUPLE DIFFERENTIAL COMPARATORS

D2901, SEPTEMBER 1985—REVISED FEBRUARY 1989

- Single- or Dual-Supply Operation
- Wide Range of Supply Voltages . . . 1.4 V to 18 V
- Very Low Supply Current Drain
300 μA Typ at 5 V
130 μA Typ at 1.4 V
- Built-In ESD Protection
- High Input Impedance . . . 10^{12} Typ
- Extremely Low Input Bias Current 5 pA Typ
- Ultrastable Low Input Offset Voltage
- Input Offset Voltage Change at Worst-Case Input Conditions Typically 0.23 $\mu\text{V}/\text{Month}$, Including the First 30 Days
- Common-Mode Input Voltage Range Includes Ground
- Outputs Compatible with TTL, MOS, and CMOS
- Pin-Compatible with LM339

description

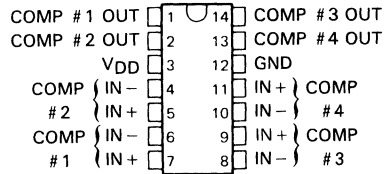
This device is fabricated using LinCMOS™ technology and consists of four independent voltage comparators; each designed to operate from a single power supply. Operation from dual supplies is also possible so long as the difference between the two supplies is 1.4 V to 18 V. Each device features extremely high input impedance (typically greater than $10^{12} \Omega$), which allows direct interface to high-impedance sources. The outputs are n-channel open-drain configurations and can be connected to achieve positive-logic wired-AND relationships. The capability of the TLC354 to operate from a 1.4-V supply makes this device ideal for low-voltage battery applications.

The TLC354 has internal electrostatic discharge (ESD) protection circuits and has been classified with a 2000-V ESD rating tested under MIL-STD-883C, Method 3015. However, care should be exercised in handling this device as exposure to ESD may result in degradation of the device parametric performance.

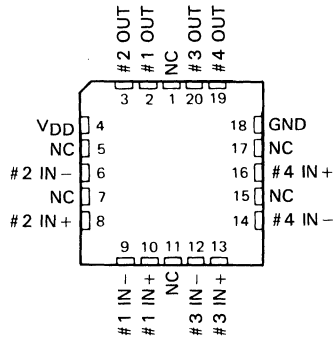
The TLC354M is characterized for operation over the full military temperature range of -55°C to 125°C . The TLC354I is characterized for operation over the industrial temperature range of -40°C to 85°C . The TLC354C is characterized for operation from 0°C to 70°C .

LinCMOS is a trademark of Texas Instruments Incorporated.

TLC354M . . . J PACKAGE
TLC354I, TLC354C . . . D OR N PACKAGE
(TOP VIEW)

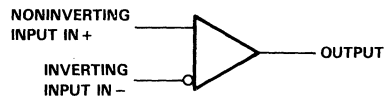


TLC354M . . . FK PACKAGE
(TOP VIEW)



NC—No internal connection

symbol (each comparator)



3

Voltage Comparators

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS
INSTRUMENTS

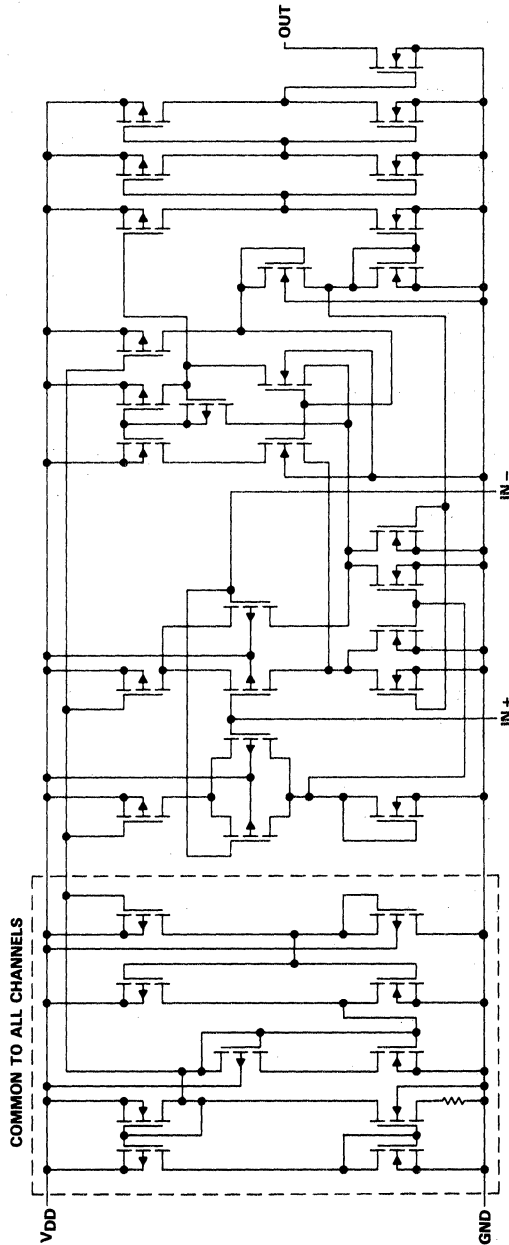
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POST OFFICE BOX 655012 • DALLAS, TEXAS 75265

TLC354M, TLC354I, TLC354C
LinCMOS™ QUADRUPLE DIFFERENTIAL COMPARATORS

equivalent schematic (each comparator)



Voltage Comparators

TLC354M, TLC354I, TLC354C LinCMOS™ QUADRUPLE DIFFERENTIAL COMPARATORS

AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE			
		SMALL-OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	5 mV	TLC354CD	—	TLC354CJ	TLC354CN
-40°C to 85°C	5 mV	TLC354ID	—	TLC354IJ	TLC354IN
-55°C to 125°C	5 mV	—	TLC354MFK	TLC354MJ	—

D packages are available taped and reeled. Add "R" suffix to device type when ordering (e.g., TLC354CDR).

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V _{DD} (see Note 1)	18 V
Differential input voltage, V _{ID} (see Note 2)	±18 V
Input voltage, V _I	V _{DD}
Input voltage range	-0.3 V to 18 V
Output voltage, V _O	18 V
Input current, I _I	±5 mA
Output current, I _O	20 mA
Duration of output short-circuit to ground (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: TLC354M	-55°C to 125°C
TLC354I	-40°C to 85°C
TLC354C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C

- NOTES: 1. All voltage values except differential voltages are with respect to network ground.
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. Short circuits from outputs to V_{DD} can cause excessive heating and eventual device destruction.

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C	DERATING FACTOR	DERATE ABOVE T _A	T _A = 70°C	T _A = 85°C	T _A = 125°C
	POWER RATING			POWER RATING	POWER RATING	POWER RATING
D	500 mW	7.6 mW/°C	84°C	500 mW	494 mW	N/A
FK	500 mW	11.0 mW/°C	104°C	500 mW	500 mW	275 mW
J (TLC354M)	500 mW	11.0 mW/°C	104°C	500 mW	500 mW	275 mW
J (TLC354I, TLC354C)	500 mW	N/A	N/A	500 mW	500 mW	N/A
N	500 mW	N/A	N/A	500 mW	500 mW	N/A

3

Voltage Comparators

Voltage Comparators



recommended operating conditions

	M-SUFFIX			I-SUFFIX			C-SUFFIX			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, V_{DD}	1.4		16	1.4		16	1.4		16	V
Common-mode input voltage, V_{IC}	$V_{DD} = 5\text{ V}$	0	3.5	0	3.5	0	3.5	0	3.5	V
	$V_{DD} = 10\text{ V}$	0	8.5	0	8.5	0	8.5	0	8.5	
Operating free-air temperature, T_A	-55		125	-40		85	0		70	°C

electrical characteristics at specified free-air temperature, $V_{DD} = 1.4\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†		TLC354M			TLC354I			TLC354C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = V_{ICR\text{ min}}$, See Note 4	25°C		2	5		2	5		2	5	mV
		Full range			10			7			6.5	
I_{IO} Input offset current		25°C		1			1			1		pA
		MAX T_A			10			1			0.3	nA
I_{IB} Input bias current		25°C		5			5			5		pA
		MAX T_A			20			2			0.6	nA
V_{ICR} Common-mode input voltage range		25°C	0 to 0.2			0 to 0.2			0 to 0.2			V
I_{OH} High-level output current	$V_{ID} = 1\text{ V}$	$V_{OH} = 5\text{ V}$	25°C		0.1		0.1			0.1		nA
		$V_{OH} = 15\text{ V}$	Full range			1			1		1	μA
V_{OL} Low-level output voltage	$V_{ID} = -0.5\text{ V}$, $I_{OL} = 0.6\text{ mA}$	25°C		100	200		100	200		100	200	mV
		Full range			200			200			200	
I_{OL} Low-level output current	$V_{ID} = -0.5\text{ V}$, $V_{OL} = 300\text{ mV}$	25°C		1	1.6		1	1.6		1	1.6	mA
I_{DD} Supply current (four comparators)	$V_{DD} = 0.5\text{ V}$, No load	25°C		130	300		130	300		130	300	μA
		Full range			400			400			400	

† All characteristics are measured with zero common-mode input voltage unless otherwise noted. Full range is -55°C to 125°C for TLC354M, -40°C to 85°C for TLC354I, and 0°C to 70°C for TLC354C. IMPORTANT: See Parameter Measurement Information.

NOTE 4: The offset voltage limits given are the maximum values required to drive the output above 1.25 V or below 150 mV with a 10-kΩ resistor between the output and V_{DD} . They can be verified by applying the limit value to the input and checking for the appropriate output state.

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS [†]		TLC354M			TLC354I			TLC354C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = V_{ICR\text{ min}}$, See Note 4		25°C	2	10	2	10	2	10	mV		
			Full range		12		13		12			
I_{IO} Input offset current			25°C	1		1		1		pA		
			MAX T_A		10		1		0.3	nA		
I_{IB} Input bias current			25°C	5		5		5		pA		
			MAX T_A		20		2		0.6	nA		
V_{ICR} Common-mode input voltage range			25°C	0 to $V_{DD}-1$		0 to $V_{DD}-1$		0 to $V_{DD}-1$		V		
			Full range	0 to $V_{DD}-1.5$		0 to $V_{DD}-1.5$		0 to $V_{DD}-1.5$				
I_{OH} High-level output current	$V_{ID} = 1\text{ V}$	$V_{OH} = 5\text{ V}$	25°C	0.1		0.1		0.1		nA		
		$V_{OH} = 15\text{ V}$	Full range		1		1		1	μA		
V_{OL} Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 4\text{ mA}$	25°C	150	400	150	400	150	400	mV			
		Full range		700		700		700				
I_{OL} Low-level output current	$V_{ID} = -1\text{ V}$, $V_{OL} = 1.5\text{ V}$	25°C	6	16	6	16	6	16	mA			
I_{DD} Supply current (four comparators)	$V_{DD} = 1\text{ V}$, No load		25°C	0.3	0.6	0.3	0.6	0.3	0.6	μA		
			Full range		0.8		0.8		0.8			

[†]All characteristics are measured with zero common-mode input voltage unless otherwise noted. Full range is -55°C to 125°C for TLC354M, -40°C to 85°C for TLC354I, and 0°C to 70°C for TLC354C. IMPORTANT: See Parameter Measurement Information.

NOTE 4: The offset voltage limits given are the maximum values required to drive the output above 4 V or below 400 mV with a 10-k Ω resistor between the output and V_{DD} . They can be verified by applying the limit value to the input and checking for the appropriate output state.

switching characteristics, $V_{DD} = 5\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
Response time	R_L connected to 5 V through 5.1 k Ω , $C_L = 15\text{ pF}$ [‡] , See Note 5	100-mV input step with 5-mV overdrive		650		ns
		TTL-level input step		200		

[‡] C_L includes probe and jig capacitance.

NOTE 5: The response time specified is the interval between the input step function and the instant when the output crosses 1.4 V.

LINCOS™ QUADRUPLE DIFFERENTIAL COMPARATORS
 TLC354M, TLC354I, TLC354C



PARAMETER MEASUREMENT INFORMATION

The digital output stage of the TLC354 can be damaged if it is held in the linear region of the transfer curve. Conventional operational amplifier/comparator testing incorporates the use of a servo-loop that is designed to force the device output to a level within this linear region. Since the servo-loop method of testing cannot be used, the following alternative for measuring parameters such as input offset voltage, common-mode rejection, etc., are offered.

To verify that the input offset voltage falls within the limits specified, the limit value is applied to the input as shown in Figure 1(a). With the noninverting input positive with respect to the inverting input, the output should be high. With the input polarity reversed, the output should be low.

A similar test can be made to verify the input offset voltage at the common-mode extremes. The supply voltages can be slewed as shown in Figure 1(b) for the V_{ICR} test, rather than changing the input voltages, to provide greater accuracy.

A close approximation of the input offset voltage can be obtained by using a binary search method to vary the differential input voltage while monitoring the output state. When the applied input voltage differential is equal, but opposite in polarity, to the input offset voltage, the output will change states.

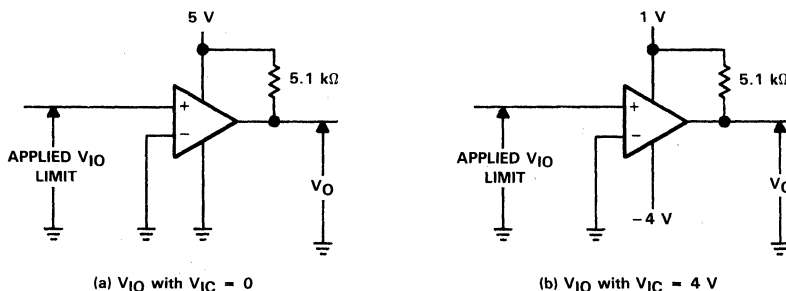


FIGURE 1. METHOD FOR VERIFYING THAT INPUT OFFSET VOLTAGE IS WITHIN SPECIFIED LIMITS

PARAMETER MEASUREMENT INFORMATION

Figure 2 illustrates a practical circuit for direct dc measurement of input offset voltage that does not bias the comparator into the linear region. The circuit consists of a switching-mode servo loop in which U1a generates a triangular waveform of approximately 20-mV amplitude. U1b acts as a buffer, with C2 and R4 removing any residual dc offset. The signal is then applied to the inverting input of the comparator under test, while the noninverting input is driven by the output of the integrator formed by U1c through the voltage divider formed by R9 and R10. The loop reaches a stable operating point when the output of the comparator under test has a duty cycle of exactly 50%, which can only occur when the incoming triangle wave is "sliced" symmetrically or when the voltage at the noninverting input exactly equals the input offset voltage.

Voltage divider R9 and R10 provides a step-up of the input offset voltage by a factor of 100 to make measurement easier. The values of R5, R8, R9, and R10 can significantly influence the accuracy of the reading; therefore, it is suggested that their tolerance level be 1% or lower.

Measuring the extremely low values of input current requires isolation from all other sources of leakage current and compensation for the leakage of the test socket and board. With a good picoammeter, the socket and board leakage can be measured with no device in the socket. Subsequently, this open-socket leakage value can be subtracted from the measurement obtained with a device in the socket to obtain the actual input current of the device.

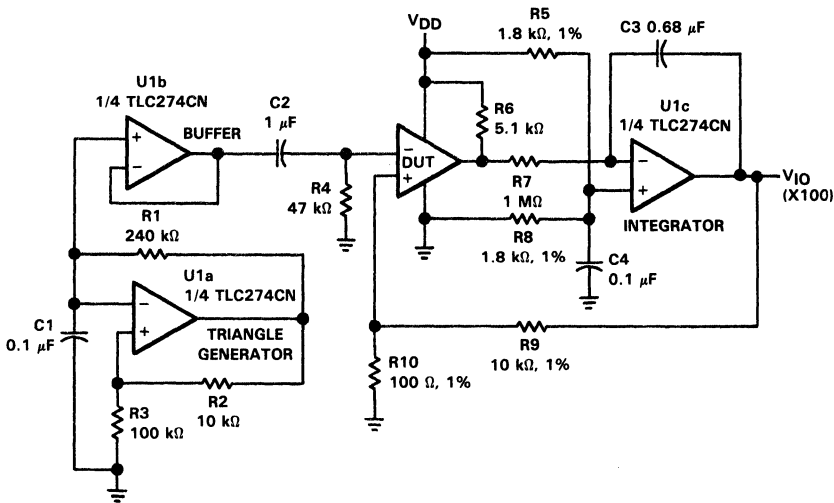
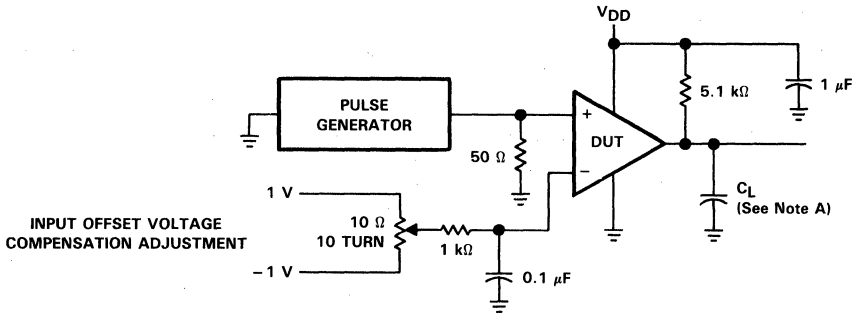


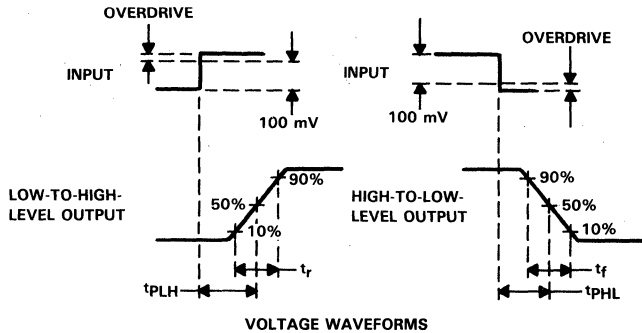
FIGURE 2. CIRCUIT FOR INPUT OFFSET VOLTAGE MEASUREMENT

PARAMETER MEASUREMENT INFORMATION

Response time is defined as the interval between the application of an input step function and the instant when the output reaches 50% of its maximum value. Response time, low-to-high-level output, is measured from the leading edge of the input pulse, while response time, high-to-low-level output, is measured from the trailing edge of the input pulse. Response-time measurement at low input signal levels can be greatly affected by the input offset voltage. The offset voltage should be balanced by the adjustment at the inverting input (as shown in Figure 3) so that the circuit is just at the transition point. Then a low signal, for example 105-mV or 5-mV overdrive, will cause the output to change state.



TEST CIRCUIT



VOLTAGE WAVEFORMS

NOTE A: C_L includes probe and jig capacitance.

FIGURE 3. RESPONSE, RISE, AND FALL TIMES CIRCUIT AND VOLTAGE WAVEFORMS

3

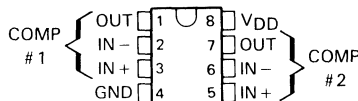
Voltage Comparators

TLC372M, TLC372I, TLC372C LinCMOS™ DUAL DIFFERENTIAL COMPARATORS

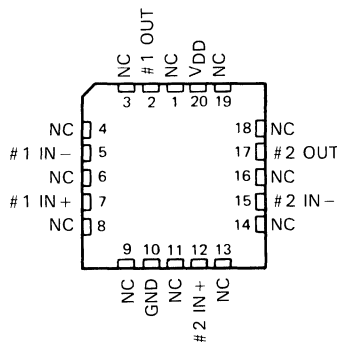
D2821, NOVEMBER 1983—REVISED SEPTEMBER 1988

- Single or Dual-Supply Operation
- Wide Range of Supply Voltages . . . 2 V to 18 V
- Very Low Supply Current Drain 150 μ A Typ at 5 V
- Fast Response Time . . . 200 ns Typ for TTL-Level Input Step
- Built-In ESD Protection
- High Input Impedance . . . $10^{12} \Omega$ Typ
- Extremely Low Input Bias Current 5 pA Typ
- Ultrastable Low Input Offset Voltage
- Input Offset Voltage Change at Worst-Case Input Conditions Typically 0.23 μ V/Month, Including the First 30 Days
- Common-Mode Input Voltage Range Includes Ground
- Outputs Compatible with TTL, MOS, and CMOS
- Pin-Compatible with LM393

TLC372M . . . JG PACKAGE
TLC372I, TLC372C . . . D, JG, OR P PACKAGE
(TOP VIEW)



TLC372M . . . FK PACKAGE
(TOP VIEW)



NC—No internal connection

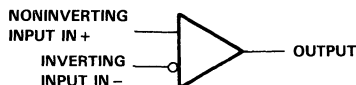
description

This device is fabricated using LinCMOS™ technology and consists of two independent voltage comparators each designed to operate from a single power supply. Operation from dual supplies is also possible so long as the difference between the two supplies is 2 to 18 V. Each device features extremely high input impedance (typically greater than $10^{12} \Omega$) allowing direct interfacing with high-impedance sources. The outputs are n-channel open-drain configurations, and can be connected to achieve positive-logic wired-AND relationships.

The TLC372 has internal electrostatic discharge (ESD) protection circuits and has been classified with a 2000-V ESD rating tested under MIL-STD-883C, Method 3015.1. However, care should be exercised in handling this device as exposure to ESD may result in a degradation of the device parametric performance.

The TLC372M is characterized for operation over the full military temperature range of -55°C to 125°C . The TLC372I is characterized for operation from -40°C to 85°C . The TLC372C is characterized for operation from 0°C to 70°C .

symbol (each comparator)



LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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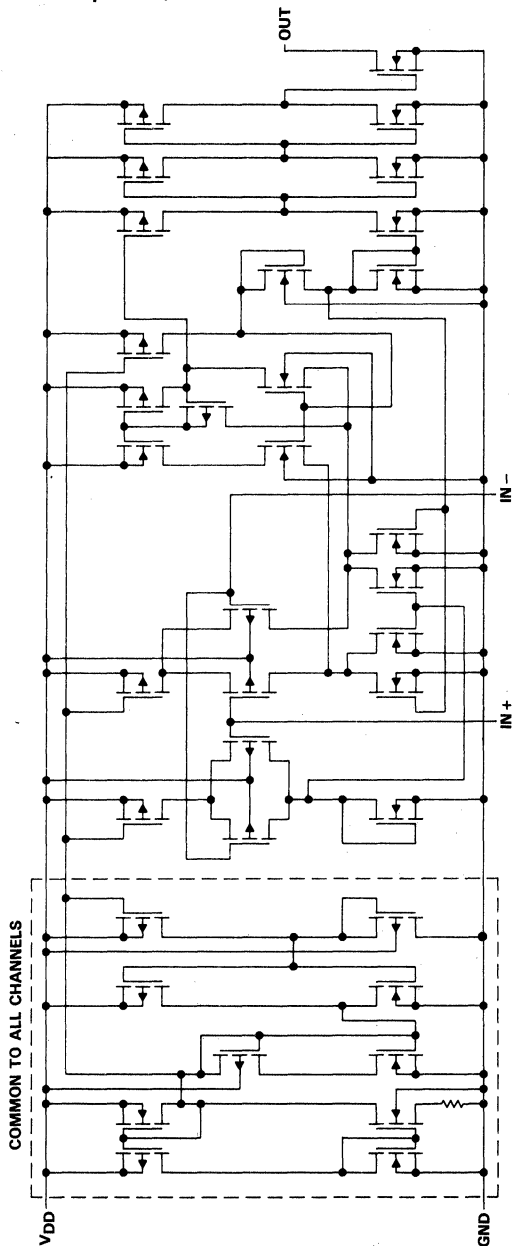
3-119

3

Voltage Comparators

TLC372M, TLC372I, TLC372C LinCMOS™ DUAL DIFFERENTIAL COMPARATORS

equivalent schematic (each comparator)



3

Voltage Comparators

TLC372M, TLC372I, TLC372C LinCMOS™ DUAL DIFFERENTIAL COMPARATORS

AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE			
		SMALL-OUTLINE (D)	CHIP-CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	5 mV	TLC372CD	—	TLC372CJG	TLC372CP
-40°C to 85°C	5 mV	TLC372ID	—	TLC372IJG	TLC372IP
-55°C to 125°C	5 mV	—	TLC372MFK	TLC372MJG	—

D packages are available taped and reeled. Add "R" suffix to device type (e.g., TLC372CDR).

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V _{DD} (see Note 1)	18 V
Differential input voltage, V _{ID} (see Note 2)	±18 V
Input voltage, V _I	-0.3 V to 18 V
Output voltage, V _O	18 V
Input current, I _I	±5 mA
Output current, I _O	20 mA
Duration of output short-circuit to ground (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: TLC372M	-55°C to 125°C
TLC372I	-40°C to 85°C
TLC372C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds, D or P package	260°C

- NOTES: 1. All voltage values except differential voltages are with respect to network ground.
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. Short circuits from outputs to V_{DD} can cause excessive heating and eventual device destruction.

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C	DERATING	DERATE	T _A = 70°C	T _A = 85°C	T _A = 125°C
	POWER RATING	FACTOR	ABOVE T _A	POWER RATING	POWER RATING	POWER RATING
D	500 mW	5.8 mW/°C	64°C	464 mW	377 mW	N/A
FK	500 mW	11.0 mW/°C	104°C	500 mW	500 mW	275 mW
JG (TLC372M)	500 mW	8.4 mW/°C	90°C	500 mW	500 mW	210 mW
JG (TLC372I, TLC372C)	500 mW	6.6 mW/°C	74°C	500 mW	429 mW	N/A
P	500 mW	N/A	N/A	500 mW	500 mW	N/A

3

Voltage Comparators

Voltage Comparators



TLC372M, TLC372I, TLC372C
 LinCMOS™ DUAL DIFFERENTIAL COMPARATORS

3-122

recommended operating conditions

	M-SUFFIX		I-SUFFIX		C-SUFFIX		UNIT	
	MIN	MAX	MIN	MAX	MIN	MAX		
Supply voltage, V_{DD}	4	16	3	16	3	16	V	
Common-mode input voltage, V_{IC}	$V_{DD} = 5\text{ V}$	0	3.5	0	3.5	0	3.5	V
	$V_{DD} = 10\text{ V}$	0	8.5	0	8.5	0	8.5	
Operating free-air temperature, T_A	-55	125	-40	85	0	70	°C	

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†		TLC372M			TLC372I			TLC372C			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX		
V_{IO} Input offset voltage	$V_{IC} = V_{ICR}\text{ min.}$	See Note 4	25°C	1	5		1	5		1	5	mV	
			Full range			10			7				6.5
I_{IO} Input offset current			25°C	1			1			1		pA	
			MAX T_A			10			1			0.3	nA
I_{IB} Input bias current			25°C		5		5			5		pA	
			MAX T_A			20			2			0.6	nA
V_{ICR} Common-mode input voltage range			25°C	0 to $V_{DD}-1$			0 to $V_{DD}-1$			0 to $V_{DD}-1$		V	
			Full range	0 to $V_{DD}-1.5$			0 to $V_{DD}-1.5$			0 to $V_{DD}-1.5$			
I_{OH} High-level output current	$V_{ID} = 1\text{ V}$		$V_{OH} = 5\text{ V}$	25°C	0.1		0.1			0.1		nA	
			$V_{OH} = 15\text{ V}$	Full range			3			1		1	μA
V_{OL} Low-level output voltage	$V_{ID} = -1\text{ V},$	$I_{OL} = 4\text{ mA}$	25°C		150	400		150	400		150	400	mV
			Full range			700			700			700	
I_{OL} Low-level output current	$V_{ID} = -1\text{ V},$	$V_{OL} = 1.5\text{ V}$	25°C	6	16		6	16		6	16	mA	
I_{DD} Supply current (two comparators)	$V_{ID} = 1\text{ V},$	No load	25°C		150	300		150	300		150	300	μA
			Full range			400			400			400	

†All characteristics are measured with zero common-mode input voltage unless otherwise noted. Full range is -55°C to 125°C for TLC372M, 0°C to 70°C for TLC372C, and -40°C to 85°C for TLC372I. IMPORTANT: See Parameter Measurement Information.

NOTE 4: The offset voltage limits given are the maximum values required to drive the output above 4 V or below 400 mV with a 10-kΩ resistor between the output and V_{DD} . They can be verified by applying the limit value to the input and checking for the appropriate output state.

switching characteristics, $V_{DD} = 5\text{ V}, T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
Response time	R_L connected to 5 V through 5.1 kΩ, $C_L = 15\text{ pF}^\ddagger$, See Note 5	100-mV input step with 5-mV overdrive		650		ns
		TTL-level input step		200		

‡ C_L includes probe and jig capacitance.

NOTE 5: The response time specified is the interval between the input step function and the instant when the output crosses 1.4 V.

PARAMETER MEASUREMENT INFORMATION

The digital output stage of the TLC372 can be damaged if it is held in the linear region of the transfer curve. Conventional operational amplifier/comparator testing incorporates the use of a servo-loop that is designed to force the device output to a level within this linear region. Since the servo-loop method of testing cannot be used, the following alternative for measuring parameters such as input offset voltage, common-mode rejection, etc., are offered.

To verify that the input offset voltage falls within the limits specified, the limit value is applied to the input as shown in Figure 1 (a). With the noninverting input positive with respect to the inverting input, the output should be high. With the input polarity reversed, the output should be low.

A similar test can be made to verify the input offset voltage at the common-mode extremes. The supply voltages can be slewed as shown in Figure 1 (b) for the V_{ICR} test, rather than changing the input voltages, to provide greater accuracy.

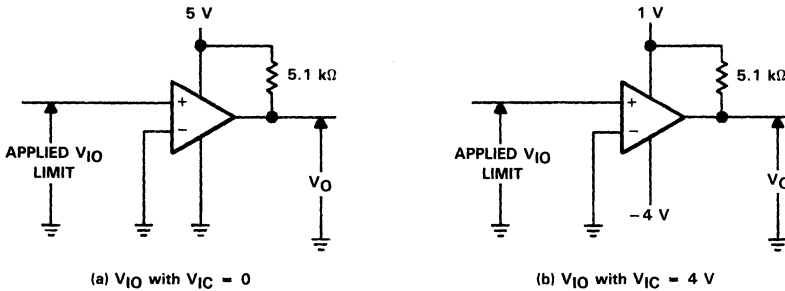


FIGURE 1. METHOD FOR VERIFYING THAT INPUT OFFSET VOLTAGE IS WITHIN SPECIFIED LIMITS

A close approximation of the input offset voltage can be obtained by using a binary search method to vary the differential input voltage while monitoring the output state. When the applied input voltage differential is equal, but opposite in polarity, to the input offset voltage, the output will change states.

3
 Voltage Comparators

PARAMETER MEASUREMENT INFORMATION

Figure 2 illustrates a practical circuit for direct dc measurement of input offset voltage that does not bias the comparator into the linear region. The circuit consists of a switching-mode servo loop in which U1a generates a triangular waveform of approximately 20-mV amplitude. U1b acts as a buffer, with C2 and R4 removing any residual dc offset. The signal is then applied to the inverting input of the comparator under test, while the noninverting input is driven by the output of the integrator formed by U1c through the voltage divider formed by R9 and R10. The loop reaches a stable operating point when the output of the comparator under test has a duty cycle of exactly 50%, which can only occur when the incoming triangle wave is "sliced" symmetrically or when the voltage at the noninverting input exactly equals the input offset voltage.

Voltage divider R9 and R10 provides a step-up of the input offset voltage by a factor of 100 to make measurement easier. The values of R5, R8, R9, and R10 can significantly influence the accuracy of the reading; therefore, it is suggested that their tolerance level be 1% or lower.

Measuring the extremely low values of input current requires isolation from all other sources of leakage current and compensation for the leakage of the test socket and board. With a good picoammeter, the socket and board leakage can be measured with no device in the socket. Subsequently, this open-socket leakage value can be subtracted from the measurement obtained with a device in the socket to obtain the actual input current of the device.

3

Voltage Comparators

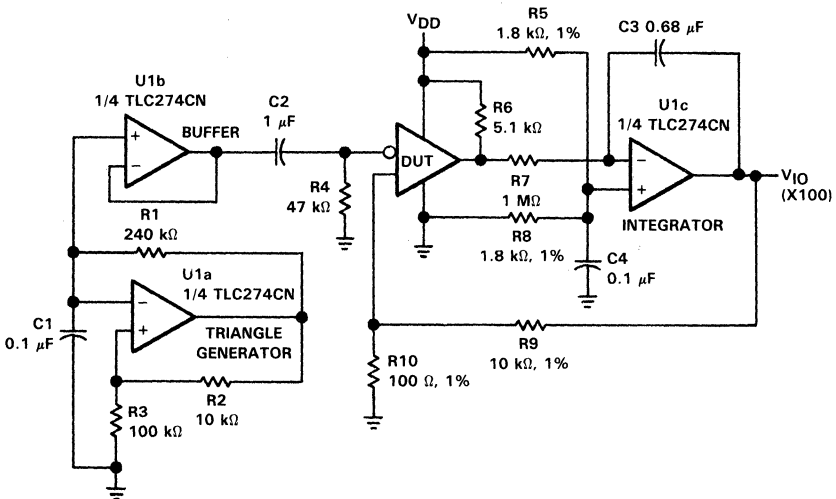
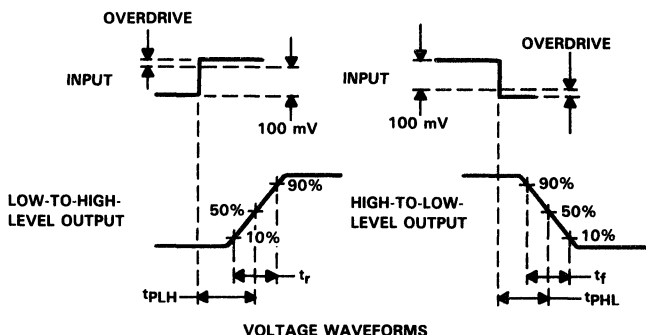
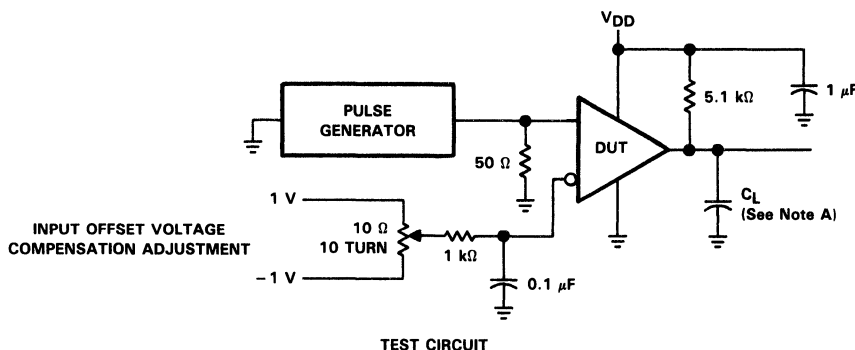


FIGURE 2. CIRCUIT FOR INPUT OFFSET VOLTAGE MEASUREMENT

PARAMETER MEASUREMENT INFORMATION

Response time is defined as the interval between the application of an input step function and the instant when the output reaches 50% of its maximum value. Response time, low-to-high-level output, is measured from the leading edge of the input pulse, while response time, high-to-low-level output, is measured from the trailing edge of the input pulse. Response-time measurement at low input signal levels can be greatly affected by the input offset voltage. The offset voltage should be balanced by the adjustment at the inverting input (as shown in Figure 3) so that the circuit is just at the transition point. Then a low signal, for example 105-mV or 5-mV overdrive, will cause the output to change state.



NOTE A: C_L includes probe and jig capacitance.

FIGURE 3. RESPONSE, RISE, AND FALL TIMES CIRCUIT AND VOLTAGE WAVEFORMS

TLC374M, TLC374I, TLC374C LinCMOS™ QUADRUPLE DIFFERENTIAL COMPARATORS

D2783, NOVEMBER 1983—REVISED SEPTEMBER 1988

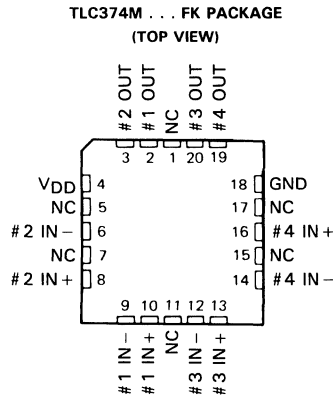
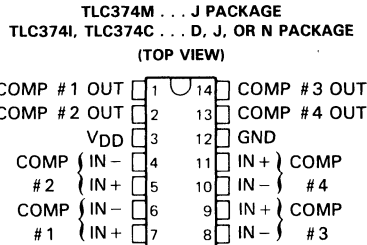
- Single or Dual-Supply Operation
- Wide Range of Supply Voltages . . . 2 V to 18 V
- Very Low Supply Current Drain 0.3 mA Typ at 5 V
- Fast Response Time . . . 200 ns Typ for TTL-Level Input Step
- Built-In ESD Protection
- High Input Impedance . . . 10^{12} Typ
- Extremely Low Input Bias Current 5 pA Typ
- Ultrastable Low Input Offset Voltage
- Input Offset Voltage Change at Worst-Case Input Conditions Typically $0.23 \mu\text{V}/\text{Month}$, Including the First 30 Days
- Common-Mode Input Voltage Range Includes Ground
- Outputs Compatible with TTL, MOS, and CMOS
- Pin-Compatible with LM339

description

This device is fabricated using LinCMOS™ technology and consists of four independent voltage comparators designed to operate from a single power supply. Operation from dual supplies is also possible so long as the difference between the two supplies is 2 to 18 V. Each device features extremely high input impedance (typically greater than $10^{12} \Omega$) allowing direct interfacing with high-impedance sources. The outputs are n-channel open-drain configurations, and can be connected to achieve positive-logic wired-AND relationships.

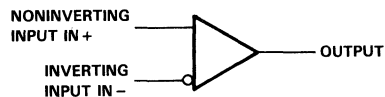
The TLC374 has internal electrostatic discharge (ESD) protection circuits and has been classified with a 2000-V ESD rating tested under MIL-STD-883C, Method 3015.1. However, care should be exercised in handling this device as exposure to ESD may result in a degradation of the device parametric performance.

The TLC374M is characterized for operation over the full military temperature range of -55°C to 125°C . The TLC374I is characterized for operation from -40°C to 85°C . The TLC374C is characterized for operation from 0°C to 70°C .



NC—No internal connection

symbol (each comparator)



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PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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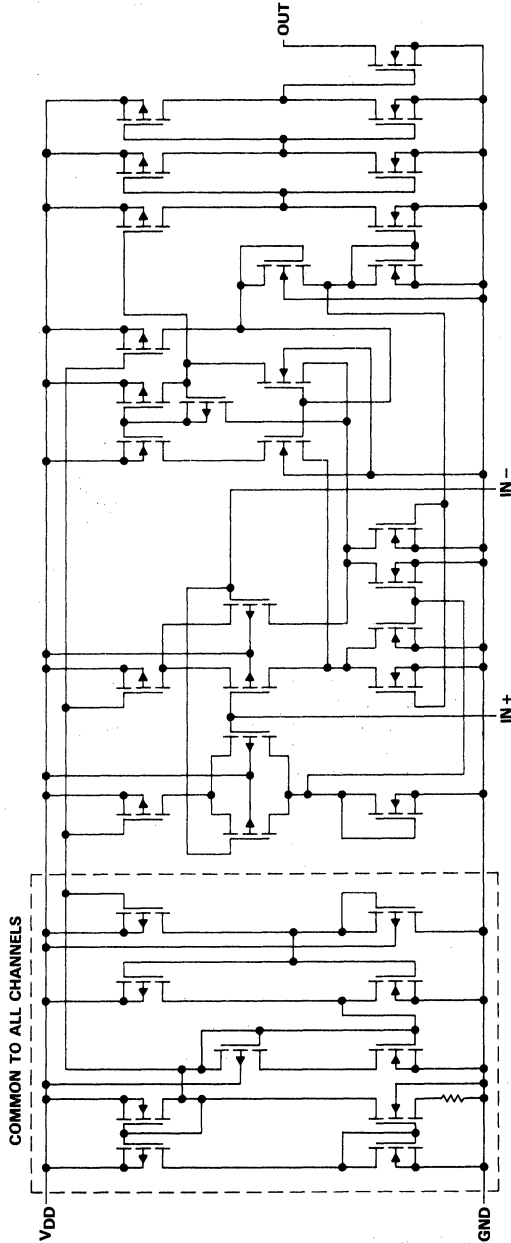
3-127

3

Voltage Comparators

TLC374M, TLC374I, TLC374C
LinCMOS™ QUADRUPLE DIFFERENTIAL COMPARATORS

equivalent schematic (each comparator)



3

Voltage Comparators

TLC374M, TLC374I, TLC374C LinCMOS™ QUADRUPLE DIFFERENTIAL COMPARATORS

AVAILABLE OPTIONS

T _A	V _{IO} MAX AT 25°C	PACKAGE			
		SMALL-OUTLINE (D)	CHIP-CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	5 mV	TLC374CD	—	TLC374CJ	TLC374CN
-40°C to 85°C	5 mV	TLC374ID	—	TLC374IJ	TLC374IN
-55°C to 125°C	5 mV	—	TLC374MFK	TLC374MJ	—

D packages are available taped and reeled. Add "R" suffix to device type (e.g., TLC374CDR).

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V _{DD} (see Note 1)	18 V
Differential input voltage, V _{ID} (see Note 2)	±18 V
Input voltage, V _I	V _{DD}
Input voltage range	-0.3 V to 18 V
Output voltage, V _O	18 V
Input current, I _I	±5 mA
Output current, I _O	20 mA
Duration of output short-circuit to ground (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: TLC374M	-55°C to 125°C
TLC374I	-40°C to 85°C
TLC374C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds, D or N package	260°C

- NOTES: 1. All voltage values except differential voltages are with respect to network ground.
 2. Differential voltages are at the noninverting input terminal with respect to the inverting input terminal.
 3. Short circuits from outputs to V_{DD} can cause excessive heating and eventual device destruction.

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C	DERATING	DERATE	T _A = 70°C	T _A = 85°C	T _A = 125°C
	POWER RATING	FACTOR	ABOVE T _A	POWER RATING	POWER RATING	POWER RATING
D	500 mW	7.6 mW/°C	84°C	494 mW	377 mW	N/A
FK	500 mW	11.0 mW/°C	104°C	500 mW	500 mW	275 mW
J (TLC374M)	500 mW	11.0 mW/°C	104°C	500 mW	500 mW	275 mW
J (TLC374I, TLC374C)	500 mW	N/A	N/A	500 mW	500 mW	N/A
N	500 mW	N/A	N/A	500 mW	500 mW	N/A

3

Voltage Comparators

recommended operating conditions

		M-SUFFIX			I-SUFFIX			C-SUFFIX			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, V_{DD}		4		16	3		16	3		16	V
Common-mode input voltage, V_{IC}	$V_{DD} = 5\text{ V}$	0		3.5	0		3.5	0		3.5	V
	$V_{DD} = 10\text{ V}$	0		8.5	0		8.5	0		8.5	
Operating free-air temperature, T_A		-55		125	-40		85	0		70	°C

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†		TLC374M			TLC374I			TLC374C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = V_{ICR}$ min. See Note 4	25°C	1 5			1 5			1 5			mV
			Full range			7			6.5			
I_{IO} Input offset current		25°C	1			1			1			pA
		MAX T_A	10			1			0.3			nA
I_{IB} Input bias current		25°C	5			5			5			pA
		MAX T_A	20			2			0.6			nA
V_{ICR} Common-mode input voltage range		25°C	0 to $V_{DD}-1$			0 to $V_{DD}-1$			0 to $V_{DD}-1$			V
		Full range	0 to $V_{DD}-1.5$			0 to $V_{DD}-1.5$			0 to $V_{DD}-1.5$			
I_{OH} High-level output current	$V_{ID} = 1\text{ V}$	$V_{OH} = 5\text{ V}$	25°C			0.1			0.1			nA
		$V_{OH} = 15\text{ V}$	Full range			1			1			μA
V_{OL} Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 4\text{ mA}$	25°C	150 400			150 400			150 400			mV
		Full range	700			700			700			
I_{OL} Low-level output current	$V_{ID} = -1\text{ V}$, $V_{OL} = 1.5\text{ V}$	25°C	6 16			6 16			6 16			mA
I_{DD} Supply current (four comparators)	$V_{ID} = 1\text{ V}$, No load	25°C	300 600			300 600			300 600			μA
		Full range	800			800			800			

† All characteristics are measured with zero common-mode input voltage unless otherwise noted. Full range is -55°C to 125°C for TLC374M, 0°C to 70°C for TLC374C, and -40°C to 85°C for TLC374I. IMPORTANT: See Parameter Measurement Information.

NOTE 4: The offset voltage limits given are the maximum values required to drive the output above 4 V or below 400 mV with a 10-kΩ resistor between the output and V_{DD} . They can be verified by applying the limit value to the input and checking for the appropriate output state.

switching characteristics, $V_{DD} = 5\text{ V}$, $T_A = 25\text{ °C}$

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
Response time	R_L connected to 5 V through 5.1 kΩ, $C_L = 15\text{ pF}$ ‡, See Note 5	100-mV input step with 5-mV overdrive	650			ns
		TTL-level input step	200			

‡ C_L includes probe and jig capacitance.

NOTE 5: The response time specified is the interval between the input step function and the instant when the output crosses 1.4 V.

PARAMETER MEASUREMENT INFORMATION

The digital output stage of the TLC374 can be damaged if it is held in the linear region of the transfer curve. Conventional operational amplifier/comparator testing incorporates the use of a servo-loop that is designed to force the device output to a level within this linear region. Since the servo-loop method of testing cannot be used, the following alternative for measuring parameters such as input offset voltage, common-mode rejection, etc., are offered.

To verify that the input offset voltage falls within the limits specified, the limit value is applied to the input as shown in Figure 1 (a). With the noninverting input positive with respect to the inverting input, the output should be high. With the input polarity reversed, the output should be low.

A similar test can be made to verify the input offset voltage at the common-mode extremes. The supply voltages can be slewed as shown in Figure 1(b) for the V_{ICR} test, rather than changing the input voltages, to provide greater accuracy.

A close approximation of the input offset voltage can be obtained by using a binary search method to vary the differential input voltage while monitoring the output state. When the applied input voltage differential is equal, but opposite in polarity, to the input offset voltage, the output will change states.

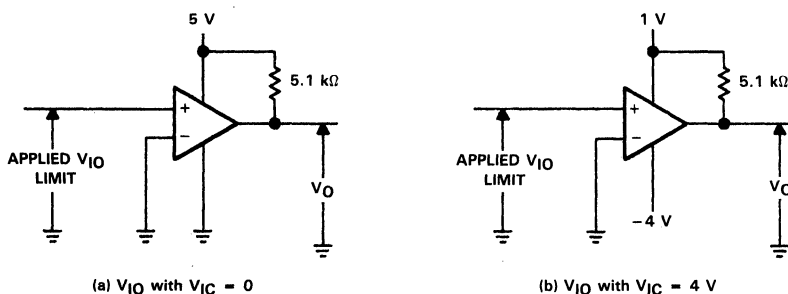


FIGURE 1. METHOD FOR VERIFYING THAT INPUT OFFSET VOLTAGE IS WITHIN SPECIFIED LIMITS

PARAMETER MEASUREMENT INFORMATION

Figure 2 illustrates a practical circuit for direct dc measurement of input offset voltage that does not bias the comparator into the linear region. The circuit consists of a switching-mode servo loop in which U1a generates a triangular waveform of approximately 20-mV amplitude. U1b acts as a buffer, with C2 and R4 removing any residual dc offset. The signal is then applied to the inverting input of the comparator under test, while the noninverting input is driven by the output of the integrator formed by U1c through the voltage divider formed by R9 and R10. The loop reaches a stable operating point when the output of the comparator under test has a duty cycle of exactly 50%, which can only occur when the incoming triangle wave is "sliced" symmetrically or when the voltage at the noninverting input exactly equals the input offset voltage.

Voltage divider R9 and R10 provides a step-up of the input offset voltage by a factor of 100 to make measurement easier. The values of R5, R8, R9, and R10 can significantly influence the accuracy of the reading; therefore, it is suggested that their tolerance level be 1% or lower.

Measuring the extremely low values of input current requires isolation from all other sources of leakage current and compensation for the leakage of the test socket and board. With a good picoammeter, the socket and board leakage can be measured with no device in the socket. Subsequently, this open-socket leakage value can be subtracted from the measurement obtained with a device in the socket to obtain the actual input current of the device.

3

Voltage Comparators

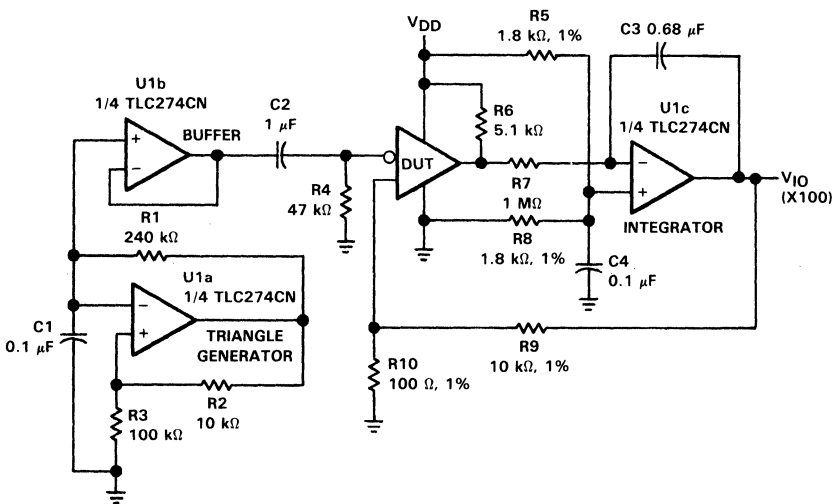
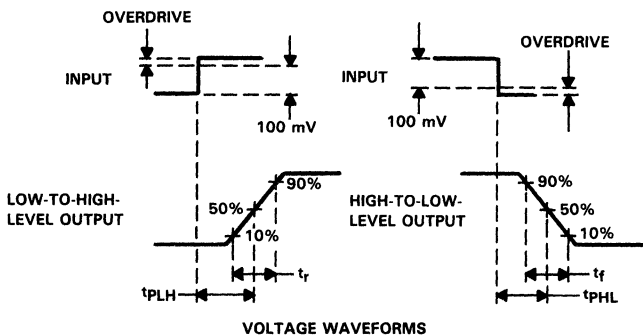
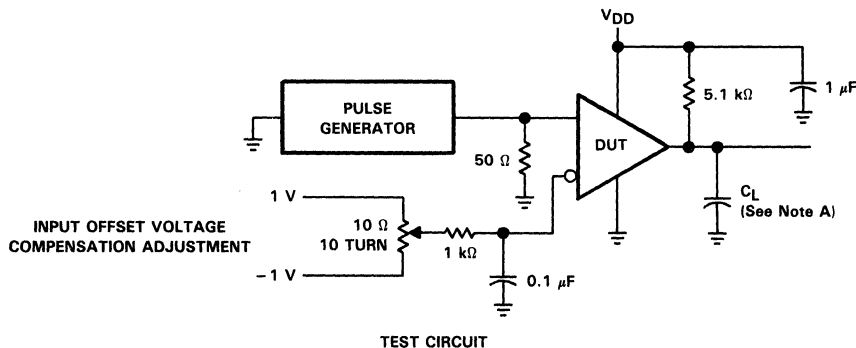


FIGURE 2. CIRCUIT FOR INPUT OFFSET VOLTAGE MEASUREMENT

PARAMETER MEASUREMENT INFORMATION

Response time is defined as the interval between the application of an input step function and the instant when the output reaches 50% of its maximum value. Response time, low-to-high-level output, is measured from the leading edge of the input pulse, while response time, high-to-low-level output, is measured from the trailing edge of the input pulse. Response-time measurement at low input signal levels can be greatly affected by the input offset voltage. The offset voltage should be balanced by the adjustment at the inverting input (as shown in Figure 3) so that the circuit is just at the transition point. Then a low signal, for example 105-mV or 5-mV overdrive, will cause the output to change state.



NOTE A: C_L includes probe and jig capacitance.

FIGURE 3. RESPONSE, RISE, AND FALL TIMES CIRCUIT AND VOLTAGE WAVEFORMS

3

Voltage Comparators

TLC393M, TLC393I, TLC393C DUAL MICROPOWER LinCMOS™ COMPARATORS

D3241, DECEMBER 1986—REVISED FEBRUARY 1989

- Very Low Power . . . 100 μ W Typ at 5 V
- Fast Response Time . . . 2.5 μ s Typ with 5 mV Overdrive
- Single Supply Operation:
TLC393M . . . 4 V to 16 V
TLC393I . . . 3 V to 16 V
TLC393C . . . 3 V to 16 V
- High Input Impedance . . . 1012 Ω Typ
- Input Offset Voltage Change at Worst Case Input Condition Typically 0.23 μ V/Month Including the First 30 Days
- On-Chip ESD Protection

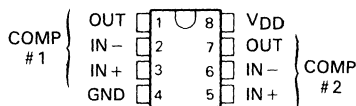
description

The TLC393 consists of two independent differential-voltage comparators designed to operate from a single supply. It is functionally similar to the LM393 but uses 1/20th the power for similar response times. The open-drain MOS output stage will interface to a variety of loads and supplies, as well as "wired" logic functions. For a similar device with a push-pull output configuration, see the TLC3702 data sheet.

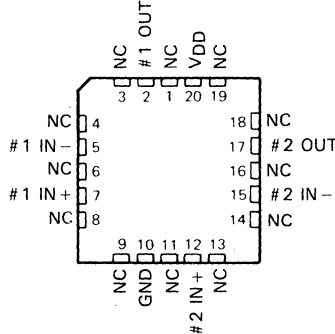
Texas Instruments LinCMOS™ process offers superior analog performance to standard CMOS processes. Along with the standard CMOS advantages of low power without sacrificing speed, high input impedance, and low bias currents, the LinCMOS™ process offers extremely stable input offset voltages, even with differential input stresses of several volts. This characteristic makes it possible to build reliable CMOS comparators.

The TLC393M is characterized for operation over the full military temperature range of -55°C to 125°C. The TLC393I is characterized for operation over the extended industrial temperature range of -40°C to 85°C. The TLC393C is characterized for operation over the commercial temperature range of 0°C to 70°C.

TLC393M . . . JG PACKAGE
TLC393I, TLC393C . . . D, JG, OR P PACKAGE
(TOP VIEW)

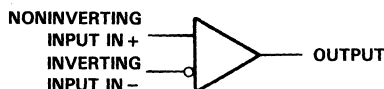


TLC393M . . . FK PACKAGE
(TOP VIEW)



NC—No internal connection

symbol (each comparator)



AVAILABLE OPTIONS

T _A	V _{IO} max at 25°C	PACKAGE			
		SMALL OUTLINE (D)	CERAMIC (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	5 mV	TLC393CD	—	TLC393CJG	TLC393CP
-40°C to 85°C	5 mV	TLC393ID	—	TLC393IJG	TLC393IP
-55°C to 125°C	5 mV	—	TLC393MFK	TLC393MJG	—

The D package is available taped and reeled. Add the suffix R to the device type. (e.g., TLC393CDR)

3

Voltage Comparators

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TEXAS
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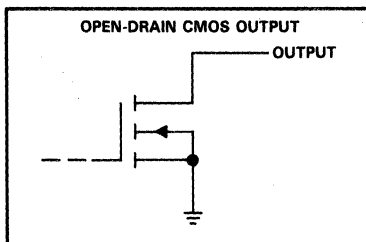
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TLC393M, TLC393I, TLC393C DUAL MICROPOWER LinCMOS™ COMPARATORS

schematic



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{DD} (see Note 1)	-0.3 to 18 V
Differential input voltage (see Note 2)	± 18 V
Input voltage, V_I	-0.3 V to V_{DD}
Output voltage, V_O	-0.3 V to V_{DD}
Input current, I_I	± 5 mA
Output current, I_O (each output)	20 mA
Total supply current into V_{DD} terminal	40 mA
Total current out of ground terminal	40 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: TLC393M	-55°C to 125°C
TLC393I	-40°C to 85°C
TLC393C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.
2. Differential voltages are at the noninverting input with respect to the inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING		POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	—
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG (TLC393M)	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
JG (All others)	825 mW	6.8 mW/°C	528 mW	429 mW	—
P	1000 mW	8.0 mW/°C	640 mW	520 mW	—

3

Voltage Comparators

TLC393M DUAL MICROPOWER LinCMOS™ COMPARATORS

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	4	5	16	V
Common-mode input voltage, V_{IC}	0	$V_{DD} - 1.5$		V
Low-level output current, I_{OL}	20			mA
Operating free-air temperature, T_A	-55	125		°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	MIN	TYP	MAX	UNIT
V_{IO} Input offset voltage	$V_{IC} = V_{ICRmin}$, $V_{DD} = 5\text{ V to } 10\text{ V}$, See Note 3	25°C	1.4	5	mV
		-55°C to 125°C		10	
I_{IO} Input offset current	$V_{IC} = 2.5\text{ V}$	25°C	1		pA
		125°C		15	nA
I_{IB} Input bias current	$V_{IC} = 2.5\text{ V}$	25°C	5		pA
		125°C		30	nA
V_{ICR} Common-mode input voltage range		25°C	0 to $V_{DD} - 1$		V
		-55°C to 125°C	0 to $V_{DD} - 1.5$		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84		dB
		125°C	84		
		-55°C	84		
k_{SVR} Supply voltage rejection ratio	$V_{DD} = 5\text{ V to } 10\text{ V}$	25°C	85		dB
		125°C	84		
		-55°C	84		
V_{OL} Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 6\text{ mA}$	25°C	300	400	mV
		125°C	800		
I_{OH} High-level output current	$V_{ID} = 1\text{ V}$, $V_O = 5\text{ V}$	25°C	0.8	40	nA
		125°C	1		μA
I_{DD} Supply current (both comparators)	No load, Outputs low	25°C	22	40	μA
		-55°C to 125°C	90		

†All characteristics are measured with zero common-mode voltage unless otherwise noted.

NOTE 3: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V with a 2.5-kΩ load to V_{DD} .

3

Voltage Comparators

TLC3931
DUAL MICROPOWER LinCMOS™ COMPARATORS

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	3	5	16	V
Common-mode input voltage, V_{IC}	-0.2	$V_{DD} - 1.5$		V
Low-level output current, I_{OL}				20 mA
Operating free-air temperature, T_A	-40	85		°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	MIN	TYP	MAX	UNIT
V_{IO} Input offset voltage	$V_{IC} = V_{ICRmin}$, $V_{DD} = 5$ V to 10 V, See Note 4	25°C	1.4	5	mV
		-40°C to 85°C	7		
I_{IO} Input offset current	$V_{IC} = 2.5$ V	25°C	1		pA
		85°C	1		nA
I_{IB} Input bias current	$V_{IC} = 2.5$ V	25°C	5		pA
		85°C	2		nA
V_{ICR} Common-mode input voltage range		25°C	0 to $V_{DD} - 1$		V
		-40°C to 85°C	0 to $V_{DD} - 1.5$		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84		dB
		85°C	84		
		-40°C	84		
kSVR Supply voltage rejection ratio	$V_{DD} = 5$ V to 10 V	25°C	85		dB
		85°C	85		
		-40°C	84		
V_{OL} Low-level output voltage	$V_{ID} = -1$ V, $I_{OL} = 6$ mA	25°C	300	400	mV
		85°C	700		
I_{OH} High-level output current	$V_{ID} = 1$ V, $V_O = 5$ V	25°C	0.8	40	nA
		85°C	1		μA
I_{DD} Supply current (both comparators)	No load, Outputs low	25°C	22	40	μA
		-40°C to 85°C	65		

† All characteristics are measured with zero common-mode voltage unless otherwise noted.

NOTE 4: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.



Voltage Comparators

TLC393C DUAL MICROPOWER LinCMOS™ COMPARATORS

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	3	5	16	V
Common-mode input voltage, V_{IC}	-0.2	$V_{DD}-1.5$		V
Low-level output current, I_{OL}				20 mA
Operating free-air temperature, T_A	0			70 °C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	MIN	TYP	MAX	UNIT
V_{IO} Input offset voltage	$V_{IC} = V_{ICRmin}$, $V_{DD} = 5\text{ V to }10\text{ V}$, See Note 4	25°C	1.4	5	mV
		0°C to 70°C	6.5		
I_{IO} Input offset current	$V_{IC} = 2.5\text{ V}$	25°C	1		pA
		70°C	0.3		nA
I_{IB} Input bias current	$V_{IC} = 2.5\text{ V}$	25°C	5		pA
		70°C	0.6		nA
V_{ICR} Common-mode input voltage range		25°C	0 to $V_{DD}-1$		V
		0°C to 70°C	0 to $V_{DD}-1.5$		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84		dB
		70°C	84		
		0°C	84		
k_{SVR} Supply voltage rejection ratio	$V_{DD} = 5\text{ V to }10\text{ V}$	25°C	85		dB
		70°C	85		
		0°C	85		
V_{OL} Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 6\text{ mA}$	25°C	300	400	mV
		70°C	650		
I_{OH} High-level output current	$V_{ID} = 1\text{ V}$, $V_O = 5\text{ V}$	25°C	0.8	40	nA
		70°C	1		μA
I_{DD} Supply current (both comparators)	No load, Outputs low	25°C	22	40	μA
		0°C to 70°C	50		

†All characteristics are measured with zero common-mode voltage unless otherwise noted.

NOTE 4: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.

3

Voltage Comparators

TLC393M, TLC393I, TLC393C
DUAL MICROPOWER LinCMOS™ COMPARATORS

switching characteristics, $V_{DD} = 5\text{ V}$, $T_A = 25^\circ\text{C}$ (see Figure 3)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{PLH} Propagation delay time, low-to-high level output	f = 10 kHz, C _L = 15 pF	Overdrive = 2 mV	4.5		μs
		Overdrive = 5 mV	2.5		
		Overdrive = 10 mV	1.7		
		Overdrive = 20 mV	1.2		
		Overdrive = 40 mV	1.1		
	V _I = 1.4 V step at IN+ pin		1.1		
t _{PHL} Propagation delay time, high-to-low level output	f = 10 kHz, C _L = 15 pF	Overdrive = 2 mV	3.6		μs
		Overdrive = 5 mV	2.1		
		Overdrive = 10 mV	1.3		
		Overdrive = 20 mV	0.85		
		Overdrive = 40 mV	0.55		
	V _I = 1.4 V step at IN+ pin		0.10		
t _{THL} Transition time, high-to-low level output	f = 10 kHz, C _L = 15 pF		20		ns

3

Voltage Comparators

PARAMETER MEASUREMENT INFORMATION

The TLC393 contains a digital output stage which, if held in the linear region of the transfer curve, can cause damage to the device. Conventional operational amplifier/comparator testing incorporates the use of a servo-loop that is designed to force the device output to a level within this linear region. Since the servo-loop method of testing cannot be used, the following alternatives for testing parameters such as input offset voltage, common-mode rejection, etc., are suggested.

To verify that the input offset voltage falls within the limits specified, the limit value is applied to the input as shown in Figure 1(a). With the noninverting input positive with respect to the inverting input, the output should be high. With the input polarity reversed, the output should be low.

A similar test can be made to verify the input offset voltage at the common-mode extremes. The supply voltages can be slewed as shown in Figure 1(b) for the V_{ICR} test, rather than changing the input voltages, to provide greater accuracy.

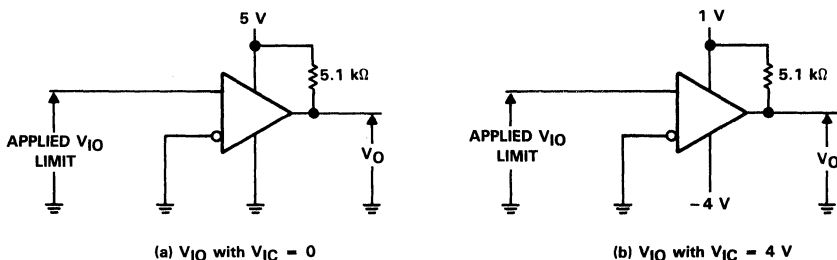


FIGURE 1. METHOD FOR VERIFYING THAT INPUT OFFSET VOLTAGE IS WITHIN SPECIFIED LIMITS

A close approximation of the input offset voltage can be obtained by using a binary search method to vary the differential input voltage while monitoring the output state. When the applied input voltage differential is equal, but opposite in polarity, to the input offset voltage, the output will change states.

**TLC393M, TLC393I, TLC393C
DUAL MICROPOWER LinCMOS™ COMPARATORS**

PARAMETER MEASUREMENT INFORMATION

Figure 2 illustrates a practical circuit for direct dc measurement of input offset voltage that does not bias the comparator into the linear region. The circuit consists of a switching-mode servo-loop in which U1A generates a triangular waveform of approximately 20-mV amplitude. U1B acts as a buffer, with C2 and R4 removing any residual dc offset. The signal is then applied to the inverting input of the comparator under test, while the noninverting input is driven by the output of the integrator formed by U1C through the voltage divider formed by R9 and R10. The loop reaches a stable operating point when the output of the comparator under test has a duty cycle of exactly 50%, which can only occur when the incoming triangle wave is "sliced" symmetrically or when the voltage at the noninverting input exactly equals the input offset voltage.

Voltage divider R9 and R10 provides a step-up of the input offset voltage by a factor of 100 to make measurement easier. The values of R5, R8, R9, and R10 can significantly influence the accuracy of the reading; therefore, it is suggested that their tolerance level be 1% or lower.

Measuring the extremely low values of input current requires isolation from all other sources of leakage current and compensation for the leakage of the test socket and board. With a good picoammeter, the socket and board leakage can be measured with no device in the socket. Subsequently, this open socket leakage value can be subtracted from the measurement obtained with a device in the socket to obtain the actual input current of the device.

3

Voltage Comparators

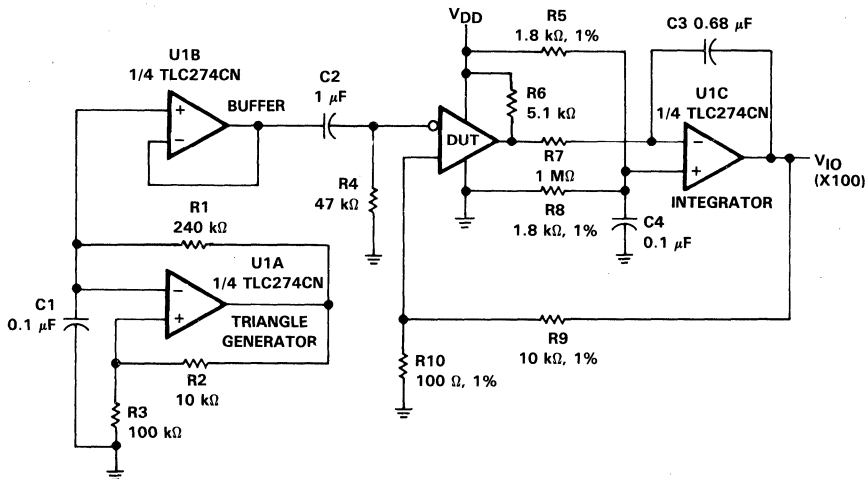
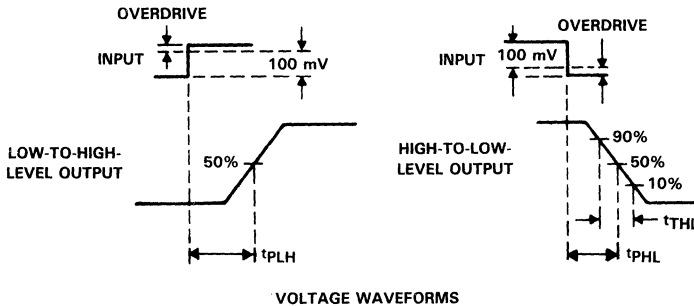
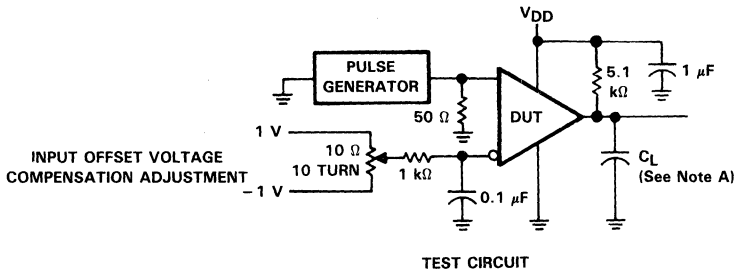


FIGURE 2. CIRCUIT FOR INPUT OFFSET VOLTAGE MEASUREMENT

PARAMETER MEASUREMENT INFORMATION

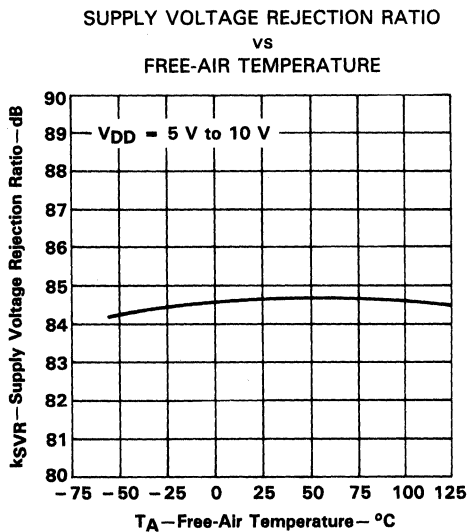
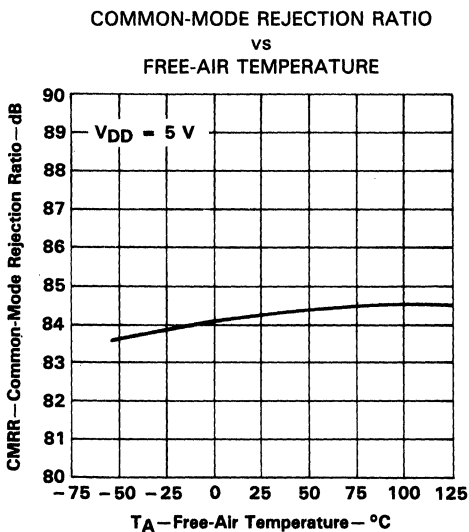
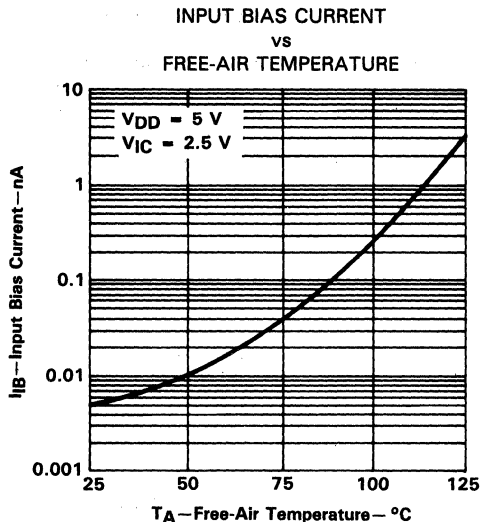
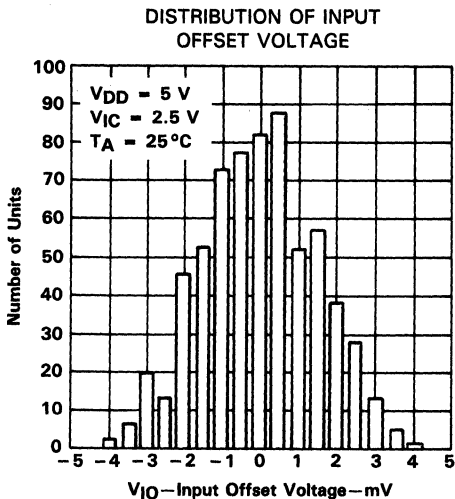
Propagation delay time is defined as the interval between the application of an input step function and the instant when the output reaches 50% of its maximum value. Propagation delay time, low-to-high-level output is measured from the leading edge of the input pulse, while propagation delay time, high-to-low-level output, is measured from the trailing edge of the input pulse. Propagation delay time measurement at low input signal levels can be greatly affected by the input offset voltage. The offset voltage should be balanced by the adjustment at the inverting input (as shown in Figure 3) so that the circuit is just at the transition point. Then a low signal, for example 105 mV or 5 mV overdrive, will cause the output to change state.



NOTE A: C_L includes probe and jig capacitance.

FIGURE 3. PROPAGATION DELAY, RISE, AND FALL TIMES
 CIRCUIT AND VOLTAGE WAVEFORMS

TYPICAL CHARACTERISTICS†



†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



Voltage Comparators

TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT CURRENT
 vs
 HIGH-LEVEL OUTPUT VOLTAGE

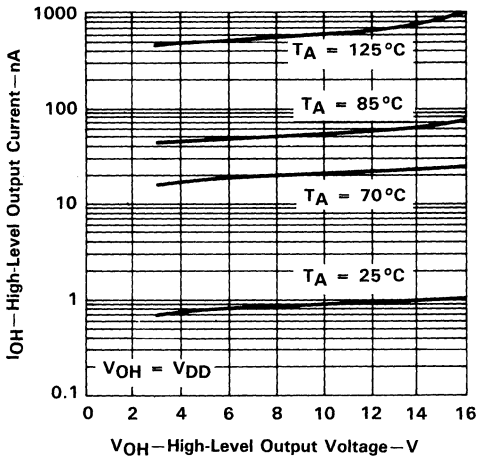


FIGURE 8

HIGH-LEVEL OUTPUT CURRENT
 vs
 FREE-AIR TEMPERATURE

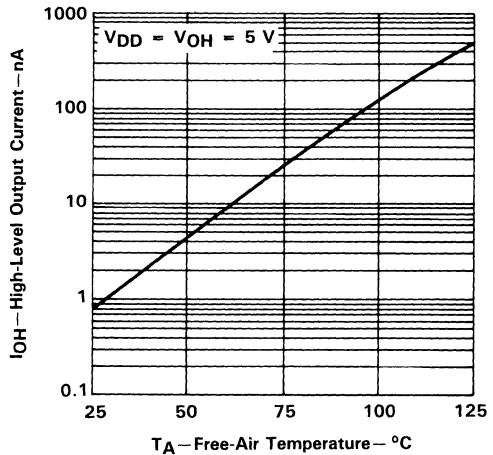


FIGURE 9

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

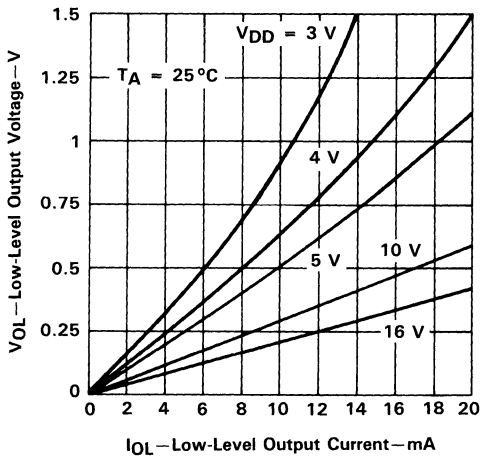


FIGURE 10

LOW-LEVEL OUTPUT VOLTAGE
 vs
 FREE-AIR TEMPERATURE

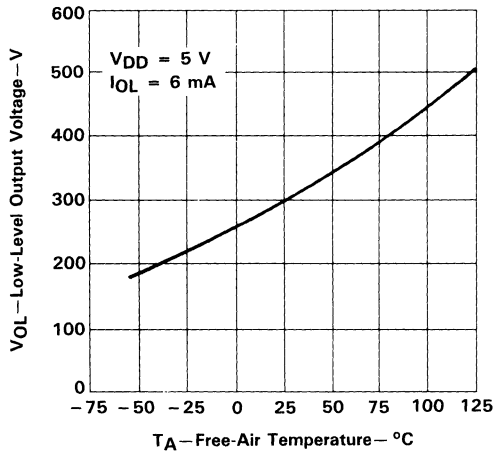


FIGURE 11

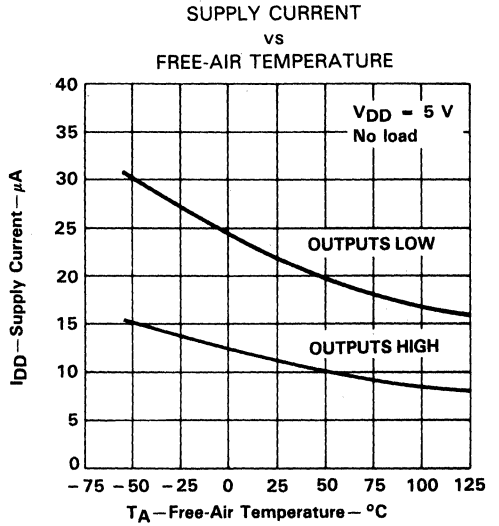
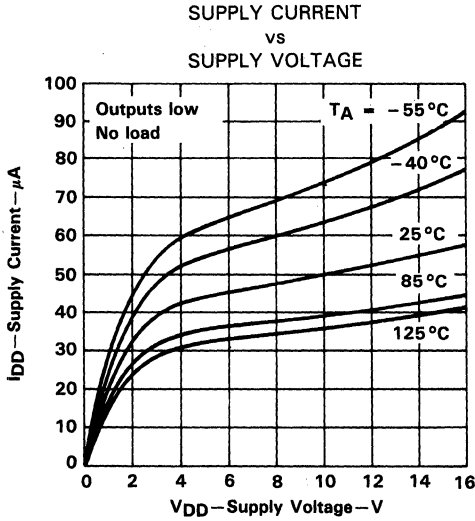
3

Voltage Comparators

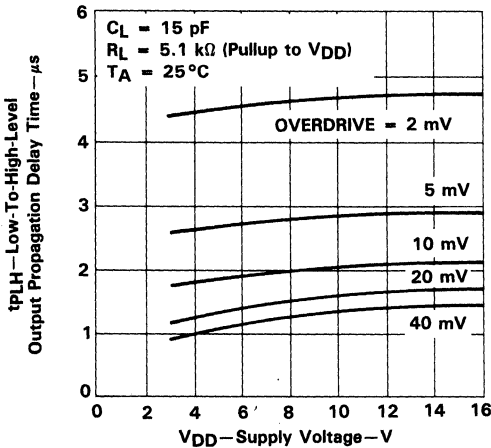
†Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC393M, TLC393I, TLC393C
DUAL MICROPOWER LinCMOS™ COMPARATORS**

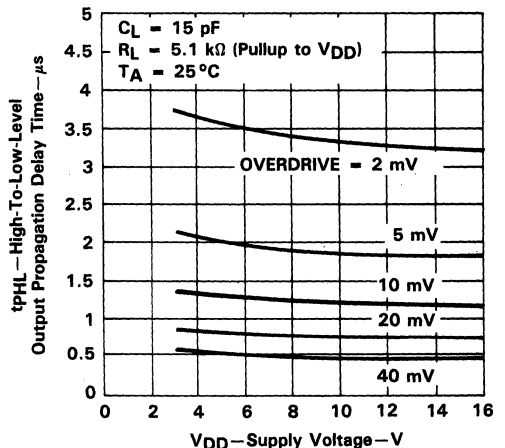
TYPICAL CHARACTERISTICS†



LOW-TO-HIGH-LEVEL
OUTPUT RESPONSE TIME
vs
SUPPLY VOLTAGE



HIGH-TO-LOW-LEVEL
OUTPUT RESPONSE TIME
vs
SUPPLY VOLTAGE



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



Voltage Comparators

TLC393M, TLC393I, TLC393C
DUAL MICROPOWER LinCMOS™ COMPARATORS

TYPICAL CHARACTERISTICS

LOW-TO-HIGH-LEVEL OUTPUT PROPAGATION DELAY FOR VARIOUS OVERDRIVE VOLTAGES

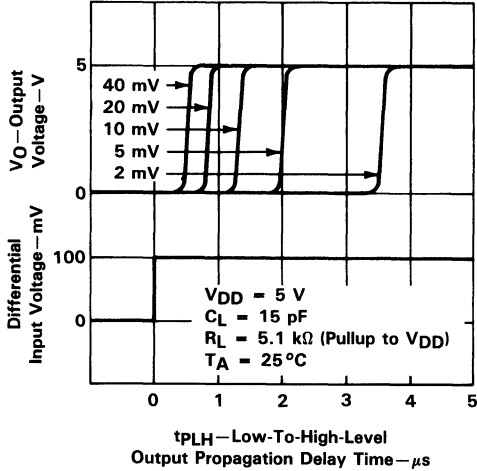


FIGURE 16

OUTPUT FALL TIME VS SUPPLY VOLTAGE

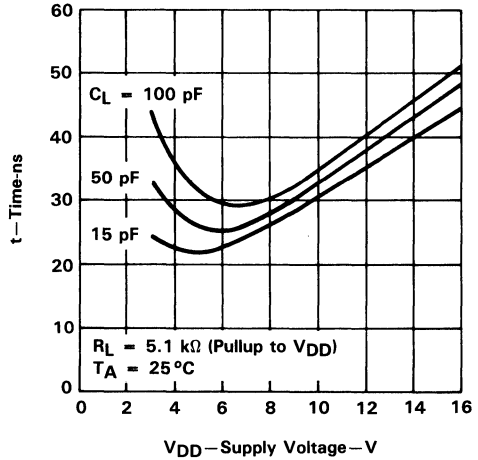


FIGURE 17

HIGH-TO-LOW-LEVEL OUTPUT PROPAGATION DELAY FOR VARIOUS OVERDRIVE VOLTAGES

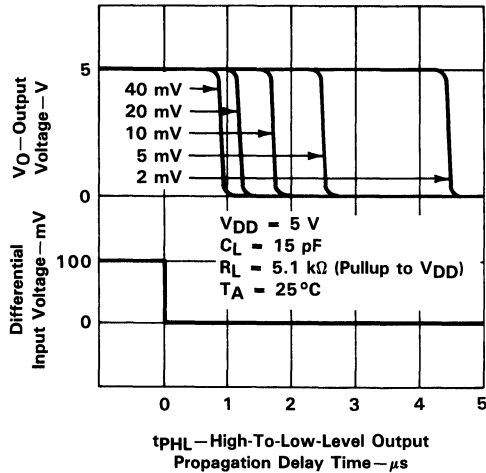


FIGURE 18

3
Voltage Comparators

TLC393M, TLC393I, TLC393C
DUAL MICROPOWER LinCMOS™ COMPARATORS

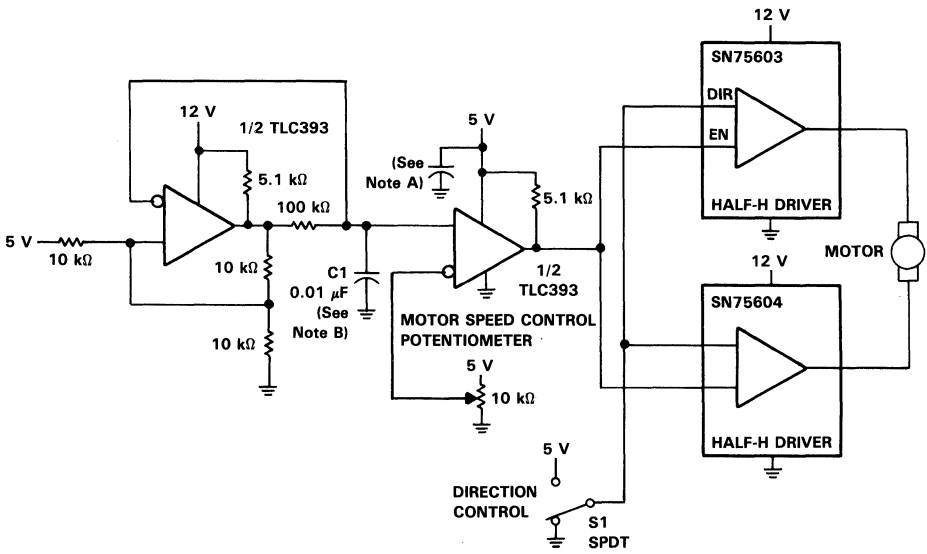
TYPICAL APPLICATION DATA

The input should always remain within the supply rails in order to avoid forward biasing the diodes in the electrostatic discharge (ESD) protection structure. If either input exceeds this range, the device will not be damaged as long as the input current is limited to less than 5 mA. To maintain the expected output state, the inputs must remain within the common-mode range. For example, at 25 °C with $V_{DD} = 5\text{ V}$, both inputs must remain between -0.2 V and 4 V to assure proper device operation.

To assure reliable operation, the supply should be decoupled with a capacitor ($0.1\ \mu\text{F}$) positioned as close to the device as possible.

The TLC393 has internal ESD-protection circuits that will prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

3 Voltage Comparators



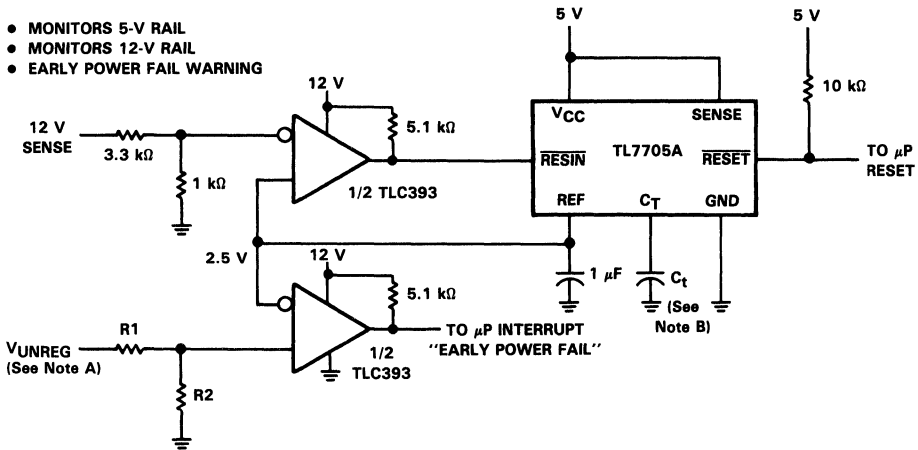
- NOTES: A. The recommended minimum capacitance is $10\ \mu\text{F}$ to eliminate common ground switching noise.
 B. Select C1 for change in oscillator frequency.

FIGURE 19. PULSE-WIDTH-MODULATED MOTOR SPEED CONTROLLER

TLC393M, TLC393I, TLC393C DUAL MICROPOWER LinCMOS™ COMPARATORS

TYPICAL APPLICATION DATA

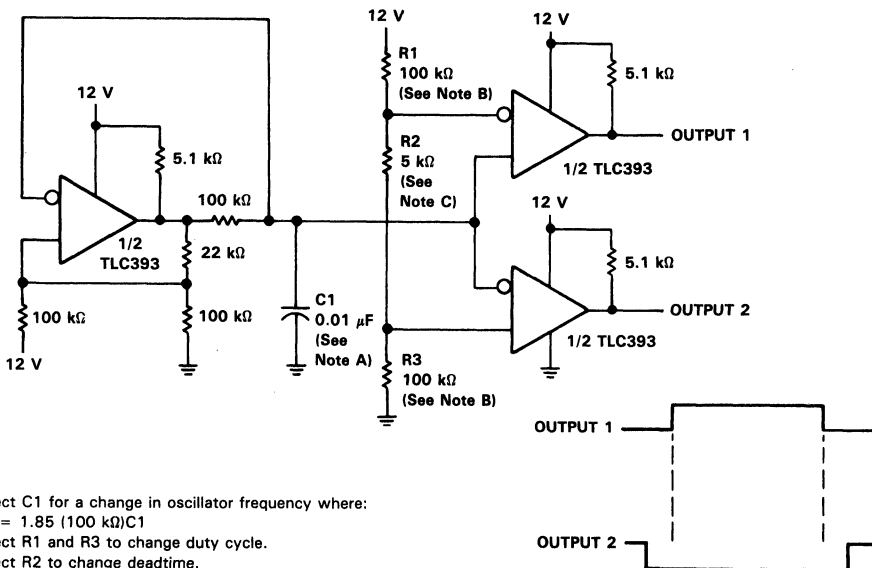
- MONITORS 5-V RAIL
- MONITORS 12-V RAIL
- EARLY POWER FAIL WARNING



NOTES: A. $V_{UNREG} = 2.5 \left(\frac{R1 + R2}{R2} \right)$

B. The value of C_t determines the time delay of reset.

FIGURE 20. ENHANCED SUPPLY SUPERVISOR



NOTES: A. Select $C1$ for a change in oscillator frequency where:

$$1/f = 1.85 (100 \text{ k}\Omega) C1$$

B. Select $R1$ and $R3$ to change duty cycle.

C. Select $R2$ to change deadtime.

FIGURE 21. TWO-PHASE NONOVERLAPPING CLOCK GENERATOR

3
Voltage Comparators

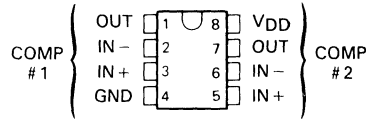
3 Voltage Comparators

TLC3702M, TLC3702I, TLC3702C DUAL MICROPOWER LinCMOS™ COMPARATORS

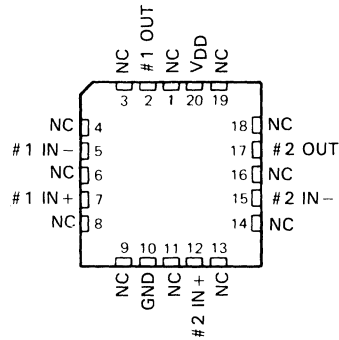
D3209, NOVEMBER 1986—REVISED JANUARY 1989

- Push-Pull CMOS Output Drives Capacitive Loads without Pull-Up Resistor, $I_O = \pm 8 \text{ mA}$
- Very Low Power ... $100 \mu\text{W Typ at } 5 \text{ V}$
- Fast Response Time ... $2.5 \mu\text{s Typ with } 5 \text{ mV Overdrive}$
- Single Supply Operation:
TLC3702M ... 4 V to 16 V
TLC3702I ... 3 V to 16 V
TLC3702C ... 3 V to 16 V
- High Input Impedance ... $10^{12} \Omega \text{ Typ}$
- Input Offset Voltage Change at Worst Case Input Condition Typically $0.23 \mu\text{V/Month}$ Including the First 30 Days
- On-Chip ESD Protection

TLC3702M ... JG PACKAGE
TLC3702I, TLC3702C ... D, JG, OR P PACKAGE
(TOP VIEW)

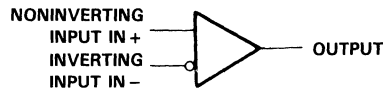


TLC3702M ... FK PACKAGE
(TOP VIEW)



NC—No internal connection

symbol (each comparator)



description

The TLC3702 consists of two independent differential-voltage comparators designed to operate from a single supply and be compatible with modern HCMOS logic systems. It is functionally similar to the LM393 but uses 1/20th the power for similar response times. The push-pull CMOS output stage will drive capacitive loads directly without a power-consuming pull-up resistor to achieve the stated response time. Eliminating the pull-up resistor not only reduces power dissipation, but also saves board space and component cost. The output stage is also fully compatible with TTL requirements.

AVAILABLE OPTIONS

T _A	V _{IO} max at 25°C	PACKAGE			
		SMALL OUTLINE (D)	CERAMIC (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	5 mV	TLC3702CD	—	TLC3702CJG	TLC3702CP
-40°C to 85°C	5 mV	TLC3702ID	—	TLC3702IJG	TLC3702IP
-55°C to 125°C	5 mV	—	TLC3702MFK	TLC3702MJG	—

The D package is available taped and reeled. Add the suffix R to the device type when ordering. (e.g., TL3702CDR)

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TEXAS
INSTRUMENTS

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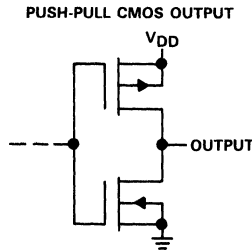
TLC3702M, TLC3702I, TLC3702C DUAL MICROWPOWER LinCMOS™ COMPARATORS

description (continued)

Texas Instruments LinCMOS™ process offers superior analog performance to standard CMOS processes. Along with the standard CMOS advantages of low power without sacrificing speed, high input impedance, and low bias currents, the LinCMOS™ process offers extremely stable input offset voltages, even with differential input stresses of several volts. This characteristic makes it possible to build reliable CMOS comparators.

The TLC3702M is characterized for operation over the full military temperature range of -55°C to 125°C . The TLC3702I is characterized for operation over the extended industrial temperature range of -40°C to 85°C . The TLC3702C is characterized for operation over the commercial temperature range of 0°C to 70°C .

schematic



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{DD} (see Note 1)	$-0.3\text{ V to }18\text{ V}$
Differential input voltage (see Note 2)	$\pm 18\text{ V}$
Input voltage, V_I	$-0.3\text{ V to }V_{DD}$
Output voltage, V_O	$-0.3\text{ V to }V_{DD}$
Input current, I_I	$\pm 5\text{ mA}$
Output current, I_O (each output)	$\pm 20\text{ mA}$
Total supply current into V_{DD} terminal	40 mA
Total current out of ground terminal	40 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: TLC3702M	$-55^{\circ}\text{C to }125^{\circ}\text{C}$
TLC3702I	$-40^{\circ}\text{C to }85^{\circ}\text{C}$
TLC3702C	$0^{\circ}\text{C to }70^{\circ}\text{C}$
Storage temperature range	$-65^{\circ}\text{C to }150^{\circ}\text{C}$
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.
2. Differential voltages are at the noninverting input with respect to the inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$	DERATING FACTOR	$T_A = 70^{\circ}\text{C}$	$T_A = 85^{\circ}\text{C}$	$T_A = 125^{\circ}\text{C}$
	POWER RATING	ABOVE $T_A = 25^{\circ}\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/ $^{\circ}\text{C}$	464 mW	377 mW	—
FK	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	715 mW	275 mW
JG (TLC3702M)	1050 mW	8.4 mW/ $^{\circ}\text{C}$	672 mW	546 mW	210 mW
JG (All others)	825 mW	6.6 mW/ $^{\circ}\text{C}$	528 mW	429 mW	—
P	1000 mW	8.0 mW/ $^{\circ}\text{C}$	640 mW	520 mW	—

TLC3702M DUAL MICROPOWER LinCMOS™ COMPARATORS

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	4	5	16	V
Common-mode input voltage, V_{IC}	0		$V_{DD}-1.5$	V
High-level output current, I_{OH}			-20	mA
Low-level output current, I_{OL}			20	mA
Operating free-air temperature, T_A	-55		125	°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	$V_{IC} = V_{ICRmin}$, $V_{DD} = 5\text{ V to }10\text{ V}$, See Note 3	25°C	1.2	5	mV
			-55°C to 125°C		10	
I_{IO}	Input offset current	$V_{IC} = 2.5\text{ V}$	25°C	1		pA
			125°C		15	nA
I_{IB}	Input bias current	$V_{IC} = 2.5\text{ V}$	25°C	5		pA
			125°C		30	nA
V_{ICR}	Common-mode input voltage range		25°C	0 to $V_{DD}-1$		V
			-55°C to 125°C	0 to $V_{DD}-1.5$		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84		dB
			125°C	83		
			-55°C	82		
kSVR	Supply voltage rejection ratio	$V_{DD} = 5\text{ V to }10\text{ V}$	25°C	85		dB
			125°C	85		
			-55°C	82		
V_{OH}	High-level output voltage	$V_{ID} = 1\text{ V}$, $I_{OH} = -4\text{ mA}$	25°C	4.5	4.7	V
			125°C	4.2		
V_{OL}	Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 4\text{ mA}$	25°C	210	300	mV
			125°C		500	
I_{DD}	Supply current (both comparators)	No load, Outputs low	25°C	18	40	μA
			-55°C to 125°C		90	

† All characteristics are measured with zero common-mode voltage unless otherwise noted.

NOTE 3: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.

Voltage Comparators

TLC37021

DUAL MICROPOWER LinCMOS™ COMPARATORS

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	3	5	16	V
Common-mode input voltage, V_{IC}	-0.2	$V_{DD}-1.5$		V
High-level output current, I_{OH}				-20 mA
Low-level output current, I_{OL}				20 mA
Operating free-air temperature, T_A	-40	85		°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	$V_{IC} = V_{ICRmin}$, $V_{DD} = 5\text{ V to }10\text{ V}$, See Note 3	25°C	1.2	5	mV
			-40°C to 85°C	7		
I_{IO}	Input offset current	$V_{IC} = 2.5\text{ V}$	25°C	1		pA
			85°C	1		nA
I_{IB}	Input bias current	$V_{IC} = 2.5\text{ V}$	25°C	5		pA
			85°C	2		nA
V_{ICR}	Common-mode input voltage range		25°C	0 to $V_{DD}-1$		V
			-40°C to 85°C	0 to $V_{DD}-1.5$		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84		dB
			85°C	84		
			-40°C	83		
k_{SVR}	Supply voltage rejection ratio	$V_{DD} = 5\text{ V to }10\text{ V}$	25°C	85		dB
			85°C	85		
			-40°C	83		
V_{OH}	High-level output voltage	$V_{ID} = 1\text{ V}$, $I_{OH} = -4\text{ mA}$	25°C	4.5	4.7	V
			85°C	4.3		
V_{OL}	Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 4\text{ mA}$	25°C	210	300	mV
			85°C	400		
I_{DD}	Supply current (both comparators)	No load, Outputs low	25°C	18	40	µA
			-40°C to 85°C	65		

† All characteristics are measured with zero common-mode voltage unless otherwise noted.

NOTE 3: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.

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Voltage Comparators

TLC3702C DUAL MICROPOWER LinCMOS™ COMPARATORS

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	3	5	16	V
Common-mode input voltage, V_{IC}	-0.2		$V_{DD}-1.5$	V
High-level output current, I_{OH}			-20	mA
Low-level output current, I_{OL}			20	mA
Operating free-air temperature, T_A	0		70	°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	$V_{IC} = V_{ICRmin}$, $V_{DD} = 5\text{ V to }10\text{ V}$, See Note 3	25°C	1.2	5	mV
			0°C to 70°C		6.5	
I_{IO}	Input offset current	$V_{IC} = 2.5\text{ V}$	25°C	1		pA
			70°C		0.3	nA
I_{IB}	Input bias current	$V_{IC} = 2.5\text{ V}$	25°C	5		pA
			70°C		0.6	nA
V_{ICR}	Common-mode input voltage range		25°C	0 to $V_{DD}-1$		V
			0°C to 70°C	0 to $V_{DD}-1.5$		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84		dB
			70°C	84		
			0°C	84		
k_{SVR}	Supply voltage rejection ratio	$V_{DD} = 5\text{ V to }10\text{ V}$	25°C	85		dB
			70°C	85		
			0°C	85		
V_{OH}	High-level output voltage	$V_{ID} = 1\text{ V}$, $I_{OH} = -4\text{ mA}$	25°C	4.5	4.7	V
			70°C	4.3		
V_{OL}	Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 4\text{ mA}$	25°C	210	300	mV
			70°C		375	
I_{DD}	Supply current (both comparators)	No load, Outputs low	25°C	18	40	µA
			0°C to 70°C		50	

† All characteristics are measured with zero common-mode voltage unless otherwise noted.

NOTE 3: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.

3

Voltage Comparators

TLC3702M, TLC3702I, TLC3702C
DUAL MICROPOWER LinCMOS™ COMPARATORS

switching characteristics, $V_{DD} = 5\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
t _{PLH}	Propagation delay time, low-to-high-level output	f = 10 kHz, C _L = 50 pF	Overdrive = 2 mV		4.5		μs
			Overdrive = 5 mV		2.7		
			Overdrive = 10 mV		1.9		
			Overdrive = 20 mV		1.4		
			Overdrive = 40 mV		1.1		
		V _I = 1.4 V step at IN+ pin			1.1		
t _{PHL}	Propagation delay time, high-to-low-level output	f = 10 kHz, C _L = 50 pF	Overdrive = 2 mV		4.0		μs
			Overdrive = 5 mV		2.3		
			Overdrive = 10 mV		1.5		
			Overdrive = 20 mV		0.95		
			Overdrive = 40 mV		0.65		
		V _I = 1.4 V step at IN+ pin			0.15		
t _f	Fall time	f = 10 kHz, C _L = 50 pF	Overdrive = 50 mV		50		ns
t _r	Rise time	f = 10 kHz, C _L = 50 pF	Overdrive = 50 mV		125		ns

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Voltage Comparators



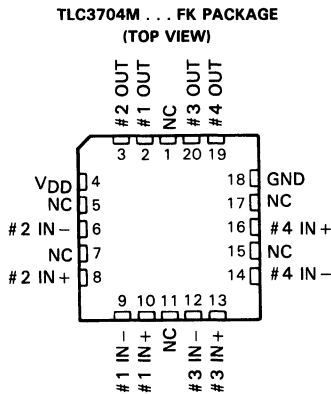
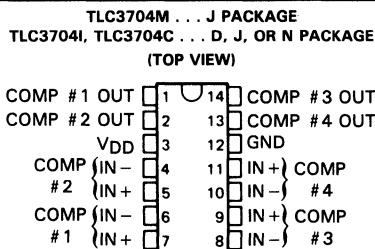
TLC3704M, TLC3704I, TLC3704C QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

D3240, NOVEMBER 1986—REVISED MARCH 1989

- Push-Pull CMOS Output Drives Capacitive Loads without Pull-Up Resistor, $I_O = \pm 8 \text{ mA}$
- Very Low Power . . . 200 μW Typ at 5 V
- Fast Response Time . . . 2.5 μs Typ with 5 mV Overdrive
- Single Supply Operation:
TLC3704M . . . 4 V to 16 V
TLC3704I . . . 3 V to 16 V
TLC3704C . . . 3 V to 16 V
- High Input Impedance . . . $10^{12} \Omega$ Typ
- Input Offset Voltage Change at Worst Case Input Condition Typically 0.23 $\mu\text{V}/\text{Month}$ Including the First 30 Days
- On-Chip ESD Protection

description

The TLC3704 consists of four independent differential-voltage comparators designed to operate from a single supply and be compatible with modern HCMOS logic systems. It is functionally similar to the LM339 but uses 1/20th the power for similar response times. The push-pull CMOS output stage will drive capacitive loads directly without a power-consuming pull-up resistor to achieve the stated response time. Eliminating the pull-up resistor not only reduces power dissipation, but also saves board space and component cost. The output stage is also fully compatible with TTL requirements.



NC—No internal connection

AVAILABLE OPTIONS

T_A	$V_{IO\max}$ at 25°C	PACKAGE			
		SMALL OUTLINE (D)	CERAMIC (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	5 mV	TLC3704CD	—	TLC3704CJ	TLC3704CN
-40°C to 85°C	5 mV	TLC3704ID	—	TLC3704IJ	TLC3704IN
-55°C to 125°C	5 mV	—	TLC3704MFK	TLC3704MJ	—

The D package is available taped and reeled. Add the suffix R to the device type. (e.g., TLC3704CDR)

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Voltage Comparators

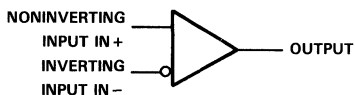
TLC3704M, TLC3704I, TLC3704C QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

description (continued)

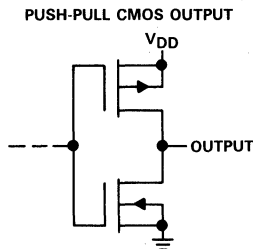
Texas Instruments LinCMOS™ process offers superior analog performance to standard CMOS processes. Along with the standard CMOS advantages of low power without sacrificing speed, high input impedance, and low bias currents, the LinCMOS™ process offers extremely stable input offset voltages, even with differential input stresses of several volts. This characteristic makes it possible to build reliable CMOS comparators.

The TLC3704M is characterized for operation over the full military temperature range of -55°C to 125°C . The TLC3704I is characterized for operation over the extended industrial temperature range of -40°C to 85°C . The TLC3704C is characterized for operation over the commercial temperature range of 0°C to 70°C .

symbol (each comparator)



schematic



3

Voltage Comparators

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{DD} (see Note 1)	-0.3 to 18 V
Differential input voltage (see Note 2)	± 18 V
Input voltage, V_I	-0.3 V to V_{DD}
Output voltage, V_O	-0.3 V to V_{DD}
Input current, I_I	± 5 mA
Output current, I_O (each output)	± 20 mA
Total supply current into V_{DD} terminal	40 mA
Total current out of ground terminal	60 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range: TLC3704M	-55°C to 125°C
TLC3704I	-40°C to 85°C
TLC3704C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.
2. Differential voltages are at the noninverting input with respect to the inverting input.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$	DERATING FACTOR	$T_A = 70^{\circ}\text{C}$	$T_A = 85^{\circ}\text{C}$	$T_A = 125^{\circ}\text{C}$
	POWER RATING	ABOVE $T_A = 25^{\circ}\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/ $^{\circ}\text{C}$	608 mW	494 mW	—
FK	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	715 mW	275 mW
J (TLC3704M)	1375 mW	11.0 mW/ $^{\circ}\text{C}$	880 mW	715 mW	275 mW
J (All others)	1025 mW	8.2 mW/ $^{\circ}\text{C}$	656 mW	533 mW	—
N	1150 mW	9.2 mW/ $^{\circ}\text{C}$	736 mW	598 mW	—

TLC3704M
QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	4	5	16	V
Common-mode input voltage, V_{IC}	0	$V_{DD} - 1.5$		V
High-level output current, I_{OH}				-20 mA
Low-level output current, I_{OL}				20 mA
Operating free-air temperature, T_A	-55	125		°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	MIN	TYP	MAX	UNIT
V_{IO} Input offset voltage	$V_{IC} = V_{ICRmin}$, $V_{DD} = 5$ V to 10 V, See Note 3	25 °C	1.2	5	mV
		-55 °C to 125 °C		10	
I_{IO} Input offset current	$V_{IC} = 2.5$ V	25 °C	1		pA
		125 °C		15	nA
I_{IB} Input bias current	$V_{IC} = 2.5$ V	25 °C	5		pA
		125 °C		30	nA
V_{ICR} Common-mode input voltage range		25 °C	0 to $V_{DD} - 1$		V
		-55 °C to 125 °C	0 to $V_{DD} - 1.5$		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25 °C		84	dB
		125 °C		83	
		-55 °C		82	
k_{SVR} Supply voltage rejection ratio	$V_{DD} = 5$ V to 10 V	25 °C		85	dB
		125 °C		85	
		-55 °C		82	
V_{OH} High-level output voltage	$V_{ID} = 1$ V, $I_{OH} = -4$ mA	25 °C	4.5	4.7	V
		125 °C	4.2		
V_{OL} Low-level output voltage	$V_{ID} = -1$ V, $I_{OL} = 4$ mA	25 °C	210	300	mV
		125 °C		500	
I_{DD} Supply current (four comparators)	No load, Outputs low	25 °C	35	80	μA
		-55 °C to 125 °C		175	

†All characteristics are measured with zero common-mode voltage unless otherwise noted.

NOTE 3: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.

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Voltage Comparators

TLC3704I
QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	3	5	16	V
Common-mode input voltage, V_{IC}	-0.2	$V_{DD}-1.5$		V
High-level output current, I_{OH}				-20 mA
Low-level output current, I_{OL}				20 mA
Operating free-air temperature, T_A	-40			85 °C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	MIN	TYP	MAX	UNIT
V_{IO} Input offset voltage	$V_{IC} = V_{ICRmin}$, $V_{DD} = 5\text{ V to }10\text{ V}$, See Note 3	25°C	1.2	5	mV
		-40°C to 85°C		7	
I_{IO} Input offset current	$V_{IC} = 2.5\text{ V}$	25°C	1		pA
		85°C		1	nA
I_{IB} Input bias current	$V_{IC} = 2.5\text{ V}$	25°C	5		pA
		85°C		2	nA
V_{ICR} Common-mode input voltage range		25°C	0 to $V_{DD}-1$		V
		-40°C to 85°C	0 to $V_{DD}-1.5$		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84		dB
		85°C	84		
		-40°C	83		
k_{SVR} Supply voltage rejection ratio	$V_{DD} = 5\text{ V to }10\text{ V}$	25°C	85		dB
		85°C	85		
		-40°C	83		
V_{OH} High-level output voltage	$V_{ID} = 1\text{ V}$, $I_{OH} = -4\text{ mA}$	25°C	4.5	4.7	V
		85°C	4.3		
V_{OL} Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 4\text{ mA}$	25°C	210	300	mV
		85°C	400		
I_{DD} Supply current (four comparators)	No load, Outputs low	25°C	35	80	μA
		-40°C to 85°C	125		

†All characteristics are measured with zero common-mode voltage unless otherwise noted.

NOTE 3: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.

3

Voltage Comparators

TLC3704C

QUADRUPLE MICROPOWER LinCMOS™ COMPARATORS

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	3	5	16	V
Common-mode input voltage, V_{IC}	-0.2		$V_{DD} - 1.5$	V
High-level output current, I_{OH}			-20	mA
Low-level output current, I_{OL}			20	mA
Operating free-air temperature, T_A	0		70	°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	MIN	TYP	MAX	UNIT
V_{IO} Input offset voltage	$V_{IC} = V_{ICRmin}$, $V_{DD} = 5\text{ V to }10\text{ V}$, See Note 3	25°C	1.2	5	mV
		0°C to 70°C		6.5	
I_{IO} Input offset current	$V_{IC} = 2.5\text{ V}$	25°C	1		pA
		70°C		0.3	nA
I_{IB} Input bias current	$V_{IC} = 2.5\text{ V}$	25°C	5		pA
		70°C		0.6	nA
V_{ICR} Common-mode input voltage range		25°C	0 to $V_{DD} - 1$		V
		0°C to 70°C	0 to $V_{DD} - 1.5$		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84		dB
		70°C	84		
		0°C	84		
k_{SVR} Supply voltage rejection ratio	$V_{DD} = 5\text{ V to }10\text{ V}$	25°C	85		dB
		70°C	85		
		0°C	85		
V_{OH} High-level output voltage	$V_{ID} = 1\text{ V}$, $I_{OH} = -4\text{ mA}$	25°C	4.5	4.7	V
		70°C	4.3		
V_{OL} Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OL} = 4\text{ mA}$	25°C	210	300	mV
		70°C		375	
I_{DD} Supply current (four comparators)	No load, Outputs low	25°C	35	80	μA
		0°C to 70°C		100	

†All characteristics are measured with zero common-mode voltage unless otherwise noted.

NOTE 3: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.

3

Voltage Comparators

TLC3704M, TLC3704I, TLC3704C
QUADRUPLÉ MICROPOWER LinCMOS™ COMPARATORS

switching characteristics, $V_{DD} = 5\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
t _{PLH} Propagation delay time, low-to-high-level output	f = 10 kHz, C _L = 50 pF	Overdrive = 2 mV		4.5		μs
		Overdrive = 5 mV		2.7		
		Overdrive = 10 mV		1.9		
		Overdrive = 20 mV		1.4		
		Overdrive = 40 mV		1.1		
		V _I = 1.4-V step at IN+ pin		1.1		
t _{PHL} Propagation delay time, high-to-low-level output	f = 10 kHz, C _L = 50 pF	Overdrive = 2 mV		4.0		μs
		Overdrive = 5 mV		2.3		
		Overdrive = 10 mV		1.5		
		Overdrive = 20 mV		0.95		
		Overdrive = 40 mV		0.65		
		V _I = 1.4-V step at IN+ pin		0.15		
t _f Fall time, high-to-low-level output	f = 10 kHz, C _L = 50 pF	Overdrive = 50 mV		50		ns
t _r Rise time, low-to-high-level output	f = 10 kHz, C _L = 50 pF	Overdrive = 50 mV		125		ns

3

Voltage Comparators

General Information

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Video Amplifiers
Hall-Effect Devices
Timers and Current Mirrors
Magnetic Memory Controllers
Sound Generators
Frequency-to-Voltage Converters
Sonar Ranging Circuits and Modules

4

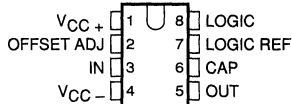
Special Functions

LF198, LF198A, LF398, LF398A PRECISION SAMPLE-AND-HOLD AMPLIFIERS

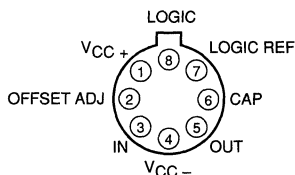
D3218, JULY 1988 – REVISED MARCH 1989

- Acquisition Time . . . 4 μ s Typ
- Gain Error . . . 0.01% Max,
0.001% Typ for LF198A, LF398A
- Input Offset Voltage . . .
1 mV Max for LF198A, 2 mV Max for LF398A
- Hold Step . . . 1 mV Max for LF198A, LF398A
- Very Low Feedthrough Attenuation Ratio at
f = 1 kHz . . . 96 dB Typ
- High Input Impedance . . . 10^{10} Ω Typ
- Logic Inputs Compatible With All Logic
Families

JG OR P PACKAGE
(TOP VIEW)



L PACKAGE
(TOP VIEW)



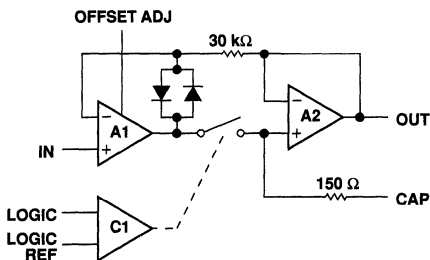
description

These sample-and-hold amplifiers use a combination of bipolar and junction FET transistors to provide precision, high speed, and long hold times. Input offset voltages as low as 1 mV (LF198A) and gain errors as low as 0.001% (LF198A, LF398A) allow these amplifiers to be used in 12-bit systems. Properly selecting the external hold capacitor optimizes the dynamic performance. Acquisition times can be as low as 4 μ s for small capacitors, while hold step and droop errors can be held below 0.1 mV and 30 μ V/s, respectively, when using larger capacitors.

The LF198 and LF398 are fixed at unity gain with 10^{10} - Ω input impedance independent of the sample or hold mode. The logic inputs are at a high differential impedance to allow easy interfacing to any logic family without ground loop problems. A separate offset adjust pin can be used to zero the input offset voltage in either the sample or hold mode. Additionally, the hold capacitor can be driven with an external signal to provide precision level shifting or "differencing" operation. The devices operate over a wide supply voltage range from ± 5 V to ± 18 V with very little change in performance. Key parameters are specified over this full supply range.

The LF198 and LF198A are characterized for operation over the full military temperature range of -55°C to 125°C . The LF398 and LF398A are characterized for operation from 0°C to 70°C .

functional block diagram



AVAILABLE OPTIONS

T _A	V _{IO} max AT 25°C	PACKAGE		
		CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)
0°C to 70°C	2 mV	—	LF398AL	LF398AP
	7 mV	LF398JG	LF398L	LF398P
-55°C to 125°C	1 mV	—	LF198AL	—
	3 mV	—	LF198L	—

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

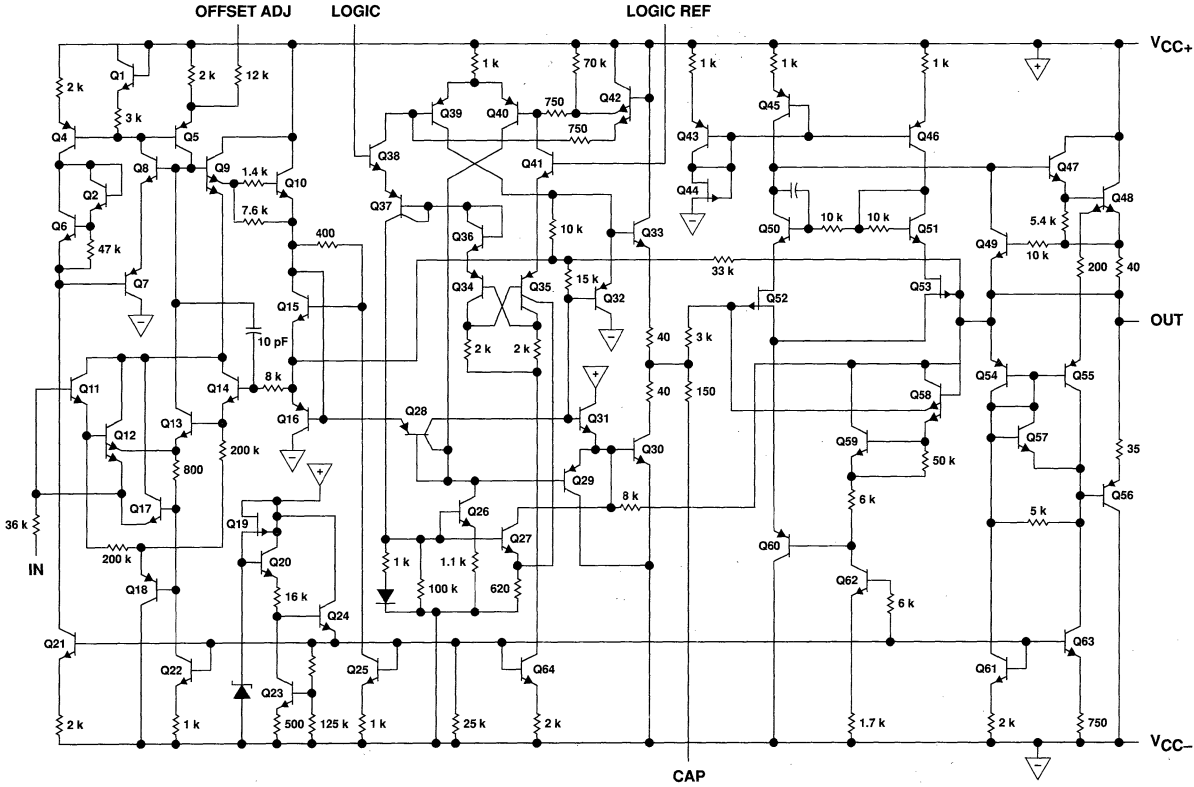
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4
Special Functions

LF198, LF198A, LF398, LF398A
PRECISION SAMPLE-AND-HOLD AMPLIFIERS

Schematic



4 Special Functions

All resistor values shown are nominal and in ohms.

LF198, LF198A, LF398, LF398A PRECISION SAMPLE-AND-HOLD AMPLIFIERS

definition of terms

Acquisition Time

The time required to acquire, within a defined error, a new analog input voltage with an output change of 10 V. Acquisition time includes output settling time and includes the time required for all internal nodes to settle so that the output is at the proper value when switched to the hold mode.

Aperture Time

The delay required between a hold command and an input analog transition so that the transition does not affect the held output.

Dynamic Sampling Error

The error introduced into the held output voltage due to a changing analog input at the time the hold command is given. Dynamic sampling error is expressed in mV with a given hold capacitor value and input slew rate. Note that this error term occurs even for long sample times.

Gain Error

The ratio of output voltage swing to input voltage swing in the sample mode expressed as a percent difference.

Hold Settling Time

The time required for the output to settle within 1 mV of final value after a hold command is initiated.

Hold Step

The voltage step at the output of the amplifier when switching from sample mode to hold mode with a constant analog input voltage and a logic swing of 5 V.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	18 V
Supply voltage, V_{CC-} (see Note 1)	-18 V
Input voltage range, V_I (any input, see Note 1)	$V_{CC\pm}$
Differential input voltage, logic to logic reference (see Note 2)	± 30 V
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Duration of hold capacitor short-circuit current	10 s
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, T_A : LF198, LF198A	-55°C to 125°C
LF398, LF398A	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: L package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: P package	260°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .
2. The logic inputs are protected to ± 30 V differential as long as the voltage on both pins does not exceed the supply voltage. For proper operation, however, both the logic and logic reference pins must be at least 2 V below the positive supply, and one of these pins must be at least 3 V above the negative supply.
3. The output may be shorted to either supply.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATING ABOVE T_A	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
JG (LF198_)	500 mW	8.4 mW/°C	90°C	500 mW	210 mW
JG (LF398_)	500 mW	6.6 mW/°C	74°C	500 mW	N/A
L (LF198_)	500 mW	6.6 mW/°C	74°C	500 mW	165 mW
L (LF398_)	500 mW	5.2 mW/°C	54°C	416 mW	N/A
P	500 mW	8.0 mW/°C	88°C	500 mW	200 mW

LF198, LF398 PRECISION SAMPLE-AND-HOLD AMPLIFIERS

electrical characteristics in sample mode, $V_{CC\pm} = \pm 15\text{ V}$, $V_I = \pm 11.5\text{ V}$, $C_H = 0.01\ \mu\text{F}$, $R_L = 10\ \text{k}\Omega$, logic reference at 0 V, logic at 2.5 V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A^\dagger	LF198			LF398			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{CC\pm} = \pm 5\text{ V to } \pm 18\text{ V}$	25°C		1	3		2	7	mV
		Full range			5			10	
I_{IB} Input bias current	$V_{CC\pm} = \pm 5\text{ V to } \pm 18\text{ V}$	25°C		5	25		10	50	nA
		Full range			75			100	
Differential logic threshold voltage		25°C	0.8	1.4	2.4	0.8	1.4	2.4	V
Input current, logic and logic reference		25°C		2	10		2	10	μA
Leakage current into hold capacitor terminal	$V_{CC\pm} = \pm 5\text{ V to } \pm 18\text{ V}$, Hold mode, See Note 4	25°C		30	100		30	200	pA
Hold capacitor charging current	$V_I - V_O = 2\text{ V}$	25°C		5			5		mA
z_i Input impedance		25°C		10^{10}			10^{10}		Ω
z_o Output impedance	Hold mode	25°C		0.5	2		0.5	4	Ω
		Full range			4			6	
Gain error (see Note 5)	$R_L = 10\ \text{k}\Omega$	25°C		0.002	0.005		0.004	0.01	%
		Full range			0.02			0.02	
Feedthrough attenuation ratio	$f = 1\ \text{kHz}$	25°C		86	96		80	96	dB
Hold step (see Note 6)	$V_O = 0$	25°C		0.5	2		0.5	2.5	mV
k_{SVR} Supply-voltage rejection ratio	$V_{CC\pm} = \pm 5\text{ V to } \pm 18\text{ V}$, $V_O = 0$	25°C		80	110		80	110	V
I_{CC} Supply current	$V_{CC\pm} = \pm 5\text{ V to } \pm 18\text{ V}$, $T_A \geq 25^\circ\text{C}$	25°C		4.5	5.5		4.5	6.5	mA
Acquisition time to 0.1% (see Note 5)	$\Delta V_O = 10\text{ V}$, $C_H = 1\ \mu\text{F}$ $\Delta V_O = 10\text{ V}$, $C_H = 0.01\ \mu\text{F}$	25°C		4			4		μs
					16			16	

[†]Full range is -55°C to 125°C for the LF198 and 0°C to 70°C for the LF398.

- NOTES: 4. The effects of a rise in junction temperature due to power dissipation or elevated ambient free-air temperature can be approximated by doubling the 25°C value for each 11°C increase in junction temperature. The specified limit applies for the full input signal range.
5. See definition of terms.
6. See definition of terms. Hold step is sensitive to stray capacitance coupling between input logic signals and the hold capacitor. Stray capacitance of 1 pF, for example, creates an additional 0.5-mV step with a 5-V logic swing and a 0.01- μF hold capacitor. The magnitude of the hold step is inversely proportional to the value of the hold capacitor.

4

Special Functions

LF198A, LF398A PRECISION SAMPLE-AND-HOLD AMPLIFIERS

electrical characteristics in sample mode, $V_{CC\pm} = \pm 15$ V, $V_I = \pm 11.5$ V, $C_H = 0.01$ μ F, $R_L = 10$ k Ω , logic reference at 0 V, logic at 2.5 V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A [†]	LF198A			LF398A			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{CC\pm} = \pm 5$ V to ± 18 V	25°C	0.5 1			1 2			mV
		Full range	2			3			
I_{IB} Input bias current		25°C	5 25			10 25			nA
		Full range	75			50			
Differential logic threshold voltage		25°C	0.8	1.4	2.4	0.8	1.4	2.4	V
Input current, logic and logic reference		25°C	2 10			2 10			μ A
Leakage current into hold capacitor terminal	$V_{CC\pm} = \pm 5$ V to ± 18 V, Hold mode, See Note 4	25°C	10 100			10 100			pA
Hold capacitor charging current	$V_I - V_O = 2$ V	25°C	5			5			mA
z_i Input impedance		25°C	10^{10}			10^{10}			Ω
z_o Output impedance	Hold mode	25°C	0.5 1			0.5 1			Ω
		Full range	4			6			
Gain error (see Note 5)	$R_L = 10$ k Ω	25°C	0.001 0.005			0.001 0.005			%
		Full range	0.01			0.01			
Feedthrough attenuation ratio	$f = 1$ kHz	25°C	86	96		86	96		dB
Hold step (see Note 6)	$V_O = 0$	25°C	0.25 1			0.25 1			mV
k_{SVR} Supply-voltage rejection ratio	$V_{CC\pm} = \pm 5$ V to ± 18 V, $V_O = 0$	25°C	90	110		90	110		V
I_{CC} Supply current	$V_{CC\pm} = \pm 5$ V to ± 18 V, $T_A \geq 25^\circ$ C	25°C	4.5 5.5			4.5 6.5			mA
		Acquisition time to 0.1% (see Note 5)	25°C	$\Delta V_O = 10$ V, $C_H = 1$ μ F		4 6		4 6	
				$\Delta V_O = 10$ V, $C_H = 0.01$ μ F		16 25		16 25	

[†]Full range is -55° C to 125° C for the LF198A and 0° C to 70° C for the LF398A.

- NOTES: 4. The effects of a rise in junction temperature due to power dissipation or elevated ambient free-air temperature can be approximated by doubling the 25°C value for each 11°C increase in junction temperature. The specified limit applies for the full input signal range.
5. See definition of terms.
6. See definition of terms. Hold step is sensitive to stray capacitance coupling between input logic signals and the hold capacitor. Stray capacitance of 1 pF, for example, creates an additional 0.5-mV step with a 5-V logic swing and a 0.01- μ F hold capacitor. The magnitude of the hold step is inversely proportional to the value of the hold capacitor.

4
Special Functions

**LF198, LF198A, LF398, LF398A
PRECISION SAMPLE-AND-HOLD AMPLIFIERS**

TYPICAL CHARACTERISTICS, $V_{CC\pm} = \pm 15\text{ V}^\dagger$

**INPUT BIAS CURRENT
VS
JUNCTION TEMPERATURE**

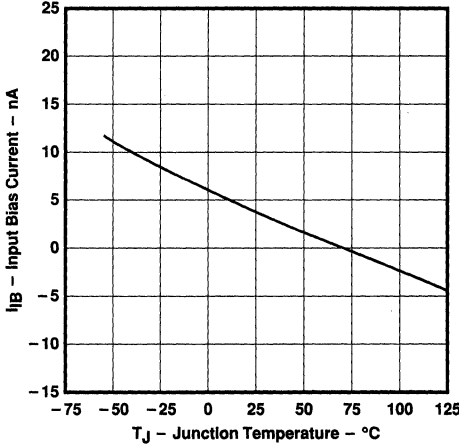


FIGURE 1

**LF198A, LF398A
CAPACITOR TERMINAL LEAKAGE CURRENT
VS
JUNCTION TEMPERATURE**

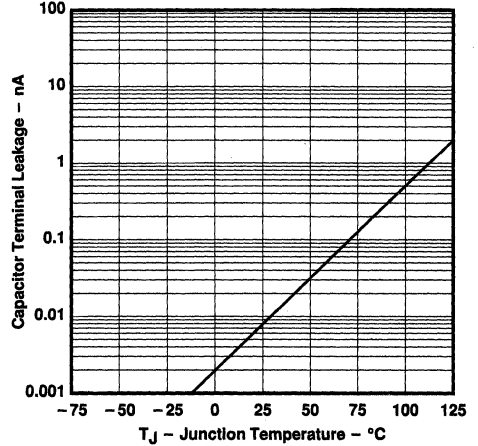


FIGURE 2

4

Special Functions

GAIN ERROR

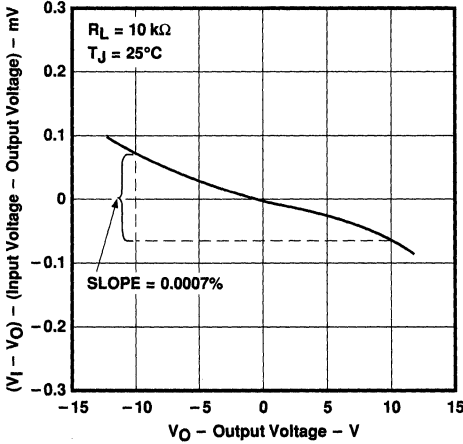


FIGURE 3

**FEEDTHROUGH ATTENUATION RATIO
VS
FREQUENCY**

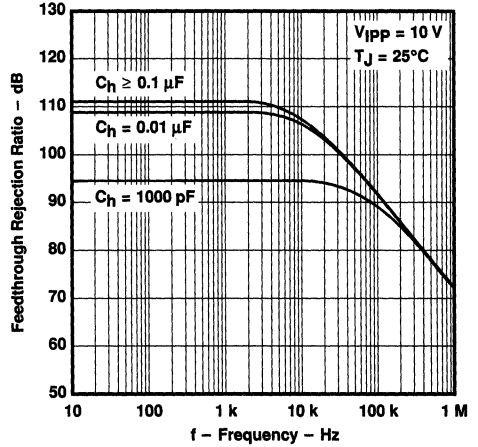


FIGURE 4

[†]Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

LF198, LF198A, LF398, LF398A PRECISION SAMPLE-AND-HOLD AMPLIFIERS

TYPICAL CHARACTERISTICS, $V_{CC\pm} = \pm 15\text{ V}^\dagger$

HOLD STEP
VS
INPUT VOLTAGE

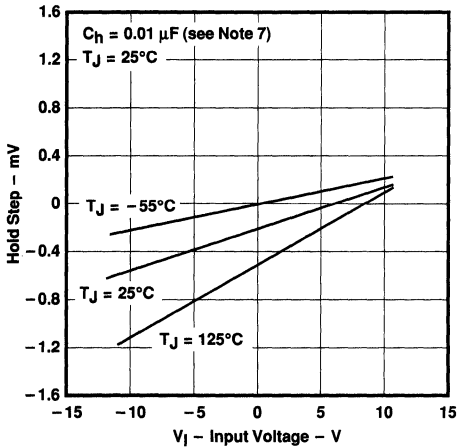


FIGURE 5

HOLD STEP
VS
HOLD CAPACITANCE

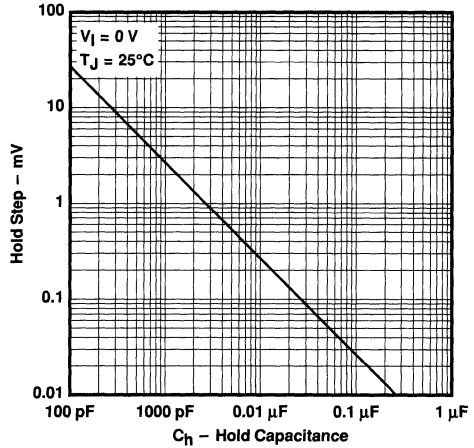


FIGURE 6

HOLD STEP
VS
LOGIC SLEW RATE

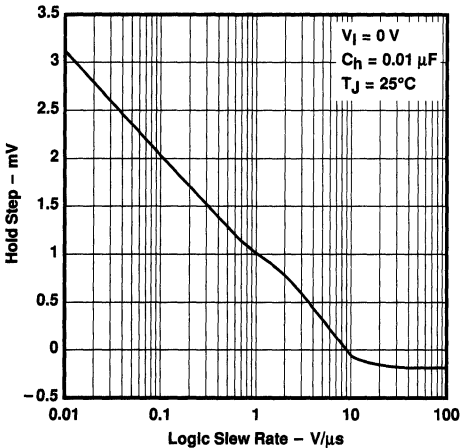


FIGURE 7

SMALL-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION and PHASE DELAY
VS
FREQUENCY

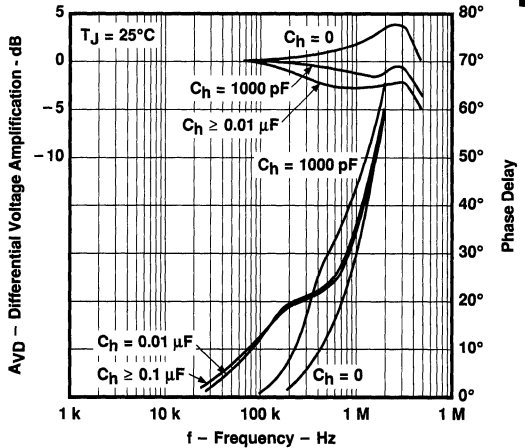


FIGURE 8

[†]Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.
NOTE 7: The amplitude of the hold step varies inversely with the value of the hold capacitor.

TYPICAL CHARACTERISTICS, $V_{CC\pm} = \pm 15\text{ V}^\dagger$

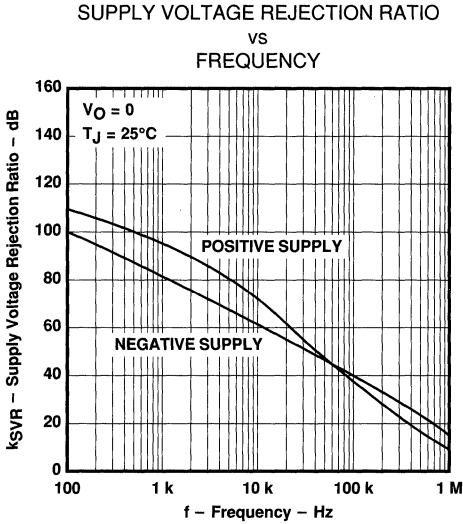


FIGURE 9

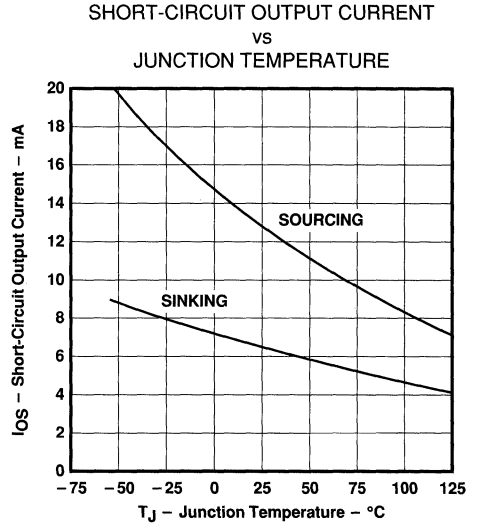


FIGURE 10

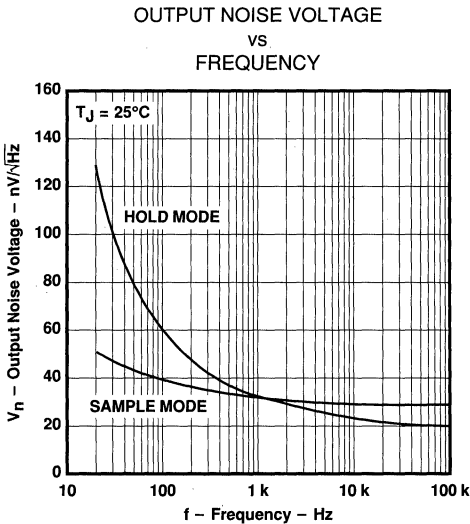


FIGURE 11

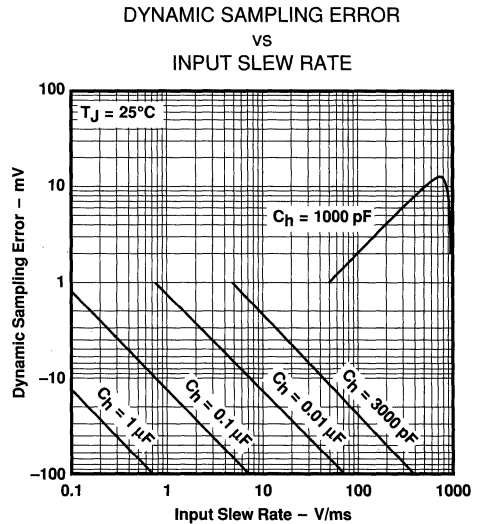


FIGURE 12

4

Special Functions

[†]Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

LF198, LF198A, LF398, LF398A PRECISION SAMPLE-AND-HOLD AMPLIFIERS

TYPICAL CHARACTERISTICS, $V_{CC\pm} = \pm 15\text{ V}^\dagger$

APERTURE TIME
vs
JUNCTION TEMPERATURE

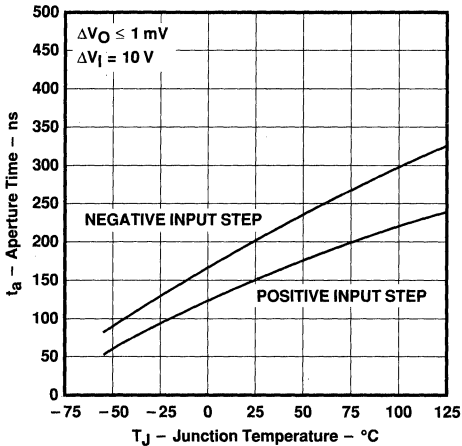


FIGURE 13

HOLD SETTLING TIME
vs
JUNCTION TEMPERATURE

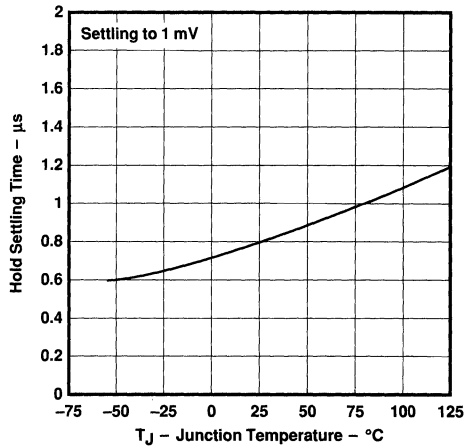


FIGURE 14

ACQUISITION TIME
vs
HOLD CAPACITANCE

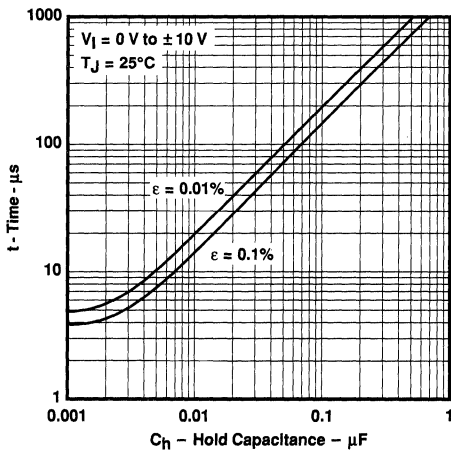


FIGURE 15

OUTPUT DROOP RATE
vs
HOLD CAPACITANCE

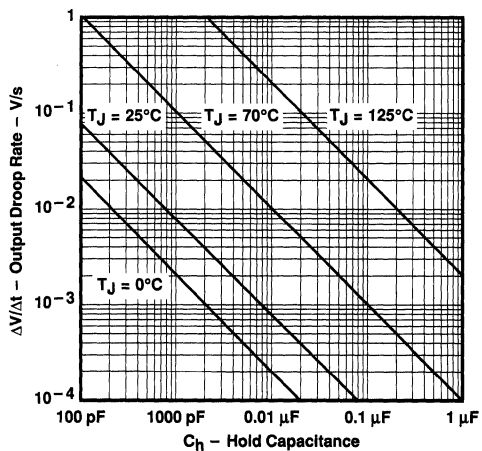


FIGURE 16

4 Special Functions

[†]Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS, $V_{CC\pm} = \pm 15 V^\dagger$

**CAPACITOR DIELECTRIC ABSORPTION
 vs
 SAMPLE TIME**

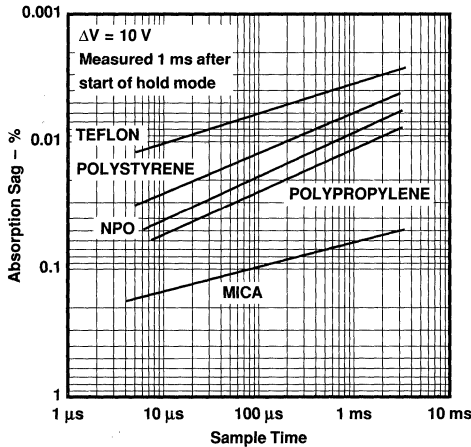


FIGURE 17

**CAPACITOR DIELECTRIC ABSORPTION
 vs
 TIME IN HOLD MODE**

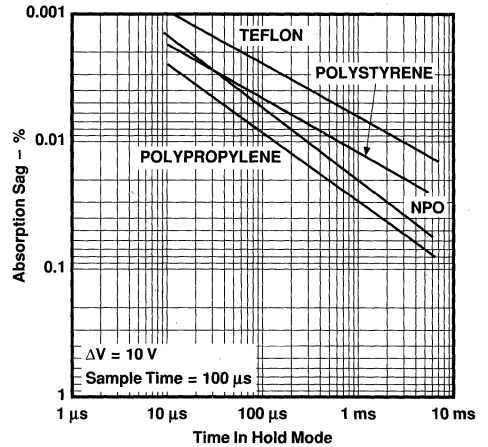


FIGURE 18

**OUTPUT TRANSIENT
 AT START OF HOLD MODE**

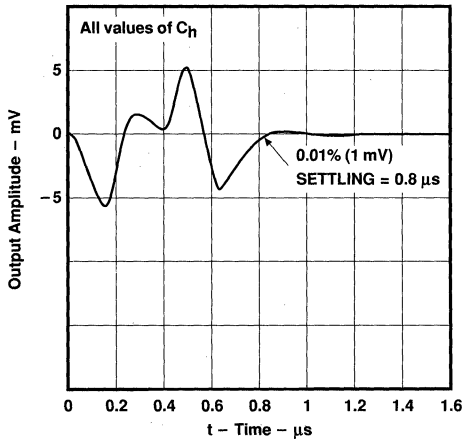


FIGURE 19

**OUTPUT TRANSIENT
 AT START OF SAMPLE PERIOD**

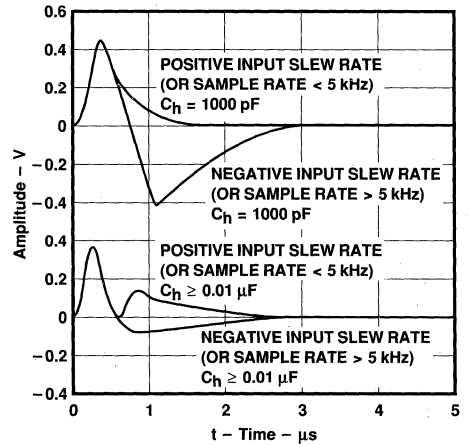


FIGURE 20

4

Special Functions

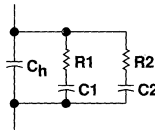
[†]Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL APPLICATION DATA

hold capacitor

For fast sample-and-hold applications, the value of the hold capacitor is critical. A low value gives fast acquisition but also increases errors due to hold step and droop caused by amplifier bias current. The capacitor should be made as large as possible consistent with acquisition time and dynamic sampling error requirements. Capacitors larger than 0.1 μF are generally not available in the low-loss dielectrics like Teflon, Polystyrene, and NPO, at least not at a reasonable price and size. Mylar, even with its poor dielectric absorption properties, may be a reasonable choice when very long sample times are used and low droop rates are needed.

Dielectric absorption in the hold capacitor can often be the major source of error in a sample-and-hold amplifier. The equivalent "circuit" of a typical capacitor with parallel RC networks used to model dielectric absorption is shown in Figure 21. In this capacitor, rapid changes in capacitor voltage are not tracked by the internal parasitic capacitors because of the resistance in series with them. This leads to a "sag" effect in the hold capacitor after a sudden change in voltage followed by a rapid switch to the hold mode. The capacitor remembers its previous state via the charge in the internal parasitic capacitance and sags back slightly toward the previous voltage. The magnitude of the sag depends upon the voltage change and the time spent sampling the new voltage. Several time constants are typically evident in the sag, although some capacitors tend to exhibit a single time constant, while others show a sag that indicates a blending of many time constants. Figures 17 and 18 show the amount of sag found after a 10-V step, with sample time at the new voltage and hold time at the new voltage as variables. It is obvious that sag problems are minimized by long sample times and short hold times. While this is often in conflict with basic sampling requirements, the sample-and-hold amplifier should be kept in the "tracking", or sampling, mode as much as possible to maximize the time the hold capacitor spends near the voltage at which it eventually "holds".



NOTE: C1 and C2 are approximately equal to 0.01C_h to 0.1C_h. R1 and R2 generate time constants of 0.1 to 50 ms with C1 and C2.

FIGURE 21. TYPICAL HOLD CAPACITOR EQUIVALENT CIRCUIT

The best capacitor for sample-and-hold applications is Teflon, which is clearly superior with regard to dielectric absorption and operates over the full temperature range (-55°C to 125°C). If size or price is a problem, the second choice for full-temperature-range operation is NPO (or COG) ceramic units. Some care must be used because not all NPO capacitors use the low-dielectric-constant ceramic necessary for low dielectric absorption. For lower temperature ($\leq 70^{\circ}\text{C}$), Polystyrene has traditionally been the best hold capacitor. The best units are cylindrical and fairly large; there seems to be a strong correlation between small size and poor dielectric performance. Polypropylene has nearly the same absorption properties as polystyrene and offers 85°C operation; it also tends to be smaller. Other standard dielectrics such as mica, glass, mylar, and ordinary ceramic are much worse with regard to dielectric absorption. Mylar is sometimes used for large values when the ratio of sample to hold time is large and extremely low droop is required.

dynamic sampling error

A significant sampling error can occur in any sample-and-hold amplifier if the input is moving when the unit is put into the hold mode. The two major causes for this error are digital delay in switch opening and analog delay across the hold capacitor. The switch opening delay is obvious and leads to a "held" output error of $(dv/dt) \times t_d$; dv/dt is the slew rate of the input signal and t_d is the switch delay. For this device, t_d is approximately 150 ns, giving a 4.5-mV error when sampling the zero crossing of a 5-V (peak) sine wave at 1 kHz ($dv/dt = A \times 2\pi f = 5 \times 2\pi \times 10^3$). The analog delay is the difference between input signal and capacitor voltage. It is determined by the RC product of the hold capacitor and the effective series resistance, which in this device is about 150 Ω .

LF198, LF198A, LF398, LF398A PRECISION SAMPLE-AND-HOLD AMPLIFIERS

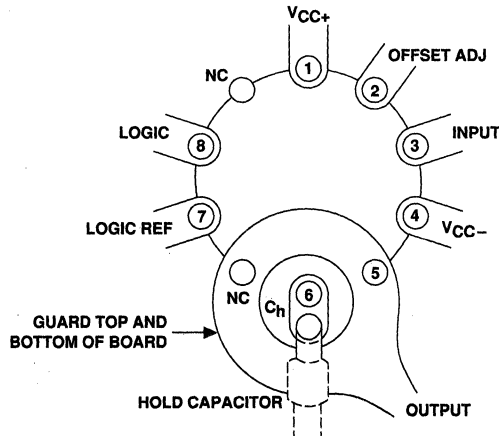
TYPICAL APPLICATION DATA

This analog delay with a 0.01- μF hold capacitor is $R \times C = 150 \times 10^{-8} = 1.5 \mu\text{s}$, or about ten times the delay of the switch. The held output is related in time to the input voltage *before* the hold command was given. The overall dynamic sampling error is the sum of the digital and analog errors. Figure 12 helps estimate these errors as a function of input slew rate and hold capacitor size.

Dynamic sampling error can be reduced by a factor of ten or more by inserting a delay in the logic input so that the "hold" command is delayed by an amount equal to the RC time constant of the device and the external hold capacitor. For a 0.01- μF hold capacitor and the 150- Ω resistor internal to this device, this value is 1.5 μs . A simple RC network can be used in front of the logic input for delays up to approximately 1 μs . Longer delays require the addition of a logic gate to speed up the rise time of the delayed signal. See "logic rise time" for further details.

hold step

Hold step is the small voltage step (after settling) seen at the output of a sample-and-hold amplifier when it is switched from the sample mode to the hold mode with a steady dc input. Hold step is typically the result of, or can be modeled as, a fixed quantity of charge transferred to the hold capacitor due to internal switching that occurs during the hold command. In the case of this device, that charge is approximately 5 pC, giving a hold step of 0.5 mV for a 0.01- μF hold capacitor and 5 mV for a 1000-pF hold capacitor ($V = Q/C$). Hold step is reasonably independent of logic amplitude if care is taken to minimize the stray capacitance between the logic input and the hold capacitor. With thoughtful layout, including the guarding technique shown in Figure 21, stray capacitance should be under 0.3 pF, limiting charge variations to less than 0.3 pC/V.



NOTE: Use 10-pin layout. The guard around C_h is tied to output.

FIGURE 22. GUARDING TECHNIQUE (BOTTOM VIEW)

Hold step varies slightly with analog input voltage (see Typical Characteristics). A typical unit changes at a rate of 0.4 pC/V. This slight variation manifests itself as a gain error when the amplifier is switched to the hold mode. With a 0.01- μF capacitor, the resulting gain error is $(0.4 \text{ pC/V}) / 0.01 \mu\text{F} = 0.004\%$. This gain error is in the opposite direction of dc (sample mode) gain error. When the hold capacitor has a high value, dc gain error dominates and gain is slightly below unity (0.002%). When the hold capacitor has a low value ($< 0.01 \mu\text{F}$), gain error induced by hold step dominates, and hold-mode gain is slightly above unity. Zeroing out hold step does not change the variation of hold step with regard to input voltage.

TYPICAL APPLICATION DATA

offset zeroing

A sample-and-hold amplifier has two distinct offset voltages. The first is the dc offset voltage of the amplifier while in the sample, or "tracking," mode. It is identical to the input offset voltage of any operational amplifier. The second offset voltage is the sum of the dc offset voltage plus a dynamic term called hold step. Hold step is a change in output voltage when the amplifier is switched from sample mode to hold mode with the input held steady. This second offset voltage is often called hold-mode offset voltage. It can be less than or much greater than the dc offset voltage, depending on the magnitude and sign of hold step.

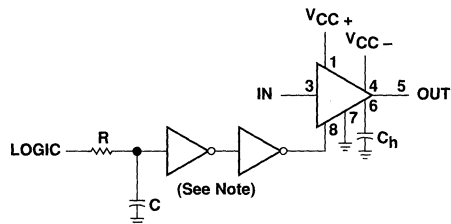
A fairly accurate model for hold step is a fixed charge injected into the hold capacitor by the switch turn-off circuitry. The magnitude of the charge is reasonably independent of logic input amplitude. The resulting change in the hold capacitor is Q/C_H . The charge Q is typically 5 pC, giving a 0.5-mV hold step with a 0.01- μ F hold capacitor. Since most sample-and-hold amplifiers are "used" (i.e., have their outputs read by an A-to-D converter), during the hold mode, hold-mode offset voltage is arguably more important than the sample-mode dc offset voltage.

Adjusting dc offset voltage is accomplished with a 1-k Ω low TC cermet potentiometer tied to V_{CC+} with 0.6 mA flowing through it and the wiper tied to pin 2. This allows pin 2 to be moved ± 300 mV around its nominal voltage (0.3 V below V_{CC+}). The offset voltage adjustment range is ± 9 mV, and the adjustment procedure nominally improves offset voltage drift when the dc offset voltage is reduced to zero. This offset method *can* be used to zero out hold-mode offset voltage, but at the expense of some induced offset voltage drift. Each millivolt of hold-step offset corrected by this method introduces 3.3 μ V/ $^{\circ}$ C drift. For 0.002- μ F hold capacitors or larger with hold step a few millivolts or less, this is a practical solution to hold-mode offset voltage. In precision, wide-temperature-range applications, or when C_H is less than 0.002 μ F, a separate hold-mode zeroing method should be used. The circuit shown in Figure 28, which uses a logic inverter and a 5-pF capacitor, is recommended.

logic fall time

Hold step is independent of logic input fall time only for fall times faster than 10 V/ μ s. For example, as logic fall time changes from 10 V/ μ s to 1 V/ μ s, hold step with a 0.01- μ F hold capacitor typically increases from 0.25 mV to 1 mV. (See Figure 7 for more data and refer to Figure 23.) If logic slew rate is not constant, use the value at the threshold point (1.5 V with respect to logic reference). An RC network will have a discharge slew rate of V_L/RC , where V_L is the logic threshold of the LF198. The delay generated by the network will be $RC \times \ln(V_{CC+}/V_L)$, where V_{CC+} is logic amplitude. For a 1- μ s delay with 5-V logic, an RC time constant of 0.8 μ s is required. This has a slew rate of 2 V/ μ s at threshold, which slightly degrades hold step. It is obvious that an RC delay network significantly longer than 1 μ s will have a large effect on hold step. If longer delays are required, they should be followed by several inverter stages or a Schmitt trigger to increase slew rate.

4
Special Functions

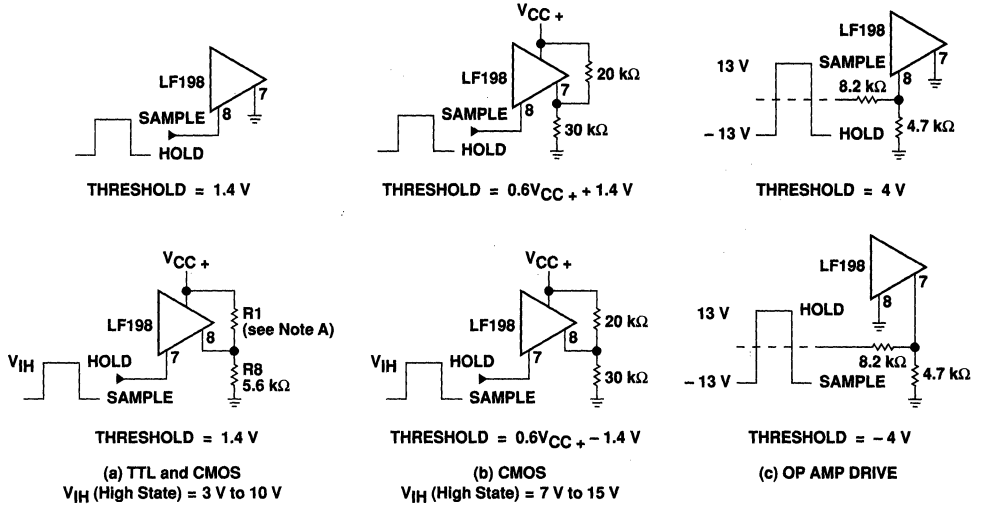


NOTE: Inverters may be eliminated for $RC \leq 3 \mu$ s.

FIGURE 23. ADDING DELAY TO LOGIC INPUT

LF198, LF198A, LF398, LF398A PRECISION SAMPLE-AND-HOLD AMPLIFIERS

TYPICAL APPLICATION DATA

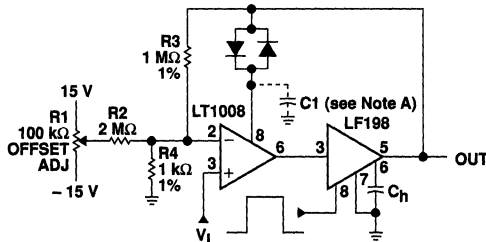


NOTES: A. Select R1 for 2.8 V at pin 8.
B. The logic input signal high level must be at least 2 V below the positive supply voltage of the device.

FIGURE 24. LOGIC INPUT CONFIGURATIONS

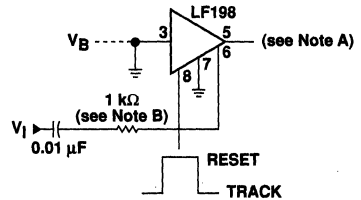
4

Special Functions



NOTE A: For lower gains, the LT1008 must be frequency compensated. Use approximately $(100/A_V)$ pF from comp2 to ground.

FIGURE 25. x1000 SAMPLE-AND-HOLD

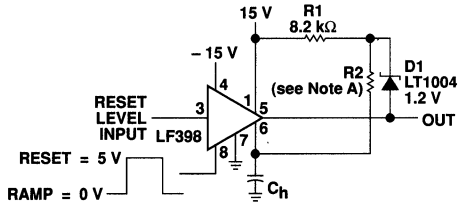


NOTES: A. $V_O = V_B + \Delta V_I$ (Hold mode).
B. This resistor protects input from surge currents but increases sample time. It can be eliminated if input is otherwise protected.
C. Output follows input in hold mode and resets to V_B in sample mode.

FIGURE 26. SAMPLE-AND-DIFFERENCE CIRCUIT

LF198, LF198A, LF398, LF398A PRECISION SAMPLE-AND-HOLD AMPLIFIERS

TYPICAL APPLICATION DATA



NOTE A: Select R2 for ramp rate $\Delta V/\Delta t = 1.2 V/(R_2)(C_h)$, $R \geq 10 \text{ k}\Omega$.

FIGURE 27. RAMP GENERATOR WITH VARIABLE RESET LEVEL

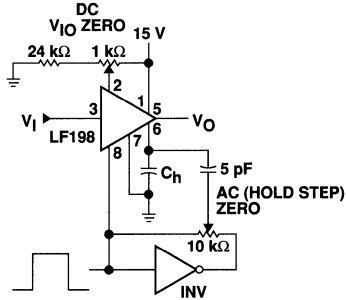
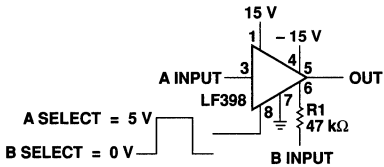
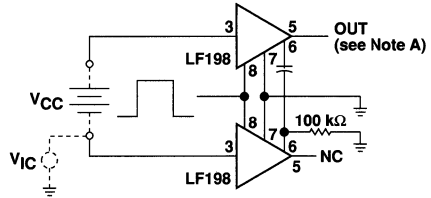


FIGURE 28. DC AND AC ZEROING



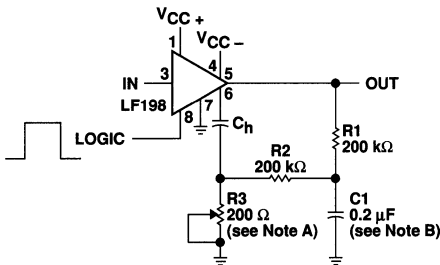
PARAMETER	A	B	UNIT
Gain	1 ± 0.02	1 ± 0.2	%
z_i	10^7	47	$\text{k}\Omega$
BW	≈ 1000	≈ 400	kHz
Crosstalk at 1 kHz	-90	-90	dB
Offset	≤ 6	≤ 75	mV

FIGURE 29. 2-CHANNEL SWITCH



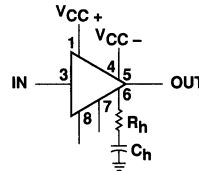
NOTE A: Output equals V_{CC} when in hold mode. Output equals $(V_{CC} + V_{IC})$ when in sample mode.

FIGURE 30. DIFFERENTIAL HOLD



NOTES: A. Adjust R3 for amplitude.
B. Select for time constant $C_1 = \tau/100 \text{ k}\Omega$.

FIGURE 31. CAPACITOR HYSTERESIS COMPENSATION



NOTE A: Select $(R_h)(C_h) \gg \frac{1}{2\pi f_{\text{MIN}}}$

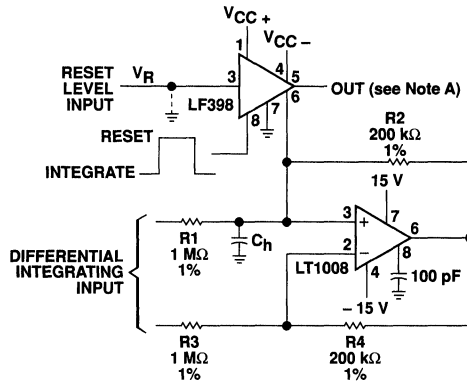
FIGURE 32. OUTPUT HOLDS AT AVERAGE OF SAMPLED INPUT

4

Special Functions

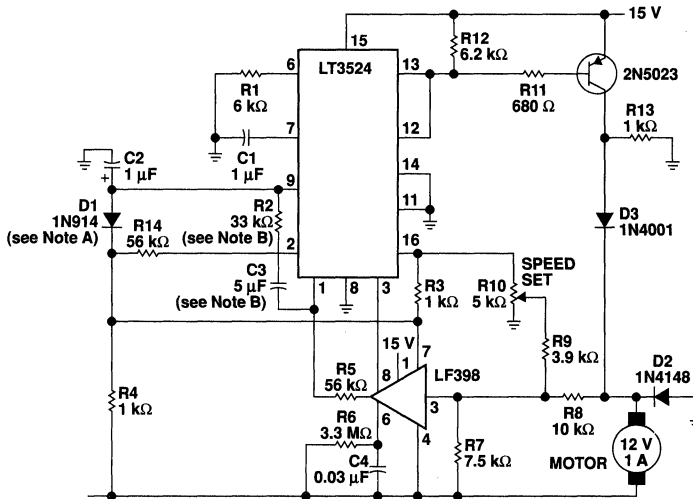
LF198, LF198A, LF398, LF398A PRECISION SAMPLE-AND-HOLD AMPLIFIERS

TYPICAL APPLICATION DATA



NOTE A: $V_O(\text{Hold mode}) = \left[\frac{1}{(R1)(C_h)} \int_0^t V_i dt \right] + V_R$

**FIGURE 35. INTEGRATOR WITH
PROGRAMMABLE RESET LEVEL**



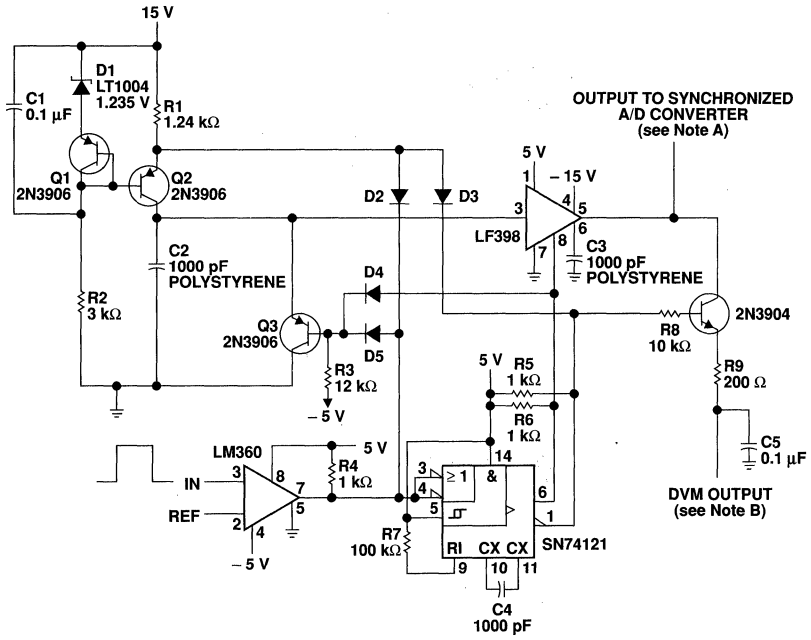
- NOTES: A. D1 is used for start-up. It limits duty cycle to approximately 75%.
 B. Select for optimum loop stability. C3 is nonpolarized.
 C. No tachometer is needed; back EMF of motor is sampled and used to control speed.

FIGURE 36. MOTOR SPEED CONTROLLER

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Special Functions

LF198, LF198A, LF398, LF398A PRECISION SAMPLE-AND-HOLD AMPLIFIERS

TYPICAL APPLICATION DATA



- NOTES: A. Read $\geq 1 \mu\text{s}$ after Q goes low.
 B. For repetitive pulses only. Increase C5 for $f \leq 10 \text{ kHz}$.
 C. D2-D5 1N914.

FIGURE 37. PULSE DURATION TO VOLTAGE CONVERTER

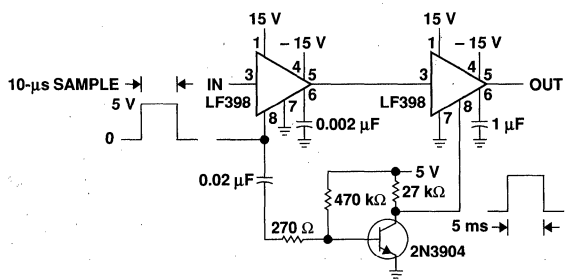


FIGURE 38. FAST-ACQUISITION, LOW-DROOP SAMPLE-AND-HOLD

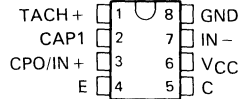
4
Special Functions

LM2907, LM2917 FREQUENCY-TO-VOLTAGE CONVERTERS

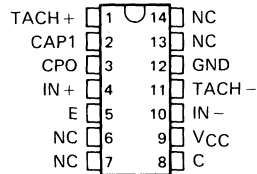
D3003, MARCH 1986—REVISED OCTOBER 1988

- Output Swings to Ground for Zero-Frequency Input
- Only One RC Network Provides Frequency Doubling for Low Ripple
- 8-Pin Versions Interface Directly to Variable-Reluctance Magnetic Pickups
- Uncommitted Collector and Emitter Outputs Provide 40-mA Sink or Source Current to Operate Relays, Solenoids, Meters, or LEDs
- Built-In Hysteresis for Noise Immunity
- Linearity Typically $\pm 0.3\%$
- 8-Pin Versions are Fully Protected from Damage Due to TACH Input Swing Above VCC and Below Ground

LM2907, LM2917 . . . D OR P PACKAGE
(TOP VIEW)



LM2907, LM2917 . . . D OR N PACKAGE
(TOP VIEW)



NC—No internal connection

applications

Over/under speed sensing
 Frequency-to-voltage conversion
 Speedometers
 Breaker-point dwell meters
 Hand-held tachometers
 Speed governors
 Cruise control
 Automotive door-lock control
 Clutch control
 Horn control
 Touch or sound switches

description

The LM2907 and LM2917 are monolithic frequency-to-voltage converters with an output circuit designed to operate a relay, lamp, or other load when the input frequency reaches or exceeds a selected rate. The converter (tachometer) section consists of a comparator driving a charge pump and offers frequency doubling for low ripple, full input protection in 8-pin versions, and an output swing to ground for a zero-frequency input. The output section consists of an operational amplifier, normally operating as a comparator, that drives an output transistor with both the collector and emitter floating. The circuit can either sink or source 40 mA of load current.

Two basic configurations of the devices are offered; an 8-pin version and a 14-pin version. The 8-pin versions have a ground-referenced tachometer input and an internal connection between the tachometer output and the operational amplifier input. The 8-pin version is well suited to single-speed or single-frequency switching or fully buffered frequency-to-voltage conversion applications. The more versatile 14-pin versions provide differential tachometer inputs and uncommitted operational amplifier inputs. In the 14-pin versions, the tachometer input can be floated and the operational amplifier becomes suitable for active filter conditioning of the tachometer output.

The LM2917 has an active shunt regulator connected across the power leads. The regulator clamps the supply voltage so that stable frequency-to-voltage and frequency-to-current conversions are possible with any supply voltage and a suitable resistor.

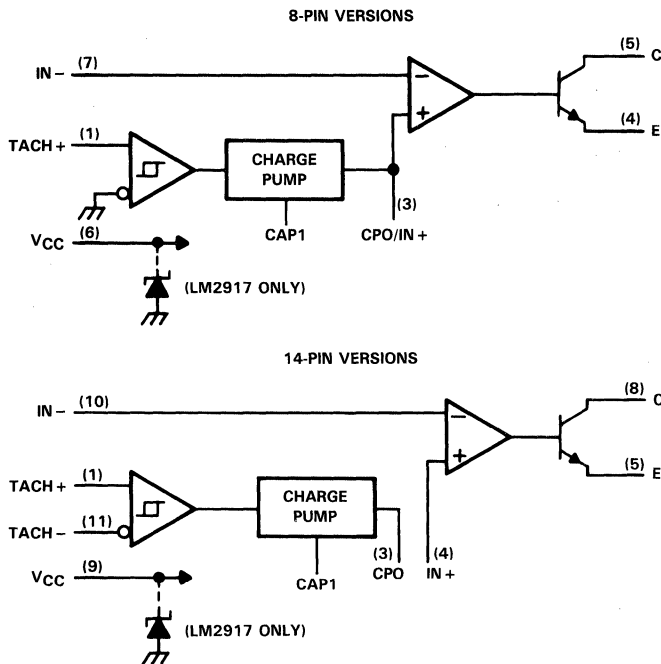
The LM2907 and LM2917 are designed for operation from -40°C to 85°C .

4

Special Functions

LM2907, LM2917 FREQUENCY-TO-VOLTAGE CONVERTERS

functional block diagrams



4

Special Functions

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, VCC: LM2907	28 V
Supply current, ICC: LM2917	25 mA
Collector-to-emitter voltage	28 V
Operational amplifier input voltage, IN+ and IN-	0 V to VCC
Tachometer input voltage: 8-pin version TACH+	0 V to 28 V
14-pin version TACH+ and TACH-	0 V to VCC
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	-40°C to 85°C
Storage temperature range	-65°C to 150°C
Lead temperature range 1,6 mm (1/16 inch) from case for 10 seconds	260°C

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C POWER RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 85°C POWER RATING
D (8 pins)	725 mW	5.8 mW/°C	377 mW
D (14 pins)	900 mW	7.2 mW/°C	468 mW
N	1000 mW	7.7 mW/°C	500 mW
P	900 mW	7.2 mW/°C	468 mW

LM2907, LM2917 FREQUENCY-TO-VOLTAGE CONVERTERS

electrical characteristics, $V_{CC} = 12\text{ V}$ (LM2907), $V_+ = 12\text{ V}$ through $470\ \Omega$ (LM2917), $T_A = 25\text{ }^\circ\text{C}$ converter (tachometer) section

PARAMETER		TEST CONDITIONS	LM2907			LM2917			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
V_T	Input threshold voltage	$V_I = 250\text{ mV}$, $f = 1\text{ kHz}$	± 10	± 15	± 40	± 10	± 15	± 40	mV	
V_{hys}	Input hysteresis (see Note 1)	$V_I = 250\text{ mV}$, $f = 1\text{ kHz}$	30			30			mV	
V_{IO}	Input offset voltage (see Note 1)	8-pin versions	$V_I = 250\text{ mV}$, $f = 1\text{ kHz}$			5	15	5	15	mV
		14-pin versions	$V_{ID} = 250\text{ mV}$, $f = 1\text{ kHz}$			3.5	10	3.5	10	
I_B	Input bias current	$V_I = \pm 50\text{ mV}$	0.1 1			0.1 1			μA	
V_{OH}	High-level output voltage, CAP1	V_I or $V_{ID} = 125\text{ mV}$	8.3			5.0			V	
V_{OL}	Low-level output voltage, CAP1	V_I or $V_{ID} = -125\text{ mV}$	2.3			1.2			V	
I_O	Output current, CAP1, CPO	CAP1 and CPO at 6 V	140	180	240				μA	
		CAP1 and CPO at 3.8 V				140	180	240		
	Leakage current, CPO	CAP1 open, CPO at 0 V, See Note 3	0.1			0.1			μA	
	Gain constant		0.9	1	1.1	0.9	1	1.1		
	Nonlinearity (see Note 2)	$f = 1\text{ kHz}$, 5 kHz, or 10 kHz	0.3 ± 1			0.3 ± 1			%	

output section

PARAMETER		TEST CONDITIONS	LM2907			LM2917			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO}	Input offset voltage	$V_I = 6\text{ V}$, See Note 3	3 10						mV
		$V_I = 3.8\text{ V}$, See Note 3				3 10			
I_B	Bias current	$V_I = 6\text{ V}$	50 500						nA
		$V_I = 3.8\text{ V}$				50 500			
A_V	Voltage amplification		200			200			V/mV
I_C	Collector output (sink) current	$V_C = 1\text{ V}$, $V_E = 0$	40	50		40	50		mA
I_E	Emitter output (source) current	$V_C = V_{CC}$, $V_E = V_{CC} - 2$	-10			-10			mA
$V_{CE(sat)}$	Collector-emitter saturation voltage	$I_C = 5\text{ mA}$	0.1 0.5			0.1 0.5			V
		$I_C = 20\text{ mA}$	1			1			
		$I_C = 50\text{ mA}$	1 1.5			1 1.5			

- NOTES: 1. Hysteresis is the algebraic difference $V_{T+} - V_{T-}$; offset voltage is the difference in magnitudes $|V_{T+}| - |V_{T-}|$. See parameter measurement information test circuits.
2. Nonlinearity is defined as the deviation of V_O at CPO for $f = 5\text{ kHz}$ from a straight line defined by the V_O at 1 kHz and V_O at 10 kHz, with $C1 = 1000\text{ pF}$, $R1 = 68\text{ k}\Omega$, $C2 = 0.22\text{ }\mu\text{F}$.
3. Pin 2 must be bypassed with a $0.001\text{-}\mu\text{F}$ capacitor to prevent oscillation for these tests.

4

Special Functions

LM2907, LM2917 FREQUENCY-TO-VOLTAGE CONVERTERS

electrical characteristics

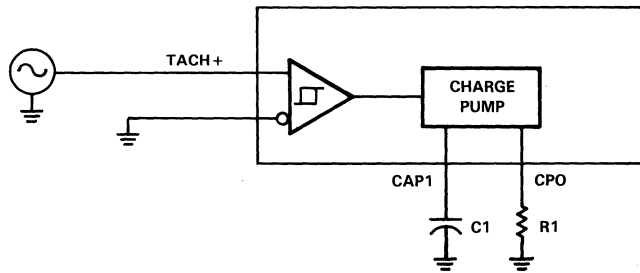
zener regulator (LM2917 only) $V_+ = 12\text{ V}$ through $470\ \Omega$, $T_A = 25^\circ\text{C}$

PARAMETER		MIN	TYP	MAX	UNIT
V_{CC}	Regulated supply voltage		7.56		V
r_s	Series resistance		10.5	15	Ω
αV_{CC}	Temperature coefficient of regulated supply voltage		1		mV/ $^\circ\text{C}$

total device (LM2907 only) $V_{CC} = 12\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER		MIN	TYP	MAX	UNIT
I_{CC}	Supply current		3.8	6	mA

PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT

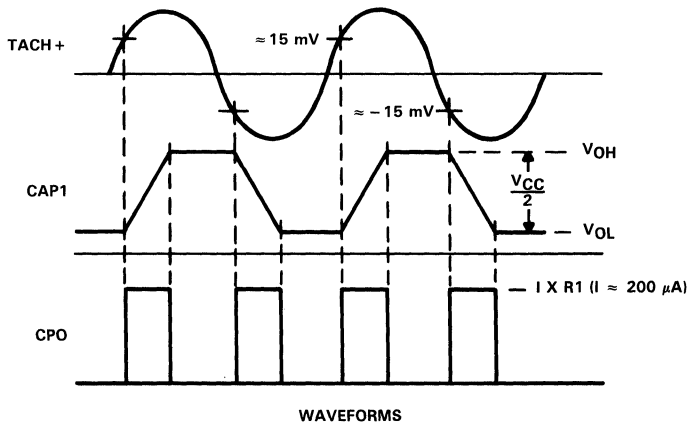


FIGURE 1. TEST CIRCUIT AND WAVEFORMS

4

Special Functions

TYPICAL APPLICATION DATA

The LM2907 and LM2917 frequency-to-voltage converter circuits are designed for maximum versatility with a minimum of external parts. The first stage of these devices is a differential comparator. The single-input 8-pin versions have one input grounded so that an input signal must swing above and below ground and exceed the input thresholds to produce an output. This version is specifically for magnetic variable-reluctance pickups, which typically provide a single-ended ac output. These single-ended inputs are fully protected against voltage swings to $\pm 28\text{ V}$, which are easily attained by these types of pickups.

The differential-input 14-pin versions provide the option of setting the input reference level and still having hysteresis around that level to provide excellent noise rejection in any application. The input protection is removed in the 14-pin versions. Therefore, neither of the differential inputs should exceed the limits of the supply voltage. An input must not go below ground without a resistance in the lead to limit the current that will flow in the epi-substrate diode. The charge pump circuit that follows the input stage produces a dc output voltage proportional to the input frequency. The charge pump circuit (see Figure 2) consists of a timing capacitor (C1), an output resistor (R1), and an integrating or filter capacitor (C2). When the input changes state (due to a suitable zero crossing or differential voltage on the input), the timing capacitor is either charged or discharged linearly with a constant current of $200\ \mu\text{A}$ through CAP1 between two voltages whose difference is $V_{CC}/2$. Within one-half cycle of the input frequency or a time equal to $1/2f$, the change in charge on C1 is equal to $(V_{CC}/2)C1$. The average amount of current pumped into or out of the capacitor is:

$$\text{CAP1 current (average)} = \frac{Q}{T} = C1 \cdot \frac{V_{CC}}{2} \cdot 2f = V_{CC} \cdot f \cdot C1$$

The output of the charge pump accurately mirrors the CAP1 current into the load resistor (R1) connected to CPO. If the pulses of current are integrated with a filter capacitor, the output voltage is the average CAP1 current times R1, and the total equation becomes:

$$V_O = V_{CC} \cdot f \cdot C1 \cdot R1 \cdot K$$

where K is the gain factor, which is typically 1.

The size of C2 is dependent only on the amount of ripple allowable and the required response time.

selection of R1, C1, and C2

To achieve optimum performance, there are some limitations to be considered in the selection of R1 and C1. The timing capacitor controls the RC time and provides internal compensation for the charge pump circuit. For very accurate operation it should be 100 pF or greater. Smaller values, especially at lower temperatures, can cause an error current through R1. $V_O/R1$ must be less than or equal to the output current at CPO, which is fixed typically at $180\ \mu\text{A}$. If R1 is too large it becomes a significant fraction of the output impedance at CPO, which degrades the linearity. In addition, ripple voltage must be considered when selecting R1. The size of C2 is directly affected by the size of R1. An expression that describes the ripple content at CPO is:

$$V_{\text{ripple}} = \frac{V_{CC}}{2} \cdot \frac{C1}{C2} \cdot (1 - V_{CC} \cdot f \cdot \frac{C1}{200}) \quad \text{volts peak-to-peak}$$

where

C1 and C2 are in farads

V_{CC} is in volts

f is in hertz.

LM2907, LM2917 FREQUENCY-TO-VOLTAGE CONVERTERS

TYPICAL APPLICATION DATA

R1 cannot be chosen independently of ripple because response time or the time it takes V_O to stabilize at a new level increases as the size of C2 increases. A compromise between ripple, response time, and linearity must be chosen carefully. As a final consideration, the maximum attainable input frequency is determined by V_{CC} , C1, and I_{cap} (current through CAP1).

$$f_{max} = \frac{I_{cap}}{C1 \cdot V_{CC}} \quad \text{hertz}$$

where

I_{cap} is typically 200 μA

C1 is in farads

V_{CC} is in volts.

zener regulator options (LM2917)

For those applications in which an output voltage or current must be obtained independently of supply voltage variations, the LM2917 can be used. The most important factor in selecting a dropping resistor for the unregulated supply is that the frequency-to-voltage converter circuit and the operational amplifier alone require approximately 3 mA at the voltage level set by the zener diode. At low supply voltages there must be some current flowing in the resistor above the 3 mA circuit current to operate the regulator. As an example, if the supply voltage varies between 9 and 16 V, a resistance of 470 Ω will minimize the zener voltage variation to typically 160 mV. If the resistance goes under 400 Ω or above 600 Ω , the zener variation quickly rises above 200 mV for the same input variation.

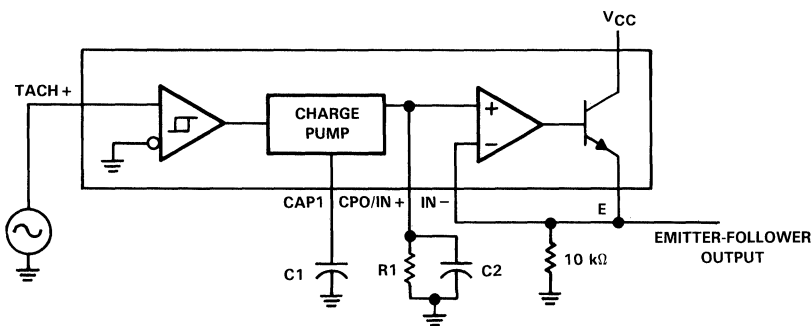


FIGURE 2. MINIMUM-COMPONENT TACHOMETER

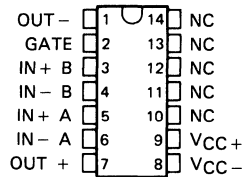
MC1445

GATE-CONTROLLED 2-CHANNEL-INPUT VIDEO AMPLIFIER

D2572, JANUARY 1980—REVISED APRIL 1988

- Differential Inputs and Outputs
- Channel Select Time . . . 20 ns Typ
- Bandwidth Typically 50 MHz
- 16-dB Minimum Gain
- Common-Mode Rejection Typically 85 dB
- Broadband Noise Typically 25 μ V

J OR N DUAL-IN-LINE PACKAGE
(TOP VIEW)



NC—No internal connection

FUNCTION TABLE

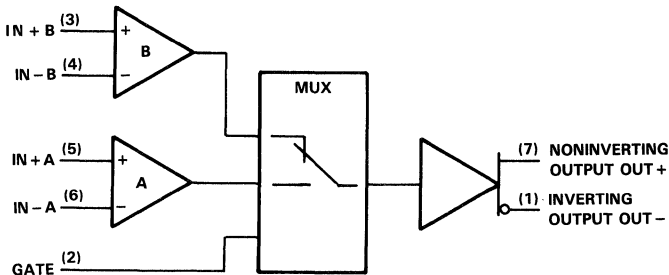
GATE INPUT	SELECT
H	Channel A
L	Channel B

description

The MC1445 is a general-purpose, gated, dual-channel wideband amplifier designed for use in video-signal mixing and switching. Channel selection is accomplished by control of the voltage level at the gate. A high logic level selects channel A; a low logic level selects channel B. The unselected channel will have a gain of one or less.

The MC1445 is characterized for operation from 0°C to 75°C.

block diagram



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	12 V
Supply voltage, V_{CC-} (see Note 1)	-12 V
Differential input voltage, V_{ID} (see Note 2)	± 5 V
Output current, I_O	± 25 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature	0°C to 75°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: N package	260°C

- NOTES: 1. Voltage values, except differential input voltage, are with respect to the midpoint of V_{CC+} and V_{CC-} .
 2. Differential input voltages are measured at a noninverting input terminal with respect to the appropriate inverting input terminal.

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Special Functions

MC1445 GATE-CONTROLLED 2-CHANNEL-INPUT VIDEO AMPLIFIER

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C POWER RATING	DERATING FACTOR	DERATE ABOVE T _A	T _A = 75°C POWER RATING
J	626 mW	8.2 mW/°C	74°C	615 mW
N	625 mW	N/A	N/A	625 mW

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V _{CC+}		5	8	V
Supply voltage, V _{CC-}		-5	-8	V
Operating free-air temperature range, T _A	0		75	°C

electrical characteristics at V_{CC+} = 5 V, V_{CC-} = -5 V, T_A = 25°C

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
A _{VS}	Large-signal single-ended voltage amplification f = 125 kHz, V _i = 20 mV	16	19.5	23	dB
BW	Bandwidth V _i = 20 mV		50		MHz
V _{IO}	Input offset voltage			7.5	mV
I _{IO}	Input offset current		2		μA
I _{IB}	Input bias current		.15	30	μA
V _{ICR}	Common-mode voltage range		±2.5		V
V _{OQ}	Quiescent output voltage		0.1		V
ΔV _{OQ}	Change in quiescent output voltage Gate input change from 5 V to 0 V		±15		mV
V _{OPP}	Maximum peak-to-peak output voltage swing f = 125 kHz, R _L = 1 kΩ	1.5	2.5		V
z _i	Input impedance f = 125 kHz	3	10		kΩ
z _o	Output impedance f = 50 kHz		25		Ω
CMRR	Common-mode rejection ratio f = 50 kHz		85		dB
V _n	Broadband equivalent input noise voltage BW = 5 Hz to 10 MHz, R _S = 50 Ω		25		μV
V _{TH}	High-level gate threshold voltage A _{VS(A)} ≥ 16 dB, A _{VS(B)} ≤ 0 dB		1.3	3	V
V _{TL}	Low-level gate threshold voltage A _{VS(B)} ≥ 16 dB, A _{VS(A)} ≤ 0 dB	0.2	0.4		V
I _{IH}	High-level gate current V _I = 5 V			4	μA
I _{IL}	Low-level gate current V _I = 0			4	mA
t _{PLH}	Propagation delay time, low-to-high-level output ΔV _I = 20 mV, 50% to 50%		6.5		ns
t _{PHL}	Propagation delay time, high-to-low-level output ΔV _I = 20 mV, 50% to 50%		6.3		ns
t _{TLH}	Transition time, low-to-high-level output ΔV _I = 20 mV, 10% to 90%		6.5		ns
t _{THL}	Transition time, high-to-low-level output ΔV _I = 20 mV, 10% to 90%		7		ns
I _{CC+}	Supply current from V _{CC+} No load, No signal		7	15	mA
I _{CC-}	Supply current from V _{CC-} No load, No signal		-7	-15	mA
P _D	Power dissipation No load, No signal		70	150	mW

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Special Functions

MC3470, MC3470A FLOPPY DISK READ-AMPLIFIER SYSTEMS

D2759, NOVEMBER 1983—REVISED FEBRUARY 1988

- Combines All Read-Amplifier Active Circuitry into One Monolithic Circuit
- Peak Shift . . . 2% Max (MC3470A)
- Designed to be Interchangeable with Motorola MC3470

description

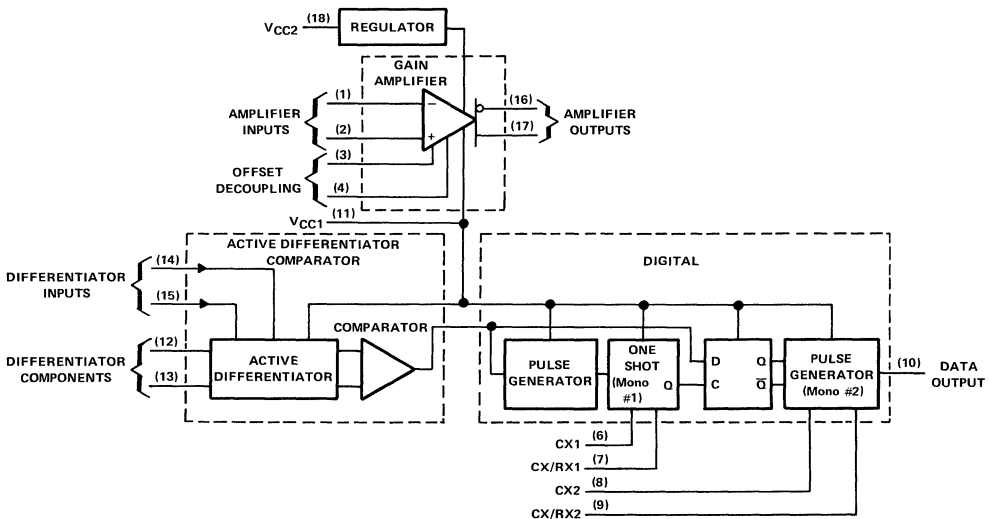
The MC3470 and MC3470A are monolithic read-amplifier systems each containing all the active circuitry necessary for obtaining digital information from floppy disk storage. They are designed to accept the ac differential signal from the magnetic head and produce a digital output pulse corresponding to each peak of the input signal. The gain stage amplifies the input waveform and applies it to an external filter network, enabling the active differentiator and time domain filter to produce the desired output.

The MC3470 and MC3470A are characterized for operation from 0°C to 70°C.

N DUAL-IN-LINE PACKAGE (TOP VIEW)

AMPLIFIER	{	1	18	V _{CC2}
INPUTS	{	2	17	AMPLIFIER
OFFSET	{	3	16	OUTPUTS
DECOUPLING	{	4	15	DIFFERENTIATOR
GND	{	5	14	INPUTS
CX1	{	6	13	DIFFERENTIATOR
CX/RX1	{	7	12	COMPONENTS
CX2	{	8	11	V _{CC1}
CX/RX2	{	9	10	DATA OUTPUT

functional block diagram



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Special Functions

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MC3470, MC3470A FLOPPY DISK READ-AMPLIFIER SYSTEMS

absolute maximum ratings over operating temperature range (unless otherwise noted)

Supply voltage, V_{CC1} (see Note 1)	7 V
Supply voltage, V_{CC2}	16 V
Input voltage range (amplifier inputs)	-0.2 V to 7 V
Output voltage, V_O (data output)	-0.2 V to 7 V
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C

NOTE 1: All voltage values are with respect to network ground terminal.

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage V_{CC1}	4.75	5	5.25	V
Supply voltage V_{CC2}	10	12	14	V
Timing capacitor CX1 (see Note 2)	150		680	pF
Timing capacitor CX2	100		800	pF
Timing resistors RX1 and RX2	1.5		10	k Ω
Timing of digital section	Monostable no. 1		500	4000
	Monostable no. 2		150	1000
Operating free-air temperature, T_A	0		70	°C

NOTE 2: To minimize current transients, CX1 should be kept as small as convenient.

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Special Functions

MC3470, MC3470A FLOPPY DISK READ-AMPLIFIER SYSTEMS

electrical characteristics over recommended ranges of supply voltages and operating free-air temperature (unless otherwise noted)

gain amplifier section

PARAMETER		TEST CONDITIONS		MIN	TYP [†]	MAX	UNIT
A _{VD}	Differential voltage amplification	MC3470	V _{id} = 5 mV rms, f = 200 kHz	80	100	120	V/V
		MC3470A		100	110	130	
I _{IB}	Input bias current				-10	-25	μA
V _{ICR}	Common-mode input voltage range		THD ≤ 5%	-0.1	to	1.5	V
V _{IDR}	Differential input voltage range		THD ≤ 5%	±25			mV
V _{OPP}	Peak-to-peak differential output voltage			3	4		V
V _{OC}	Common-mode output voltage	V _I = 0,	V _{ID} = 0		3		V
V _{OD}	Differential output offset voltage	V _I = 0, T _A = 25°C	V _{ID} = 0,			0.4	V
I _{OS}	Short-circuit output current (each amplifier output)	Output shorted to ground			-8		mA
		Output shorted to V _{CC1}		2.8	4		
r _i	Small-signal input resistance	T _A = 25°C		100	250		kΩ
r _o	Small-signal output resistance (single-ended)	V _{CC1} = 5 V, T _A = 25°C	V _{CC2} = 12 V,		15		Ω
BW	Bandwidth (3 dB)	V _{id} = 2 mV rms, V _{CC2} = 12 V,	V _{CC1} = 5 V, T _A = 25°C		5		MHz
CMRR	Common-mode rejection ratio	V _{CC1} = 5 V, A _{VD} = 40 dB, T _A = 25°C	V _I PP = 200 mV, f = 100 kHz,		50		dB
k _{SVR}	Supply voltage rejection ratio	A _{VD} = 40 dB, T _A = 25°C	V _{CC1} = 5 ± 0.25 V, V _{CC2} = 12 V		50		dB
			V _{CC1} = 5 V, V _{CC2} = 12 ± 2 V		60		
V _n	Equivalent input noise voltage	BW = 10 Hz to 1 MHz, T _A = 25°C			15		μV

[†]All typical values are at V_{CC1} = 5 V, V_{CC2} = 12 V, T_A = 25°C.

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Special Functions

MC3470, MC3470A

FLOPPY DISK READ-AMPLIFIER SYSTEMS

active-differentiator section

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
I_{sink}	Sink current at pins 12 and 13	$V_{\text{OD}} = V_{\text{CC1}}$		1	1.4		mA
Peak shift	MC3470	$V_{\text{CC1}} = 5 \text{ V}$,	$V_{\text{CC2}} = 12 \text{ V}$,			5%	
	MC3470A	$V_{\text{IDPP}} = 1 \text{ V}$,	$f = 250 \text{ kHz}$,			2%	
		$I_{\text{cap}} = 500 \mu\text{A}$, See Figure 1					
r_{id}	Differential input resistance				30		$k\Omega$
r_{od}	Differential output resistance				40		Ω

digital section

PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
V_{OH}	High-level output voltage (pin 10)	$V_{\text{CC1}} = 4.75 \text{ V}$,	$V_{\text{CC2}} = 12 \text{ V}$,	2.7			V
		$I_{\text{OH}} = -0.4 \text{ mA}$					
V_{OL}	Low-level output voltage (pin 10)	$V_{\text{CC1}} = 4.75 \text{ V}$,	$V_{\text{CC2}} = 12 \text{ V}$,			0.5	V
		$I_{\text{OL}} = 8 \text{ mA}$					
I_{CC1}	Supply current from V_{CC1}	$V_{\text{CC1}} = 5.25 \text{ V}$			35	50	mA
I_{CC2}	Supply current from V_{CC2}	$V_{\text{CC2}} = 14 \text{ V}$			4.5	10	mA

timing characteristics over recommended ranges of supply voltages and operating free-air temperature (unless otherwise noted) (see Figure 2)

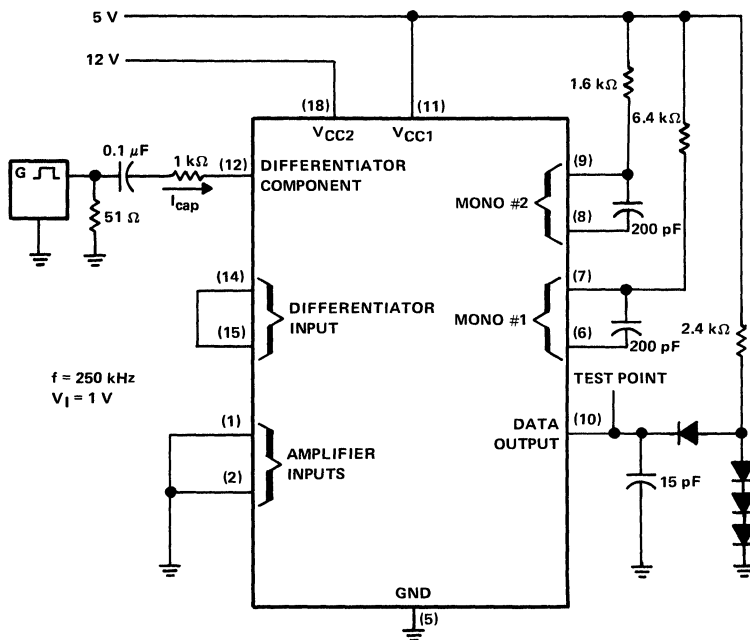
PARAMETER		TEST CONDITIONS		MIN	TYP†	MAX	UNIT
t_{r}	Rise time (pin 10)					20	ns
t_{f}	Fall time (pin 10)					25	ns
	Timing accuracy of monostable no. 1 compared to $0.625 \text{ RX1} \cdot \text{CX1} + 200 \text{ ns}$	$\text{RX1} = 1.5 \text{ k}\Omega$ to $10 \text{ k}\Omega$, $\text{CX1} = 150 \text{ pF}$ to 680 pF		85%		115%	
	Timing accuracy of monostable no. 2 compared to $0.625 \text{ RX2} \cdot \text{CX2}$	$\text{RX2} = 1.5 \text{ k}\Omega$ to $10 \text{ k}\Omega$, $\text{CX2} = 100 \text{ pF}$ to 800 pF		85%		115%	

†All typical values are at $V_{\text{CC1}} = 5 \text{ V}$, $V_{\text{CC2}} = 12 \text{ V}$, $T_{\text{A}} = 25^{\circ}\text{C}$.

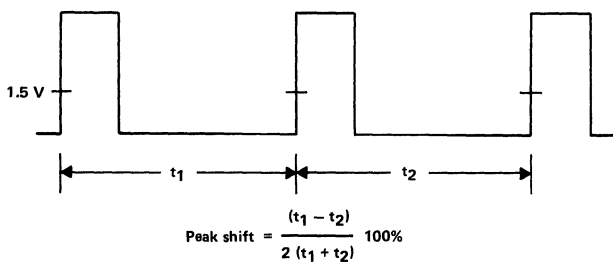
4

Special Functions

PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

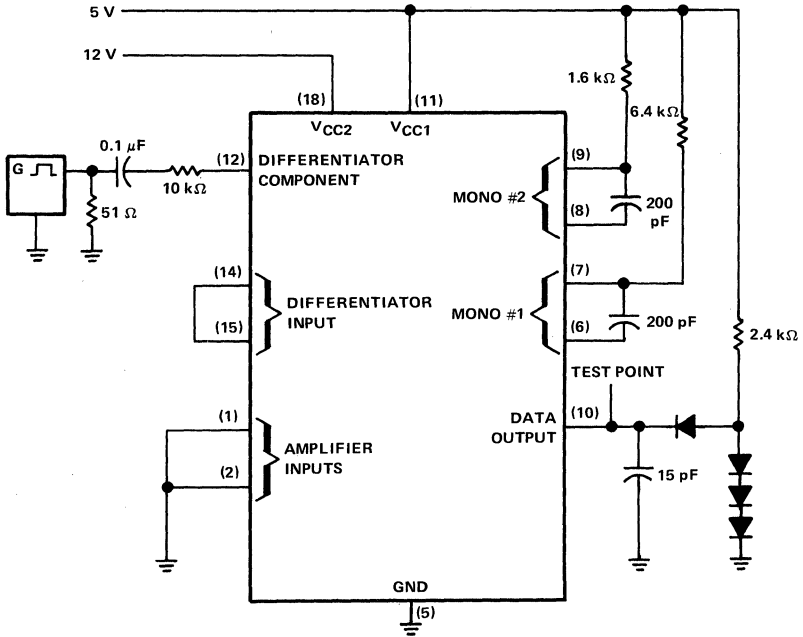
FIGURE 1. PEAK SHIFT

4

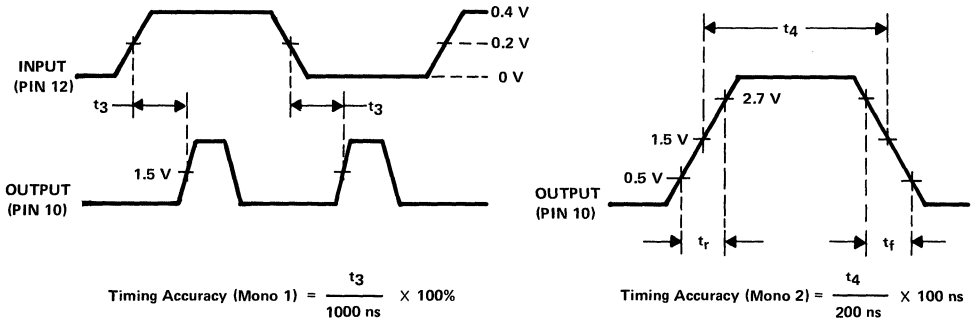
Special Functions

**MC3470, MC3470A
FLOPPY DISK READ-AMPLIFIER SYSTEMS**

PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

FIGURE 2. TIMING ACCURACY

4

Special Functions

TYPICAL CHARACTERISTICS

NORMALIZED POWER SUPPLY CURRENT
vs
FREE-AIR TEMPERATURE

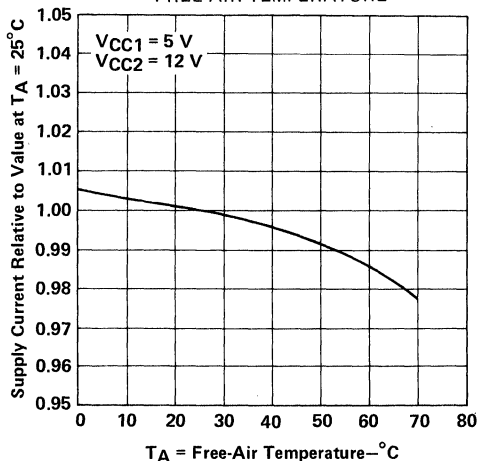


FIGURE 3

NORMALIZED VOLTAGE AMPLIFICATION
vs
FREE-AIR TEMPERATURE

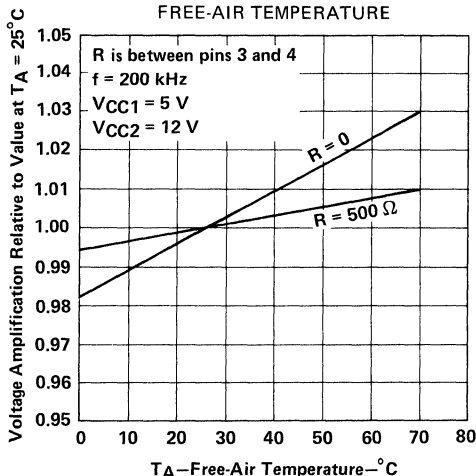


FIGURE 4

NORMALIZED TIME DELAY (PIN 12 TO PIN 10)
vs
FREE-AIR TEMPERATURE

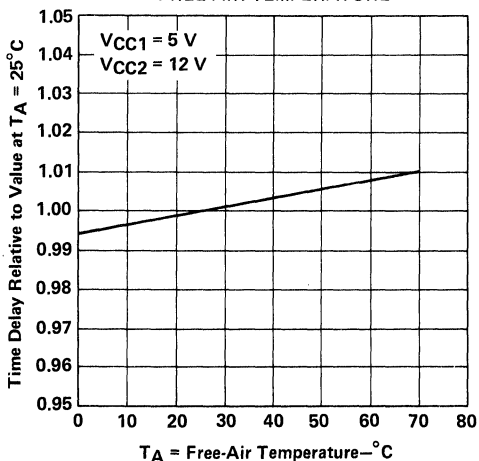


FIGURE 5

VOLTAGE AMPLIFICATION (NORMALIZED)
and PHASE SHIFT
vs
FREQUENCY

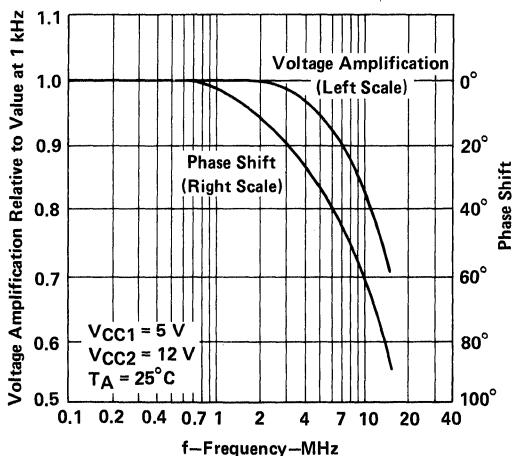


FIGURE 6

**MC3470, MC3470A
FLOPPY DISK READ-AMPLIFIER SYSTEMS**

TYPICAL APPLICATION INFORMATION

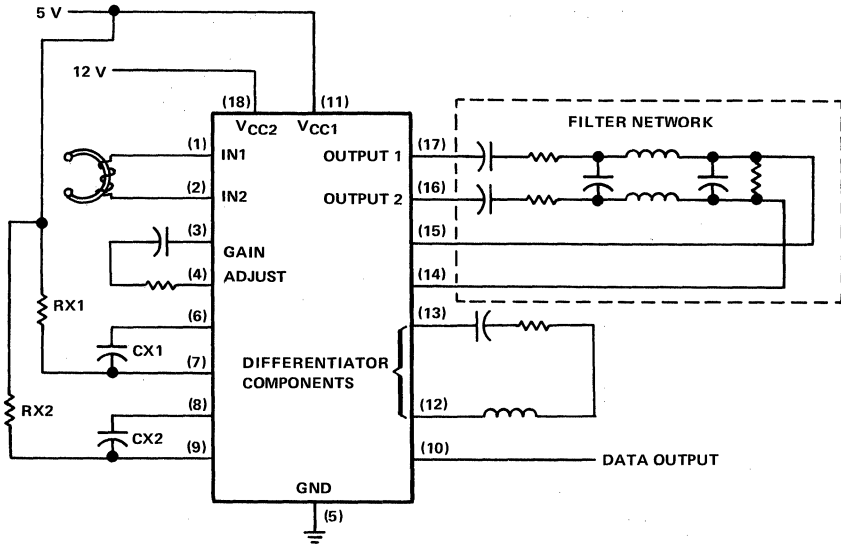


FIGURE 7

4

Special Functions

SE555, SE555C, SA555, NE555 PRECISION TIMERS

D1669, SEPTEMBER 1973—REVISED OCTOBER 1988

- Timing from Microseconds to Hours
- Astable or Monostable Operation
- Adjustable Duty Cycle
- TTL-Compatible Output Can Sink or Source Up to 200 mA
- Functionally Interchangeable with the Signetics SE555, SE555C, SA555, NE555; Have Same Pinout

SE555C FROM TI IS NOT RECOMMENDED FOR NEW DESIGNS

description

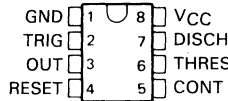
These devices are monolithic timing circuits capable of producing accurate time delays or oscillation. In the time-delay or monostable mode of operation, the timed interval is controlled by a single external resistor and capacitor network. In the astable mode of operation, the frequency and duty cycle may be independently controlled with two external resistors and a single external capacitor.

The threshold and trigger levels are normally two-thirds and one-third, respectively, of V_{CC} . These levels can be altered by use of the control voltage terminal. When the trigger input falls below the trigger level, the flip-flop is set and the output goes high. If the trigger input is above the trigger level and the threshold input is above the threshold level, the flip-flop is reset and the output is low. The reset input can override all other inputs and can be used to initiate a new timing cycle. When the reset input goes low, the flip-flop is reset and the output goes low. Whenever the output is low, a low-impedance path is provided between the discharge terminal and ground.

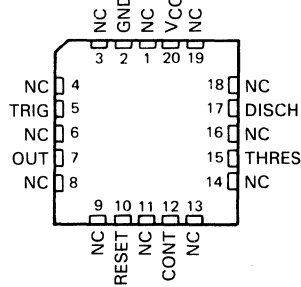
The output circuit is capable of sinking or sourcing current up to 200 mA. Operation is specified for supplies of 5 to 15 V. With a 5-V supply, output levels are compatible with TTL inputs.

The SE555 and SE555C are characterized for operation over the full military range of -55°C to 125°C . The SA555 is characterized for operation from -40°C to 85°C , and the NE555 is characterized for operation from 0°C to 70°C .

SE555, SE555C . . . JG PACKAGE
SA555, NE555 . . . D, JG, OR P PACKAGE
(TOP VIEW)

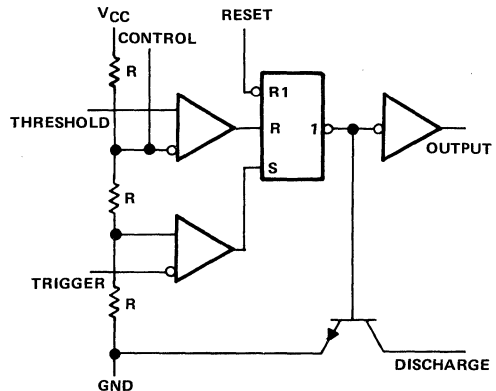


SE555, SE555C . . . FK PACKAGE
(TOP VIEW)



NC—No internal connection

functional block diagram



Reset can override Trigger, which can override Threshold.

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Special Functions

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SE555, SE555C, SA555, NE555 PRECISION TIMERS

AVAILABLE OPTIONS

T _A RANGE	V _{thres} MAX V _{CC} = 15 V	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	11.2 V	NE555D		NE555JG	NE555P
-40°C to 85°C	11.2 V	SA555D		SA555JG	SA555P
-55°C to 125°C	10.6 V 11.2 V		SE555FK SE555CFK	SE555JG SE555CJG	

The D package is available taped and reeled. Add the suffix R to the device type (e.g., NE555DR).

FUNCTION TABLE

RESET	TRIGGER VOLTAGE†	THRESHOLD VOLTAGE†	OUTPUT	DISCHARGE SWITCH
Low	Irrelevant	Irrelevant	Low	On
High	< 1/3 V _{DD}	Irrelevant	High	Off
High	> 1/3 V _{DD}	> 2/3 V _{DD}	Low	On
High	> 1/3 V _{DD}	< 2/3 V _{DD}	As previously established	

†Voltage levels shown are nominal.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V _{CC} (See Note 1)	18 V
Input voltage (control, reset, threshold, and trigger)	V _{CC}
Output current	±225 mA
Continuous total dissipation	see Dissipation Rating Table
Operating free-air temperature range: SE555, SE555C	-55°C to 125°C
SA555	-40°C to 85°C
NE555	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C

NOTE 1: All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C	DERATING	T _A = 70°C	T _A = 85°C	T _A = 125°C
	POWER RATING	FACTOR ABOVE T _A = 25°C	POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	N/A
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG (SE555, SE555C)	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
JG (SA555, NE555)	825 mW	6.6 mW/°C	528 mW	429 mW	N/A
P	1000 mW	8.0 mW/°C	640 mW	520 mW	N/A

recommended operating conditions

	SE555		SE555C		SA555		NE555		UNIT
	MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, V _{CC}	4.5	18	4.5	16	4.5	16	4.5	16	V
Input voltage (control, reset, threshold, and trigger)	V _{CC}		V _{CC}		V _{CC}		V _{CC}		V
Output current	±200		±200		±200		±200		mA
Operating free-air temperature, T _A	-55	125	-55	125	-40	85	0	70	°C

4

Special Functions

**SE555, SE555C, SA555, NE555
PRECISION TIMERS**

electrical characteristics at 25 °C free-air temperature, V_{CC} = 5 V to 15 V (unless otherwise noted)

PARAMETER	TEST CONDITIONS		SE555			SE555C, SA555, NE555			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
Threshold voltage level	V _{CC} = 15 V		9.4	10	10.6	8.8	10	11.2	V
	V _{CC} = 5 V		2.7	3.3	4	2.4	3.3	4.2	
Threshold current (see Note 2)			30			30			nA
Trigger voltage level	V _{CC} = 15 V		4.8	5	5.2	4.5	5	5.6	V
	V _{CC} = 5 V		1.45	1.67	1.9	1.1	1.67	2.2	
Trigger current	Trigger at 0 V		0.5			0.5			μA
Reset voltage level			0.3	0.7	1	0.3	0.7	1	V
Reset current	Reset at V _{CC}		0.1			0.1			mA
	Reset at 0 V		-0.4			-1			
Discharge switch off-state current			20			20			nA
Control voltage (open circuit)	V _{CC} = 15 V		9.6	10	10.4	9	10	11	V
	V _{CC} = 5 V		2.9	3.3	3.8	2.6	3.3	4	
Low-level output voltage	V _{CC} = 15 V	I _{OL} = 10 mA	0.1		0.15	0.1		0.25	V
		I _{OL} = 50 mA	0.4		0.5	0.4		0.75	
		I _{OL} = 100 mA	2		2.2	2		2.5	
		I _{OL} = 200 mA	2.5			2.5			
	V _{CC} = 5 V	I _{OL} = 5 mA	0.1		0.2	0.1		0.35	
		I _{OL} = 8 mA	0.15		0.25	0.15		0.4	
High-level output voltage	V _{CC} = 15 V	I _{OH} = -100 mA	13	13.3		12.75	13.3	V	
		I _{OH} = -200 mA	12.5			12.5			
	V _{CC} = 5 V	I _{OH} = -100 mA	3	3.3		2.75	3.3		
Supply current	Output low, No load	V _{CC} = 15 V	10		12	10		15	mA
		V _{CC} = 5 V	3		5	3		6	
	Output high, No load	V _{CC} = 15 V	9		10	9		13	
		V _{CC} = 5 V	2		4	2		5	

NOTE 2: This parameter influences the maximum value of the timing resistors R_A and R_B in the circuit of Figure 12. For example, when V_{CC} = 5 V, the maximum value is R = R_A + R_B ≈ 3.4 MΩ, and for V_{CC} = 15 V, the maximum value is 10 MΩ.

operating characteristics, V_{CC} = 5 V and 15 V

PARAMETER		TEST CONDITIONS†	SE555			SE555C, SA555, NE555			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
Initial error of timing interval‡	Each timer, monostable§	T _A = 25 °C	0.5	1.5		1	3	%	
	Each timer, astable¶		1.5		2.25				
Temperature coefficient of timing interval	Each timer, monostable§	T _A = MIN to MAX	30	100		50		ppm/°C	
	Each timer, astable¶		90		150				
Supply voltage sensitivity of timing interval	Each timer, monostable§	T _A = 25 °C	0.05	0.2		0.1	0.5	% / V	
	Each timer, astable¶		0.15		0.3				
Output pulse rise time	C _L = 15 pF,		100	200		100	300	ns	
Output pulse fall time	T _A = 25 °C		100	200		100	300		

†For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡Timing interval error is defined as the difference between the measured value and the average value of a random sample from each process run.

§Values specified are for a device in a monostable circuit similar to Figure 9, with component values as follow: R_A = 2 kΩ to 100 kΩ, C = 0.1 μF.

¶Values specified are for a device in an astable circuit similar to Figure 12, with component values as follow: R_A = 1 kΩ to 100 kΩ, C = 0.1 μF.

4

Special Functions



TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

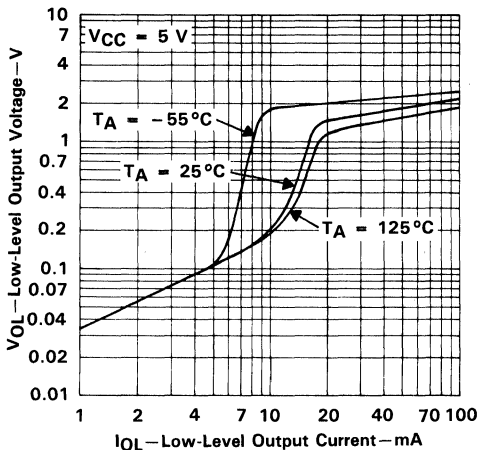


FIGURE 1

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

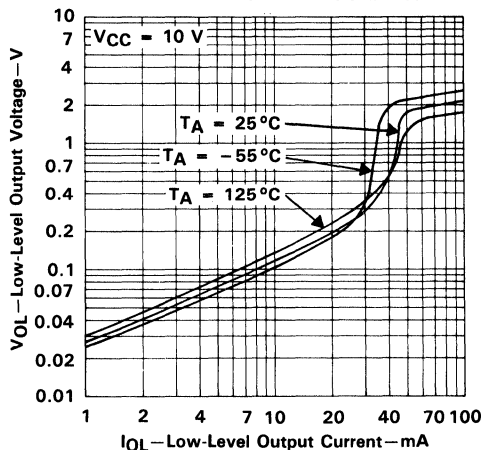


FIGURE 2

LOW-LEVEL OUTPUT VOLTAGE
 vs
 LOW-LEVEL OUTPUT CURRENT

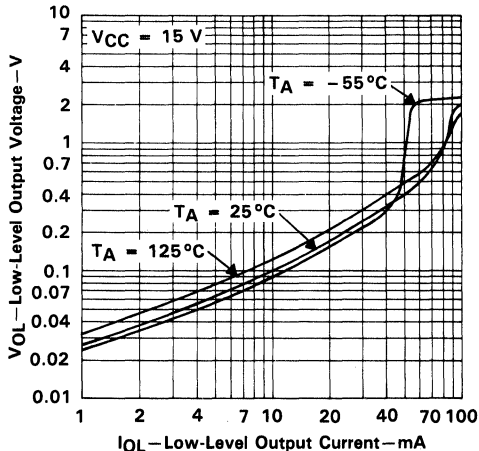


FIGURE 3

DROP BETWEEN SUPPLY VOLTAGE AND OUTPUT
 vs
 HIGH-LEVEL OUTPUT CURRENT

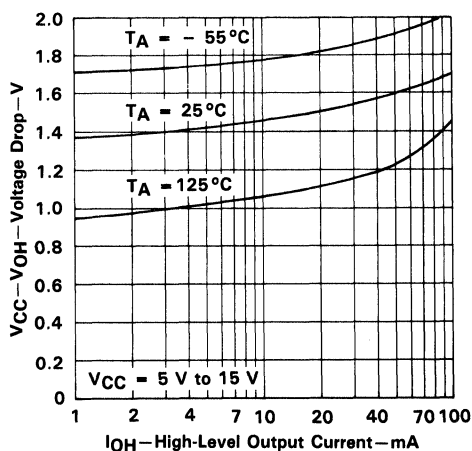


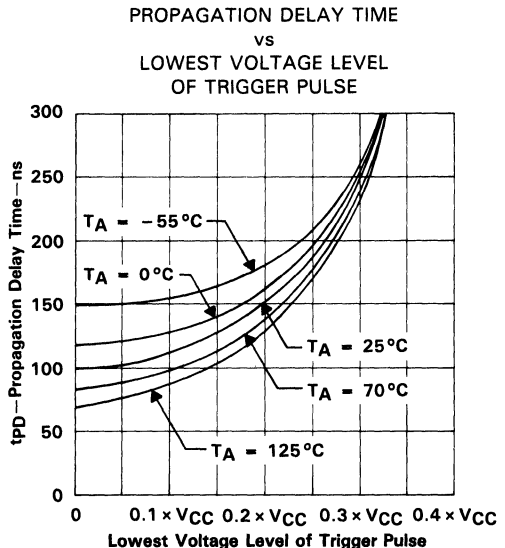
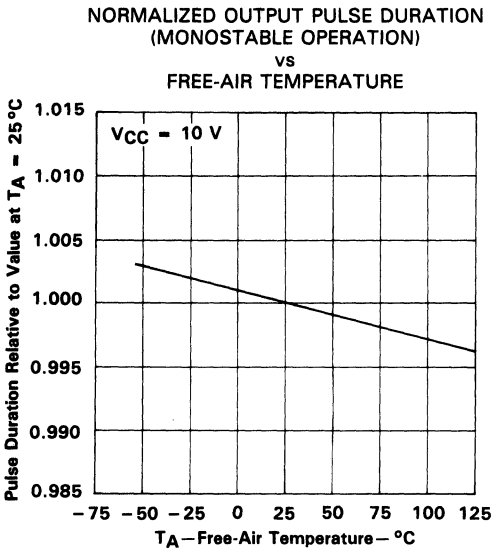
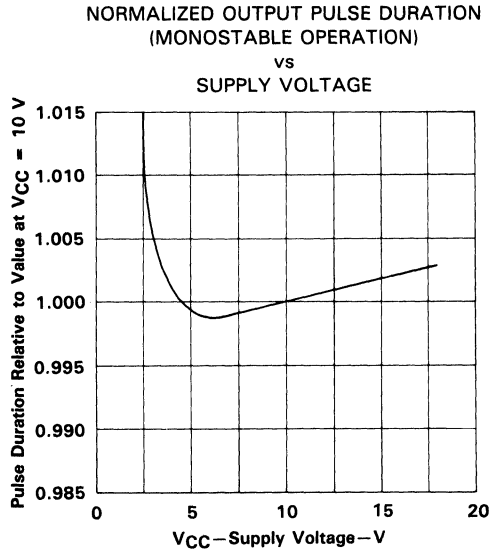
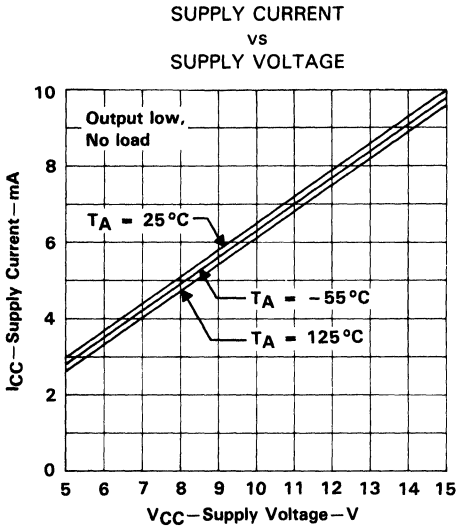
FIGURE 4

†Data for temperatures below 0°C and above 70°C are applicable for SE555 circuits only.

4

Special Functions

TYPICAL CHARACTERISTICS†



**4
Special Functions**

†Data for temperatures below 0°C and above 70°C are applicable for SE555 circuits only.

TYPICAL APPLICATION DATA

monostable operation

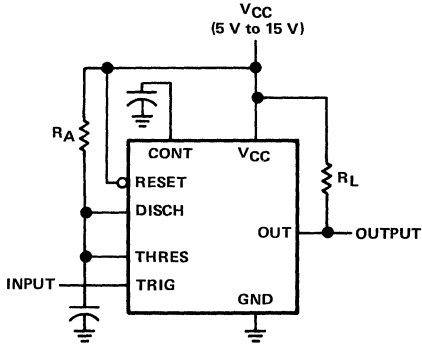


FIGURE 9. CIRCUIT FOR MONOSTABLE OPERATION

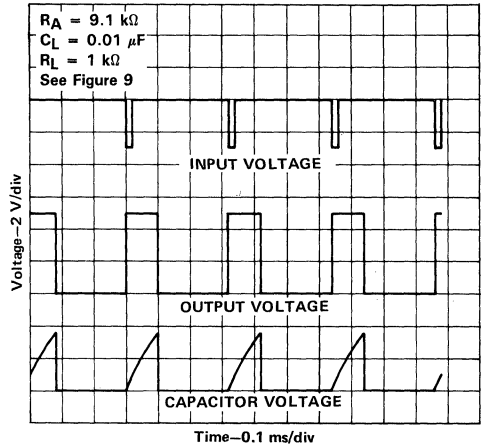


FIGURE 10. TYPICAL MONOSTABLE WAVEFORMS

For monostable operation, any of these timers may be connected as shown in Figure 9. If the output is low, application of a negative-going pulse to the trigger input sets the flip-flop (\bar{Q} goes low), drives the output high, and turns off Q1. Capacitor C is then charged through R_A until the voltage across the capacitor reaches the threshold voltage of the threshold input. If the trigger input has returned to a high level, the output of the threshold comparator will reset the flip-flop (\bar{Q} goes high), drive the output low, and discharge C through Q1.

Monostable operation is initiated when the trigger input voltage falls below the trigger threshold. Once initiated, the sequence ends only if the trigger input is high at the end of the timing interval. Because of the threshold level and saturation voltage of Q1, the output pulse duration is approximately $t_w = 1.1 R_A C$. Figure 11 is a plot of the time constant for various values of R_A and C. The threshold levels and charge rates are both directly proportional to the supply voltage, V_{CC} . The timing interval is therefore independent of the supply voltage, so long as the supply voltage is constant during the time interval.

Applying a negative-going trigger pulse simultaneously to the reset and trigger terminals during the timing interval discharges C and re-initiates the cycle, commencing on the positive edge of the reset pulse. The output is held low as long as the reset pulse is low. To prevent false triggering, when the reset input is not used, it should be connected to V_{CC} .

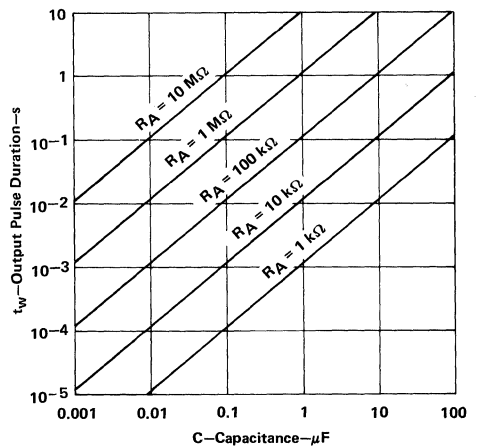


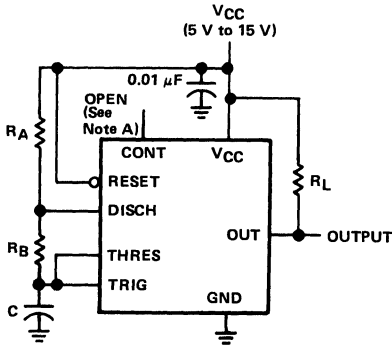
FIGURE 11. OUTPUT PULSE DURATION vs CAPACITANCE

4

Special Functions

TYPICAL APPLICATION DATA

astable operation



NOTE A: Decoupling the control voltage input to ground with a capacitor may improve operation. This should be evaluated for individual applications.

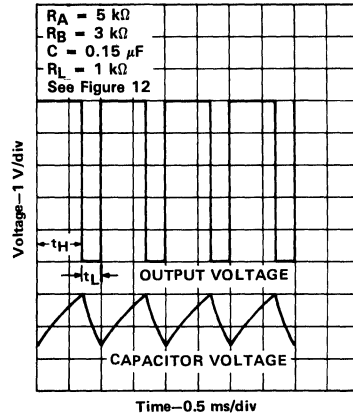


FIGURE 12. CIRCUIT FOR ASTABLE OPERATION

FIGURE 13. TYPICAL ASTABLE WAVEFORMS

As shown in Figure 12, adding a second resistor, R_B , to the circuit of Figure 9 and connecting the trigger input to the threshold input causes the timer to self-trigger and run as a multivibrator. The capacitor C will charge through R_A and R_B and then discharge through R_B only. The duty cycle may be controlled, therefore, by the values of R_A and R_B .

This astable connection results in capacitor C charging and discharging between the threshold-voltage level ($\approx 0.67 \cdot V_{CC}$) and the trigger-voltage level ($\approx 0.33 \cdot V_{CC}$). As in the monostable circuit, charge and discharge times (and therefore the frequency and duty cycle) are independent of the supply voltage.

Figure 13 shows typical waveforms generated during astable operation. The output high-level duration t_H and low-level duration t_L may be calculated as follows:

$$t_H = 0.693 (R_A + R_B) C$$

$$t_L = 0.693 (R_B) C$$

Other useful relationships are shown below.

$$\text{period} = t_H + t_L = 0.693 (R_A + 2R_B) C$$

$$\text{frequency} \approx \frac{1.44}{(R_A + 2R_B) C}$$

$$\text{Output driver duty cycle} = \frac{t_L}{t_H + t_L} = \frac{R_B}{R_A + 2R_B}$$

$$\text{Output waveform duty cycle} = \frac{t_H}{t_H + t_L} = 1 - \frac{R_B}{R_A + 2R_B}$$

$$\text{Low-to-high ratio} = \frac{t_L}{t_H} = \frac{R_B}{R_A + R_B}$$

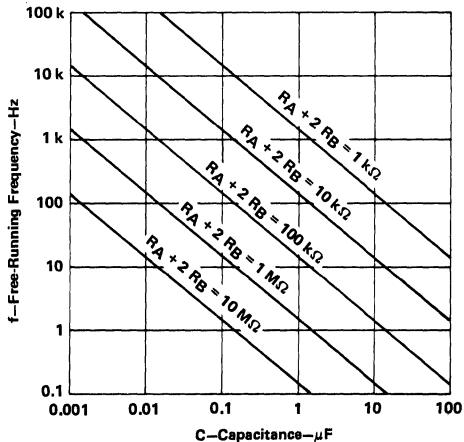


FIGURE 14. FREE-RUNNING FREQUENCY

TYPICAL APPLICATION DATA

missing-pulse detector

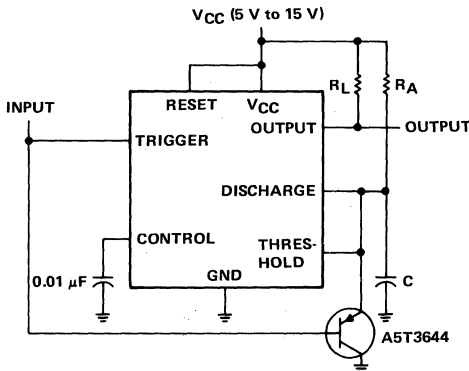


FIGURE 15. CIRCUIT FOR MISSING-PULSE DETECTOR

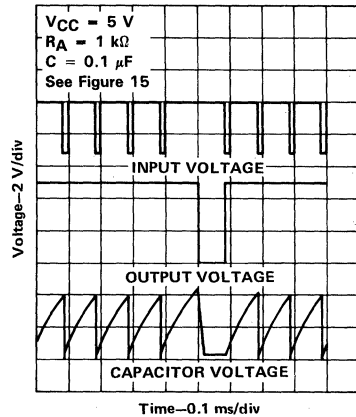


FIGURE 16. MISSING-PULSE DETECTOR WAVEFORMS

The circuit shown in Figure 15 may be used to detect a missing pulse or abnormally long spacing between consecutive pulses in a train of pulses. The timing interval of the monostable circuit is continuously retriggered by the input pulse train as long as the pulse spacing is less than the timing interval. A longer pulse spacing, missing pulse, or terminated pulse train permits the timing interval to be completed, thereby generating an output pulse as illustrated in Figure 16.

4 frequency divider

By adjusting the length of the timing cycle, the basic circuit of Figure 9 can be made to operate as a frequency divider. Figure 17 illustrates a divide-by-3 circuit that makes use of the fact that retriggering cannot occur during the timing cycle.

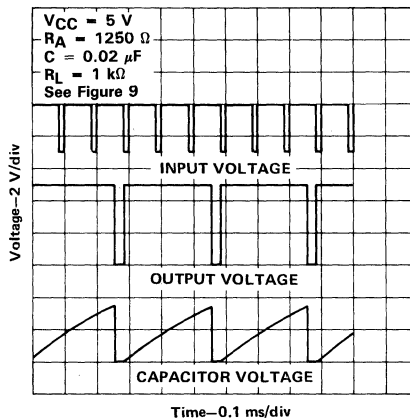
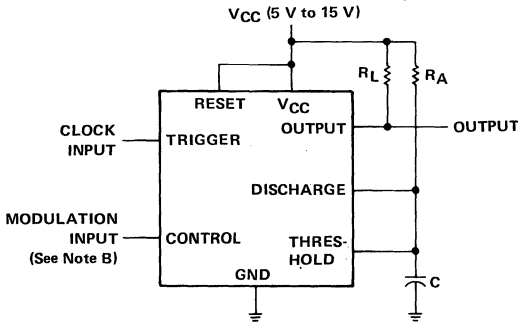


FIGURE 17. DIVIDE-BY-THREE CIRCUIT WAVEFORMS

TYPICAL APPLICATION DATA

pulse-width modulation



NOTE B: The modulating signal may be direct or capacitively coupled to the control terminal. For direct coupling, the effects of modulation source voltage and impedance on the bias of the timer should be considered.

FIGURE 18. CIRCUIT FOR PULSE-WIDTH MODULATION

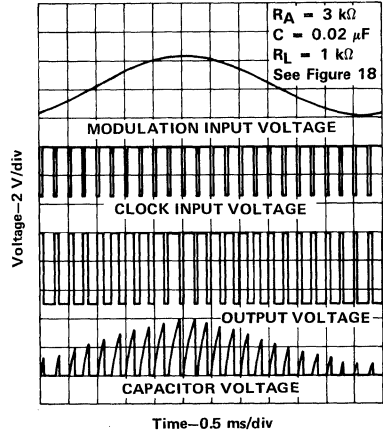
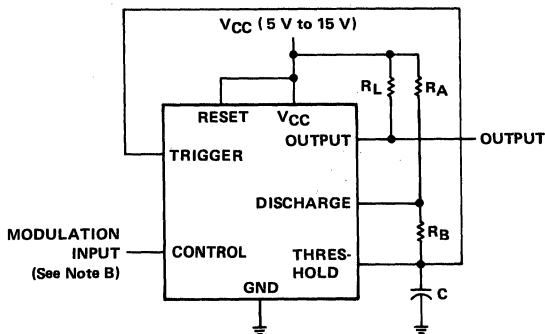


FIGURE 19. PULSE-WIDTH MODULATION WAVEFORMS

The operation of the timer may be modified by modulating the internal threshold and trigger voltages, which is accomplished by applying an external voltage (or current) to the control pin. Figure 18 shows a circuit for pulse-width modulation. A continuous input pulse train triggers the monostable circuit, and a control signal modulates the threshold voltage. Figure 19 illustrates the resulting output pulse-width modulation. While a sine-wave modulation signal is illustrated, any wave shape could be used.

TYPICAL APPLICATION DATA

pulse-position modulation



NOTE B: The modulating signal may be direct or capacitively coupled to the control terminal. For direct coupling, the effects of modulation source voltage and impedance on the bias of the timer should be considered.

FIGURE 20. CIRCUIT FOR PULSE-POSITION MODULATION

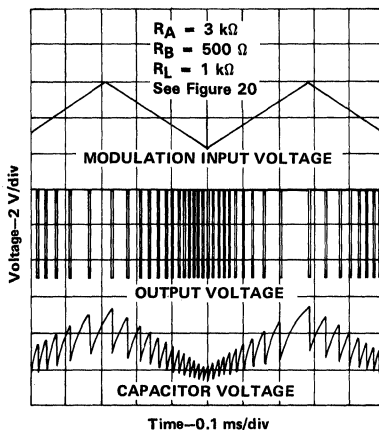
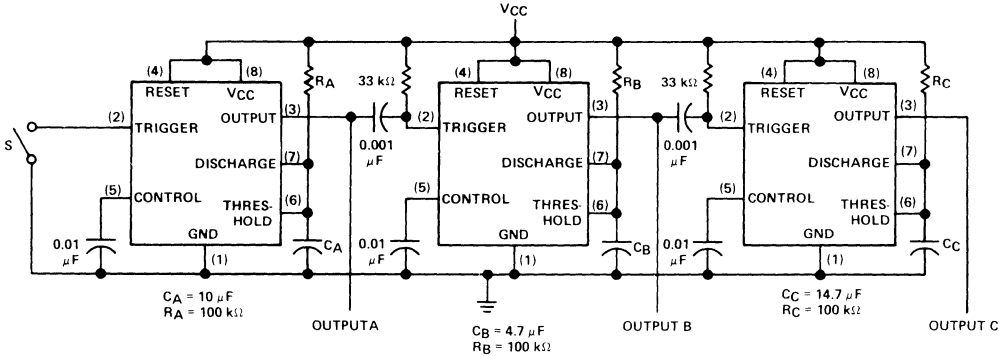


FIGURE 21. PULSE POSITION-MODULATION WAVEFORMS

As shown in Figure 20, any of these timers may be used as a pulse-position modulator. This application modulates the threshold voltage, and thereby the time delay, of a free-running oscillator. Figure 21 illustrates a triangular-wave modulation signal for such a circuit; however, any wave shape could be used.

TYPICAL APPLICATION DATA

sequential timer



S closes momentarily at $t = 0$.

FIGURE 22. SEQUENTIAL TIMER CIRCUIT

Many applications, such as computers, require signals for initializing conditions during start-up. Other applications, such as test equipment, require activation of test signals in sequence. These timing circuits may be connected to provide such sequential control. The timers may be used in various combinations of astable or monostable circuit connections, with or without modulation, for extremely flexible waveform control. Figure 22 illustrates a sequencer circuit with possible applications in many systems, and Figure 23 shows the output waveforms.

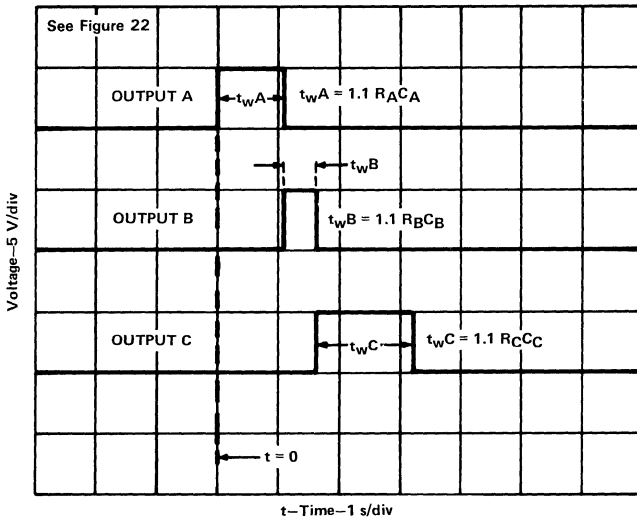


FIGURE 23. SEQUENTIAL TIMER WAVEFORMS

4

Special Functions

SE556, SE556C, SA556, NE556 DUAL PRECISION TIMERS

D2440, APRIL 1978—REVISED OCTOBER 1988

- Two Precision Timing Circuits per Package
- Astable or Monostable Operation
- TTL-Compatible Output Can Sink or Source Up to 150 mA
- Active Pull-Up or Pull-Down
- Designed to be Interchangeable with Signetics SE556, SE556C, SA556, NE556

APPLICATIONS

Precision Timer from Microseconds to Hours	Sequential Timer Pulse Generator
Pulse-Shaping Circuit	Time-Delay Circuit
Missing-Pulse Detector	Frequency Divider
Tone-Burst Generator	Appliance Timer
Pulse-Width Modulator	Industrial Controls
Pulse-Position Modulator	Touch-Tone Encoder

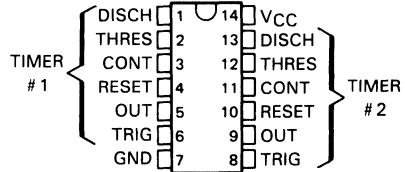
**SE556C FROM TI IS NOT
RECOMMENDED FOR NEW DESIGNS**

description

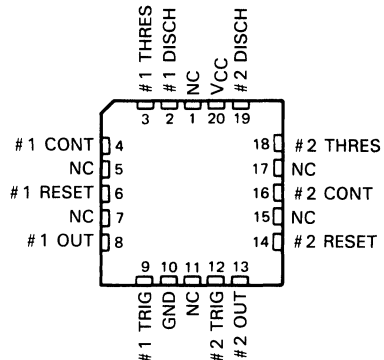
These devices provide two monolithic, independent timing circuits of the SE555, SE555C, SA555, or NE555 type in each package. These circuits can be operated in the astable or the monostable mode with external resistor-capacitor timing control. The basic timing provided by the RC time constant may be actively controlled by modulating the bias of the control voltage input.

The threshold and trigger levels are normally two-thirds and one-third respectively of V_{CC} . These levels can be altered by use of the control voltage terminal. When the trigger input falls below trigger level, the flip-flop is set and the output goes high. If the trigger input is above the trigger level and the threshold input is above the threshold level, the flip-flop is reset and the output is low. The reset input can override all other inputs and can be used to initiate a new timing cycle. When the reset input goes low, the flip-flop is reset and the output goes low. Whenever the output is low, a low impedance path is provided between the discharge terminal and ground.

SE556, SE556C . . . J PACKAGE
SA556, NE556 . . . D, J, OR N PACKAGE
(TOP VIEW)

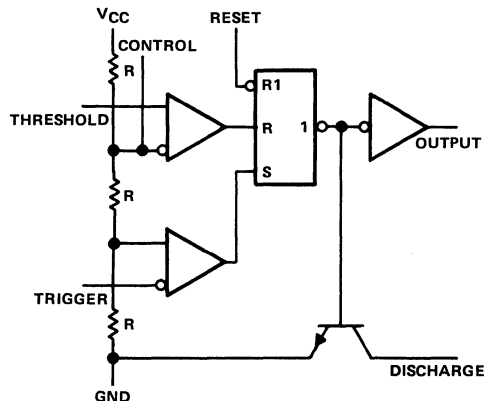


SE556, SE556C . . . FK PACKAGE
(TOP VIEW)



NC—No internal connection

functional block diagram (each timer)



Reset can override Trigger, which can override Threshold.

4

Special Functions

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

SE556, SE556C, SA556, NE556 DUAL PRECISION TIMERS

The SE556 and SE556C are characterized for operation over the full military range of -55°C to 125°C . The SA556 is characterized for operation from -40°C to 85°C , and the NE556 is characterized for operation from 0°C to 70°C .

AVAILABLE OPTIONS

T _A RANGE	V _{thres} MAX V _{CC} = 15 V	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	11.2 V	NE556D		NE556J	NE556N
-40°C to 85°C	11.2 V	SA556D		SA556J	SA556N
-55°C to 125°C	10.6 V 11.2 V		SE556FK SE556CFK	SE556J SE556CJ	

FUNCTION TABLE

RESET	TRIGGER VOLTAGE†	THRESHOLD VOLTAGE†	OUTPUT	DISCHARGE SWITCH
Low	Irrelevant	Irrelevant	Low	On
High	$< 1/3 V_{DD}$	Irrelevant	High	Off
High	$> 1/3 V_{DD}$	$> 2/3 V_{DD}$	Low	On
High	$> 1/3 V_{DD}$	$< 2/3 V_{DD}$	As previously established	

†Voltage levels shown are nominal.

The D package is available taped and reeled. Add the suffix R to the device type (e.g., NE556DR).

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V _{CC} (See Note 1)	18 V
Input voltage (control, reset, threshold, and trigger)	V _{CC}
Output current	± 225 mA
Continuous total dissipation	see Dissipation Rating Table
Operating free-air temperature range: SE556, SE556C	-55°C to 125°C
SA556	-40°C to 85°C
NE556	0°C to 70°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C

NOTE 1: All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C	DERATING FACTOR	T _A = 70°C	T _A = 85°C	T _A = 125°C
	POWER RATING	ABOVE T _A = 25°C	POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	N/A
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J (SE556, SE556C)	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J (SA556, NE556)	1025 mW	8.2 mW/°C	656 mW	533 mW	N/A
N	1575 mW	12.6 mW/°C	1008 mW	819 mW	N/A

4

Special Functions

SE556, SE556C, SA556, NE556 DUAL PRECISION TIMERS

recommended operating conditions

	SE556		SE556C		SA556		NE556		UNIT
	MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, V_{CC}	4.5	18	4.5	18	4.5	16	4.5	16	V
Input voltage (control, reset, threshold, and trigger)	V_{CC}		V_{CC}		V_{CC}		V_{CC}		V
Output current	± 200		± 200		± 200		± 200		mA
Operating free-air temperature, T_A	-55	125	-55	125	-40	85	0	70	°C

electrical characteristics at 25 °C free-air temperature, $V_{CC} = 5\text{ V to }15\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	SE556			SE556C, SA556, NE556			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX		
Threshold voltage level	$V_{CC} = 15\text{ V}$	9.4	10	10.6	8.8	10	11.2	V	
	$V_{CC} = 5\text{ V}$	2.7	3.3	4	2.4	3.3	4.2		
Threshold current (see Note 2)			30	250		30	250	nA	
Trigger voltage level	$V_{CC} = 15\text{ V}$	4.8	5	5.2	4.5	5	5.6	V	
	$V_{CC} = 5\text{ V}$	1.45	1.67	1.9	1.1	1.67	2.2		
Trigger current	Trigger at 0 V		0.5	0.9		0.5	2	μA	
Reset voltage level		0.3	0.7	1	0.3	0.7	1	V	
Reset current	Reset at V_{CC}		0.1	0.4		0.1	0.4	mA	
	Reset at 0 V		-0.4	-1		-0.4	-1.5		
Discharge switch off-state current			20	100		20	100	nA	
Control voltage (open circuit)	$V_{CC} = 15\text{ V}$	9.6	10	10.4	9	10	11	V	
	$V_{CC} = 5\text{ V}$	2.9	3.3	3.8	2.6	3.3	4		
Low-level output voltage	$V_{CC} = 15\text{ V}$	$I_{OL} = 10\text{ mA}$		0.1	0.15		0.1	0.25	V
		$I_{OL} = 50\text{ mA}$		0.4	0.5		0.4	0.75	
		$I_{OL} = 100\text{ mA}$		2	2.2		2	2.5	
		$I_{OL} = 200\text{ mA}$		2.5			2.5		
	$V_{CC} = 5\text{ V}$	$I_{OL} = 5\text{ mA}$		0.1	0.15		0.1	0.25	
		$I_{OL} = 8\text{ mA}$		0.15	0.25		0.15	0.3	
High-level output voltage	$V_{CC} = 15\text{ V}$	$I_{OH} = -100\text{ mA}$	13	13.3		12.75	13.3	V	
		$I_{OH} = -200\text{ mA}$		12.5			12.5		
	$V_{CC} = 5\text{ V}$	$I_{OH} = -100\text{ mA}$	3	3.3		2.75	3.3		
Supply current	Output low, No load	$V_{CC} = 15\text{ V}$		20	24		20	30	mA
		$V_{CC} = 5\text{ V}$		6	10		6	12	
	Output high, No load	$V_{CC} = 15\text{ V}$		18	20		18	26	
		$V_{CC} = 5\text{ V}$		4	8		4	10	

NOTE 2: This parameter influences the maximum value of the timing resistors R_A and R_B in the circuit of Figure 1. For example, when $V_{CC} = 5\text{ V}$, the maximum value is $R = R_A + R_B \approx 3.4\text{ M}\Omega$, and for $V_{CC} = 15\text{ V}$, the maximum value is $\approx 10\text{ M}\Omega$.

4
Special Functions

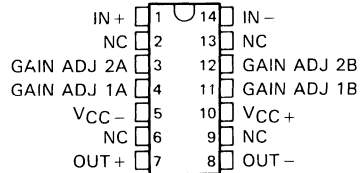
SE592, NE592, NE592A DIFFERENTIAL VIDEO AMPLIFIERS

D2667, FEBRUARY 1984—REVISED FEBRUARY 1988

- 90-MHz Bandwidth
- Adjustable Gain to 400
- No Frequency Compensation Required
- Adjustable Passband
- Designed to be Interchangeable with Signetics SE592 and NE592

NE592, NE592A . . . D OR N PACKAGE
SE592 . . . J PACKAGE

(TOP VIEW)



NC—No internal connection

DEVICE TYPE	TEMPERATURE RANGE	A _{VD} RANGE (GAIN OPTION 1)
SE592	-55°C to 125°C	300—500
NE592	0°C to 70°C	250—600
NE592A	0°C to 70°C	400—600

description

These devices are monolithic two-stage amplifiers with differential inputs and differential outputs.

Internal series-shunt feedback provides wide bandwidth, low phase distortion, and excellent gain stability. Emitter-follower outputs enable the devices to drive capacitive loads, and all stages are current-source biased to obtain high common-mode and supply-voltage rejection ratios.

Fixed differential amplification of 100 or 400 may be selected without external components; or amplification may be adjusted from 0 to 400 by the use of a single external resistor connected between the gain-adjustment pins 1A and 1B. External frequency-compensating components are not required for any gain option.

The devices are particularly useful in magnetic-tape or disc-file systems using phase or NRZ encoding and in high-speed thin-film or plated-wire memories. Other applications include general purpose video and pulse amplifiers where wide bandwidth, low phase shift, and excellent gain stability are required.

The SE592 is characterized for operation over the full military temperature range of -55°C to 125°C. The NE592 and NE592A are characterized for operation from 0°C to 70°C.

4

Special Functions

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS
INSTRUMENTS

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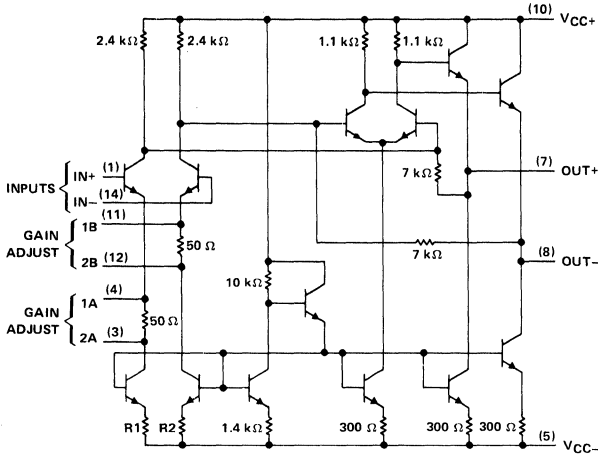
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4-53

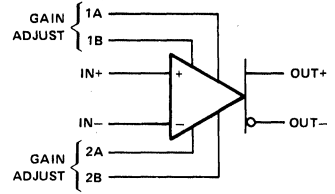
SE592, NE592, NE592A

DIFFERENTIAL VIDEO AMPLIFIERS

schematic



symbol



All resistor values shown are in ohms and nominal.
 In NE592 or SE592, R1 = 500 Ω, R2 = 500 Ω.
 In NE592A, R1 = 600 Ω, R2 = 600 Ω.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage V_{CC+} (see Note 1)	8 V
Supply voltage V_{CC-} (see Note 1)	-8 V
Differential input voltage	± 5 V
Common-mode input voltage	± 6 V
Output current	10 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range: SE592	-55°C to 125°C
NE592, NE592A	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

NOTE 1: All voltage values except differential input voltages are with respect to the midpoint between V_{CC+} and V_{CC-} .

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING			POWER RATING	POWER RATING
D	500 mW	N/A	N/A	500 mW	
J	500 mW	11 mW/°C	105°C	500 mW	275 mW
N	500 mW	N/A	N/A	500 mW	

4
Special Functions

SE592, NE592, NE592A DIFFERENTIAL VIDEO AMPLIFIERS

recommended operating conditions

	SE592			NE592 NE592A			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, V_{CC+}	3	6	8	3	6	8	V
Supply voltage, V_{CC-}	-3	-6	-8	-3	-6	-8	V
Operating free-air temperature, T_A	-55		125	0		70	°C

electrical characteristics at 25°C operating free-air temperature, $V_{CC+} = 6\text{ V}$, $V_{CC-} = -6\text{ V}$ (unless otherwise noted)

PARAMETER	TEST FIGURE	TEST CONDITIONS	GAIN OPTION†	SE592			UNIT	
				MIN	TYP	MAX		
AVD Large-signal differential voltage amplification	1	$V_{OPP} = 3\text{ V}$, $R_L = 2\text{ k}\Omega$	1	300	400	500	V/V	
			2	90	100	110		
BW Bandwidth (-3 dB)	2	$V_{OPP} = 1\text{ V}$	1	40			MHz	
			2	90				
I_{IO} Input offset current			1, 2, or 3	0.4	3		μA	
I_{IB} Input bias current			1, 2, or 3	9		20	μA	
V_{ICR} Common-mode input voltage range	3		1, 2, or 3	± 1			V	
V_{OC} Common-mode output voltage	1	$R_L = \infty$	1, 2, or 3	2.4	2.9	3.4	V	
V_{OO} Output offset voltage	1	$V_{IO} = 0$, $R_L = \infty$	1	1.5			V	
			2	1				
			3	0.35	0.75			
V_{OPP} Maximum peak-to-peak output voltage swing	1	$R_L = 2\text{ k}\Omega$	1, 2, or 3	3	4		V	
r_i Input resistance			1	4			$\text{k}\Omega$	
			2	20	30			
r_o Output resistance				20			Ω	
C_i Input capacitance				2			pF	
CMRR Common-mode rejection ratio	3	$V_{IC} = \pm 1\text{ V}$, $f = 100\text{ kHz}$	2	60	86		dB	
	3	$V_{IC} = \pm 1\text{ V}$, $f = 5\text{ MHz}$	2	60				
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IO}$)	4	$\Delta V_{CC+} = \pm 0.5\text{ V}$, $\Delta V_{CC-} = \pm 0.5\text{ V}$	2	50	70		dB	
V_n Broadband equivalent noise voltage	4	$BW = 1\text{ kHz to } 10\text{ MHz}$	1, 2, or 3	12			μV	
t_{pd} Propagation delay time	2	$\Delta V_O = 1\text{ V}$	1	7.5			ns	
			2	6	10			
t_r Rise time	2	$\Delta V_O = 1\text{ V}$	1	10.5			ns	
			2	4.5				
$I_{sink(max)}$ Maximum output sink current			1, 2, or 3	3	4		mA	
I_{CC} Supply current		No load, No signal	1, 2, or 3	18			24	mA

†The gain option is selected as follows:

Gain Option 1 . . Gain Adjust pin 1A is connected to pin 1B, pins 2A and 2B are open.

Gain Option 2 . . Gain Adjust pin 2A is connected to pin 2B, pins 1A and 1B are open.

Gain Option 3 . . All Gain Adjust pins are open.

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Special Functions

SE592

DIFFERENTIAL VIDEO AMPLIFIER

electrical characteristics over recommended operating free-air temperature range, $V_{CC+} = 6\text{ V}$, $V_{CC-} = -6\text{ V}$ (unless otherwise noted)

PARAMETER	TEST FIGURE	TEST CONDITIONS	GAIN OPTION†	SE592			UNIT
				MIN	TYP	MAX	
A _{VD}	Large-signal differential voltage amplification	1	V _{OPP} = 3 V	1	200	600	V/V
				2	80	120	
I _{IO}	Input offset current		1 or 2			5	μA
I _{IB}	Input bias current		1 or 2			40	μA
V _{ICR}	Common-mode input voltage range	3		1 or 2	±1		V
V _{OO}	Output offset voltage	1	V _{ID} = 0, R _L = ∞	1		1.5	V
				2		1.2	
				3		1	
V _{OPP}	Maximum output voltage peak-to-peak swing	1	R _L = 2 kΩ	1 or 2	2.5		V
r _i	Input resistance			2	8		kΩ
CMRR	Common-mode rejection ratio	3	V _{IC} = ±1 V, f = 100 kHz	2	50		dB
k _{SVR}	Supply voltage rejection ratio (ΔV _{CC} /ΔV _{IO})	4	ΔV _{CC+} = ±0.5 V, ΔV _{CC-} = ±0.5 V	2	50		dB
I _{sink(max)}	Maximum output sink current			1, 2, or 3	2.5		mA
I _{CC}	Supply current	1	No load, No signal	1, 2, or 3		27	mA

†The gain option is selected as follows:

Gain Option 1 . . Gain Adjust pin 1A is connected to pin 1B; pins 2A and 2B are open.

Gain Option 2 . . Gain Adjust pin 2A is connected to pin 2B; pins 1A and 1B are open.

Gain Option 3 . . All Gain Adjust pins are open.

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Special Functions

electrical characteristics at 25°C operating free-air temperature, $V_{CC+} = 6\text{ V}$, $V_{CC-} = -6\text{ V}$ (unless otherwise noted)

PARAMETER	TEST FIGURE	TEST CONDITIONS	GAIN OPTION†	NE592			NE592A			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
A _{VD} Large signal differential voltage amplification	1	V _{OPP} = 3 V, R _L = 2 kΩ	1	250	400	600	400	440	600	V/V
			2	80	100	120	80	100	120	
BW Bandwidth (-3 dB)	2	V _{OPP} = 1 V	1		40			40	MHz	
			2		90		90			
I _{IQ} Input offset current		V _{IC} = 0	1, 2, or 3		0.4	5		0.4	5	μA
I _{IB} Input bias current		V _{IC} = 0	1, 2, or 3		9	30		10	30	μA
V _{ICR} Common-mode input voltage range	3		1, 2, or 3		±1			±1		V
V _{OC} Common-mode output voltage	1	R _L = ∞		2.4	2.9	3.4	2.4	2.9	3.4	V
V _{OO} Output offset voltage	1	V _{ID} = 0, R _L = ∞	1 or 2			1.5			1.5	V
			3		0.35	0.75		0.35	0.75	
V _{OPP} Maximum peak-to-peak output voltage swing	1	R _L = 2 kΩ	1, 2, or 3	3	4		3	4		V
r _i Input resistance			1		4			4		kΩ
			2	10	30		10	30		
r _o Output resistance					20			20		Ω
C _i Input capacitance					2			2		pF
CMRR Common-mode rejection ratio	3	V _{IC} = ±1 V, f = 100 kHz	2	60	86		60	86		dB
	3	V _{IC} = ±1 V, f = 5 MHz	2		60			60		
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} /ΔV _{IO})	4	ΔV _{CC+} = ±0.5 V, ΔV _{CC-} = ±0.5 V	2	50	70		50	70		dB
V _n Broadband equivalent noise voltage	4	BW = 1 kHz to 10 MHz	1, 2, or 3		12			12		μV
t _{pd} Propagation delay time	2	ΔV _O = 1 V	1		7.5			7.5		ns
			2		6	10		6	10	
t _r Rise time	2	ΔV _O = 1 V	1		10.5			10.5		ns
			2		4.5	12		4.5	12	
I _{sink(max)} Maximum output sink current			1, 2, or 3	3	4		3	4		mA
I _{CC} Supply current		No load, No signal	1, 2, or 3		18	24		19	24	mA

†The gain option is selected as follows:

Gain Option 1 . . Gain Adjust pin 1A is connected to pin 1B, pins 2A and 2B are open.

Gain Option 2 . . Gain Adjust pin 2A is connected to pin 2B, pins 1A and 1B are open.

Gain Option 3 . . All Gain Adjust pins are open.

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electrical characteristics over recommended operating free-air temperature range, $V_{CC+} = 6\text{ V}$, $V_{CC-} = -6\text{ V}$
(unless otherwise noted)

PARAMETER	TEST FIGURE	TEST CONDITIONS	GAIN OPTION†	NE592			NE592A			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
A_{VD} Large signal differential voltage amplification	1	$V_{OPP} = 3\text{ V}$	1	250		600	400		600	V/V
			2	80		120	80		120	
I_{IQ} Input offset current			1 or 2			6			6	μA
I_{IB} Input bias current			1 or 2			40			40	μA
V_{ICR} Common-mode input voltage range	3		1 or 2			± 1			± 1	V
V_{OO} Output offset voltage	1	$V_{ID} = 0, R_L = \infty$	1 or 2			1.5			1.5	V
			3			1			1	
V_{OPP} Maximum output voltage peak-to-peak swing	1	$R_L = 2\text{ k}\Omega$	1 or 2	2.8			2.8			V
r_i Input resistance			2	8			8			$\text{k}\Omega$
CMRR Common-mode rejection ratio	3	$V_{IC} = \pm 1\text{ V}, f = 100\text{ kHz}$	2	50			50			dB
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IQ}$)	4	$\Delta V_{CC+} = \pm 0.5\text{ V}, \Delta V_{CC-} = \pm 0.5\text{ V}$	2	50			50			dB
$I_{sink(max)}$ Maximum output sink current			1, 2, or 3				2.8	4		mA
I_{CC} Supply current	1	No load, No signal	1, 2, or 3			27			27	mA

†The gain option is selected as follows:

Gain Option 1 . . Gain Adjust pin 1A is connected to pin 1B, pins 2A and 2B are open.

Gain Option 2 . . Gain Adjust pin 2A is connected to pin 2B, pins 1A and 1B are open.

Gain Option 3 . . All Gain Adjust pins are open.

SE592, NE592, NE592A
DIFFERENTIAL VIDEO AMPLIFIERS

PARAMETER MEASUREMENT INFORMATION

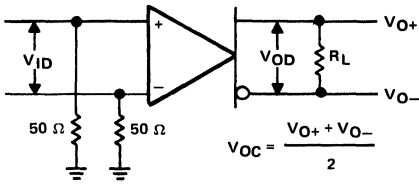


FIGURE 1

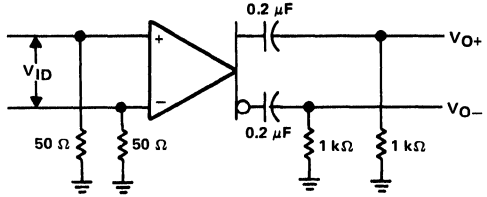


FIGURE 2

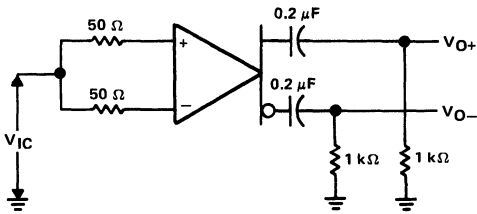


FIGURE 3

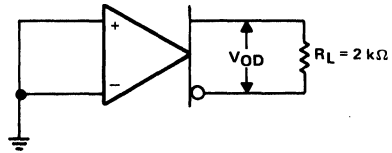
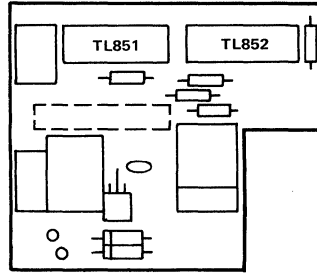


FIGURE 4

SN28827 SONAR RANGING MODULE

D2780, OCTOBER 1983—REVISED MARCH 1988

- Accurate Sonar Ranging from 6 Inches to 35 Feet
- Drives 50-kHz Electrostatic Transducer with No Additional Interface
- Operates from Single Supply
- Accurate Clock Output Provided for External Use
- Selective Echo Exclusion
- TTL-Compatible
- Multiple Measurement Capability
- Uses TL851 and TL852 Sonar Ranging Integrated Circuits



See schematic, Figure 4, for terminal assignments.

description

The SN28827 is an economical sonar ranging module that can drive a 50-kHz, 300-V electrostatic transducer with no additional interface. This module, with a simple interface, is able to measure distances of from 6 inches to 35 feet. The typical absolute accuracy is $\pm 2\%$ at one foot or greater.

This module has an external blanking input that allows selective echo exclusion for operation in a multiple-echo mode. The module is able to differentiate echos from objects that are only three inches apart. The digitally controlled-gain, variable-bandwidth amplifier minimizes noise and sidelobe detection in sonar applications.

The module has an accurate ceramic-resonator-controlled 420-kHz time-base generator. An output based on the 420-kHz time base is provided for external use. The sonar transmit output is 16 pulses at a frequency of 49.4 kHz.

The SN28827 operates over a supply voltage range of from 4.5 V to 6.8 V and is characterized for operation from 0°C to 40°C.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Voltage from any pin to ground (see Note 1)	7 V
Voltage from any pin except XDCR to V _{CC} (see Note 1)	-7 V to 0.5 V
Operating free-air temperature range	0°C to 40°C
Storage temperature range	-40°C to 85°C

NOTE 1: The XDCR pin may be driven from -1 V to 300 V typical with respect to ground.

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Special Functions

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SN28827 SONAR RANGING MODULE

recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, V_{CC}	4.5	6.8	V
High-level input voltage, V_{IH}	BLNK, BINH, INIT		V
Low-level input voltage, V_{IL}	BLNK, BINH, INIT		0.6
ECHO and OSC output voltage		6.8	V
Delay time, power up to INIT high	5		ms
Recycle period	80		ms
Operating free-air temperature, T_A	0	40	°C

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP [†]	MAX	UNIT
Input current	BLNK, BINH, INIT	$V_I = 2.1$ V			1	mA
High-level output current, I_{OH}	ECHO, OSC	$V_{OH} = 5.5$ V			100	μ A
Low-level output voltage, V_{OL}	ECHO, OSC	$I_{OL} = 1.6$ mA			0.4	V
Transducer bias voltage				150		V
Transducer output voltage (peak-to-peak)				300		V
Number of cycles for XDCR output to reach 300 V		$C = 500$ pF			7	
Internal blanking interval		See Note 2		2.38		ms
Frequency during 16-pulse transmit period	OSC output	See Note 2		49.4		kHz
	XMIT output			49.4		
Frequency after 16-pulse transmit period	OSC output	See Note 2		93.3		kHz
	XMIT output			0		
Supply current, I_{CC}	During transmit period				2000	mA
	After transmit period				100	

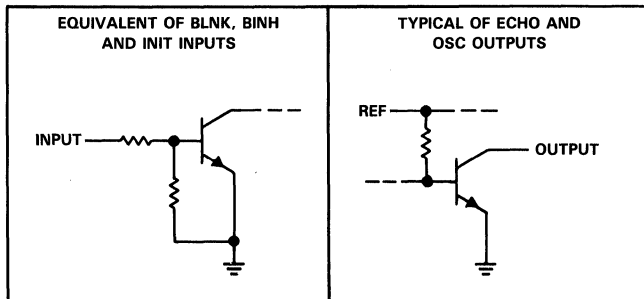
[†]All typical values are at $V_{CC} = 5$ V and $T_A = 25^\circ$ C.

NOTE 2: These typical values apply for a 420-kHz ceramic resonator.

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Special Functions

schematics of inputs and outputs



operation with Polaroid electrostatic transducer

There are two basic modes of operation for the SN28827 Sonar ranging module: single-echo mode and multiple-echo mode. The application of power (V_{CC}), the activation of the Initiate (INIT) input, and the resulting transmit output, and the use of the Blanking Inhibit (BINH) input are basically the same for either mode of operation. After applying power (V_{CC}), a minimum of 5 ms must elapse before the INIT input can be taken high. During this time, all internal circuitry is reset and the internal oscillator stabilizes. When INIT is taken high, drive to the Transducer (XDCR) output occurs. Sixteen pulses at 49.4 kHz with 300-V amplitude will excite the transducer as transmission occurs. At the end of the 16 transmit pulses, a dc bias of 150 V will remain on the transducer as recommended for optimum operation by the transducer manufacturer.

In order to eliminate ringing of the transducer from being detected as a return signal, the receive (REC) input of the ranging control IC is inhibited by internal blanking for 2.38 ms after the initiate signal. If a reduced blanking time is desired, then the BINH input can be taken high to end the blanking of the Receive input anytime prior to internal blanking. This may be desired to detect objects closer than 1.33 feet corresponding to 2.38 ms and may be done if transducer damping is sufficient that ringing is not detected as a return signal.

In the single-echo mode of operation (Figure 1), all that must be done next is to wait for the return of the transmitted signal, traveling at approximately 0.9 ms per foot out and back. The returning signal is amplified and appears as a high-logic-level echo output. The time between INIT going high and the echo (ECHO) output going high is proportional to the distance of the target from the transducer. If desired, the cycle can now be repeated by returning INIT to a low-logic level and then taking it high when the next transmission is desired.

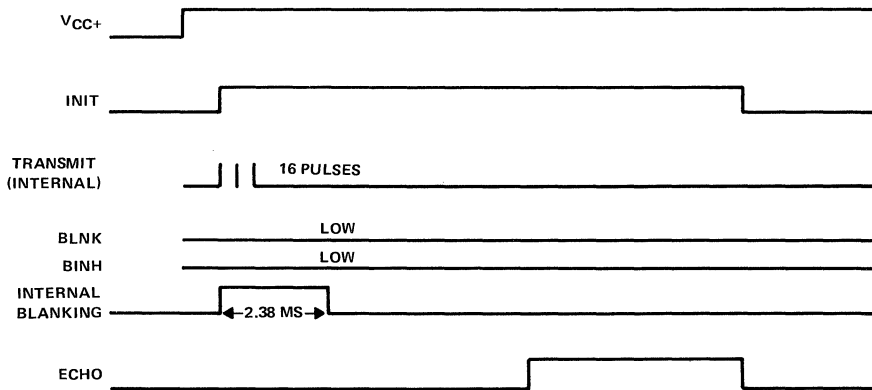


FIGURE 1. EXAMPLE OF A SINGLE-ECHO-MODE CYCLE WITHOUT BLANKING INPUT

If there is more than one target and multiple echos are to be detected from a single transmission, then the cycle is slightly different (Figure 2). After receiving the first return signal, which causes the ECHO output to go high, the Blanking (BLNK) input must be taken high then back low to reset the Echo output for the next return signal. The blanking signal must be at least 0.44 ms in duration to account for all 16 returning pulses from the first target and allow for internal delay times. This corresponds to the two targets being 3 inches apart.

**SN28827
SONAR RANGING MODULE**

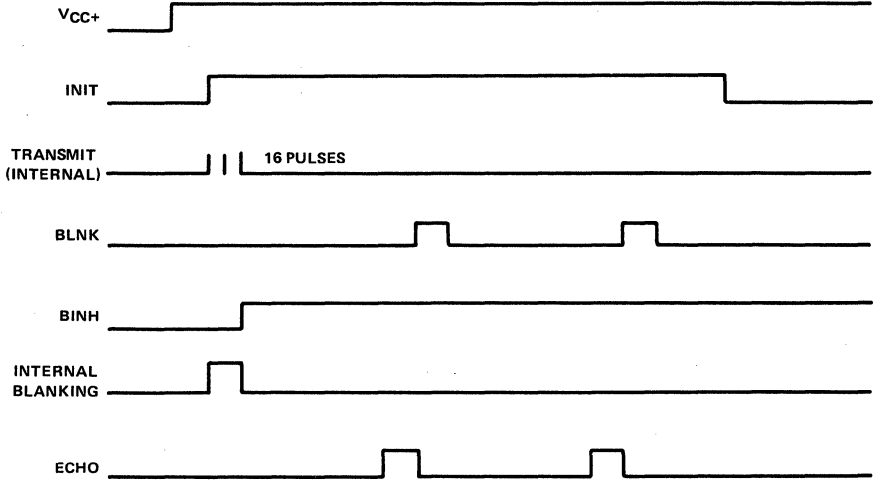


FIGURE 2. EXAMPLE OF A MULTIPLE-ECHO-MODE CYCLE WITH BLANKING INPUT

During a cycle starting with INIT going high, the receiver amplifier gain is incremented higher at discrete times (Figure 3) since the transmitted signal is attenuated with distance. At approximately 38 ms, the maximum gain is attained. For this reason, sufficient gain may not be available for objects greater than

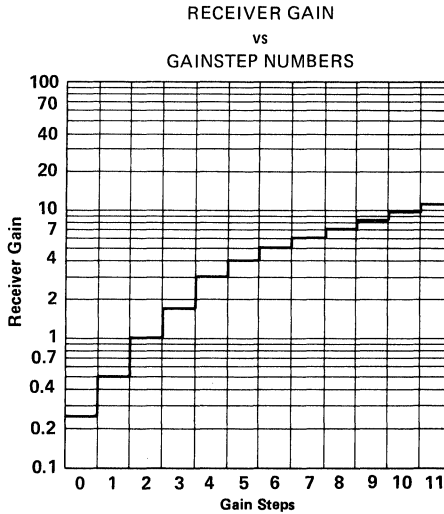


FIGURE 3

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Special Functions

SN28827 SONAR RANGING MODULE

35 feet away. Although gain can be increased by varying R1 (Figure 4), there is a limit to which the gain can be increased for reliable module operation. This will vary from application to application. The modules are "kitted" prior to their final test during manufacture. This is necessary because the desired gain distribution is much narrower than the module gain distribution if all were kitted with one value resistor. As kitted, these modules will perform satisfactorily in most applications. As a rule of thumb, the gain can be increased by up to a factor of 4, if required, by increasing R1 correspondingly. Gain is directly proportional to R1.

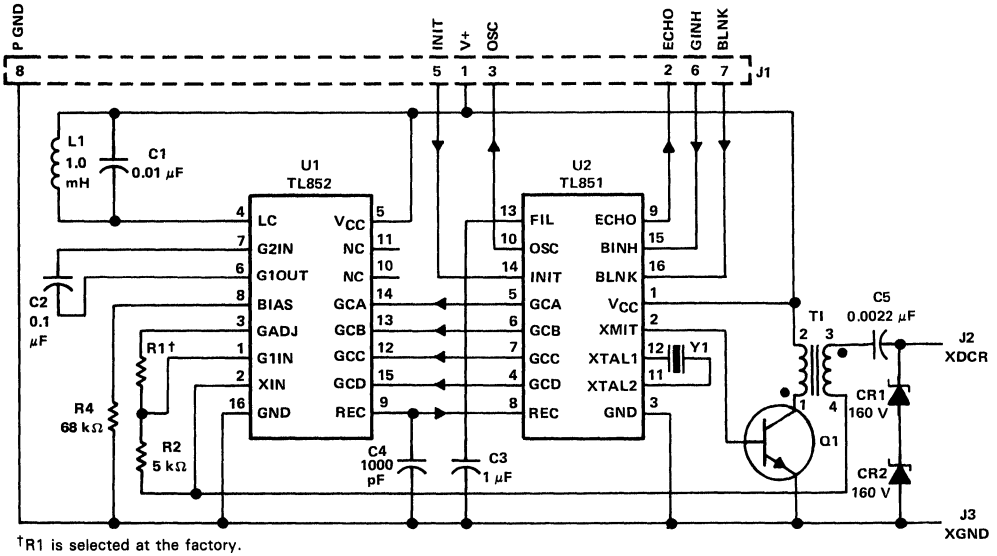
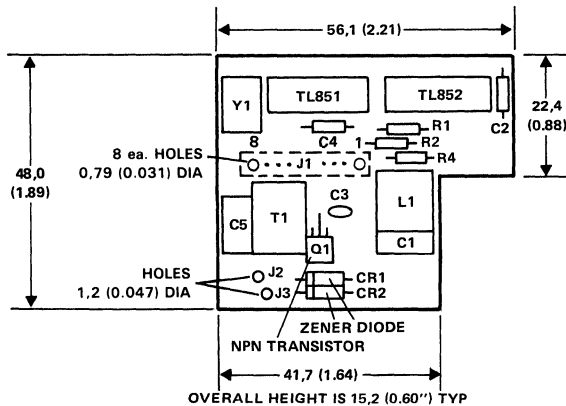


FIGURE 4. SCHEMATIC



NOTE: All dimensions are in millimeters and parenthetically in inches.

FIGURE 5. COMPONENT LAYOUT AND DIMENSIONS OF MODULE

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Special Functions

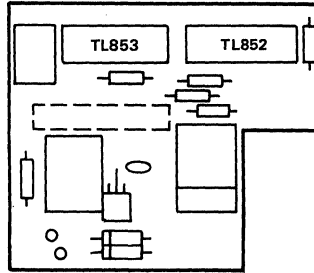
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Special Functions

SN28828 SONAR RANGING MODULE

D2842, JANUARY 1985

- Accurate Sonar Ranging from 6 Inches to 35 Feet
- Drives 40-kHz Piezoelectric Transducer
- Operates from Single Supply
- Accurate Clock Output Provided for External Use
- Selective Echo Exclusion
- TTL-Compatible
- Multiple Measurement Capability
- Uses TL852 and TL853 Sonar Ranging Integrated Circuits



See schematic, Figure 4, for terminal assignments.

description

The SN28828 is an economical sonar ranging module that can drive a 40-kHz piezoelectric transducer with no additional interface. This module, with a simple interface, is able to measure distances ranging from 6 inches to 35 feet. The typical absolute accuracy is $\pm 2\%$ at one foot or greater.

This module has an external blanking input that allows selective echo exclusion for operation in a multiple-echo mode. The module is able to differentiate echos from objects that are only 3.5 inches apart. The digitally controlled-gain, variable-bandwidth amplifier minimizes noise and side-lobe detection in sonar applications.

The module has an accurate ceramic-resonator-controlled 420-kHz time-base generator. An output based on the 420-kHz time base is provided for external use. The sonar transmit output is 16 pulses at a frequency of 40 kHz.

The SN28828 operates over a supply voltage range of from 4.5 V to 6.8 V and is characterized for operation from 0°C to 40°C.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Voltage from any pin to ground (see Note 1)	7 V
Voltage from any pin except XDCR to VCC (see Note 1)	-7 V to 0.5 V
Operating free-air temperature range	0°C to 40°C
Storage temperature range	-40°C to 85°C

NOTE 1: The XDCR pin may be driven to ± 35 V typical with respect to ground.

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Special Functions

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SN28828 SONAR RANGING MODULE

recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, V_{CC}	4.5	6.8	V
High-level input voltage, V_{IH}	BLNK, BINH, INIT		V
Low-level input voltage, V_{IL}	BLNK, BINH, INIT		V
ECHO and OSC output voltage		0.6	V
ECHO and OSC output voltage		6.8	V
Delay time, power up to INIT high	5		ms
Recycle period	80		ms
Operating free-air temperature, T_A	0	40	°C

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP [†]	MAX	UNIT
Input current	BLNK, BINH, INIT			1	mA
High-level output current, I_{OH}	ECHO, OSC			100	μ A
Low-level output voltage, V_{OL}	ECHO, OSC			0.4	V
Transducer output voltage (peak-to-peak)				70	V
Internal blanking interval	See Note 2		2.46		ms
Frequency during 16-pulse transmit period	OSC output			40	kHz
	XMIT output			40	
Frequency after 16-pulse transmit period	OSC output			93.3	kHz
	XMIT output			0	
Supply current, I_{CC}	During transmit period			350	mA
	After transmit period			100	

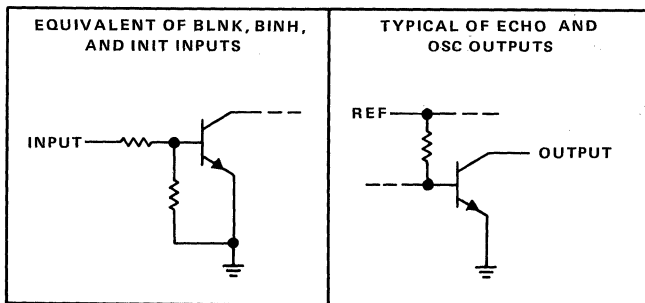
[†]Typical values are at $V_{CC} = 5$ V and $T_A = 25^\circ\text{C}$.

NOTE 2: These typical values apply for a 420-kHz ceramic resonator.

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Special Functions

schematics of inputs and outputs



operation with 40-kHz piezoelectric transducer

There are two basic modes of operation for the SN28828 Sonar ranging module: single-echo mode and multiple-echo mode. The application of power (V_{CC}), the activation of the Initiate (INIT) input and the resulting transmit output, and the use of the Blanking Inhibit (BINH) input are basically the same for either mode of operation. After applying power (V_{CC}), a minimum of 5 ms must elapse before the INIT input can be taken high. During this time, all internal circuitry is reset and the internal oscillator stabilizes. When INIT is taken high, drive to the Transducer (XDCR) output occurs. Sixteen pulses at 40 kHz with 70-V peak-to-peak amplitude will excite the transducer as transmission occurs.

In order to eliminate ringing of the transducer from being detected as a return signal, the Receive (REC) input of the ranging control IC is inhibited by internal blanking for 2.46 ms after the initiate signal. If a reduced blanking time is desired, then the BINH input can be taken high to end the blanking of the Receive input anytime prior to internal blanking. This may be desired to detect objects closer than 1.37 feet corresponding to 2.46 ms and may be done if transducer damping is sufficient that ringing is not detected as a return signal.

In the single-echo mode of operation (Figure 1), all that must be done next is to wait for the return of the transmitted signal, traveling at approximately 0.9 ms per foot out and back. The returning signal is amplified and appears as a high-logic-level echo output. The time between INIT going high and the Echo (ECHO) output going high is proportional to the distance of the target from the transducer. If desired, the cycle can now be repeated by returning INIT to a low-logic level and then taking it high when the next transmission is desired.

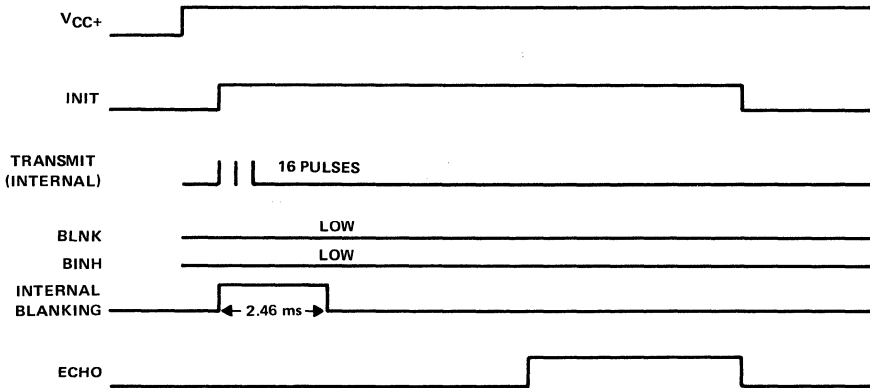


FIGURE 1. EXAMPLE OF A SINGLE-ECHO-MODE CYCLE WITHOUT BLANKING INPUT

If there is more than one target, and multiple echos are to be detected from a single transmission, then the cycle is slightly different (Figure 2). After receiving the first return signal, which causes the ECHO output to go high, the Blanking (BLNK) input must be taken high then back low to reset the Echo output for the next return signal. The blanking signal must be at least 0.52 ms in duration to account for all 16 returning pulses from the first target and allow for internal delay times. This corresponds to the two targets being 3.5 inches apart.

**SN28828
SONAR RANGING MODULE**

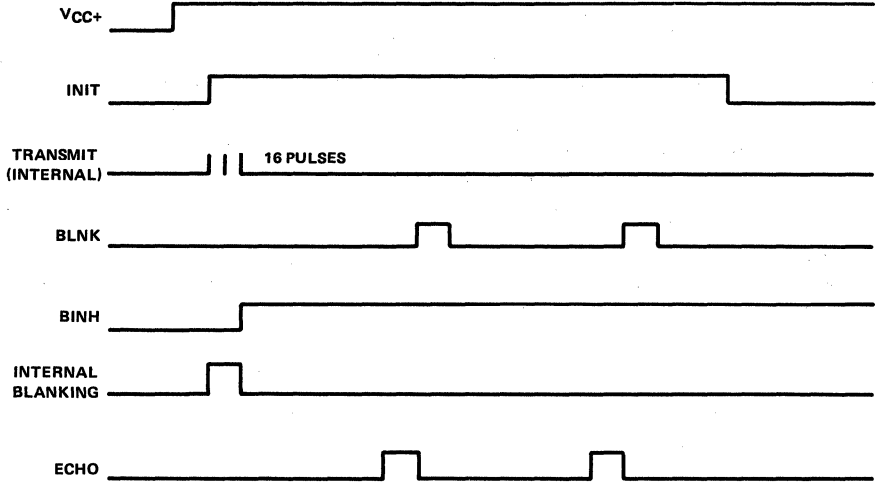


FIGURE 2. EXAMPLE OF A MULTIPLE-ECHO-MODE CYCLE WITH BLANKING INPUT

During a cycle starting with INIT going high, the receiver amplifier gain is incremented higher at discrete times (Figure 3) since the transmitted signal is attenuated with distance. At approximately 38 ms, the maximum gain is attained. For this reason, sufficient gain may not be available for objects greater than

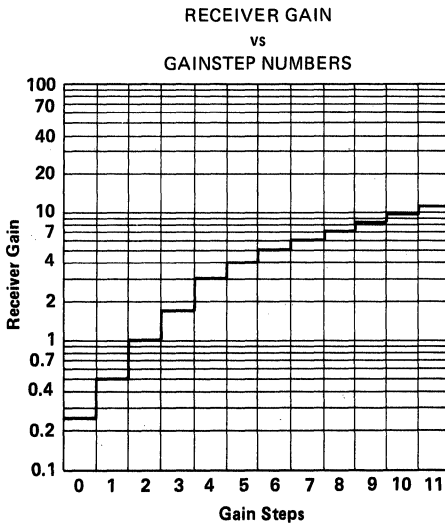


FIGURE 3

4

Special Functions

35 feet away. Although gain can be increased by varying R1 (Figure 4), there is a limit to which the gain can be increased for reliable module operation. This will vary from application to application. The modules are "kitted" prior to their final test during manufacture. This is necessary because the desired gain distribution is much narrower than the module gain distribution if all were kitted with one value resistor. As kitted, these modules will perform satisfactorily in most applications. As a rule of thumb, the gain can be increased by up to a factor of 3, if required, by increasing R1 correspondingly. Gain is directly proportional to R1.

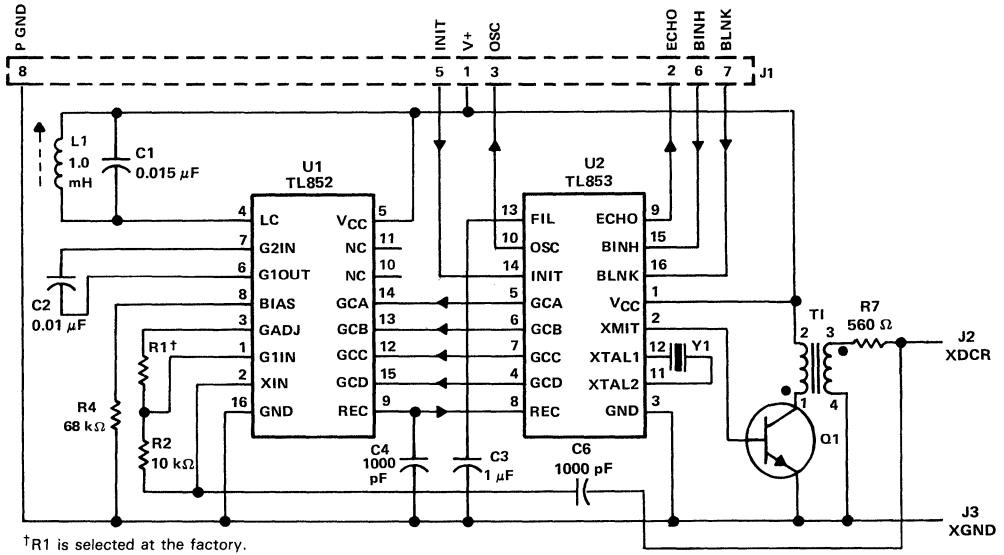
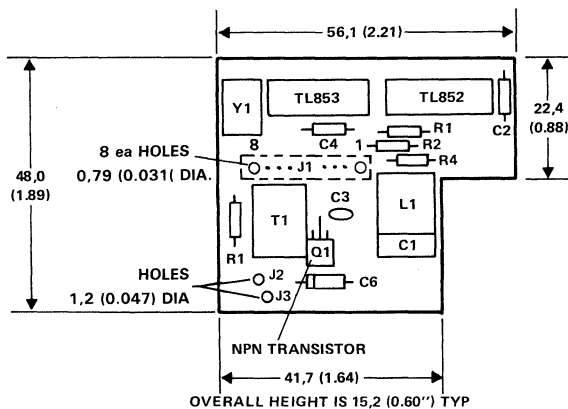


FIGURE 4. SCHEMATIC



NOTE: All dimensions are in millimeters and parenthetically in inches.

FIGURE 5. COMPONENT LAYOUT AND DIMENSIONS OF MODULE

4
Special Functions

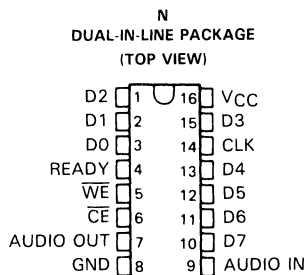
4

Special Functions

SN76494, SN76494A, SN76496, SN76496A PROGRAMMABLE TONE/NOISE GENERATOR

D2801, JUNE 1984 – REVISED JANUARY 1989

- Each Circuit Contains 3 Programmable Tone Generators
- Programmable White-Noise Generator
- Programmable Attenuation
- Simultaneous Sounds
- Up to 500 kHz Clock Input for SN76494 and 4 MHz for SN76496
- External Audio Input for SN76496 May Be Summed with Internally Generated Tones
- The SN76494A and SN76496A are Interchangeable with the SN76494 and SN76496, Respectively



description

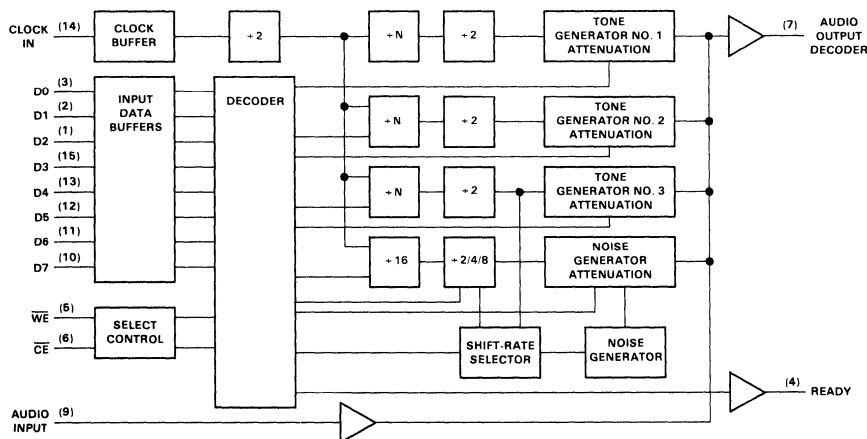
The SN76494 and SN76494A digital complex sound generators are integrated injection logic (I²L) tone generators designed to provide low-cost tone or noise generation capability in microprocessor systems. The SN76494 and SN76494A are data-bus-based input-output peripheral devices that interface the microprocessor through 8 data lines and 3 control lines.

The SN76494 and SN76494A are identical to the SN76496 and SN76496A except that the maximum clock input frequency for SN76494 and SN76494A is 500 kHz and for SN76496 and SN76496A, it is 4 MHz. A "divide-by-eight" stage is deleted from the SN76496 and SN76496A circuitry so that only 4 clock pulses are required to load the data into the SN76494 and SN76494A, compared to 32 pulses for the SN76496 and SN76496A.

Either of these devices may also be used as a replacement for the SN76489A in all applications if pin 9 is left open or grounded. The output load must be limited to 10 mA.

When audio input is not desired in the SN76494, SN76494A, SN76496 or SN76496A, the audio input pin should be grounded.

functional block diagram



PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

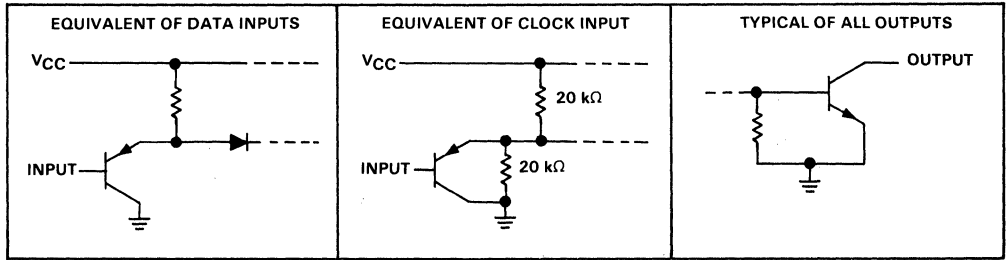

TEXAS
INSTRUMENTS

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SN76494, SN76494A, SN76496, SN76496A PROGRAMMABLE TONE/NOISE GENERATOR

schematics of inputs and outputs



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	7 V
Input voltage, V_I : Audio input	0.9 V
All other inputs	7 V
Output current at pin 7	10 mA
Continuous total power dissipation at (or below) 25°C free-air temperature (see Note 2)	1150 mW
Operating free-air temperature	0°C to 70°C
Storage temperature range	-55°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTES: 1. All voltage values are with respect to network ground terminal.

2. For operation above 25°C free-air temperature, derate linearly to 736 mW at 70°C at the rate of 9.2 mW/°C.

recommended operating conditions

		SN76494, SN76494A			SN76496, SN76496A			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	
V_{CC}	Supply voltage	4.5	5	5.5	4.5	5	5.5	V
V_{IH}	High-level input voltage	2			2			V
V_{IL}	Low-level input voltage	0.8			0.8			V
I_I	Audio input current	0			1.8			mA
V_{OH}	High-level output voltage (pin 4)	5.5			5.5			V
I_{OL}	Low-level output current (pin 4)	2			2			mA
f_{clock}	Input clock frequency	0.5			4			MHz
$t_d(WE)$	Delay time, \overline{CE} low to \overline{WE} low	0			0			ns
t_{su}	Setup time, data before $\overline{WE}\downarrow$ or $\overline{CE}\downarrow$	0			0			ns
t_h	Hold time, data after $READY\uparrow$	0			0			ns
T_A	Operating free-air temperature	70			70			°C

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Special Functions

SN76494, SN76494A, SN76496, SN76496A PROGRAMMABLE TONE/NOISE GENERATOR

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP†	MAX	UNIT	
I_{OH}	High-level output current (pin 4)	$V_O = 5.5\text{ V}$			10	μA	
I_{IH}	High-level input current (All digital inputs)	$V_I = V_{CC}$			10	μA	
I_{IL}	Low-level input current	$V_I = 0$		-25	-175	μA	
				-10	-70		
V_{IB}	Input bias voltage, audio (pin 9)	$R = 4.7\text{ k}\Omega$ to V_{CC}	0.5	0.7	0.9	V	
V_{OH}	High-level output voltage (pin 7)				5.5	V	
V_{OL}	Low-level output voltage (pin 4)	$I_{OL} = 2\text{ mA}$		0.25	0.4	V	
V_{OPP}	Peak-to-peak output voltage (pin 7)	$V_{CC} = 5\text{ V}$, Attenuation: Generator under test = 0 dB All other generators = 30 dB	260			mV	
I_{CC}	Supply current			30	50	mA	
Attenuation		See Table 1	2 dB NOM	1	2	3	dB
			4 dB NOM	3	4	5	
			8 dB NOM	7	8	9	
			16 dB NOM	15	16	17	
C_i	Input capacitance					15	pF

† Typical values are at $V_{CC} = 5\text{ V}$, $T_A = 25^\circ\text{C}$.

switching characteristics, $V_{CC} = 5\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{PHL}	Propagation delay time, high-to-low level RDY output from \overline{CE}		90	150	ns
t_{PHL}	Propagation delay time high-to-low level, RDY output from WE	$C_L = 225\text{ pF}$, $R_L = 2\text{ k}\Omega$ to V_{CC}	90		ns
t_{PLH}	Propagation delay time low-to-high level, RDY output from CLK		90		ns

4

Special Functions

PARAMETER MEASUREMENT INFORMATION

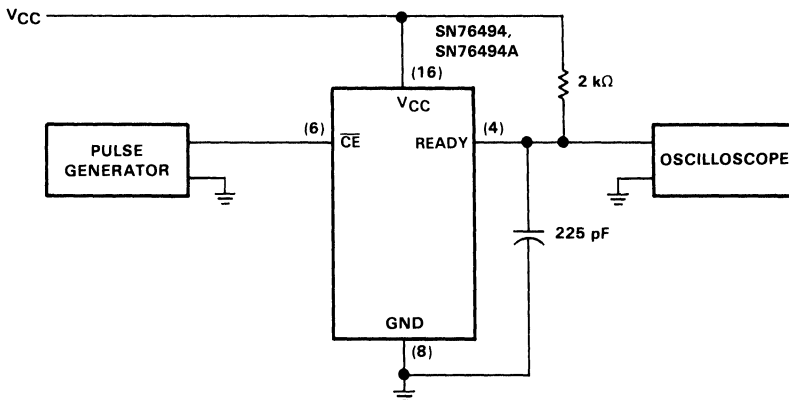


FIGURE 1. t_{pHL} TEST CIRCUIT

SN76494, SN76494A, SN76496, SN76496A PROGRAMMABLE TONE/NOISE GENERATOR

pin assignments and functions

SIGNATURE	PIN	I/O	DESCRIPTION
\overline{CE}	6	I	Chip Enable. When chip enable is low, the device is operational, input terminals are enabled, and data may be entered.
D0 (MSB)	3	I	D0 through D7 – Input data bus
D1	2	I	
D2	1	I	
D3	15	I	
D4	13	I	
D5	12	I	
D6	11	I	
D7 (LSB)	10	I	
VCC	16	I	Supply voltage (5 V nom)
GND	8	O	Ground reference
CLOCK	14	I	Input Clock
\overline{WE}	5	I	Write Enable. When \overline{CE} is enabled and \overline{WE} is active (low), input on the data bus is accepted. \overline{CE} and \overline{WE} must be held low until READY returns high (four clock cycles for the SN76494 and SN76494A or 32 clock cycles for the SN76496 and SN76496A). If \overline{WE} remains low throughout four additional clock cycles for the SN76494 and SN76494A (32 clock cycles for the SN76496 and SN76496A) a new write cycle will be initiated.
READY	4	O	When low, READY indicates that a write cycle is in progress; data on the input bus must remain valid until READY returns high.
AUDIO IN	9	I	Audio input from external source
AUDIO OUT	7	O	Audio Drive Out

PRINCIPLES OF OPERATION

tone generators

Each tone generator consists of a frequency synthesis section and an attenuation section. The frequency synthesis section requires 10 bits of information (F0-F9) to define half the period of the desired frequency (f). F0 is the most significant bit and F9 is the least significant bit. This information is loaded into a 10-stage tone counter, which counts down at an N/2 rate where N is the input clock frequency. When the tone counter counts down to zero, a borrow signal is produced. This borrow signal toggles the frequency flip-flop and also reloads the tone counter. Thus, the period of the desired frequency is twice the value of the period register.

The frequency can be calculated by the following:

$$f = \frac{N}{4n} \text{ for SN76494 and SN76494A, or } f = \frac{N}{32n} \text{ for SN76496 and SN76496A}$$

where N = clock in Hz
n = 10-bit binary number

The output level of each tone/noise generator may be selected by programming a four stage attenuator. The attenuator values, along with their bit position in the data word, are shown in Table 1. Multiple attenuation control bits may be true simultaneously. Thus, the maximum attenuation is 30 dB.

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Special Functions

SN76494, SN76494A, SN76496, SN76496A PROGRAMMABLE TONE/NOISE GENERATOR

TABLE 1. ATTENUATION CONTROL

BIT POSITION				WEIGHT (In dB)
A0	A1	A2	A3	
0	0	0	1	2
0	0	1	0	4
0	1	0	0	8
1	0	0	0	16
1	1	1	1	OFF

noise generator

The noise generator consists of a noise source and an attenuator. The noise source is a shift register with an exclusive OR-feedback network. The feedback network has provisions to protect the shift register from being locked in the zero state.

TABLE 2. NOISE FEEDBACK CONTROL

FEEDBACK	CONFIGURATION
0	"Periodic" noise
1	"White" noise

Whenever the noise control register is changed, the shift register is cleared. The shift register will shift at one of four rates as determined by the two NF bits. The fixed shift rates are derived from the input clock.

TABLE 3. NOISE GENERATOR FREQUENCY CONTROL

BITS		SHIFT RATE
NF0	NF1	
0	0	N/64
0	1	N/128
1	0	N/256
1	1	Tone generator #3 output

The output of the noise source is connected to a programmable attenuator as shown in Figure 4.

output buffer/amplifier

The output buffer is a conventional operational amplifier summing circuit. It sums the three tone generator outputs, the noise generator output, and any audio input through pin 9. The output buffer will generate up to 10 mA.

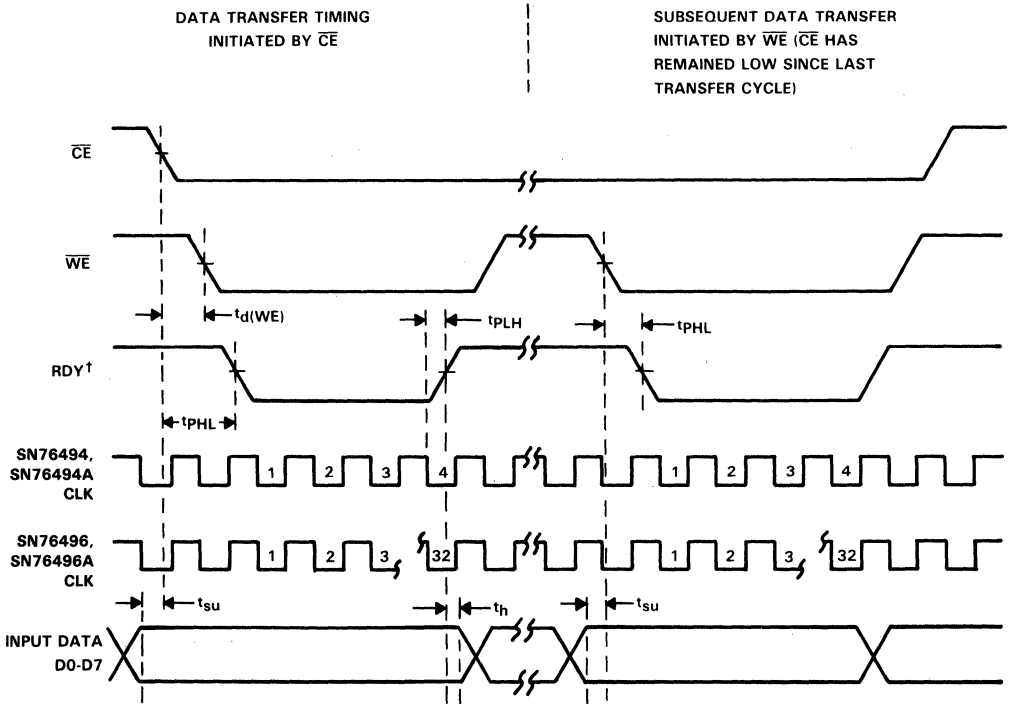
To prevent oscillations in the output buffer, the output (pin 7) should be decoupled. This is done by putting 10 ohms in series with 0.1 μ F from pin 7 to ground (see Figure 3).

data transfer

The microprocessor selects the SN76494, SN76494A, SN76496, or SN76496A by taking \overline{CE} low (low voltage). Unless \overline{CE} is low, no data transfer can occur. When \overline{CE} is low, the \overline{WE} signal strobes the contents of the data bus to the appropriate control register. The data bus contents must be valid at this time.

SN76494, SN76494A, SN76496, SN76496A PROGRAMMABLE TONE/NOISE GENERATOR

The SN76494 and SN76494A require approximately 4 clock cycles to load the data into the control register. The SN76496 and SN76496A require approximately 32 clock cycles. The open-collector READY output is used to synchronize the microprocessor to this transfer and is pulled to the false state (low) immediately following the falling edge of \overline{CE} (or \overline{WE} when data transfer is initiated by \overline{WE}). READY will go high upon completion of the data transfer cycle. The data transfer timing is shown below.



† \overline{WE} must be returned high (inactive) within 4 clock pulses for the SN76494 and SN76494A (32 clock pulses for the SN76496 and SN76496A) after RDY returns high. Otherwise, a new data transfer cycle will be initiated.

FIGURE 2. DATA TRANSFER TIMING

TABLE 4. FUNCTION TABLE

INPUTS		OUTPUT
\overline{CE}	\overline{WE}	READY
L	L	L
L	H	L
H	L	H
H	H	H

This table is valid when the device is:

- (1) not being clocked, and
- (2) is initialized by pulling \overline{WE} and \overline{CE} high.

SN76494, SN76494A, SN76496, SN76496A PROGRAMMABLE TONE/NOISE GENERATOR

CPU interface to SN76494, SN76494A, SN76496 or SN76496A

The microprocessor interfaces with the SN76494, SN76494A, SN76496, or SN76496A by means of the 8 data lines and 3 control lines (\overline{WE} , \overline{CE} and \overline{READY}). Each tone generator requires 10 bits of information to select the frequency and four bits of information to select the attenuation. A frequency selection requires a double-byte transfer, while an attenuator selection requires a single-byte transfer.

If no other control registers on the chip are accessed, a tone generator may be rapidly updated by initially sending both bytes of frequency and register data, followed by just the second byte of data for succeeding values. The register address is latched on the chip, so the data will continue going into the same register. This allows the six most significant bits to be quickly modified for frequency sweeps.

control registers

The devices have 8 internal registers that are used to control the 3 tone generators and the noise source. During all data transfers to the devices, the first byte contains a 3-bit field that determines the destination control register. The register address codes are shown in Table 5.

TABLE 5. REGISTER ADDRESS FIELD

R0	R1	R2	DESTINATION CONTROL REGISTER
0	0	0	Tone 1 Frequency
0	0	1	Tone 1 Attenuation
0	1	0	Tone 2 Frequency
0	1	1	Tone 2 Attenuation
1	0	0	Tone 3 Frequency
1	0	1	Tone 3 Attenuation
1	1	0	Noise Control
1	1	1	Noise Attenuation

data formats

The formats required to transfer data are shown below.

REG ADDR				DATA			
1	R0	R1	R2	F6	F7	F8	F9
BIT 0 FIRST BYTE				BIT 7			

DATA							
0	x	F0	F1	F2	F3	F4	F5
BIT 0		SECOND BYTE					BIT 7

UPDATE NOISE SOURCE (SINGLE BYTE TRANSFER)

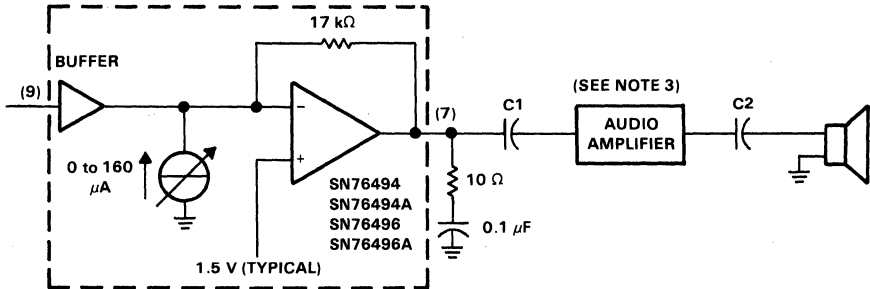
REG ADDR				SHIFT			
1	R0	R1	R2	x	FB	NF0	NF1
BIT 0				BIT 7			

UPDATE ATTENUATOR (SINGLE BYTE TRANSFER)

REG ADDR				DATA			
1	R0	R1	R2	A0	A1	A2	A3
BIT 0				BIT 7			

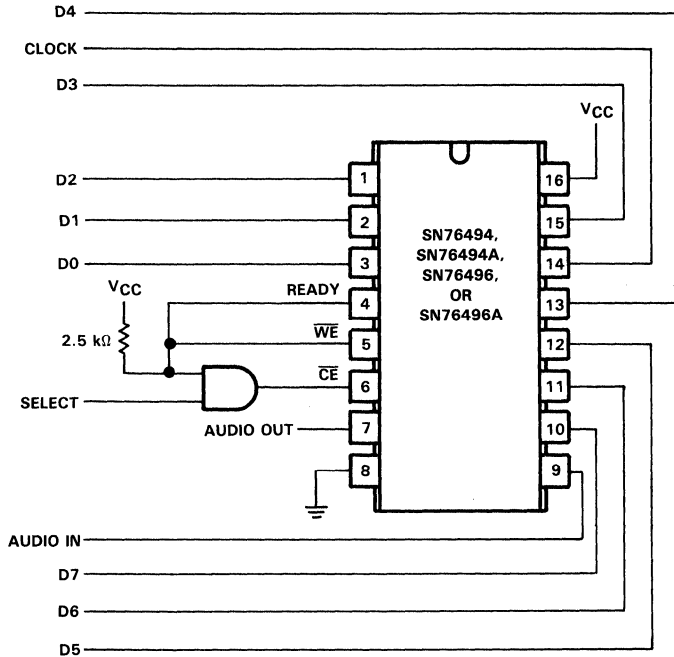
SN76494, SN76494A, SN76496, SN76496A PROGRAMMABLE TONE/NOISE GENERATOR

TYPICAL APPLICATION DATA



NOTE 3: The capacitance values of C1 and C2 are determined by the frequency response desired and the audio amplifier used.

FIGURE 3. EXTERNAL AUDIO OUTPUT INTERFACE



- NOTES: 4. The data lines must be latched so that the data remains on them at least 32 clock cycles for the SN76496 and SN76496A or (4 clock cycles for the SN76494 and SN76494A) after the select line goes low.
5. The select pulse should be a negative-going pulse with minimum duration of 150 ns.

FIGURE 4. MICROCOMPUTER PARALLEL PORT INTERFACE

4

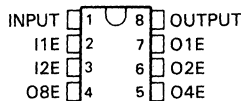
Special Functions

TL010I, TL010C ADJUSTABLE-RATIO CURRENT MIRRORS

D2738, SEPTEMBER 1983—REVISED APRIL 1988

- 33 Distinct Input-to-Output Emitter Ratios from 3:1 to 1:15
- Wide Input current Range: 1 μ A to 3 mA
- 35-Volt Output Capability
- High Output Impedance

P DUAL-IN-LINE PACKAGE
(TOP VIEW)



description

The TL010 is a Wilson current mirror that provides output current in a selectable fixed ratio to the input current. The ratio is substantially independent of changes in load, voltages, and temperature. Selecting the ratio consists of connecting appropriate input-emitter pins and output-emitter pins to ground as shown in Figure 1.

The TL010 is designed to operate with up to 3 mA input current if all three input-emitter pins are used. It will also operate at voltages up to 35 V.

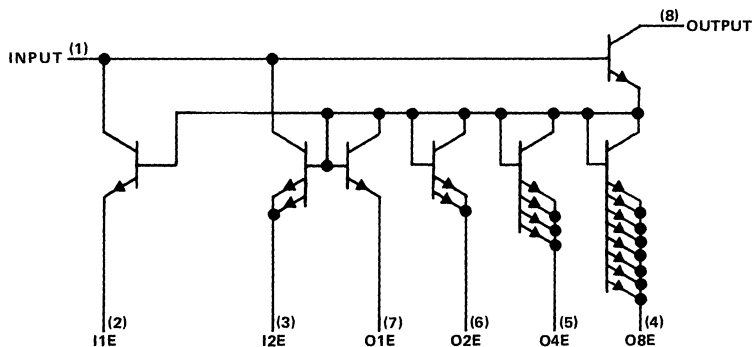
The TL010I is characterized for operation from -40°C to 85°C . The TL010C is characterized for operation from 0°C to 70°C .

typical values of current ratio at $T_A = 25^{\circ}\text{C}^{\dagger}$

EMITTER RATIO $m:n^{\dagger}$	CURRENT RATIO $h_F = I_O/I_I$	EMITTER RATIO $m:n^{\dagger}$	CURRENT RATIO $h_F = I_O/I_I$	EMITTER RATIO $m:n^{\dagger}$	CURRENT RATIO $h_F = I_O/I_I$
1:15	14.1	1:6	5.78	3:8	2.61
1:14	13.2	2:11	5.34	2:5	2.43
1:13	12.3	1:5	4.82	3:7	2.26
1:12	11.4	3:14	4.53	1:2	1.98
1:11	10.5	2:9	4.38	3:5	1.64
1:10	9.55	3:13	4.21	2:3	1.45
1:9	8.62	1:4	3.89	3:4	1.32
1:8	7.72	3:11	3.57	1:1	0.99
2:15	7.23	2:7	3.40	3:2	0.663
1:7	6:71	3:10	3:25	2:1	0.50
2:13	6.29	1:3	2.90	3:1	0.332

$^{\dagger}m$ is the number of input emitters used, n is the number of output emitters used.

schematic



TL010I, TL010C ADJUSTABLE-RATIO CURRENT MIRRORS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Output voltage (see Note 1)	45 V
Input current	5 mA
Continuous total dissipation at (or below) 25°C free-air temperature (see Note 2)	725 mW
Operating free-air temperature range: TL010I	-40°C to 85°C
TL010C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTES: 1. Input and output voltages are with respect to the common terminal. Neither voltage should be more negative than -0.3 V.
2. For operation above 25°C free-air temperature, derate linearly at the rate of 8.0 mW/°C.

recommended operating conditions

	TL010I		TL010C		UNIT
	MIN	MAX	MIN	MAX	
Output voltage, V_O	5	35	5	35	V
Input voltage, V_I	0.6	1.7	0.65	1.6	V
Input current per input emitter, I_I	0.001	1	0.001	1	mA
Operating free-air temperature, T_A	-40	85	0	70	°C

electrical characteristics over recommended ranges of operating free-air temperature and output voltage (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TL010I			TL010C			UNIT	
		MIN	TYP‡	MAX	MIN	TYP‡	MAX		
V_I Input voltage	$I_I = m \times 1 \mu\text{A}$	1			1			V	
		1.1			1.1				
		1.25			1.25				
		1.4			1.4				
h_F Current ratio (I_O/I_I)	$I_I = \text{MIN to MAX}$	$m:n = 1:8$	6.97	7.72	8.13	7.05	7.72	8.13	V
		$m:n = 1:4$	3.61	3.89	4.05	3.64	3.89	4.05	
		$m:n = 1:2$	1.84	1.98	2.07	1.88	1.98	2.07	
		$m:n = 1:1$	0.89	0.99	1.08	0.94	0.99	1.04	
		$m:n = 2:1$	0.46	0.50	0.56	0.475	0.50	0.525	
αh_F Temperature coefficient of current ratio	$I_I = \text{MIN to MAX}$	300			300			ppm/°C	
Output-to-input isolation	$I_I = \text{MIN to MAX}, f = 1 \text{ kHz}$	60			60			dB	
$V_{O(th)}$ Output threshold voltage§	$I_I = \text{MIN to MAX}$	$T_A = \text{MIN}$	1.1			1.05			V
		$T_A = 25^\circ\text{C}$	1			1			
r_o Output resistance¶	$F = 1 \text{ kHz}$	$I_I = m \times 10 \mu\text{A}$	200 m/n			200 m/n			MΩ
		$I_I = m \times 100 \mu\text{A}$	20 m/n			20 m/n			
		$I_I = m \times 1 \text{ mA}$	2 m/n			2 m/n			
f_{max} Maximum operating frequency#	$I_I = m \times 1 \text{ mA}, R_L = 500 \Omega$	10			10			MHz	

† m is the number of input emitters, n is the number of output emitters. For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ All typical values are at $T_A = 25^\circ\text{C}$.

§ Output threshold voltage is the voltage at which the current ratio is equal to 90% of its value at $V_O = 15 \text{ V}$.

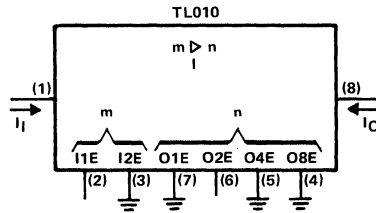
¶ The output resistance is directly proportional to the number of input emitters divided by the number of output emitters (m/n).

Maximum operating frequency is the frequency at which the output current is down 3 dB from its low-frequency value.

4

Special Functions

TYPICAL APPLICATION INFORMATION



See Notes 3 and 4

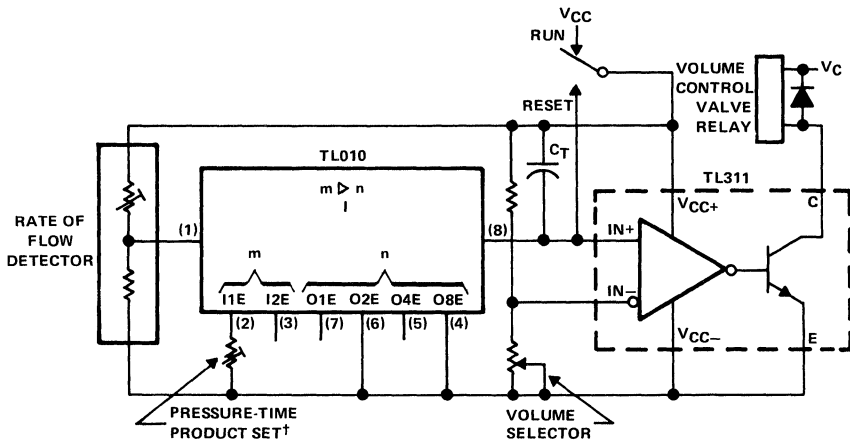
- NOTES: 3. Selected emitters must be grounded as close as possible to the package to avoid unstable device behavior. Using the fixed-Beta model, the current ratio for a current mirror of m input emitters and n output emitters may be calculated as

$$\frac{I_O}{I_I} = \frac{\beta^2 n + \beta(n + m)}{\beta^2 m + (\beta + 1)(m + n)}$$

Second-order effects, such as on-chip self-heating, may slightly perturb the observed ratio from the calculated value.

4. At high current levels, a small capacitor (270 pF) may be required between the input and output terminals to improve stability.

FIGURE 1. CURRENT MIRROR SET FOR A CURRENT RATIO OF 2:13



† Adjust for a mirror of 11.9

FIGURE 2. TYPICAL APPLICATION CIRCUIT

In the application shown in Figure 2, the problem is to measure a precise volume of liquid flowing through a line and shut off the flow with a relay when limit is reached. For the particular volume to be measured and the pressure detector used, a current gain of 11.9 is required. By setting the TL010 for a gain of 10 with the emitter selection, the exact gain of 11.9 may be obtained by adjusting the pressure-time product control.

4

Special Functions

SERIES TL011, TL012, TL014A, TL021 FIXED-RATIO N-P-N CURRENT MIRRORS

D2614, FEBRUARY 1984—REVISED OCTOBER 1988

- **Wide Input Current Range:**
1 μ A to 1 mA
- **35-Volt Output Capability**
- **High Output Impedance**
- **Current-Ratio Tolerances Over Full Temperature Range;**
± 8% for I Suffix
± 7% for C Suffix
- **Typically Less Than ± 1% Error at 25 °C**

LP PACKAGE
(TOP VIEW)



TEMPERATURE RANGE	INPUT-TO-OUTPUT CURRENT RATIO			
	1:1	1:2	1:4	2:1
-40 °C to 85 °C	TL011I	TL012I		TL021I
0 °C to 70 °C	TL011C	TL012C	TL014AC	TL021C

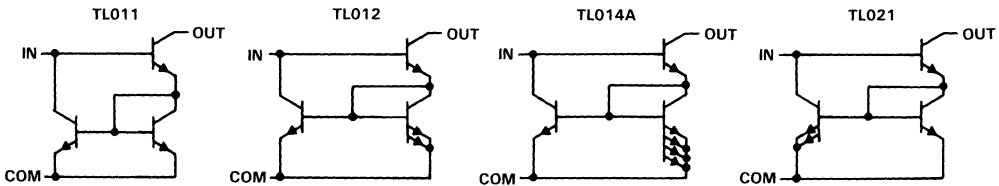
description

The TL011, TL012, TL014A, and TL021 are Wilson current mirrors with output currents in fixed proportion to the input currents and substantially independent of changes in voltage, load, and temperature. These devices make use of the tight matching properties of identical bipolar transistors on a monolithic integrated circuit chip to achieve current-ratio accuracy typically better than 98%.

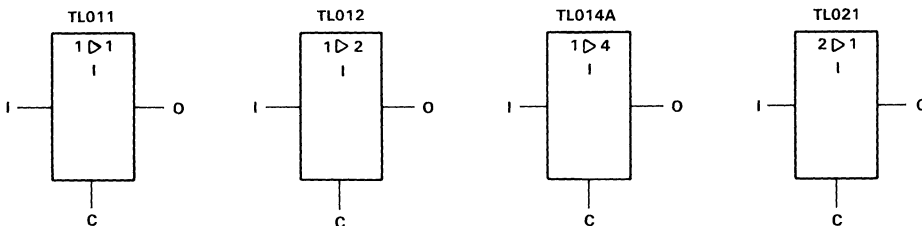
Current mirrors are used extensively in linear integrated circuit designs as active loads for operational-amplifier stages and as current sources for other stages. The TL011 family gives the designer this same capability with no sacrifice in accuracy or stability.

The TL011, TL012, and TL014A are designed to operate with input currents up to 1 mA and output voltage up to 35 V. The TL021 is designed for 2 mA and 35 V.

schematics



symbols



4

Special Functions

**SERIES TL011, TL012, TL014A, TL021
FIXED-RATIO N-P-N CURRENT MIRRORS**

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Output voltage (see Note 1)	45 V
Input current	5 mA
Continuous total dissipation at (or below) 25°C free-air temperature (see Note 2)	775 mW
Operating free-air temperature range: TL011I, TL012I, TL021I	-40°C to 85°C
TL011C, TL012C, TL014AC, TL021C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. Input and output voltages are with respect to the common terminal. Neither voltage should be more negative than -0.3 V.
 2. For operation above 25°C free-air temperature, derate linearly at the rate of 6.2 mW/°C. The LP package dissipation rating was based on thermal resistance, $R_{\theta JA}$, measured in still air with the device mounted in an Augat socket. The bottom of the package was 10 mm (0.375 in.) above the socket.

recommended operating conditions

		TL0__I		TL0__C, AC		UNIT
		MIN	MAX	MIN	MAX	
Output voltage, V_O		5	35	5	35	V
Input current, I_O	TL021	0.002	2	0.002	2	mA
	All others	0.001	1	0.001	1	
Operating free-air temperature, T_A		-40	85	0	70	°C

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Special Functions

electrical characteristics over recommended ranges of operating free-air temperature and output voltage (unless otherwise noted)

PARAMETER		TEST CONDITIONS		TL011			TL012			TL014A			TL021			UNIT		
				MIN	TYP [†]	MAX	MIN	TYP [†]	MAX	MIN	TYP [†]	MAX	MIN	TYP [†]	MAX			
V _I	Input voltage	I _I = 1 μA		1			1			1			1			V		
		I _I = 2 μA											1					
		I _I = 10 μA		1.1			1.1			1.1								
		I _I = 20 μA											1.1					
		I _I = 100 μA		1.25			1.25			1.25								
		I _I = 200 μA											1.25					
		I _I = 1 mA		1.4			1.4			1.4								
I _I = 2 mA											1.4							
h _F	Current ratio (I _O /I _I)	TLO_I	I _I = MIN to MAX [‡]		0.92	1	1.08	1.84	2	2.16	3.68	4	4.32	0.46	0.5	0.54		
		TLO_C, AC			0.93	1	1.07	1.86	2	2.14	3.72	4	4.28	0.465	0.5	0.535		
α _{hF}	Temperature coefficient of current ratio	I _I = MIN to MAX		50			100			200			200			ppm/°C		
Output-to-input isolation		I _I = MIN to MAX, f = 1 kHz		80			80			80			80			dB		
V _{O(th)}	Output threshold voltage [§]	TLO_I	I _I = MIN to MAX		T _A = -40°C			1.35			1.35			1.35			V	
		TLO_C, AC			T _A = 0°C			1.25			1.25			1.25				
		All			T _A = 25°C			1.2			1.2			1.2				
r _o	Output resistance	f = 1 kHz		I _I = 10 μA		200			100			50						MΩ
				I _I = 20 μA								200						
				I _I = 100 μA		20			10			5						
				I _I = 200 μA											20			
				I _I = 1 mA		2			1			0.5						
				I _I = 2 mA											2			
f _{max}	Maximum operating frequency [¶]	I _I = MAX		R _L = 500 Ω		10			10			10			MHz			

[†]All typical values are at T_A = 25°C.

[‡]For test conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

[§]Output threshold voltage is the voltage at which the current ratio is equal to 90% of its value at V_O = 15 V.

[¶]Maximum operating frequency is the frequency at which the output current is down 3 dB from its low frequency value.

SERIES TL011, TL012, TL014A, TL021
FIXED-RATIO N-P-N CURRENT MIRRORS

Special Functions

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**SERIES TL011, TL012, TL014A, TL021
FIXED-RATIO N-P-N CURRENT MIRRORS**

TYPICAL CHARACTERISTICS

TL011
CURRENT RATIO
vs
FREE-AIR TEMPERATURE

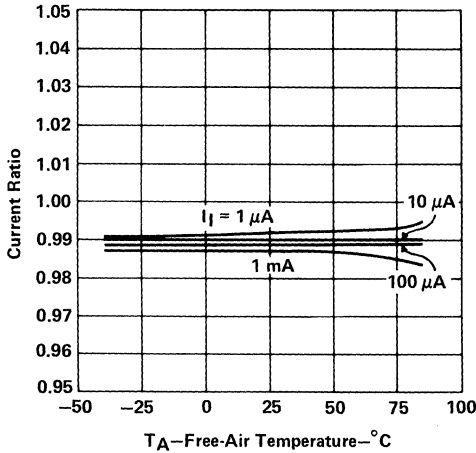


FIGURE 1

TL012
CURRENT RATIO
vs
FREE-AIR TEMPERATURE

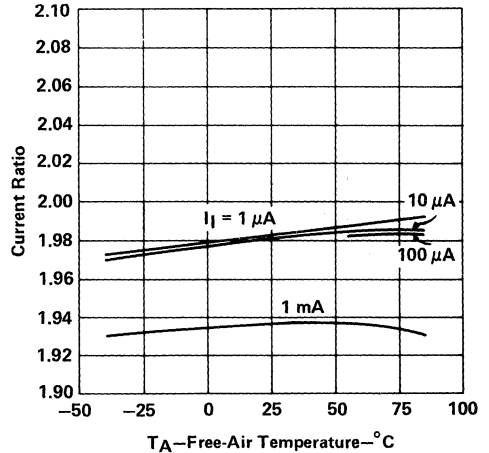


FIGURE 2

TL014A
CURRENT RATIO
vs
FREE-AIR TEMPERATURE

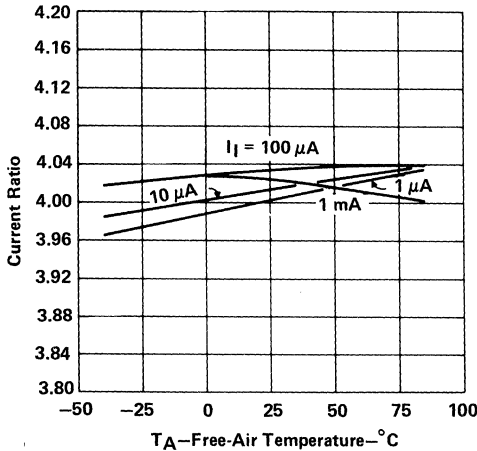


FIGURE 3

TL021
CURRENT RATIO
vs
FREE-AIR TEMPERATURE

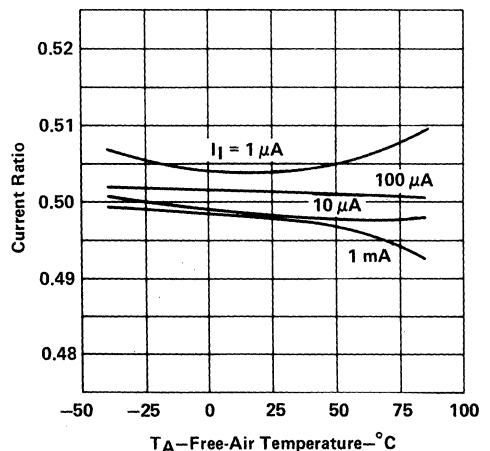


FIGURE 4

4

Special Functions

SERIES TL011, TL012, TL014A, TL021 FIXED-RATIO N-P-N CURRENT MIRRORS

TYPICAL APPLICATIONS INFORMATION

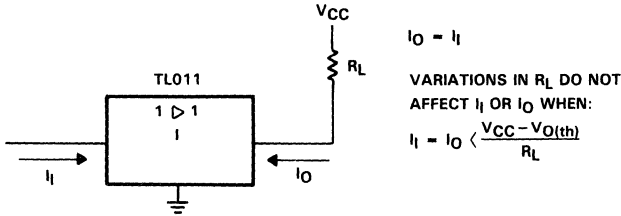
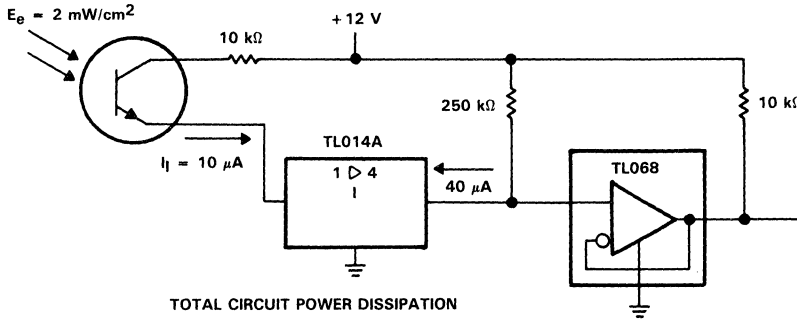


FIGURE 5. BASIC CURRENT BUFFER



Idle condition: $P_D = 1.5 \text{ mW}$ typical
 On condition: $P_D = 12.5 \text{ mW}$ typical
 $10 \mu\text{A}$ from phototransistor provides a V_O swing of 10 V at 1 mA.

FIGURE 6. PHOTOTRANSISTOR PREAMPLIFIER

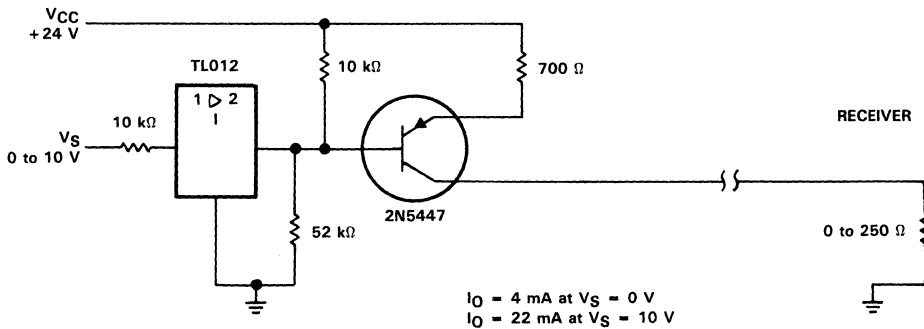


FIGURE 7. TWO-WIRE LINEAR CURRENT-MODE TRANSMITTER

4

Special Functions

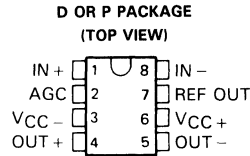
4

Special Functions

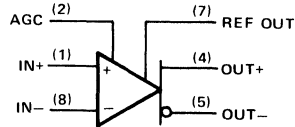
TL026C DIFFERENTIAL VIDEO AMPLIFIER WITH AGC

D2790, JUNE 1985—REVISED OCTOBER 1988

- Low Output Common-Mode Sensitivity to AGC Voltages
- Input and Output Impedances Independent of AGC Voltage
- Peak Gain . . . 38 dB Typ
- Wide AGC Range . . . 50 dB Typ
- 3-dB Bandwidth . . . 50 MHz
- Other Characteristics Similar to NE592 and μ A733



symbol



description

This device is a monolithic two-stage video amplifier with differential inputs and outputs.

Internal feedback provides wide bandwidth, low phase distortion, and excellent gain stability. Variable gain based on signal summation provides large AGC control over a wide bandwidth with low harmonic distortion. Emitter-follower outputs enable the device to drive capacitive loads. All stages are current-source biased to obtain high common-mode and supply-voltage rejection ratios. The gain may be electronically attenuated by applying a control voltage to the AGC pin. No external frequency compensation components are required.

This device is particularly useful in TV and radio IF and RF AGC circuits, as well as magnetic-tape and disk-file systems where AGC is needed. Other applications include video and pulse amplifiers where a large AGC range, wide bandwidth, low phase shift, and excellent gain stability are required.

The TL026C is characterized for operation from 0°C to 70°C.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	8 V
Supply voltage, V_{CC-} (see Note 1)	-8 V
Differential input voltage	± 5 V
Common-mode input voltage	± 6 V
Output current	10 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltages are with respect to the midpoint between V_{CC+} and V_{CC-} except differential input and output voltages.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$
	POWER RATING	DERATING FACTOR	POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW

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Special Functions

TLO26C

DIFFERENTIAL VIDEO AMPLIFIER WITH AGC

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{CC+}	3	6	8	V
Supply voltage, V_{CC-}	-3	-6	-8	V
Operating free-air temperature, T_A	0		70	°C

electrical characteristics at 25 °C operating free-air temperature, $V_{CC\pm} = \pm 6\text{ V}$, $V_{AGC} = 0$, REF OUT pin open (unless otherwise specified)

PARAMETER	TEST FIGURE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
A_{VD} Large-signal differential voltage amplification	1	$V_{O(PP)} = 3\text{ V}$, $R_L = 2\text{ k}\Omega$	65	85	105	V/V
ΔA_{VD} Change in voltage amplification	1	$V_{IPP} = 28.5\text{ mV}$, $R_L = 2\text{ k}\Omega$, $V_{AGC} - V_{ref} = \pm 180\text{ mV}$		-50		dB
V_{ref} Voltage at REF OUT		$I_{ref} = -1\text{ mA}$ to $100\text{ }\mu\text{A}$	1.3		1.5	V
BW Bandwidth (-3 dB)	2	$V_{O(PP)} = 1\text{ V}$, $V_{AGC} - V_{ref} = \pm 180\text{ mV}$		50		MHz
I_{IQ} Input offset current				0.4	5	μA
I_{IB} Input bias current				10	30	μA
V_{ICR} Common-mode input voltage range	3		± 1			V
V_{OC} Common-mode output voltage	1	$R_L = \infty$	3.25	3.75	4.25	V
ΔV_{OC} Change in common-mode output voltage	1	$V_{AGC} = 0$ to 2 V , $R_L = \infty$			300	mV
V_{OO} Output offset voltage	1	$V_{ID} = 0$, $R_L = \infty$			0.75	V
$V_{O(PP)}$ Maximum peak-to-peak output voltage swing	1	$R_L = 2\text{ k}\Omega$	3	4		V
r_i Input resistance at AGC, IN+, or IN-			10	30		k Ω
r_o Output resistance				20		Ω
CMRR Common-mode rejection ratio	3	$V_{IC} = \pm 1\text{ V}$, $f = 100\text{ kHz}$	60	86		dB
	3	$V_{IC} = \pm 1\text{ V}$, $f = 5\text{ MHz}$		60		
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IQ}$)	4	$\Delta V_{CC+} = \pm 0.5\text{ V}$, $\Delta V_{CC-} = \pm 0.5\text{ V}$	50	70		dB
V_n Broadband equivalent noise voltage	4	BW = 1 kHz to 10 MHz		12		μV
t_{pd} Propagation delay time	2	$\Delta V_O = 1\text{ V}$		6	10	ns
t_r Rise time	2	$\Delta V_O = 1\text{ V}$		4.5	12	ns
$I_{sink(max)}$ Maximum output sink current		$V_{ID} = 1\text{ V}$, $V_O = 3\text{ V}$	3	4		mA
I_{CC} Supply current		No load, No signal	22	27		mA

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Special Functions

TL026C DIFFERENTIAL VIDEO AMPLIFIER WITH AGC

electrical characteristics over recommended operating free-air temperature range, $V_{CC\pm} = \pm 6\text{ V}$, $V_{AGC} = 0$, REF OUT pin open (unless otherwise specified)

PARAMETER	TEST FIGURE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
A_{VD}	1	$V_{O(PP)} = 3\text{ V}$, $R_L = 2\text{ k}\Omega$	55	115		V/V
I_{IO}				6		μA
I_{IB}				40		μA
V_{ICR}	3		± 1			V
V_{OO}	1	$V_{ID} = 0$, $R_L = \infty$		1.5		V
$V_{O(PP)}$	1	$R_L = 2\text{ k}\Omega$	2.8			V
r_i				8		$\text{k}\Omega$
CMRR	3	$V_{IC} = \pm 1\text{ V}$, $f = 100\text{ kHz}$	50			dB
k_{SVR}	4	$\Delta V_{CC+} = \pm 0.5\text{ V}$, $\Delta V_{CC-} = \pm 0.5\text{ V}$	50			dB
$I_{sink(max)}$		$V_{ID} = 1\text{ V}$, $V_O = 3\text{ V}$	2.8	4		mA
I_{CC}	1	No load, No signal			30	mA

PARAMETER MEASUREMENT INFORMATION

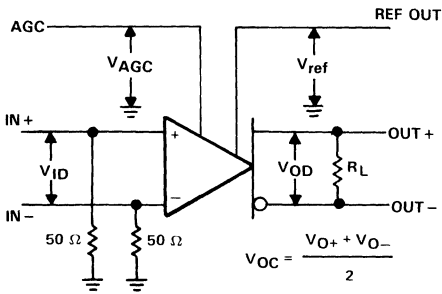


FIGURE 1

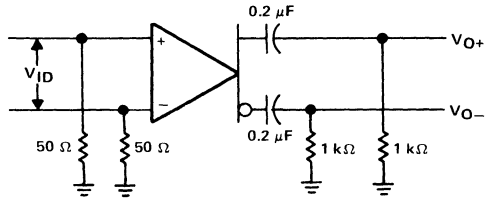


FIGURE 2

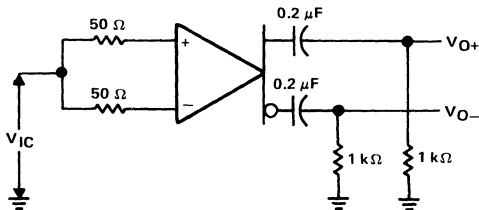


FIGURE 3

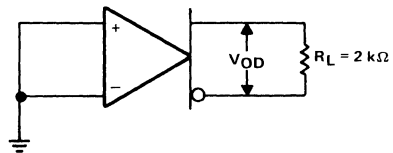


FIGURE 4

4

Special Functions

TL026C
DIFFERENTIAL VIDEO AMPLIFIER WITH AGC

TYPICAL CHARACTERISTICS

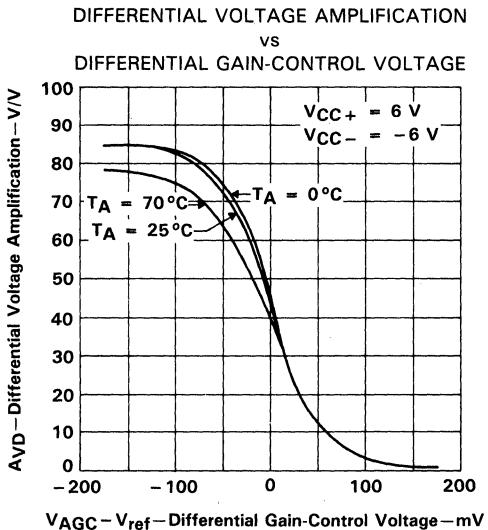


FIGURE 5

4

Special Functions

TYPICAL APPLICATION INFORMATION

gain characteristics

Figure 5 shows the differential voltage amplification versus the differential gain-control voltage ($V_{AGC} - V_{ref}$). V_{AGC} is the absolute voltage applied to the AGC input and V_{ref} is the dc voltage at the REF OUT output. As V_{AGC} increases with respect to V_{ref} , the TL026C gain changes from maximum to minimum. As shown in Figure 5 for example, V_{AGC} would have to vary from approximately 180 mV less than V_{ref} to approximately 180 mV greater than V_{ref} to change the gain from maximum to minimum. The total signal change in V_{AGC} is defined by the following equation.

$$\begin{aligned}\Delta V_{AGC} &= V_{ref} + 180 \text{ mV} - (V_{ref} - 180 \text{ mV}) & (1) \\ \Delta V_{AGC} &= 360 \text{ mV}\end{aligned}$$

However, because V_{AGC} varies as the ac AGC signal varies and also differentially around V_{ref} , then V_{AGC} should have an ac signal component and a dc component. To preserve the dc and thermal tracking of the device, this dc voltage must be generated from V_{ref} . To apply proper bias to the AGC input, the external circuit used to generate V_{AGC} must combine these two voltages. Figures 6 and 7 show two circuits that will perform this operation and are easy to implement. The circuits use a standard dual operational amplifier for AGC feedback. By providing rectification and the required feedback gain, these circuits are also complete AGC systems.

circuit operation

Amplifier A1 amplifies and inverts the rectified and filtered AGC signal voltage V_C producing output voltage V_1 . Amplifier A2 is a differential amplifier that inverts V_1 again and adds the scaled V_{ref} voltage. This conditioning makes V_{AGC} the sum of the signal plus the scaled V_{ref} . As the signal voltage increases, V_{AGC} increases and the gain of the TLC026C is reduced. This maintains a constant output level.

feedback circuit equations

Following the AGC input signal (Figures 6 and 7) from the OUT output through the feedback amplifiers to the AGC input produces the following equation:

1. AC output to diode D1, assuming sinusoidal signals

$$V_O = V_{OP} (\sin (wt)) \quad (2)$$

where:

$$V_{OP} = \text{peak voltage of } V_O$$

2. Diode D1 and capacitor C1 output

$$V_C = V_{OP} - V_F \quad (3)$$

where:

$$V_F = \text{forward voltage drop of D1}$$

$$V_C = \text{voltage across capacitor C1}$$

3. A1 output

$$V_1 = -\frac{R_2}{R_1} V_C \quad (4)$$

TL026C

DIFFERENTIAL VIDEO AMPLIFIER WITH AGC

TYPICAL APPLICATION INFORMATION

4. A2 output (R3 = R4)

$$V_{AGC} = \frac{R2}{R1} V_c + 2 \frac{R6}{R5 + R6} V_{ref} \quad (5)$$

Amplifier A2 inverts V1 producing a positive AGC signal voltage. Therefore, the input voltage to the TL026C AGC pin consists of an AGC signal equal to:

$$\frac{R2}{R1} V_c \quad (6)$$

and a dc voltage derived from V_{ref} , defined as the quiescent value of V_{AGC} .

$$V_{AGC}(q) = 2 \frac{R6}{R5 + R6} V_{ref} \quad (7)$$

For the initial resistor calculations, V_{ref} is assumed to be typically 1.4 V making quiescent V_{AGC} approximately 1.22 V ($V_{AGC}(q) = V_{ref} - 180$ mV). This voltage allows the TLO26C to operate at maximum gain under no-signal and low-signal conditions. In addition, with V_{ref} used as both internal and external reference, its variation from device to device automatically adjusts the overall bias and makes AGC operation essentially independent of the absolute value of V_{ref} . The resistor divider needs to be calculated only once and is valid for the full tolerance of V_{ref} .

output voltage limits (see Figures 6 and 7)

The output voltage level desired must fall within the following limits.

1. Because the data sheet minimum output swing is 3 V peak-to-peak using a 2-k Ω load resistor, the user-selected design limit for the peak output swing should not exceed 1.5 V.
2. The voltage drop of the rectifying diode determines the lower voltage limit. When a silicon diode is used, this voltage is approximately 0.7 V. The output voltage V_O must have sufficient amplitude to exceed the rectifying diode drop. A schottky diode can be used to reduce the V_O level required.

gain calculations for a peak output voltage of 1 V

A peak output voltage of 1 V was chosen for gain calculations because it is approximately midway between the limits of conditions 1 and 2 in the preceding paragraph.

Using equation 3 ($V_c = V_{OP} - V_d$), V_c is calculated as follows:

$$V_c = 1 \text{ V} - 0.7 \text{ V}$$

$$V_c = 0.3 \text{ V}$$

Therefore, the gain of A1 must produce a voltage V1 that is equal to or greater than the total change in V_{AGC} for maximum TL026C gain change.

With a total change in V_{AGC} of 360 mV and using equation 4, the calculation is as follows:

$$-\frac{V1}{V_c} = \frac{\Delta V_{AGC}}{V_c} = \frac{R2}{R1} = \frac{0.36}{0.3} = 1.2$$

If R1 is 10 k Ω , R2 is 1.2 times R1 or 12 k Ω .

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Special Functions

TYPICAL APPLICATION INFORMATION

Since the output voltage for this circuit must be between 0.85 V and 1.3 V, the component values in Figures 6 and 7 provide a nominal 1-V peak output limit. This limit is the best choice to allow for temperature variations of the diode and minimum output voltage specification.

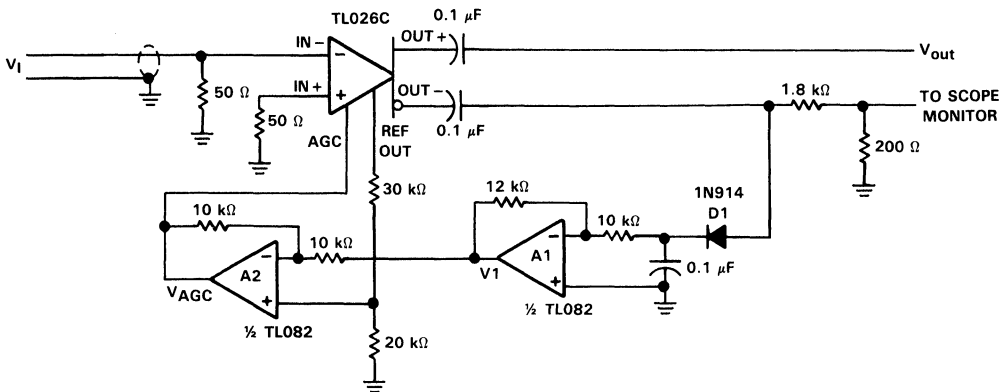
The circuit values in Figures 6 and 7 will produce the best results in this general application. Because of rectification and device input constraints, the circuit in Figure 6 will not provide attenuation and has about 32 dB of control range. The circuit shown in Figure 7 will have approximately 25% variation in the peak output voltage limit due to the variation in gain of the TL592 device to device. In addition, if a lower output voltage is desired, the output of the TL026C can be used for approximately 40 mV of controlled signal.

considerations for the use of the TL026C

To obtain the most reliable results, RF breadboarding techniques must be used. A groundplane board should be used and power supplies should be bypassed with 0.1- μ F capacitors. Input leads and output leads should be as short as possible and separated from each other.

A peak input voltage greater than 200 mV will begin to saturate the input stages of the TL026C and, while the circuit is in the AGC mode, the output signal may become distorted.

To observe the output signal of TL026C or TL592, low-capacitance FET probes or the output voltage divider technique shown in Figure 6 should be used.

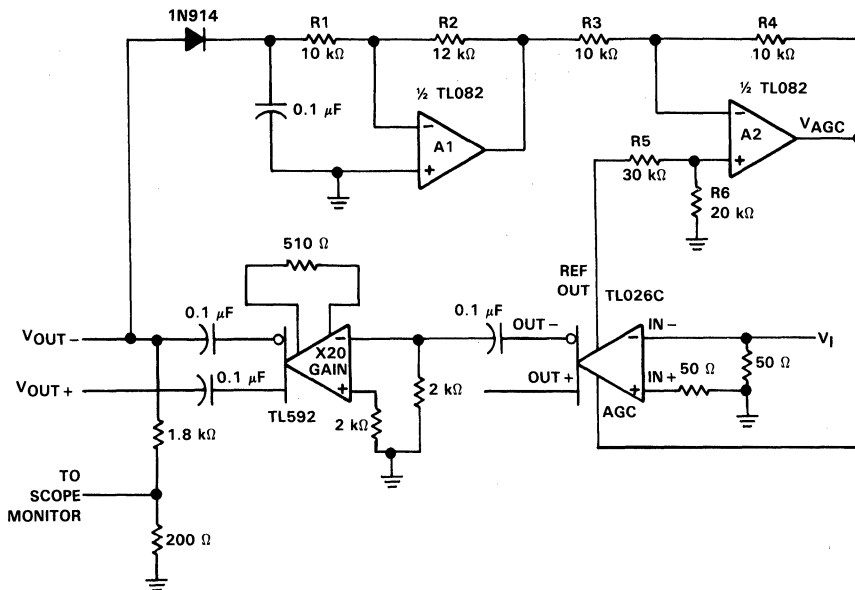


NOTE: $V_{CC+} = 6\text{ V}$ and $V_{CC-} = -6\text{ V}$ for TL026C and amplifiers A1 and A2.

FIGURE 6. TYPICAL APPLICATION CIRCUIT WITH NO ATTENUATION

TL026C DIFFERENTIAL VIDEO AMPLIFIER WITH AGC

TYPICAL APPLICATION INFORMATION



NOTE: $V_{CC+} = 6\ \text{V}$ and $V_{CC-} = -6\ \text{V}$ for TL026C and amplifiers A1 and A2.

FIGURE 7. TYPICAL APPLICATION CIRCUIT WITH ATTENUATION

4

Special Functions

TL027M, TL027C DIFFERENTIAL VIDEO AMPLIFIER WITH AGC

D2888, JUNE 1985—REVISED JANUARY 1989

- Low Output Common-Mode Sensitivity to AGC Voltages
- Input and Output Impedances Independent of AGC Voltage
- Wide AGC Range . . . 50 dB Typ
- 3-dB Bandwidth . . . 50 MHz
- Other Characteristics Similar to NE592 and μ A733

DEVICE FEATURES

	GAIN	AGC
Gain Option 1	50 dB	50 dB
Gain Option 2	38 dB	50 dB

description

This device is a monolithic two-stage video amplifier with differential inputs and outputs.

Internal feedback provides wide bandwidth, low phase distortion, and excellent gain stability. Variable gain based on signal summation provides large AGC control over a wide bandwidth with low harmonic distortion. Emitter-follower outputs enable the device to drive capacitive loads. All stages are current-source biased to obtain high common-mode and supply-voltage rejection ratios. The gain may be electronically attenuated by applying a control voltage to the AGC pins. No external frequency compensation components are required.

This device is particularly useful in TV and radio IF and RF AGC circuits, as well as magnetic-tape and disk-file systems where AGC is needed. Other applications include video and pulse amplifiers for which a large AGC range, wide bandwidth, low phase shift, and excellent gain stability are required.

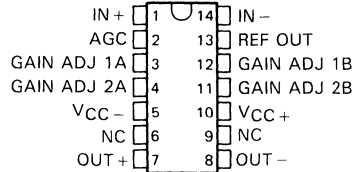
The TL027M is characterized for operation over the full military temperature range of -55°C to 125°C . The TL027C is characterized for operation from 0°C to 70°C .

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	8 V
Supply voltage, V_{CC-} (see Note 1)	-8 V
Differential input voltage	± 5 V
Common-mode input voltage	± 6 V
Output current	10 mA
Continuous total power dissipation	500 mW
Operating free-air temperature: TL027M	-55°C to 125°C
TL027C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

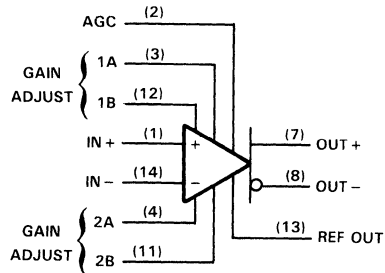
NOTE 1: All voltage values are with respect to the midpoint between V_{CC+} and V_{CC-} except differential input and output voltages.

D, J, OR N PACKAGE (TOP VIEW)



NC — No internal connection

symbol



4

Special Functions

TL027M, TL027C

DIFFERENTIAL VIDEO AMPLIFIER WITH AGC

recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, V_{CC+}		3	6	8	V
Supply voltage, V_{CC-}		-3	-6	-8	V
Operating free-air temperature, T_A	TL027M	-55		125	°C
	TL027C	0		70	

electrical characteristics at 25 °C operating free-air temperature, $V_{CC\pm} = \pm 6$ V, $V_{AGC} = 0$, REF OUT pin open (unless otherwise noted)

PARAMETER		TEST FIGURE	TEST CONDITIONS	GAIN OPTION†	MIN	TYP	MAX	UNIT
A_{VD}	Large-signal differential voltage amplification	1	$V_{O(PP)} = 3$ V, $R_L = 2$ k Ω	1	200	300	400	V/V
				2	65	85	105	
ΔA_{VD1}	Change in voltage amplification	1	$V_{I(PP)} = 7.5$ mV, $R_L = 2$ k Ω , $V_{AGC} - V_{ref} = \pm 180$ mV	1	-50			dB
ΔA_{VD2}	Change in voltage amplification	1	$V_{I(PP)} = 28.5$ mV, $R_L = 2$ k Ω , $V_{AGC} - V_{ref} = \pm 180$ mV	2	-50			dB
V_{ref}	Voltage at REF OUT		$I_{ref} = 1$ mA to 100 μ A	1	1.3	1.5		V
BW	Bandwidth (-3 dB)	2	$V_{O(PP)} = 1$ V, $V_{AGC} - V_{ref} = \pm 180$ mV	1	20			MHz
				2	50			
I_{IO}	Input offset current			1 or 2	0.4		5	μ A
I_{IB}	Input bias current			1 or 2	10		30	μ A
V_{ICR}	Common-mode input voltage range	3		2	± 1			V
V_{OC}	Common-mode output voltage	1	$R_L = \infty$	1 or 2	3.25	3.75	4.25	V
ΔV_{OC}	Change in common-mode output voltage	1	$V_{AGC} = 0$ to 2 V, $R_L = \infty$	1 or 2			300	mV
V_{OO}	Output offset voltage	1	$V_{ID} = 0$, $R_L = \infty$	1 or 2			0.75	V
$V_{O(PP)}$	Maximum peak-to-peak output voltage swing	1	$R_L = 2$ k Ω	1 or 2	3	4		V
r_i	Input resistance			1	4			k Ω
				2	10	30		
r_o	Output resistance				20			Ω
C_i	Input capacitance				2			pF
CMRR	Common-mode rejection ratio	3	$V_{IC} = \pm 1$ V, $f < 100$ kHz	2	60	86		dB
		3	$V_{IC} = \pm 1$ V, $f = 5$ MHz	2	60			
kSVR	Supply-voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IO}$)	4	$\Delta V_{CC+} = \pm 0.5$ V, $\Delta V_{CC-} = \pm 0.5$ V	2	50	70		dB
V_n	Broadband equivalent input noise voltage	4	BW = 1 kHz to 10 MHz	1 or 2	12			μ V
t_{pd}	Propagation delay time	2	$\Delta V_O = \pm 1$ V	1	7.5			ns
				2	6	10		
t_r	Rise time	2	$\Delta V_O = \pm 1$ V	1	10.5			ns
				2	4.5	12		
$I_{sink(max)}$	Maximum output sink current		$V_{ID} = 1$ V, $V_O = 3$ V	1 or 2	3	4		mA
I_{CC}	Supply current		No load, No signal	1 or 2	22		27	mA

† The gain option is selected as follows:

Gain Option 1. . . Gain Adjust pin 1A is connected to pin 1B, pins 2A and 2B are open.

Gain Option 2. . . Gain Adjust pin 2A is connected to pin 2B, pins 1A and 1B are open.

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Special Functions

TLO27M, TLO27C DIFFERENTIAL VIDEO AMPLIFIER WITH AGC

electrical characteristics over recommended operating free-air temperature range, $V_{CC\pm} = \pm 6\text{ V}$, $V_{AGC} = 0\text{ V}$, REF OUT pin open (unless otherwise noted)

PARAMETER	TEST FIGURE	TEST CONDITIONS	GAIN OPTION [†]	MIN	TYP	MAX	UNIT
A _{VD}	1	$V_{O(PP)} = 3\text{ V}$, $R_L = 2\text{ k}\Omega$	1	150	450		V/V
			2	55	115		
I _O			1 or 2			6	μA
I _B			1 or 2			40	μA
V _{ICR}	3		2	± 1			V
V _{OO}	1	$V_{ID} = 0$, $R_L = \infty$	1 or 2			1.5	V
V _{O(PP)}	1	$R_L = 2\text{ k}\Omega$	1 or 2	2.8			V
r _i			2	8			k Ω
CMRR	3	$V_{IC} = \pm 1\text{ V}$, $f < 100\text{ kHz}$	2	50			dB
k _{SVR}	4	$\Delta V_{CC+} = \pm 0.5\text{ V}$, $\Delta V_{CC-} = \pm 0.5\text{ V}$	2	50			dB
I _{sink(max)}		$V_{ID} = 1\text{ V}$, $V_O = 3\text{ V}$	1 or 2	2.8	4		mA
I _{CC}	1	No load, No signal	1 or 2			30	mA

[†] The gain option is selected as follows:

- Gain Option 1. . . Gain Adjust pin 1A is connected to pin 1B, pins 2A and 2B are open.
- Gain Option 2. . . Gain Adjust pin 2A is connected to pin 2B, pins 1A and 1B are open.

PARAMETER MEASUREMENT INFORMATION

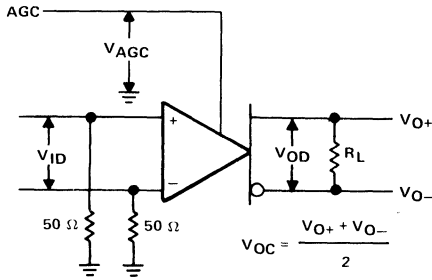


FIGURE 1

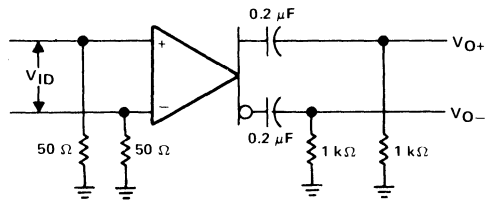


FIGURE 2

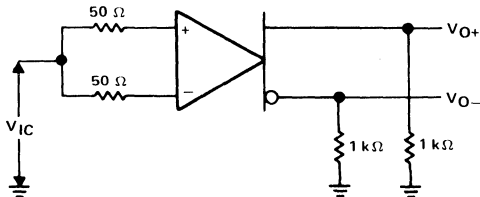


FIGURE 3

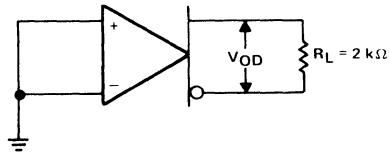


FIGURE 4

TL027M, TL027C
DIFFERENTIAL VIDEO AMPLIFIER WITH AGC

TYPICAL CHARACTERISTICS

DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 DIFFERENTIAL GAIN-CONTROL VOLTAGE

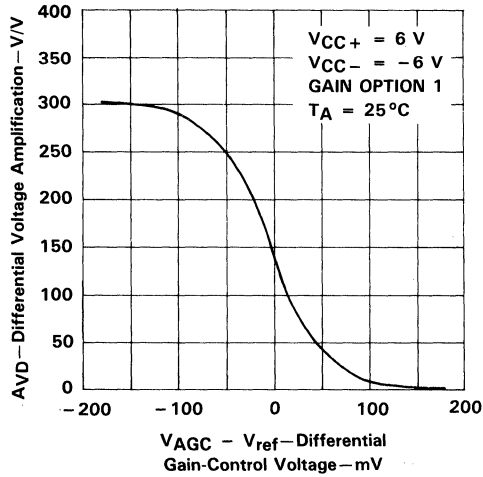


FIGURE 5

DIFFERENTIAL VOLTAGE AMPLIFICATION
 vs
 DIFFERENTIAL GAIN-CONTROL VOLTAGE

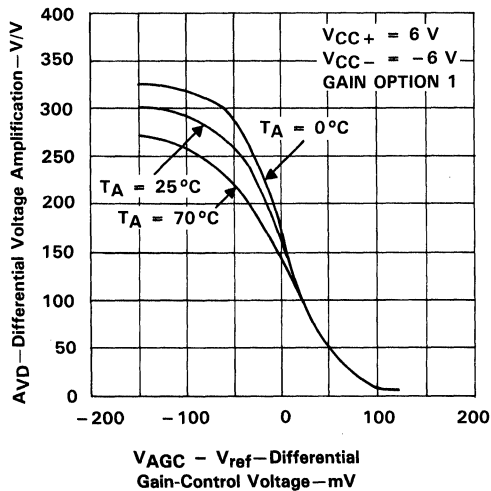


FIGURE 6

4

Special Functions

TYPICAL APPLICATION INFORMATION

gain characteristics

Figure 5 and 6 show the differential voltage amplification versus the differential gain-control voltage ($V_{AGC} - V_{ref}$). V_{AGC} is the absolute voltage applied to the AGC input and V_{ref} is the dc voltage at the REF OUT output. As V_{AGC} increases with respect to V_{ref} , the TL027C gain changes from maximum to minimum. As shown in Figure 5 for example, V_{AGC} would have to vary from approximately 180 mV less than V_{ref} to approximately 180 mV greater than V_{ref} to change the gain from maximum to minimum. The total signal change in V_{AGC} is defined by the following equation.

$$\Delta V_{AGC} = V_{ref} + 180 \text{ mV} - (V_{ref} - 180 \text{ mV}) \quad (1)$$

$$\Delta V_{AGC} = 360 \text{ mV}$$

However, because V_{AGC} varies as the ac AGC signal varies and also differentially around V_{ref} , then V_{AGC} should have an ac signal component and a dc component. To preserve the dc and thermal tracking of the device, this dc voltage must be generated from V_{ref} . To apply proper bias to the AGC input, the external circuit used to generate V_{AGC} must combine these two voltages. Figures 7 and 8 show two circuits that will perform this operation and are easy to implement. The circuits use a standard dual operational amplifier for AGC feedback. By providing rectification and the required feedback gain, these circuits are also complete AGC systems.

circuit operation

Amplifier A1 amplifies and inverts the rectified and filtered AGC signal voltage V_C producing output voltage V1. Amplifier A2 is a differential amplifier that inverts V1 again and adds the scaled V_{ref} voltage. This conditioning makes V_{AGC} the sum of the signal plus the scaled V_{ref} . As the signal voltage increases, V_{AGC} increases and the gain of the TLC027C is reduced. This maintains a constant output level.

feedback circuit equations

Following the AGC input signal (Figures 7 and 8) from the OUT – output through the feedback amplifiers to the AGC input produces the following equations.

1. AC output to diode D1, assuming sinusoidal signals

$$V_O = V_{OP} (\sin (wt)) \quad (2)$$

where:

$$V_{OP} = \text{peak voltage of } V_O$$

2. Diode D1 and capacitor C1 output

$$V_C = V_{OP} - V_F \quad (3)$$

where:

$$V_F = \text{forward voltage drop of D1}$$

$$V_C = \text{voltage across capacitor C1}$$

3. A1 output

$$V1 = - \frac{R2}{R1} V_C \quad (4)$$

TL027M, TL027C

DIFFERENTIAL VIDEO AMPLIFIER WITH AGC

TYPICAL APPLICATION INFORMATION

4. A2 output ($R3 = R4$)

$$V_{AGC} = \frac{R2}{R1} V_c + 2 \frac{R1}{R5 + R6} V_{ref} \quad (5)$$

Amplifier A2 inverts V1 producing a positive AGC signal voltage. Therefore, the input voltage to the TL027C AGC pin consists of an AGC signal equal to:

$$\frac{R2}{R1} V_c \quad (6)$$

and a dc voltage derived from V_{ref} , defined as the quiescent value of V_{AGC} .

$$V_{AGC}(q) = 2 \frac{R6}{R5 + R6} V_{ref} \quad (7)$$

For the initial resistor calculations, V_{ref} is assumed to be typically 1.4 V making quiescent V_{AGC} approximately 1.22 V ($V_{AGC}(q) = V_{ref} - 180$ mV). This voltage allows the TL027C to operate at maximum gain under no-signal and low-signal conditions. In addition, with V_{ref} used as both internal and external reference, its variation from device to device automatically adjusts the overall bias and makes AGC operation essentially independent of the absolute value of V_{ref} . The resistor divider needs to be calculated only once and is valid for the full tolerance of V_{ref} .

output voltage limits (see Figures 7 and 8)

The output voltage level desired must fall within the following limits:

1. Because the data sheet minimum output swing is 3 V peak-to-peak using a 2-k Ω load resistor, the user-selected design limit for the peak output swing should not exceed 1.5 V.
2. The voltage drop of the rectifying diode determines the lower voltage limit. When a silicon diode is used, this voltage is approximately 0.7 V. The output voltage V_O must have sufficient amplitude to exceed the rectifying diode drop. A schottky diode can be used to reduce the V_O level required.

gain calculations for a peak output voltage of 1 V

A peak output voltage of 1 V was chosen for gain calculations because it is approximately midway between the limits of conditions 1 and 2 in the preceding paragraph.

Using equation 3 ($V_c = V_{OP} - V_d$), V_c is calculated as follows:

$$V_c = 1 \text{ V} - 0.7 \text{ V}$$

$$V_c = 0.3 \text{ V}$$

Therefore, the gain of A1 must produce a voltage V1 that is equal to or greater than the total change in V_{AGC} for maximum TL027C gain change.

With a total change in V_{AGC} of 360 mV and using equation 4, the calculation is as follows:

$$-\frac{V1}{V_c} = \frac{\Delta V_{AGC}}{V_c} = \frac{R2}{R1} = \frac{0.36}{0.3} = 1.2$$

If R1 is 10 k Ω , R2 is 1.2 times R1 or 12 k Ω .

TL027M, TL027C DIFFERENTIAL VIDEO AMPLIFIER WITH AGC

TYPICAL APPLICATION INFORMATION

Since the output voltage for this circuit must be between 0.85 V and 1.3 V, the component values in Figures 7 and 8 provide a nominal 1-V peak output limit. This limit is the best choice to allow for temperature variations of the diode and minimum output voltage specification.

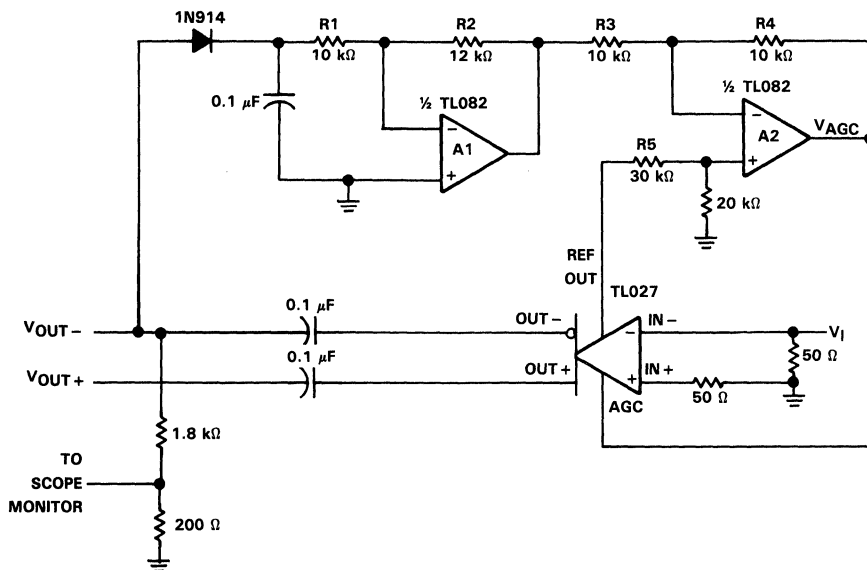
The circuit values in Figures 7 and 8 will produce the best results in this general application. Because of rectification and device input constraints, the circuit in Figure 7 will not provide attenuation and has about 32 dB of control range. The circuit shown in Figure 8 will have approximately 25% variation in the peak output voltage limit due to the variation in gain of the TL592 device to device. In addition, if a lower output voltage is desired, the output of the TL027C can be used for approximately 40 mV of controlled signal.

considerations for the use of the TL027C

To obtain the most reliable results, RF breadboarding techniques must be used. A groundplane board should be used and power supplies should be bypassed with 0.1- μ F capacitors. Input leads and output leads should be as short as possible and separated from each other.

A peak input voltage greater than 200 mV will begin to saturate the input stages of the TL027C and, while the circuit is in the AGC mode, the output signal may become distorted.

To observe the output signal of TL027C or TL592, low-capacitance FET probes or the output voltage divider technique shown in Figure 7 should be used.



NOTES: A. $V_{CC+} = 6$ V and $V_{CC-} = -6$ V for TL027C and amplifiers A1 and A2.

B. On the TL027, short pin 3 to pin 12 and pin 4 to pin 11. Connect pins 6 and 9 to ground.

FIGURE 7. TYPICAL APPLICATION CIRCUIT WITH NO ATTENUATION

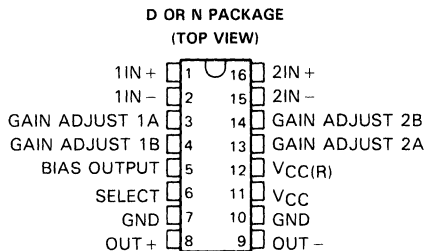
4

Special Functions

TL040C 2-CHANNEL MULTIPLEXED VIDEO AMPLIFIER

D3002, MARCH 1986—REVISED DECEMBER 1988

- **Designed for Use with the TL041 Magnetic Field Pulse Detector**
- **Wide Bandwidth . . . 20 MHz Typ**
- **Low Noise . . . Less than 8 μ V Typ**
- **Independently Adjustable Channel Gains . . . Up to 450 Typ**
- **No Frequency Compensation Required**
- **Internal Voltage Source Eliminates External Components**
- **Input Channel Select Pin is Compatible with TTL and CMOS**
- **Low Power Dissipation . . . 150 mW Typ**



CHANNEL SELECT TABLE

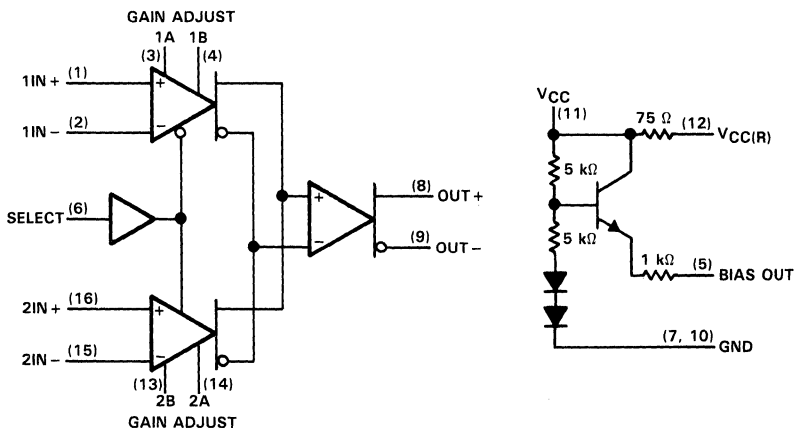
SELECT	CHANNEL
L	1
H	2

description

The TL040 is a two-channel multiplexed video amplifier designed for use with magnetic pulse detectors in streaming tape drives. The circuit design eliminates many external components, and the D package allows substantial reduction in circuit board area. The gain of each channel is a function of the resistance across its gain-adjust pins (A-B) with maximum gain occurring when the terminals are shorted.

The $V_{CC(R)}$ pin provides supply voltage decoupling required by some designs. The BIAS OUT pin provides a voltage source for other circuits that is approximately equal to $1/2 V_{CC}$.

functional block diagram



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Special Functions

TLO40C

2-CHANNEL MULTIPLEXED VIDEO AMPLIFIER

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V _{CC} (see Note 1)	14 V
Input voltage range	-0.2 V to V _{CC} + 0.2 V
Continuous total power dissipation	600 mW
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltages except differential voltages are with respect to the ground terminals.

recommended operating conditions

	MIN	TYP	MAX	UNIT
Supply voltage, V _{CC}	10.8	12	13.2	V
Common-mode input voltage (diff inputs), V _{IC}	5	6	7	V
High-level input voltage, SELECT input, V _{IH}	2			V
Low-level input voltage, SELECT input, V _{IL}			0.8	V
Output sink current (diff outputs), I _{sink}			1.5	mA
Operating free-air temperature, T _A	0		70	°C

electrical characteristics of selected channel at T_A = 25°C, V_{CC} = 12 V, R_{AB} = 0, R_L = 2 kΩ (unless otherwise noted)

PARAMETER	TEST FIGURE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
A _{VD} Large-signal differential voltage amplification	1		300	530	600	V/V
Channel amplification mismatch	1			1%		
Large-signal differential voltage attenuation	1	ΔV _I = 50 mV on unselected input		60		dB
V _{OC} Common-mode output voltage	1	R _L = ∞		8.5		V
V _{OPP} Maximum peak-to-peak output voltage swing	1			4		V
BW Bandwidth (-3 dB)	2			20		MHz
I _{IQ} Input offset current	1			0.1	3	μA
I _{IB} Input bias current	1			6	17	μA
V _{OD} Differential output voltage	1	R _L = ∞, V _{ID} = 0		0.2		V
r _i Input resistance (differential inputs)				4		kΩ
CMRR Common-mode rejection ratio	3	V _{IC} = 5 V to 7 V	60	100		dB
k _{SVR} Supply-voltage rejection ratio (ΔV _{CC} /ΔV _{IQ})	4	V _{CC} = 10.8 V to 13.2 V	50	70		dB
V _n Broadband equivalent input noise voltage	4			<5		μV
I _{IH} High-level input current, Select input		V _{IH} = 2.7 V			-0.4	mA
I _{IL} Low-level input current, Select input		V _{IL} = 0.4 V			20	μA
t _{pd} Propagation delay time (differential inputs)	2	ΔV _O = 1 V		15		ns
t _r Output rise time	2	ΔV _O = 1 V		20		ns
I _{CC} Supply current	1			12	15	mA
Bias output voltage	1		5	6	7	V

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Special Functions

PARAMETER MEASUREMENT INFORMATION

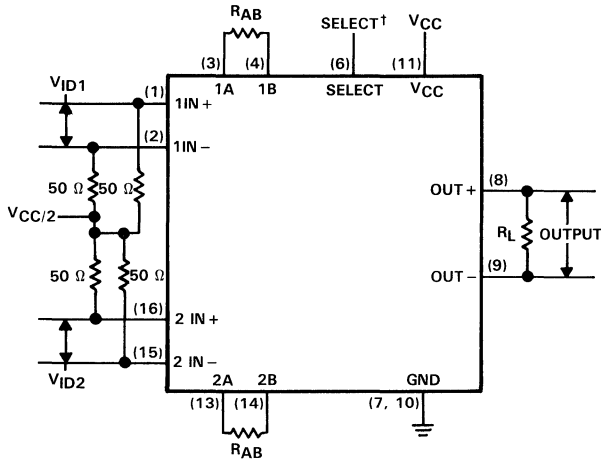


FIGURE 1

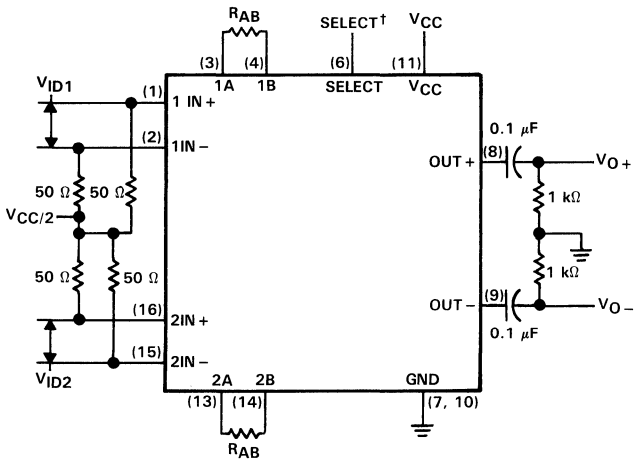


FIGURE 2

†Select input must be at proper logic level to select desired input channel.

TLO40C
2-CHANNEL MULTIPLEXED VIDEO AMPLIFIER

PARAMETER MEASUREMENT INFORMATION (continued)

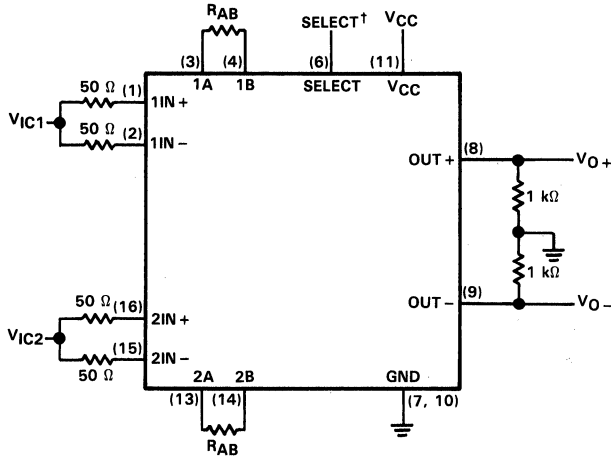


FIGURE 3

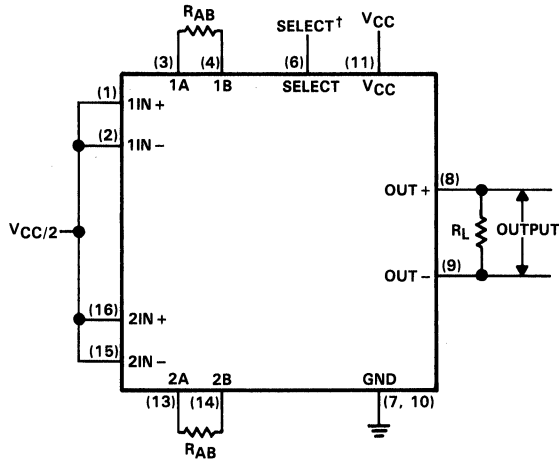


FIGURE 4

†Select input must be at proper logic level to select desired input channel.

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Special Functions

TYPICAL CHARACTERISTICS

LARGE-SIGNAL DIFFERENTIAL
VOLTAGE AMPLIFICATION
vs
GAIN-ADJUST RESISTANCE

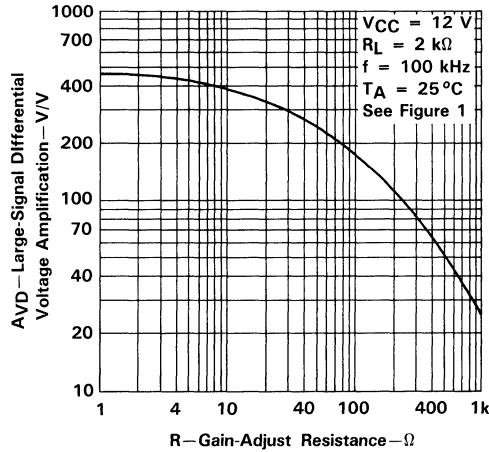


FIGURE 5

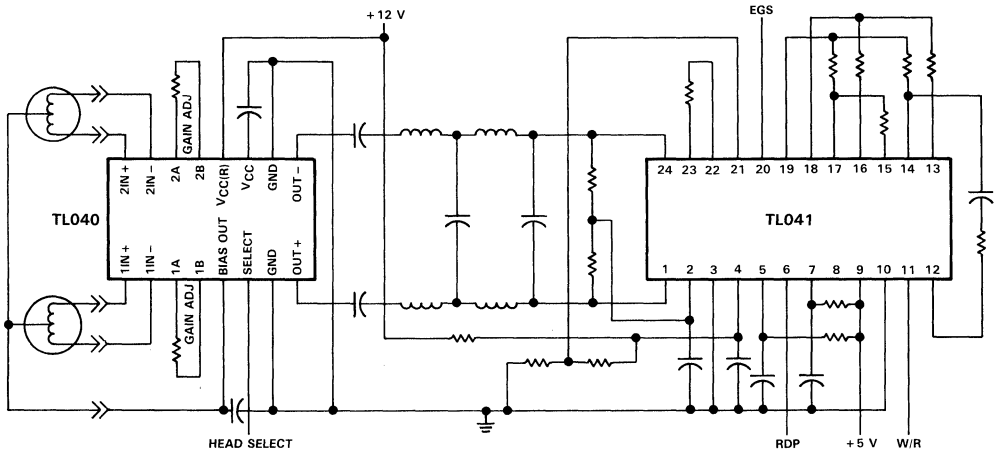


FIGURE 6. READ SIGNAL CIRCUIT FOR A STREAMING TAPE DRIVE

4
Special Functions

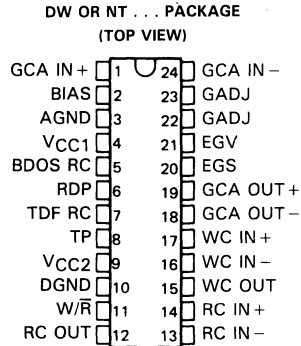
4

Special Functions

TL041C TAPE READ SIGNAL CONDITIONER

D3024, AUGUST 1987

- Designed for Signal Processing in Streaming-Tape Memory Units in Combination with TL040 Two-Channel Video Amplifier
- Space-Saving LSI Circuits Include:
 - Two High-Speed Differential Comparators
 - Time-Domain Filter
 - Bidirectional One-Shot Multivibrator
 - Gain-Controlled Video Amplifier with Differential Inputs and Outputs
- Amplifier and Comparator
 - Bandwidth . . . 20 MHz Typical
- Maximum Data Rate at Read Data Pulse (RDP) . . . 1.4 Mb/s Typical
- Available in 300-mil Dual-In-Line and "Small Outline" Plastic Packages



description

The TL041 is a magnetic tape read signal conditioner designed for use with the TL040 video amplifier. When combined, these devices amplify the low-signal output from a streaming-tape playback head and reconstruct the data as originally written on the tape. The TL041C includes a gain-controlled amplifier, two comparators, read/write select logic, a time-domain filter, and a bidirectional one-shot multivibrator.

The amplifier has differential inputs, differential outputs, and electronic gain control. A special feature of the electronic gain control is the Electronic Gain Select (EGS). When the EGS input is high, the Electronic Gain Voltage (EGV) input is driven low and amplifier gain is determined by the value of the resistor connected between the Gain Adjust (GADJ) pins. When the EGS input is low, the gain set by the resistor is increased by an amount determined by the voltage applied to the EGV pin.

To accommodate different magnetic tape output signal levels, the amplifier gain may be switched by logic at the EGS input, controlled manually with an adjustable voltage at the EGV input, or automatically adjusted with an automatic gain control (AGC) circuit applying a control voltage to the EGV input.

The comparator functions are controlled by a logic input to the Write/Read (W/R) select input. With the W/R input low, the read comparator output (usually connected as a zero-crossing detector) is sent to the time-domain filter. When W/R is high, the write comparator output is used to provide write amplitude verification in a typical read-after-write function.

The time-domain filter helps to ensure the input data is valid. A capacitor in series with a resistor, connected to the time-domain filter pin (TDF RC), begins charging at the leading edge of an input pulse from the read comparator. If the input pulse does not remain high for one RC time constant, the pulse is considered invalid and no signal is passed to the bidirectional one-shot multivibrator (BDOS). However, if the input pulse remains high for longer than one RC time constant, the pulse is considered valid and the signal is passed through the time-domain filter to trigger the BDOS. When triggered, the BDOS provides a pulse to the Read Data Pulse (RDP) output. The RDP output pulse duration is determined by a resistor-capacitor network connected to the BDOS RC pin.

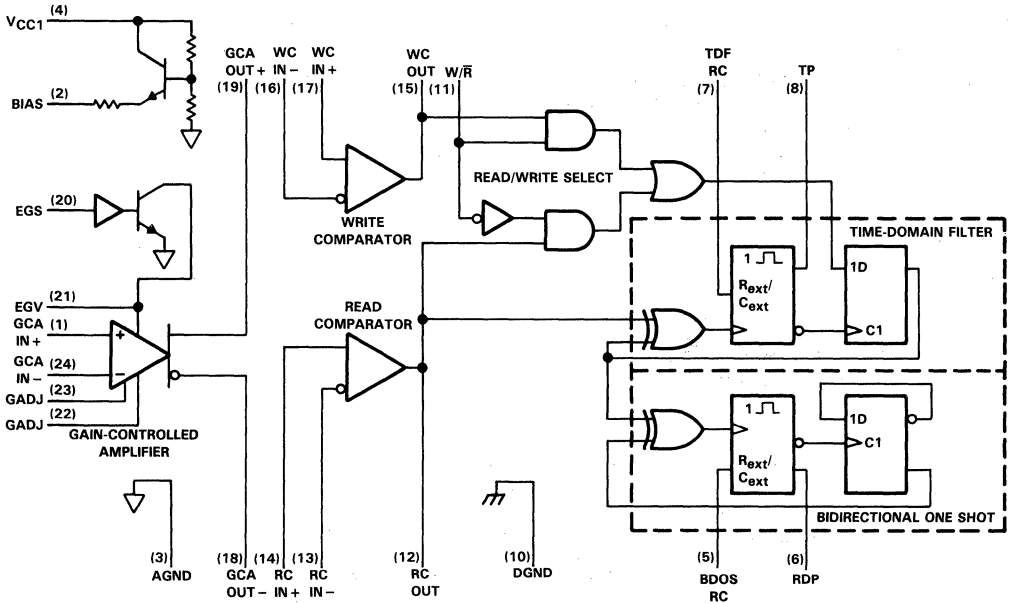
The TL041C is characterized for operation from 0°C to 70°C.

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Special Functions

TL041C TAPE READ SIGNAL CONDITIONER

functional block diagram



FUNCTION TABLE

INPUT CONDITIONS			I/O NAME	I/O CONDITION
EGS	W/R	DIFFERENTIAL INPUTS WRITE OR READ COMPARATOR		
	X	RC IN+ > RC IN-	RC OUT	H
	X	RC IN- > RC IN+	RC OUT	L
	L	X	RC OUT	Input to time-domain filter
	X	WC IN+ > WC IN-	WC OUT	H
	X	WC IN- > WC IN+	WC OUT	L
	H	X	WC OUT	Input to time-domain filter
H		X	EGV	L
L		X	EGV	Input

4

Special Functions

PIN		DESCRIPTION
NAME	NO.	
AGND	3	Analog ground
BDOS RC	5	Bidirectional one-shot resistor and capacitor
BIAS	2	Output bias voltage
DGND	10	Digital ground
EGS	20	Electronic gain select
EGV	21	Electronic gain voltage
GCA IN -	24	Gain-controlled amplifier, inverting input
GCA IN +	1	Gain-controlled amplifier, noninverting input
GADJ	22	Gain adjust
GADJ	23	Gain adjust
GCA OUT -	18	Gain-controlled amplifier, inverting output
GCA OUT +	19	Gain-controlled amplifier, noninverting output
RC IN -	13	Read comparator, inverting input
RC IN +	14	Read comparator, noninverting input
RC OUT	12	Read comparator out
RDP	6	Read data pulse
TDF RC	7	Time-domain filter resistor and capacitor
TP	8	Test point
VCC1	4	Analog collector supply voltage
VCC2	9	Digital collector supply voltage
WC IN -	16	Write comparator, inverting input
WC IN +	17	Write comparator, noninverting input
WC OUT	15	Write comparator out
W/R	11	Write/read

TL041C

TAPE READ SIGNAL CONDITIONER

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage: V_{CC1} (see Note 1)	14 V
V_{CC2}	7 V
Input voltage range: Amplifier and comparators	AGND - 0.2 V to $V_{CC1} + 0.2$ V
Multivibrators and logic	AGND - 0.2 V to $V_{CC2} + 0.2$ V
Input current: EGV (see Note 2)	± 2 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from the case for 10 seconds	260°C

- NOTES: 1. All voltages except differential voltages are with respect to network ground terminals (AGND and DGND tied together).
 2. Driving EGV high from a low-impedance source ($> \pm 2$ mA capability) with EGS high can result in damage to the device.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING
DW	1350 mW	10.8 mW/°C	864 mW/°C
NT	1700 mW	13.6 mW/°C	1088 mW/°C

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{CC1}	10.8	12	13.2	V
Supply voltage, V_{CC2}	4.5	5	5.5	V
High-level input voltage, V_{IH}	EGS or W/ \bar{R}		2	V
Low-level input voltage, V_{IL}	EGS or W/ \bar{R}		0.8	V
Input voltage, V_I	EGS		0	10
	EGV		0	$0.8V_{CC1}$
Common-mode input voltage to gain-control amplifier, V_{IC}	4			V
High-level output current, I_{OH}	WC OUT, RC OUT, TP, or RDP		-400	μA
Low-level output current, I_{OL}	WC OUT, RC OUT, TP, or RDP		8	mA
Pulse duration, t_w	40			ns
External timing resistance, (see Note 3)	TDF or BDOS RC		5	25
External timing capacitance	TDF or BDOS RC)		0.01	0.1
Operating free-air temperature, T_A	0			70
				°C

NOTE 3: Some high resistance and capacitance combinations may produce abnormal output waveforms.

4 Special Functions

electrical characteristics at $V_{CC1} = 12\text{ V}$, $V_{CC2} = 5\text{ V}$, $V_{IC}(GIC) = V_{bias}$, $R_{ADJ} = 5\text{ k}\Omega$, EGS at high level, EGV at 0 V, $r_i = 50\text{ }\Omega$, $R_L = 2\text{ k}\Omega$, $T_A = 25\text{ }^\circ\text{C}$ (unless otherwise noted)

gain-controlled amplifier

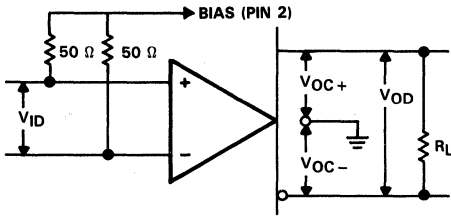
PARAMETER		TEST FIGURE	TEST CONDITIONS	MIN	TYP	MAX	UNIT		
V_{OO}	Output offset voltage	1	$V_{ID} = 0$, $V_{OD} = V_O$		0.35	0.75	V		
V_{OPP}	Maximum differential output voltage	1	$V_{ID} = 1\text{ V}$, $V_{OPP} = V_O$	3	4		V		
A_{VD}	Large-signal differential voltage amplification	1	$V_{ID} = 20\text{ mV}$, $V_{id} = 20\text{ mV}$, EGS low, $f = 455\text{ kHz}$	EGS high		8	14	20	V/V
			EGV at 4 V		19				V/V
			EGV at 9.6 V		90				
CMRR	Common-mode rejection ratio	2	$V_{IC} = 2\text{ V to }5\text{ V}$	60	80		dB		
V_{IC}	Common-mode input voltage	2		2		5	V		
V_{OC}	Common-mode output voltage	1	$V_{ID} = 0$	4	5	6	V		
I_{IO}	Input offset current	1	$I_{IB+} - I_{IB-}$		0.2	3	μA		
I_O	Output current, sink			1.5	2		mA		
I_{IB}	Input bias current	1	$(I_{IB+} + I_{IB-})/2$		5	17	μA		
$V_O(\text{BIAS})$	Bias output voltage	1		3	4	5	V		
$z_O(\text{BIAS})$	Bias output impedance				1		k Ω		
z_i	Input impedance				30		k Ω		
BW	Bandwidth (-3 dB)	3			20		MHz		
k $_{SVR}$	Supply voltage rejection ratio	4	$V_{CC1} = 10.8\text{ V to }13.2\text{ V}$	50	70		dB		
I_{CC1}	Supply current from V_{CC1}		$V_{CC1} = 13.2\text{ V}$, No signal		32	45	mA		

logic section

PARAMETER		TEST FIGURE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{OH}	High-level output voltage		$V_{CC2} = 4.5\text{ V}$, $I_{OH} = -400\text{ }\mu\text{A}$, $V_{ID} = 0.1\text{ V}$	2.7	3.5		V
V_{OL}	Low-level output voltage		$V_{CC2} = 4.5\text{ V}$, $I_{OL} = 8\text{ mA}$, $V_{ID} = 0.1\text{ V}$		260	500	mV
V_{ICR}	Common-mode input voltage, comparators			2		7	V
I_{IH}	High-level input current	EGS	$V_{I(EGS)} = 2.7\text{ V}$		120	200	μA
		W/R	$V_{I(W/R)} = 2.7\text{ V}$			20	
I_{IL}	Low-level input current	EGS	$V_{I(EGS)} = 0.4\text{ V}$			-20	μA
		W/R	$V_{I(W/R)} = 0.4\text{ V}$			-400	
I_{CC2}	Supply current from V_{CC2}		$V_{CC2} = 5.5\text{ V}$, No signal		22	31	mA
	Response time		100-mV step, 5-mV overdrive		50		ns
t_w	Pulse duration of one-shots (TP, RDP)		$R_{ext} = 5\text{ k}\Omega$, $C_{ext} = 100\text{ pF}$		360		ns
			$R_{ext} = 20\text{ k}\Omega$, $C_{ext} = 33\text{ pF}$		460		

TL041C
TAPE READ SIGNAL CONDITIONER

PARAMETER MEASUREMENT INFORMATION



$V_{OD} = V_{OD}$ with $V_{ID} = 0$
 $V_{OPP} = V_{OD}$ with $V_{ID} = 1\text{ V}$
 $A_{VD} = \frac{V_{OD}}{V_{ID}}$ with $V_{ID} = 20\text{ mV}$
 $V_{OC} = \frac{V_{OC+} + V_{OC-}}{2}$ with $V_{ID} = 0$

FIGURE 1

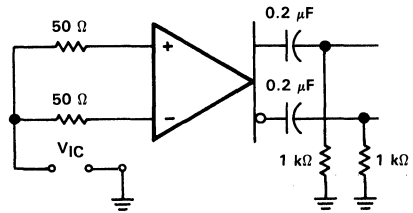


FIGURE 2

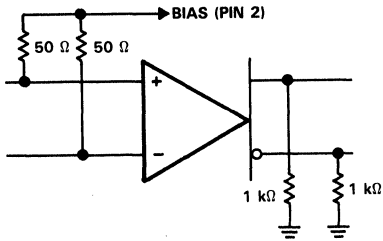


FIGURE 3

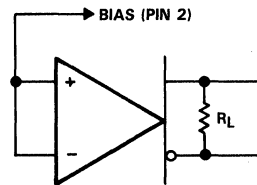


FIGURE 4

4

Special Functions

TYPICAL CHARACTERISTICS

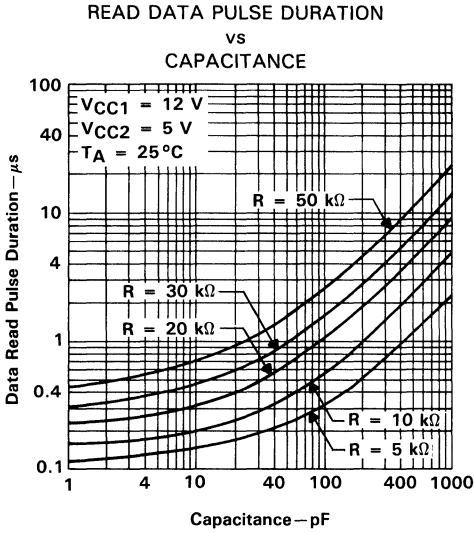


FIGURE 5

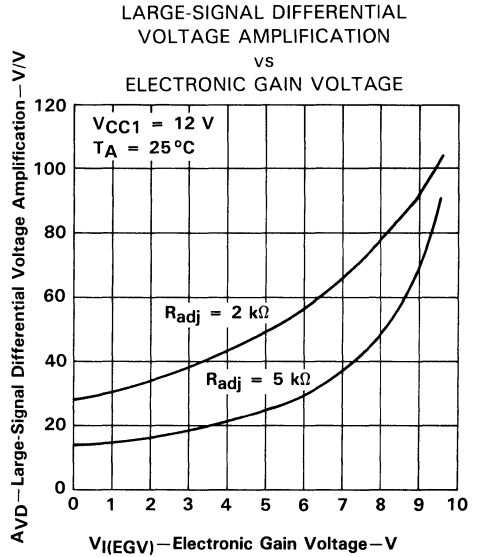


FIGURE 6

TYPICAL APPLICATION DATA

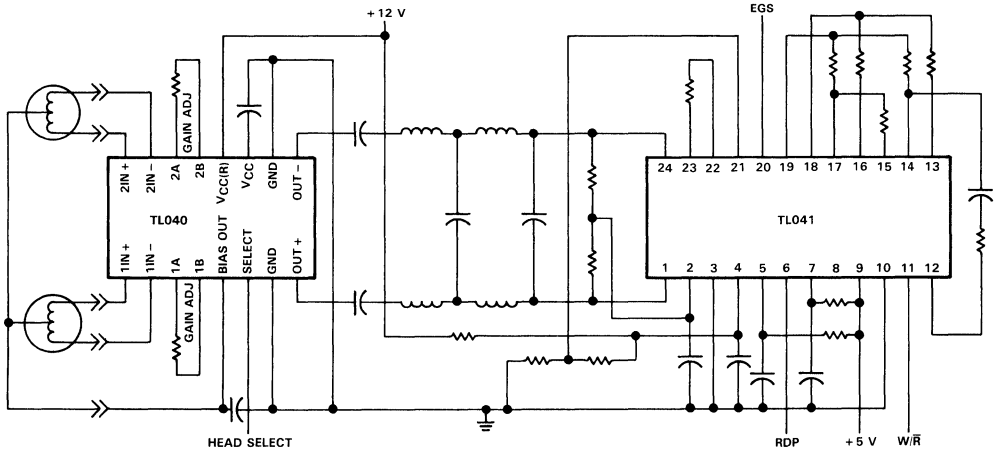


FIGURE 7. READ SIGNAL CIRCUIT FOR A STREAMING TAPE DRIVE

4

Special Functions

TL170C SILICON HALL-EFFECT SWITCH

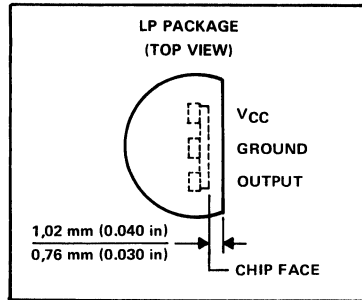
D2408, DECEMBER 1977, REVISED APRIL 1988

- Magnetic-Field Sensing Hall-Effect Input
- On-Off Hysteresis
- Small Size
- Solid-State Technology
- Open-Collector Output

description

The TL170C is a low-cost magnetically operated electronic switch that utilizes the Hall Effect to sense steady-state magnetic fields. Each circuit consists of a Hall-Effect sensor, signal conditioning and hysteresis functions, and an output transistor integrated into a monolithic chip. The outputs of these circuits can be directly connected to many different types of electronic components.

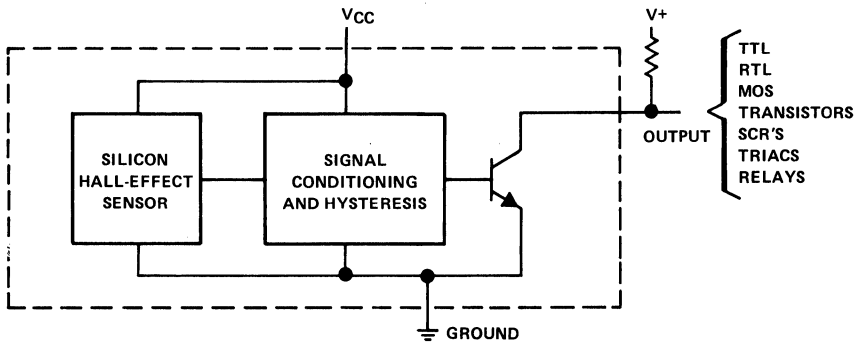
The TL170C is characterized for operation over the temperature range of 0°C to 70°C.



FUNCTION TABLE ($T_A = 25^\circ\text{C}$)

FLUX DENSITY	OUTPUT
$\leq -25 \text{ mT}$	Off
$-25 \text{ mT} < B < 25 \text{ mT}$	Undefined
$\geq 25 \text{ mT}$	On

functional block diagram



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	7 V
Output voltage	30 V
Output current	20 mA
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1.6 mm (1/16 inch) from case for 10 seconds	260°C
Magnetic flux density	unlimited

NOTE 1. Voltage values are with respect to network ground terminal.

4
Special Functions

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TL170C SILICON HALL-EFFECT SWITCH

electrical characteristics at specified free-air temperature, $V_{CC} = 5\text{ V} \pm 5\%$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
B_{T+}	Threshold of positive-going magnetic flux density †	25 °C		25	mT ‡	
		0 °C to 70 °C		35		
B_{T-}	Threshold of negative-going magnetic flux density †	25 °C		-25 §	mT ‡	
		0 °C to 70 °C		-35 §		
$B_{T+} - B_{T-}$	Hysteresis			20	mT ‡	
I_{OH}	High-level output current	$V_{OH} = 20\text{ V}$			100	μA
V_{OL}	Low-level output voltage	$V_{CC} = 4.75\text{ V}$, $I_{OL} = 16\text{ mA}$			0.4	V
I_{CC}	Supply current	$V_{CC} = 5.25\text{ V}$	Output low		6	mA
			Output high		4	

† Threshold values are those levels of magnetic flux density at which the output changes state. For the TL170C, a level more positive than B_{T+} causes the output to go to a low level and a level more negative than B_{T-} causes the output to go to a high level. See Figures 1 and 2.

‡ The unit of magnetic flux density in the International System of Units (SI) is the tesla (T). The tesla is equal to one weber per square meter. Values expressed in milliteslas may be converted to gauss by multiplying by ten.

§ The algebraic convention, where the most negative limit is designated as minimum, is used in this data sheet for flux-density threshold levels only.

The north pole of a magnet is the pole that is attracted by the geographical north pole. The north pole of a magnet repels the north-seeking pole of a compass. By accepted magnetic convention, lines of flux emanate from the north pole of a magnet and enter the south pole.

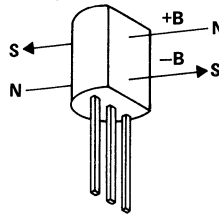
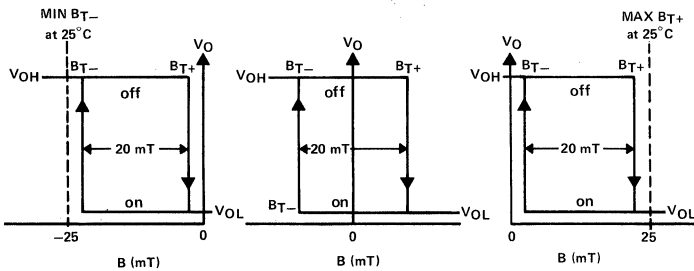


FIGURE 1. DEFINITION OF MAGNETIC FLUX POLARITY



The positive-going threshold (B_{T+}) may be a negative or positive B level at which a positive-going (decreasing negative or increasing positive) flux density results in the TL170 output turn-on. The negative-going threshold is a positive or negative B level at which a negative-going (decreasing positive or increasing negative) flux density results in the TL170 turning off.

FIGURE 2. REPRESENTATIVE CURVES OF V_O vs B

4

Special Functions

TL172C NORMALLY OFF SILICON HALL-EFFECT SWITCH

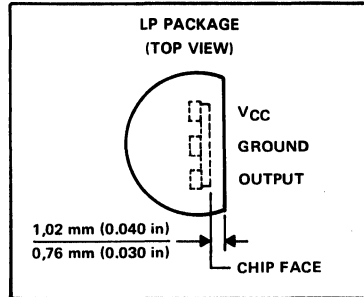
D2490, AUGUST 1977—REVISED APRIL 1988

- Magnetic-Field Sensing Hall-Effect Input
- On-Off Hysteresis
- Small Size
- Solid-State Technology
- Open-Collector Output
- Normally Off Switch

description

The TL172C is a low-cost magnetically operated normally off electronic switch that utilizes the Hall Effect to sense the presence of a magnetic field. Each circuit consists of a Hall-Effect sensor, signal conditioning and hysteresis functions, and an output transistor integrated into a monolithic chip. A magnetic field of sufficient strength in the positive direction will cause the TL172C output to be in a low-impedance state. Otherwise, the output will present a high impedance. The output of this circuitry connected to many different types of electronic components.

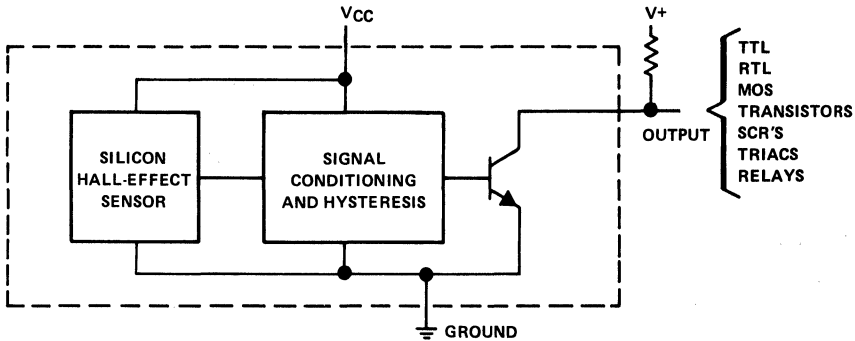
The TL172C is characterized for operation over the temperature range of 0°C to 70°C.



FUNCTION TABLE

FLUX DENSITY	OUTPUT
$\leq 10 \text{ mT}$	Off
$10 \text{ mT} < B < 60 \text{ mT}$	Undefined
$\geq 60 \text{ mT}$	On

functional block diagram



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Special Functions

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TL172C

NORMALLY OFF SILICON HALL-EFFECT SWITCH

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	7 V
Output voltage	30 V
Output current	20 mA
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C
Magnetic flux density	unlimited

NOTE 1: Voltage values are with respect to network ground terminal.

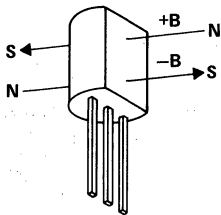
electrical characteristics over rated operating free-air temperature range, $V_{CC} = 5 V \pm 5\%$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
B_{T+}	Threshold of positive-going magnetic flux density†			60	mT‡
B_{T-}	Threshold of negative-going magnetic flux density†	10			mT‡
$B_{T+} - B_{T-}$	Hysteresis		23		mT‡
I_{OH}	High-level output current	$V_{OH} = 20 V$		100	μA
V_{OL}	Low-level output voltage	$V_{CC} = 4.75 V, I_{OL} = 16 mA$		0.4	V
I_{CC}	Supply current	$V_{CC} = 5.25 V$		6	mA

†Threshold values are those levels of magnetic flux density at which the output changes state. For the TL172C, a level more positive than B_{T+} causes the output to go to a low level, and a level more negative than B_{T-} causes the output to go to a high level. See Figures 1 and 2.

‡The unit of magnetic flux density in the International System of Units (SI) is the tesla (T). The tesla is equal to one weber per square meter. Values expressed in milliteslas may be converted to gauss by multiplying by ten.

4 Special Functions



The north pole of a magnet is the pole that is attracted by the geographical north pole. The north pole of a magnet repels the north-seeking pole of a compass. By accepted magnetic convention, lines of flux emanate from the north pole of a magnet and enter the south pole.

FIGURE 1. DEFINITION OF MAGNETIC FLUX POLARITY

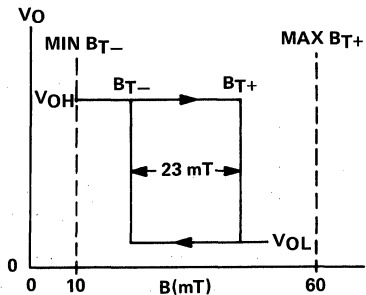
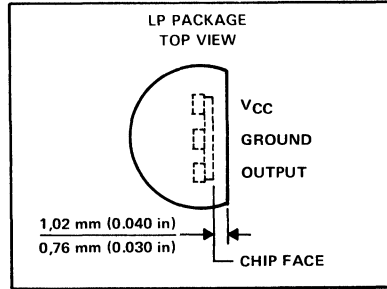


FIGURE 2. REPRESENTATIVE CURVE OF V_O vs B

TL173I, TL173C LINEAR HALL-EFFECT SENSORS

D2526, MARCH 1979—REVISED APRIL 1988

- Output Voltage Linear with Applied Magnetic Field
- Sensitivity Constant Over Wide Operating Temperature Range
- Solid-State Technology
- Three-Terminal Device
- Senses Static or Dynamic Magnetic Fields

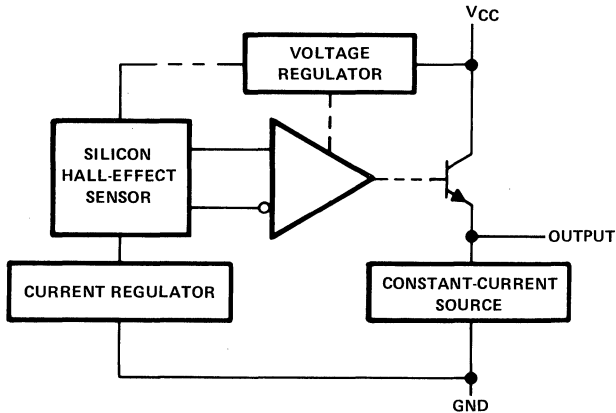


description

The TL173I and TL173C are low-cost magnetic-field sensors designed to provide a linear output voltage proportional to the magnetic field they sense. These monolithic circuits incorporate a Hall element as the primary sensor along with a voltage reference and a precision amplifier. Temperature stabilization and internal trimming circuitry yield a device that features high overall sensitivity accuracy with less than 5% error over its operating temperature range.

The TL173I is characterized for operation from -20°C to 85°C . The TL173C is characterized for operation from 0°C to 70°C .

functional block diagram



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	25 V
Continuous total dissipation at (or below) 25°C free-air temperature (see Note 2)	775 mW
Operating free-air temperature range: TL173I	-20°C to 85°C
TL173C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C
Magnetic flux density	unlimited

- NOTES: 1. Voltage values are with respect to network ground terminal.
2. For operation above 25°C free-air temperature, derate linearly at the rate of $6.2\text{ mW}/^{\circ}\text{C}$.

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Special Functions

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TL173I, TL173C LINEAR HALL-EFFECT SENSORS

recommended operating conditions

	TL173I			TL173C			UNIT
	MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, V_{CC}	10.8	12	13.2	10.8	12	13.2	V
Magnetic flux density, B	±50			±50			mT
Output current, I_O	Sink			0.5			mA
	Source			-2			
Operating free-air temperature, T_A	-20	85		0	70		°C

electrical characteristics over full range of recommended operating conditions (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	MIN	TYP‡	MAX	UNIT
V_O Output voltage	$I_O = -2 \text{ mA to } 0.5 \text{ mA}$,	5.8	6	6.2	V
k_{SVS} Supply voltage sensitivity ($\Delta V_O / \Delta V_{CC}$)	$B = 0 \text{ mT}^{\S}$, $T_A = 25^\circ\text{C}$	18			mV/V
S Magnetic sensitivity ($\Delta V_O / \Delta B$)	$B = -50 \text{ to } 50 \text{ mT}^{\S}$, $T_A = 25^\circ\text{C}$	13.5	15	18	V/T [§]
ΔS Magnetic sensitivity change with temperature	$\Delta T_A = 25^\circ\text{C to MIN or MAX}$	±5			%
I_{CC} Supply current	$B = 0 \text{ mT}^{\S}$, $I_O = 0$	8		12	mA
f_{max} Maximum operating frequency		100			kHz

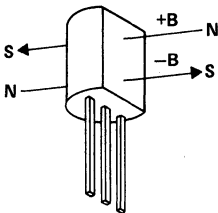
† For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡ Typical values are at $V_{CC} = 12 \text{ V}$ and $T_A = 25^\circ\text{C}$.

§ The unit of magnetic flux density in the International System of Units (SI) is the tesla (T). The tesla is equal to one weber per square meter. Values expressed in milliteslas may be converted to gauss by multiplying by ten, e.g., 50 millitesla = 500 gauss.

4

Special Functions



The north pole of a magnet is the pole that is attracted by the geographical north pole. The north pole of a magnet repels the north-seeking pole of a compass. By accepted magnetic convention, lines of flux emanate from the north pole of a magnet and enter the south pole.

FIGURE 1. DEFINITION OF MAGNETIC FLUX POLARITY

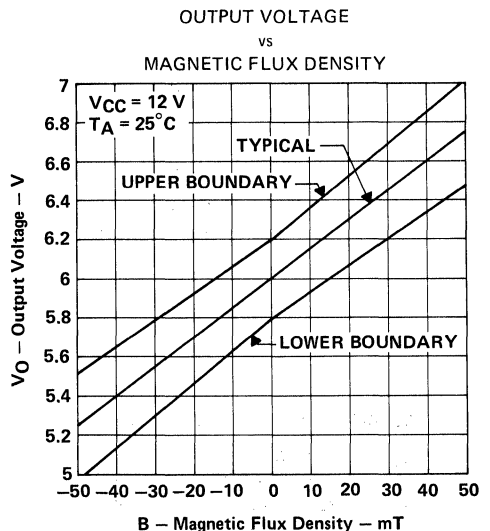


FIGURE 2

TYPICAL APPLICATION DATA

The circuit in Figure 3 may be used to set the output voltage at zero field strength to exactly 6 V (using R1), and to set the sensitivity to exactly -15 V/T (using R2), as depicted in Figure 4.

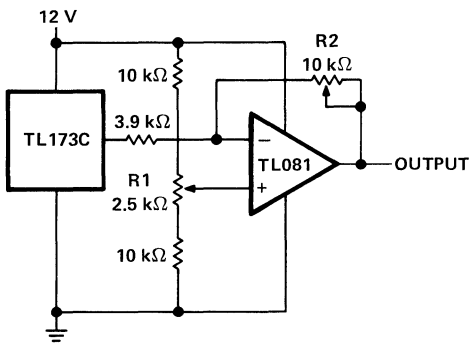


FIGURE 3. COMPENSATION CIRCUIT

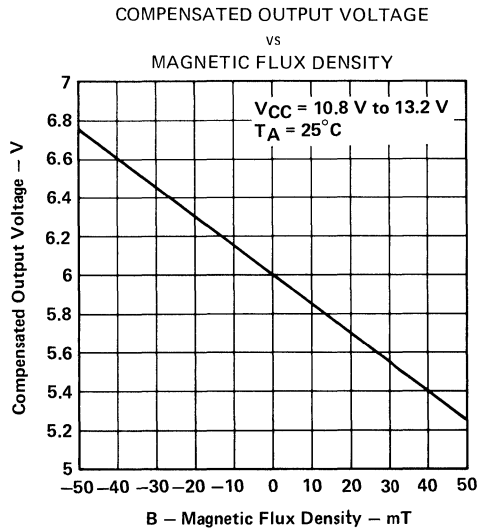


FIGURE 4

4

Special Functions

TL441AM LOGARITHMIC AMPLIFIER

D956, JUNE 1976—REVISED FEBRUARY 1989

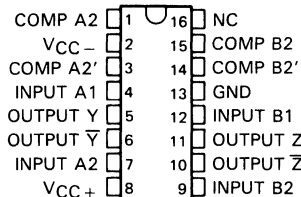
- Excellent Dynamic Range
- Wide Bandwidth
- Built-In Temperature Compensation
- Log Linearity (30 dB Sections) . . . 1 dB Typ
- Wide Input Voltage Range

description

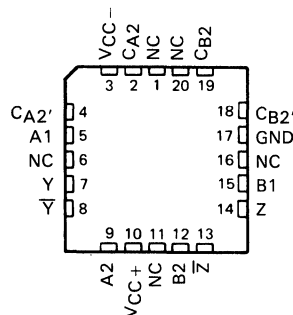
This monolithic amplifier circuit contains four 30-dB logarithmic stages. Gain in each stage is such that the output of each stage is proportional to the logarithm of the input voltage over the 30-dB input voltage range. Each half of the circuit contains two of these 30-dB stages summed together in one differential output that is proportional to the sum of the logarithms of the input voltages of the two stages. The four stages may be interconnected to obtain a theoretical input voltage range of 120 dB. In practice, this permits the input voltage range to be typically greater than 80 dB with log linearity of ± 0.5 dB (see application data). Bandwidth is from dc to 40 MHz.

This circuit is useful in military weapons systems, broadband radar, and infrared reconnaissance systems. It serves for data compression and analog compensation. This logarithmic amplifier is used in log IF circuitry as well as video and log amplifiers. The TL441AM is characterized for operation over the full military temperature range of -55°C to 125°C .

J PACKAGE
(TOP VIEW)

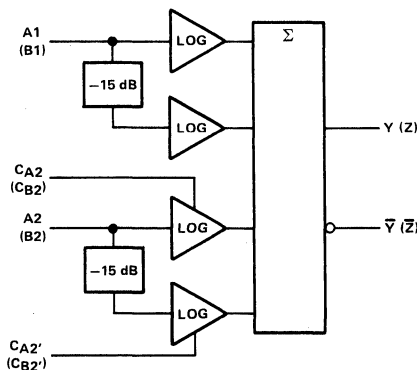


FK PACKAGE
(TOP VIEW)



NC—No internal connection

functional block diagram (one half)



$Y \propto \log A1 + \log A2$; $Z \propto \log B1 + \log B2$
 where: A1, A2, B1, and B2 are in dBV, 0 dBV = 1 V.
 CA2, CA2', CB2, and CB2' are detector compensation inputs.

4
Special Functions

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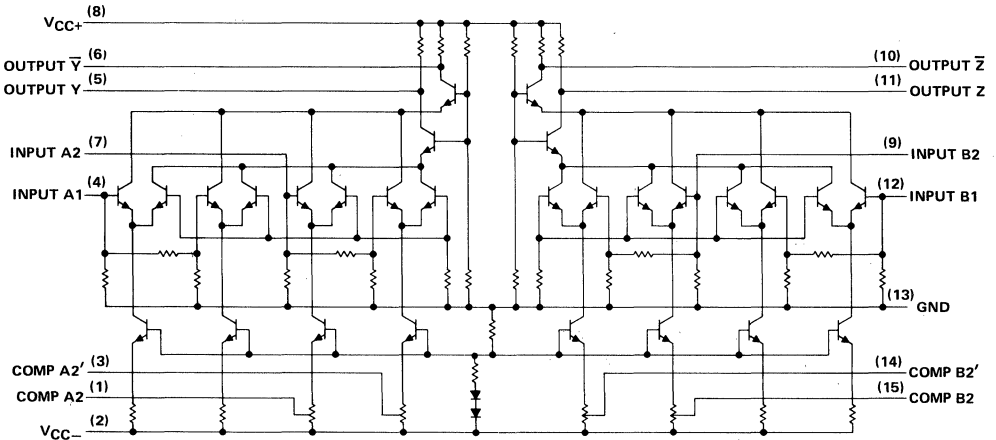


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TL441AM LOGARITHMIC AMPLIFIER

schematic



Pin numbers shown are for the J package.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltages (see Note 1): V_{CC+}	8 V
V_{CC-}	-8 V
Input voltage (see Note 1)	6 V
Output sink current (any one output)	30 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	-55 °C to 125 °C
Storage temperature range	-65 °C to 150 °C
Case temperature for 60 seconds: FK package	260 °C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300 °C

NOTE: 1. All voltages, except differential output voltages, are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
FK	500 mW	11.0 mW/°C	104 °C	500 mW	275 mW
J	500 mW	11.0 mW/°C	104 °C	500 mW	275 mW

recommended operating conditions

	MIN	NOM	MAX	UNIT
Peak-to-peak input voltage for each 30-dB stage	0.01		1	V
Operating free-air temperature, T_A	-55		125	°C

4

Special Functions

electrical characteristics, $V_{CC+} = 6\text{ V}$, $V_{CC-} = -6\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST FIGURE	MIN	TYP	MAX	UNIT
Differential output offset voltage	1		± 25	± 70	mV
Quiescent output voltage	2	5.45	5.6	5.85	V
DC scale factor (differential output), each 3-dB stage, -35 dBV to -5 dBV	3	7	8	11	mV/dB
AC scale factor (differential output)			8		mV/dB
DC error at -20 dBV (midpoint of -35 dBV to -5 dBV range)	3		1	2.6	dB
Input impedance			500		Ω
Output impedance			200		Ω
Rise time, 10% to 90% points, $C_L = 24\text{ pF}$	4		20	35	ns
Supply current from V_{CC+}	2	14.5	18.5	23	mA
Supply current from V_{CC-}	2	-6	-8.5	-10.5	mA
Power dissipation	2	123	162	201	mW

electrical characteristics over operating free-air temperature range, $V_{CC+} = 6\text{ V}$, $V_{CC-} = -6\text{ V}$ (unless otherwise noted)

PARAMETER	TEST FIGURE	MIN	TYP	MAX	UNIT
Differential output offset voltage	1		± 100		mV
Quiescent output voltage	2	5.3		5.85	V
DC scale factor (differential output) each 30-dB stage, -35 dBV to -5 dBV	3	7		11	mV/dB
DC error at -20 dBV (midpoint of -35 dBV to -5 dBV range)	3			4	dB
				3	
Supply current from V_{CC+}	2	10		31	mA
Supply current from V_{CC-}	2	-4.5		-15	mA
Power dissipation	2	87		276	mW

PARAMETER MEASUREMENT INFORMATION

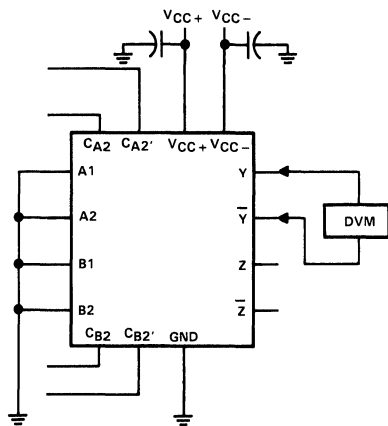


FIGURE 1

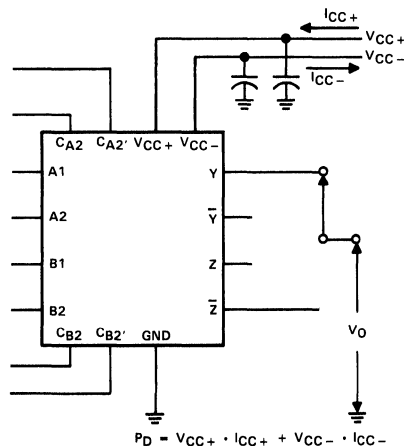
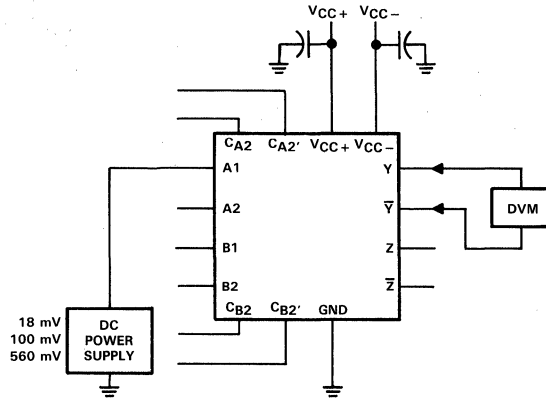


FIGURE 2

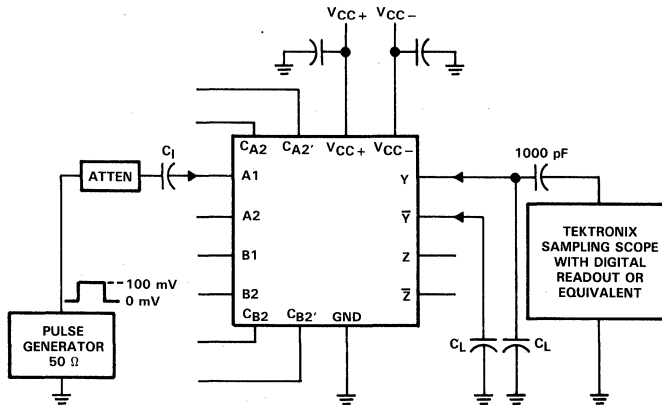
PARAMETER MEASUREMENT INFORMATION



$$\text{Scale Factor} = \frac{[V_{\text{out}}(560 \text{ mV}) - V_{\text{out}}(18 \text{ mV})] \text{ mV}}{30 \text{ dB}}$$

$$\text{Error} = \frac{|V_{\text{out}}(100 \text{ mV}) - 0.5 V_{\text{out}}(560 \text{ mV}) - 0.5 V_{\text{out}}(18 \text{ mV})|}{\text{Scale Factor}}$$

FIGURE 3



- NOTES: A. The input pulse has the following characteristics:
 $t_w = 200 \text{ ns}$, $t_r \leq 2 \text{ ns}$, $t_f \leq 2 \text{ ns}$, $\text{PRR} \leq 10 \text{ MHz}$.
 B. Capacitor C_1 consists of three capacitors in parallel: $1 \mu\text{F}$, $0.1 \mu\text{F}$, and $0.01 \mu\text{F}$.
 C. C_L includes probe and jig capacitance.

FIGURE 4

4

Special Functions

TYPICAL CHARACTERISTICS

DIFFERENTIAL OUTPUT OFFSET VOLTAGE
vs
FREE-AIR TEMPERATURE

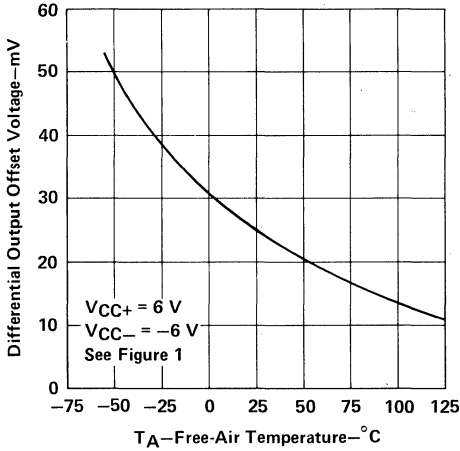


FIGURE 5

QUIESCENT OUTPUT VOLTAGE
vs
FREE-AIR TEMPERATURE

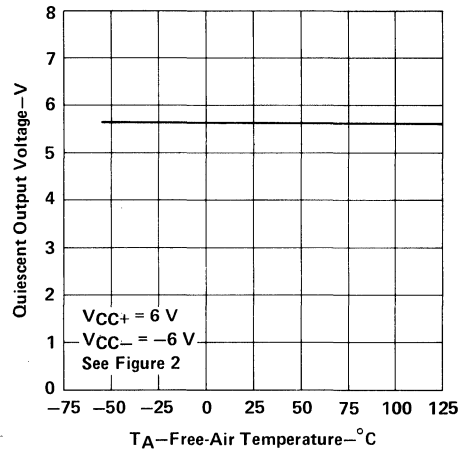


FIGURE 6

DC SCALE FACTOR
vs
FREE-AIR TEMPERATURE

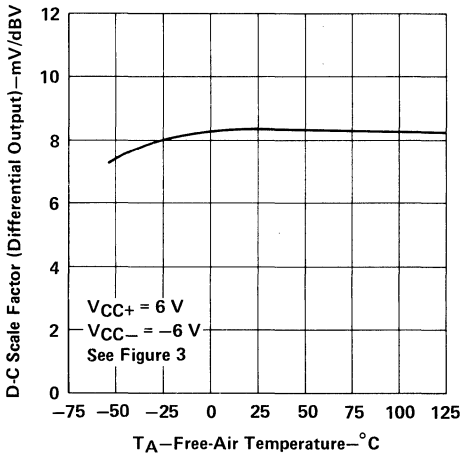


FIGURE 7

DC ERROR
vs
FREE-AIR TEMPERATURE

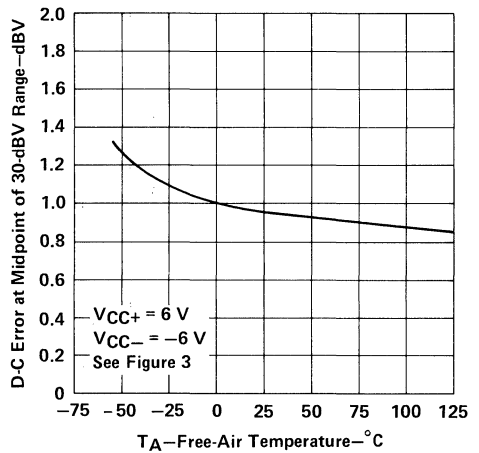


FIGURE 8

TYPICAL CHARACTERISTICS

OUTPUT RISE TIME
 vs
 LOAD CAPACITANCE

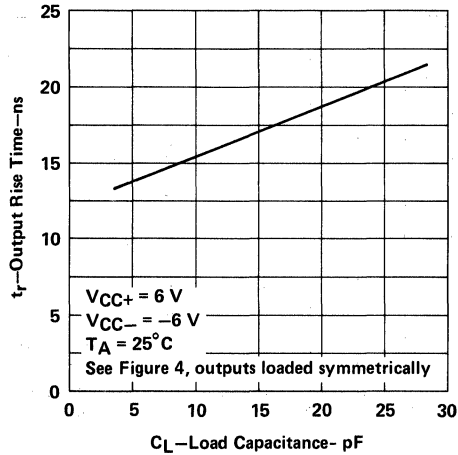


FIGURE 9

POWER DISSIPATION
 vs
 FREE-AIR TEMPERATURE

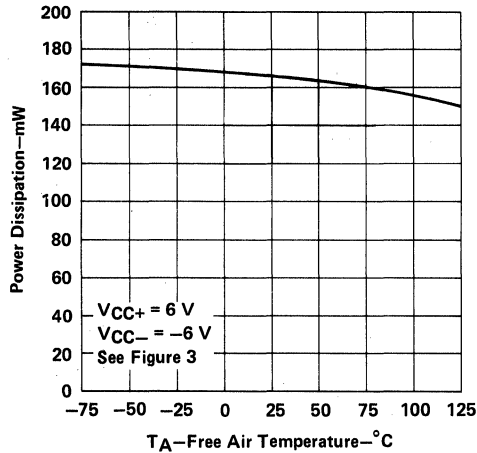


FIGURE 10

TYPICAL APPLICATION DATA

Although designed for high-performance applications such as broadband radar infrared detection and weapons systems, this device has a wide range of applications in data compression and analog computation.

basic logarithmic function

The basic logarithmic response is derived from the exponential current-voltage relationship of collector current and base-emitter voltage. This relationship is given in the equation:

$$m \cdot V_{BE} = \ln [(I_C + I_{CES})/I_{CES}]$$

where:

- I_C = collector current
- I_{CES} = collector current at $V_{BE} = 0$
- $m = q/kT$ (in V^{-1})
- V_{BE} = base-emitter voltage

The differential input amplifier allows dual-polarity inputs, is self-compensating for temperature variations, and is relatively insensitive to common-mode noise.

Four compensation points are made available to allow slight variations in the gain (slope) of the two individual 15-dB stages of input A2 and B2. By slightly changing the voltage on any of the compensation pins from its quiescent value, the gain of that particular 15-dB stage can be adjusted to match the other 15-dB stage in the pair. The compensation pins may also be used to match the transfer characteristics of input A2 to A1 or B2 to B1.

The log stages in each half of the circuit are summed by directly connecting their collectors together and summing through a common-base output stage. The two sets of output collectors are used to give two log outputs, Y and \bar{Y} (or Z and \bar{Z}) which are equal in amplitude but opposite in polarity. This increases the versatility of the device.

By proper choice of external connections, linear amplification, linear attenuation, and many different applications requiring logarithmic signal processing are possible.

functional block diagram

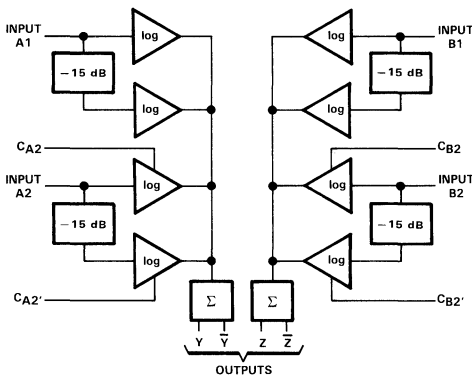


FIGURE 11

logarithmic sections

As can be seen from the schematic, there are eight differential pairs. Each pair is a 15-dB log subsection, and each input feeds two pairs for a range of 30-dB per stage.

input levels

The recommended input voltage range of any one stage is given as 0.01 V to 1 V. Input levels in excess of 1 V may result in a distorted output. When several log sections are summed together, the distorted area of one section overlaps with the next section and the resulting distortion is insignificant. However, there is a limit to the amount of overdrive that may be applied. As the input drive reaches ± 3.5 V, saturation occurs, clamping the collector-summing line and severely distorting the output. Therefore, the signal to any input must be limited to approximately ± 3 V to ensure a clean output.

output levels

Differential-output-voltage levels are low, generally less than 0.6 V. As demonstrated in Figure 12, the output swing and the slope of the output response can be adjusted by varying the gain by means of the slope control. The coordinate origin may also be adjusted by positioning the offset of the output buffer.

TL441AM LOGARITHMIC AMPLIFIER

TYPICAL APPLICATION DATA

circuits

Figures 12 through 19 show typical circuits using this logarithmic amplifier. Operational amplifiers not otherwise designated are TLC271. For operation at higher frequencies, the TL592 is recommended instead of the TLC271.

TYPICAL TRANSFER CHARACTERISTICS

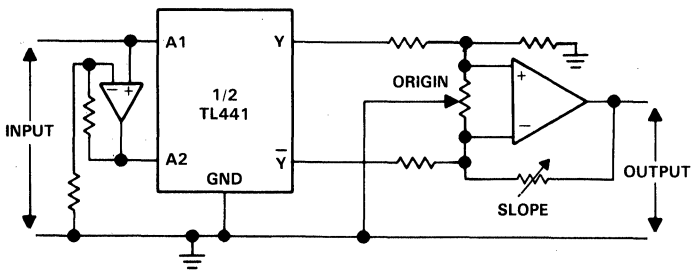
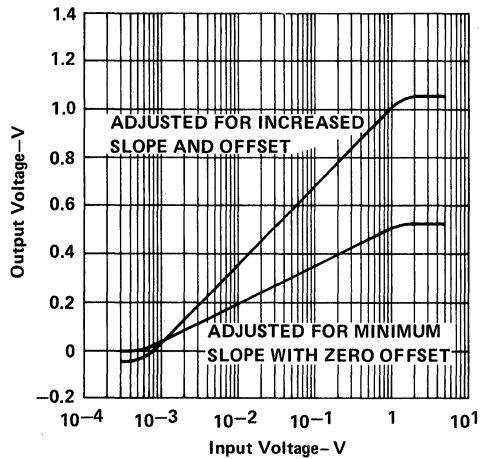


FIGURE 12. OUTPUT SLOPE AND ORIGIN ADJUSTMENT

4

Special Functions

TYPICAL APPLICATION DATA

**TRANSFER CHARACTERISTICS
OF TWO TYPICAL INPUT STAGES**

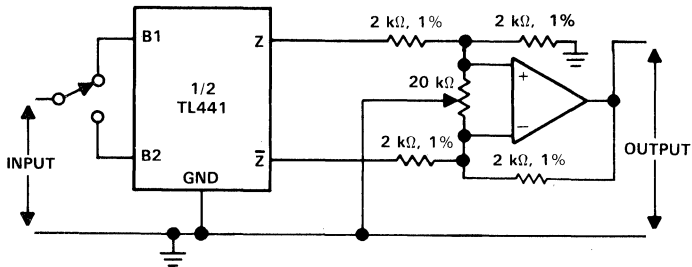
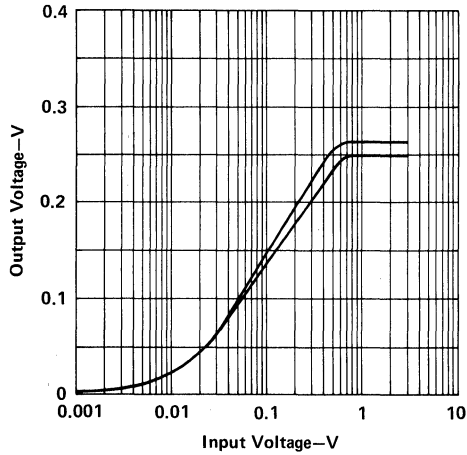


FIGURE 13. UTILIZATION OF SEPARATE STAGES

TYPICAL APPLICATION DATA

TRANSFER CHARACTERISTICS
WITH BOTH SIDES PARALLELED

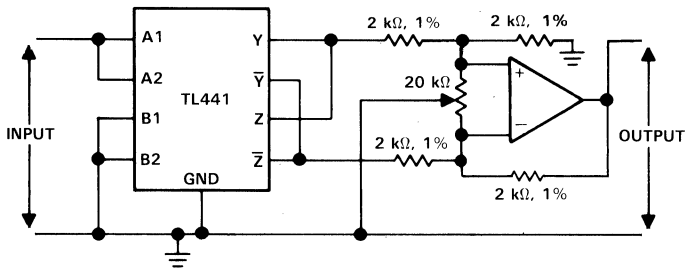
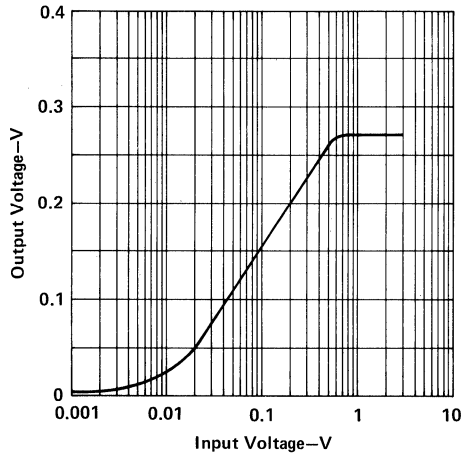
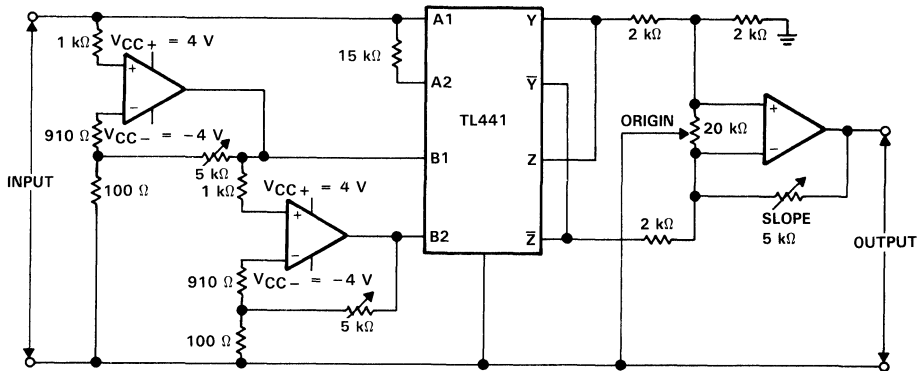
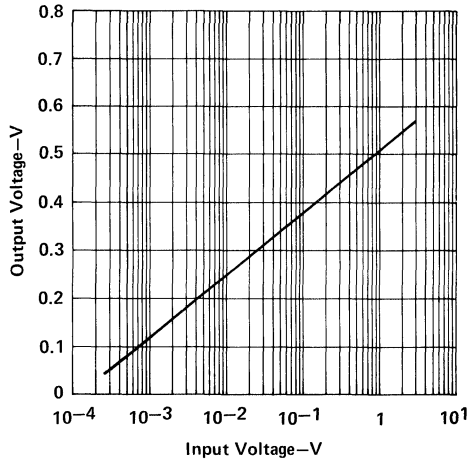


FIGURE 14. UTILIZATION OF PARALLELED INPUTS

TYPICAL APPLICATION DATA

TRANSFER CHARACTERISTICS



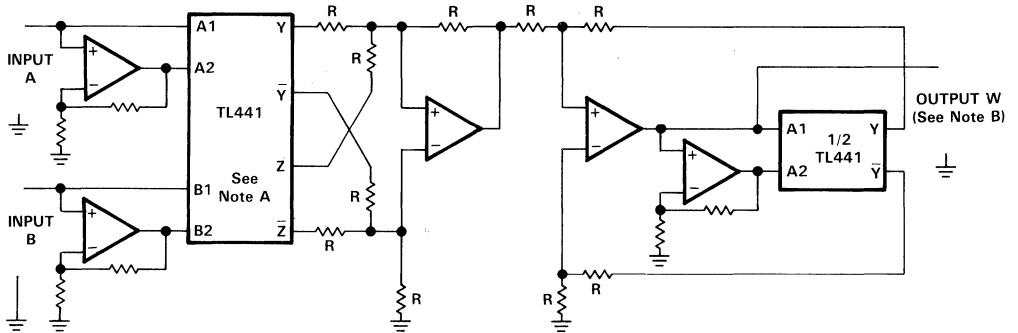
- NOTES: A. Inputs are limited by reducing the supply voltages for the input amplifiers to ± 4 V.
 B. The gains of the input amplifiers are adjusted to achieve smooth transitions.

Figure 15. LOGARITHMIC AMPLIFIER WITH INPUT VOLTAGE RANGE GREATER THAN 80 dB

**4
Special Functions**

TL441AM LOGARITHMIC AMPLIFIER

TYPICAL APPLICATION DATA



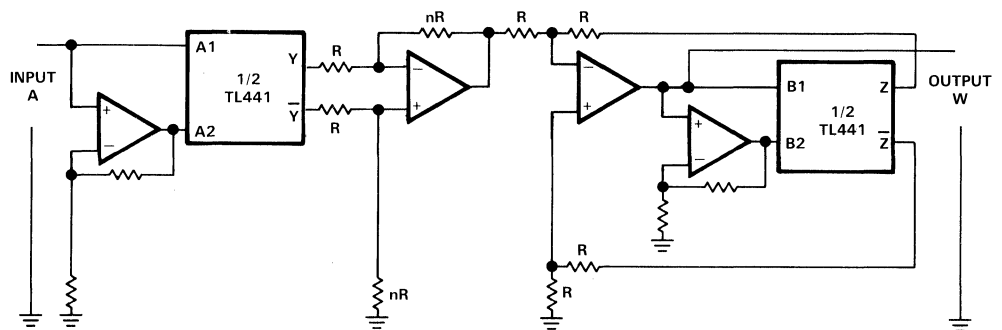
- NOTES: A. Connections shown are for multiplication. For division, Z and \bar{Z} connections are reversed.
 B. Output W may need to be amplified to give actual product or quotient of A and B.
 C. R designates resistors of equal value, typically 2 kΩ to 10 kΩ.

Multiplication: $W = A \cdot B \Rightarrow \log W = \log A + \log B$, or $W = a(\log_a A + \log_a B)$

Division: $W = A/B \Rightarrow \log W = \log A - \log B$, or $W = a(\log_a A - \log_a B)$

FIGURE 16. MULTIPLICATION OR DIVISION

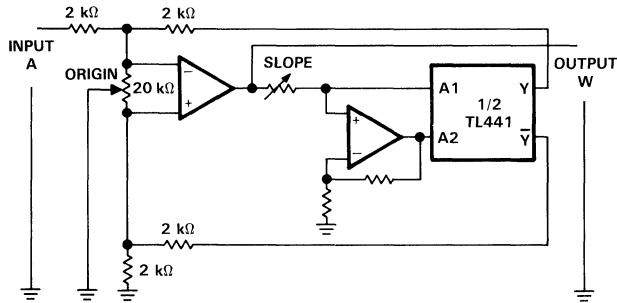
4 Special Functions



- NOTE: R designates resistors of equal value, typically 2 kΩ to 10 kΩ. The power to which the input variable is raised is fixed by setting nR. Output W may need to be amplified to give the correct value.
 Exponential: $W = A^n \Rightarrow \log W = n \log A$, or $W = a(n \log_a A)$

FIGURE 17. RAISING A VARIABLE TO A FIXED POWER

TYPICAL APPLICATION DATA



NOTE: Adjust the slope to correspond to the base "a".
Exponential to any base: $W = a$

FIGURE 18. RAISING A FIXED NUMBER TO A VARIABLE POWER

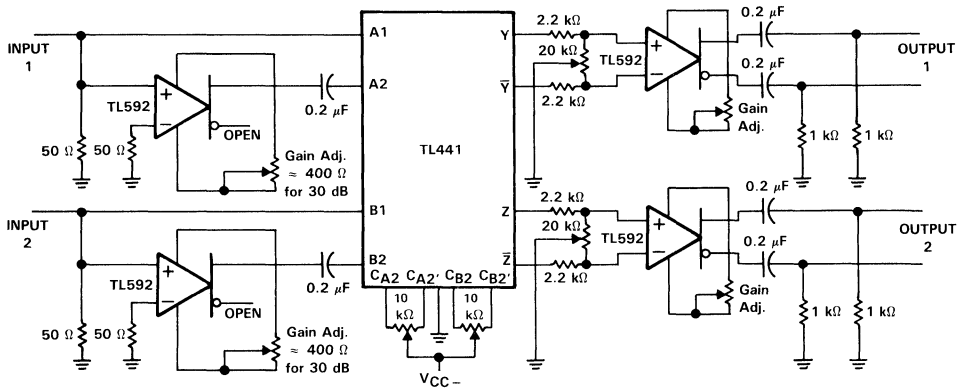


FIGURE 19. DUAL-CHANNEL RF LOGARITHMIC AMPLIFIER WITH 50-dB INPUT RANGE PER CHANNEL AT 10 MHz

4
Special Functions

4

Special Functions

TL592, TL592A DIFFERENTIAL VIDEO AMPLIFIERS

D2668, NOVEMBER 1983—REVISED MAY 1988

- 8-Pin Version of NE592 . . . Saves Printed Circuit Board Space
- Adjustable Gain to 400
- No Frequency Compensation Required
- Adjustable Passband

DEVICE TYPE	TEMPERATURE RANGE	A _{VD} RANGE (GAIN OPTION 1)
TL592	0°C to 70°C	250—600
TL592A	0°C to 70°C	400—600

description

This device is a monolithic two-stage video amplifier with differential inputs and differential outputs.

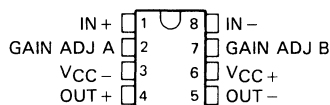
Internal series-shunt feedback provides wide bandwidth, low phase distortion, and excellent gain stability. Emitter-follower outputs enable the device to drive capacitive loads. All stages are current-source biased to obtain high common-mode and supply-voltage rejection ratios.

Fixed differential amplification of nominally 400 may be selected without external components, or amplification may be adjusted from 0 to approximately 400 by the use of a single external resistor connected between the gain adjustment pins A and B. No external frequency-compensating components are required for any gain option.

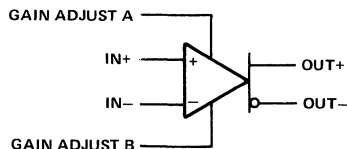
The device is particularly useful in magnetic-tape or disc-file systems using phase or NRZ encoding and in high-speed thin-film or plated-wire memories. Other applications include general-purpose video and pulse amplifiers where wide bandwidth, low phase shift, and excellent gain stability are required.

The TL592 and TL592A are characterized for operation from 0°C to 70°C.

D OR P PACKAGE
(TOP VIEW)

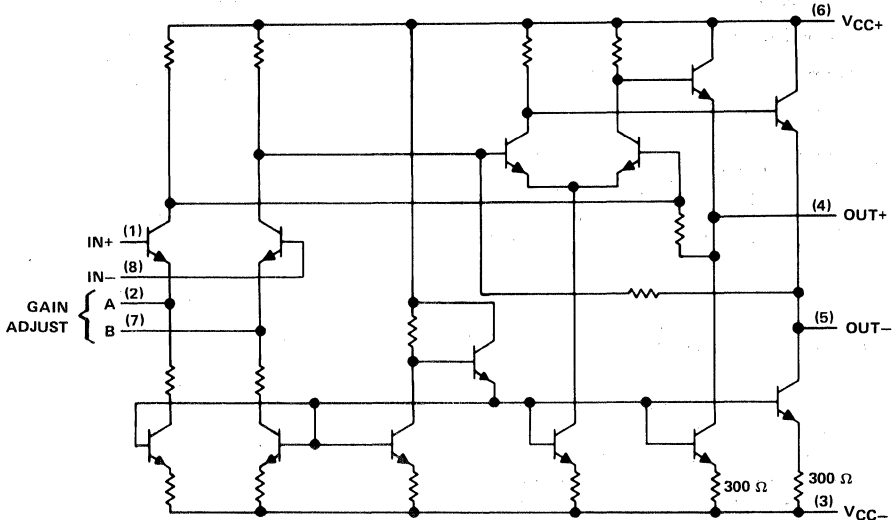


symbol



TL592, TL592A DIFFERENTIAL VIDEO AMPLIFIERS

schematic



absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	8 V
Supply voltage, V_{CC-} (see Note 1)	-8 V
Differential input voltage	± 5 V
Voltage range, any input	V_{CC+} to V_{CC-}
Output current	10 mA
Continuous total power dissipation at 70°C	500 mW
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values except differential input voltages are with respect to the midpoint between V_{CC+} and V_{CC-} .

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{CC+}	3	6	8	V
Supply voltage, V_{CC-}	-3	-6	-8	V
Operating free-air temperature, T_A	0		70	°C

4

Special Functions

electrical characteristics at specified free-air temperature, $V_{CC+} = 6\text{ V}$, $V_{CC-} = -6\text{ V}$, $R_L = 2\text{ k}\Omega$ (unless otherwise noted)

PARAMETER	TEST FIGURE	TEST CONDITIONS		GAIN OPTION†	TL592			TL592A			UNIT	
					MIN	TYP	MAX	MIN	TYP	MAX		
A _{VD}	Large-signal differential voltage amplification	1	$V_{OPP} = 3\text{ V}$, $R_L = 2\text{ k}\Omega$	25°C	1	250	400	600	400	440	600	V/V
				0°C to 70°C		250		600	400		600	
BW	Bandwidth (-3 dB)	2	$V_{OPP} = 1\text{ V}$	25°C	1	50			50		MHz	
I _{IO}	Input offset current		$V_{IC} = 0$	25°C	1 or 2	0.4	5		0.4	5	μA	
				0°C to 70°C			6		6			
I _{IB}	Input bias current		$V_{IC} = 0$	25°C	1 or 2	9	30		10	30	μA	
				0°C to 70°C			40		40			
V _{ICR}	Common-mode input voltage range	3		25°C	1 or 2	±1			±1			V
				0°C to 70°C		±1			±1			
V _{OC}	Common-mode output voltage	1	$R_L = \infty$	25°C	2	2.4	2.9	3.4	2.4	2.9	3.4	V
V _{OO}	Output offset voltage	1	$V_{ID} = 0$, $R_L = \infty$	25°C	2	0.35	0.75		0.35	0.75	V	
				0°C to 70°C			1.5		1.5			
V _{OPP}	Peak-to-peak output voltage swing	1	$R_L = 2\text{ k}\Omega$	25°C	1	3	4		3	4	V	
				0°C to 70°C		2.8		2.8				
z _i	Input impedance		$V_{OD} = 1\text{ V}$, $f = 1\text{ kHz to } 10\text{ MHz}$	25°C	1	4			3.6			kΩ
				0°C to 70°C		3.6			3.3			
CMRR	Common-mode rejection ratio	3	$V_{IC} = \pm 1\text{ V}$	f = 100 kHz	1	60	86		60	86	dB	
				f = 5 MHz		60		60				
				f = 100 kHz		50		50				
				f = 5 MHz		60		60				
k _{SVR}	Supply voltage rejection ratio ($\Delta V_{CC}/\Delta V_{IO}$)	4	$\Delta V_{CC+} = \pm 0.5\text{ V}$, $\Delta V_{CC-} = \pm 0.5\text{ V}$	25°C	1	50	70		50	70	dB	
				0°C to 70°C		50		50				
V _n	Broadband equivalent input noise voltage	4	$BW = 1\text{ kHz to } 10\text{ MHz}$	25°C	1 or 2	12			12	μV		
t _{pd}	Propagation delay time	2	$\Delta V_O = 1\text{ V}$	25°C	2	7.5			7.5	ns		
t _r	Rise time	2	$\Delta V_O = 1\text{ V}$	25°C	2	10.5			10.5	ns		
I _{sink(max)}	Maximum output sink current				1, 2, or 3	3	4		3	4	mA	
I _{CC}	Supply current		No load, No signal	25°C	1 or 2	18	24		19	24	mA	
				0°C to 70°C			27		27			

†The gain option is selected as follows:

- Gain Option 1 . . . Gain adjust pin A is connected to pin B.
- Gain Option 2 . . . Gain adjust pins A and B are open.

TL592, TL592A
DIFFERENTIAL VIDEO AMPLIFIERS

PARAMETER MEASUREMENT INFORMATION

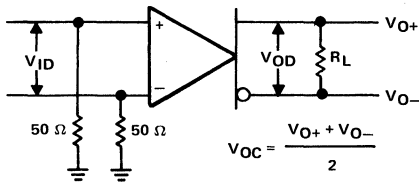


FIGURE 1

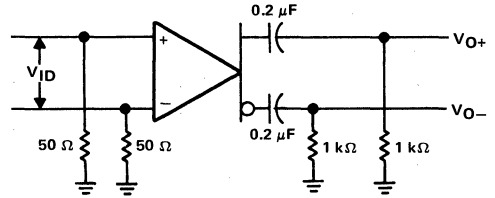


FIGURE 2

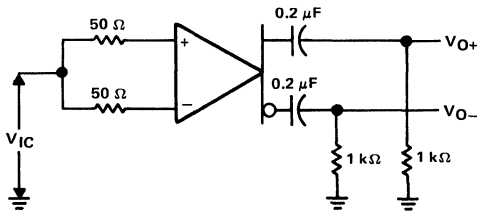


FIGURE 3

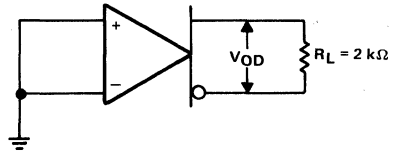


FIGURE 4

4

Special Functions

TL592B DIFFERENTIAL VIDEO AMPLIFIER

D2668, JUNE 1985—REVISED APRIL 1988

- Adjustable Gain to 400 Typ
- No Frequency Compensation Required
- Low Noise . . . 3 μ V Typ V_N

description

This device is a monolithic two-stage video amplifier with differential inputs and differential outputs. It features internal series-shunt feedback that provides wide bandwidth, low phase distortion, and excellent gain stability. Emitter-follower outputs enable the device to drive capacitive loads. All stages are current-source biased to obtain high common-mode and supply-voltage rejection ratios.

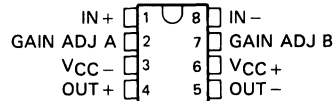
The differential gain is typically 400 when the gain adjust pins are connected together, or amplification may be adjusted from near 0 to 400 by the use of a single external resistor connected between the gain adjustment pins A and B. No external frequency-compensating components are required for any gain option.

The device is particularly useful in magnetic-tape or disk-file systems using phase or NRZ encoding and in high-speed thin-film or plated-wire memories. Other applications include general-purpose video and pulse amplifiers.

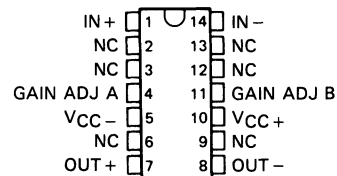
The device achieves low equivalent noise voltage through special processing and a new circuit layout incorporating input transistors with low base resistance.

The TL592B is characterized for operation from 0°C to 70°C.

D8[†] OR P PACKAGE
(TOP VIEW)



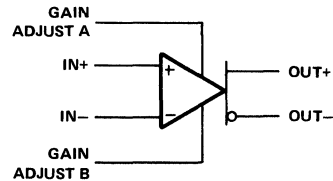
D14[†] OR N PACKAGE
(TOP VIEW)



NC—No internal connection

[†]D8 and D14 are the codes used to differentiate the 8-pin and 14-pin versions, respectively.

symbol

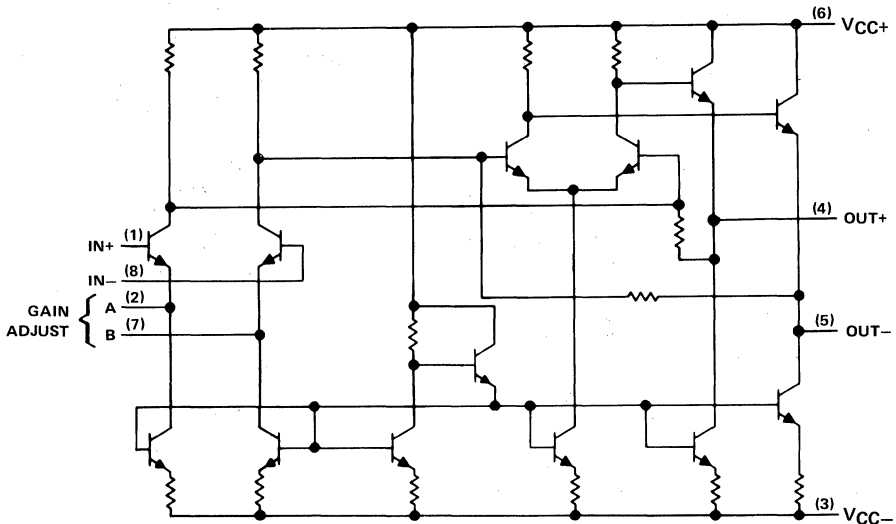


4

Special Functions

TL592B DIFFERENTIAL VIDEO AMPLIFIER

schematic



absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	8 V
Supply voltage, V_{CC-}	-8 V
Differential input voltage	± 5 V
Voltage range, any input	V_{CC+} to V_{CC-}
Output current	10 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTES: 1. All voltage values except differential input voltages are with respect to the midpoint between V_{CC+} and V_{CC-} .

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE T_A	$T_A = 70^\circ\text{C}$ POWER RATING
D8	530 mW	5.8 mW/°C	59°C	464 mW
D14	530 mW	N/A	N/A	530 mW
N	530 mW	N/A	N/A	530 mW
P	530 mW	N/A	N/A	530 mW

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{CC+}	3	6	8	V
Supply voltage, V_{CC-}	-3	-6	-8	V
Operating free-air temperature, T_A	0		70	°C

4

Special Functions

TL592B
DIFFERENTIAL VIDEO AMPLIFIER

electrical characteristics at specified free-air temperature, $V_{CC+} = 6\text{ V}$, $V_{CC-} = -6\text{ V}$, $R_L = 2\text{ k}\Omega$ (unless otherwise noted)

PARAMETER		TEST FIGURE	TEST CONDITIONS [†]		MIN	TYP	MAX	UNIT		
A _{VD}	Large-signal differential voltage amplification	1	V _{OPP} = 3 V, R _{AB} = 0	R _L = 2 kΩ,	25°C	300	400	500	V/V	
					0°C to 70°C	250		600		
A _{VD2}	Large-signal differential voltage amplification	1	V _{OPP} = 3 V, R _{AB} = 1 kΩ	R _L = 2 kΩ,	25°C		13	V/V		
BW	Bandwidth (-3 dB)	2	V _{OPP} = 1 V, R _{AB} = 0		25°C		50	MHz		
I _{IO}	Input offset current				25°C		0.4	5	μA	
					0°C to 70°C					6
I _{IB}	Input bias current				25°C		9	30	μA	
					0°C to 70°C					40
V _{ICR}	Common-mode input voltage range	3			25°C		±1		V	
					0°C to 70°C			±1		
V _{OC}	Common-mode output voltage	1	R _L = ∞		25°C	2.4	2.9	3.4	V	
V _{OO}	Output offset voltage	1	V _{ID} = 0, R _{AB} = ∞, R _L = ∞		25°C		0.35	0.75	V	
					0°C to 70°C					1.5
V _{OPP}	Peak-to-peak output voltage swing	1	R _L = 2 kΩ	R _{AB} = 0	25°C		3	4	V	
					0°C to 70°C					2.8
r _i	Input resistance		V _{OD} = 1 V,	R _{AB} = 0	25°C			4	kΩ	
					0°C to 70°C					3.6
r _o	Output resistance				0°C to 70°C			30	Ω	
C _i	Input capacitance				25°C			5	pF	
CMRR	Common-mode rejection ratio	3	V _{IC} = ±1 V, R _{AB} = 0		f = 100 kHz	25°C		60	86	dB
					f = 5 MHz				60	
					f = 100 kHz	0°C to 70°C			50	
					f = 5 MHz				60	
k _{SVR}	Supply voltage rejection ratio (ΔV _{CC} /ΔV _{IO})	4	ΔV _{CC+} = ±0.5 V, ΔV _{CC-} = ±0.5 V, R _{AB} = 0		25°C		50	70	dB	
					0°C to 70°C			50		
V _n	Broadband equivalent input noise voltage	4		BW = 1 kHz to 10 MHz	25°C			3	μV	
t _{pd}	Propagation delay time	2		ΔV _O = 1 V	25°C			7.5	ns	
t _r	Rise time	2		ΔV _O = 1 V	25°C			10.5	ns	
I _{sink(max)}	Maximum output sink current		V _{ID} = 1 V,	V _O = 3 V				3	4	mA
I _{CC}	Supply current		No load,	No signal	25°C			18	24	mA
					0°C to 70°C					

[†]R_{AB} is the gain-adjustment resistor connected between gain-adjust pins A and B. If not specified for a particular parameter, its value is irrelevant to that parameter.

4
Special Functions

TL592B DIFFERENTIAL VIDEO AMPLIFIER

PARAMETER MEASUREMENT INFORMATION

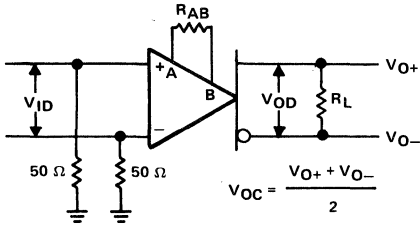


FIGURE 1

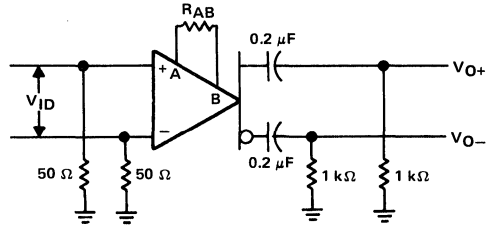


FIGURE 2

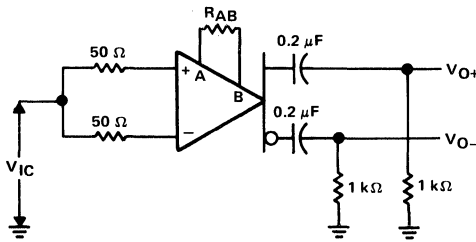


FIGURE 3

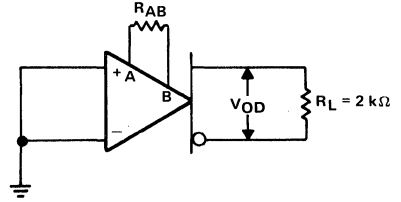


FIGURE 4

TYPICAL CHARACTERISTICS

4

Special Functions

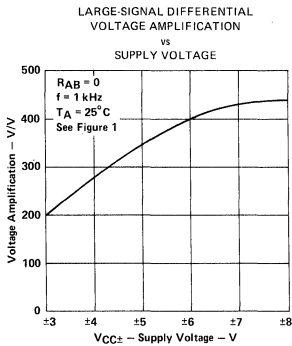


FIGURE 5

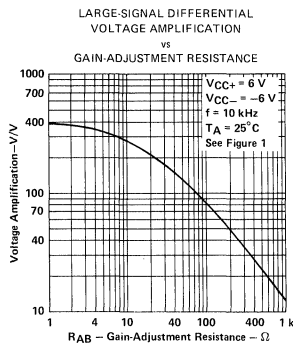


FIGURE 6

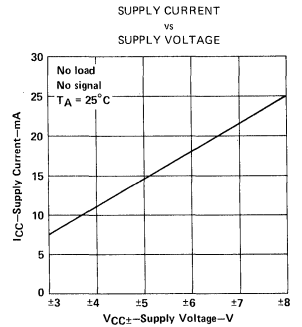


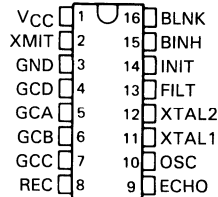
FIGURE 7

TL851 SONAR RANGING CONTROL

D2760, SEPTEMBER 1983—REVISED MARCH 1988

- **Designed for Use with the TL852 in Sonar Ranging Modules Like the SN28827**
- **Operates with Single Supply**
- **Accurate Clock Output for External Use**
- **Synchronous 4-Bit Gain Control Output**
- **Internal 1.2-V Level Detector for Receive**
- **TTL-Compatible**
- **Interfaces to Electrostatic or Piezoelectric Transducers**

N DUAL-IN-LINE PACKAGE
(TOP VIEW)



description

The TL851 is an economical digital I²L ranging control integrated circuit designed for use with the Texas Instruments TL852 Sonar ranging receiver integrated circuit.

The TL851 is designed for distance measurement from six inches to 35 feet. The device has an internal oscillator that uses a low-cost external ceramic resonator. With a simple interface and a 420-kHz ceramic resonator, the device will drive a 50-kHz electrostatic transducer.

The device cycle begins when Initiate (INIT) is taken to the high logic level. There must be at least 5 ms from initial power up (V_{CC}) to the first initiate signal in order for all the device internal latches to reset and for the ceramic-resonator-controlled oscillator to stabilize. The device will transmit a burst of 16 pulses each time INIT is taken high.

The oscillator output (OSC) is enabled by INIT. The oscillator frequency is the ceramic resonator frequency divided by 8.5 for the first 16 cycles (during transmit) and then the oscillator frequency changes to the ceramic resonator frequency divided by 4.5 for the remainder of the device cycle.

When used with an external 420 kilohertz ceramic resonator, the device internal blanking disables the receive input (REC) for 3.8 ms after initiate to exclude false receive inputs that may be caused by transducer ringing. The internal blanking feature also eliminates echos from objects closer than 1.3 feet from the transducer. If it is necessary to detect objects closer than 1.3 feet, then the internal blanking may be shortened by taking the blanking inhibit (BINH) high, enabling the receive input. The blanking input (BLNK) may be used to disable the receive input and reset ECHO to a low logic level at any time during the device cycle for selective echo exclusion or for a multiple-echo mode of operation.

The device provides a synchronous 4-bit gain control output (12 steps) designed to control the gain of the TL852 sonar ranging receiver integrated circuit. The digital gain control waveforms are shown in Figure 2 with the nominal transition times from INIT listed in the Gain Control Output Table.

The threshold of the internal receive level detector is 1.2 volts. The TL851 operates over a supply voltage range of 4.5 volts to 6.8 volts and is characterized for operation from 0°C to 40°C.

4

Special Functions

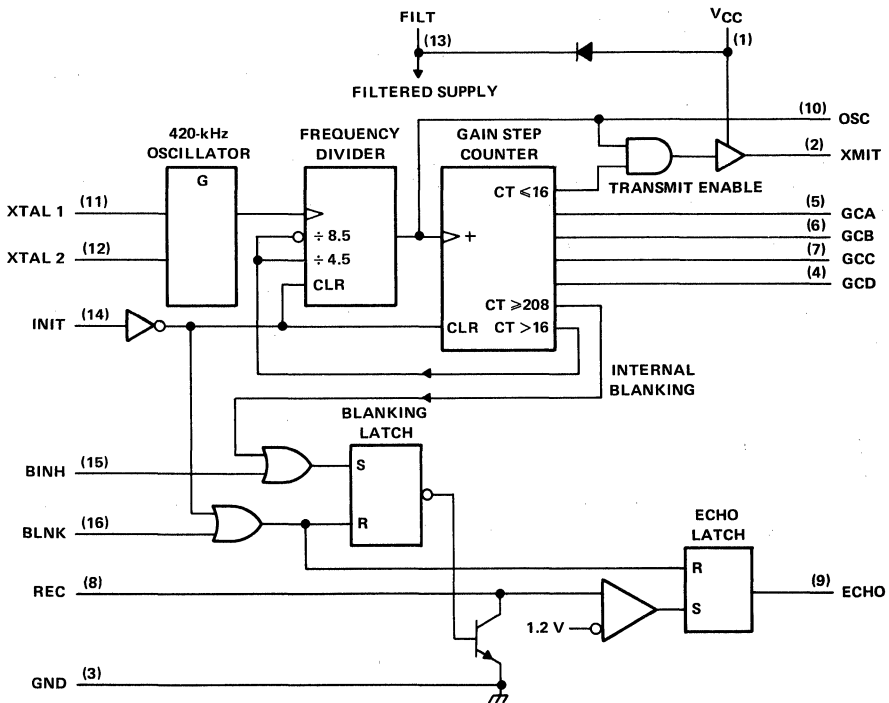
**TL851
SONAR RANGING CONTROL**

GAIN CONTROL OUTPUT TABLE

STEP NUMBER	GCD	GCC	GCB	GCA	TIME (ms) FROM INITIATE †
0	L	L	L	L	2.38 ms
1	L	L	L	H	5.12 ms
2	L	L	L	L	7.87 ms
3	L	L	H	H	10.61 ms
4	L	H	L	L	13.35 ms
5	L	H	L	H	16.09 ms
6	L	H	H	L	18.84 ms
7	L	H	H	H	21.58 ms
8	H	L	L	L	27.07 ms
9	H	L	L	H	32.55 ms
10	H	L	H	L	38.04 ms
11	H	L	H	H	INIT I

† This is the time to the end of the indicated step and assumes a nominal 420-kHz ceramic resonator.

functional block diagram



**4
Special Functions**

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Voltage at any pin with respect to GND -0.5 V to 7 V
 Voltage at any pin with respect to V_{CC} -7 V to 0.5 V
 Continuous power dissipation at (or below) 25 °C free-air temperature (see Note 1) 1150 mW
 Operating free-air temperature range 0 °C to 70 °C
 Storage temperature range -65 °C to 150 °C
 Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds 260 °C

NOTE 1: For operation above 25 °C, derate linearly at the rate of 9.2 mW/°C.

recommended operating conditions

		MIN	MAX	UNIT
Supply voltage, V _{CC}		4.5	6.8	V
High-level input voltage, V _{IH}	BLNK, BINH, INIT	2.1		V
Low-level input voltage, V _{IL}	BLNK, BINH, INIT		0.6	V
Delay time, power up to INIT high		5		ms
Operating free-air temperature, T _A		0	40	°C

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER		TEST CONDITIONS	MIN	TYP [†]	MAX	UNIT
Input current	BLNK, BINH, INIT	V _I = 2.1 V			1	mA
High-level output current, I _{OH}	ECHO, OSC, GCA, GCB, GCC, GCD	V _{OH} = 5.5 V			100	μA
Low-level output voltage, V _{OL}	ECHO, OSC, GCA, GCB, GCC, GCD	I _{OL} = 1.6 mA			0.4	V
On-state output current	SMIT output	V _O = 1 V		-140		mA
Internal blanking interval	REC input			2.38 [‡]		ms
Frequency during 16-pulse transmit period	OSC output			49.4 [‡]		kHz
	XMIT output			49.4 [‡]		
Frequency after 16-pulse transmit period	OSC output			93.3 [‡]		kHz
	XMIT output			0		
Supply current, I _{CC}	During transmit period				260	mA
	After transmit period				55	

[†]Typical values are at V_{CC} = 5 V and T_A = 25 °C.

[‡]These typical values apply for a 420-kHz ceramic resonator.

TL851 SONAR RANGING CONTROL

schematics of inputs and outputs

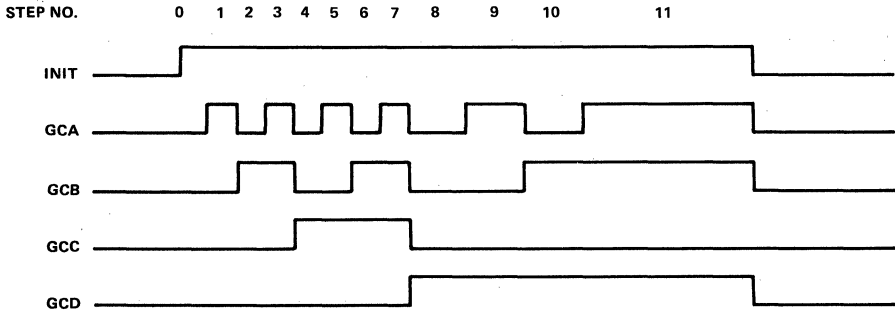
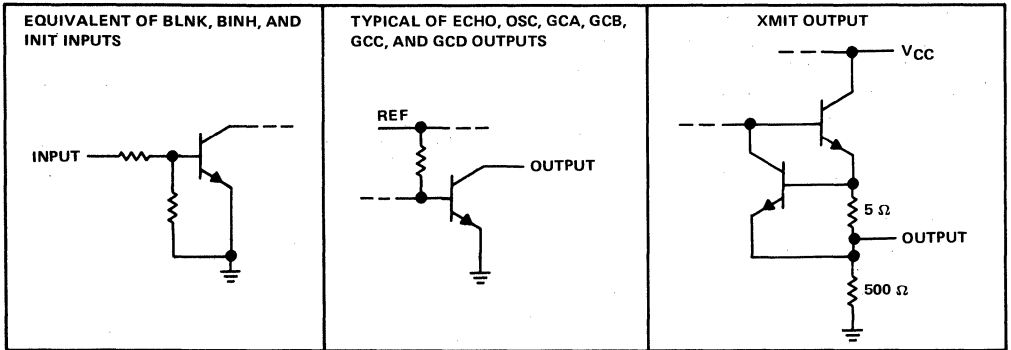


FIGURE 1. DIGITAL GAIN CONTROL WAVEFORMS

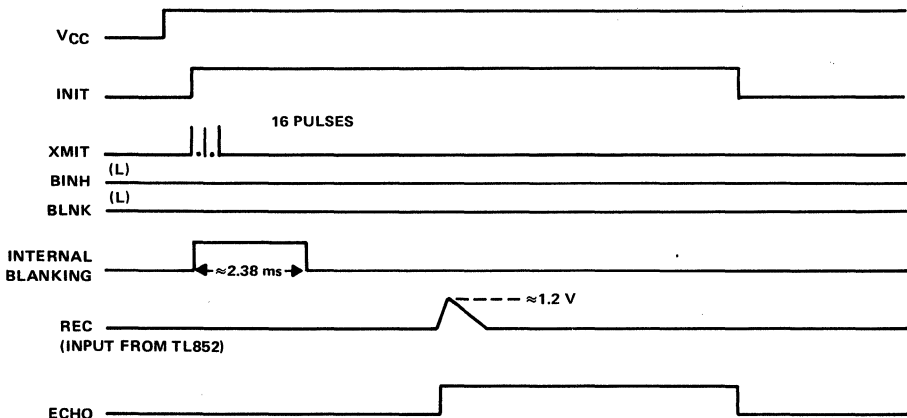


FIGURE 2. EXAMPLE OF SINGLE-ECHO-MODE CYCLE WHEN USED WITH THE TL852 RECEIVER AND 420-KHz CERAMIC RESONATOR

4

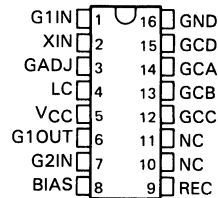
Special Functions

TL852 SONAR RANGING RECEIVER

D2779, SEPTEMBER 1983—REVISED MARCH 1988

- **Designed for Use with the TL851 in Sonar Ranging Modules Like the SN28827**
- **Digitally Controlled Variable-Gain Variable-Bandwidth Amplifier**
- **Operational Frequency Range of 20 kHz to 90 kHz**
- **TTL-Compatible**
- **Operates from Power Sources of 4.5 V to 6.8 V**
- **Interfaces to Electrostatic or Piezoelectric Transducers**
- **Overall Gain Adjustable with One External Resistor**

N DUAL-IN-LINE PACKAGE
(TOP VIEW)



NC—No internal connection

description

The TL852 is an economical sonar ranging receiver integrated circuit for use with the TL851 control integrated circuit. A minimum of external components is required for operation, and this amplifier easily interfaces to Polaroid's 50-kilohertz electrostatic transducer. An external 68-kilohm $\pm 5\%$ resistor from pin 8 (Bias) to pin 16 (GND) provides the internal biasing reference. Amplifier gain can be set with a resistor from pin 1 (G1IN) to pin 3 (GADJ). Required amplifier gain will vary for different applications. Using the detect-level measurement circuit of Figure 1, a nominal peak-to-peak value of 230 millivolts input during gain step 2 is recommended for most applications. For reliable operation, a level no lower than 50 millivolts should be used. The recommended detect level of 230 millivolts can be obtained for most amplifiers with an R1 value between 5 kilohms and 20 kilohms.

Digital control of amplifier gain is provided with gain control inputs on pins 12 through 15. These inputs must be driven synchronously (all inputs stable within 0.1 microsecond) to avoid false receive output signals due to invalid logic counts. This can be done easily with the TL851 control IC. A plot showing relative gain for the various gain steps versus time can be seen in Figure 2. To dampen ringing of the 50-kilohertz electrostatic transducer, a 5-kilohm resistor from pin 1 (GAIN) to pin 2 (XIN) is recommended.

An external parallel combination of inductance and capacitance between pin 4 (LC) and pin 5 (VCC) provides an amplifier with an externally controlled gain and Q. This not only allows control of gain to compensate for attenuation of signal with distance, but also maximizes noise and sidelobe rejection. Care must be taken to accurately tune the L-C combination at operating frequency or gain and Q will be greatly reduced at higher gain steps.

AC coupling between stages of the amplifier is accomplished with a 0.01-microfarad capacitor for proper biasing.

The receive output is normally held at a low level by an internal 1-microampere current source. When an input of sufficient amplitude is received, the output is driven alternately by the 1-microampere discharge current and a 50-microampere charging current. A 1000-picofarad capacitor is required from the receive output (pin 9) to ground (pin 16) to integrate the received signal so that one or two noise pulses will not be recognized.

Pin 2 (XIN) provides clamping for the transformer secondary when used for transducer transmit drive as shown in Figure 4 of the SN28827 data sheet.

The TL852 is characterized for operation from 0°C to 40°C

4

Special Functions

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**TEXAS
INSTRUMENTS**

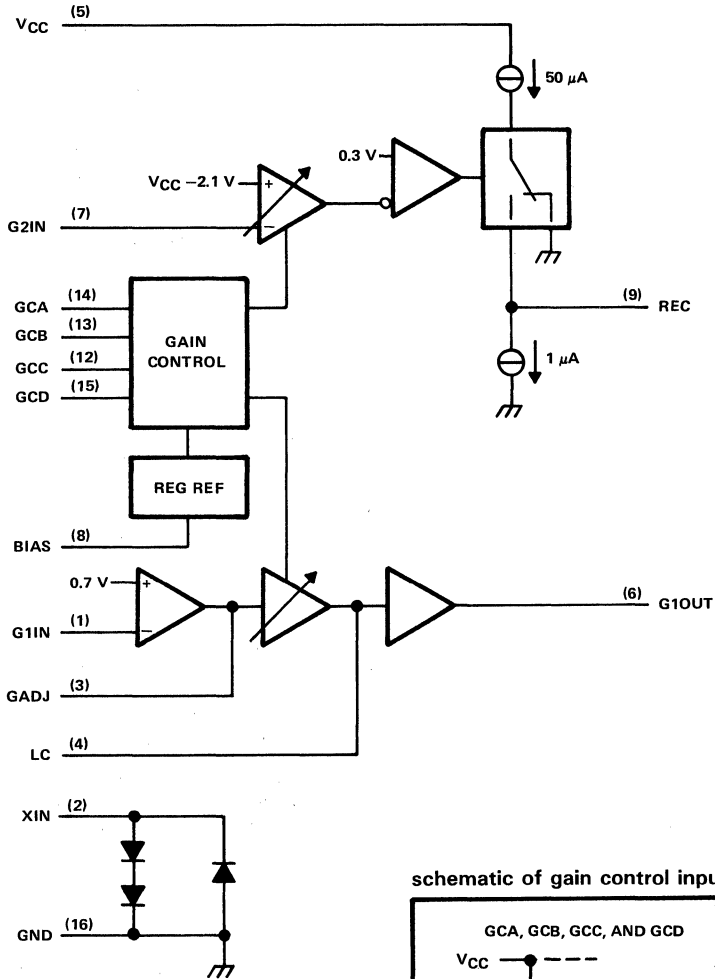
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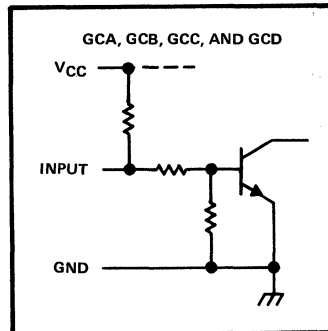
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TL852 SONAR RANGING RECEIVER

functional block diagram



schematic of gain control inputs



4

Special Functions

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Voltage at any pin with respect to GND	-0.5 V to 7 V
Voltage at any pin with respect to V _{CC}	-7 V to 0.5 V
XIN input current (50% duty cycle)	±60 mA
Continuous power dissipation at (or below) 25°C free-air temperature (see Note 1)	1150 mW
Operating free-air temperature range	-40°C to 85°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: For operation above 25°C, derate linearly at the rate of 9.2 mW/°C.

recommended operating conditions

		MIN	MAX	UNIT
Supply voltage, V _{CC}		4.5	6.8	V
High-level input voltage, V _{IH}	GCA, GCB, GCC, GCD	2.1		V
Low-level input voltage, V _{IL}			0.6	V
Bias resistor between pins 8 and 16		64	72	kΩ
Operating free-air temperature, T _A		0	40	°C

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER	TEST CONDITIONS	MIN	TYP [†]	MAX	UNIT
Input clamp voltage at XIN	I _I = 40 mA			2.5	V
	I _I = -40 mA			-1.5	
Open-circuit input voltage at GCA, GCB, GCC, GCD	V _{CC} = 5 V, I _I = 0		2.5		V
High-level input current, I _{IH} , into GCA, GCB, GCC, GCD	V _{CC} = 5 V, V _{IH} = 2 V		-0.5		mA
Low-level input current, I _{IL} , into GCA, GCB, GCC, GCD	V _{CC} = 5 V, V _{IL} = 0			-3	mA
Receive output current	I _{G2IN} = -100 μA, V _O = 0.3 V		1		μA
	I _{G2IN} = 100 μA, V _O = 0.1 V		-50		
Supply current, I _{CC}				45	mA

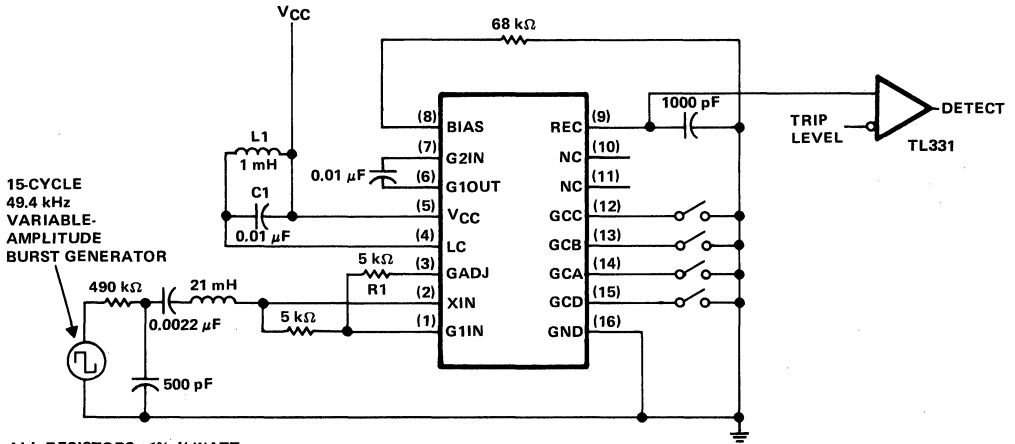
[†]Typical values are at V_{CC} = 5 V and T_A = 25°C.

TL852 SONAR RANGING RECEIVER

TYPICAL APPLICATION INFORMATION

detect level vs gain step

Detect level is measured by applying a 15-cycle burst of 49.4 kilohertz square wave just after the beginning of the gain step to be tested. The least burst amplitude that makes the REC pin reach the trip level is defined to be the detect level. System gain is then inversely proportional to detect level. See the test circuit in Figure 1.



ALL RESISTORS $\pm 1\%$, $\frac{1}{4}$ WATT
 ALL CAPACITORS $\pm 1\%$, FILM
 L1 Q > 60 at 50 kHz
 C1 Q > 500 at 50 kHz

4

Special Functions

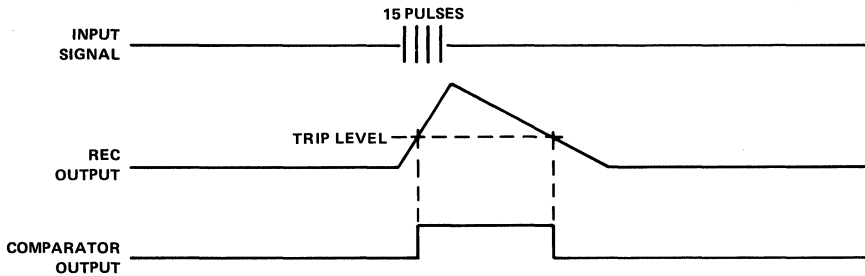


FIGURE 1. DETECT-LEVEL MEASUREMENT CIRCUIT AND WAVEFORMS

TYPICAL APPLICATION INFORMATION

GAIN STEP TABLE

GCD	GCC	GCB	GCA	STEP NUMBER
L	L	L	L	0
L	L	L	H	1
L	L	H	L	2
L	L	H	H	3
L	H	L	L	4
L	H	L	H	5
L	H	H	L	6
L	H	H	H	7
H	L	L	L	8
H	L	L	H	9
H	L	H	L	10
H	L	H	H	11

RECEIVER GAIN
vs
GAIN STEP NUMBERS

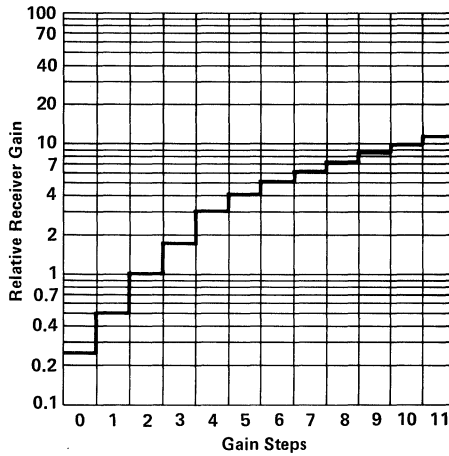


FIGURE 2

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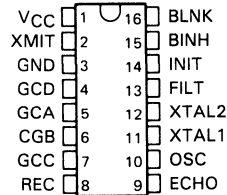
Special Functions

TL853 SONAR RANGING CONTROL

D2843, DECEMBER 1984—REVISED MARCH 1988

- Designed for Use with the TL852 in Sonar Ranging Modules Like the SN28828
- Operates with Single Supply
- Accurate Clock Output for External Use
- Synchronous 4-Bit Gain Control Output
- Internal 1.2-V Level Detector for Receive
- TTL-Compatible
- Interface to 40-kHz Piezoelectric or Electrostatic Transducers

N DUAL-IN-LINE PACKAGE (TOP VIEW)



description

The TL853 is an economical digital I²L ranging control integrated circuit designed for use with the Texas Instruments TL852 Sonar ranging receiver integrated circuit.

The TL853 is designed for distance measurement ranging from six inches to 35 feet. The device has an internal oscillator that uses a low-cost external ceramic resonator. With a simple interface and a 420-kHz ceramic resonator, the device will drive a 40-kHz piezoelectric transducer.

The device cycle begins when Initiate (INIT) is taken to the high logic level. There must be at least 5 ms from initial power up (V_{CC}) to the first initiate signal in order for all the device internal latches to reset and for the ceramic-resonator-controlled oscillator to stabilize. The device will transmit a burst of 16 pulses each time INIT is taken high.

The oscillator output (OSC) is enabled by INIT. The oscillator frequency is the ceramic resonator frequency divided by 10.5 for the first 16 cycles (during transmit) and then the oscillator frequency changes to the ceramic resonator frequency divided by 4.5 for the remainder of the device cycle.

When used with an external 420-kilohertz ceramic resonator, the device internal blanking disables the receive input (REC) for 2.46 ms after initiate to exclude false receive inputs that may be caused by transducer ringing. The internal blanking feature also eliminates echos from objects closer than 1.37 feet from the transducer. If it is necessary to detect objects closer than 1.37 feet, then the internal blanking may be shortened by taking the blanking inhibit (BINH) high, enabling the receive input. The blanking input (BLNK) may be used to disable the receive input and reset ECHO to a low logic level at any time during the device cycle for selective echo exclusion or for a multiple-echo mode of operation.

The device provides a synchronous 4-bit gain control output (12 steps) designed to control the gain of the TL852 sonar ranging receiver integrated circuit. The digital gain control waveforms are shown in Figure 2 with the nominal transition times from INIT listed in the Gain Control Output Table.

The threshold of the internal receive level detector is 1.2 volts. The TL853 operates over a supply voltage range of 4.5 volts to 6.8 volts and is characterized for operation from 0°C to 40°C.

4
Special Functions

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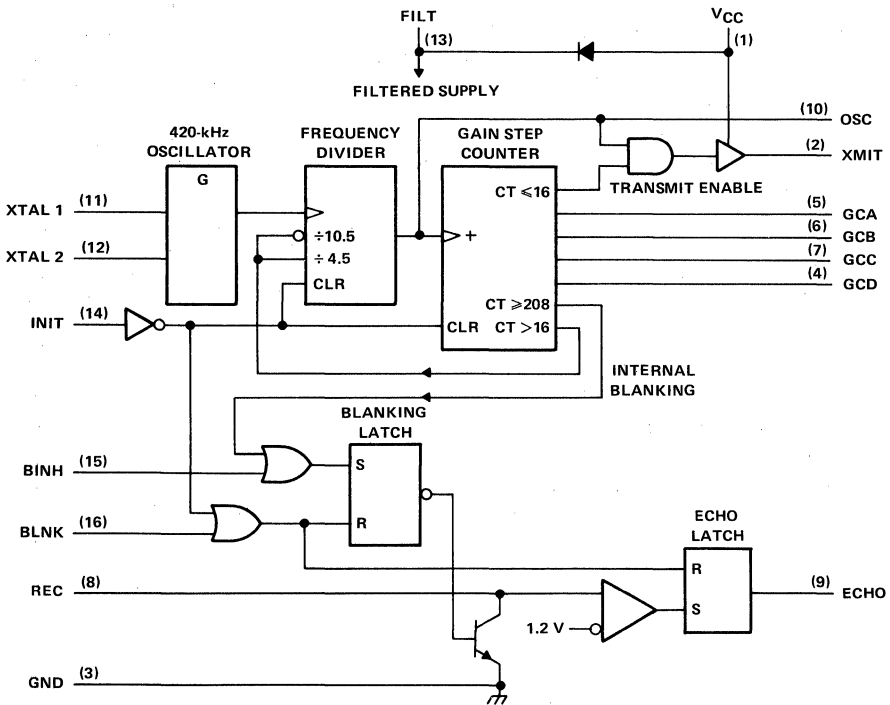
TL853 SONAR RANGING CONTROL

GAIN CONTROL OUTPUT TABLE

STEP NUMBER	CGD	GCC	GCB	GCA	TIME (ms) FROM INITIATE†
0	L	L	L	L	2.46 ms
1	L	L	L	H	5.2 ms
2	L	L	H	L	7.94 ms
3	L	L	H	H	10.69 ms
4	L	H	L	L	13.43 ms
5	L	H	L	H	16.17 ms
6	L	H	H	L	18.91 ms
7	L	H	H	H	21.66 ms
8	H	L	L	L	27.14 ms
9	H	L	L	H	32.63 ms
10	H	L	H	L	38.11 ms
11	H	L	H	H	INIT ↓

†This is the time to the end of the indicated step and assumes a nominal 420-kHz ceramic resonator.

functional block diagram



4

Special Functions

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Voltage at any pin with respect to GND	-0.5 V to 7 V
Voltage at any pin with respect to V _{CC}	-7 V to 0.5 V
Continuous power dissipation at (or below) 25°C free-air temperature (see Note 1)	1150 mW
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: For operation above 25°C, derate linearly at the rate of 9.2 mW/°C.

recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, V _{CC}	4.5	6.8	V
High-level input voltage, V _{IH}	BLNK, BINH, INIT	2.1	V
Low-level input voltage, V _{IL}	BLNK, BINH, INIT	0.6	V
Delay time, power up to INIT high	5		ms
Operating free-air temperature, T _A	0	40	°C

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature

PARAMETER		TEST CONDITIONS	MIN	TYP [†]	MAX	UNIT
Input current	BLNK, BINH, INIT	V _I = 2.1 V			1	mA
High-level output current, I _{OH}	ECHO, OSC, GCA, GCB, GCC, GCD	V _{OH} = 5.5 mA			100	μA
Low-level output voltage, V _{OL}	ECHO, OSC, GCA, GCB, GCC, GCD	I _{OL} = 1.6 mA			0.4	V
On-state output current	XMIT output	V _O = 1 V	-140			mA
Internal blanking interval	REC input		2.46 [‡]			ms
Frequency during 16-pulse transmit period	OSC output			40 [‡]		kHz
	XMIT output			40 [‡]		
Frequency after 16-pulse transmit period	OSC output			93.3 [‡]		kHz
	XMIT output			0		
Supply current, I _{CC}	During transmit period				260	mA
	After transmit period				55	

[†]Typical values are at V_{CC} = 5 V and T_A = 25°C.

[‡]These typical values apply for a 420-kHz ceramic resonator.

TL853 SONAR RANGING CONTROL

schematics of inputs and outputs

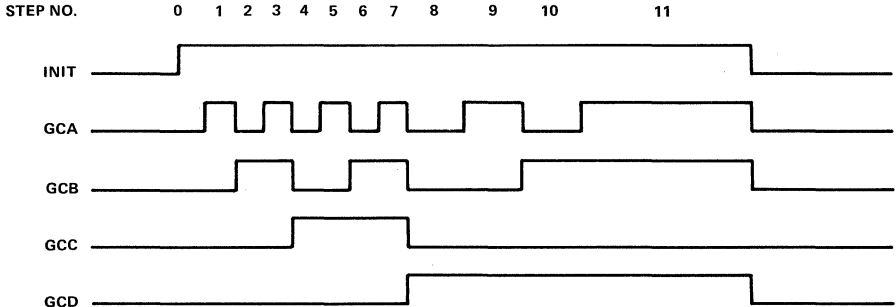
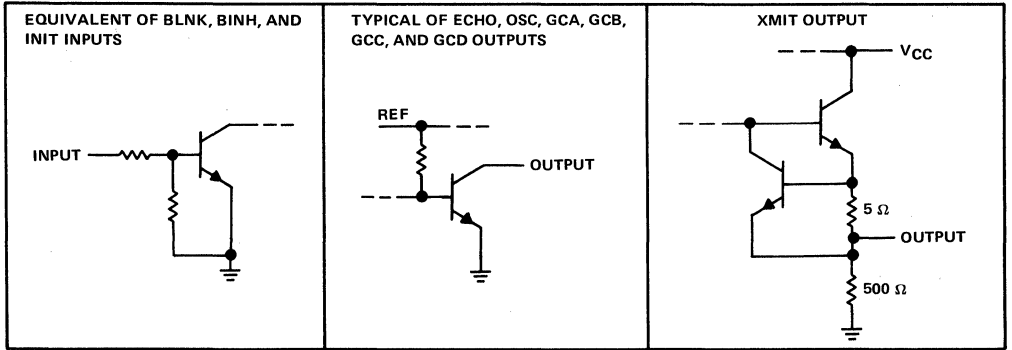


FIGURE 1. DIGITAL GAIN CONTROL WAVEFORMS

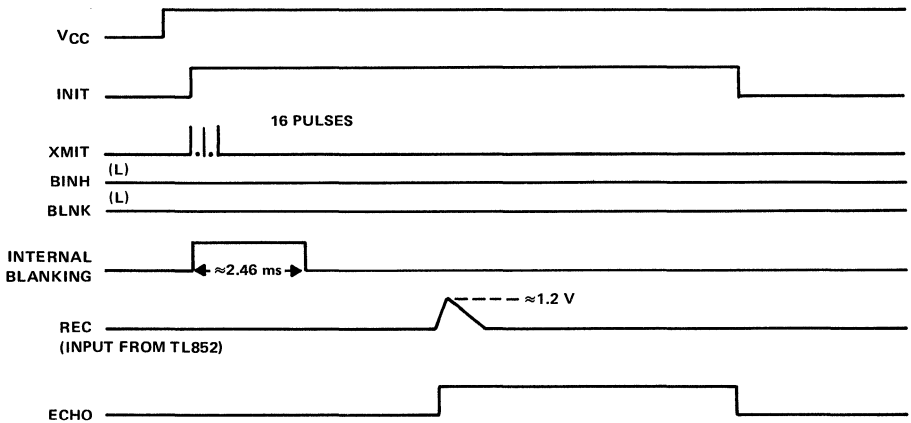


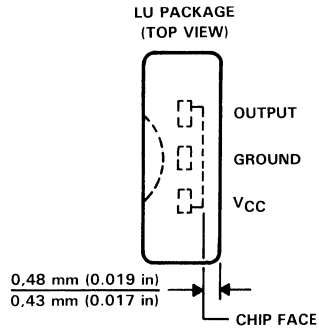
FIGURE 2. EXAMPLE OF SINGLE-ECHO-MODE CYCLE WHEN USED WITH THE TL852 RECEIVER AND 420-kHz CERAMIC RESONATOR

4 Special Functions

TL3013C SILICON HALL-EFFECT SWITCH

D2903, SEPTEMBER 1985—REVISED APRIL 1988

- Magnetic-Field Sensing Hall-Effect Input
- On-Off Hysteresis
- Small Size
- Standard Bipolar Technology Minimizes ESD Susceptibility
- $I_{OL} \dots 20 \text{ mA}$ Min at $V_{OL} = 0.4 \text{ V}$
- $I_{OH} \dots 1 \mu\text{A}$ Max at $V_{OH} = 24 \text{ V}$
- Buried Hall-Effect Cell Reduces Threshold Drift Caused by Temperature Variation and Aging
- Direct Replacement for the Sprague UGN3013



FUNCTION TABLE ($0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$)

FLUX DENSITY	OUTPUT
$B \leq 2.5 \text{ mT}$ (25 G)	Off
2.5 mT (25 G) $< B < 45 \text{ mT}$ (450 G)	Undefined
$B \geq 45 \text{ mT}$ (450 G)	On

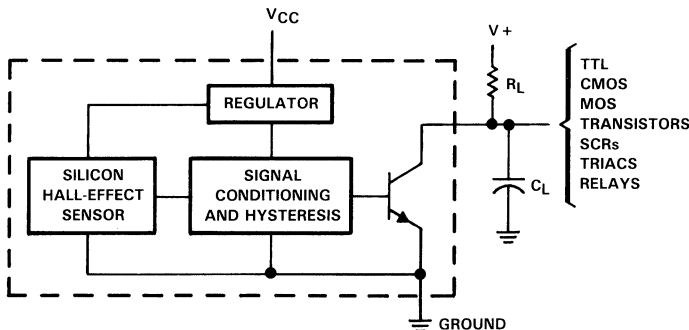
The unit of magnetic density in the International System of Units (SI) is the tesla (T). The tesla is equal to one weber per square meter. Values expressed in milliteslas may be converted to gauss (G) by multiplying by ten.

description

The TL3013C is a low-cost magnetically operated electronic switch that utilizes the Hall effect to sense magnetic fields. Each circuit consists of a Hall-effect sensor, signal conditioning and hysteresis functions, and an output transistor integrated into a monolithic chip. The outputs of these circuits can be directly connected to many different types of electronic components.

The TL3013C is characterized for operation over the temperature range of 0°C to 70°C .

functional block diagram



4

Special Functions

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TL3013C SILICON HALL-EFFECT SWITCH

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	40 V
Output voltage	40 V
Output current	30 mA
Magnetic flux density	unlimited
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1. Voltage values are with respect to the network ground terminal.

electrical characteristics over operating free-air temperature range, $V_{CC} = 4.5 V$ to $24 V$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
B_{OP}	Operate-point magnetic flux density (see Figure 2)		30	45	mT†
B_{RP}	Release-point magnetic flux density (see Figure 2)	2.5	22.5		mT†
B_{hys}	Hysteresis ($B_{OP} - B_{RP}$)	3	7.5		mT†
α_B	Temperature coefficient of B_{OP} and B_{RP}		± 0.25		%/°C
V_{OL}	Low-level output voltage			0.4	V
I_{OH}	High-level output current			1	μA
I_{CC}	Supply current		3	7	mA

†The unit of magnetic density in the International System of Units (SI) is the tesla (T). The tesla is equal to one weber per square meter. Values expressed in milliteslas may be converted to gauss (G) by multiplying by ten.

4 Special Functions

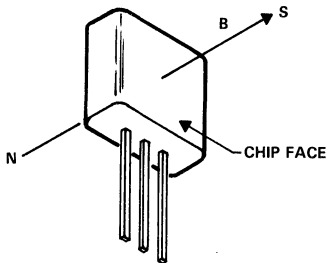


FIGURE 1. DEVICE ORIENTATION IN A MAGNETIC FIELD

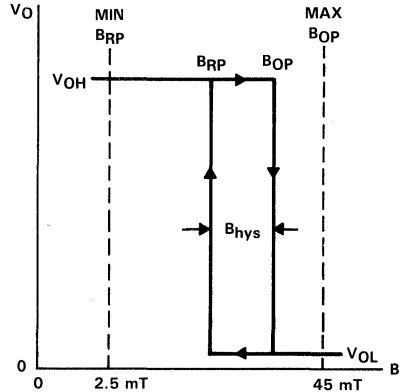


FIGURE 2. REPRESENTATIVE CURVE OF V_O vs B

switching characteristics at $V_{CC} = 12 V$, $T_A = 25^\circ C$ (unless otherwise noted)

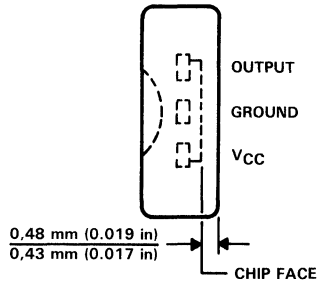
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_r	$R_L = 820 \Omega$, $C_L = 20 pF$		350		ns
t_f			85		

TL3019C SILICON HALL-EFFECT SWITCH

D2903, JULY 1985—REVISED APRIL 1988

- Magnetic-Field Sensing Hall-Effect Input
- On-Off Hysteresis
- Small Size
- Standard Bipolar Technology Minimizes ESD Susceptibility
- $I_{OL} \dots 20 \text{ mA Min at } V_{OL} = 0.4 \text{ V}$
- $I_{OH} \dots 1 \mu\text{A Max at } V_{OH} = 24 \text{ V}$
- Buried Hall-Effect Cell Reduces Threshold Drift Caused by Temperature Variation and Aging
- Direct Replacement for the Sprague UGN3019

LU PACKAGE
(TOP VIEW)



description

The TL3019C is a low-cost magnetically operated electronic switch that utilizes the Hall Effect to sense magnetic fields. Each circuit consists of a Hall-Effect sensor, signal conditioning and hysteresis functions, and an output transistor integrated into a monolithic chip. The outputs of these circuits can be directly connected to many different types of electronic components.

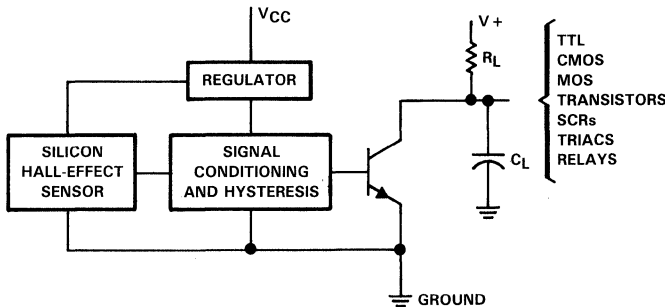
The TL3019C is characterized for operation over the temperature range of 0°C to 70°C .

FUNCTION TABLE ($0^{\circ}\text{C} \leq T_A \leq 70^{\circ}\text{C}$)

FLUX DENSITY	OUTPUT
$B \leq 12.5 \text{ mT (125 G)}$	Off
$12.5 \text{ mT (125 G)} < B < 50 \text{ mT (500 G)}$	Undefined
$B \geq 50 \text{ mT (500 G)}$	On

The unit of magnetic density in the International System of Units (SI) is the tesla (T). The tesla is equal to one weber per square meter. Values expressed in milliteslas may be converted to gauss (G) by multiplying by ten.

functional block diagram



4

Special Functions

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TL3019C SILICON HALL-EFFECT SWITCH

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	40 V
Output voltage	40 V
Output current	30 mA
Magnetic flux density	unlimited
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1. Voltage values are with respect to the network ground terminal.

electrical characteristics over operating free-air temperature range, $V_{CC} = 4.5 \text{ V to } 24 \text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
B_{OP}	Operate-point magnetic flux density (see Figure 2)			42	50	mT^\dagger
B_{RP}	Release-point magnetic flux density (see Figure 2)		12.5	30		mT^\dagger
B_{hys}	Hysteresis ($B_{OP} - B_{RP}$)		5	12		mT^\dagger
α_B	Temperature coefficient of B_{OP} and B_{RP}			± 0.25		%/°C
V_{OL}	Low-level output voltage	$I_{OL} = 20 \text{ mA}$			0.4	V
I_{OH}	High-level output current	$V_{OH} = 24 \text{ V}$			1	μA
I_{CC}	Supply current			3	7	mA

[†]The unit of magnetic density in the International System of Units (SI) is the tesla (T). The tesla is equal to one weber per square meter. Values expressed in milliteslas may be converted to gauss (G) by multiplying by ten.

4

Special Functions

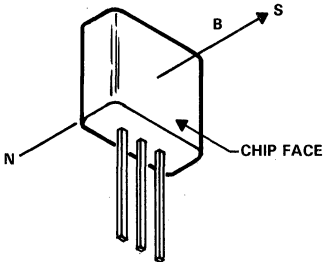


FIGURE 1. DEVICE ORIENTATION IN A MAGNETIC FIELD

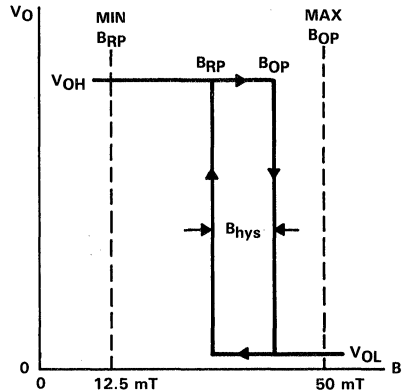


FIGURE 2. REPRESENTATIVE CURVE OF V_O vs B

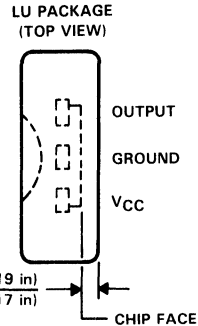
switching characteristics at $V_{CC} = 12 \text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_r	Output rise time		350		ns
t_f	Output fall time		85		
	$R_L = 820 \Omega$, $C_L = 20 \text{ pF}$				

TL3020C SILICON HALL-EFFECT SWITCH

D2903, OCTOBER 1985—REVISED APRIL 1988

- Magnetic-Field Sensing Hall-Effect Input
- On-Off Hysteresis
- Small Size
- Standard Bipolar Technology Minimizes ESD Susceptibility
- $I_{OL} \dots 20 \text{ mA Min at } V_{OL} = 0.4 \text{ V}$
- $I_{OH} \dots 1 \mu\text{A Max at } V_{OH} = 24 \text{ V}$
- Buried Hall-Effect Cell Reduces Threshold Drift Caused by Temperature Variation and Aging
- Direct Replacement for the Sprague UGN3020



FUNCTION TABLE ($0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$)

FLUX DENSITY	OUTPUT
$B \leq 5 \text{ mT (50 G)}$	Off
$5 \text{ mT (50 G)} < B < 35 \text{ mT (350 G)}$	Undefined
$B \geq 35 \text{ mT (350 G)}$	On

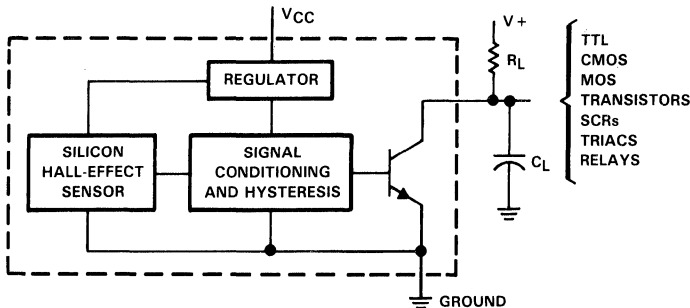
The unit of magnetic density in the International System of Units (SI) is the tesla (T). The tesla is equal to one weber per square meter. Values expressed in milliteslas may be converted to gauss (G) by multiplying by ten.

description

The TL3020C is a low-cost magnetically operated electronic switch that utilizes the Hall effect to sense magnetic fields. Each circuit consists of a Hall-effect sensor, signal conditioning and hysteresis functions, and an output transistor integrated into a monolithic chip. The outputs of these circuits can be directly connected to many different types of electronic components.

The TL3020C is characterized for operation over the temperature range of 0°C to 70°C .

functional block diagram



4

Special Functions

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TL3020C SILICON HALL-EFFECT SWITCH

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	40 V
Output voltage	40 V
Output current	30 mA
Magnetic flux density	unlimited
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1. Voltage values are with respect to the network ground terminal.

electrical characteristics over operating free-air temperature range, $V_{CC} = 4.5 V$ to $24 V$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
B_{OP}	Operate-point magnetic flux density (see Figure 2)			22	35	mT†
B_{RP}	Release-point magnetic flux density (see Figure 2)		5	16.5		mT†
B_{hys}	Hysteresis ($B_{OP} - B_{RP}$)		2	5.5		mT†
α_B	Temperature coefficient of B_{OP} and B_{RP}		± 0.25			%/°C
V_{OL}	Low-level output voltage	$I_{OL} = 20$ mA			0.4	V
I_{OH}	High-level output current	$V_{OH} = 24$ V			1	μ A
I_{CC}	Supply current		3	7		mA

†The unit of magnetic density in the International System of Units (SI) is the tesla (T). The tesla is equal to one weber per square meter. Values expressed in milliteslas may be converted to gauss (G) by multiplying by ten.

4 Special Functions

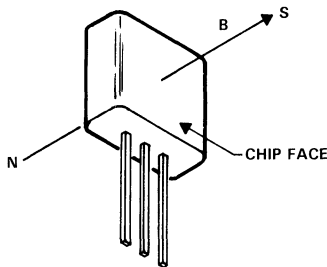


FIGURE 1. DEVICE ORIENTATION IN A MAGNETIC FIELD

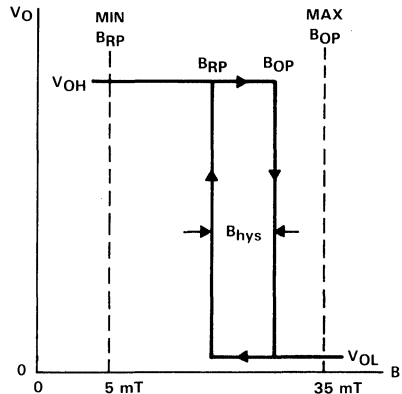


FIGURE 2. REPRESENTATIVE CURVE OF V_O vs B

switching characteristics at $V_{CC} = 12 V$, $T_A = 25^\circ C$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_r	$R_L = 820 \Omega$, $C_L = 20$ pF	350			ns
t_f		85			

TL3101I, TL3101C SILICON HALL-EFFECT SWITCH

APRIL 1985—REVISED APRIL 1988

- Magnetic-Field Sensing Hall-Effect Input
- On-Off Hysteresis
- Small Size
- Solid-State Technology
- Open-Collector Output
- Buried Hall-Effect Cell Reduces Threshold Drift Caused By Temperature Variation and Aging

description

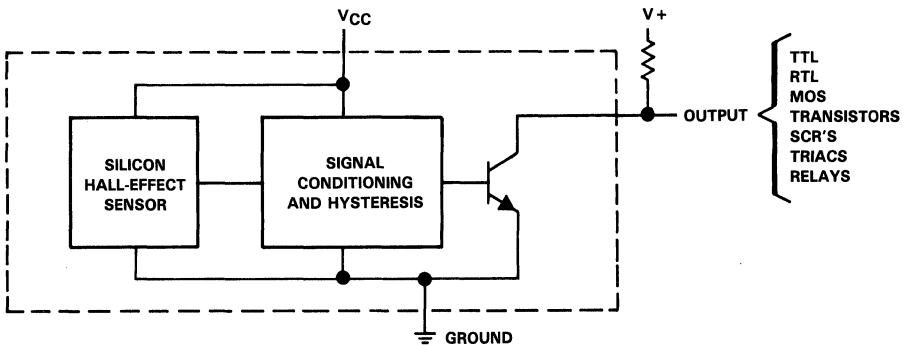
The TL3101I and TL3101C are low-cost magnetically operated electronic switches that utilize the Hall Effect to sense steady-state magnetic fields. Each circuit consists of a Hall-Effect sensor, signal conditioning and hysteresis functions, and an output transistor integrated into a monolithic chip. The outputs of these circuits can be directly connected to many different types of electronic components.

The TL3101C is characterized for operation over the temperature range of 0°C to 70°C. The TL3101I is characterized for operation over the range of -20°C to 85°C.

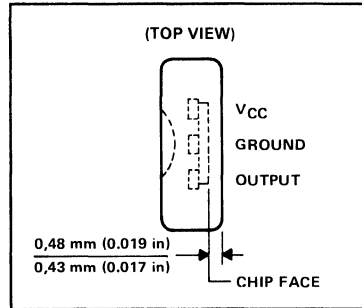
FUNCTION TABLE 0°C ≥ T_A ≤ 70°C

FLUX DENSITY	OUTPUT
≤ -25 mT	Off
-25 mT < B < 25 mT	Undefined
≥ 25 mT	On

functional block diagram



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TL3101I, TL3101C SILICON HALL-EFFECT SWITCH

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	7 V
Output voltage	30 V
Output current	20 mA
Operating free-air temperature range: TL3101C	0°C to 70°C
TL3101I	-20°C to 85°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C
Magnetic flux density	unlimited

NOTE 1: Voltage values are with respect to network ground terminal.

electrical characteristics at specified free-air temperature, $V_{CC} = 5 V \pm 5%$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
B_{T+}	Threshold of positive going magnetic flux density [†]		0°C to 70°C	0	25		mT [‡]
			-20°C to 85°C	0	35		
B_{T-}	Threshold of negative going magnetic flux density [†]		0°C to 70°C	-25 [‡]	0		mT [‡]
			-20°C to 85°C	-35 [‡]	0		
$B_{T+} - B_{T-}$	Hysteresis		0°C to 70°C	5	20		mT [‡]
I_{OH}	High-level output current	$V_{OH} = 20 V$	0°C to 70°C			100	μA
V_{OL}	Low-level output voltage	$V_{CC} = 4.75, I_{OL} = 16 mA$	0°C to 70°C			0.4	V
I_{CC}	Supply current	$V_{CC} = 5.25 V$	Output low			6	mA
			Output high	0°C to 70°C		4	

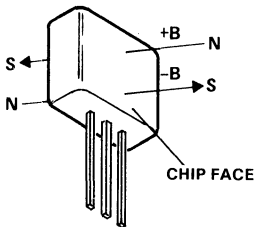
[†] Threshold values are those levels of magnetic flux density at which the output changes state. For the TL3101, a level more positive than B_{T+} causes the output to a low level and a level more negative than B_{T-} causes the output to go to a high level. See Figures 1 and 2.

[‡] The algebraic convention, where the most negative limit is designated as minimum, is used in this data sheet for flux density threshold levels only.

[§] The unit of magnetic flux density in the International System of Units (SI) is the tesla (T). The tesla is equal to one weber per square meter. Values expressed in milliteslas may be converted to gauss by multiplying by ten.

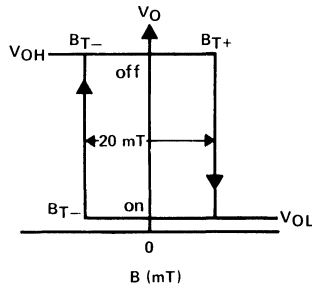
4

Special Functions



The north pole of a magnet is the pole that is attracted by the geographical north pole. The north pole of a magnet repels the north-seeking pole of a compass. By accepted magnetic convention, lines of flux emanate from the north pole of a magnet and enter the south pole.

FIGURE 1. DEFINITION OF MAGNETIC FLUX POLARITY



The positive-going threshold (B_{T+}) is a positive B level at which a positive-going flux density results in the TL3101 output turning on. The negative-going threshold is negative B level at which a negative-going flux density results in the TL3101 turning off.

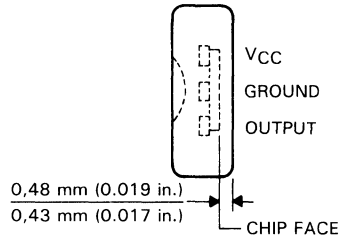
FIGURE 2. REPRESENTATIVE CURVES OF V_O vs B

TL3103I, TL3103C LINEAR HALL-EFFECT SENSORS

D3184, MAY 1985—REVISED FEBRUARY 1989

- Output Voltage Linear with Applied Magnetic Field
- Sensitivity Stable Over Wide Operating Temperature Range
- Buried Hall Cell Reduces Changes Due to Temperature Variation and Aging
- Solid-State Technology
- Three-Terminal Device
- Senses Static or Dynamic Magnetic Fields

LU PACKAGE
(TOP VIEW)

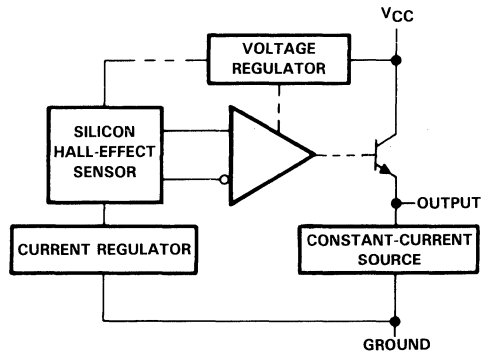


description

The TL3103I and TL3103C are low-cost magnetic-field sensors designed to provide a linear output voltage proportional to the magnetic field they sense. These monolithic circuits incorporate a Hall element as the primary sensor along with a voltage reference and a precision amplifier. Temperature stabilization and internal trimming circuitry yield a device that features high overall sensitivity accuracy with less than 5% error over its operating temperature range.

The TL3103I is characterized for operation from -20°C to 85°C . The TL3103C is characterized for operation from 0°C to 70°C .

functional block diagram



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	25 V
Continuous total dissipation at (or below) 25°C free-air temperature (see Note 2)	775 mW
Operating free-air temperature range: TL3103I	-20°C to 85°C
TL3103C	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C
Magnetic flux density	unlimited

- NOTES: 1. Voltage values are with respect to network ground terminal.
2. For operation above 25°C free-air temperature, derate linearly at the rate of $6.2\text{ mW}/^{\circ}\text{C}$.

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TL3103I, TL3103C LINEAR HALL-EFFECT SENSORS

recommended operating conditions

		TL3103I			TL3103C			UNIT		
		MIN	NOM	MAX	MIN	NOM	MAX			
Supply voltage, V_{CC}		9	12	15	9	12	15	V		
Magnetic flux density, B		±50			±50			mT		
Output current, I_O	Sink	0.5			0.5			mA		
	Source	-2			-2					
Operating free-air temperature, T_A		-20			85			0	70	°C

electrical characteristics over recommended ranges of supply voltage and magnetic flux density, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	MIN	TYP‡	MAX	UNIT		
V_O Output voltage	$I_O = -2\text{ mA to } 0.5\text{ mA}$, $B = 0\text{ mT}^{\S}$	5.8	6	6.2	V		
k_{SVS} Supply voltage sensitivity ($\Delta V_O/\Delta V_{CC}$)		18	mV/V				
S Magnetic sensitivity ($\Delta V_O/\Delta B$)	$B = -50\text{ to } 50\text{ mT}^{\S}$	14	16	18	V/T [§]		
ΔS Magnetic sensitivity change with temperature	$\Delta T_A = 25^\circ\text{C to MIN or MAX}$				±5	%	
I_{CC} Supply current	$B = 0\text{ mT}^{\S}$, $I_O = 0$				8	12	mA
f_{max} Maximum operating frequency					100	kHz	

†For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡Typical values are at $V_{CC} = 12\text{ V}$ at $T_A = 25^\circ\text{C}$.

§The unit of magnetic flux density in the International System of Units (SI) is the tesla (T). The tesla is equal to one weber per square meter. Values expressed in milliteslas may be converted to gauss by multiplying by ten, e.g., 50 millitesla = 500 gauss.

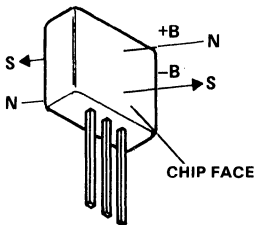


FIGURE 1. DEFINITION OF MAGNETIC FLUX POLARITY

The north pole of a magnet is the pole that is attracted by the geographical north pole. The north pole of a magnet repels the north-seeking pole of a compass. By accepted magnetic convention, lines of flux emanate from the north pole of a magnet and enter the south pole.

OUTPUT VOLTAGE vs MAGNETIC FLUX DENSITY

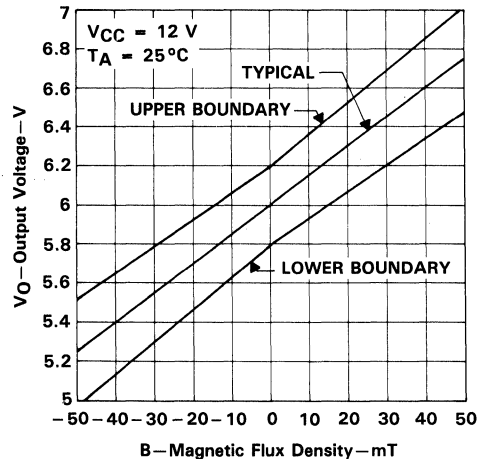


FIGURE 2

TYPICAL APPLICATION DATA

The circuit in Figure 3 may be used to set the output voltage at zero field strength to exactly 6 V (using R1) and to set the sensitivity to exactly -15 V/T (using R2) as depicted in Figure 4.

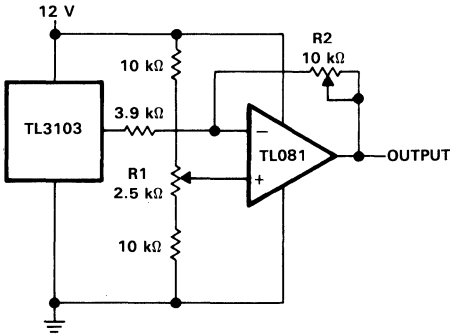


FIGURE 3. COMPENSATION CIRCUIT

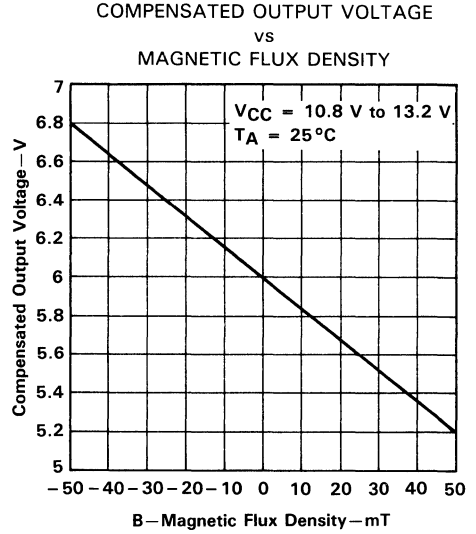


FIGURE 4

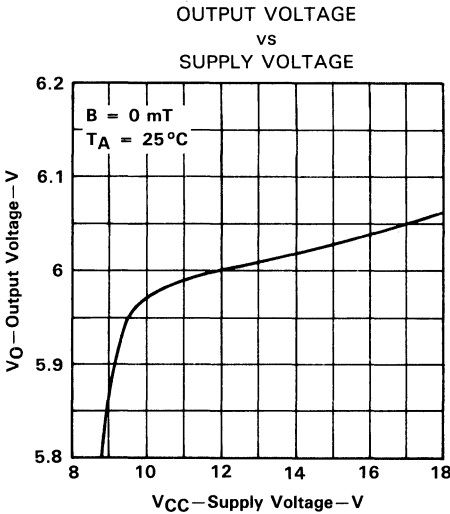


FIGURE 5

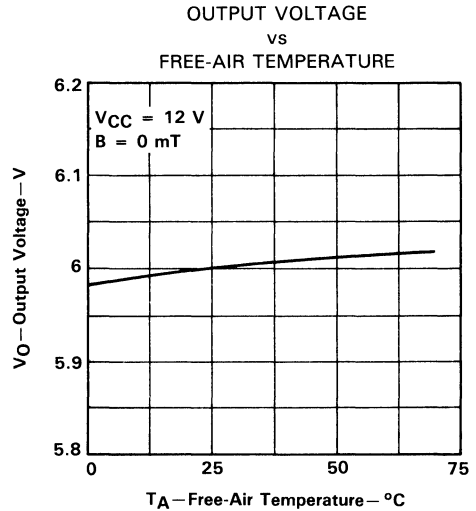


FIGURE 6

TL3103I, TL3103C
LINEAR HALL-EFFECT SENSORS

TYPICAL APPLICATION DATA

MAGNETIC SENSITIVITY
 vs
 SUPPLY VOLTAGE

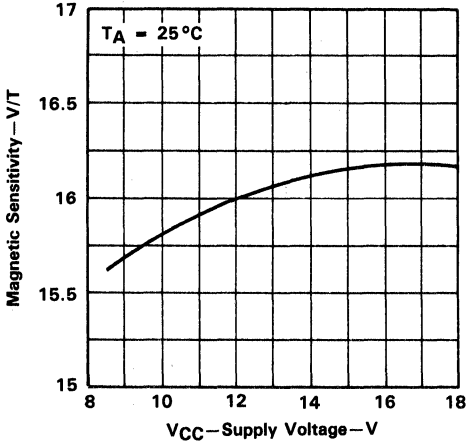


FIGURE 7

MAGNETIC SENSITIVITY
 vs
 FREE-AIR TEMPERATURE

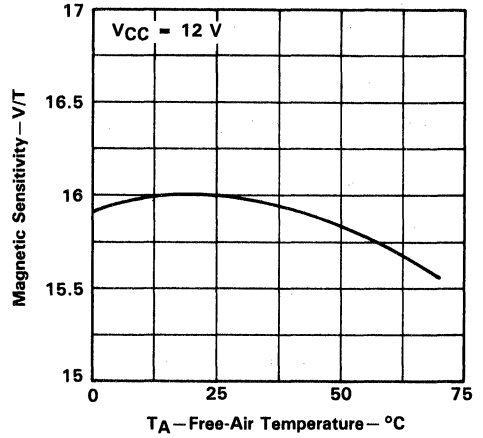


FIGURE 8

linear hall-effect sensor in isolated feedback applications

purpose

The purpose of the circuit in Figure 9 is to demonstrate the capability of the linear hall-effect sensor to provide isolated sensing.

4

Special Functions

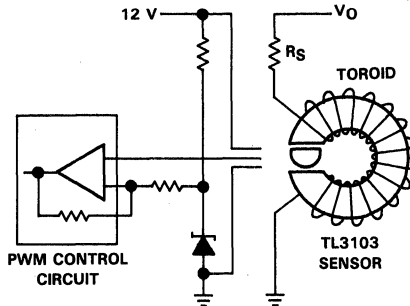


FIGURE 9. ISOLATE FEEDBACK

linear sensor

The TL3103 senses the presence of a magnetic field. In the absence of a magnetic field, the TL3103 output voltage is 6 V. As the sensor senses the presence of a magnetic field, its output varies proportionally at 16 V/T.

toroid

The permeability (μ) of any given material is the relationship of the magnetic flux density (B) to the magnetic field intensity (H). The magnetic field intensity of a toroid is given by the expression:

$$H = \frac{NI}{L} \quad (1)$$

Where: NI is number of turns times the current.
L is mean length of the toroid.

The expression for the magnetic flux density then becomes:

$$B = \mu \frac{NI}{L} \quad (2)$$

With an air gap, the basic expression is altered to the following:

$$B_{\text{gap}} = \mu_0 \frac{NI}{L + Kg} K \quad (3)$$

Where: μ_0 is the permeability of air = 12.57×10^{-7} .
K is the relative permeability of the toroid (μ/μ_0).
g is the length of the air gap in mils.

If the relative permeability of the toroid is large ($Kg > K$), the mean length of the toroid becomes insignificant and the expression for flux density reduces to:

$$B_{\text{gap}} = \mu_0 \frac{NI}{g} \quad (4)$$

circuit

As previously discussed, the output of the TL3103 is:

$$V_{\text{sense}} = 6 \text{ V} + (16 \text{ V/T}) \left(\mu_0 \frac{NI}{g} \right) \quad (5)$$

This shows how the output of the TL3103 varies with the ampere turns of the toroid. For some relative numbers, Figure 9 uses an Arnold toroid #A393163-2 with a 156-mil air gap. The magnetic field created in the air gap is:

$$B \text{ (gauss)} = 1.92 \text{ NI (ampere turns)} \quad (6)$$

Therefore, the variation in the output of the TL3103 is:

$$\Delta V_{\text{sense}} \text{ (mV)} = (1.6) (1.92 \text{ NI}) \quad (7)$$

The sensitivity of the TL3103 to the current in the windings can therefore be altered by the number of turns composing the windings.

ΔV_{sense}	N	ΔI
614 mV	20	10 mA
614 mV	200	1 mA
614 mV	2000	100 μ A

TL3103I, TL3103C LINEAR HALL-EFFECT SENSORS

power supply application

Since the output of the TL3103 varies proportionally to the current flow in the toroid, this leads to the obvious application of current sensing. The features presented by this approach are as follows:

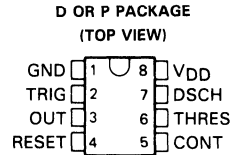
1. Minimum power loss in the sensing element:

$$P_{\text{loss}} = I_2^2 R_{\text{toroid}} \quad (R < 0.01 \Omega \text{ for } 20 \text{ turns}) \quad (8)$$

2. Isolated feedback, no passive connection required.

Another application of this concept provides a linear isolated feedback of the output voltage. This is accomplished by connecting a resistor in series with the toroid terminated to ground. In this configuration, the current in the toroid is determined by the output voltage ($I_S = V_O/R_S$), therefore, the output variation of the TL3103 is proportional to the output voltage.

- Very Low Power Consumption . . . 1 mW
Typ at $V_{DD} = 5\text{ V}$
- Capable of Operation in Astable Mode
- CMOS Output Capable of Swinging Rail to Rail
- High Output-Current Capability
. . . Sink 100 mA Typ
. . . Source 10 mA Typ
- Output Fully Compatible with CMOS, TTL, and MOS
- Low Supply Current Reduces Spikes During Output Transitions
- High-Impedance Inputs . . . $10^{12}\ \Omega$ Typ
- Single-Supply Operation from 1 V to 18 V
- Functionally Interchangeable with the NE555; Has Same Pinout



description

The TLC551 is a monolithic timing circuit fabricated using TI's LinCMOS™ process, which provides full compatibility with CMOS, TTL, and MOS logic and operation at frequencies up to 2 MHz. Accurate time delays and oscillations are possible with smaller, less-expensive timing capacitors than the NE555 because of the high input impedance. Power consumption is low across the full range of power supply voltages.

Like the NE555, the TLC551 has a trigger level approximately one-third of the supply voltage and a threshold level approximately two-thirds of the supply voltage. These levels can be altered by use of the control voltage terminal. When the trigger input falls below the trigger level, the flip-flop is set and the output goes high. If the trigger input is above the trigger level and the threshold input is above the threshold level, the flip-flop is reset and the output is low. The reset input can override all other inputs and can be used to initiate a new timing cycle. If the reset input is low, the flip-flop is reset and the output is low. Whenever the output is low, a low-impedance path is provided between the discharge terminal and ground.

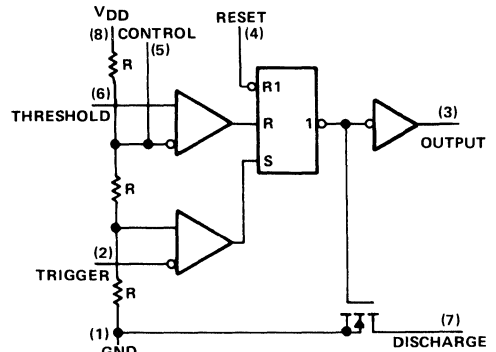
While the CMOS output is capable of sinking over 100 mA and sourcing over 10 mA, the TLC551 exhibits greatly reduced supply-current spikes during output transitions. This minimizes the need for the large decoupling capacitors required by the NE555.

These devices have internal electrostatic discharge (ESD) protection circuits that will prevent catastrophic failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. However, care should be exercised in handling these devices as exposure to ESD may result in a degradation of the device parametric performance.

All unused inputs should be tied to an appropriate logic level to prevent false triggering.

The TLC551C is characterized for operation from 0°C to 70°C.

functional block diagram



Reset can override Trigger, which can override Threshold.

TLC551C
LinCMOS™ TIMER

AVAILABLE OPTIONS

T _A	V _T MAX AT 25°C	PACKAGE	
		SMALL-OUTLINE (D)	PLASTIC DIP (P)
0°C to 70°C	3.8 V	TLC551CD	TLC551CP

D package is available taped-and-reeled. Add "R" suffix to device type when ordering (i.e. TLC551CDR).

FUNCTION TABLE

RESET VOLTAGE†	TRIGGER VOLTAGE†	THRESHOLD VOLTAGE†	OUTPUT	DISCHARGE SWITCH
< MIN	Irrelevant	Irrelevant	Low	On
> MAX	< MIN	Irrelevant	High	Off
> MAX	> MAX	> MAX	Low	On
> MAX	> MAX	< MIN	As previously established	

† For conditions shown as MIN or MAX, use the appropriate value specified under electrical characteristics.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V _{DD} (see Note 1)	18 V
Input voltage range (any input)	-0.3 V to V _{DD}
Sink current, discharge or output	150 mA
Source current, output	15 mA
Continuous total dissipation at (or below) 25°C free-air temperature	460 mW
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C

NOTE 1: All voltage values are with respect to network ground terminal.

4

Special Functions

electrical characteristics at specified free-air temperature, $V_{DD} = 1\text{ V}$

PARAMETER	TEST CONDITIONS†		MIN	TYP	MAX	UNIT
Threshold voltage level		25 °C	0.475	0.67	0.85	V
		Full range	0.45		0.875	
Threshold current		25 °C		10		pA
		MAX		75		
Trigger voltage level		25 °C	0.15	0.33	0.425	V
		Full range	0.1		0.45	
Trigger current		25 °C		10		pA
		MAX		75		
Reset voltage level		25 °C	0.4	0.7	1	V
		Full range	0.3		1	
Reset current		25 °C		10		pA
		MAX		75		
Control voltage (open-circuit) as a percentage of supply voltage		MAX	66.7%			
Discharge switch on-state voltage	$I_{OL} = 100\ \mu\text{A}$	25 °C		0.02	0.15	V
		Full range			0.2	
Discharge switch off-state current		25 °C		0.1		nA
		MAX		0.5		
Low-level output voltage	$I_{OL} = 100\ \mu\text{A}$	25 °C		0.03	0.2	V
		Full range			0.25	
High-level output voltage	$I_{OH} = -10\ \mu\text{A}$	25 °C	0.6	0.98		V
		Full range	0.6			
Supply current		25 °C		15	100	μA
		Full range			150	

†Full range (MIN to MAX) is 0 °C to 70 °C.

TLC551C
LinCMOS™ TIMER

electrical characteristics at specified free-air temperature, VDD = 2 V

PARAMETER	TEST CONDITIONS†	MIN	TYP	MAX	UNIT
Threshold voltage level	25°C	0.95	1.33	1.65	V
	Full range	0.85		1.75	
Threshold current	25°C	10			pA
	MAX	75			
Trigger voltage level	25°C	0.4	0.67	0.95	V
	Full range	0.3		1.05	
Trigger current	25°C	10			pA
	MAX	75			
Reset voltage level	25°C	0.4	1.1	1.5	V
	Full range	0.3		1.8	
Reset current	25°C	10			pA
	MAX	75			
Control voltage (open-circuit) as a percentage of supply voltage	MAX	66.7%			
Discharge switch on-state voltage	I _{OL} = 1 mA	25°C	0.03	0.2	V
		Full range		0.25	
Discharge switch off-state current		25°C	0.1		nA
		MAX	0.5		
Low-level output voltage	I _{OL} = 1 mA	25°C	0.07	0.3	V
		Full range		0.35	
High-level output voltage	I _{OH} = -300 μA	25°C	1.5	1.9	V
		Full range	1.5		
Supply current		25°C	65	250	μA
		Full range		400	

†Full range (MIN to MAX) is 0°C to 70°C.

4

Special Functions

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS†		MIN	TYP	MAX	UNIT
Threshold voltage level		25 °C	2.8	3.3	3.8	V
		Full range	2.7		3.9	
Threshold current		25 °C		10		pA
		MAX		75		
Trigger voltage level		25 °C	1.36	1.66	1.96	V
		Full range	1.26		2.06	
Trigger current		25 °C		10		pA
		MAX		75		
Reset voltage level		25 °C	0.4	1.1	1.5	V
		Full range	0.3		1.8	
Reset current		25 °C		10		pA
		MAX		75		
Control voltage (open-circuit) as a percentage of supply voltage		MAX	66.7%			
Discharge switch on-state voltage	$I_{OL} = 10\text{ mA}$	25 °C		0.14	0.5	V
		Full range			0.6	
Discharge switch off-state current		25 °C		0.1		nA
		MAX		0.5		
		Full range				
Low-level output voltage	$I_{OL} = 8\text{ mA}$	25 °C		0.21	0.4	V
		Full range			0.5	
	$I_{OL} = 5\text{ mA}$	25 °C		0.13	0.3	
		Full range			0.4	
	$I_{OL} = 3.2\text{ mA}$	25 °C		0.08	0.3	
		Full range			0.35	
High-level output voltage	$I_{OH} = -1\text{ mA}$	25 °C	4.1	4.8		V
		Full range	4.1			
Supply current		25 °C		170	350	μA
		Full range			500	

†Full range (MIN to MAX) is 0 °C to 70 °C.

TLC551C
LinCMOS™ TIMER

electrical characteristics at specified free-air temperature, V_{DD} = 15 V

PARAMETER	TEST CONDITIONS†	MIN	TYP	MAX	UNIT
Threshold voltage level	25 °C	9.45	10	10.55	V
	Full range	9.35		10.65	
Threshold current	25 °C		10		pA
	MAX		75		
Trigger voltage level	25 °C	4.65	5	5.35	V
	Full range	4.55		5.45	
Trigger current	25 °C		10		pA
	MAX		75		
Reset voltage level	25 °C	0.4	1.1	1.5	V
	Full range	0.3		1.8	
Reset current	25 °C		10		pA
	MAX		75		
Control voltage (open-circuit) as a percentage of supply voltage	MAX		66.7%		
Discharge switch on-state voltage	I _{OL} = 100 mA		0.77	1.7	V
	Full range			1.8	
Discharge switch off-state current	25 °C		0.1		nA
	MAX		0.5		
Low-level output voltage	I _{OL} = 100 mA	25 °C	1.28	3.2	V
		Full range		3.6	
	I _{OL} = 50 mA	25 °C	0.63	1	
		Full range		1.3	
	I _{OL} = 10 mA	25 °C	0.12	0.3	
		Full range		0.4	
High-level output voltage	I _{OH} = -10 mA	25 °C	12.5	14.2	V
		Full range	12.5		
	I _{OH} = -5 mA	25 °C	13.5	14.6	
		Full range	13.5		
	I _{OH} = -1 mA	25 °C	14.2	14.9	
		Full range	14.2		
Supply current	25 °C		360	600	μA
	Full range			800	

† Full range (MIN to MAX) is 0 °C to 70 °C.

4

Special Functions

electrical characteristics at specified free-air temperature, $V_{DD} = 18\text{ V}$

PARAMETER	TEST CONDITIONS†		MIN	TYP	MAX	UNIT
Threshold voltage level	25°C		11.4	12	12.6	V
	Full range		10.9		12.7	
Threshold current	25°C			10		pA
	MAX			75		
Trigger voltage level	25°C		5.6	6	6.4	V
	Full range		5.5		6.5	
Trigger current	25°C			10		pA
	MAX			75		
Reset voltage level	25°C		0.4	1.1	1.5	V
	Full range		0.3		1.8	
Reset current	25°C			10		pA
	MAX			75		
Control voltage (open-circuit) as a percentage of supply voltage	MAX		66.7%			
Discharge switch on-state voltage	$I_{OL} = 100\text{ mA}$	25°C		0.72	1.5	V
		Full range			1.6	
Discharge switch off-state current	25°C			0.1		nA
	MAX			0.5		
Low-level output voltage	$I_{OL} = 3.2\text{ mA}$	25°C		0.04	0.3	V
		Full range			0.35	
High-level output voltage	$I_{OH} = -1\text{ mA}$	25°C	17.3	17.9		V
		Full range	17.3			
Supply current	25°C			420	600	μA
	Full range				800	

†Full range (MIN to MAX) is 0°C to 70°C.

operating characteristics, $V_{DD} = 5\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Initial error of timing interval†	$V_{DD} = 5\text{ V to }15\text{ V}$, $R_A = R_B = 1\text{ k}\Omega\text{ to }100\text{ k}\Omega$, $C_T = 0.1\text{ }\mu\text{F}$, See Note 2		1%	3%	
Supply voltage sensitivity of timing interval				0.1	0.5
Output pulse rise time	$R_L = 10\text{ M}\Omega$, $C_L = 10\text{ pF}$		20	75	ns
Output pulse fall time			15	60	
Maximum frequency in astable mode	$R_A = 470\text{ }\Omega$, $R_B = 200\text{ }\Omega$, $C_T = 200\text{ pF}$, See Note 2	1.2	1.8		MHz

†Timing interval error is defined as the difference between the measured value and the nominal value of a random sample.

NOTE 2: R_A , R_B , and C_T are as defined in Figure 1.

TYPICAL APPLICATION DATA

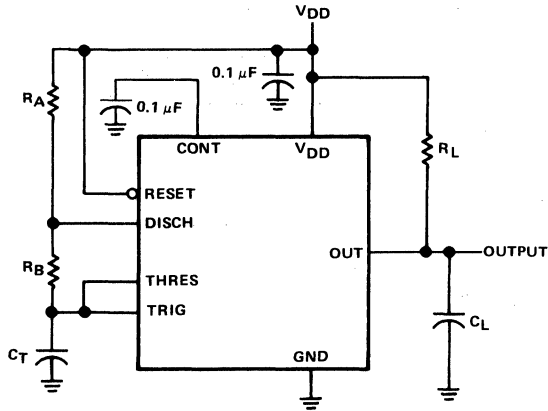


FIGURE 1. CIRCUIT FOR ASTABLE OPERATION

4

Special Functions

TLC552C DUAL LinCMOS™ TIMER

D2796, FEBRUARY 1984—REVISED MAY 1988

- Very Low Power Consumption . . . 2 mW Typ at $V_{DD} = 5\text{ V}$
- Capable of Operation in Astable Mode
- CMOS Output Capable of Swinging Rail to Rail
- High Output-Current Capability
 . . . Sink 100 mA Typ
 . . . Source 10 mA Typ
- Output Fully Compatible with CMOS, TTL, and MOS
- Low Supply Current Reduces Spikes During Output Transitions
- High-Impedance Inputs . . . $10^{12}\ \Omega$ Typ
- Single-Supply Operation from 1 V to 18 V
- Functionally Interchangeable with the NE555; Has Same Pinout

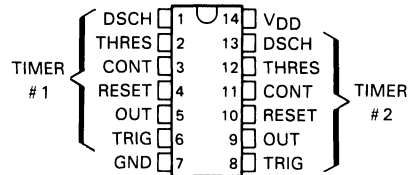
description

The TLC552 is a monolithic timing circuit fabricated using TI's LinCMOS™ process, which provides full compatibility with CMOS, TTL, and MOS logic and operation at frequencies up to 2 MHz. Accurate time delays and oscillations are possible with smaller, less-expensive timing capacitors than the NE555 because of the high input impedance. Power consumption is low across the full range of power supply voltages.

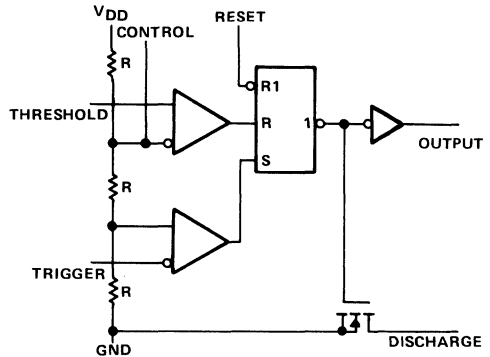
Like the NE555, the TLC552 has a trigger level approximately one-third of the supply voltage and a threshold level approximately two-thirds of the supply voltage. These levels can be altered by use of the control voltage terminal. When the trigger input falls below the trigger level, the flip-flop is set and the output goes high. If the trigger input is above the trigger level and the threshold input is above the threshold level, the flip-flop is reset and the output is low. The reset input can override all other inputs and can be used to initiate a new timing cycle. If the reset input is low, the flip-flop is reset and the output is low. Whenever the output is low, a low-impedance path is provided between the discharge terminal and ground.

While the CMOS output is capable of sinking over 100 milliamperes and sourcing over 10 milliamperes, the TLC552 exhibits greatly reduced supply-current spikes during output transitions. This minimizes the need for the large decoupling capacitors required by the NE555.

D OR N PACKAGE
(TOP VIEW)



functional block diagram (each timer)



Reset can override Trigger and Threshold.
Trigger can override Threshold.

AVAILABLE OPTIONS

SYMBOLIZATION		OPERATING TEMPERATURE RANGE	V_T MAX at 25°C
DEVICE	PACKAGE SUFFIX		
TLC552C	D,N	0°C to 70°C	3.8 mV

The D packages are available taped and reeled. Add the suffix R to the device type when ordering. (i.e. TLC552CDR)

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4

Special Functions

TLC552C

DUAL LinCMOS™ TIMER

description (continued)

These devices have internal electrostatic discharge (ESD) protection circuits that will prevent catastrophic failures at voltages up to 2000 volts as tested under MIL-STD-883C, Method 3015.2. However, care should be exercised in handling these devices as exposure to ESD may result in a degradation of the device parametric performance.

All unused inputs should be tied to an appropriate logic level to prevent false triggering.

The TLC552C is characterized for operation from 0°C to 70°C.

FUNCTION TABLE

RESET VOLTAGE†	TRIGGER VOLTAGE†	THRESHOLD VOLTAGE†	OUTPUT	DISCHARGE SWITCH
< MIN	Irrelevant	Irrelevant	Low	On
> MAX	< MIN	Irrelevant	High	Off
> MAX	> MAX	> MAX	Low	On
> MAX	> MAX	< MIN	As previously established	

†For conditions shown as MIN or MAX, use the appropriate value specified under electrical characteristics.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{DD} (see Note 1)	18 V
Input voltage range (any input)	-0.3 V to V_{DD}
Sink current, discharge or output	150 mA
Source current, output	15 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C

4

NOTE 1: All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	POWER RATING	DERATING FACTOR	ABOVE T_A
D	950 mW	7.6 mW/°C	25°C
N	1150 mW	9.2 mW/°C	25°C

recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, V_{DD}	1	18	V
Operating free-air temperature, T_A	0	70	°C

Special Functions

electrical characteristics at specified free-air temperature, V_{DD} = 1 V

PARAMETER	TEST CONDITIONS†	MIN	TYP	MAX	UNIT	
Threshold voltage level	25°C	0.475	0.67	0.85	V	
	Full range	0.45		0.875		
Threshold current	25°C		10		pA	
	MAX		75			
Trigger voltage level	25°C	0.15	0.33	0.425	V	
	Full range	0.1		1.45		
Trigger current	25°C		10		pA	
	MAX		75			
Reset voltage level	25°C	0.4	0.7	1	V	
	Full range	0.3		1		
Reset current	25°C		10		pA	
	MAX		75			
Control voltage (open-circuit) as a percentage of supply voltage	MAX		66.7%			
Discharge switch on-state voltage	I _{OL} = 100 μA		25°C	0.02	0.15	V
			Full range		0.2	
Discharge switch off-state current			25°C	0.1		nA
			MAX	0.5		
Low-level output voltage	I _{OL} = 100 μA		25°C	0.03	0.2	V
			Full range		0.25	
High-level output voltage	I _{OH} = -10 μA		25°C	0.6	0.98	V
			Full range	0.6		
Supply current			25°C	30	200	μA
			Full range		300	

† Full range (MIN to MAX) is 0°C to 70°C.

TLC552C
DUAL LinCMOS™ TIMER

electrical characteristics at specified free-air temperature, V_{DD} = 2 V

PARAMETER	TEST CONDITIONS†	MIN	TYP	MAX	UNIT
Threshold voltage level	25 °C	0.95	1.33	1.65	V
	Full range	0.85		1.75	
Threshold current	25 °C		10		pA
	MAX		75		
Trigger voltage level	25 °C	0.4	0.67	0.95	V
	Full range	0.3		1.05	
Trigger current	25 °C		10		pA
	MAX		75		
Reset voltage level	25 °C	0.4	1.1	1.5	V
	Full range	0.3		1.8	
Reset current	25 °C		10		pA
	MAX		75		
Control voltage (open-circuit) as a percentage of supply voltage	MAX		66.7%		
Discharge switch on-state voltage	I _{OL} = 1 mA	25 °C	0.03	0.2	V
		Full range		0.25	
Discharge switch off-state current		25 °C	0.1		nA
		MAX	0.5		
Low-level output voltage	I _{OL} = 1 mA	25 °C	0.07	0.3	V
		Full range		0.35	
High-level output voltage	I _{OH} = -300 μA	25 °C	1.5	1.9	V
		Full range	1.5		
Supply current		25 °C	130	500	μA
		Full range		800	

† Full range (MIN to MAX) is 0 °C to 70 °C.

4

Special Functions

TLC552C
DUAL LinCMOS™ TIMER

electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS†	MIN	TYP	MAX	UNIT
Threshold voltage level	25°C	2.8	3.3	3.8	V
	Full range	2.7		3.9	
Threshold current	25°C		10		pA
	MAX		75		
Trigger voltage level	25°C	1.36	1.66	1.96	V
	Full range	1.26		2.06	
Trigger current	25°C		10		pA
	MAX		75		
Reset voltage level	25°C	0.4	1.1	1.5	V
	Full range	0.3		1.8	
Reset current	25°C		10		pA
	MAX		75		
Control voltage (open-circuit) as a percentage of supply voltage	MAX	66.7%			
Discharge switch on-state voltage	$I_{OL} = 10\text{ mA}$	25°C	0.14	0.5	V
	Full range			0.6	
Discharge switch off-state current		25°C	0.1		nA
		MAX	0.5		
Low-level output voltage	$I_{OL} = 8\text{ mA}$	25°C	0.21	0.4	V
		Full range		0.5	
	$I_{OL} = 5\text{ mA}$	25°C	0.13	0.3	
		Full range		0.4	
	$I_{OL} = 3.2\text{ mA}$	25°C	0.08	0.3	
		Full range		0.35	
High-level output voltage	$I_{OH} = -1\text{ mA}$	25°C	4.1	4.8	V
		Full range	4.1		
Supply current		25°C	340	700	μA
		Full range		1000	

† Full range (MIN to MAX) is 0°C to 70°C.

4

Special Functions

TLC552C
DUAL LinCMOS™ TIMER

electrical characteristics at specified free-air temperature, $V_{DD} = 15\text{ V}$

PARAMETER	TEST CONDITIONS†	MIN	TYP	MAX	UNIT
Threshold voltage level	25 °C	9.45	10	10.55	V
	Full range	9.35		10.65	
Threshold current	25 °C		10		pA
	MAX		75		
Trigger voltage level	25 °C	4.65	5	5.35	V
	Full range	4.55		5.45	
Trigger current	25 °C		10		pA
	MAX		75		
Reset voltage level	25 °C	0.4	1.1	1.5	V
	Full range	0.3		1.8	
Reset current	25 °C		10		pA
	MAX		75		
Control voltage (open-circuit) as a percentage of supply voltage	MAX		66.7%		
Discharge switch on-state voltage	$I_{OL} = 100\text{ mA}$	25 °C	0.77	1.7	V
		Full range		1.8	
Discharge switch off-state current		25 °C	0.1		nA
		MAX	0.5		
Low-level output voltage	$I_{OL} = 100\text{ mA}$	25 °C	1.28	3.2	V
		Full range		3.6	
	$I_{OL} = 50\text{ mA}$	25 °C	0.63	1	
		Full range		1.3	
	$I_{OL} = 10\text{ mA}$	25 °C	0.12	0.3	
		Full range		0.4	
High-level output voltage	$I_{OH} = -10\text{ mA}$	25 °C	12.5	14.2	V
		Full range	12.5		
	$I_{OH} = -5\text{ mA}$	25 °C	13.5	14.6	
		Full range	13.5		
	$I_{OH} = -1\text{ mA}$	25 °C	14.2	14.9	
		Full range	14.2		
Supply current		25 °C	0.72	1.2	mA
		Full range		1.6	

† Full range (MIN to MAX) is 0 °C to 70 °C.

electrical characteristics at specified free-air temperature, V_{DD} = 18 V

PARAMETER	TEST CONDITIONS†	MIN	TYP	MAX	UNIT
Threshold voltage level	25 °C	11.4	12	12.6	V
	Full range	10.9		12.7	
Threshold current	25 °C		10		pA
	MAX		75		
Trigger voltage level	25 °C	5.6	6	6.4	V
	Full range	5.5		6.5	
Trigger current	25 °C		10		pA
	MAX		75		
Reset voltage level	25 °C	0.4	1.1	1.5	V
	Full range	0.3		1.8	
Reset current	25 °C		10		pA
	MAX		75		
Control voltage (open-circuit) as a percentage of supply voltage	MAX	66.7%			
Discharge switch on-state voltage	I _{OL} = 100 mA	25 °C	0.72	1.5	V
		Full range		1.6	
Discharge switch off-state current		25 °C	0.1		nA
		MAX	0.5		
Low-level output voltage	I _{OL} = 3.2 mA	25 °C	0.04	0.3	V
		Full range		0.35	
High-level output voltage	I _{OH} = -1 mA	25 °C	17.3	17.9	V
		Full range	17.3		
Supply current		25 °C	0.84	1.2	mA
		Full range		1.6	

† Full range (MIN to MAX) is 0 °C to 70 °C.

operating characteristics, V_{DD} = 5 V, T_A = 25 °C (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Initial error of timing interval‡	V _{DD} = 5 V to 15 V,		1%	3%	
Supply voltage sensitivity of timing interval	R _A = R _B = 1 kΩ to 100 kΩ, C _T = 0.1 μF, See Note 2		0.1	0.5	%/V
Output pulse rise time	R _L = 10 MΩ, C _L = 10 pF		20	75	ns
Output pulse fall time			15	60	
Maximum frequency in astable mode	R _A = 470 Ω, R _B = 200 Ω, C _T = 200 pF, See Note 2	1.2	2.8		MHz

NOTE 2: R_A, R_B, and C_T are as defined in Figure 1.

‡ Timing interval error is defined as the difference between the measured value and the nominal value of a random sample.

TYPICAL APPLICATION DATA

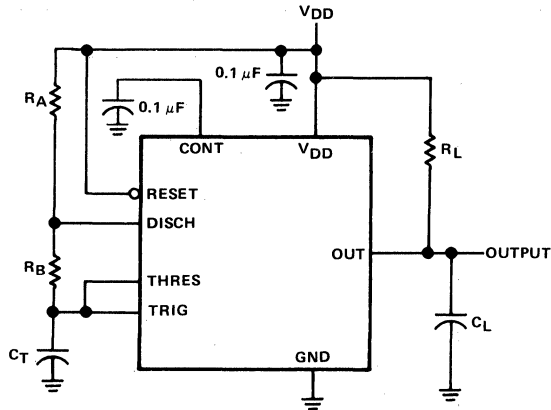


FIGURE 1. CIRCUIT FOR ASTABLE OPERATION

4

Special Functions

TLC555M, TLC555I, TLC555C LinCMOS™ TIMERS

D2784, SEPTEMBER 1983—REVISED OCTOBER 1988

- Very Low Power Consumption . . . 1 mW
Typ at $V_{DD} = 5\text{ V}$
- Capable of Operation in Astable Mode
- CMOS Output Capable of Swinging Rail to Rail
- High Output-Current Capability
. . . Sink 100 mA Typ
. . . Source 10 mA Typ
- Output Fully Compatible with CMOS, TTL,
and MOS
- Low Supply Current Reduces Spikes During
Output Transitions
- High-Impedance Inputs . . . $10^{12}\ \Omega$ Typ
- Single-Supply Operation from 2 V to 18 V
- Functionally Interchangeable with the
NE555; Has Same Pinout

description

The TLC555 is a monolithic timing circuit fabricated using TI's LinCMOS™ process, which provides full compatibility with CMOS, TTL, and MOS logic and operation at frequencies up to 2 MHz. Accurate time delays and oscillations are possible with smaller, less-expensive timing capacitors than the NE555 because of the high input impedance. Power consumption is low across the full range of power supply voltage.

Like the NE555, the TLC555 has a trigger level approximately one-third of the supply voltage and a threshold level approximately two-thirds of the supply voltage. These levels can be altered by use of the control voltage terminal. When the trigger input falls below the trigger level, the flip-flop is set and the output goes high. If the trigger input is above the trigger level and the threshold input is above the threshold level, the flip-flop is reset and the output is low. The reset input can override all other inputs and can be used to initiate a new timing cycle. If the reset input is low, the flip-flop is reset and the output is low. Whenever the output is low, a low-impedance path is provided between the discharge terminal and ground.

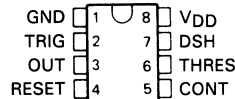
While the CMOS output is capable of sinking over 100 mA and sourcing over 10 mA, the TLC555 exhibits greatly reduced supply-current spikes during output transitions. This minimizes the need for the large decoupling capacitors required by the NE555.

These devices have internal electrostatic discharge (ESD) protection circuits that will prevent catastrophic failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015. However, care should be exercised in handling these devices, as exposure to ESD may result in degradation of the device parametric performance.

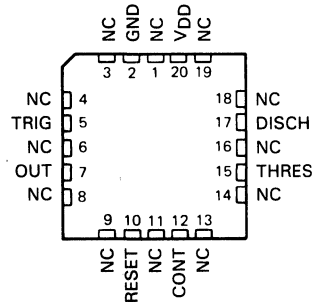
All unused inputs should be tied to an appropriate logic level to prevent false triggering.

The TLC555M is characterized for operation over the full military temperature range of $-55\text{ }^{\circ}\text{C}$ to $125\text{ }^{\circ}\text{C}$. The TLC555I is characterized for operation from $-40\text{ }^{\circ}\text{C}$ to $85\text{ }^{\circ}\text{C}$. The TLC555C is characterized for operation from $0\text{ }^{\circ}\text{C}$ to $70\text{ }^{\circ}\text{C}$.

TLC555M . . . JG PACKAGE
TLC555I, TLC555C . . . D OR P PACKAGE
(TOP VIEW)



TLC555M . . . FK PACKAGE
(TOP VIEW)



NC—No internal connection

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**TEXAS
INSTRUMENTS**

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4-195

TLC555M, TLC555I, TLC555C LinCMOS™ TIMERS

AVAILABLE OPTIONS

T _A RANGE	V _{CC} RANGE	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	2 V to 18 V	TLC555CD			TLC555CP
-40°C to 85°C	3 V to 18 V	TLC555ID			TLC555IP
-55°C to 125°C	5 V to 18 V		TLC555MFK	TLC555MJG	

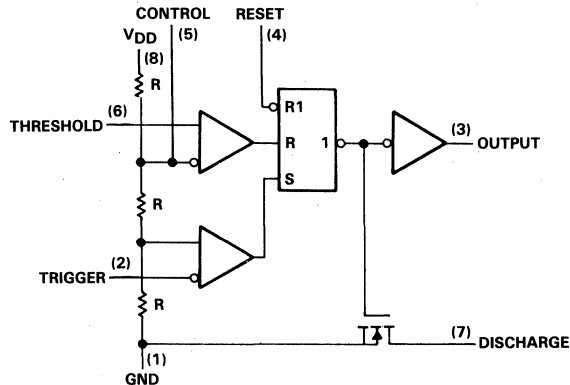
The D package is available taped and reeled. Add the suffix R to the device type (e.g., TLC555CDR).

FUNCTION TABLE

RESET VOLTAGE†	TRIGGER VOLTAGE†	THRESHOLD VOLTAGE†	OUTPUT	DISCHARGE SWITCH
< MIN	Irrelevant	Irrelevant	Low	On
> MAX	< MIN	Irrelevant	High	Off
> MAX	> MAX	> MAX	Low	On
> MAX	> MAX	< MIN	As previously established	

†For conditions shown as MIN or MAX, use the appropriate value specified under electrical characteristics.

functional block diagram



Pin numbers are for all packages except FK.
Reset can override Trigger, which can override Threshold.

4

Special Functions

TLC555M, TLC555I, TLC555C
LinCMOS™ TIMERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

		TLC555M	TLC555I	TLC555C	UNIT
Supply voltage (see Note 1)		18	18	18	V
Input voltage		-0.3 to V_{DD}	-0.3 to V_{DD}	-0.3 to V_{DD}	V
Sink current, discharge or output		150	150	150	mA
Source current, output		15	15	15	mA
Continuous total power dissipation		See Dissipation Rating Table			
Operating free-air temperature range		-55 to 125	-40 to 85	0 to 70	°C
Storage temperature range		-65 to 150	-65 to 150	-65 to 150	°C
Case temperature for 60 seconds	FK package	260			°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	JG package	300			
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package		260	260	

NOTE 1: All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	N/A
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	N/A

TLC555I, TLC555C LinCMOS™ TIMERS

electrical characteristics at specified free-air temperature, VDD = 3 V for TLC555I, VDD = 2 V for TLC555C

PARAMETER	TEST CONDITIONS†	TLC555I			TLC555C			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX		
Threshold voltage level	25 °C	1.6		2.4	0.95	1.33	1.65	V	
	Full range	1.5		2.5	0.85		1.75		
Threshold current	25 °C	10			10			pA	
	MAX	150			75				
Trigger voltage level	25 °C	0.71	1.0	1.29	0.4	0.67	0.95	V	
	Full range	0.61		1.39	0.3		1.05		
Trigger current	25 °C	10			10			pA	
	MAX	150			75				
Reset voltage level	25 °C	0.4	1.1	1.5	0.4	1.1	1.5	V	
	Full range	0.3		1.8	0.3		2		
Reset current	25 °C	10			10			pA	
	MAX	150			75				
Control voltage (open-circuit) as a percentage of supply voltage	MAX	66.7%			66.7%				
Discharge switch on-state voltage	I _{OL} = 1 mA	25 °C	0.03	0.2	0.03	0.2		V	
		Full range	0.375			0.25			
Discharge switch off-state current		25 °C	0.1			0.1			nA
		MAX	120			0.5			
Low-level output voltage	I _{OL} = 1 mA	25 °C	0.07	0.3	0.07	0.3		V	
		Full range	0.4			0.35			
High-level output voltage	I _{OH} = -300 μA	25 °C	1.5	1.9	1.5	1.9		V	
		Full range	2.5			1.5			
Supply current		25 °C	250			250			μA
		Full range	500			400			

† Full range (MIN to MAX) is -40 °C to 85 °C for TLC555I and 0 °C to 70 °C for TLC555C.

NOTE 2: These values apply for the expected operating configurations in which the Threshold terminal is connected directly to the Discharge terminal or to the Trigger terminal.

4

Special Functions

electrical characteristics at specified free-air temperature, VDD = 5 V

PARAMETER	TEST CONDITIONS †	TLC555M			TLC555I			TLC555C			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX		
Threshold voltage level	25 °C	2.8	3.3	3.8	2.8	3.3	3.8	2.8	3.3	3.8	V	
	Full range	2.7		3.9	2.7		3.9	2.7		3.9		
Threshold current	25 °C	10			10			10			pA	
	MAX	5000			150			75				
Trigger voltage level	25 °C	1.36	1.66	1.96	1.36	1.66	1.96	1.36	1.66	1.96	V	
	Full range	1.26		2.06	1.26		2.06	1.26		2.06		
Trigger current	25 °C	10			10			10			pA	
	MAX	5000			150			75				
Reset voltage level	25 °C	0.4	1.1	1.5	0.4	1.1	1.5	0.4	1.1	1.5	V	
	Full range	0.3		1.8	0.3		1.8	0.3		1.8		
Reset current	25 °C	10			10			10			pA	
	MAX	5000			150			75				
Control voltage (open-circuit) as a percentage of supply voltage	MAX	66.7%			66.7%			66.7%				
Discharge switch on-state voltage	IOL = 10 mA	25 °C	0.14	0.5	0.14	0.5	0.14	0.5			V	
		Full range		0.6		0.6		0.6		0.6		
Discharge switch off-state current		25 °C	0.1			0.1			0.1			nA
		MAX	120			120			0.5			
Low-level output voltage	IOL = 8 mA	25 °C	0.21	0.4	0.21	0.4	0.21	0.4			V	
		Full range		0.6		0.5		0.5		0.5		
	IOL = 5 mA	25 °C	0.13	0.3	0.13	0.3	0.13	0.3				
		Full range		0.45		0.4		0.4		0.4		
	IOL = 3.2 mA	25 °C	0.08	0.3	0.08	0.3	0.08	0.3				
		Full range		0.4		0.35		0.35		0.35		
High-level output voltage	IOH = -1 mA	25 °C	4.1	4.8	4.1	4.8	4.1	4.8			V	
		Full range	4.1		4.1		4.1					
Supply current	See Note 2	25 °C	170	350	170	350	170	350			μA	
		Full range		700		600		500		500		

† Full range (MIN to MAX) is -55 °C to 125 °C for TLC555M, -40 °C to 85 °C for TLC555I, and 0 °C to 70 °C for TLC555C.

NOTE 2: These values apply for the expected operating configurations in which the Threshold terminal is connected directly to the Discharge terminal or to the Trigger terminal.

TLC555M, TLC555I, TLC555C LinCMOS™ TIMERS

electrical characteristics at specified free-air temperature, VDD = 15 V

PARAMETER	TEST CONDITIONS†	TLC555M			TLC555I			TLC555C			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX		
Threshold voltage level	25 °C	9.45	10	10.55	9.45	10	10.55	9.45	10	10.55	V	
	Full range	9.35		10.65	9.35		10.65	9.35		10.65		
Threshold current	25 °C	10			10			10			pA	
	MAX	5000			150			75				
Trigger voltage level	25 °C	4.65	5	5.35	4.65	5	5.35	4.65	5	5.35	V	
	Full range	4.55		5.45	4.55		5.45	4.55		5.45		
Trigger current	25 °C	10			10			10			pA	
	MAX	5000			150			75				
Reset voltage level	25 °C	0.4	1.1	1.5	0.4	1.1	1.5	0.4	1.1	1.5	V	
	Full range	0.3		1.8	0.3		1.8	0.3		1.8		
Reset current	25 °C	10			10			10			pA	
	MAX	5000			150			75				
Control voltage (open-circuit) as a percentage of supply voltage	MAX	66.7%			66.7%			66.7%				
Discharge switch on-state voltage	I _{OL} = 100 mA	25 °C	0.77	1.7	0.77	1.7	0.77	1.7	0.77	1.7	V	
		Full range		1.8		1.8		1.8		1.8		
Discharge switch off-state current		25 °C	0.1			0.1			0.1			nA
		MAX	120			120			0.5			
Low-level output voltage	I _{OL} = 100 mA	25 °C	1.28	3.2	1.28	3.2	1.28	3.2	1.28	3.2	V	
		Full range		3.8		3.7		3.6		3.6		
	I _{OL} = 50 mA	25 °C	0.63	1	0.63	1	0.63	1	0.63	1		
		Full range		1.5		1.4		1.3		1.3		
	I _{OL} = 10 mA	25 °C	0.12	0.3	0.12	0.3	0.12	0.3	0.12	0.3		
		Full range		0.45		0.4		0.4		0.4		
High-level output voltage	I _{OH} = -10 mA	25 °C	12.5	14.2	12.5	14.2	12.5	14.2	12.5	14.2	V	
		Full range	12.5		12.5		12.5		12.5			
	I _{OH} = -5 mA	25 °C	13.5	14.6	13.5	14.6	13.5	14.6	13.5	14.6		
		Full range	13.5		13.5		13.5		13.5			
	I _{OH} = -1 mA	25 °C	14.2	14.9	14.2	14.9	14.2	14.9	14.2	14.9		
		Full range	14.2		14.2		14.2		14.2			
Supply current	See Note 2	25 °C	360	600	360	600	360	600	360	600	μA	
		Full range		1000		900		800		800		

† Full range (MIN to MAX) is -55 °C to 125 °C for TLC555M, -40 °C to 85 °C for TLC555I, and 0 °C to 70 °C for TLC555C.

NOTE 2: These values apply for the expected operating configurations in which the Threshold terminal is connected directly to the Discharge terminal or to the Trigger terminal.

4

Special Functions

electrical characteristics at specified free-air temperature, $V_{DD} = 18\text{ V}$

PARAMETER	TEST CONDITIONS†	TLC556M			TLC556I			TLC556C			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX		
Threshold voltage level	25°C	11.4	12	12.6	11.4	12	12.6	11.4	12	12.6	V	
	Full range	10.9		12.7	10.9		12.7	10.9		12.7		
Threshold current	25°C	10			10			10			pA	
	MAX	5000			150			75				
Trigger voltage level	25°C	5.6	6	6.4	5.6	6	6.4	5.6	6	6.4	V	
	Full range	5.5		6.5	5.5		6.5	5.5		6.5		
Trigger current	25°C	10			10			10			pA	
	MAX	5000			150			75				
Reset voltage level	25°C	0.4	1.1	1.5	0.4	1.1	1.5	0.4	1.1	1.5	V	
	Full range	0.3		1.8	0.3		1.8	0.3		1.8		
Reset current	25°C	10			10			10			pA	
	MAX	5000			150			75				
Control voltage (open-circuit) as a percentage of supply voltage	MAX	66.7%			66.7%			66.7%				
Discharge switch on-state voltage	$I_{OL} = 100\text{ mA}$	25°C	0.72	1.5	0.72	1.5	0.72	1.5	0.72	1.5	V	
	Full range			1.6		1.6		1.6		1.6		
Discharge switch off-state current		25°C	0.1			0.1			0.1			nA
	MAX	120			120			0.5				
Low-level output voltage	$I_{OL} = 3.2\text{ mA}$	25°C	0.04	0.3	0.04	0.3	0.04	0.3	0.04	0.3	V	
	Full range			0.4		0.35		0.35		0.35		
High-level output voltage	$I_{OH} = -1\text{ mA}$	25°C	17.3	17.9	17.3	17.9	17.3	17.9	17.3	17.9	V	
	Full range		17.3		17.3		17.3		17.3			
Supply current	See Note 2	25°C	600			600			600			μA
	Full range		1000			900			800			

†Full range (MIN to MAX) is -55°C to 125°C for TLC556M, -40°C to 85°C for TLC556I, and 0°C to 70°C for TLC556C.

NOTE 2: These values apply for the expected operating configuration in which the Threshold terminal is connected directly to the Discharge terminal or to the Trigger terminal.

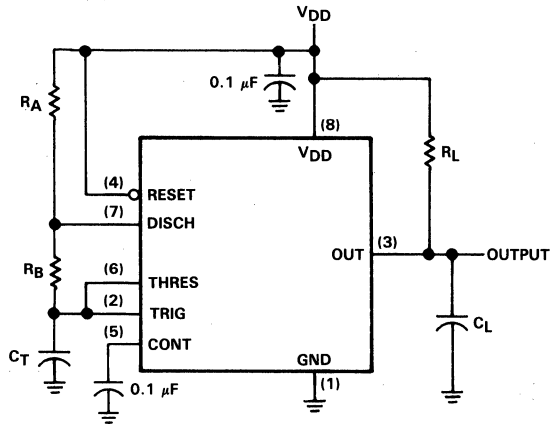
operating characteristics, $V_{DD} = 5\text{ V}$, $T_A = 25^{\circ}\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Initial error of timing interval‡	$V_{DD} = 5\text{ V to }15\text{ V}$,		1%	3%	
Supply voltage sensitivity of timing interval	$R_A = R_B = 1\text{ k}\Omega$ to $100\text{ k}\Omega$,		0.1	0.5	%/V
	$C_T = 0.1\text{ }\mu\text{F}$, See Note 3				
Output pulse rise time	$R_L = 10\text{ M}\Omega$, $C_L = 10\text{ pF}$		20	75	ns
Output pulse fall time			15	60	
Maximum frequency in astable mode	$R_A = 470\text{ }\Omega$, $R_B = 200\text{ }\Omega$,	1.2	2.1		MHz
	$C_T = 200\text{ pF}$, See Note 3				

‡Timing interval error is defined as the difference between the measured value and the average value of a random sample from each process run.

NOTE 3: R_A , R_B , and C_T are as defined in Figure 1.

TYPICAL APPLICATION DATA



Pin numbers are for all packages except FK.

FIGURE 1. CIRCUIT FOR ASTABLE OPERATION

4

Special Functions

TLC556M, TLC556I, TLC556C DUAL LinCMOS™ TIMERS

D2796, FEBRUARY 1984—REVISED OCTOBER 1988

- Very Low Power Consumption . . . 2 mW Typ at $V_{DD} = 5\text{ V}$
- Capable of Operation in Astable Mode
- CMOS Output Capable of Swinging Rail to Rail
- High Output-Current Capability . . . Sink 100 mA Typ . . . Source 10 mA Typ
- Output Fully Compatible with CMOS, TTL, and MOS
- Low Supply Current Reduces Spikes During Output Transitions
- High-Impedance Inputs . . . $10^{12}\ \Omega$ Typ
- Single-Supply Operation from 2 V to 18 V
- Functionally Interchangeable with the NE556; Has Same Pinout

description

The TLC556 is a monolithic timing circuit fabricated using TI's LinCMOS™ process, which provides full compatibility with CMOS, TTL, and MOS logic and operation at frequencies up to 2 MHz. Accurate time delays and oscillations are possible with smaller, less-expensive timing capacitors than the NE556 because of the high input impedance. Power consumption is low across the full range of power supply voltages.

Like the NE556, the TLC556 has a trigger level approximately one-third of the supply voltage and a threshold level approximately two-thirds of the supply voltage. These levels can be altered by use of the control voltage terminal. When the trigger input falls below the trigger level, the flip-flop is set and the output goes high. If the trigger input is above the trigger level and the threshold input is above the threshold level, the flip-flop is reset and the output is low. The reset input can override all other inputs and can be used to initiate a new timing cycle. If the reset input is low, the flip-flop is reset and the output is low. Whenever the output is low, a low-impedance path is provided between the discharge terminal and ground.

While the CMOS output is capable of sinking over 100 mA and sourcing over 10 mA, the TLC556 exhibits greatly reduced supply-current spikes during output transitions. This minimizes the need for the large decoupling capacitors required by the NE556.

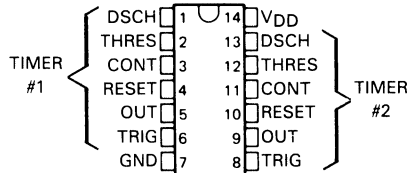
These devices have internal electrostatic discharge (ESD) protection circuits that will prevent catastrophic failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015. However, care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.

All unused inputs should be tied to an appropriate logic level to prevent false triggering.

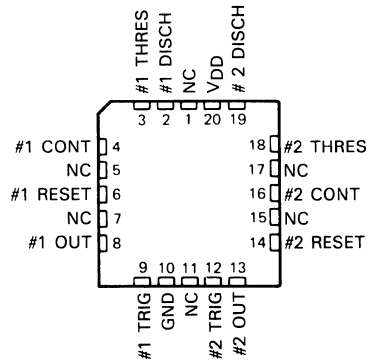
The TLC556M is characterized for operation over the full military temperature range of $-55\text{ }^{\circ}\text{C}$ to $125\text{ }^{\circ}\text{C}$. The TLC556I is characterized for operation from $-40\text{ }^{\circ}\text{C}$ to $85\text{ }^{\circ}\text{C}$. The TLC556C is characterized for operation from $0\text{ }^{\circ}\text{C}$ to $70\text{ }^{\circ}\text{C}$.

LinCMOS is a trademark of Texas Instruments Incorporated.

TLC556M . . . J PACKAGE
TLC556I, TLC556C . . . D OR N PACKAGE
(TOP VIEW)



TLC0556M . . . FK PACKAGE
(TOP VIEW)



NC—No internal connection

4

Special Functions

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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TLC556M, TLC556I, TLC556C DUAL LinCMOS™ TIMERS

AVAILABLE OPTIONS

TA RANGE	VCC RANGE	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	2 V to 18 V	TLC556CD			TLC556CN
-40°C to 85°C	3 V to 18 V	TLC556ID			TLC556IN
-55°C to 125°C	5 V to 18 V		TLC556MFK	TLC556MJ	

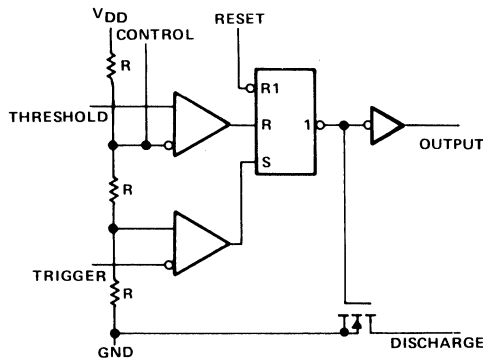
The D package is available taped and reeled. Add the suffix R to the device type (e.g., TLC556CDR).

FUNCTION TABLE

RESET VOLTAGE†	TRIGGER VOLTAGE†	THRESHOLD VOLTAGE†	OUTPUT	DISCHARGE SWITCH
<MIN	Irrelevant	Irrelevant	Low	On
>MAX	<MIN	Irrelevant	High	Off
>MAX	>MAX	>MAX	Low	On
>MAX	>MAX	<MIN	As previously established	

†For conditions shown as MIN or MAX, use the appropriate value specified under electrical characteristics.

functional block diagram (each timer)



Reset can override Trigger and Threshold.
Trigger can override Threshold.

4

Special Functions

TLC556M, TLC556I, TLC556C DUAL LinCMOS™ TIMERS

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	TLC556M	TLC556I	TLC556C	UNIT
Supply voltage (see Note 1)	18	18	18	V
Input voltage	-0.3 to V_{DD}	-0.3 to V_{DD}	-0.3 to V_{DD}	V
Sink current, discharge or output	150	150	150	mA
Source current, output	15	15	15	mA
Continuous total power dissipation	See Dissipation Rating Table			
Operating free-air temperature range	-55 to 125	-40 to 85	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	°C
Case temperature for 60 seconds	FK package	260		°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J package	300		
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or N package		260	

NOTE 1: All voltage values are with respect to network ground terminal.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	N/A
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW	N/A

TLC556I, TLC556C DUAL LinCMOS™ TIMERS

electrical characteristics at specified free-air temperature, $V_{DD} = 3\text{ V}$ for TLC556I, $V_{DD} = 2\text{ V}$ for TLC556C

PARAMETER	TEST CONDITIONS†	TLC556I			TLC556C			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX		
Threshold voltage level	25 °C	1.6	2.0	2.4	0.95	1.33	1.65	V	
	Full range	1.5		2.5	0.85		1.75		
Threshold current	25 °C	10			10			pA	
	MAX	150			75				
Trigger voltage level	25 °C	0.71	1.0	1.29	0.4	0.67	0.95	V	
	Full range	0.61		1.39	0.3		1.05		
Trigger current	25 °C	10			10			pA	
	MAX	150			75				
Reset voltage level	25 °C	0.4	1.1	1.5	0.4	1.1	1.5	V	
	Full range	0.3		1.8	0.3		1.8		
Reset current	25 °C	10			10			pA	
	MAX	150			75				
Control voltage (open-circuit) as a percentage of supply voltage	MAX	66.7%			66.7%				
Discharge switch on-state voltage	$I_{OL} = 1\text{ mA}$	25 °C	0.03	0.2	0.04	0.2	V		
	Full range			0.375		0.25			
Discharge switch off-state current		25 °C	0.1			0.1			nA
		MAX	120			0.5			
Low-level output voltage	$I_{OL} = 1\text{ mA}$	25 °C	0.07	0.3	0.07	0.3	V		
	Full range			0.4		0.35			
High-level output voltage	$I_{OH} = -300\text{ }\mu\text{A}$	25 °C	1.5	1.9	1.5	1.9	V		
	Full range		2.5		1.5				
Supply current		25 °C	130	500	130	500	μA		
	Full range			1000		800			

† Full range (MIN to MAX) is $-40\text{ }^{\circ}\text{C}$ to $85\text{ }^{\circ}\text{C}$ for TLC556I and $0\text{ }^{\circ}\text{C}$ to $70\text{ }^{\circ}\text{C}$ for TLC556C.

NOTE 2: These values apply for the expected operating configurations in which the Threshold terminal is connected directly to the Discharge terminal or to the Trigger terminal.

4

Special Functions

TLC556M, TLC556I, TLC556C
DUAL LinCMOS™ TIMERS

electrical characteristics at specified free-air temperature, V_{DD} = 5 V

PARAMETER	TEST CONDITIONS†	TLC556M			TLC556I			TLC556C			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX		
Threshold voltage level	25°C	2.8	3.3	3.8	2.8	3.3	3.8	2.8	3.3	3.8	V	
	Full range	2.7		3.9	2.7		3.9	2.7		3.9		
Threshold current	25°C	10			10			10			pA	
	MAX	5000			150			75				
Trigger voltage level	25°C	1.36	1.66	1.96	1.36	1.66	1.96	1.36	1.66	1.96	V	
	Full range	1.26		2.06	1.26		2.06	1.26		2.06		
Trigger current	25°C	10			10			10			pA	
	MAX	5000			150			75				
Reset voltage level	25°C	0.4	1.1	1.5	0.4	1.1	1.5	0.4	1.1	1.5	V	
	Full range	0.3		1.8	0.3		1.8	0.3		1.8		
Reset current	25°C	10			10			10			pA	
	MAX	5000			150			75				
Control voltage (open-circuit) as a percentage of supply voltage	MAX	66.7%			66.7%			66.7%				
Discharge switch on-state voltage	I _{OL} = 10 mA	25°C	0.15	0.5	0.15	0.5		0.15	0.5		V	
	Full range			0.6		0.6			0.6			
Discharge switch off-state current		25°C	0.1			0.1			0.1			nA
	MAX	120			2			0.5				
Low-level output voltage	I _{OL} = 8 mA	25°C	0.21	0.4	0.21	0.4		0.21	0.4		V	
		Full range		0.6		0.5			0.5			
	I _{OL} = 5 mA	25°C	0.13	0.3	0.13	0.3		0.13	0.3			
		Full range		0.45		0.4			0.4			
	I _{OL} = 3.2 mA	25°C	0.08	0.3	0.08	0.3		0.08	0.3			
		Full range		0.4		0.35			0.35			
High-level output voltage	I _{OH} = -1 mA	25°C	4.1	4.8	4.1	4.8		4.1	4.8		V	
		Full range	4.1		4.1			4.1				
Supply current	See Note 2	25°C	340	700	340	700		340	700		μA	
		Full range		1400		1200			1000			

†Full range (MIN to MAX) is -55°C to 125°C for TLC556M, -40°C to 85°C for TLC556I, and 0°C to 70°C for TLC556C.

NOTE 2: These values apply for the expected operating configurations in which the Threshold terminal is connected directly to the Discharge terminal or to the Trigger terminal.

TLC556M, TLC556I, TLC556C

DUAL LinCMOS™ TIMERS

electrical characteristics at specified free-air temperature, VDD = 15 V

PARAMETER	TEST CONDITIONS†	TLC556M			TLC556I			TLC556C			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Threshold voltage level	25°C	9.45	10	10.55	9.45	10	10.55	9.45	10	10.55	V
	Full range	9.35		10.65	9.35		10.65	9.35		10.65	
Threshold current	25°C	10			10			10			pA
	MAX	5000			150			75			
Trigger voltage level	25°C	4.65	5	5.35	4.65	5	5.35	4.65	5	5.35	V
	Full range	4.55		5.45	4.55		5.45	4.55		5.45	
Trigger current	25°C	10			10			10			pA
	MAX	5000			150			75			
Reset voltage level	25°C	0.4	1.1	1.5	0.4	1.1	1.5	0.4	1.1	1.5	V
	Full range	0.3		1.8	0.3		1.8	0.3		1.8	
Reset current	25°C	10			10			10			pA
	MAX	5000			150			75			
Control voltage (open-circuit) as a percentage of supply voltage	MAX	66.7%			66.7%			66.7%			
Discharge switch on-state voltage	I _{OL} = 100 mA	25°C	0.8	1.7	0.8	1.7	0.8	1.7	0.8	1.7	V
	Full range			1.8		1.8		1.8		1.8	
Discharge switch off-state current	25°C	0.1			0.1			0.1			nA
	MAX	120			2			0.5			
Low-level output voltage	I _{OL} = 100 mA	25°C	1.28	3.2	1.28	3.2	1.28	3.2	1.28	3.2	V
		Full range		3.8		3.7		3.6			
	I _{OL} = 50 mA	25°C	0.63	1	0.63	1	0.63	1	0.63	1	
		Full range		1.5		1.4		1.3			
	I _{OL} = 10 mA	25°C	0.12	0.3	0.12	0.3	0.12	0.3	0.12	0.3	
		Full range		0.45		0.4		0.4			
High-level output voltage	I _{OH} = -10 mA	25°C	12.5	14.2	12.5	14.2	12.5	14.2	12.5	14.2	V
		Full range	12.5		12.5		12.5				
	I _{OH} = -5 mA	25°C	13.5	14.6	13.5	14.6	13.5	14.6	13.5	14.6	
		Full range	13.5		13.5		13.5				
	I _{OH} = -1 mA	25°C	14.2	14.9	14.2	14.9	14.2	14.9	14.2	14.9	
		Full range	14.2		14.2		14.2				
Supply current	See Note 2	25°C	0.72	1.2	0.72	1.2	0.72	1.2	0.72	1.2	mA
		Full range		2		1.8		1.6			

†Full range (MIN to MAX) is -55°C to 125°C for TLC556M, -40°C to 85°C for TLC556I, and 0°C to 70°C for TLC556C.

NOTE 2: These values apply for the expected operating configurations in which the Threshold terminal is connected directly to the Discharge terminal or to the Trigger terminal.

4

Special Functions

TLC556M, TLC556I, TLC556C DUAL LinCMOS™ TIMERS

electrical characteristics at specified free-air temperature, VDD = 18 V

PARAMETER	TEST CONDITIONS†	TLC556M			TLC556I			TLC556C			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX		
Threshold voltage level	25°C	11.4	12	12.6	11.4	12	12.6	11.4	12	12.6	V	
	Full range	10.9		12.7	10.9		12.7	10.9		12.7		
Threshold current	25°C	10			10			10			pA	
	MAX	5000			150			75				
Trigger voltage level	25°C	5.6	6	6.4	5.6	6	6.4	5.6	6	6.4	V	
	Full range	5.5		6.5	5.5		6.5	5.5		6.5		
Trigger current	25°C	10			10			10			pA	
	MAX	5000			150			75				
Reset voltage level	25°C	0.4	1.1	1.5	0.4	1.1	1.5	0.4	1.1	1.5	V	
	Full range	0.3		1.8	0.3		1.8	0.3		1.8		
Reset current	25°C	10			10			10			pA	
	MAX	5000			150			75				
Control voltage (open-circuit) as a percentage of supply voltage	MAX	66.7%			66.7%			66.7%				
Discharge switch on-state voltage	I _{OL} = 100 mA	25°C	0.73	1.5	0.73	1.5	0.73	1.5		V		
		Full range		1.6		1.6		1.6				
Discharge switch off-state current		25°C	0.1		0.1		0.1			nA		
		MAX	120		2		0.5					
Low-level output voltage	I _{OL} = 3.2 mA	25°C	0.04	0.3	0.04	0.3	0.04	0.3		V		
		Full range		0.4		0.35		0.35				
High-level output voltage	I _{OH} = -1 mA	25°C	17.3	17.9	17.3	17.9	17.3	17.9		V		
		Full range	17.3		17.3		17.3					
Supply current	See Note 2	25°C	1.2			1.2			1.2			mA
		Full range	2			1.8			1.6			

†Full range (MIN to MAX) is -55°C to 125°C for TLC556M, -40°C to 85°C for TLC556I, and 0°C to 70°C for TLC556C.

NOTE 2: These values apply for the expected operating configuration in which the Threshold terminal is connected directly to the Discharge terminal or to the Trigger terminal.

4

Special Functions

TLC556M, TLC556I, TLC556C

DUAL LinCMOS™ TIMERS

operating characteristics, $V_{DD} = 5\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Initial error of timing interval†	$V_{DD} = 5\text{ V to }15\text{ V}$,		1%	3%	
Supply voltage sensitivity of timing interval	$R_A = R_B = 1\text{ k}\Omega$ to $100\text{ k}\Omega$, $C_T = 0.1\ \mu\text{F}$, See Note 3		0.1	0.5	%/V
Output pulse rise time	$R_L = 10\text{ M}\Omega$, $C_L = 10\text{ pF}$		20	75	ns
Output pulse fall time			15	60	
Maximum frequency in astable mode	$R_A = 470\ \Omega$, $R_B = 200\ \Omega$, $C_T = 200\ \text{pF}$, See Note 3	1.2	2.1		MHz

† Timing interval error is defined as the difference between the measured value and the average value of a random sample from each process run.

NOTE 3: R_A , R_B , and C_T are as defined in Figure 3.

TYPICAL CHARACTERISTICS

DISCHARGE SWITCH ON-STATE RESISTANCE
vs
FREE-AIR TEMPERATURE

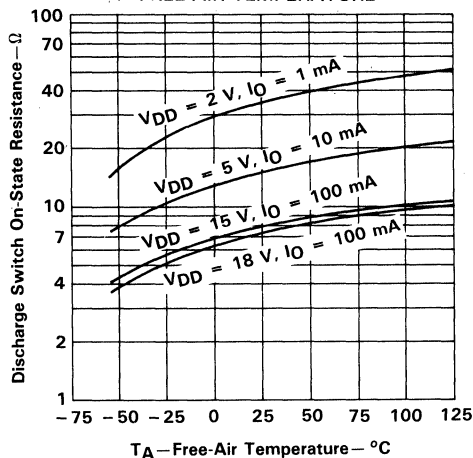
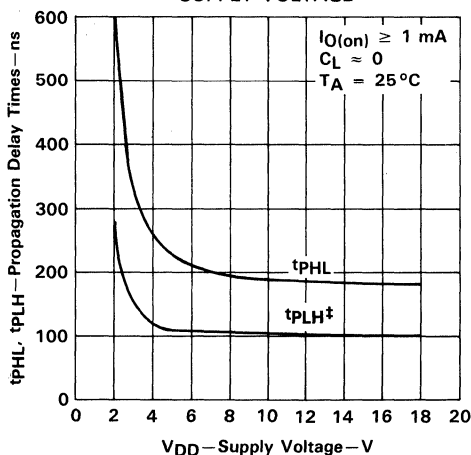


FIGURE 1

PROPAGATION DELAY TIMES
TO DISCHARGE OUTPUT
FROM TRIGGER AND THRESHOLD
SHORTED TOGETHER
vs
SUPPLY VOLTAGE



‡ The effects of the load resistance on these values must be taken into account separately.

FIGURE 2

TYPICAL APPLICATION DATA

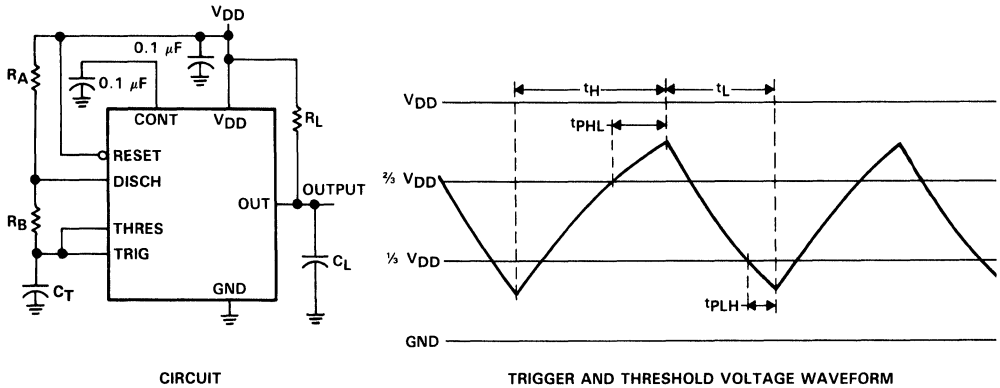


FIGURE 3. ASTABLE OPERATION

Connecting the trigger input to the threshold input, as shown in Figure 3, causes the timer to run as a multivibrator. The capacitor C_T charges through R_A and R_B to the trigger voltage level (approximately $0.67V_{DD}$) and then discharges through R_B only to the value of the threshold voltage level (approximately $0.33V_{DD}$). The output is high during the charging cycle (t_H) and low during the discharge cycle (t_L). The duty cycle is controlled by the values of R_A , R_B , and C_T , as shown in the equations below.

$$t_H \approx C_T (R_A + R_B) \ln 2 \quad (\ln 2 = 0.693)$$

$$t_L \approx C_T R_B \ln 2$$

$$\text{Period} = t_H + t_L \approx C_T (R_A + 2R_B) \ln 2$$

$$\text{Output driver duty cycle} = \frac{t_L}{t_H + t_L} \approx 1 - \frac{R_B}{R_A + 2R_B}$$

$$\text{Output waveform duty cycle} = \frac{t_H}{t_H + t_L} \approx \frac{R_B}{R_A + 2R_B}$$

The $0.1\text{-}\mu\text{F}$ capacitor at the control pin in Figure 3 decreases the period by about 10%.

The formulas shown above do not allow for any propagation delay from the trigger and threshold inputs to the discharge output. These delay times add directly to the period and create differences between calculated and actual values that increase with frequency. In addition, the discharge output resistance r_{on} adds to R_B to provide another source of error in the calculation when R_B is very low or r_{on} is very high.

The equations below provide better agreement with measured values.

$$t_H = C_T (R_A + R_B) \ln \left[3 - \exp \left(\frac{-t_{PLH}}{C_T (R_B + r_{on})} \right) \right] + t_{PLH}$$

$$t_L = C_T (R_B + r_{on}) \ln \left[3 - \exp \left(\frac{-t_{PLH}}{C_T (R_A + R_B)} \right) \right] + t_{PLH}$$

TLC556M, TLC556I, TLC556C DUAL LinCMOS™ TIMERS

The preceding equations and those given earlier are similar in that a time constant is multiplied by the logarithm of a number or function. The limit values of the logarithmic terms must be between $\ln 2$ at low frequencies and $\ln 3$ at extremely high frequencies. For a duty cycle close to 50%, an appropriate constant for the logarithmic terms can be substituted with good results. Duty cycles less than 50% $\frac{t_H}{t_H + t_L}$ will require that $\frac{t_H}{t_L} < 1$ and possibly $R_A \leq r_{on}$. These conditions can be difficult to obtain.

In monostable applications, the trip point of the trigger input can be set by a voltage applied to the control pin. An input voltage between 10% and 80% of the supply voltage from a resistor divider with at least 500 μA bias provides good results.

4

Special Functions

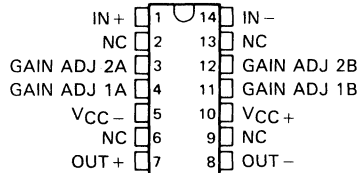


μA733M, μA733C DIFFERENTIAL VIDEO AMPLIFIERS

D922, NOVEMBER 1970—REVISED APRIL 1988

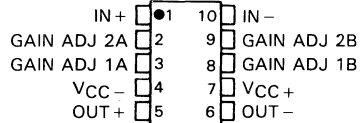
- 200-MHz Bandwidth
- 250-kΩ Input Resistance
- Selectable Nominal Amplification of 10, 100, or 400
- No Frequency Compensation Required
- Designed to be Interchangeable with Fairchild μA733M and μA733C

μA733M . . . J DUAL-IN-LINE PACKAGE
μA733C . . . D OR N PACKAGE
(TOP VIEW)



NC—No internal connection

UA733M . . . U FLAT PACKAGE
(TOP VIEW)



description

The μA733 is a monolithic two-stage video amplifier with differential inputs and differential outputs.

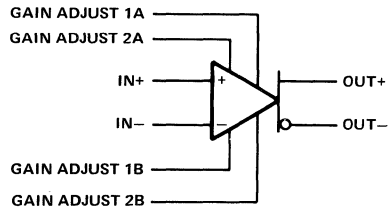
Internal series-shunt feedback provides wide bandwidth, low phase distortion, and excellent gain stability. Emitter-follower outputs enable the device to drive capacitive loads and all stages are current-source biased to obtain high common-mode and supply-voltage rejection ratios.

Fixed differential amplification of 10, 100, or 400 may be selected without external components, or amplification may be adjusted from 10 to 400 by the use of a single external resistor connected between 1A and 1B. No external frequency-compensating components are required for any gain option.

The device is particularly useful in magnetic-tape or disc-file systems using phase or NRZ encoding and in high-speed thin-film or plated-wire memories. Other applications include general purpose video and pulse amplifiers where wide bandwidth, low phase shift, and excellent gain stability are required.

The μA733M is characterized for operation over the full military temperature range of -55°C to 125°C ; the μA733C is characterized for operation from 0°C to 70°C .

symbol



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	μA733M	UA733C	UNIT
Supply voltage V_{CC+} (See Note 1)	8	8	V
Supply voltage V_{CC-} (See Note 1)	-8	-8	V
Differential input voltage	± 5	± 5	V
Common-mode input voltage	± 6	± 6	V
Output current	10	10	mA
Continuous total power dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-55 to 125	0 to 70	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J or U package	300	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or N package	260	$^{\circ}\text{C}$

NOTE 1. All voltage values, except differential input voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .

PRODUCTION DATA documents contain information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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Special Functions

uA733M, uA733C

DIFFERENTIAL VIDEO AMPLIFIERS

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C POWER RATING	DERATING FACTOR	DERATE ABOVE T _A	T _A = 70°C POWER RATING	T _A = 125°C POWER RATING
D	500 mW	N/A	N/A	500 mW	N/A
J (uA733M)	500 mW	11.0 mW/°C	104°C	500 mW	275 mW
N	500 mW	N/A	N/A	500 mW	N/A
U	500 mW	5.4 mW/°C	57°C	432 mW	135 mW

electrical characteristics, V_{CC+} = 6 V, V_{CC-} = -6 V, T_A = 25°C

PARAMETER	TEST FIGURE	TEST CONDITIONS	GAIN OPTION†	uA733M			uA733C			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
A _{VD} Large-signal differential voltage amplification	1	V _{OD} = 1 V	1	300	400	500	250	400	600	V/V
			2	90	100	110	80	100	120	
			3	9	10	11	8	10	12	
BW Bandwidth	2	R _S = 50 Ω	1	50			50			MHz
			2	90			90			
			3	200			200			
I _{IO} Input offset current			Any	0.4	3	0.4	5	μA		
I _{IB} Input bias current			Any	9	20	9	30	μA		
V _{ICR} Common-mode input voltage range	1		Any	±1			±1			V
V _{OC} Common-mode output voltage	1		Any	2.4	2.9	3.4	2.4	2.9	3.4	V
V _{OO} Output offset voltage	1		1	0.6	1.5	0.6	1.5			V
			2 & 3	0.35	1	0.35	1.5			
V _{OPP} Maximum peak-to-peak output voltage swing	1		Any	3	4.7	3	4.7		V	
r _i Input resistance	3	V _{OD} ≤ 1 V	1	4			4			kΩ
			2	20	24	10	24			
			3	250			250			
r _o Output resistance				20		20		Ω		
C _i Input capacitance	3	V _{OD} ≤ 1 V	2	2			2			pF
CMRR Common-mode rejection ratio	4	V _{IC} = ±1 V, f ≤ 100 kHz	2	60	86	60	86			dB
		V _{IC} = ±1 V, f = 5 MHz	2	70			70			
k _{SVR} Supply voltage rejection ratio (ΔV _{CC} /ΔV _{IO})	1	ΔV _{CC+} = ±0.5 V, ΔV _{CC-} = ±0.5 V	2	50	70	50	70		dB	
V _n Broadband equivalent input noise voltage	5	BW = 1 kHz to 10 MHz	Any	12			12			μV
t _{pd} Propagation delay time	2	R _S = 50 Ω, Output voltage step = 1 V	1	7.5			7.5			ns
			2	6.0	10	6.0	10			
			3	3.6			3.6			
t _r Rise time	2	R _S = 50 Ω, Output voltage step = 1 V	1	10.5			10.5			ns
			2	4.5	10	4.5	12			
			3	2.5			2.5			
I _{sink(max)} Maximum output sink current			Any	2.5	3.6	2.5	3.6		mA	
I _{CC} Supply current		No load, No signal	Any	16	24	16	24		mA	

† The gain option is selected as follows:

Gain Option 1 . . . Gain-adjust pin 1A is connected to pin 1B, and pins 2A and 2B are open.

Gain Option 2 . . . Gain-adjust pin 1A and pin 1B are open, pin 2A is connected to pin 2B.

Gain Option 3 . . . All four gain-adjust pins are open.

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Special Functions

uA733M, uA733C DIFFERENTIAL VIDEO AMPLIFIERS

electrical characteristics, $V_{CC+} = 6\text{ V}$, $V_{CC-} = -6\text{ V}$, $T_A = -55^\circ\text{C}$ to 125°C for uA733M, 0°C to 70°C for uA733C

PARAMETER	TEST FIGURE	TEST CONDITIONS	GAIN OPTION†	uA733M		uA733C		UNIT
				MIN	MAX	MIN	MAX	
A _{VD} Large-signal differential voltage amplification	1	V _{OD} = 1 V	1	200	600	250	600	V/V
			2	80	120	80	120	
			3	8	12	8	12	
I _O Input offset current			Any		5		6	μA
I _B Input bias current			Any		40		40	μA
V _{ICR} Common-mode input voltage range	1		Any	±1		±1		V
V _{OO} Output offset voltage	1		1		1.5		1.5	V
			2 & 3		1.2		1.5	
V _{OPP} Maximum peak-to-peak output voltage swing	1		Any	2.5		2.8		V
r _i Input resistance	3	V _{OD} ≤ 1 V	2	8		8		kΩ
CMRR Common-mode rejection ratio	4	V _{IC} = ±1 V, f ≤ 100 kHz	2	50		50		dB
			2					
k _{SVR} Supply voltage rejection ratio (ΔV _{CC} /ΔV _{IO})	1	ΔV _{CC+} = ±0.5 V, ΔV _{CC-} = ±0.5 V	2	50		50		dB
I _{sink(max)} Maximum output sink current			Any	2.2		2.5		mA
I _{CC} Supply current		No load, No signal	Any		27		27	mA

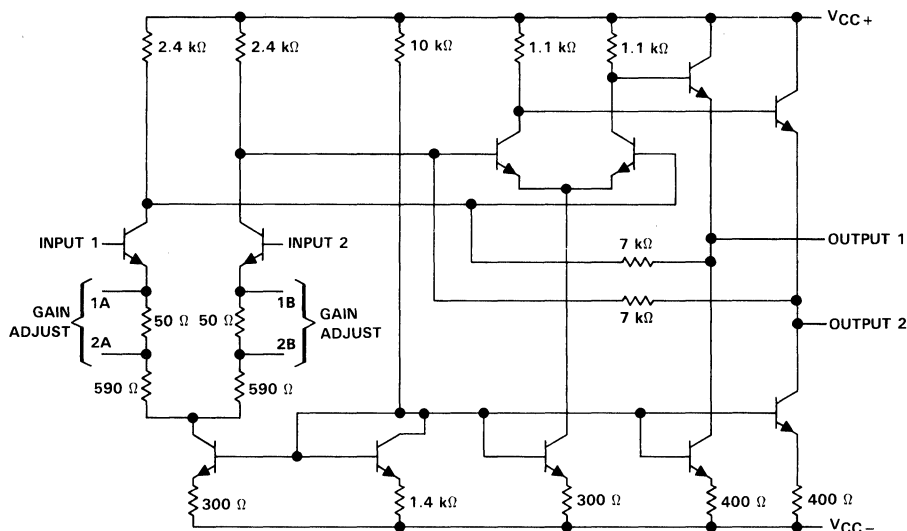
†The gain option is selected as follows:

Gain Option 1 . . . Gain-adjust pin 1A is connected to pin 1B, and pins 2A and 2B are open.

Gain Option 2 . . . Gain-adjust pin 1A and pin 1B are open, pin 2A is connected to pin 2B.

Gain Option 3 . . . All four gain-adjust pins are open.

schematic



Component values shown are nominal.

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Special Functions

DEFINITION OF TERMS

Large-Signal Differential Voltage Amplification (A_{VD}) The ratio of the change in voltage between the output terminals to the change in voltage between the input terminals producing it.

Bandwidth (BW) The range of frequencies within which the differential gain of the amplifier is not more than 3 dB below its low-frequency value.

Input Offset Current (I_{IO}) The difference between the currents into the two input terminals with the inputs grounded.

Input Bias Current (I_{IB}) The average of the currents into the two input terminals with the inputs grounded.

Input Voltage Range (V_I) The range of voltage that if exceeded at either input terminal will cause the amplifier to cease functioning properly.

Common-Mode Output Voltage (V_{OC}) The average of the d-c voltages at the two output terminals.

Output Offset Voltage (V_{OO}) The difference between the d-c voltages at the two output terminals when the input terminals are grounded.

Maximum Peak-to-Peak Output Voltage Swing (V_{OPP}) The maximum peak-to-peak output voltage swing that can be obtained without clipping. This includes the unbalance caused by output offset voltage.

Input Resistance (r_i) The resistance between the input terminals with either input grounded.

Output Resistance (r_o) The resistance between either output terminal and ground.

Input Capacitance (C_i) The capacitance between the input terminals with either input grounded.

Common-Mode Rejection Ratio (CMRR) The ratio of differential voltage amplification to common-mode voltage amplification. This is measured by determining the ratio of a change in input common-mode voltage to the resulting change in input offset voltage.

Supply Voltage Rejection Ratio (k_{SVR}) The absolute value of the ratio of the change in power supply voltages to the change in input offset voltage. For these devices, both supply voltages are varied symmetrically.

Equivalent Input Noise Voltage (V_n) The voltage of an ideal voltage source (having an internal impedance equal to zero) in series with the input terminals of the device that represents the part of the internally generated noise that can properly be represented by a voltage source.

Propagation Delay Time (t_{pd}) The interval between the application of an input voltage step and its arrival at either output, measured at 50% of the final value.

Rise Time (t_r) The time required for an output voltage step to change from 10% to 90% of its final value.

Maximum Output Sink Current ($I_{sink(max)}$) The maximum available current into either output terminal when that output is at its most negative potential.

Supply Current (I_{CC}) The average of the magnitudes of the two supply currents I_{CC1} and I_{CC2} .

PARAMETER MEASUREMENT INFORMATION

test circuits

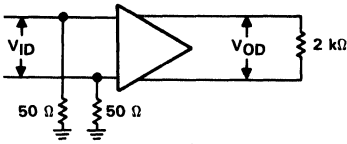


FIGURE 1

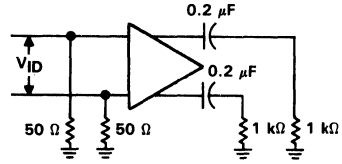


FIGURE 2

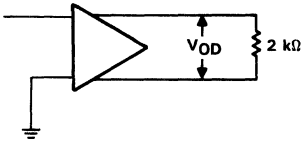


FIGURE 3

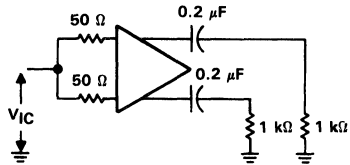


FIGURE 4

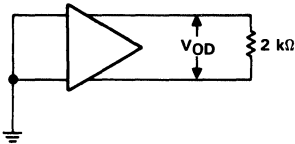
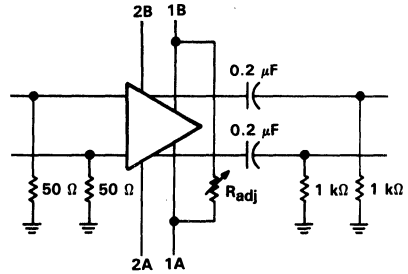


FIGURE 5



VOLTAGE AMPLIFICATION ADJUSTMENT

FIGURE 6

uA733M, uA733C
DIFFERENTIAL VIDEO AMPLIFIERS

TYPICAL CHARACTERISTICS

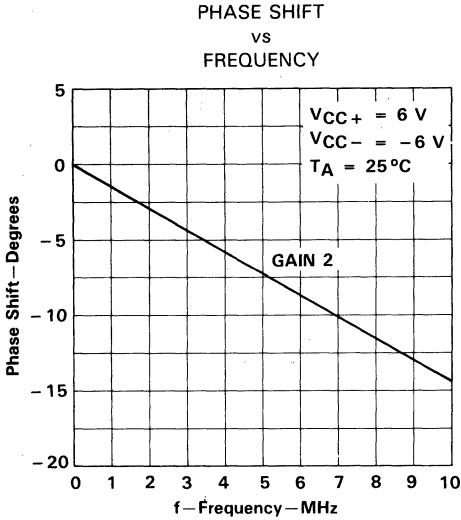


FIGURE 7

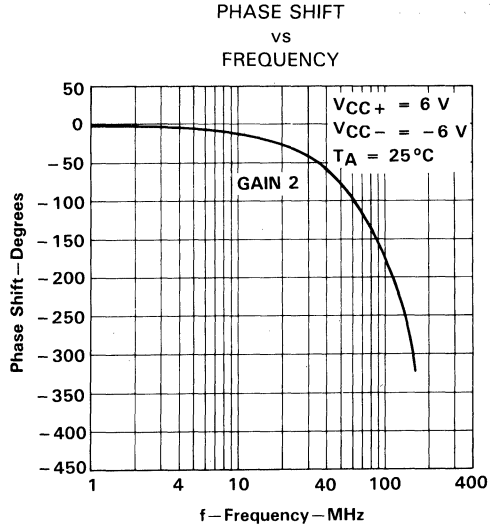


FIGURE 8

VOLTAGE AMPLIFICATION
(SINGLE-ENDED OR DIFFERENTIAL)
vs
TEMPERATURE

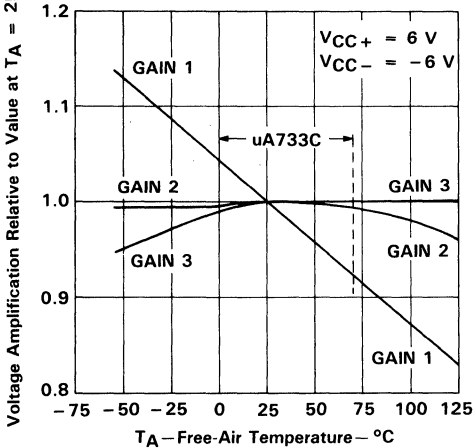


FIGURE 9

VOLTAGE AMPLIFICATION
(SINGLE-ENDED OR DIFFERENTIAL)
vs
SUPPLY VOLTAGE

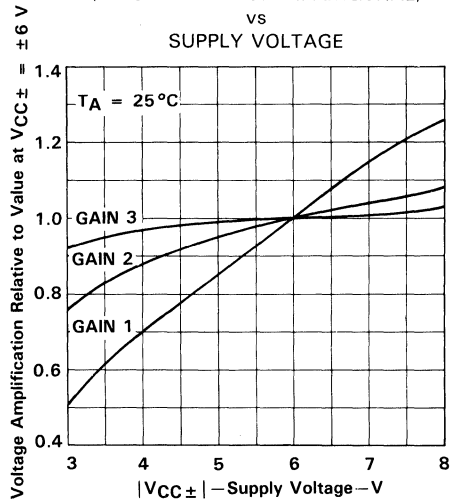


FIGURE 10

4

Special Functions

TYPICAL CHARACTERISTICS

DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
RESISTANCE BETWEEN G1A AND G1B

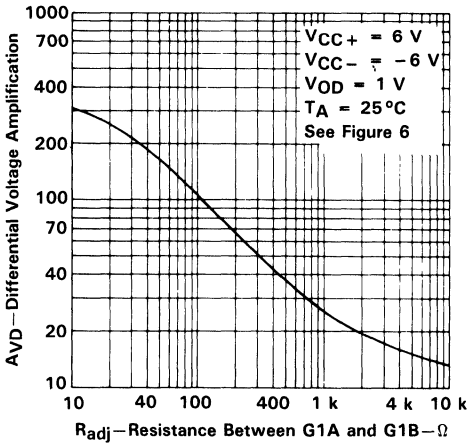


FIGURE 11

SINGLE-ENDED VOLTAGE AMPLIFICATION
vs
FREQUENCY

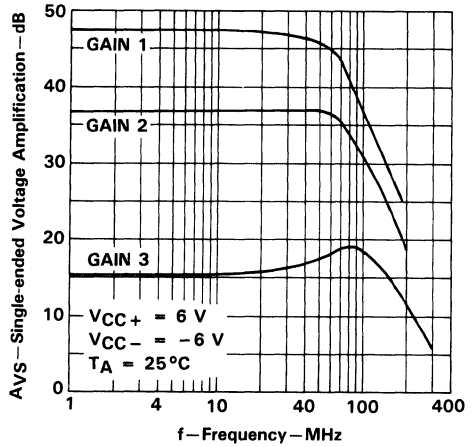


FIGURE 12

SUPPLY CURRENT
vs
FREE-AIR TEMPERATURE

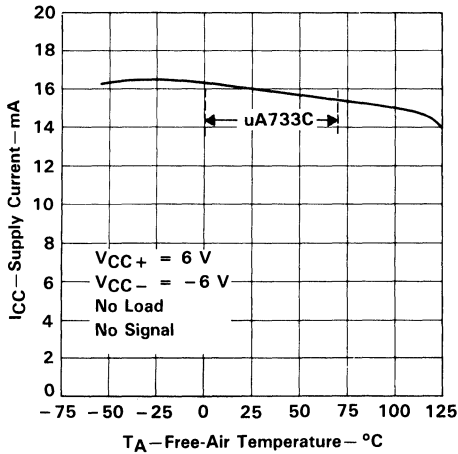


FIGURE 13

SUPPLY CURRENT
vs
SUPPLY VOLTAGE

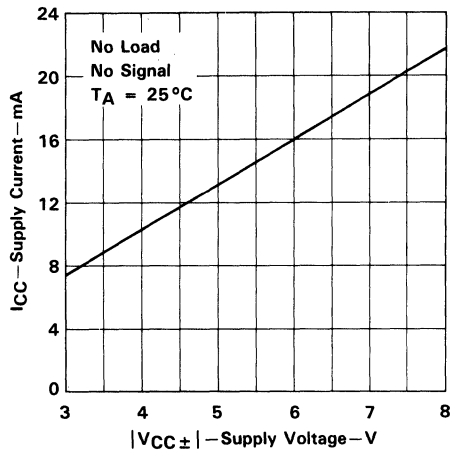


FIGURE 14

uA733M, uA733C
DIFFERENTIAL VIDEO AMPLIFIERS

TYPICAL CHARACTERISTICS

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE

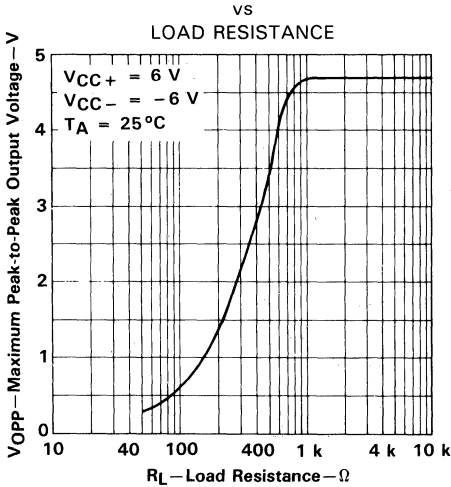


FIGURE 15

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE

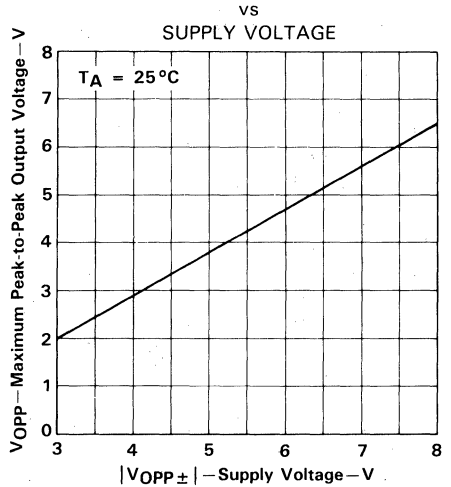


FIGURE 16

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE

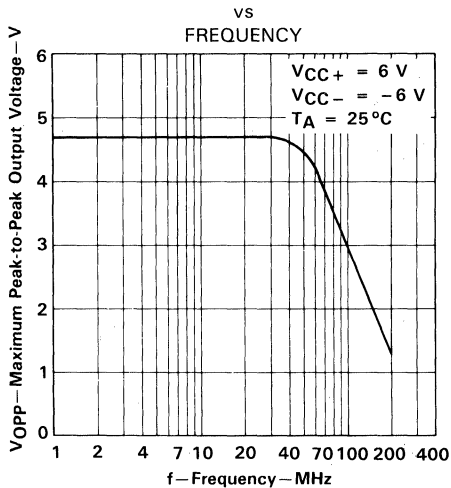


FIGURE 17

INPUT RESISTANCE

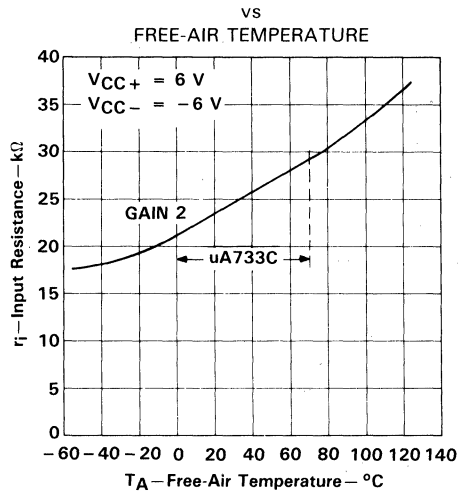


FIGURE 18

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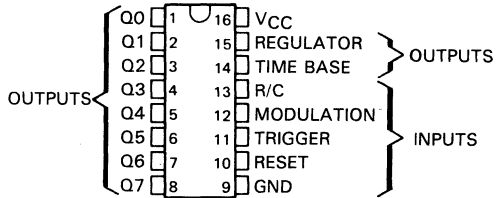
Special Functions

uA2240C PROGRAMMABLE TIMER/COUNTER

D2442, JUNE 1978 — REVISED MAY 1988

- Accurate Timing from Microseconds to Days
- Programmable Delays from 1 Time Constant to 255 Time Constants
- Outputs Compatible with TTL and CMOS
- Wide Supply-Voltage Range
- External Sync and Modulation Capability

N DUAL-IN-LINE PACKAGE
(TOP VIEW)



description

These circuits consist of a time-base oscillator, and eight-bit counter, a control flip-flop, and a voltage regulator. The frequency of the time-base oscillator is set by the time constant of an external resistor and capacitor at pin 13 and can be synchronized or modulated by signals applied to the modulation input. The output of the time-base section is applied directly to the input of the counter section and also appears at pin 14 (time base). The time-base pin may be used to monitor the frequency of the oscillator, to provide an output pulse to other circuitry, or (with the time-base section disabled) to drive the counter input from an external source. The counter input is activated on a negative-going transition. The reset input stops the time-base oscillator and sets each binary output, Q0 through Q7, and the time-base output to a TTL high level. After resetting, the trigger input starts the oscillator and all Q outputs go low. Once triggered, the uA2240C will ignore any signals at the trigger input until it is reset.

The uA2240C timer/counter may be operated in the free-running mode or with output-signal feedback to the reset input for automatic reset. Two or more binary outputs may be connected together to generate complex pulse patterns, or each output may be used separately to provide eight output frequencies. Using two circuits in cascade can provide precise time delays of up to three years.

The uA2240C is characterized for operation from 0°C to 70°C.

AVAILABLE OPTIONS

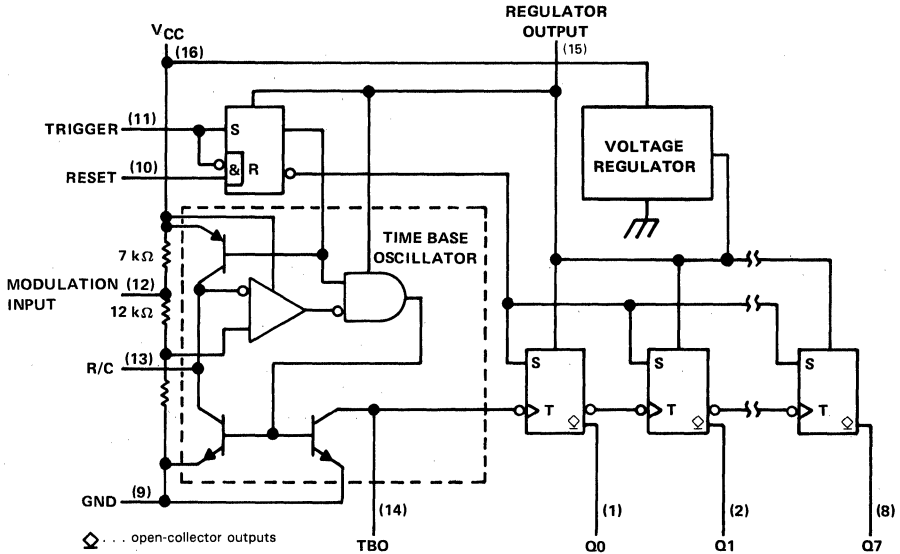
SYMBOLIZATION		OPERATING TEMPERATURE RANGE	V _T MAX at 25°C
DEVICE	PACKAGE SUFFIX		
uA2240C	N	0°C to 70°C	2 V

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Special Functions

uA2240C PROGRAMMABLE TIMER/COUNTER

functional block diagram



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC} (see Note 1)	18 V
Output voltage: Q0 thru Q7	18 V
Output current: Q0 thru Q7	10 mA
Regulator output current	-5 mA
Continuous dissipation at (or below) 25°C free-air temperature	650 mW
Operating free-air temperature range	0°C to 70°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: Voltage values are with respect to the network ground terminal.

recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{CC} (see Note 2)	4		14	V
Timing resistor	0.001		10	MΩ
Timing capacitor	0.01		1000	μF
Counter input frequency (Pin 14)		1.5		MHz
Pull-up resistor, time-base output		20		kΩ
Trigger and reset input pulse voltage	2	3		V
Trigger and reset input pulse duration	2			μs
External clock input pulse voltage	3			V
External clock input pulse duration	1			μs

NOTE 2: For operation with $V_{CC} \leq 4.5$ V, short regulator output to V_{CC} .

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Special Functions

electrical characteristics at 25°C free-air temperature

PARAMETER	TEST CIRCUIT	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Regulator output voltage	1	V _{CC} = 5 V, Trigger and reset open or grounded	3.9	4.4		V
	2	V _{CC} = 15 V, Trigger and reset open or grounded	5.8	6.3	6.8	
Modulation input open circuit voltage	1	V _{CC} = 5 V, Trigger and reset open or grounded	2.8	3.5	4.2	V
		V _{CC} = 15 V, Trigger and reset open or grounded		10.5		
Trigger threshold voltage	1	V _{CC} = 5 V, Reset at 0 V		1.4	2	V
High-level trigger current	1	V _{CC} = 5 V, Trigger at 2 V, Reset at 0 V		10		μA
Reset threshold voltage	1	V _{CC} = 5 V, Trigger at 0 V		1.4	2	V
High-level reset current	1	V _{CC} = 5 V, Trigger at 0 V		10		μA
Counter input (time base) threshold voltage	2	V _{CC} = 5 V, Trigger and reset open or grounded	1	1.4		V
Low-level output current, Q0 thru Q7	2	V _{CC} = 5 V, Trigger at 2 V, Reset at 0 V, V _{OL} < 0.4 V	2	4		mA
High-level output current, Q0 thru Q7	2	V _{OH} = 15 V, Reset at 2 V, Trigger at 0 V		0.01	15	μA
Supply current	1	V _{CC} = 5 V, Trigger at 0 V, Reset at 5 V		4	7	mA
	1	V _{CC} = 15 V, Trigger at 0 V, Reset at 5 V		13	18	
	3	V ₊ = 4 V		1.5		

operating characteristics at 25°C free-air temperature (unless otherwise noted)

PARAMETER	TEST CIRCUIT	TEST CONDITIONS†	MIN	TYP	MAX	UNIT
Initial error of time base‡	1	V _{CC} = 5 V, Trigger at 5 V, Reset at 0 V		±0.5	±5	%
Temperature coefficient of time-base period	1	T _A = 0°C to 70°C	V _{CC} = 5 V	-200		ppm/°C
			V _{CC} = 15 V	-80		
Supply voltage sensitivity of time-base period	1	V _{CC} ≥ 8 V		-0.08	-0.3	%/V
Time-base output frequency	1	V _{CC} = 5 V, R = MIN, C = MIN		130		kHz
Propagation delay time		see Note 3	From trigger input		1	μs
			From reset input		0.8	
Output rise time	2	R _L = 3 kΩ, C _L = 10 pF	Q0 thru Q7		180	ns
Output fall time					180	

†For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

‡This is the time-base period error due only to the uA2240C and expressed as a percentage of nominal (1.00 RC).

NOTE 3: Propagation delay time is measured from the 50% point on the leading edge of an input pulse to the 50% point on the leading edge of the resulting change of state at Q0.

**uA2240C
PROGRAMMABLE TIMER/COUNTER**

PARAMETER MEASUREMENT INFORMATION

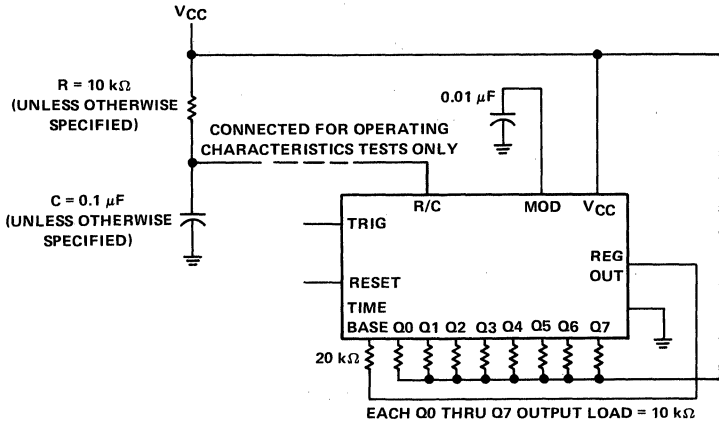


FIGURE 1. GENERAL TEST CIRCUIT

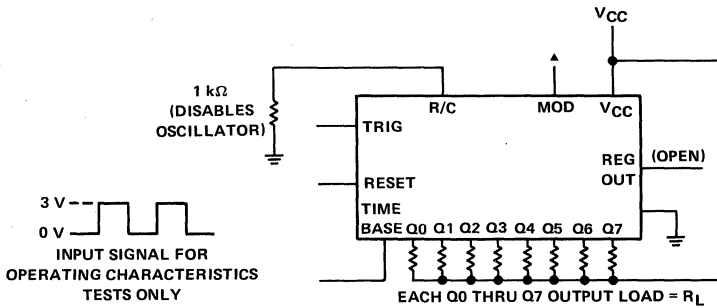


FIGURE 2. COUNTER TEST CIRCUIT

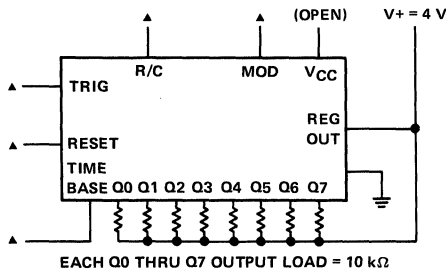


FIGURE 3. REDUCED-POWER TEST CIRCUIT (TIME BASE DISABLED)

▲These connections may be open or grounded for this test.

4

Special Functions

TYPICAL CHARACTERISTICS

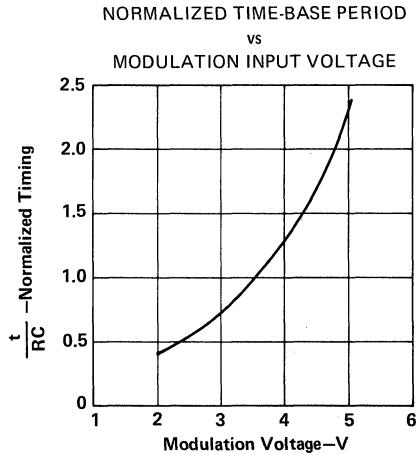
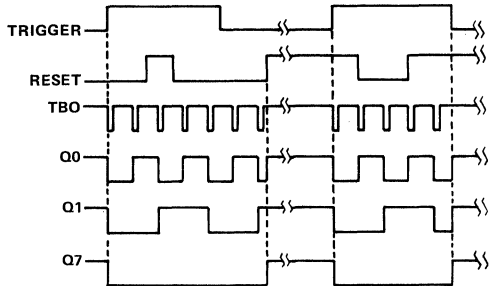


FIGURE 4

TYPICAL APPLICATION INFORMATION

Figure 5 shows voltage waveforms for typical operation of the uA2240C. If both reset and trigger inputs are low during power-up, the timer/counter will be in a reset state with all binary (Q) outputs high and the oscillator stopped. In this state, a high level on the trigger input starts the time-base oscillator. The initial negative-going pulse from the oscillator sets the Q outputs to low logic levels at the beginning of the first time-base period. The uA2240C will ignore any further signals at the trigger input until after a reset signal is applied to the reset input. With the trigger input low, a high level at the reset input will set Q outputs high and stop the time-base oscillator. If the reset signal occurs while the trigger input is high, the reset is ignored. If the reset input remains high when the trigger input goes low, the uA2240C will reset.



**FIGURE 5. TIMING DIAGRAM OF
OUTPUT WAVEFORMS**

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Special Functions

**uA2240C
PROGRAMMABLE TIMER/COUNTER**

TYPICAL APPLICATION INFORMATION

In monostable applications of the uA2240C, one or more of the binary outputs will be connected to the reset terminal as shown in Figure 6. The binary outputs are open-collector stages that can be connected together to a common pull-up resistor to provide a "wired-OR" function. The combined output will be low as long as any one of the outputs is low. This type of arrangement can be used for time delays that are integer multiples of the time-base period. For example, if Q5 ($2^5 = 32$) only is connected to the reset input, every trigger pulse will generate a 32-period active-low output. Similarly, if Q0, Q4, and Q5 are connected to reset, each trigger pulse creates a 49-period delay.

In astable operation, the uA2240C will free-run from the time it is triggered until it receives an external reset signal.

The period of the time-base oscillator is equal to the RC time constant of an external resistor and capacitor connected as shown in Figure 6 when the modulation input is open (approximately 3.5 V internal, see Figure 4). Under conditions of high supply voltage ($V_{CC} > 7$ V) and low value of timing capacitor ($C < 0.1 \mu\text{F}$), the pulse duration of the time-base oscillator may be too short to properly trigger the counters. This situation can be corrected by adding a 300-pF capacitor between the time-base output and ground. The time-base output (TBO) is an open-collector output that requires a 20-k Ω pull-up resistor to Pin 15 for proper operation. The time-base pin may also be used as an input to the counters for an external time-base or as an active-low inhibit input to interrupt counting without resetting.

The modulation input varies the ratio of the time-base period to the RC time constant as a function of the dc bias voltage (see Figure 4). It can also be used to synchronize the timer/counter to an external clock or sync signal.

The regulator output is used internally to drive the binary counters and the control logic. This terminal can also be used to supply voltage to additional uA2240C devices to minimize power dissipation when several timer circuits are cascaded. For circuit operation with an external clock, the regulator output can be used as the V_{CC} input terminal to power down the internal time base and reduce power dissipation. When supply voltages less than 4.5 V are used with the internal time base, Pin 15 should be shorted to Pin 16.

4

Special Functions

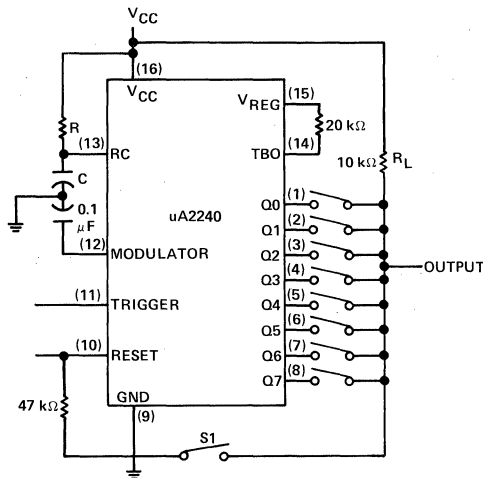


FIGURE 6. BASIC CONNECTIONS FOR TIMING APPLICATIONS

General Information

1

Operational Amplifiers

2

Voltage Comparators

3

Special Functions

4

Product Previews

5

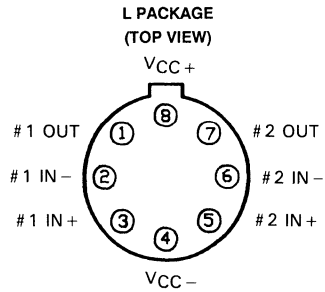
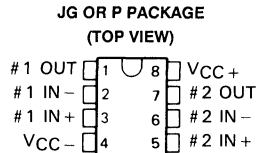
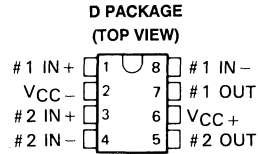
Mechanical Data

6

LT1013 DUAL PRECISION OP AMP

D3237, MAY 1988—REVISED FEBRUARY 1989

- **Single-Supply Operation:**
Input Voltage Range Extends to Ground
Output Swings to Ground While Sinking
Current
- **Input Offset Voltage . . . 150 μ A Max at 25°C**
for LT1013AM, LT1013AC
- **Offset Voltage Temperature Coefficient . . .**
2 μ V/°C Max for LT1013AM, LT1013AC
- **Input Offset Current . . . 0.8 nA Max at 25°C**
for LT1013AM, LT1013AC
- **High Gain . . . 1.5 V/ μ V Min ($R_L = 600 \Omega$),**
0.8 V/ μ V Min ($R_L = 2 \text{ k}\Omega$) for LT1013AM,
LT1013AC
- **Low Supply Current . . . 0.5 mA Max at 25°C**
for LT1013AM, LT1013AC
- **Low Peak-To-Peak Noise Voltage . . .**
0.55 μ V Typ
- **Low Current Noise . . . 0.07 pA/ $\sqrt{\text{Hz}}$ Typ**



Pin 4 (L package) is in electrical contact with the case.

description

The LT1013 is a precision dual operational amplifier with an 8-pin industry-standard configuration. It features low offset voltage temperature coefficient, high gain, low supply current, and low noise.

The LT1013 can be operated from a single 5-V power supply; the common-mode input voltage range includes ground, and the output can also swing to within a few millivolts of ground. Crossover distortion, so apparent in previous single-supply designs, is eliminated. The LT1013 is fully specified for both dual ± 15 -V and single 5-V supplies.

The LT1013AM and LT1013M are characterized for operation over the full military temperature range of -55°C to 125°C . The LT1013AC, LT1013C, and LT1013D are characterized for operation from 0°C to 70°C .

AVAILABLE OPTIONS

T _A	V _{IQ} MAX at 25°C	PACKAGE			
		SMALL OUTLINE (D)	CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)
0°C to 70°C	150 μ V		LT1013ACJG	LT1013ACL	
	300 μ V		LT1013CJG	LT1013CL	LT1013CP
-55°C to 125°C	800 μ V	LT1013DD			LT1013DP
	150 μ V		LT1013AMJG	LT1013AML	
	300 μ V		LT1013MJG	LT1013ML	

The D package is available taped and reeled. Add the suffix R to the device type (e.g., LT1013DDR).

Product Previews

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LT1013 DUAL PRECISION OP AMP

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1)	22 V
Supply voltage, V_{CC-}	-22 V
Differential input voltage	± 30 V
Input voltage range, V_I	$V_{CC-} - 5$ V to V_{CC+}
Duration of output short-circuit at (or below) 25°C	unlimited
Operating free-air temperature range: LT1013AM, LT1013M	-55°C to 125°C
LT1013AC, LT1013C, LT1013D	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: JG or L package	300°C

NOTE 1: All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .

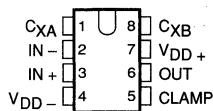


LTC1052, LTC7652 CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

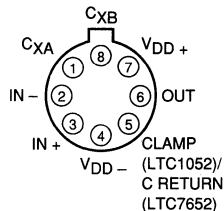
D3238, MAY 1988 – REVISED MARCH 1989

- Input Offset Voltage . . . 5 μV Max at 25°C
- Temperature Coefficient of Input Offset Voltage . . . 0.01 $\mu\text{V}/^\circ\text{C}$ Typ
- Long-Term Drift of Input Offset Voltage . . . 100 nV/mo Typ
- Maximum Input Bias Current . . . 30 pA at 25°C
- Minimum Differential Voltage Amplification Over Full Temperature Range . . . 120 dB
- Minimum Common-Mode Rejection Ratio Over Full Temperature Range . . . 120 dB
- Minimum Supply Voltage Rejection Ratio Over Full Temperature Range . . . 120 dB
- Single-Supply Operation from 4.75 V to 16 V (Input Voltage Range Extends to Ground)
- External Capacitors Can Be Returned to $V_{\text{DD-}}$ with No Noise Degradation

LTC1022 . . . JG OR P PACKAGE
(TOP VIEW)



LTC1052, LTC7652 . . . L PACKAGE
(TOP VIEW)



Pin 4 (L package) is in electrical contact with the case.

description

The LTC1052 and LTC7652 are low-noise chopper-stabilized operational amplifiers manufactured using CMOS silicon-gate technology. The devices are well-suited for applications such as thermocouple amplifiers, strain-gauge amplifiers, low-level signal processing, and medical instrumentation.

Chopper stabilization constantly corrects input offset voltage errors, including both errors in the initial input offset voltage and errors in input offset voltage due to time, temperature, and common-mode input voltage. The chopper circuitry is internal and completely transparent to the user. Only two external capacitors are required to alternately sample and hold the offset correction voltage and the amplified input signal.

Low-frequency (1/f) noise is also improved by the chopping technique. Instead of noise increasing continuously at a rate of 3 dB/octave, the internal chopping causes noise to decrease at low frequencies. Picoampere input currents further enhance the performance of these devices.

The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The C-suffix devices are characterized for operation from -40°C to 85°C .

AVAILABLE OPTIONS

T _A	PACKAGE		
	CERAMIC DIP (JG)	METAL CAN (L)	PLASTIC DIP (P)
-40°C to 85°C	LTC1052CJG	LTC1052CL LTC7652CL	LTC1052CP
-55°C to 125°C	LTC1052MJG	LTC1052ML	LTC1052MP

LTC1052, LTC7652 CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

electrical characteristics at specified free-air temperature, $V_{DD\pm} = \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A †	LTC1052M			LTC1052C, LTC7652C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.5 5			0.5 5			μV
α_{VIO} Temperature coefficient of input offset voltage		Full range	0.01 0.05			0.01 0.05			$\mu\text{V}/^\circ\text{C}$
Long-term drift of input offset voltage		25°C	100			100			nV/mo
I_{IO} Input offset current		25°C	5 30			5 30			pA
		Full range	2000			350			
I_{IB} Input bias current		25°C	1 30			1 30			pA
	Full range	1000			175				
V_{ICR} Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7			-5 to 2.7			V
V_{OM} Maximum peak output voltage swing	$R_L = 100\ \text{k}\Omega$, See Note 4	25°C	4.95			4.95			V
	$R_L = 10\ \text{k}\Omega$, See Note 4	Full range	4.7			4.7			
A_{VD} Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	120 150			120 150			dB
		Full range	120			120			
f_{ch} Internal chopping frequency		25°C	330			330			Hz
On-state clamp current	$R_L = 100\ \text{k}\Omega$	25°C	100			100			μA
		Full range	25			25			
Off-state clamp current	$V_O = -4\ \text{V to } 4\ \text{V}$	25°C	10 100			10 100			pA
		Full range	2			1			
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR}\ \text{min}, R_S = 50\ \Omega$	25°C	120 140			120 140			dB
		Full range	120			120			
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{CC\pm} / \Delta V_{IO}$)	$V_{CC\pm} = \pm 2.375\ \text{V to } \pm 8\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C	120 150			120 150			dB
		Full range	120			120			
I_{DD} Supply current	$V_O = 0, \text{No load}$	25°C	1.7 2			1.7 2			mA
		Full range	3			3			

†Full range is -55°C to 125°C for the LTC1052M and -40°C to 85°C for the LTC1052C and LTC7652C.

NOTE 4: Output clamp is not connected.

operating characteristics, $V_{DD\pm} = \pm 5\ \text{V}, T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slew rate	$R_L = 10\ \text{k}\Omega, C_L = 50\ \text{pF}$		4		V/ μs
V_{NPP} Peak-to-peak equivalent input noise voltage	$R_S = 100\ \Omega$ to 10 Hz		1.5		μV
	$R_S = 100\ \Omega$ to 1 Hz		0.5		
I_n Input noise current (see Note 5)	$f = 10\ \text{Hz}$		0.6		fA/ $\sqrt{\text{Hz}}$
GBP Gain bandwidth product			1.2		MHz

NOTE 5: Equivalent input noise current is calculated as follows: $I_n = (2q \times I_{IB})^{1/2}$, where $q = 1.6 \times 10^{-19}$.

TLE2021, TLE2022, TLE2024 Excaltbur HIGH-SPEED LOW-POWER PRECISION OPERATIONAL AMPLIFIERS

D3197, JANUARY 1989

- Excellent Input Offset Voltage Stability with Temperature . . . 2 $\mu\text{V}/^\circ\text{C}$ Typ
- Long-Term Drift of Input Offset Voltage . . . 0.005 $\mu\text{V}/\text{mo}$ Typ
- High Slew Rate . . . 0.9 V/ μs Typ
- High Unity-Gain Bandwidth . . . 2 MHz Typ
- Low Supply Current . . . 200 $\mu\text{A}/\text{Amplifier}$
- Phase-Reversal Protection
- Stable Supply Current with Temperature . . . 0.08 $\mu\text{A}/^\circ\text{C}$ Typ
- Full Electrical Parameters Specified at $V_{\text{CC}\pm} = \pm 15\text{ V}$ and $V_{\text{CC}} = 5\text{ V}$ to GND
- Common-Mode Input Voltage Range Includes the Negative Rail

description

The TLE2021, TLE2022, and TLE2024 are high-precision, high-speed, low-power operational amplifiers using Texas Instruments patent-pending Excaltbur process. Available in standard-pinout single, dual, and quad configurations, these devices offer improved slew rate and unity-gain bandwidth performance over the popular OP21, OP22I, and OP42I. A wide variety of packaging options is available, including small-outline (SO) and chip carrier versions for high-density systems applications.

The complementary bipolar Excaltbur process uses isolated vertical P-N-P transistors that yield dramatic improvement in gain-bandwidth product and slew rate compared to similar devices. The addition of a patent-pending bias circuit in conjunction with this process results in unsurpassed parameter stability with both time and temperature. This means that a "precision" device remains a precision device even with extreme changes in temperature and over years of use.

The combination of excellent dc performance with a common-mode input voltage range that includes the negative rail makes these devices an ideal choice for low-level signal conditioning applications in either single-supply or split-supply configurations. Additionally, these devices offer phase-reversal protection circuitry that eliminates an unexpected change in output states when one of the inputs goes below the negative supply rail.

The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C . The I-suffix devices are characterized for operation from -40°C to 85°C , and the C-suffix devices are characterized for operation from 0°C to 70°C .

General Information

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Operational Amplifiers

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Voltage Comparators

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ORDERING INSTRUCTIONS

Electrical characteristics presented in this data book, unless otherwise noted, apply for the circuit type(s) listed in the page heading regardless of package. The availability of a circuit function in a particular package is denoted by an alphabetical reference above the pin-connection diagram(s). These alphabetical references refer to mechanical outline drawings shown in this section.

Factory orders for circuits described in this data book should include a four-part type number as explained in the following example.

EXAMPLE: TL 514M J /883B

Prefix

MUST CONTAIN TWO OR THREE LETTERS

SN . . . TI Special Functions or Interface Products
 TL, TLE TI Linear Products
 TLC TI Linear Silicon-Gate CMOS Products

STANDARD SECOND-SOURCE PREFIXES

LT . . Linear Technology uA Fairchild/National
 LTC . Linear Technology OP PMI
 MC Motorola RC, RM, or RV . Raytheon
 LF, LM, LP, NE, SA, or SE . . Signetics
 or MF National

Unique Circuit Description Including Temperature Range

MUST CONTAIN TWO TO SIX CHARACTERS
 (From Individual Data Sheets)

Examples: 853 2652A
 1078 27M2AC

Package

MUST CONTAIN ONE OR TWO LETTERS

D, DW, FK, FN, J, JG, L, LP, LU, N, NT, P, U, W
 (From Pin-Connection Diagrams on Individual Data Sheet)

MIL-STD-883B, Method 5004, Class B

OMIT/883B WHEN NOT APPLICABLE

Circuits are shipped in one of the carriers below. Unless a specific method of shipment is specified by the customer (with possible additional costs), circuits will be shipped in the most practical carrier.

Small Outline (D, DW)
 Dual-In-Line (J, JG, N, NT, P)
 —Slide Magazines
 —A-Channel Plastic Tubing
 —Sectioned Cardboard Box
 —Individual Cardboard Box

Chip Carriers (FK, FN)
 —Anti-Static Plastic Tubing
 Flat (U, W)
 —Milton Ross Carriers

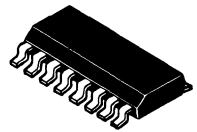
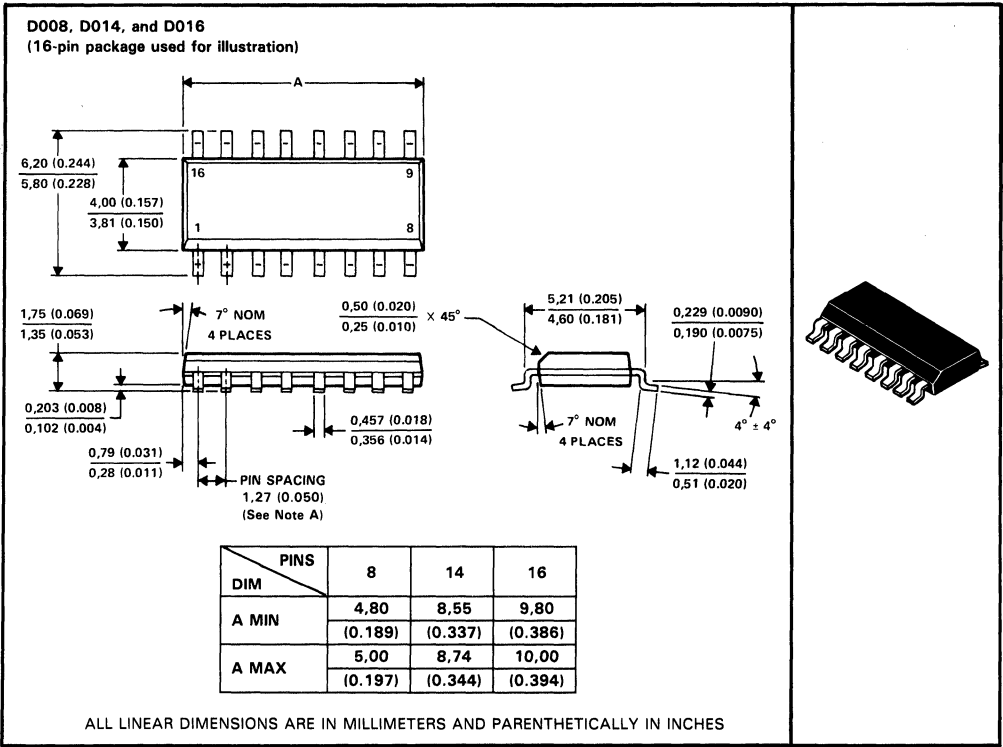
Plug-In (L, LP, LU)
 —Sectional Cardboard Box
 —Individual Cardboard Box

Mechanical Data



D008, D014, and D016 plastic "small outline" packages

Each of these "small outline" packages consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. Leads require no additional cleaning or processing when used in soldered assembly.



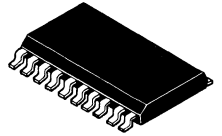
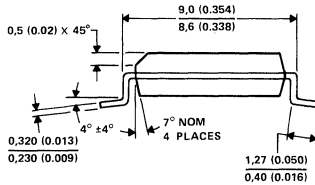
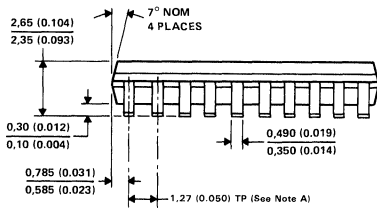
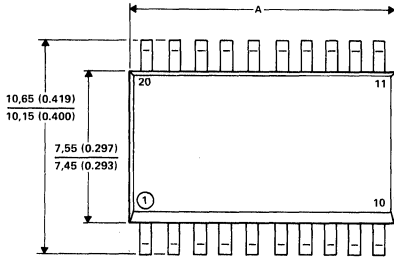
- NOTES: A. Leads are within 0,25 (0.010) radius of true position at maximum material dimension.
 B. Body dimensions do not include mold flash or protrusion.
 C. Mold flash or protrusion shall not exceed 0,15 (0.006).
 D. Lead tips to be planar within ±0,051 (0.002) exclusive of solder.

MECHANICAL DATA

DW016, DW020, DW024, and DW028 plastic "small outline" packages

Each of these "small outline" packages consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. Leads require no additional cleaning or processing when used in soldered assembly.

DW016, DW020, DW024, and DW028
(20-pin package used for illustration)



DIM	PINS			
	16	20	24	28†
A MIN	10,16 (0.400)	12,70 (0.500)	15,29 (0.602)	17,68 (0.696)
A MAX	10,36 (0.408)	12,90 (0.508)	15,49 (0.610)	17,88 (0.704)

ALL LINEAR DIMENSIONS ARE IN MILLIMETERS AND PARENTHETICALLY IN INCHES

- †The 28-pin package drawing is presently classified as Advance Information.
- NOTES: A. Leads are within 0,25 (0.010) radius of true position at maximum material dimension.
 B. Body dimensions do not include mold flash or protrusion.
 C. Mold flash or protrusion shall not exceed 0,15 (0.006).
 D. Lead tips to be planar within ±0,051 (0.002) exclusive of solder.

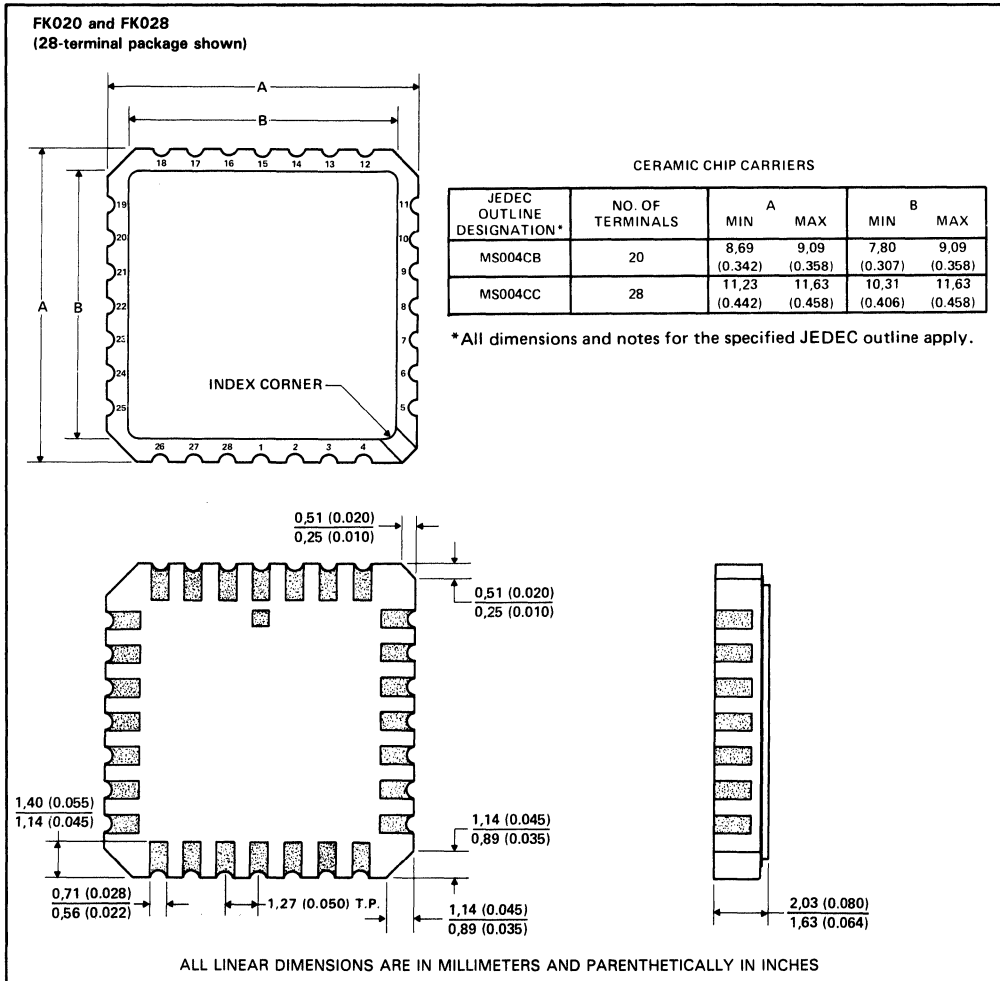
Mechanical Data

6

FK020 and FK028 ceramic chip carrier packages

Each of these hermetically sealed chip carrier packages has a three-layer ceramic base with a metal lid and braze seal. The packages are intended for surface mounting on solder lands on 1,27 (0.050-inch) centers. Terminals require no additional cleaning or processing when used in soldered assembly.

FK package terminal assignments conform to JEDEC Standards 1 and 2.



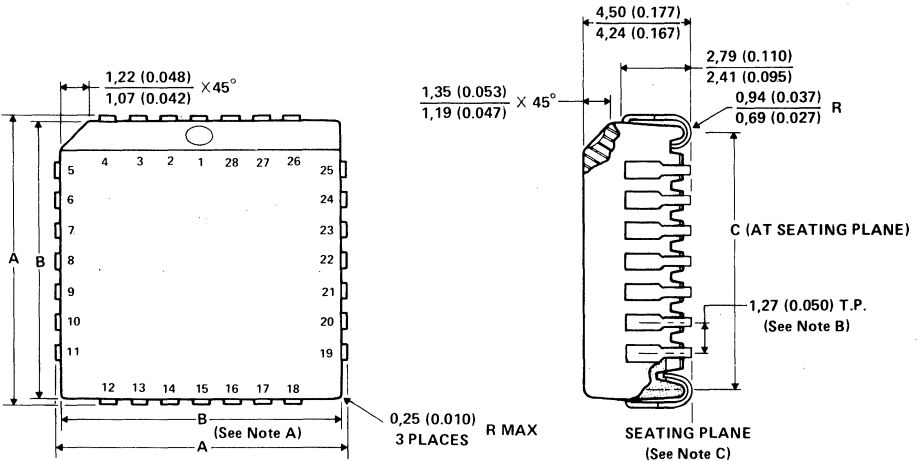
Mechanical Data

MECHANICAL DATA

FN020, FN028, FN044, FN068, and FN084 plastic chip carrier packages

Each of these chip carrier packages consists of a circuit mounted on a lead frame and encapsulated within an electrically nonconductive plastic compound. The compound withstands soldering temperatures with no deformation, and circuit performance characteristics remain stable when the devices are operated in high-humidity conditions. The packages are intended for surface mounting on solder lands on 1,27 (0.050) centers. Leads require no additional cleaning or processing when used in soldered assembly.

FN020, FN028, FN044, FN068, and FN084
(28-terminal package used for illustration)



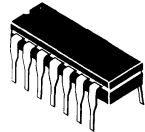
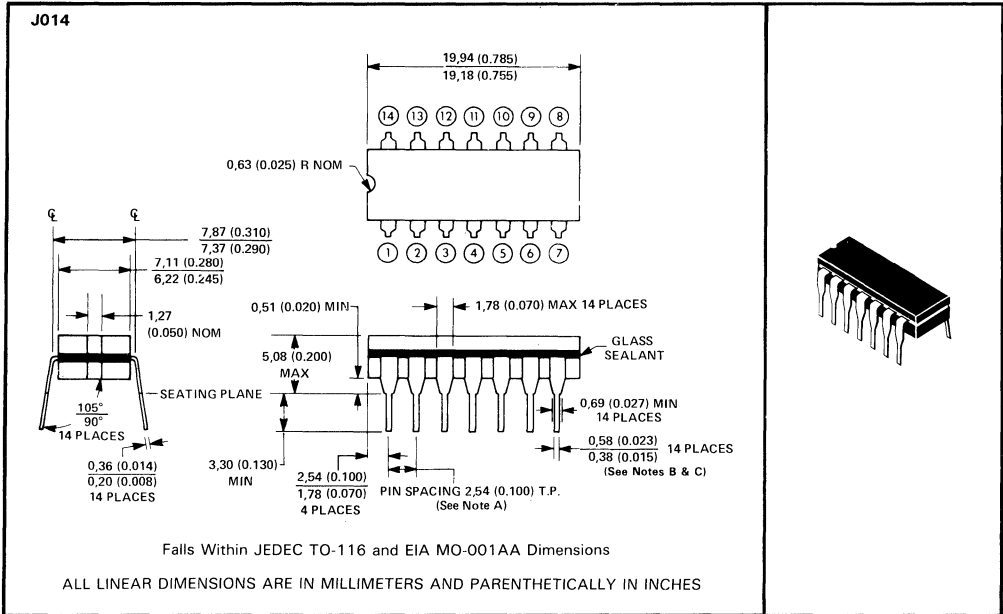
NO. OF TERMINALS	A		B		C	
	MIN	MAX	MIN	MAX	MIN	MAX
20	9,78 (0.385)	10,03 (0.395)	8,89 (0.350)	9,04 (0.356)	7,87 (0.310)	8,38 (0.330)
28	12,32 (0.485)	12,57 (0.495)	11,43 (0.450)	11,58 (0.456)	10,41 (0.410)	10,92 (0.430)
44	17,40 (0.685)	17,65 (0.695)	16,51 (0.650)	16,66 (0.656)	15,49 (0.610)	16,00 (0.630)
68	25,02 (0.985)	25,27 (0.995)	24,13 (0.950)	24,33 (0.956)	23,11 (0.910)	23,62 (0.930)
84	30,10 (1.185)	30,35 (1.195)	29,21 (1.150)	29,41 (1.158)	27,69 (1.090)	28,70 (1.130)

ALL LINEAR DIMENSIONS ARE IN MILLIMETERS AND PARENTHETICALLY IN INCHES

- NOTES: A. Centerline of center pin each side is within 0,10 (0.004) of package centerline as determined by dimension B.
B. Location of each pin is within 0,127 (0.005) of true position with respect to center pin on each side.
C. The lead contact points are planar within 0,10 (0.004).

J014 ceramic dual-in-line package

This hermetically sealed dual-in-line package consists of a ceramic base, ceramic cap, and a lead frame. Hermetic sealing is accomplished with glass. The package is intended for insertion in mounting-hole rows on 7,62 (0.300) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Tin-plated ("bright-dipped") leads require no additional cleaning or processing when used in soldered assembly.

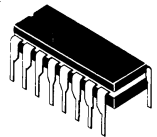
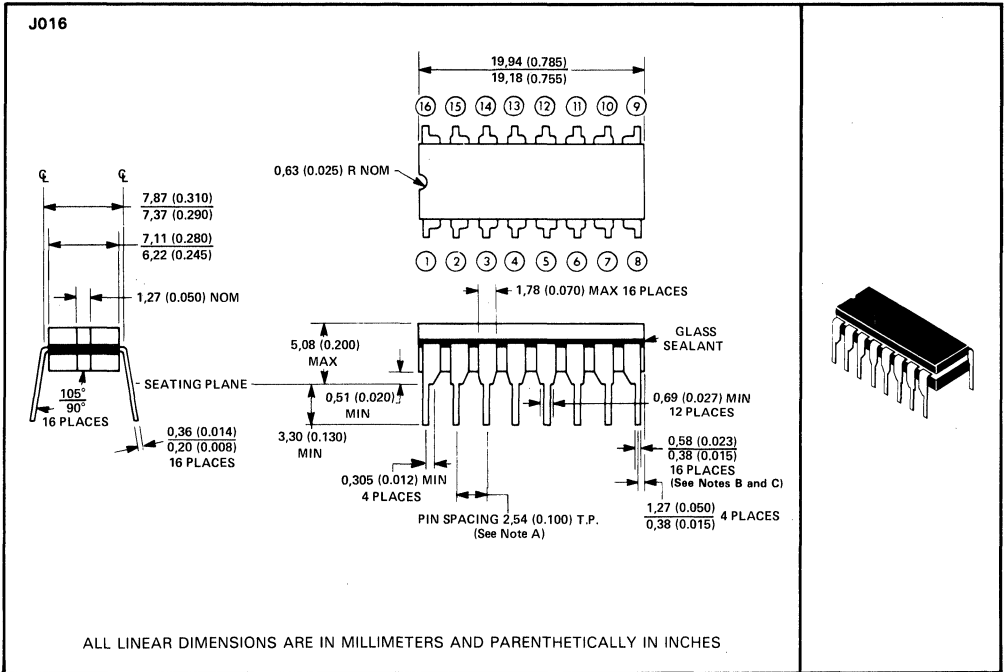


- NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.
 B. This dimension does not apply for solder-dipped leads.
 C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above the seating plane.

MECHANICAL DATA

J016 ceramic dual-in-line package

This hermetically sealed dual-in-line package consists of a ceramic base, ceramic cap, and a lead frame. Hermetic sealing is accomplished with glass. The package is intended for insertion in mounting-hole rows on 7,62 (0.300) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Tin-plated ("bright-dipped") leads require no additional cleaning or processing when used in soldered assembly.

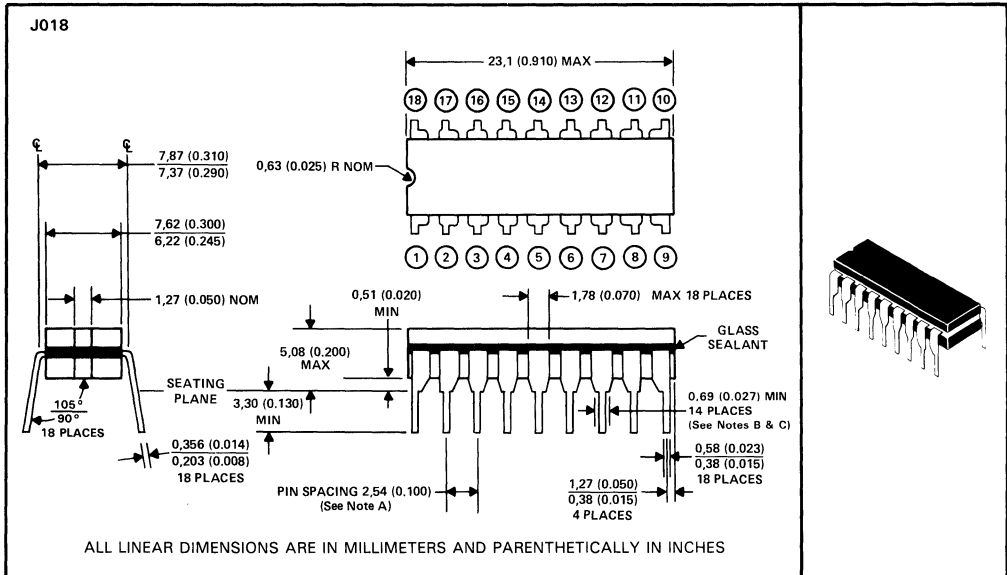


Mechanical Data

- NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.
 B. This dimension does not apply for solder-dipped leads.
 C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above the seating plane.

J018 ceramic dual-in-line package

This hermetically sealed dual-in-line package consists of a ceramic base, ceramic cap, and a lead frame. Hermetic sealing is accomplished with glass. The package is intended for insertion in mounting-hole rows on 7,62 (0.300) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Tin-plated ("bright-dipped") leads require no additional cleaning or processing when used in soldered assembly.



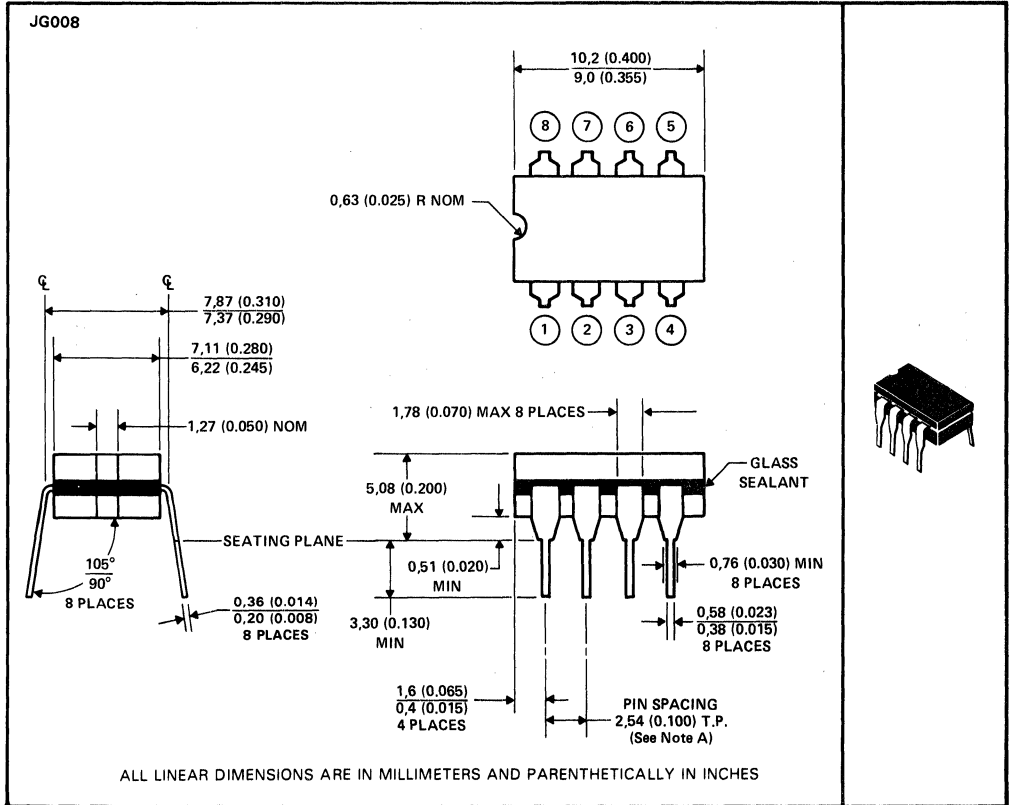
- NOTES:
- A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.
 - B. This dimension does not apply for solder-dipped leads.
 - C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above the seating plane.



MECHANICAL DATA

JG008 ceramic dual-in-line package

This hermetically sealed dual-in-line package consists of a ceramic base, ceramic cap, and an 8-pin lead frame. The package is intended for insertion in mounting-hole rows 7,62 (0.300) centers (see Note A). Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering.



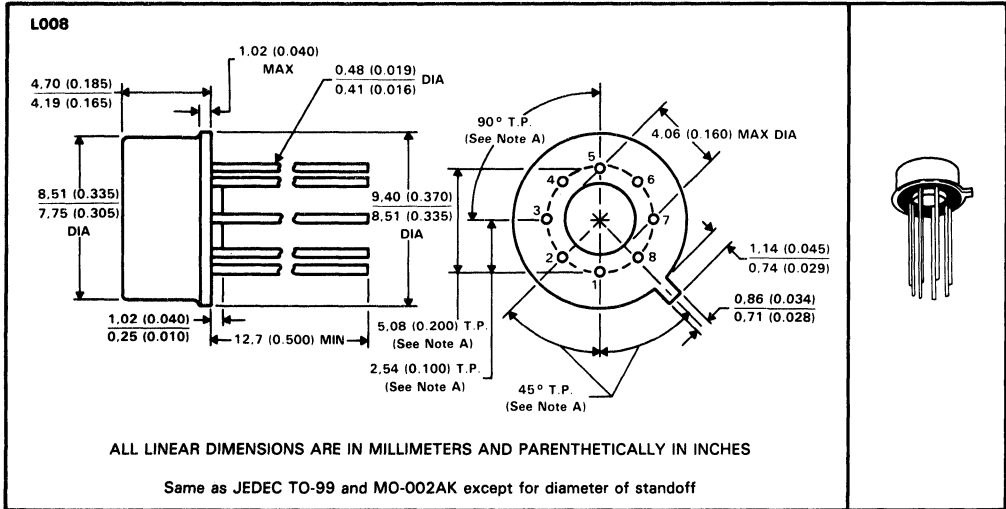
NOTE A: Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.

Mechanical Data

6

L008 metal cylindrical package

This hermetically sealed cylindrical package consists of a welded metal base and cap with individual leads secured by an insulating glass sealant. The gold-plated leads (—00) require no additional cleaning or processing when used in soldered assembly.



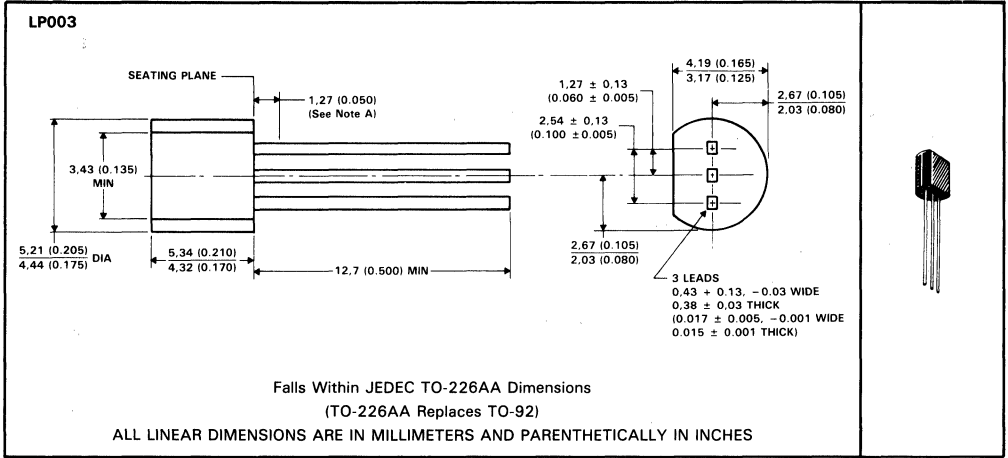
NOTE A: Each lead is located within 0,18 (0.007) of its true position at maximum material condition.



MECHANICAL DATA

LP003 cylindrical plastic package

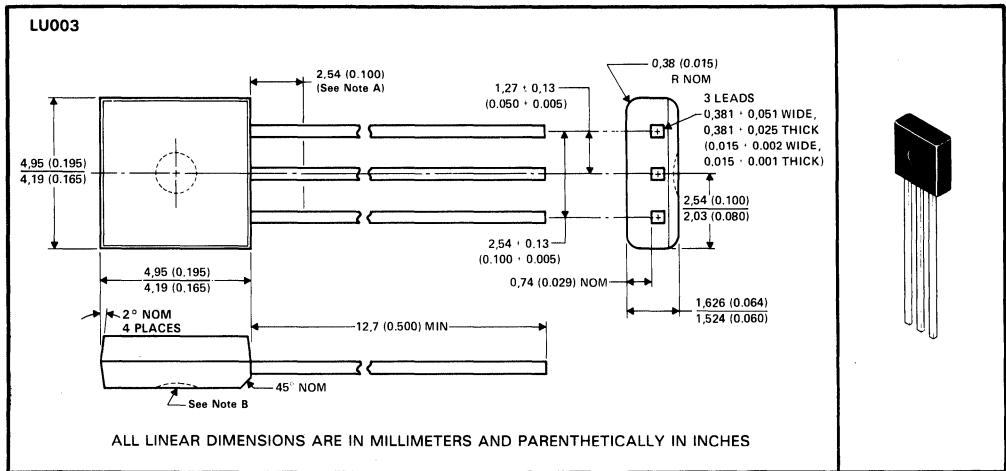
This package consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation and circuit performance characteristics remain stable when operated in high-humidity conditions. Leads require no additional cleaning or processing when used in soldered assembly.



NOTE A: Lead dimensions are not controlled within this area.

LU003 plastic package

This package consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation and circuit performance characteristics remain stable when operated in high-humidity conditions. Leads require no additional cleaning or processing when used in soldered assembly.



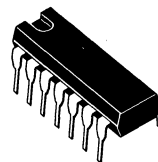
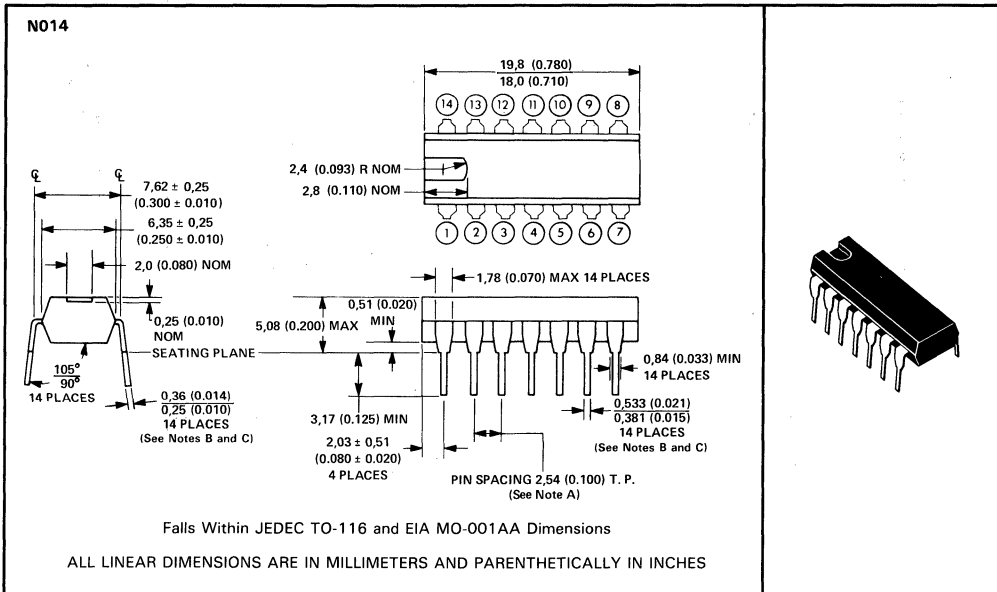
- NOTES: A. Lead dimensions are not controlled in this area.
B. The shape, dimensions, and placement of the index mark is not guaranteed.



MECHANICAL DATA

N014 plastic dual-in-line package

This dual-in-line package consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. The package is intended for insertion in mounting-hole rows on 7,62 (0.300) centers (see Note A). Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Leads require no additional cleaning or processing when used in soldered assembly.



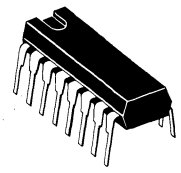
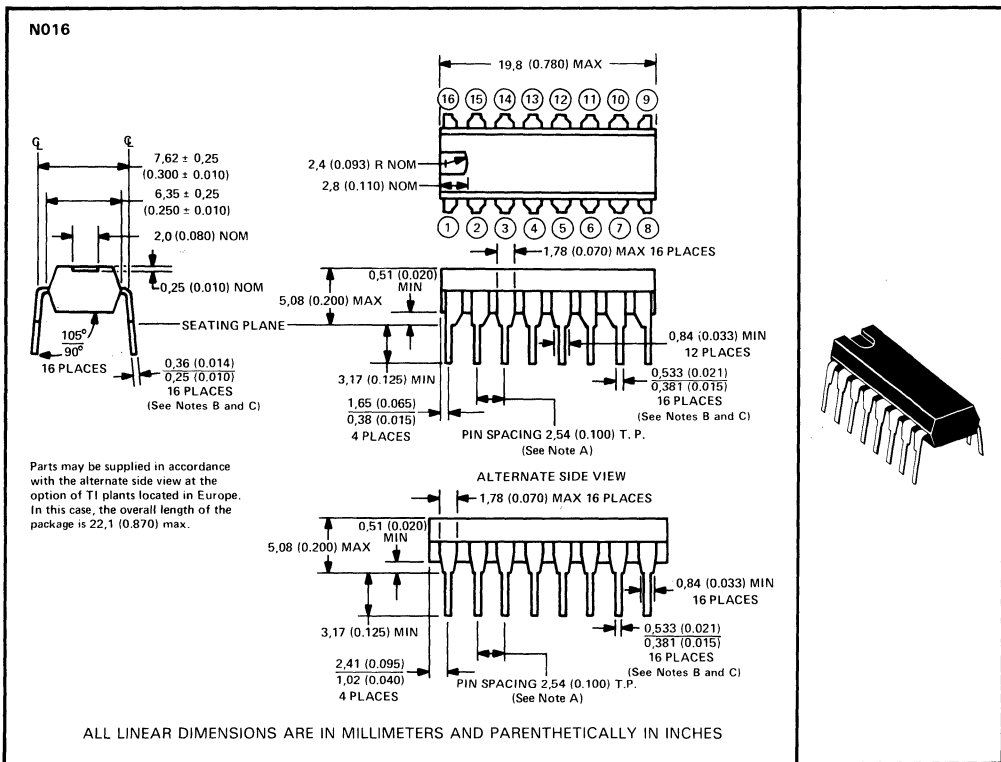
- NOTES:
- A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.
 - B. This dimension does not apply for solder-dipped leads.
 - C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above seating plane.

Mechanical Data



N016 plastic dual-in-line package

This dual-in-line package consists of a circuit mounted on a lead frame and encapsulated within an electrically nonconductive plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. The package is intended for insertion in mounting-hole rows on 7,62 (0.300) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Leads require no additional cleaning or processing when used in soldered assembly.



- NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.
 B. This dimension does not apply for solder-dipped leads.
 C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above seating plane.

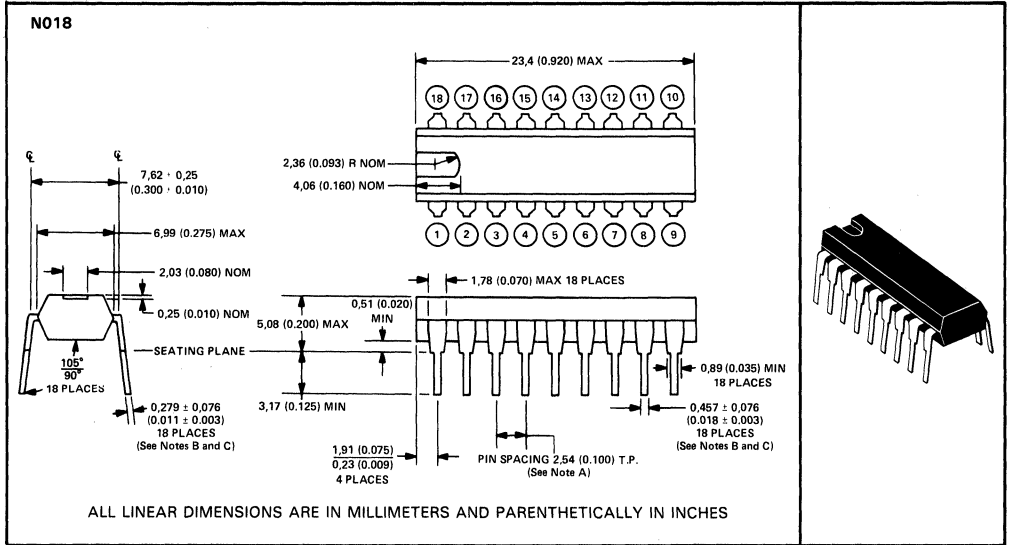
Mechanical Data

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MECHANICAL DATA

N018 plastic dual-in-line package

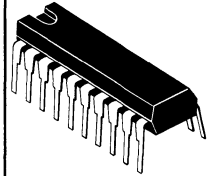
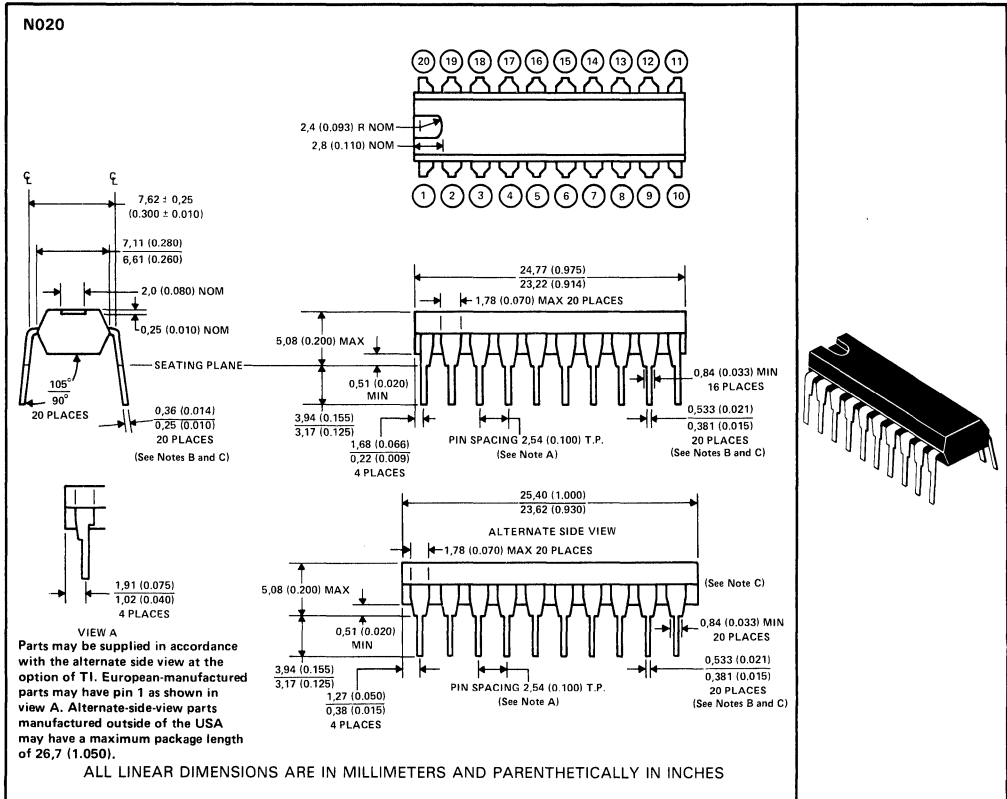
This dual-in-line package consists of a circuit mounted on a lead frame and encapsulated within an electrically nonconductive plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. The package is intended for insertion in mounting-hole rows on 7,62 (0.300) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Leads require no additional cleaning or processing when used in soldered assembly.



- NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.
 B. This dimension does not apply for solder-dipped leads.
 C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above seating plane.

N020 plastic dual-in-line package

This dual-in-line package consists of a circuit mounted on a lead frame and encapsulated within an electrically nonconductive plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. The package is intended for insertion in mounting-hole rows on 7,62 (0.300) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Leads require no additional cleaning or processing when used in soldered assembly.



- NOTES:
- A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.
 - B. This dimension does not apply for solder-dipped leads.
 - C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above seating plane.

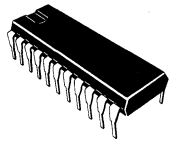
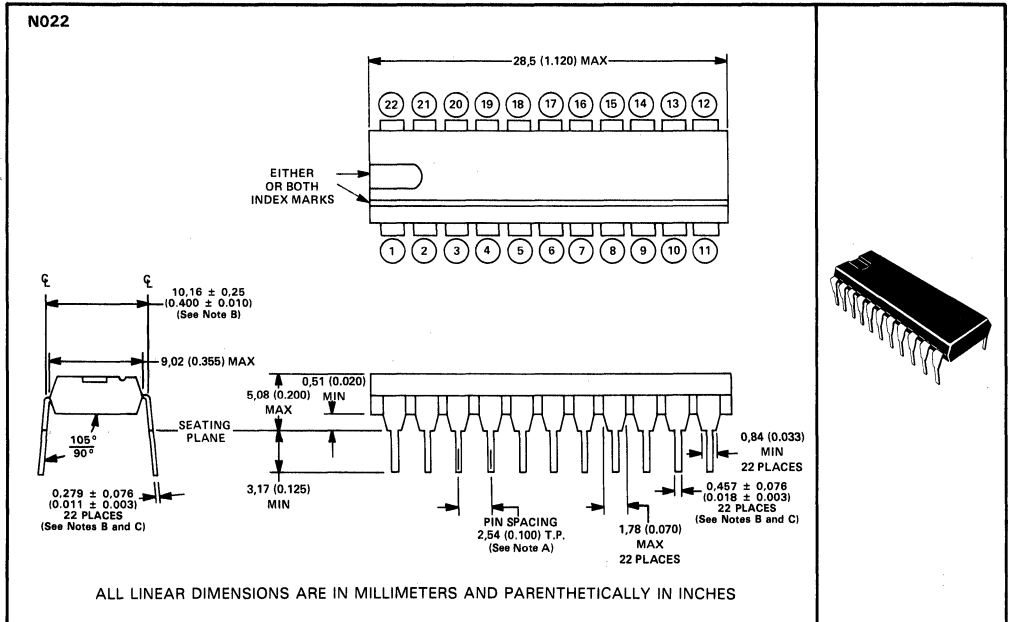
Mechanical Data



MECHANICAL DATA

N022 plastic dual-in-line package

This dual-in-line package consists of a circuit mounted on a lead frame and encapsulated within an electrically nonconductive plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. The package is intended for insertion in mounting-hole rows on 10,16 (0.400) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Leads require no additional cleaning or processing when used in soldered assembly.



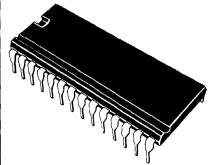
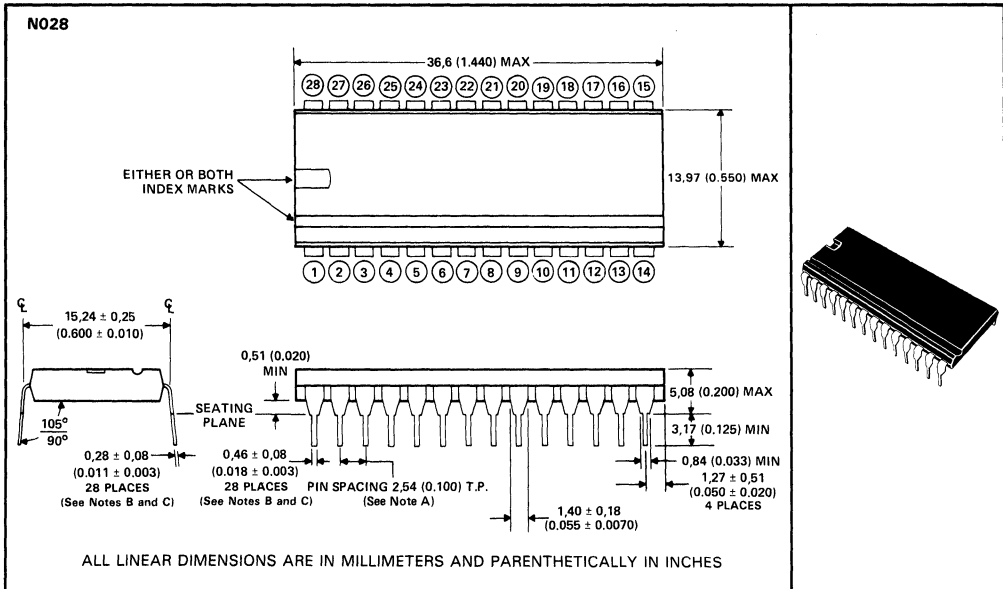
Mechanical Data

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- NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.
 B. This dimension does not apply for solder-dipped leads.
 C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above seating plane.

N028 plastic dual-in-line package

This dual-in-line package consists of a circuit mounted on a lead frame and encapsulated within an electrically nonconductive plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. The package is intended for insertion in mounting-hole rows on 15,24 (0.600) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Leads require no additional cleaning or processing when used in soldered assembly.



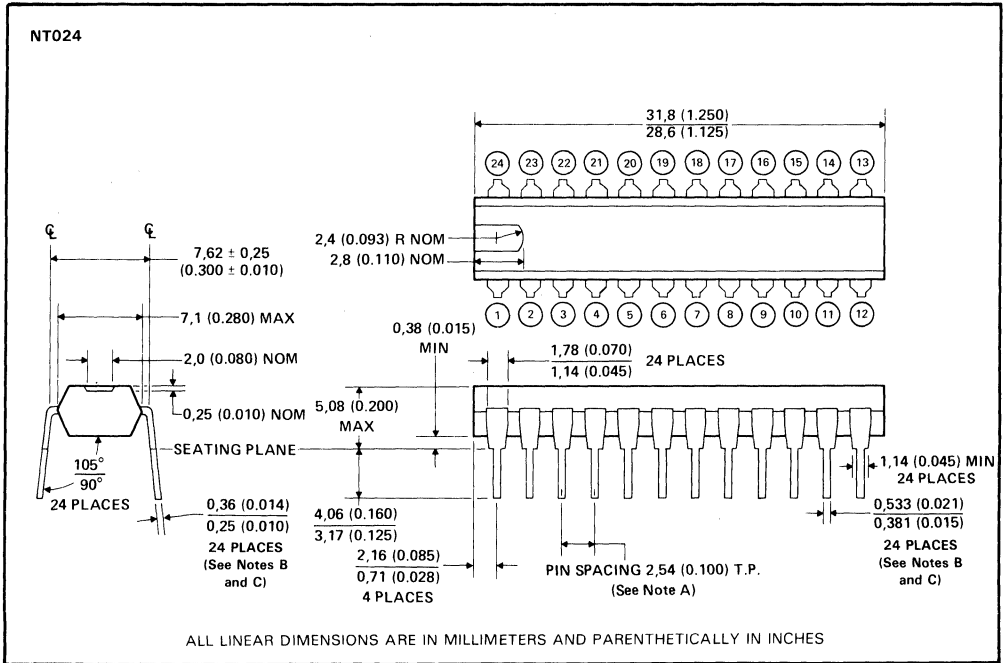
- NOTES: A. Each pin centerline is located within 0,25 (0,010) of its true longitudinal position.
 B. This dimension does not apply for solder-dipped leads.
 C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0,020) above seating plane.

MECHANICAL DATA

NT024 plastic dual-in-line package

This dual-in-line package consists of a circuit mounted on a lead frame and encapsulated within an electrically nonconductive plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. The package is intended for insertion in mounting-hole rows on 7,62 (0.300) centers. Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Leads require no additional cleaning or processing when used in soldered assembly.

NOTE: For all except 24-pin packages, the letter N is used by itself since only the 24-pin package is available in more than one row-spacing. For the 24-pin package, the 7,62 (0.300) version is designated NT; the 15,24 (0.600) version is designated NW. If no second letter or row-spacing is specified, the package is assumed to have 15,24 (0.600) row-spacing.



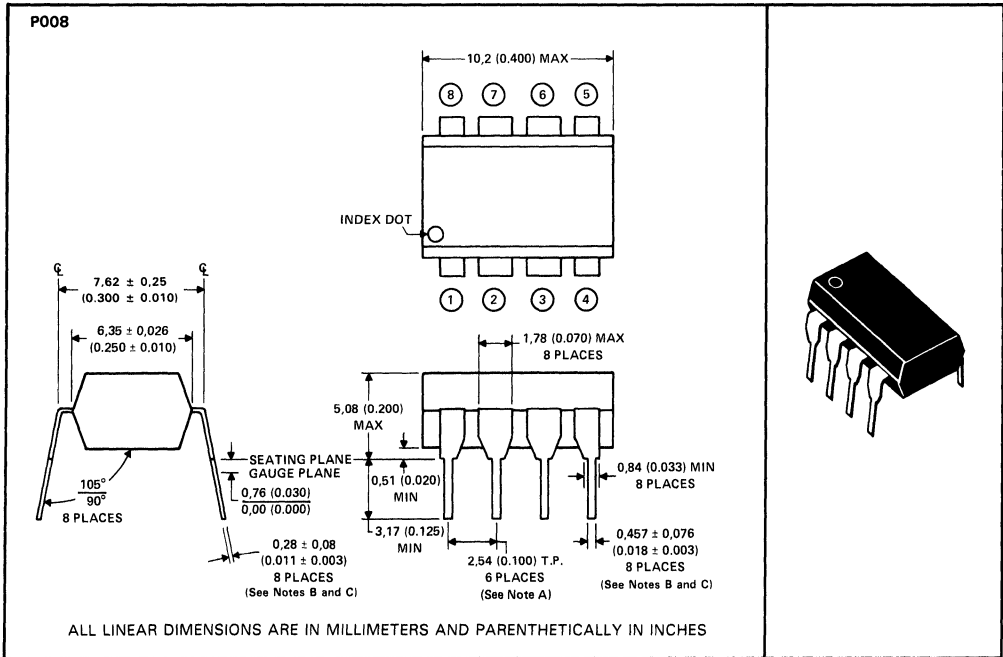
- NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.
 B. This dimension does not apply for solder-dipped leads.
 C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above the seating plane.

Mechanical Data



P008 dual-in-line plastic package

This package consists of a circuit mounted on an 8-pin lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high-humidity conditions. The package is intended for insertion in mounting-hole rows on 7,62 (0.300) centers (See Note A). Once the leads are compressed and inserted, sufficient tension is provided to secure the package in the board during soldering. Solder-plated leads require no additional cleaning or processing when used in soldered assembly.

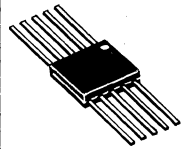
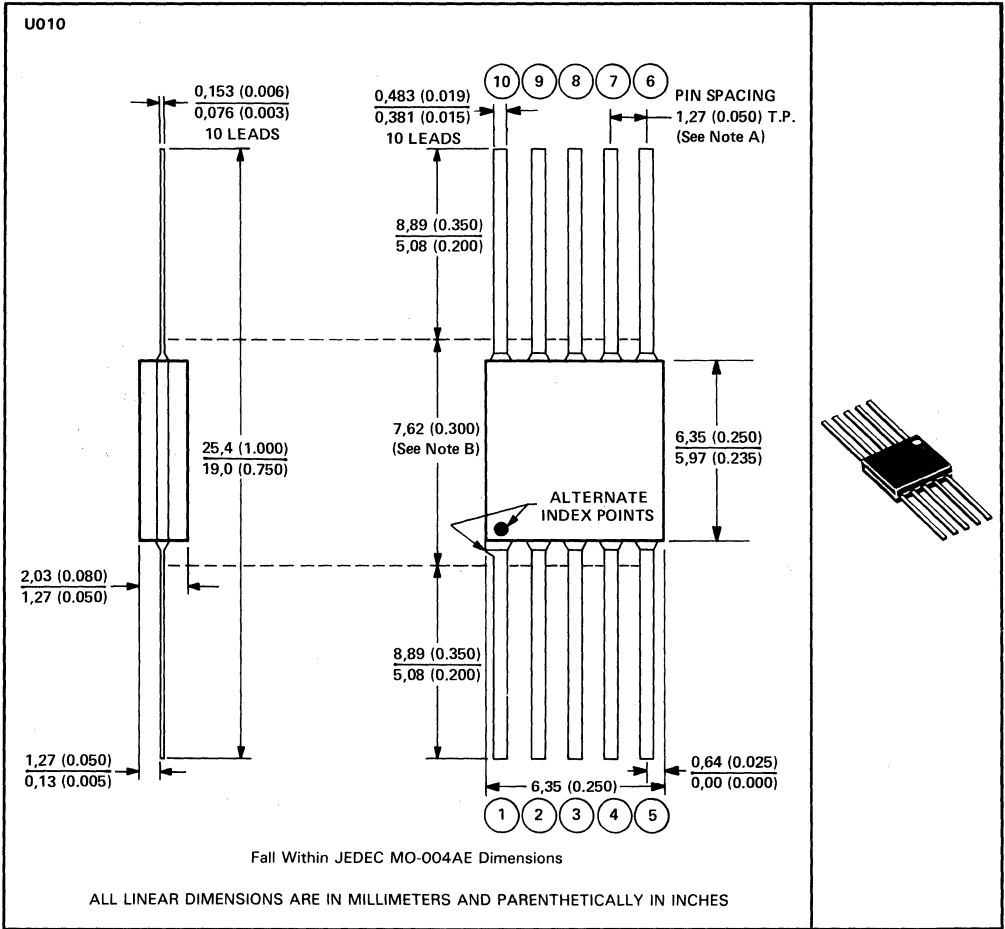


- NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.
 B. This dimension does not apply for solder-dipped leads.
 C. When solder-dipped leads are specified, dipped area of the lead extends from the lead tip to at least 0,51 (0.020) above seating plane.

MECHANICAL DATA

U010 ceramic flat package

This flat package consists of a ceramic base, ceramic cap, and lead frame. Circuit bars are alloy mounted. Hermetic sealing is accomplished with glass. Leads require no additional cleaning or processing when used in soldered assembly.



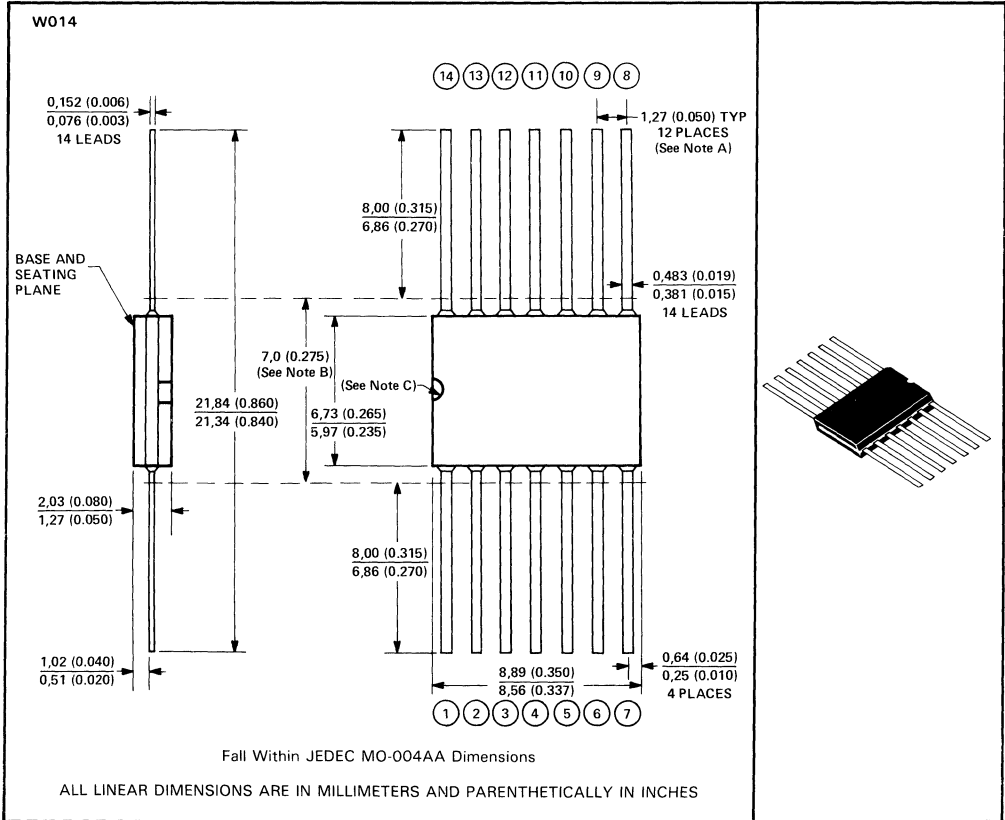
Mechanical Data

6

NOTES: A. Leads are within 0,13 (0.005) radius of true position (T.P.) at maximum material conditions.
B. This dimension determines a zone within which all body and lead irregularities lie.

W014 ceramic flat package

This hermetically sealed flat package consists of an electrically nonconductive ceramic base and cap and a lead frame. Hermetic sealing is accomplished with glass. Leads require no additional cleaning or processing when used in soldered assembly.



- NOTES: A. Leads are within 0,13 (0.005) radius of true position (T.P.) at maximum material condition.
 B. This dimension determines a zone within which all body and lead irregularities lie.
 C. Index point is provided on cap for terminal identification only.

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